

Properties of Signals

PS-1

RECALL

→ analog signals typically represented using real-valued functions of a real variable, t , as in

$$x(t), u(t), y(t)$$

→ we can use the mathematical notation

$$x: \mathbb{R} \rightarrow \mathbb{R}$$

as well.

→ discrete-time signals are represented using real-valued functions of an integer variable, n , as in

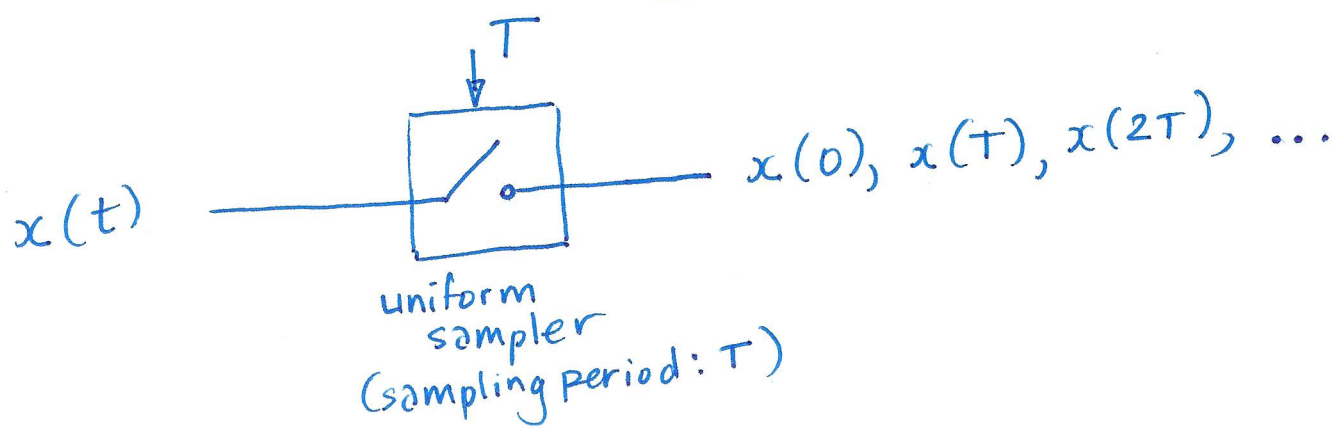
$$x(n), u(n), y(n)$$

→ mathematically:

$$x: \mathbb{Z} \rightarrow \mathbb{R}$$

→ discrete-time signals can also be referred to as sequences.

- while discrete-time signals are things "onto themselves" they are often viewed as being derived from analog signals in a process known as sampling.



and we define

↗ "for all"
 $\forall n \quad x_d(n) = x(nT)$

thus the discrete-time signal, $x_d(n)$, is obtained from the analog signal, $x(t)$.

Recall: Shannon's Sampling Theorem

The analog (continuous-time) signal can be recovered ("perfectly reconstructed") from $x_d(n)$ provided that:

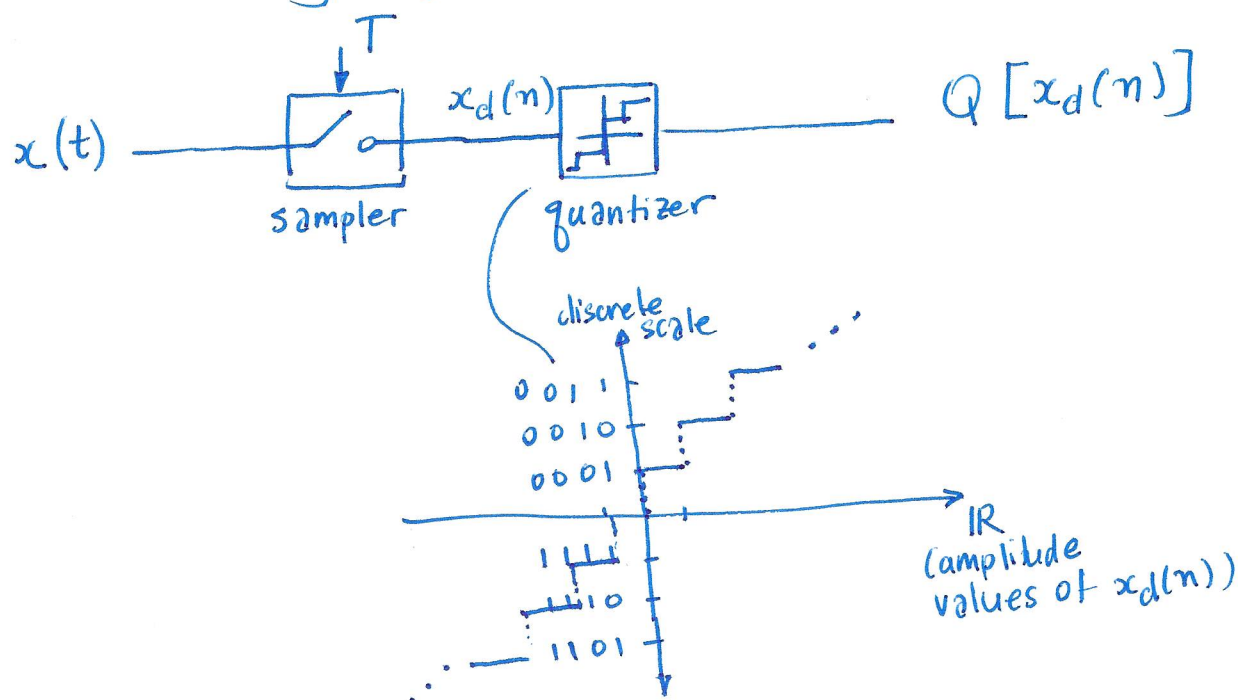
- i) $x(t)$ is band-limited with band limit, $\Omega_B > 0$
 (i.e., if $X(j\Omega) = \mathcal{F}[x(t)]$, then $|X(j\Omega)| = 0 \quad \forall |\Omega| \geq \Omega_B$)

- ii) $\Omega_s = \frac{2\pi}{T_s} > \Omega_B$, where T_s is the sampling period. ▮

Amplitude Quantization

PS-3

If, in addition to sampling (or "time quantization" or "temporal quantization") we quantize the amplitude of the analog signal, a digital signal is produced



we are often concerned with the resolution of the discretized amplitude scale, often expressed as the number of bits needed to represent the staircase characteristic. ▮

Elementary Signals

PS-4

- earlier in the course, we reviewed elementary signals in both continuous- and discrete-time, used for a variety of analyses:

→ continuous-time: delta function, unit step function

$$\int_{-\epsilon}^{\epsilon} \delta(t) dt = 1$$

for all $\epsilon > 0$

→ $\delta(t)$, $u_s(t)$
→ many ways to define $\delta(t)$!

→ discrete-time: unit sample sequence, unit step sequence

$$\delta[n] = \begin{cases} 1, & n=0 \\ 0, & n \neq 0 \end{cases}$$

$\forall n \in \mathbb{Z}$

→ $\delta(n)$, $u_s(n)$

we can write:

$$u_s(n) = \sum_{k=-\infty}^n \delta(k)$$

→ homework:
prove this

- we also make extensive use of sinusoids, both real- and complex-valued (note: a complex-value can be seen as a pair of real values, the Real and Imaginary components)

→ CT: $\sin(\omega_0 t + \varphi)$, $\cos(\omega_0 t + \varphi)$, $e^{j(\omega_0 t + \varphi)}$

NOTE: Euler's Theorem

$$e^{j\theta} = \cos \theta + j \sin \theta$$

$$\therefore e^{j(\omega_0 t + \varphi)} = \cos(\omega_0 t + \varphi) + j \sin(\omega_0 t + \varphi)$$

→ DT: $\sin(\omega_0 n + N)$, $\cos(\omega_0 n + N)$, $e^{j(\omega_0 n + N)}$ PS-5

Consider: $x(n) = Ae^{j(\omega_0 + 2\pi r)n}$, $A \in \mathbb{R}$, $r \in \mathbb{Z}$,
for all $n \in \mathbb{Z}$.

so we have:

$$x(n) = A e^{j\omega_0 n} \underbrace{e^{j(2\pi r)n}}$$

$$e^{j(2\pi r)n} = \cos(2\pi r n) + j \sin(2\pi r n)$$

$$= \cos(2\pi L) + j \sin(2\pi L),$$

where $L = rn$, is just
an integer

$$\Rightarrow e^{j(2\pi r)n} = 1 + j(0) = 1$$

$$\therefore x(n) = A e^{j\omega_0 n}$$

→ so we see that complex exponentials in discrete-time are "periodic in ω ", with frequencies differing by integer multiples of 2π , are indistinguishable from each other.

Q: is this true in continuous-time? Why or why not?

Temporal Periodicity

PS-6

CT: $x(t)$ is periodic with period T if,
for any $t \in \mathbb{R}$,

$$x(t) = x(t+T)$$

→ the period T is also a real number.

DT: $x(n)$ is periodic with period N if,
for any $n \in \mathbb{Z}$,

$$x(n) = x(n+N),$$

where $N \in \mathbb{Z}$.

EX.// $x(t) = 3\cos(\Omega_0 t + 7)$

• find the period T :

→ if T exists, it satisfies:

$$x(t) = x(t+T) \quad \forall t \in \mathbb{R}$$

$$3\cos(\Omega_0 t + 7) = 3\cos(\Omega_0 t + \Omega_0 T + 7)$$

NOTE: $\cos(\theta)$ is periodic with 2π , so that
 $\cos(\theta) = \cos(\theta + 2\pi)$

setting $\theta = \Omega_0 t + 7$, and

$$2\pi = \Omega_0 T,$$

we find $\boxed{T = \frac{2\pi}{\Omega_0}}$

setting $T = \frac{2\pi}{\Omega_0}$ yields:

PS-7

$$\begin{aligned} x(t) &= 3 \cos \left[\Omega_0 t + \Omega_0 \left(\frac{2\pi}{\Omega_0} \right) + \gamma \right] \\ &= 3 \cos \left[\Omega_0 t + \gamma + 2\pi \right] \\ &= 3 \left[\cos(\Omega_0 t + \gamma) \cos(2\pi) - \sin(\Omega_0 t + \gamma) \sin(2\pi) \right] \\ &= 3 \cos(\Omega_0 t + \gamma) \end{aligned}$$

which proves that the period is $T = \frac{2\pi}{\Omega_0}$ \square

EX. $x(n) = 3 \cos(\omega_0 n + \gamma)$

we want to find N such that

$$x(n) = x(n+N)$$

for all n . Recall that N itself must be an integer.

$$\begin{aligned} \text{write: } x(n+N) &= 3 \cos [\omega_0(n+N) + \gamma] \\ &= 3 \cos \left[\underbrace{(\omega_0 n + \gamma)}_0 + \underbrace{\omega_0 N}_{2\pi k}, \text{ for some } k \in \mathbb{Z} \right] \end{aligned}$$

\therefore we want

$$\omega_0 N = 2\pi k$$

$$\Rightarrow N = \frac{2\pi k}{\omega_0}$$

However, in order for N to be an integer, $\frac{2\pi k}{\omega_0}$ must be an integer; PS-8

if we write $\omega_0 = 2\pi f_0$, then

$$\frac{2\pi k}{\omega_0} = \frac{2\pi k}{2\pi f_0} = \frac{k}{f_0}$$

so we need

$$\frac{k}{f_0} = m, \text{ for some } m \in \mathbb{Z}$$

$$\Rightarrow f_0 = \frac{k}{m}$$

$\therefore f_0$ must be a ratio of two integers,
in other words, a rational number. \square

Even & Odd Signals

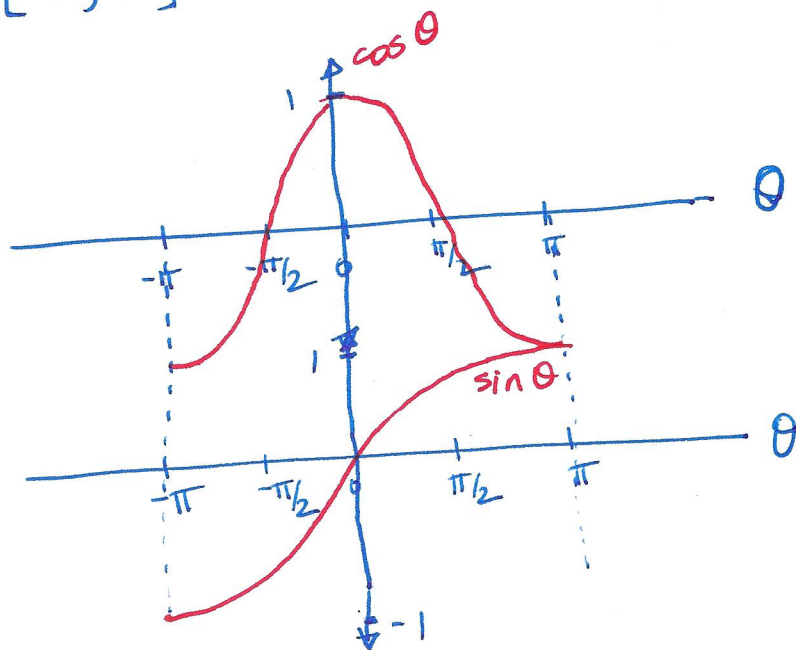
PS-9

- a symmetry property:

i) $f(x)$ is even if $f(x) = f(-x)$

ii) $f(x)$ is odd if $f(x) = -f(-x)$

- consider $\cos(\theta)$ & $\sin(\theta)$ in the domain $\theta \in [-\pi, \pi]$



which is even, which is odd?

Aperiodic Signals

→ signals which are not periodic!

→ in the frequency domain, such signals have continuous (smooth) rather than discrete spectra — recall that only periodic signals can be described exactly by a Fourier Series