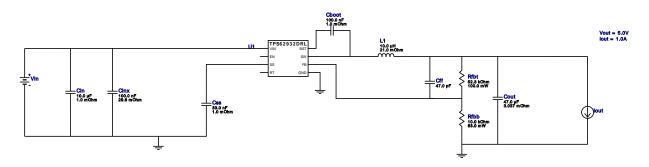
VinMin = 24.0V VinMax = 24.0V Vout = 5.0V Iout = 1.0A Device = TPS62932DRLR Topology = Buck Created = 2024-02-15 11:28:19.149 BOM Cost = \$1.22 BOM Count = 10 Total Pd = 0.26W

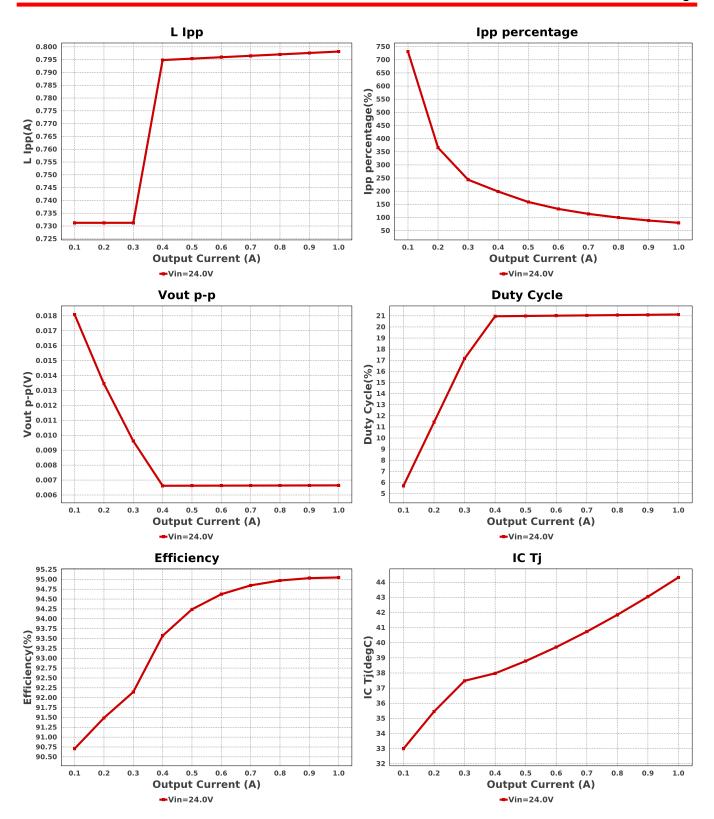
# WEBENCH® Design Report

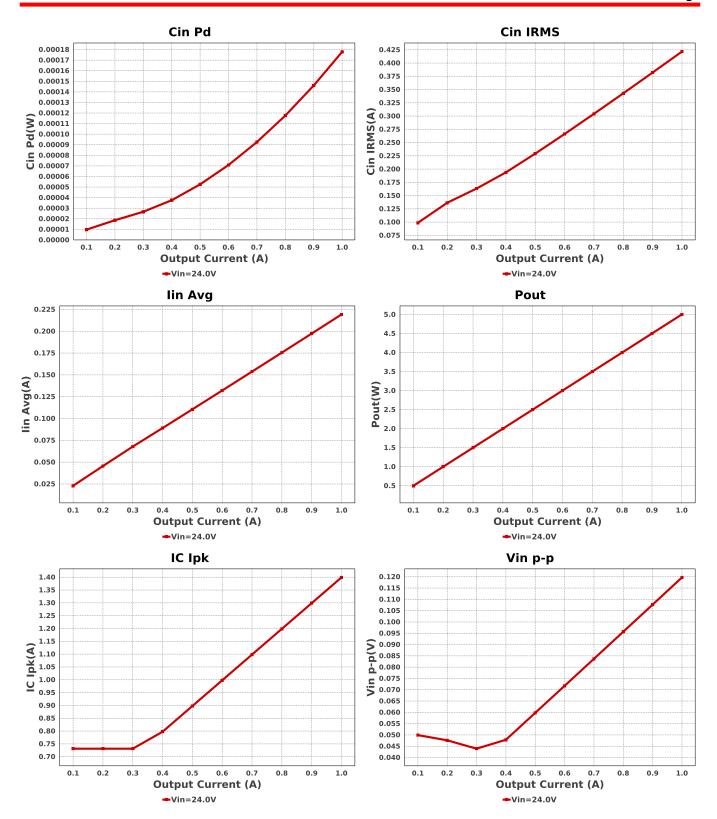
Design: 2 TPS62932DRLR TPS62932DRLR 24V-24V to 5.00V @ 1A

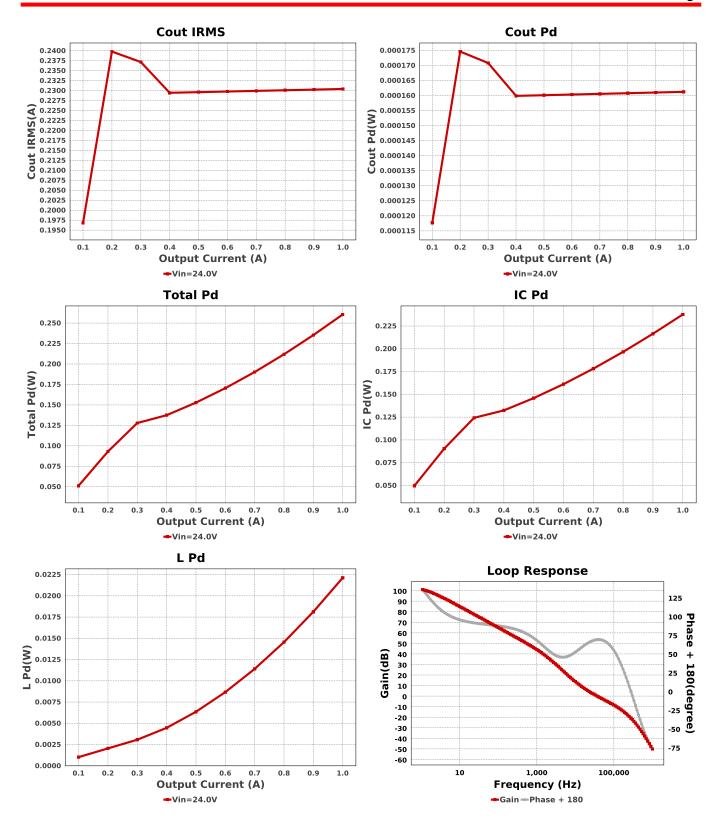


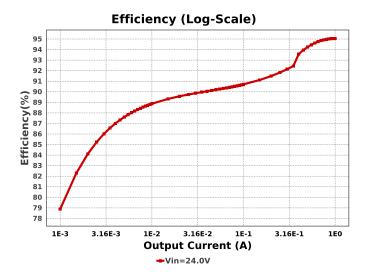
### **Electrical BOM**

Name	Manufacturer	Part Number	Properties	Qty	Price	Footprint
Cboot	MuRata	GRM155R71A104KA01D Series= X7R	Cap= 100.0 nF ESR= 1.0 mOhm VDC= 10.0 V IRMS= 0.0 A	1	\$0.01	0402 3 mm <sup>2</sup>
Cff	MuRata	GRM0335C1E470JA01D Series= C0G/NP0	Cap= 47.0 pF VDC= 25.0 V IRMS= 0.0 A	1	\$0.01	0201 2 mm <sup>2</sup>
Cin	TDK	C3216X5R1H106K160AB Series= X5R	Cap= 10.0 uF ESR= 1.0 mOhm VDC= 50.0 V IRMS= 4.9 A	1	\$0.24	1206_180 11 mm <sup>2</sup>
Cinx	TDK	CGA3E2X7R1H104K080AA Series= X7R	Cap= 100.0 nF ESR= 29.6 mOhm VDC= 50.0 V IRMS= 971.99 mA	1	\$0.01	0603 5 mm <sup>2</sup>
Cout	MuRata	GRM32ER61C476KE15L Series= X5R	Cap= 47.0 uF ESR= 3.037 mOhm VDC= 16.0 V IRMS= 4.59346 A	1	\$0.17	1210_280 15 mm <sup>2</sup>
Css	MuRata	GRM155R71A333KA01D Series= X7R	Cap= 33.0 nF ESR= 1.0 mOhm VDC= 10.0 V IRMS= 0.0 A	1	\$0.01	0402 3 mm <sup>2</sup>
L1	Bourns	SRR1208-100ML	L= 10.0 μH 21.0 mOhm	1	\$0.56	SRR1208 216 mm <sup>2</sup>
Rfbb	Vishay-Dale	CRCW040210K0FKED Series= CRCWe3	Res= 10.0 kOhm Power= 63.0 mW Tolerance= 1.0%	1	\$0.01	0402 3 mm <sup>2</sup>
Rfbt	Yageo	RC0603FR-0752K3L Series= ?	Res= 52.3 kOhm Power= 100.0 mW Tolerance= 1.0%	1	\$0.01	0603 5 mm <sup>2</sup>
U1	Texas Instruments	TPS62932DRLR	Switcher	1	\$0.19	DRL0008A-MFG 9 mm <sup>2</sup>









## **Operating Values**

#	Name	Value	Cotogoni	Describe the co
	Name	v alue	Category	Description
1.	Cin IRMS	421.686 mA	Capacitor	Input capacitor RMS ripple current
2.	Cin Pd	177.82 μW	Capacitor	Input capacitor power dissipation
3.	Cout IRMS	230.409 mA	Capacitor	Output capacitor RMS ripple current
4.	Cout Pd	161.23 μW	Capacitor	Output capacitor power dissipation
5.	IC lpk	1.399 A	IC '	Peak switch current in IC
6.	IC Pd	237.65 mW	IC	IC power dissipation
7.	IC Ti	44.319 degC	IC	IC junction temperature
8.	IC Tolerance	16.0 mV	IC	IC Feedback Tolerance
9.	ICThetaJA Effective	60.25 degC/W	IC	Effective IC Junction-to-Ambient Thermal Resistance
10.	lin Avg	219.19 mA	iC	Average input current
11.	Ipp percentage	79.816 %	Inductor	Inductor ripple current percentage (with respect to average inductor current)
12	L lpp	798.16 mA	Inductor	Peak-to-peak inductor ripple current
	L Pd	22.115 mW	Inductor	Inductor power dissipation
	Cin Pd	177.82 μW	Power	Input capacitor power dissipation
	Cout Pd	161.23 μW	Power	Output capacitor power dissipation
	IC Pd	237.65 mW	Power	IC power dissipation
-	L Pd	22.115 mW	Power	Inductor power dissipation
18.	Total Pd	260.503 mW	Power	Total Power Dissipation
_				
19.	BOM Count	10	System Information	Total Design BOM count
20.	Cross Freq	37.068 kHz	System Information	Bode plot crossover frequency
21.	Duty Cycle	21.122 %	System Information	Duty cycle
22.	Efficiency	95.048 %	System Information	Steady state efficiency
23.	FootPrint	271.0 mm <sup>2</sup>	System Information	Total Foot Print Area of BOM components
24.	Frequency	500.0 kHz	System Information	Switching frequency
25.	Gain Marg	-21.051 dB	System Information	Bode Plot Gain Margin
26.	Inductor ripple current	40 0 %	System	Custom Inductor ripple current (% of average inductor current)
20.	requirement used for Inductor selection	40.0 /0	Information	requirement used for Inductor selection
27.	lout	1.0 A	System Information	lout operating point
28.	lout transient step used for Cout calculations	d 500.0 mA	System Information	Custom Transient current step requirement that was used for Cout selection (A).
29.	Low Freq Gain	100.953 dB	System Information	Gain at 1Hz
30.	Mode	CCM	System Information	Conduction Mode
31.	Overshoot Value	7.887 mV	System Information	Theoretical Vout Overshoot Value
32.	Phase Marg	69.931 deg	System Information	Bode Plot Phase Margin
33.	Pout	5.0 W	System Information	Total output power
34.	Total BOM	\$1.22	System Information	Total BOM Cost

#	Name	Value	Category	Description
35.	Undershoot Value	27.052 mV	System Information	Theoretical Vout Undershoot Value
36.	Vin	24.0 V	System Information	Vin operating point
37.	Vin p-p	119.704 mV	System Information	Peak-to-peak input voltage
38.	Vout	5.0 V	System Information	Operational Output Voltage
39.	Vout Actual	4.984 V	System Information	Vout Actual calculated based on selected voltage divider resistors
40.	Vout Ripple requirement used for Cout calculations	1.0 %	System Information	Custom maximum output ripple requirement that was used for Cout selection(% of Vout).
41.	Vout Tolerance	3.73 %	System Information	Vout Tolerance based on IC Tolerance (no load) and voltage divider resistors if applicable
42.	Vout p-p	6.646 mV	System Information	Peak-to-peak output ripple voltage
43.	Vout transient requirement used for Cout calculations	3.0 %	System Information	Custom Transient voltage change requirement that was used for Cout selection (% of Vout).

### **Design Inputs**

Name	Value	Description	
lout	1.0	Maximum Output Current	
VinMax	24.0	Maximum input voltage	
VinMin	24.0	Minimum input voltage	
Vout	5.0	Output Voltage	
base_pn	TPS62932	Base Product Number	
source	DC	Input Source Type	
Ta	30.0	Ambient temperature	

## WEBENCH® Assembly

#### Component Testing

Some published data on components in datasheets such as Capacitor ESR and Inductor DC resistance is based on conservative values that will guarantee that the components always exceed the specification. For design purposes it is usually better to work with typical values. Since this data is not always available it is a good practice to measure the Capacitance and ESR values of Cin and Cout, and the inductance and DC resistance of L1 before assembly of the board. Any large discrepancies in values should be electrically simulated in WEBENCH to check for instabilities and thermally simulated in WebTHERM to make sure critical temperatures are not exceeded.

#### Soldering Component to Board

If board assembly is done in house it is best to tack down one terminal of a component on the board then solder the other terminal. For surface mount parts with large tabs, such as the DPAK, the tab on the back of the package should be pre-tinned with solder, then tacked into place by one of the pins. To solder the tab town to the board place the iron down on the board while resting against the tab, heating both surfaces simultaneously. Apply light pressure to the top of the plastic case until the solder flows around the part and the part is flush with the PCB. If the solder is not flowing around the board you may need a higher wattage iron (generally 25W to 30W is enough).

#### Initial Startup of Circuit

It is best to initially power up the board by setting the input supply voltage to the lowest operating input voltage 24.0V and set the input supply's current limit to zero. With the input supply off connect up the input supply to Vin and GND. Connect a digital volt meter and a load if needed to set the minimum lout of the design from Vout and GND. Turn on the input supply and slowly turn up the current limit on the input supply. If the voltage starts to rise on the input supply continue increasing the input supply current limit while watching the output voltage. If the current increases on the input supply, but the voltage remains near zero, then there may be a short or a component misplaced on the board. Power down the board and visually inspect for solder bridges and recheck the diode and capacitor polarities. Once the power supply circuit is operational then more extensive testing may include full load testing, transient load and line tests to compare with simulation results.

#### Load Testing

The setup is the same as the initial startup, except that an additional digital voltmeter is connected between Vin and GND, a load is connected between Vout and GND and a current meter is connected in series between Vout and the load. The load must be able to handle at least rated output power + 50% (7.5 watts for this design). Ideally the load is supplied in the form of a variable load test unit. It can also be done in the form of suitably large power resistors. When using an oscilloscope to measure waveforms on the prototype board, the ground leads of the oscilloscope probes should be as short as possible and the area of the loop formed by the ground lead should be kept to a minimum. This will help reduce ground lead inductance and eliminate EMI noise that is not actually present in the circuit.

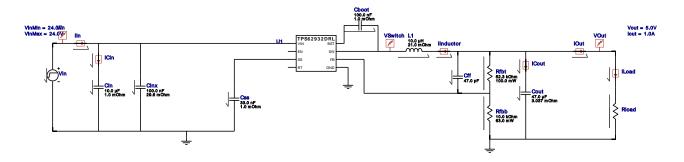


## **WEBENCH**<sup>®</sup> Electrical Simulation Report

Design Id = 2

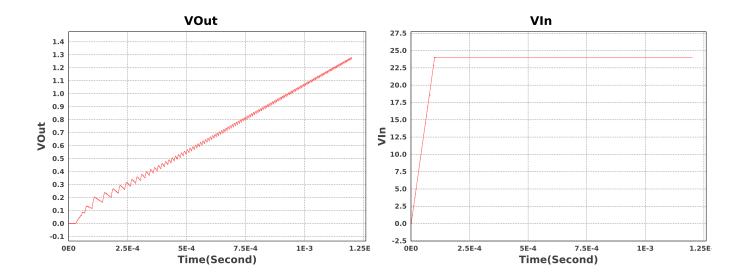
 $sim_id = 2$ 

Simulation Type = Startup



#### Simulation Parameters

#	Name	Parameter Name	Description	Values
		IC	Initial Voltage	0 V
2.	Rload	R	Load Resistance	5.0 ohm



#### Design Assistance

- 1. Master key: 37C3C50233B7D137384961381C9450C6[v1]
- $2. \ \textbf{TPS62932} \ \textbf{Product Folder: http://www.ti.com/product/TPS62932: contains the data sheet and other resources.} \\$

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