

Fourier Series

Fourier Series Representation of Continuous Time periodic signals

Signals can have dual personalities.

Time domain perspective

- (i) Depicts its wave form
- (ii) Signal width (duration)
- (iii) Rate at which the waveform decays

Frequency Domain perspectives.

- (i) in Terms of sinusoidal components
- (ii) their relative amplitudes and phases

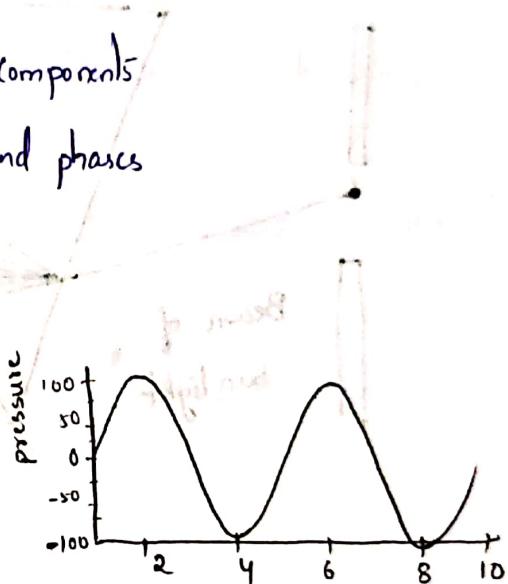
Time Domain

→ Despicts its wave form

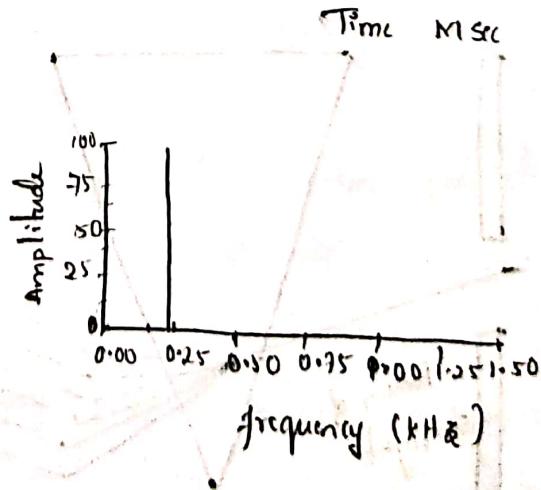
$$\text{Amplitude} = 100$$

frequency = number of cycles in

$$\text{One second} = 200 \text{ Hz}$$

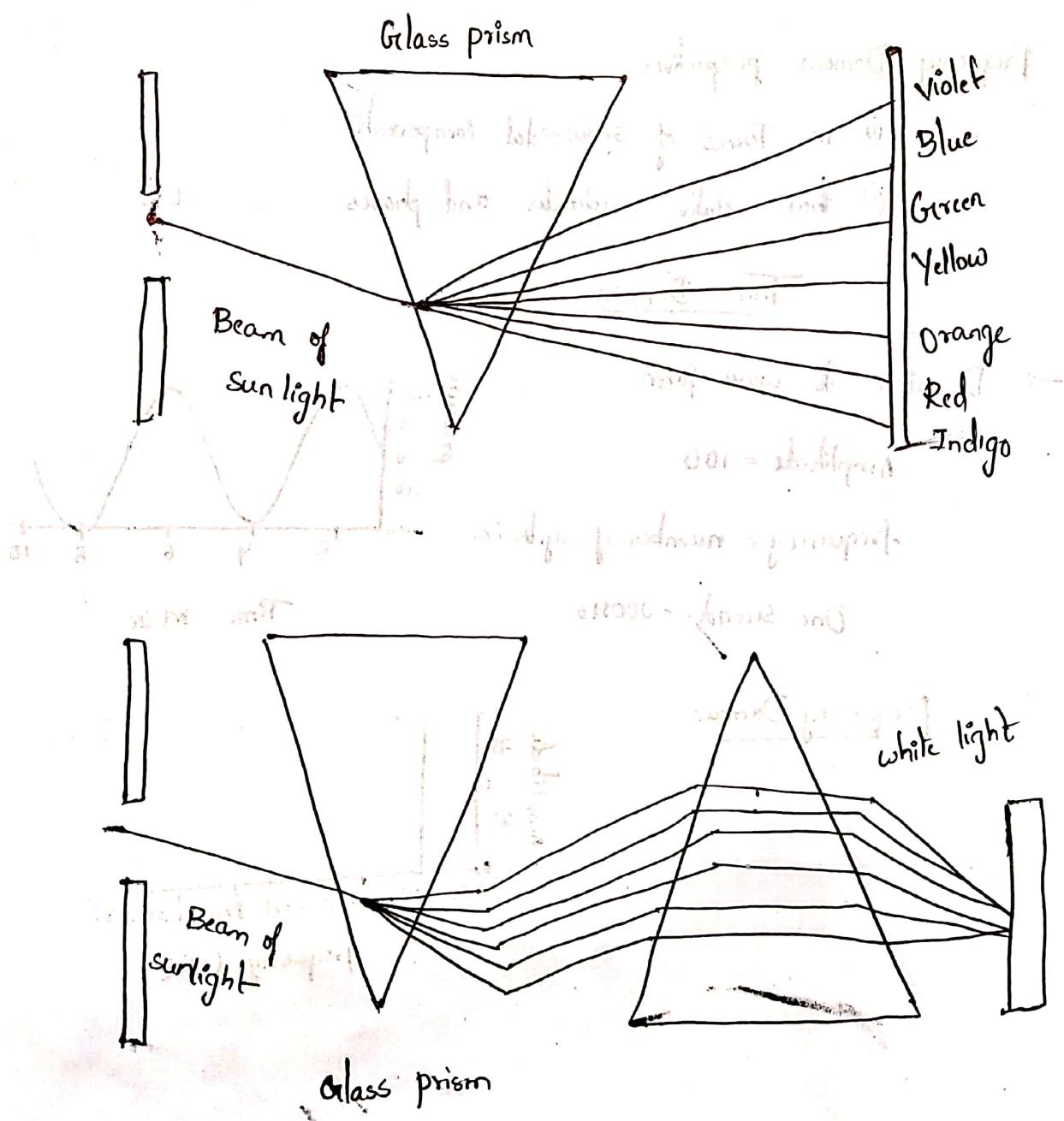


Frequency Domain



Fourier Series Representation

- Resolving the signal into its frequency (sinusoids) components is called as frequency analysis of the signal.
- The Decomposition of the signals in terms of sinusoidal or complex exponential components which are to be represented in.
- Which such a Decomposition, a signal is said to be represented in frequency Domain.



Fourier Series Representation

- in our treatment of frequency analysis, we will develop the proper mathematical tools (prism) for decomposition of the signals (light) into Sinusoidal frequency components (colours)
- the basic motivation for developing the frequency analysis tool is to provide a mathematical & pictorial representation for the frequency components that are contained in any given signal.

Fourier Series

- The basic mathematical representation of periodic signal is the Fourier series
- which is a linear weighted sum of harmonically related sinusoids or complex exponentials.
- The Decomposition of the periodic Signals into sinusoids or complex exponentials.

Fourier Transforms

- Such a Decomposition of the Non-periodic / Energy Signals is called Fourier transform representation

Fourier Series Representation of Continuous-time Period Signals

- A signal is periodic, if for some positive value of T , $x(t) = x(t+T)$ for all T
- the fundamental period of $x(t)$ is the minimum positive, non zero value of T for which the above equation is satisfied and the value $\omega_0 = 2\pi/T$ is referred to as the fundamental frequency or Angular frequency

Periodic Signals

Sinusoidal signals & periodic complex exponentials

$$x(t) = \cos \omega_0 t \quad \text{and} \quad x(t) = e^{j\omega_0 t}$$

- Both of these signals are periodic with fundamental frequency ω_0 & fundamental period $T = \frac{2\pi}{\omega_0}$

$$\text{Let } \phi_k(t) = e^{jk(2\pi/T)t}, \quad k = 0, \pm 1, \pm 2, \dots$$

is the set of harmonically related complex exponentials

- Each of these signals has a fundamental frequency that is a multiple of ω_0 and therefore each is periodic with period T .

- thus, a linear combination of harmonically related Complex Exponentials of the form.

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t} = \sum_{k=-\infty}^{\infty} a_k e^{jk(2\pi/T)t}$$

is also periodic with period T

The term for $k=0$ is a constant

- The terms for $k=+1$ & $k=-1$ both have fundamental frequency equal to ω_0 and are collectively referred to as fundamental components or the first harmonic.

- The two terms for $k=+2$ & $k=-2$ are periodic with half the period (or, equivalently, twice the frequency) of the fundamental components and are referred to as the second harmonic.
- more generally, the components for $k=+N$ & $k=-N$ are referred to as the N th harmonic components

→ thus a linear combination of harmonically related complex exponentials of the form

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{j k \omega_0 t} = \sum_{k=-\infty}^{+\infty} a_k c_k e^{j k (\omega_0 t + \phi_k)}$$

→ the representation of a periodic signal in the form of equation is referred to as the Fourier Series representation

Dirichlet's Condition

Condition-1 $x(t)$ is absolutely integrable over one period

(i.e.)

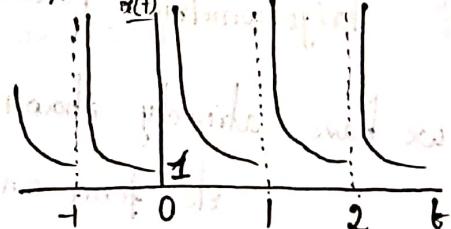
$$\int_{-\infty}^{\infty} |x(t)| dt < \infty$$

Condition-2 In a finite time interval, $x(t)$ has a finite number of maxima and minima

Condition-3 In a finite time interval, $x(t)$ has only a finite number of discontinuities

→ A periodic signal that violates the first Dirichlet condition is

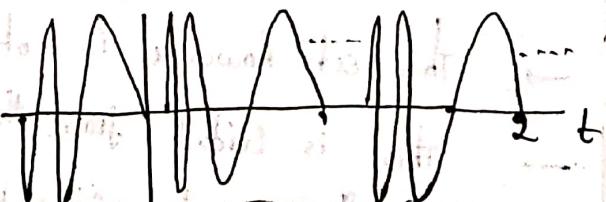
$$x(t) = \frac{1}{t}, \text{ for } 0 < t \leq 1$$



→ An example of a function that meets condition 1 but not

condition 2 is

$$x(t) = \sin\left(\frac{2\pi}{T} t\right) \text{ for } t \leq 1$$



with an infinite number of

maxima and minima in the interval

A Signal periodic with period 8 that violates the third
 Dirichlet condition (for $0 < t < 8$ the value of $x(t)$
 decreases by a factor of 2 whenever the distance from t to

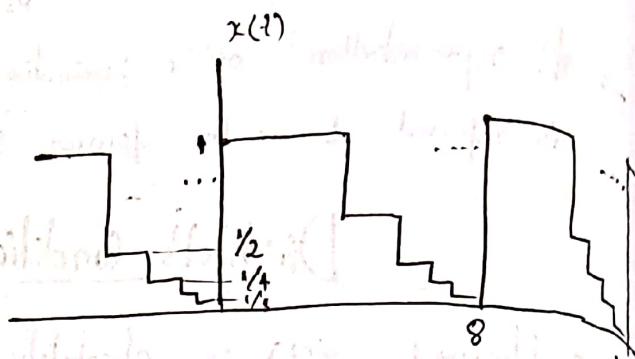
8 decreases by a factor 2

$$x(t) = 1, \quad 0 \leq t < 4$$

$$x(t) = \frac{1}{2}, \quad 4 \leq t < 6$$

$$x(t) = \frac{1}{4}, \quad 6 \leq t < 7$$

$$x(t) = \frac{1}{8}, \quad 7 \leq t < 7.5$$



Types of Fourier Series Representations

Two Different ways of Expressing the same periodic function They are

(i) Trigonometric Fourier Series (TFS) Representation

(ii) Exponential Fourier Series (EFS) Representation

Trigonometric Fourier Series (TFS) Representation

→ we have already shown that function $\sin \omega_0 t, \sin 2\omega_0 t, \dots$
 etc form an Orthogonal set over any interval

$$(t_0, t_0 + T), T = \frac{2\pi}{\omega_0}$$

→ this set however is not complete

→ this is evident from the fact a function $\cos \omega_0 t$ is
 orthogonal to $\sin \omega_0 t$ over the same interval.

→ Hence to complete the set, we must include cosine as well as
 sine functions

→ It can be shown that the composite set of functions consisting a set $\cos nw_0 t$ and $\sin nw_0 t$ for ($n = 0, 1, 2, \dots$) forms a complete orthogonal set.

→ Note that for $n=0$, $\sin nw_0 t$ is zero but $\cos nw_0 t$ is one thus we have a completed orthogonal set represented by functions $1, \cos w_0 t, \cos 2w_0 t, \dots, \cos nw_0 t; \sin w_0 t, \sin 2w_0 t, \dots, \sin nw_0 t, \dots$ etc

→ It therefore follows that any function $f(t)$ can be represented in terms of these functions over any interval $(t_0, t_0 + T) T = 2\pi/w_0$

thus.
$$f(t) = a_0 + a_1 \cos w_0 t + a_2 \cos 2w_0 t + \dots + a_n \cos nw_0 t + \dots + b_1 \sin w_0 t + b_2 \sin 2w_0 t + \dots + b_n \sin nw_0 t + \dots$$

$$(t_0 < t < t_0 + \frac{2\pi}{w_0}) \rightarrow 0$$

For convenience we shall denote $\frac{2\pi}{w_0}$ by T

The equation can be expressed as

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos nw_0 t + b_n \sin nw_0 t) \quad (t_0 < t < t_0 + T) \rightarrow 0$$

Equation 2 is the trigonometric Fourier series representation of

$f(t)$ over an interval $(t_0, t_0 + T)$

$$a_n = \frac{1}{T} \int_{t_0}^{t_0+T} f(t) \cos nw_0 t dt$$

$$b_n = \frac{1}{T} \int_{t_0}^{t_0+T} f(t) \sin nw_0 t dt$$

If we let $n=0$ in a_n we get

$a_0 = \frac{1}{T} \int_{t_0}^{(t_0+T)} f(t) dt$

We also have

$$\int_{t_0}^{(t_0+T)} \cos^2 n\omega_0 t dt = \int_{t_0}^{(t_0+T)} \sin^2 n\omega_0 t dt = \frac{T}{2}$$

therefore

$$a_n = \frac{2}{T} \int_{t_0}^{(t_0+T)} f(t) \cos n\omega_0 t dt$$

$$b_n = \frac{2}{T} \int_{t_0}^{(t_0+T)} f(t) \sin n\omega_0 t dt$$

The constant term a_0 in the series is given by

$$a_0 = \frac{1}{T} \int_{t_0}^{(t_0+T)} f(t) dt$$

The constant term a_0 is the average value of $f(t)$ over the interval $(t_0, t_0 + T)$. Thus a_0 is the DC component of $f(t)$.

Over this interval

Exponential Fourier Series (EFS) Representation

It can be shown easily that a set of exponential functions $\{e^{jn\omega_0 t}\}$, ($n = 0, \pm 1, \pm 2, \dots$) is orthogonal over an interval $(t_0, t_0 + 2\pi/\omega_0)$ for any value of t_0 . Note that this is a set of complex functions. We can demonstrate the orthogonality of this set by considering the integral

$$I = \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} (e^{jn\omega_0 t}) (e^{jm\omega_0 t})^* dt$$

$$= \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} e^{jn\omega_0 t} e^{-jm\omega_0 t} dt$$

If $n=m$ the integral in the integral I is given by

$$I = \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} dt = \frac{2\pi}{\omega_0}$$

If $n \neq m$ the integral I is given by

$$I = \frac{1}{j(n-m)\omega_0} e^{j(n-m)\omega_0 t} \Big|_{t_0}^{t_0 + \frac{2\pi}{\omega_0}}$$

$$= \frac{1}{j(n-m)\omega_0} e^{j(n-m)\omega_0 t} \left[e^{j2\pi(n-m)} - 1 \right]$$

Since both n and m are integers $e^{j2\pi(n-m)}$ is equal to

Unity, and hence the integral is zero and $I=0$

Thus $\int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} e^{jn\omega_0 t} (e^{jn\omega_0 t})^* dt = \begin{cases} \frac{2\pi}{\omega_0} & m=n \\ 0 & m \neq n \end{cases}$

$$\frac{2\pi}{\omega_0} = T$$

It is evident from eq that the set of functions

$\{e^{jn\omega_0 t}\}$ ($n = 0, \pm 1, \pm 2, \dots$) over the interval $(t_0, t_0 + T)$ where $T = \frac{2\pi}{\omega_0}$

is orthogonal. Over the interval $(t_0, t_0 + T)$ where $T = \frac{2\pi}{\omega_0}$

Further, it can be shown that this is a complete set. It is

therefore possible to represent an arbitrary function $s(t)$ by a linear combination of exponential functions over an interval $(t_0, t_0 + T)$.

This is a complete set. It is therefore possible to represent an arbitrary function $f(t)$ by a linear combination of Exponential functions over an interval (t_0, t_0+T) .

$$f(t) = F_0 + F_1 e^{j\omega_0 t} + F_2 e^{j2\omega_0 t} + \dots + F_n e^{jn\omega_0 t} + \dots + F_{-1} e^{-j\omega_0 t} + F_{-2} e^{-j2\omega_0 t} + \dots + F_{-n} e^{-jn\omega_0 t} + \dots$$

for

$$f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t} \quad (t_0 < t < t_0+T)$$

Known as Exponential Fourier Series representation of $f(t)$ over the interval (t_0, t_0+T) .

The Various Coefficients in Series $F_n = \int_{t_0}^{t_0+T} f(t) e^{jn\omega_0 t} dt$

is given by

$$\begin{aligned} F_n &= \int_{t_0}^{t_0+T} f(t) e^{jn\omega_0 t} dt \\ &= \frac{1}{T} \int_0^T f(t) e^{jn\omega_0 t} dt \end{aligned}$$

Summarizing the results - Any given function $f(t)$ may be expressed as a discrete sum of exponential functions $\{e^{jn\omega_0 t}\}$, ($n = 0, \pm 1, \pm 2, \dots$) over an interval $t_0 < t < t_0+T$ ($\omega_0 = \frac{2\pi}{T}$) in radians per second.

$$f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t} \quad (t_0 < t < t_0 + T)$$

$$F_n = \frac{1}{T} \int_{t_0}^{t_0 + T} f(t) e^{-jn\omega_0 t} dt$$

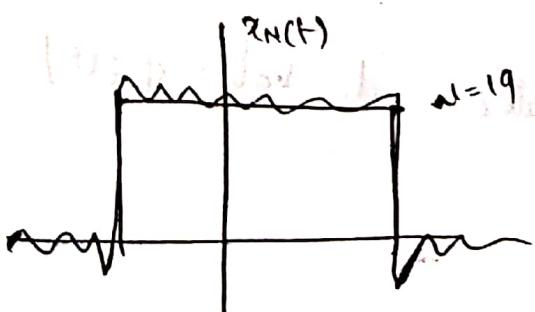
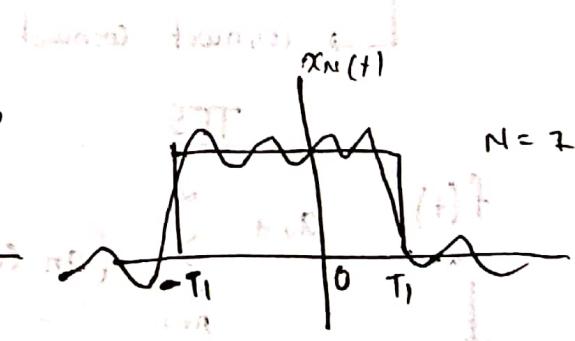
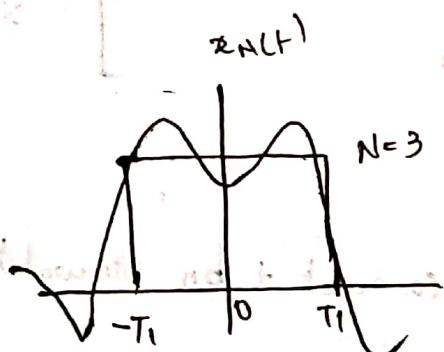
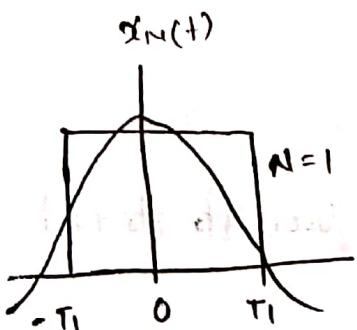
Relationship between TFS & EFS Representations

It should be noted that the trigonometric and the exponential Fourier Series are not two different types of Series but two different types of Series ways of Expressing the same Series. The coefficients of One Series can be Obtained from those of the other.

$$a_0 = F_0$$

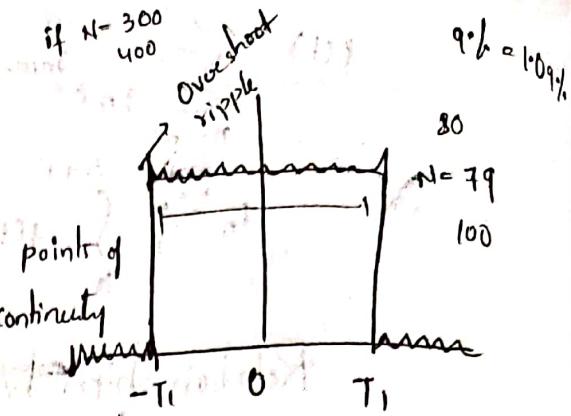
$$a_n = F_n + F_{-n} \quad \& \quad F_n = \frac{1}{2}(a_n - j b_n)$$

$$b_n = j(F_n - F_{-n})$$



GIBB's phenomenon

Convergence of the Fourier Series
representation of a Square wave



Fourier Series

Classification of Fourier Series

- (i) Set of mutually Orthogonal functions forms a Complete set
- then any function (periodic/non-periodic) can be Expressed in terms of these Orthogonal functions

\rightarrow Sinusoid, Cosinusoid \rightarrow orthogonal] Only Sinusoid \rightarrow not complete
 \rightarrow Sin, Sgn Only Cos X
 \rightarrow (Cos nωt, Cos nωt) Sin nωt, Cos nωt - complete

TFS

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega t + b_n \sin n\omega t)$$

Periodic

$$a_0 = \frac{1}{T} \int_{t_0}^{t_0+T} f(t) dt \rightarrow \text{avg value} = \text{dc value of } f(t)$$

$$a_n = \frac{2}{T} \int_{t_0}^{t_0+T} f(t) \cos n\omega t dt$$

$$b_n = \frac{2}{T} \int_{t_0}^{t_0+T} f(t) \sin n\omega t dt$$

EFS

→ Complex exponential function = $e^{j\omega t}$

set of complex exponential functions

$$= e^{j\omega t}, e^{j2\omega t}, \dots, e^{j(n-1)\omega t}$$

\downarrow
1st harmonic 2nd

3rd

$$N = [0, \pm 1, \pm 2, \dots]$$

$$= e^{-j\omega t}, e^{-j2\omega t}, \dots, e^{-j(n-1)\omega t}$$

TFS

$\sin, \cos \rightarrow$ form complete

$\frac{\sin \omega t}{\sin \omega t} \rightarrow$ set

whether these orthogonal or not

$$\phi_k(t) = e^{\pm j\omega_0 t}$$

→ check for Orthogonality of complex exponential

$$T = \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} f_1(t) f_2^*(t) dt = 0$$

$$f_1(t) = e^{j\omega_0 t} \quad f_2^*(t) = a + jb$$

$$f_2(t) = te^{j\omega_0 t} \quad f_2^*(t) = e^{-j\omega_0 t}$$

$$T = \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} e^{j\omega_0 t} \cdot e^{-j\omega_0 t} dt = \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} dt$$

$$= \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} (ae^{-jt}) dt + \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} (be^{-jt}) dt$$

$$= \frac{1}{(n-m)w_0} \left[e^{j(n-m)w_0 t + \frac{2\pi}{w_0}} - e^{-j(n-m)w_0 t} \right]$$

n, m are integers

$$f = \frac{1}{(n-m)w_0} \left[e^{j(n-m)w_0 t} \left(e^{\frac{j(n-m)2\pi}{w_0}} - 1 \right) \right] \quad e^{j2\pi k}$$

$$\frac{\cos k + j \sin k}{1}$$

$$f = \frac{1}{(n-m)w_0} \left[e^{j(n-m)w_0 t} \int_{t_1}^t f_1(t) e^{-jnw_0 t} dt \right]$$

Orthogonal set

$$f = 0 \quad \int_{t_1}^t f_1(t) f_2^*(t) dt = 0$$

(i) Orthogonal (ii) Complete $\rightarrow e^{jnw_0 t} \rightarrow n = 0, \pm 1, \pm 2, \pm 3, \dots$

$$e^{jnw_0 t} = \cos nw_0 t + j \sin nw_0 t \quad \text{complete set}$$

(iii) Any function can be expressed by this set of mutually

orthogonal complex exponential functions \rightarrow EFS

Sinusoids TFS

$$f(t) \cong f_0 + f_1 e^{jw_0 t} + f_2 e^{j2w_0 t} + \dots + f_n e^{jn w_0 t} + \dots$$

$$f_{-1} e^{-jw_0 t} + f_{-2} e^{-j2w_0 t} + \dots + f_{-n} e^{-jn w_0 t}$$

$$f_i(t) = f_0 + \sum_{n=0}^{\infty} f_n e^{jn w_0 t} \quad (n \neq 0)$$

$$f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t} \rightarrow \text{EFS of } (t_0, t_0 + \frac{2\pi}{\omega_0})$$

over one period

$$F_n = \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} f(t) \cdot e^{-jn\omega_0 t} dt$$

$$\frac{\int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} e^{jn\omega_0 t} \cdot e^{-jn\omega_0 t} dt}{\int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} dt} = 0$$

$$c_{12} = \frac{\int f_1(t) \cdot f_2^*(t) dt}{\int f_1(t) f_2^*(t) dt}$$

$$e^0 = \int f_1 dt$$

$$\Rightarrow F_n = \frac{1}{T} \int_{t_0}^{t_0 + T} f(t) e^{-jn\omega_0 t} dt \rightarrow \text{EFS of } f(t) \text{ over } T$$

$$n=0 \Rightarrow F_0 = \frac{1}{T} \int_{t_0}^{t_0 + T} f(t) dt \Rightarrow a_0 \text{ avg value}$$

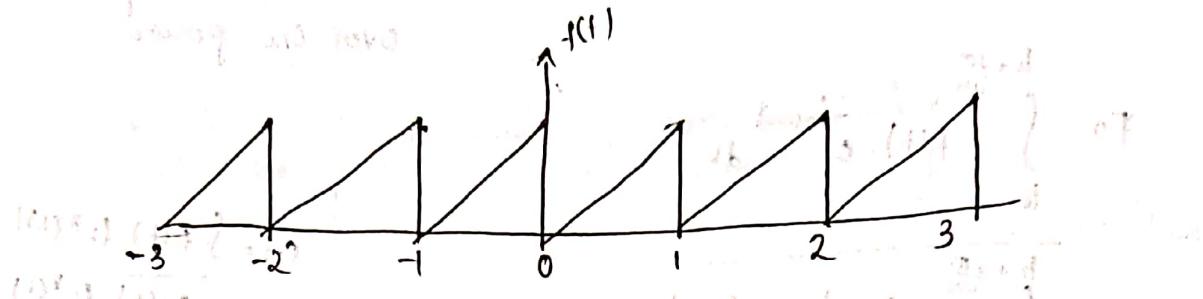
$$\text{DC value} = \frac{f_1 \cdot f_2(t)}{(1-e^{-j\omega_0 T})}$$

$$\text{EFS } f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t} \rightarrow$$

$$\text{where } F_n = \frac{1}{T} \int_{t_0}^{t_0 + T} f(t) e^{-jn\omega_0 t} dt$$

$$\text{TFS } f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega_0 t + b_n \sin n\omega_0 t)$$

Problem



$f(t) \rightarrow$ periodic

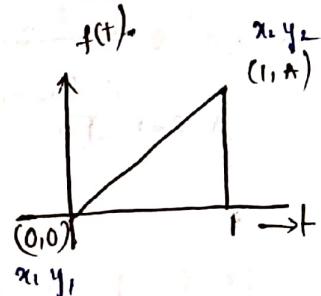
$$T = 1$$

$$\begin{aligned} 0-1 &= 1 \\ 2-1 &= 1 \\ 3-1 &= 1 \end{aligned}$$

Sol $TFS = f(t) \approx a_0 + \sum_{n=1}^{\infty} (\text{an cos} n\omega t + \text{bn sin} n\omega t)$

$$a_0 = \frac{1}{T} \int_0^T f(t) dt =$$

$$y - y_1 = m(x - x_1)$$



$$m = \frac{y_2 - y_1}{x_2 - x_1} \rightarrow \text{slope}$$

$$f(t) - 0 = \frac{1-0}{1-0}(t-0)$$

$$f(t) = At \quad (0 \leq t \leq 1)$$

$$a_0 = \frac{1}{T} \int_0^T At dt$$

$$A \int_0^1 t dt$$

$$A \left[\frac{t^2}{2} \right]_0^1$$

$$\left(\frac{At^2}{2} \right)_0^1$$

$$\left[\frac{At^2}{2} \right]$$

$$a_0 = \frac{A}{2}$$

$$a_n = \frac{2}{T} \int_0^T f(t) \cos(n\omega_0 t) dt$$

$$T = \frac{2\pi}{\omega_0} \Rightarrow \frac{2\pi}{T} = 100$$

$$= \frac{2}{T} \int_0^T A t \cos(n\omega_0 t) dt$$

$$S_{UV} = Uf_V - S_U f_V$$

$$S_{UV} = Uf_V - S_U f_V$$

$$= 2A \left[t \cdot \frac{\sin 2\pi nt}{2\pi n} - \int_0^t \frac{\sin 2\pi nt}{2\pi n} dt \right]$$

$$= 2A \left[t \cdot \frac{\sin 2\pi nt}{2\pi n} + \frac{\cos 2\pi nt}{(2\pi n)^2} \right]_0^T$$

$$a_n = 2A \left[\frac{1}{(2\pi n)^2} - \frac{1}{(2\pi n)^2} \right] = 0$$

$$2A \left(\frac{1}{2\pi n} - 0 - 0 \right)$$

$$\frac{A}{\pi n}$$

$$b_n = \frac{A}{\pi n}$$

$$b_n = \frac{2}{T} \int_0^T f(t) \sin(n\omega_0 t) dt$$

$$= 2A \left[t \cdot \frac{\cos 2\pi nt}{2\pi n} + \int_0^t \frac{\cos 2\pi nt}{(2\pi n)} dt \right] \left| \begin{array}{l} 2A \left(\frac{1}{2\pi n} - 0 - 0 \right) \\ -\frac{A}{\pi n} \end{array} \right.$$

$$= 2A \left[t \cdot \frac{\cos 2\pi nt}{2\pi n} + \frac{\sin 2\pi nt}{(2\pi n)^2} \right]_0^T \left| \begin{array}{l} b_n = \frac{A}{\pi n} \\ \end{array} \right.$$

$$a_0 = \frac{A}{2}$$

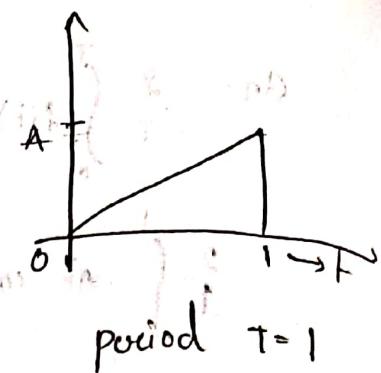
$$a_n = 0 \quad b_n = -\frac{A}{\pi n}$$

$$f(t) = \frac{A}{2} + \sum_{n=1}^{\infty} \frac{A}{\pi n} \sin(2\pi nt) \quad \text{for } (0 \leq t \leq 1)$$

$$f(t) = \frac{A}{2} + -\frac{A}{\pi n} \sum_{n=1}^{\infty} \sin(2\pi nt)$$

TFS

$$F(t) = \frac{A}{2} + \sum_{n=1}^{\infty} \left(\frac{A}{\pi n} \right) \sin(n\pi t)$$



$$F(t) = \frac{A}{2} - \frac{A}{\pi} \sum_{n=1}^{\infty} \frac{\sin(2\pi nt)}{n}$$

$$= \frac{A}{2} - \frac{A}{\pi} \sin(2\pi t) - \frac{A}{2} \sin(\pi t)$$

$$- \frac{1}{3} \sin(6\pi t) \quad \text{TTFS}$$

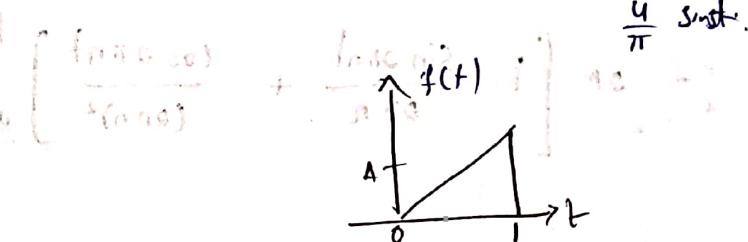
EFS?

$$f(t) = \sum_{n=-\infty}^{\infty} f_n e^{jn\omega_0 t}$$

$$f_0 + f_n = ?$$

$$f_0 = \frac{1}{T} \int_0^T f(t) dt$$

$$= \frac{1}{T} \int_0^1 At dt = \frac{A}{2} = a_0$$



$$0 = \left[\frac{1}{(a_0 + a_1)} + \frac{i}{(a_1 - a_0)} \right] j(t) \quad j(t) = At \quad \text{for } 0 < t < 1$$

$T=1$

$$\omega_0 = \frac{2\pi}{T} = 2\pi$$

$$f_n = \frac{1}{T} \int_0^T f(t) e^{-jn\omega_0 t} dt$$

$$f_n = \frac{1}{T} \int_0^1 At e^{-jn2\pi t} dt$$

$$= A \left[t \cdot \frac{e^{-jn2\pi t}}{-j2\pi n} - \int_0^1 \frac{e^{-jn2\pi t}}{(-j2\pi n)} dt \right]$$

$$= A \left[\frac{-t \cdot e^{-jn2\pi t}}{2j\pi n} \Big|_0^1 + \frac{e^{-jn2\pi t}}{(-j2\pi n)^2} \Big|_0^1 \right]$$

$$e^{-j2\pi n} = 1$$

$$f_n = -\frac{A}{j2\pi n}$$

$$f(t) = \frac{A}{2} - A \sum_{n=1}^{\infty} \frac{e^{-jn2\pi t}}{j2\pi n}$$

$$= A \left[\frac{-1}{j2\pi n} + \frac{1}{(2\pi n)^2} - \frac{0+1}{(2\pi n)^2} \right]$$

$$= A \left[-\frac{1}{j2\pi n} + \frac{1}{2\pi n^2} - \frac{1}{2\pi n^2} \right]$$

$$f(t) = \frac{A}{2} - \frac{A}{j2} \sum_{n=-\infty}^{\infty} \frac{e^{-jn\pi nt}}{\pi n} \quad \text{EFS}$$

$$f(t) = \frac{A}{2} + \frac{A}{2} \sum_{n=1}^{\infty} \frac{\sin(n\pi t)}{\pi n} \quad \text{TFS}$$

$$f_n = \frac{A}{j2\pi n}$$

Relationship between TFS & EFS

$$\text{EFS } f(t) = \sum_{n=-\infty}^{\infty} f_n e^{jn\omega_0 t} \quad \text{where}$$

$$f_n = \alpha_n + j\beta_n$$

$$f_n = \frac{1}{T} \int_0^T f(t) e^{-jn\omega_0 t} dt \rightarrow ① \quad F_1 = \omega_1 + j\beta_1$$

$$F_{-1} = \omega_{-1} + j\beta_{-1}$$

$$F_n = \frac{1}{T} \int_0^T f(t) e^{-jn\omega_0 t} dt \rightarrow ②$$

$$f_n = F_n^*$$

$$\rightarrow f(t) = F_0 + F_1 e^{j\omega_0 t} + F_2 e^{j2\omega_0 t} + \dots \quad F_n e^{jn\omega_0 t}$$

$$F_1 e^{-j\omega_0 t} + F_2 e^{-j2\omega_0 t} + \dots \quad F_n e^{-jn\omega_0 t}$$

$$f_n = \alpha_n + j\beta_n = F_0 + (\omega_1 + j\beta_1) e^{j\omega_0 t} + (\omega_2 + j\beta_2) e^{j2\omega_0 t} + \dots + (\alpha_n + j\beta_n) e^{jn\omega_0 t}$$

$$F_n = \alpha_n - j\beta_n \quad \text{and} \quad f_n = \alpha_n + j\beta_n$$

$$+ (\alpha - j\beta_1) e^{-j\omega_0 t} + (\alpha_2 - j\beta_2) e^{-j2\omega_0 t} + \dots + (\alpha_n - j\beta_n) e^{-jn\omega_0 t}$$

$$= F_0 + \alpha_1 (e^{j\omega_0 t} + e^{-j\omega_0 t}) + \alpha_2 (e^{j2\omega_0 t} + e^{-j2\omega_0 t}) + \dots$$

$$+ j\beta_1 (e^{j\omega_0 t} - e^{-j\omega_0 t}) + j\beta_2 (e^{j2\omega_0 t} - e^{-j2\omega_0 t}) + \dots$$

$$= F_0 + \alpha_1 \cos(\omega_0 t) + \alpha_2 \cos(2\omega_0 t) + \dots + j\beta_1 \sin(\omega_0 t) + j\beta_2 (\sin(2\omega_0 t)) + \dots$$

$$+ j\beta_1 \sin(\omega_0 t) + j\beta_2 (\sin(2\omega_0 t)) + \dots$$

$$= F_0 + \sum_{n=1}^{\infty} (2\alpha_n \cos(n\omega_0 t) - 2\beta_n \sin(n\omega_0 t)) \xrightarrow{EFS}$$

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos(n\omega_0 t) + b_n \sin(n\omega_0 t)) \rightarrow ④$$

$$a_0 = F_0$$

$$\alpha_n = 2\alpha_n$$

$$b_n = -2\beta_n$$

$$= j(F_n - F_{-n})$$

$$F_n = 2\alpha_n j\beta_n$$

$$F_{-n} = 2n - j\beta_n$$

$$F_n + F_{-n} = 2\alpha_n$$

$$F_n + F_{-n}, F_0 \rightarrow$$

$$a_n = F_n + F_{-n}$$

$$b_n = j(F_n - F_{-n})$$

$$F_n - F_{-n} = 2j\beta_n$$

$$F_n = -\frac{A}{j2\pi n}$$

$$F_n = \frac{-A}{j2\pi n} + \frac{A}{j2\pi n}$$

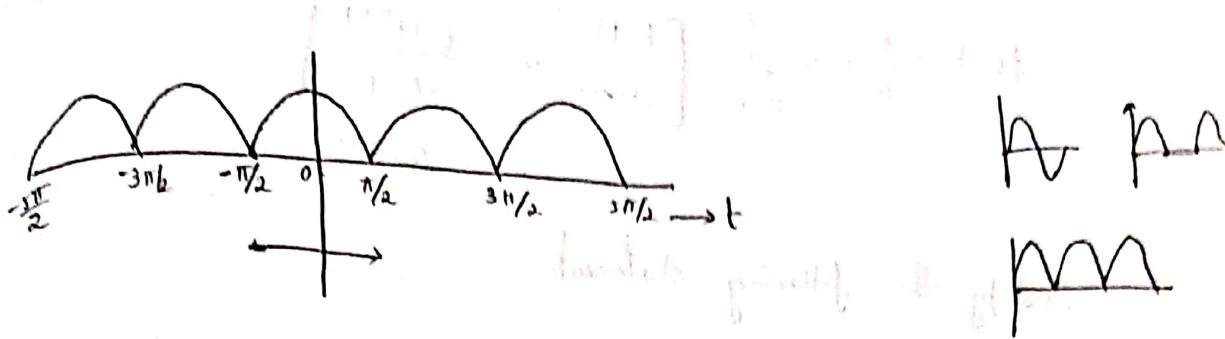
$$F_n = \frac{-A}{j2\pi n}$$

$$F_n = \frac{-A}{j2\pi n(n)} = \frac{A}{j2\pi n} \Rightarrow F_n + F_{-n} = 0 = a_n$$

$$j(F_n - F_{-n}) = j\left(\frac{-A}{j2\pi n} - \frac{A}{j2\pi n}\right)$$

$$= j\left(\frac{-2A}{j2\pi n}\right) = \frac{-A}{\pi n} = b_n$$

Full wave rectified cosine function



Period $T = \pi$ and one cycle starts at $t = 0$

$$f(t) = a_0 + \sum a_n \cos(n\omega_0 t) + b_n \sin(n\omega_0 t)$$

$$a_0 = \frac{1}{T} \int_0^T f(t) dt$$

$$= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \cos t dt$$

$$= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \sin t dt$$

$$= \frac{2}{\pi} \left(\sin \frac{\pi}{2} - \sin \left(-\frac{\pi}{2} \right) \right) = \frac{2}{\pi} (1 - (-1)) = \frac{4}{\pi} = 2$$

$$\omega_0 = \frac{2\pi}{T}$$

$$a_n = \frac{1}{T} \int_{-\pi/2}^{\pi/2} f(t) \cos(n\omega_0 t) dt$$

$$= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \cos t \cos(2n\pi) dt$$

$$= 2 \cos A \cos B = A(A+B) + \cos(A-B)$$

$$\frac{1}{T} \int_{-\pi/2}^{\pi/2} \frac{\cos((1+2n)t) + \cos((2n-1)t)}{2} dt$$

$$\frac{1}{2\pi} \int_{-\pi/2}^{\pi/2} (\cos((1+2n)t) dt \cdot \frac{1}{2\pi} \int_{-\pi/2}^{\pi/2} \cos((2n-1)t) dt$$

$$a_n = \frac{2}{\pi} \left[\frac{(-1)^n}{2n+1} + \frac{(-1)^{n+1}}{2n-1} \right]$$

$$b_n = 0 \rightarrow \text{odd function} \rightarrow \text{fourier series} (-\pi/a \leq t \leq \frac{\pi}{a})$$

$$f(t) = \frac{a_0}{2} + 2 \sum_{n=1}^{\infty} \left[\frac{a_n}{2n+1} + \frac{b_n}{2n-1} \right]$$

(cos 2nt)

Justify the following statements

- (i) Odd function have Only Sin terms ($a_n = 0, b_n \neq 0$)
- (ii) Even function have only cosine terms ($a_n \neq 0, b_n = 0$)
- (iii) function with wave symmetry having only odd harmonics
($a_n \neq 0, b_n \neq 0$ but $n = \text{odd}$)

Definition → Even function → $f(t) = f(-t)$ (cosine)
 $f(-t) = f(t) \rightarrow$ symmetrical about vertical axis

Odd → $f(t) \rightarrow f(-t) = -f(t) \rightarrow$ Anti symmetric about vertical axis

Even \times even = even

even \times odd = odd

odd \times odd = even

We know that

- ↳ cosnwot → is an even function
- ↳ sinnwot → is an odd function

$f(t) \rightarrow$ periodic

$$I_2 \int_{-T}^T f_e(t) dt = \int_0^T f_e(t) dt + \int_{-T}^0 f_e(t) dt$$

$t = -t$

$$\left[\frac{(1-t)}{2} + \frac{(1+t)}{2} \right] dt$$

$$I = \int_0^T f_e(t) dt + \int_0^T f_o(t) dt = 2 \int_0^T f_e(t) dt$$

$I = f(t) = \text{periodic}, f_0(t)$

$$I = \int_0^T f_0(t) dt$$

$$I = \int_{-T}^0 f_0(t) dt + \int_0^T f_0(t) dt$$

$$I = \int_{-T}^0 f_0(-t) dt + \int_0^T f_0(t) dt$$

$$= \int_0^T f_0(t) dt + \int_0^T f_0(t) dt$$

$$I = 0 \quad \begin{cases} \text{Even} & \int_0^T f_e(t) dt = 2 \int_0^T f_e(t) dt \\ \text{Odd} & \int_0^T f_o(t) dt = 0 \end{cases}$$

$f(t) \Rightarrow \text{even function} - f_e(t) \rightarrow$

$$f_e(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos \omega n t + b_n \sin \omega n t) \rightarrow \text{TFS}$$

$$a_n = \frac{2}{T} \int_{-T/2}^{T/2} f_e(t) \cos \omega n t dt$$

$\downarrow \text{Even} \times \text{Even} \rightarrow \text{Even function}$

$$\text{so } f_e(t) \approx \text{area } T/2$$

$$a_0 = \frac{2}{T} \int_{-T/2}^{T/2} f(t) dt$$

periodic wave plot $T/2$

$$a_n = \frac{4}{T} \int_0^{T/2} f_e(t) \cdot \cos \omega n t dt$$



$$b_n = \frac{2}{T} \int_{-T/2}^{T/2} f_c(t) \sin n\omega_0 t dt$$

even \times odd \rightarrow odd

$$I = \int_{-T/2}^{T/2} \text{odd function} dt = 0$$

$$b_n = 0$$

(i.e.) $f_c(t)$ for Even $f_n \rightarrow$ only cosine, Terms are existing.

$f(t) = f_0(t) \rightarrow$ periodic \rightarrow TFS/EFS

$$f_0(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega_0 t + b_n \sin n\omega_0 t)$$

$$a_n = \frac{2}{T} \int_{-T/2}^{T/2} f_0(t) \cos n\omega_0 t dt = 0$$

$$\text{odd} \times \overline{\text{even}} \rightarrow \text{odd}$$

$$b_n = \frac{2}{T} \int_{-T/2}^{T/2} f_0(t) \sin n\omega_0 t dt$$

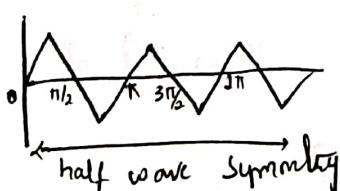
odd \times odd

For odd function \leftarrow (i) if \rightarrow odd and \leftarrow (ii)

$$= \frac{4}{T} \int_0^{T/2} f_0(t) \sin n\omega_0 t dt$$

for odd function have Only Sine terms, ($b_n \neq 0, a_n = 0$)

functions with half wave symmetry



$$a_n = \frac{2}{T} \int_0^T f(t) \cos n\omega_0 t dt$$

$$\text{even} = f(t) = f(-t)$$

$$\text{odd} = f(-t) = -f(t)$$

for half wave symmetry

$$f(t \pm T/2) = -f(t)$$

$$f(t) = -f(t \pm T/2)$$

$$= (T/2 + T/2)$$

$$a_n = \frac{2}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \, dt + \frac{2}{T} \int_{T/2}^T f(t) \cos n\omega_0 t \, dt$$

$$= \frac{2}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \, dt + \int_{T/2}^T f(t + T/2) \cos n\omega_0 (t + T/2) \, dt$$

$$f(t \pm T/2) = -f(t)$$

$$a_n = \frac{2}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \, dt - \frac{2}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \cos n\pi t \, dt$$

$n = \text{even}$ $T = 0$

$n = \text{odd}$

$$= \frac{4}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \, dt$$

odd harmonics

$b_n = 0$ for $n = \text{even}$

$$= \frac{4}{T} \int_0^{T/2} f(t) \sin n\omega_0 t \, dt = \text{odd value of } n$$

Half wave symmetry $\rightarrow a_n = 0$ for $n = \text{even}$

$$\frac{4}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \, dt \quad \text{for } n = \text{odd}$$

$b_n = 0$ for $n = \text{even}$

$$= \frac{4}{T} \int_0^{T/2} f(t) \sin n\omega_0 t \, dt \quad \text{for } n = \text{odd}$$

have only odd harmonics

even \rightarrow cosine terms ($a_n \neq 0$) $b_n = 0$

odd \rightarrow sine terms ($b_n \neq 0$)

EFS (co-efficients) $f(t) = \sin \omega_0 t$

$$f(t) = \sum_{n=-\infty}^{\infty} f_n e^{j\omega_0 t} \quad (t_0, t_0+T)$$

$$= F_0 e^{-j\omega_0 t} + F_1 e^{-j\omega_0 t} + F_0 \rightarrow \text{dc value / arg value}$$

$$F_2 e^{j\omega_0 t} + F_1 e^{j\omega_0 t} \quad (\text{1st harmonic})$$

$$\begin{aligned} & \downarrow \\ & \text{1st harmonic} \quad \downarrow \text{1st harmonic} \quad \text{TFS} \rightarrow a_0, a_n, b_n \\ & \text{2nd harmonic} \quad \downarrow \quad \text{EFS} \quad \text{co-efficients} \end{aligned}$$

$$\text{EFS} = \frac{F_n}{2} \quad \text{co-efficients}$$

$F_0 \rightarrow \text{dc component}$

$$a = \frac{F_0}{2} \quad e^{j\omega_0 t}, e^{-j\omega_0 t} \rightarrow \text{1st harmonic} \quad \text{EFS co-efficient}$$

$$\frac{F_2 - F_{-2}}{2} \quad e^{j\omega_0 t}, e^{-j\omega_0 t} \rightarrow \text{EFS} \quad |F_n|$$

Same in magnitude different phase

$$F_2, F_{-2} \rightarrow \text{EFS coefficients}$$

$$F_n, F_{-n} \rightarrow \text{EFS coefficients}$$

$$\frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j}$$

$$= \frac{1}{2j} e^{j\omega_0 t} - \frac{1}{2j} e^{-j\omega_0 t}$$

$$F_0 = \frac{1}{2j} \quad \text{1st harmonic} \quad f_0 \rightarrow \text{dc}$$

$$F_1 = -\frac{1}{2j} \quad \left[\frac{F_2 - F_{-2}}{2} \right] \quad \text{Zero}$$

$$|F_1| = |F_{-1}|$$

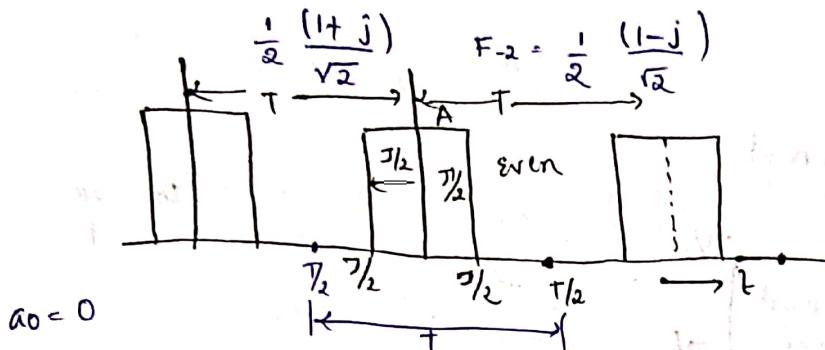
$$* f(t) = 1 + \sin \omega_0 t + 2 \cos \omega_0 t + \cos(2\omega_0 t + \pi/4)$$

$$1 + \frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j} + \frac{j(e^{j\omega_0 t} + e^{-j\omega_0 t})}{2} + \frac{e^{j\omega_0 t + \pi/4} + e^{-j\omega_0 t + \pi/4}}{2}$$

$$1 + \left(1 + \frac{1}{2j}\right) e^{j\omega_0 t} + \left(1 - \frac{1}{2j}\right) e^{-j\omega_0 t} + \frac{1}{2} e^{j2\omega_0 t} \cdot e^{j\pi/4} + \frac{1}{2} e^{-j2\omega_0 t} \cdot e^{-j\pi/4}$$

$$F_0 = 1, \quad F_1 = 1 + \frac{1}{2j}, \quad F_{-1} = \left(1 - \frac{1}{2j}\right)$$

$$F_2 = \frac{1}{2} e^{j\pi/4}, \quad F_{-2} = \frac{1}{2} e^{-j\pi/4}$$



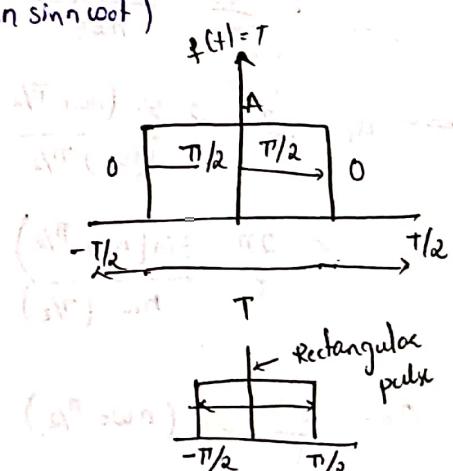
$$\int_{-\pi/2}^{\pi/2} f(t) = A \begin{cases} \text{for } -\pi/2 < t < \pi/2 \\ = 0 \text{ otherwise} \end{cases}$$

IFS

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega_0 t + b_n \sin n\omega_0 t)$$

$$a_0 = \frac{1}{T} \int_{-\pi/2}^{\pi/2} f(t) dt$$

$$f(t) = A \begin{cases} (-\pi/2, \pi/2) \\ (-\pi/2 < t < \pi/2) \\ = 0 \text{ otherwise} \end{cases}$$



$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega_0 t + b_n \sin n\omega_0 t)$$

$$a_0 = \frac{1}{T} \int_{-\pi/2}^{\pi/2} f(t) dt$$

$$f(t) = A \begin{cases} (-\pi/2, \pi/2) \\ (-\pi/2 < t < \pi/2) \\ = 0 \text{ otherwise} \end{cases}$$

0 otherwise

$$a_0 = \frac{1}{T} \int_{-\pi/2}^{\pi/2} A dt > \frac{At}{T} \int_{-\pi/2}^{\pi/2} dt = \frac{At}{T} = a_0 \quad \begin{matrix} T = \text{width} \\ T = \text{period} \end{matrix}$$

$$\begin{aligned} a_n &= \frac{2}{T} \int_{-\pi/2}^{\pi/2} f(t) \cos n\omega_0 t dt \\ &= \frac{2}{T} \int_{-\pi/2}^{\pi/2} A \cdot \cos n\omega_0 t dt \\ &= \frac{2A}{T} \left[\frac{\sin n\omega_0 t}{n\omega_0} \right]_{-\pi/2}^{\pi/2} \end{aligned}$$

$$= \frac{2A}{T} \left[\frac{\sin n\omega_0 \pi/2 + \sin n\omega_0 (-\pi/2)}{n\omega_0} \right]$$

$$= \frac{4A}{T} \frac{\sin n\omega_0 \pi/2}{n\omega_0}$$

$$\Rightarrow a_n = \frac{2}{T} \frac{2 \sin(n\omega_0 \pi/2)}{(n\omega_0 \pi/2)} \times \frac{\pi/2}{\pi/2}$$

$$a_n = \frac{2T}{T} \text{sa}(n\omega_0 \pi/2)$$

$$b_n = \frac{2}{T} \int_{-\pi/2}^{\pi/2} f(t) \sin n\omega_0 t dt$$

$$= 0$$

$$f(t) = \frac{At}{T} + \sum_{n=1}^{\infty} \frac{2T}{T} \text{sa}(n\omega_0 \pi/2) \cdot \cos(n\omega_0 t)$$

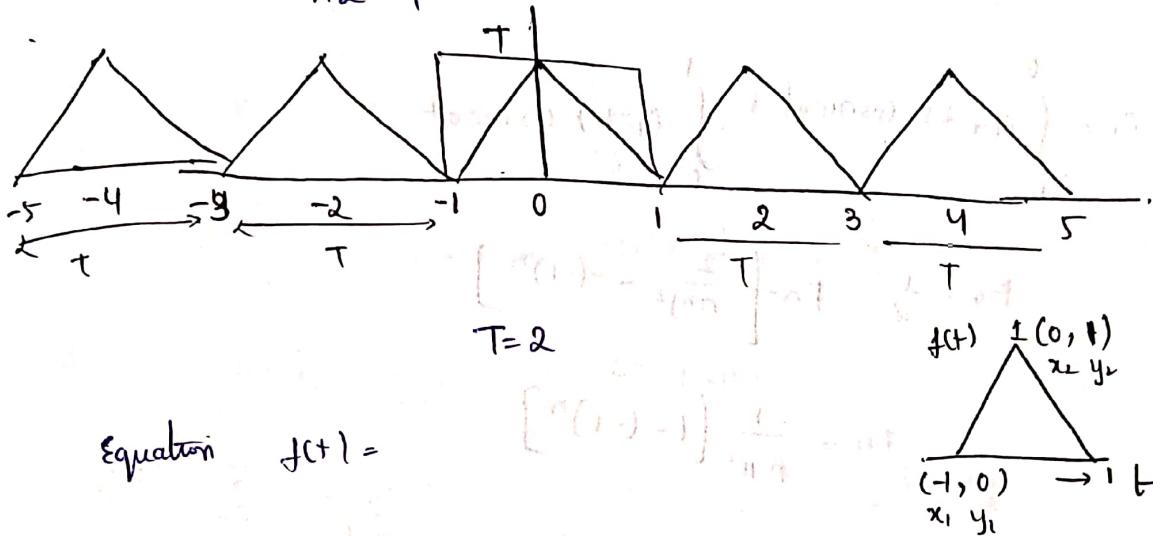
$$\underline{EFS} = f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t}$$

$$F_n = \frac{1}{T} \int_0^T f(t) e^{-jn\omega_0 t} dt$$

$$F_n = \frac{A\pi}{T} \operatorname{sa}(n\omega_0 T/2)$$

$$\frac{\sin x}{x} = \operatorname{sa}(x)$$

$$f(t) = \sum_{n=-\infty}^{\infty} \frac{A\pi}{T} \operatorname{sa}(n\omega_0 T/2) e^{jn\omega_0 t}$$



$$\frac{y - y_1}{x - x_1} = \frac{y_2 - y_1}{x_2 - x_1} \Rightarrow f(t) - 0 = 1 \left[\frac{t - 0}{1 - 0} \right] \Rightarrow f(t) = t$$

$$= \frac{1-0}{0+1}$$

$$f(t) = 1+t \quad -1 < t < 0$$

$$\frac{f(t)-1}{t-0} = \frac{0-1}{1-0}$$

$$f(t) - 1 = -t$$

$$f(t) = 1-t \quad 0 \leq t < 1$$

$$f(t) = 1-t \quad \text{for } 0 < t < 1$$

$$f(t) = 1+t \quad -1 < t < 0$$

$$f(t) = \int -|t|$$

$$a_0 = \frac{1}{T} \int_{-T}^T f(t) dt$$

$$a_n = \frac{2}{T} \int_{-T}^T f(t) \cos n\omega_0 t dt$$

$$b_n = \frac{2}{T} \int_{-T}^T f(t) \sin n\omega_0 t dt$$

$$a_n = \int_{-1}^0 (1+t) \cos n\omega_0 t + \int_0^1 (1-t) \cos n\omega_0 t$$

$$F_0 = \frac{1}{2}, \quad F_n = \left[\frac{2}{n\pi} - (-1)^n \right]$$

$$f_n = \frac{1}{n^2\pi^2} [1 - (-1)^n]$$

$$f(t) = \frac{1}{2} + \sum_{n=1}^{\infty} \frac{1}{n^2\pi^2} [1 - (-1)^n] e^{\frac{j n \omega_0 t}{\pi}}$$

$$\omega_0 = \frac{2\pi}{T}$$

Properties of Fourier Series

$$f(t) = \sum_{n=-\infty}^{\infty} f_n e^{jn\omega_0 t}$$

$$f(t) \xrightarrow{\text{f.s.}} f_n$$

$$f_n = \frac{1}{T} \int_0^T f(t) e^{-jn\omega_0 t} dt$$

$$f(t) \xrightarrow{\text{f.s.}} f_n$$

If $f(t) \rightarrow$ periodic $\rightarrow T$

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega_0 t + b_n \sin n\omega_0 t)$$

$$f(t) = \sum_{n=-\infty}^{\infty} f_n e^{jn\omega_0 t}$$

Linearity property of F.S

Linearity → two properties → additivity
→ homogeneity (or) scaling

$$x(t) \xrightarrow{F.S} a_n$$

$$y(t) \xrightarrow{F.S} b_n$$

$$x(t) + y(t) \xrightarrow{F.S} a_n + b_n \quad ? \quad \text{additivity } Ax(t) + By(t)$$

$$x(t) \xrightarrow{F.S} a_n$$

$$\text{scalar 'A'} \quad x(t) \xrightarrow{F.S} A \cdot a_n$$

$$\text{If } x(t) \xrightarrow{F.S} a_n \text{ & } y(t) \xrightarrow{F.S} b_n \text{ then}$$

$$z(t) = Ax(t) + By(t) \rightarrow Aa_n + Bb_n$$

$$f(t) = \sum_{n=0}^{\infty} f_n e^{jnw_0 t} \quad f_n = \frac{1}{T} \int_0^T f(t) e^{-jnw_0 t} dt$$

$$\text{Let } z(t) = Ax(t) + By(t)$$

$$z(t) \xrightarrow{F.S} c_n$$

$$c_n = \frac{1}{T} \int_0^T z(t) e^{-jnw_0 t} dt$$

$$c_n = \frac{1}{T} \int_0^T (Ax(t) + By(t)) e^{-jnw_0 t} dt$$

$$c_n = \frac{1}{T} \int_0^T Ax(t) e^{-jnw_0 t} dt + \frac{1}{T} \int_0^T By(t) e^{-jnw_0 t} dt$$

$$c_n = A \frac{1}{T} \int_0^T x(t) e^{-jnw_0 t} dt + B \frac{1}{T} \int_0^T y(t) e^{-jnw_0 t} dt$$

$$c_n = Aa_n + Bb_n$$

$$z(t) \xrightarrow{F.S} c_n$$

$$Ax(t) + By(t) \xrightarrow{\text{F.S}} Aa_n + Bb_n$$

linearity property

$$x(t) \xrightarrow{\text{F.S}} a_n$$

$$y(t) \xrightarrow{\text{F.S}} b_n$$

$$Ax(t) + By(t) \rightarrow Aa_n + Bb_n$$

$$A = B = 1$$

$$x(t) = \sum_{n=-\infty}^{\infty} a_n e^{j\omega_0 t} \rightarrow \textcircled{1}$$

$$a_n = \frac{1}{T} \int x(t) e^{-jn\omega_0 t} dt \rightarrow \textcircled{2}$$

analysis $\rightarrow x(t)$

synthesis \rightarrow combining all the frequency a_n

Eq. ① synthesis a_1

Eq. ② analysis a_2

② Time shifting property of F.S

if $x(t) \rightarrow$ periodic with period $T \xrightarrow{\text{F.S}} a_n$ then if $x(t)$ is shifted by t_0 seconds $\rightarrow x(t-t_0) \rightarrow ?$

$$a_n = \frac{1}{T} \int x(t) e^{-jn\omega_0 t} dt$$

$$b_n \rightarrow x(t-t_0) \Rightarrow$$

$$\text{Let } t-t_0 = T$$

$$\Rightarrow t = (t_0 + T)$$

$$t = (t_0 + T) \quad \text{if } x(t) \text{ is periodic}$$

$$dt = dT$$

$$b_n = \frac{1}{T} \int_{-T}^T x(\tau) e^{-jn\omega_0 (\tau+t_0)} d\tau$$

$$b_n = \frac{1}{T} \int_{-T}^T x(\tau) e^{-jn\omega_0 \tau} \cdot e^{-jn\omega_0 t_0} d\tau$$

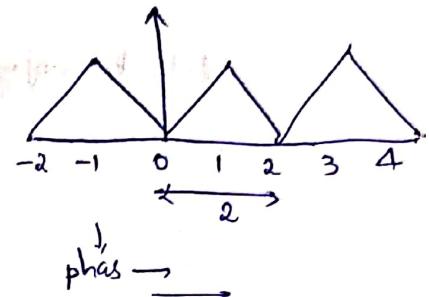
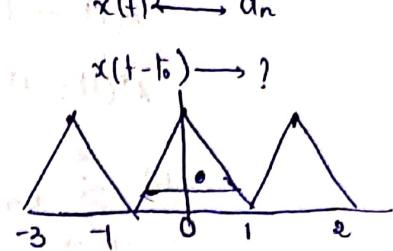
$$= e^{-jn\omega_0 t_0} \frac{1}{T} \int_{-T}^T x(\tau) e^{-jn\omega_0 \tau} d\tau$$

$$b_n = e^{-jn\omega_0 t_0} a_n$$

both $x(t)$ & $x(t-t_0)$ → same magnitude

but $x(t-t_0)$ phase change

$$e^{-jn\omega_0 t_0}$$



$$x(t) \xrightarrow{\text{F.S.}} a_n$$

$$x(t-t_0) \xrightarrow{\text{F.S.}} e^{jn\omega_0 t_0} a_n$$

$$\text{mag } x(t-t_0) |e^{jn\omega_0 t_0}| |a_n|$$

$$|e^{jn\theta}| = 1 \quad a_n = a_n$$

$$\text{mag } |x(t)| = |a_n| =$$

2) * Time Reversal

$$x(t) \rightarrow T \rightarrow a_n$$

$$x(t) \rightarrow T \rightarrow a_{-n}$$

$$x(t) = \sum_{n=-\infty}^{\infty} a_n e^{jn\omega_0 t}$$

$$x(t) = \text{odd} \Rightarrow x(-t) = -x(t)$$

$$a_{-n} = -a_n$$

$$\text{let } [-n=m] \Rightarrow n=-m$$

$$x(-t) = \sum_{m=-\infty}^{\infty} a_m e^{jm\omega_0 t}$$

$$x(t) \rightarrow a_n$$

$$\text{If } x(t) = \text{even} \Rightarrow x(-t) = x(t)$$

$$a_{-n} = a_n$$

3)

Conjugation & Conjugation Symmetry

$$x(t) \rightarrow T \rightarrow a_n \text{ then}$$

$$x^*(t) \rightarrow T \rightarrow ?$$

$$x(t) = \sum_{n=-\infty}^{\infty} a_n e^{j n \omega t}$$

$$\text{take the conjugation both sides } x^*(t) = \sum_{n=-\infty}^{\infty} a_n^* e^{-j n \omega t}$$

$$= \sum_{m=1}^{\infty} a_m^* e^{j m \omega t}$$

$$x^*(t) = \sum_{n=-\infty}^{\infty} a_n^* e^{-j n \omega t} \quad (e^{j\theta}) \rightarrow \cos \theta + j \sin \theta$$

$$(e^{j\theta}) \rightarrow \cos \theta + j \sin \theta$$

$$\text{so } x^*(t) \rightarrow a_n^*$$

$$\text{If } x(t) \Rightarrow \text{real} \quad x^*(t) \rightarrow \text{real}$$

$$\text{both } x(t) = x^*(t)$$

$$x = a + jb$$

$$x^* = a - jb$$

$$y^* = b$$

$$\text{and } a = a^* \rightarrow a_n = a_n^* \rightarrow a^* = a_n$$

$$\text{if } x(t) \rightarrow \text{real} = a_n = a_n^* \quad (x(t) = x^*(t))$$

$$(1) x(-t) = \text{conjugate of } x(t) \\ \text{and } a_n = a_{-n} \Rightarrow \boxed{a_n^* = a_{-n}}$$

Conjugate Symmetric

$$|a_n| = |a_n^*| =$$

$$\text{If } x(t) \rightarrow \text{real + Even} \Rightarrow x(-t) = x(t)$$

$$a_{-n} = a_n$$

$$\boxed{a_n = a_n^* = a_{-n}}$$

Parserval's Relation of periodic signals

$$x(t) \rightarrow T \rightarrow a_n$$

parseval's \rightarrow periodic \rightarrow power signals

$$\text{avg power } P_{\text{avg}} = \frac{1}{T} \int_0^T |x(t)|^2 dt$$

avg power in time domain

periodic \rightarrow F.S \rightarrow a_n

$$P_{\text{avg}} = \frac{1}{T} \int_0^T |x(t)|^2 dt$$

$$= \frac{1}{T} \int_0^T x(t) x^*(t) dt$$

we know that $x(t) = \sum_{n=-\infty}^{\infty} a_n e^{jnw_0 t}$

$$x^*(t) = \sum_{n=-\infty}^{\infty} a^*(n) e^{-jnw_0 t}$$

$$P_{\text{avg}} = \frac{1}{T} \int_0^T x(t) \left(\sum_{n=-\infty}^{\infty} a^*(n) e^{-jnw_0 t} \right) dt$$

$$= \sum_{n=-\infty}^{\infty} a^*(n) \frac{1}{T} \int_0^T x(t) e^{-jnw_0 t} dt$$

$$= \sum_{n=-\infty}^{\infty} a^*(n) a_n = \sum_{n=-\infty}^{\infty} |a_n|^2 = |a_0|^2 + |a_1|^2 + |a_2|^2 + \dots + |a_n|^2$$

$$\frac{1}{T} \int_0^T |x(t)|^2 dt = \sum_{n=-\infty}^{\infty} |a_n|^2$$

→ avg power in the periodic signal Equals to

→ avg power in all of its harmonics components

The sum of the avg powers in all of

$$\boxed{\frac{1}{T} \int_0^T |x(t)|^2 dt = \sum_{n=-\infty}^{\infty} |a_n|^2}$$

Parserval's Relation

Time Scaling

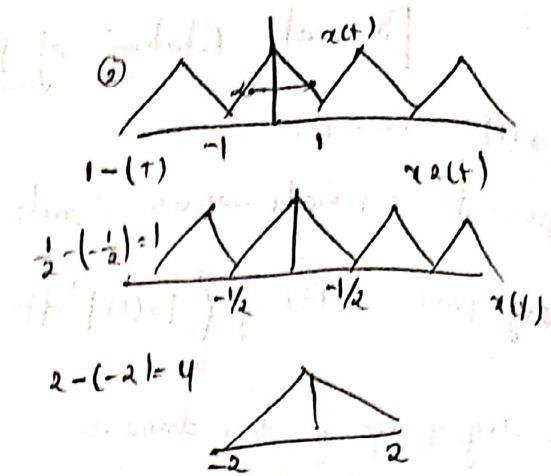
No change in

But T will be changing

$$T = \frac{2\pi}{\omega_0}$$

Period - $\frac{T}{\alpha}$

$$\frac{\omega_0}{T} \rightarrow \frac{\omega_0}{\frac{T}{\alpha}} = \alpha \omega_0$$



Parseval's Relation :- Explain the relation in power signal in domain & freq Domain (F.S co-efficients)

$x(t) \rightarrow$ periodic \rightarrow power signals \rightarrow power relation

$$x(t) \rightarrow a_n \rightarrow \frac{1}{T} \int_T |x(t)|^2 dt$$

$$= \left(\sum_{n=-\infty}^{\infty} |a_n|^2 \right) \text{avg power}$$

$$= (a_0)^2 + (a_1)^2 + \dots + |a_n|^2$$

F 1st harmonic n-th freq.

avg power = sum of the powers of all freq components

Properties of Fourier Series

$$x(t) + y(t) \rightarrow x(t) + y(t) \xrightarrow{T'} \sum_{n=-\infty}^{\infty} [a_n + b_n]$$

① Linearity property $\rightarrow Ax(t) + By(t) \rightarrow Aa_n + Bb_n$

② Time Reversal $\rightarrow x(t) \rightarrow a_n \Rightarrow a(-t) \rightarrow a_{-n}$

③ Time shifting $\rightarrow x(t) \rightarrow a_n \Rightarrow x(t-t_0) \rightarrow e^{-j\omega_0 t_0} a_n$

④ Time Scaling $\rightarrow x(t) \rightarrow a_n \Rightarrow x(\alpha t)$

$$3) \text{ Conjugation} \rightarrow x(t) \rightarrow a_n \Rightarrow x^*(t) \rightarrow a_{-n}$$

If $x(t)$ real $\Rightarrow x(t) = x^*(t) \Rightarrow a_n = a_{-n}^* \rightarrow a_n = a_n^*$

$x(t) = \text{real} + \text{even} \Rightarrow x(t) = x(-t) \Rightarrow a_n = a_{-n} = a_n^*$

Gibb's phenomenon

$f(t)/x(t) \rightarrow f(t) = \sum_{n=-\infty}^{\infty} a_n e^{jnw_0 t}$

periodic \rightarrow approximated by Sinusoids / complex Exponential

If the no. of terms (functions) in the approximation increases
MSE \downarrow

The occurrence of overshoot / Ripple at the point of discontinuity in the approximation of f-s it is called as Gibbs phenomenon

(Concept of Negative frequency)

$$f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t}$$

$$= \underbrace{F_0}_{DC} + F_1 e^{j\omega_0 t} + F_2 e^{j2\omega_0 t} + \dots + F_n e^{jn\omega_0 t} + F_{-1} e^{-j\omega_0 t} + F_{-2} e^{-j2\omega_0 t} + \dots + F_{-n} e^{-jn\omega_0 t}$$

F_1 = F.S coefficient of 1st freq component

$$F_1 = " " " " " 0 = j\omega_0$$

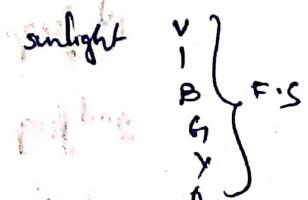
$$f_1 = \omega_0$$

$$F_1 = \omega_0$$

F_2 = F.S coefficient of 2nd freq component

$$F_2 = " " " " "$$

F_n/F_1 = F.S coefficient of nth component



$$F_1 = a+jb$$

$$F_1 = a-jb$$

$$|F_1| = |F_1|$$

phasors are diff (opposite) in direction but magnitudes are same

freq = no. of cycles/sec \rightarrow ② real & imaginary part

\rightarrow Negative freq = 2 but phase is opposite

$$e^{j\omega_0 t} \& e^{-j\omega_0 t}$$

positive freq \rightarrow construct real signal

In Hz Analysis \rightarrow +ve & -ve

mathematical model freq components

$$(e^{j\omega_0 t} + e^{-j\omega_0 t}) = 2 \cos \omega_0 t$$

obtaining complex fourier spectrum

Spectrum \rightarrow set of frequencies

Fourier \rightarrow F.S / F.T

Complex \rightarrow the F.S co-efficients of any periodic signal is complex

Freq present 0, ω_0 , $-\omega_0$, $2\omega_0$, $-2\omega_0$ \dots $n\omega_0$, $-n\omega_0$

$f(t)$:

$$\omega_0 = \frac{2\pi}{T}$$

$$f(t) = F_0 + F_1 e^{j\omega_0 t} + F_2 e^{j2\omega_0 t} + \dots + F_n e^{jn\omega_0 t} + F_1 e^{-j\omega_0 t} + F_2 e^{-j2\omega_0 t} + \dots + F_n e^{-jn\omega_0 t}$$

dc component freq = 0

1st freq.

$$\text{freq} = \omega_0, \text{ or } -\omega_0 \rightarrow F_1$$

$$\omega_0 = 0 = F_0$$

$$\omega = \omega_0 = F_1, F_{-1}$$

2nd freq

$$\text{freq} = 2\omega_0, -2\omega_0 \rightarrow F_2$$

$$2\omega_0 = F_2, F_{-2}$$

3rd freq

$$\text{freq} = 3\omega_0, -3\omega_0 \rightarrow F_3$$

$$F_3 = F_{-3}$$

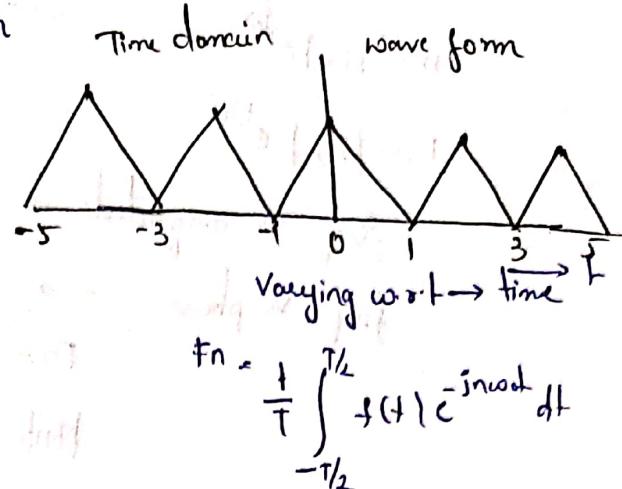
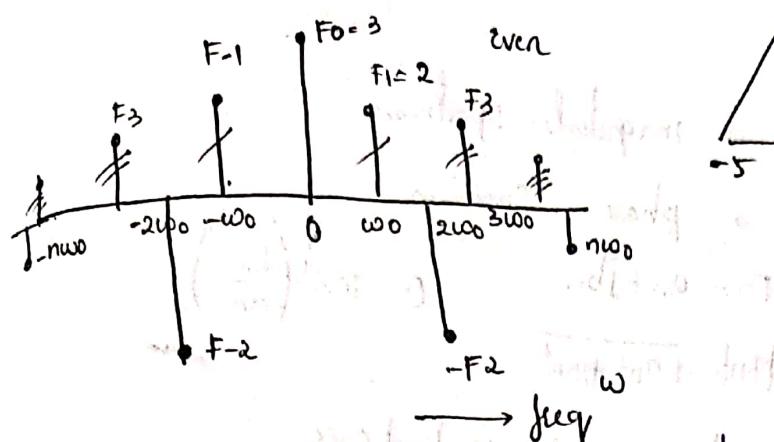
n th freq

$$= n\omega_0, -n\omega_0 \rightarrow F_n$$

$$F_n = F_{-n}$$

$f(t) \rightarrow$ periodic \rightarrow time domain \rightarrow frequency domain

Freq Domain representation



magnitude Versus frequency components

$$F_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jnwot} dt$$

freq vs magnitude \rightarrow spectrum

magnitude spectrum $|F_n| = |F_{-n}| = \text{same}$

complex fourier spectrum $\sum f(t) e^{-jnwot} dt$

$$n = 0, \pm 1, \pm 2, \pm 3, \dots, nw0 \checkmark$$

Only discrete freq $w_0, 2w_0, 3w_0, \dots$ discrete spectrum

$$f_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jnwot} dt$$

line spectrum
magnitude spectrum

$$F_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jnwot} dt$$

$$|F_n| = |F_{-n}| = \text{same}$$

$$F_n = |F_n| e^{-j\theta}$$

$$F_{-n} = |F_n| e^{+j\theta}$$

$F_n \rightarrow$ complex quantity

$$a_0 = 1, a_1 = \left(1 + \frac{1}{2}j\right) = a + jb$$

$$a_1 \left(1 + \frac{1}{2}j\right)$$

$$a_2 = \frac{\sqrt{2}}{4} (1-j)$$

$$a_3 = \frac{\sqrt{2}}{4} (1+j)$$

Complex nature of $F_n \rightarrow$ two Spectrums

$$F_n = |F_n| e^{j\theta}$$

$$F_n = |F_n| e^{j\theta}$$

freq vs magnitude \rightarrow magnitude spectrum

freq vs phase \rightarrow phase spectrum

$$F_n = a_n + j b_n$$

$$\theta = \tan^{-1} \left(\frac{b_n}{a_n} \right)$$

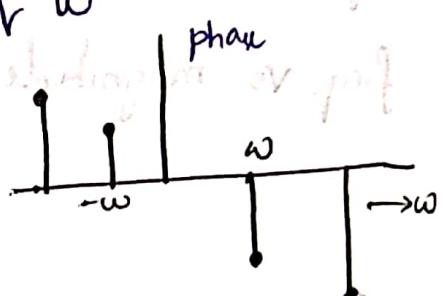
$$|F_n| = \sqrt{a_n^2 + b_n^2}$$

magnitude spectrum \rightarrow symmetric about vertical axis

even function of freq ω

phase spectrum \rightarrow anti-symmetric

odd function of freq ω .



Summary Fourier Series

$f(t) \rightarrow$ periodic \rightarrow all sum of complex exponentials / sinusoidal signal
 ei ωt complete \downarrow
 sin n ωt & cos n ωt .
 Complete set of orthogonal functions

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos n\omega t + b_n \sin n\omega t) \rightarrow \text{TFS}$$

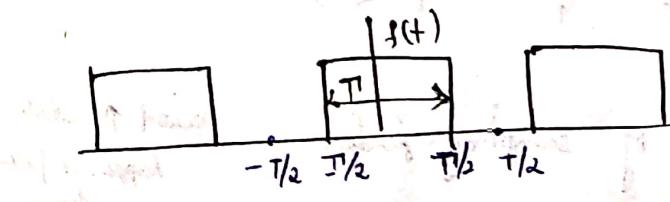
$$a_0 = \frac{1}{T} \int_{-T/2}^{T/2} f(t) dt$$

$$a_n = \frac{2}{T} \int_{-T/2}^{T/2} f(t) \cos n\omega t dt$$

$$b_n = \frac{2}{T} \int_{-T/2}^{T/2} f(t) \sin n\omega t dt$$

$$f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega t} \quad \left. \begin{array}{l} F_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jn\omega t} dt \end{array} \right\} \text{EFS}$$

Fourier Transform Representation



$f(t)$ periodic function T basic function / rectangular pulse

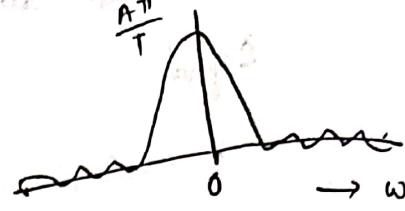
$$\text{EFS} = f(t) = \sum_{n=1}^{\infty} F_n e^{jn\omega t}$$

$$F_n = \frac{AT}{T} \operatorname{sinc}\left(\frac{n\omega_0 T}{2}\right)$$

$$f(t) = \sum_{n=-\infty}^{\infty} \frac{AT}{T} \operatorname{sa}\left(\frac{n\omega_0 T}{2}\right) e^{jn\omega_0 t}$$

$$\rightarrow \operatorname{sa}(x) = \frac{\sin x}{x}$$

$$\operatorname{sa}(0) = 1$$



$$\frac{\sin(n\omega_0 T/2)}{n\omega_0 T/2} = \operatorname{sa}\left(\frac{n\omega_0 T}{2}\right)$$

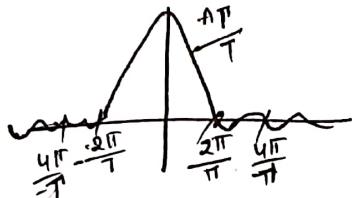
$$\frac{\sin n\omega_0 T/2}{n\omega_0 T/2} = 0$$

$$\Rightarrow \frac{\sin n\omega_0 T}{2} = 0$$

$$= \sin \frac{n\omega_0 T}{2} = \sin(n\pi)$$

$$\frac{n\omega_0 T}{2} = \pm n\pi$$

$$n\omega_0 T = \pm \frac{2n\pi}{T}$$



$$n\omega_0 = \pm \frac{2n\pi}{T}$$

$$= \frac{2\pi}{T}, \frac{4\pi}{T}, \frac{6\pi}{T}, \dots, \frac{2n\pi}{T}$$

T-width
T-period

$$f(t) = \sum \frac{AT}{T} \operatorname{sa}\left(\frac{n\omega_0 T}{2}\right) e^{jn\omega_0 t}$$

$$n=0 \quad \operatorname{sa}(0)=1 \quad \frac{AT}{T}, \frac{AT}{T}, \operatorname{sa}\left(\frac{\omega_0 T}{2}\right), \frac{\pi T}{T} \operatorname{sa}\left(\frac{2\omega_0 T}{2}\right)$$

$$\frac{AT}{T}$$

→ magnitude of sampling function period T what happen in frequency domain

$$T = \frac{1}{20} \text{ width/k}$$

$$T = \frac{1}{4} \text{ period}$$

$$\frac{T}{T} = \frac{\frac{1}{20}}{\frac{1}{4}} = \frac{1}{5}$$

$$3\text{ zeros at } \pm \frac{2n\pi}{T} = \frac{2n\pi}{20} \pm 40n\pi$$

$$T \rightarrow \text{period of pulse}$$

$$T \rightarrow \text{width of pulse}$$

$$\frac{n\omega_0 T}{2} = \pm n\pi$$

$$n\omega_0 T = \pm 2n\pi$$

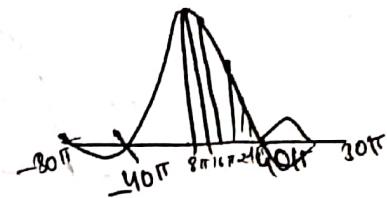
$$n\omega_0 = \pm 40n\pi$$

$$- \pm 40 \text{ rad}, \pm 80\pi \text{ rad}$$

$$T = \frac{1}{20} \text{ width}$$

$$T = \frac{1}{4}$$

$$T = \frac{1}{2}, 3\pi, \pm 40\pi, \pm 80\pi, \pm 120\pi$$



$$\omega_0 = \frac{2\pi}{T} = \frac{2\pi}{1/20} = 40\pi$$

$$4\pi, 8\pi, 12\pi, 16\pi, 20\pi, 24\pi, 28\pi, 32\pi, 36\pi, 40\pi$$

$$T=1 \Rightarrow T=\frac{1}{20}$$

10 freq components

$$\omega_0 = \frac{2\pi}{T} = 2\pi$$

$$T=50 \Rightarrow \frac{0\pi}{50} = 0$$

$$0.02 \times 2\pi$$

$$T = d \\ f = \frac{1}{T} = \frac{1}{d} = 0$$

no spacing b/w freq components

$T \uparrow \Rightarrow$ spectrum

AS T increases frequency components \uparrow frequency Spacing decreases

$T = \text{large} \uparrow$ close to Unity

$$\omega = \frac{2\pi}{T} = 0$$

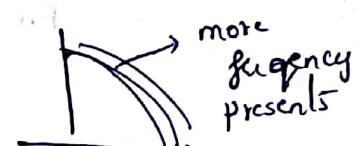
there no spacing between successive freq components

freq = continuous

$$\frac{1}{4} = 5$$

$$\frac{1}{2} = 10$$

$$T = \frac{1}{f}$$

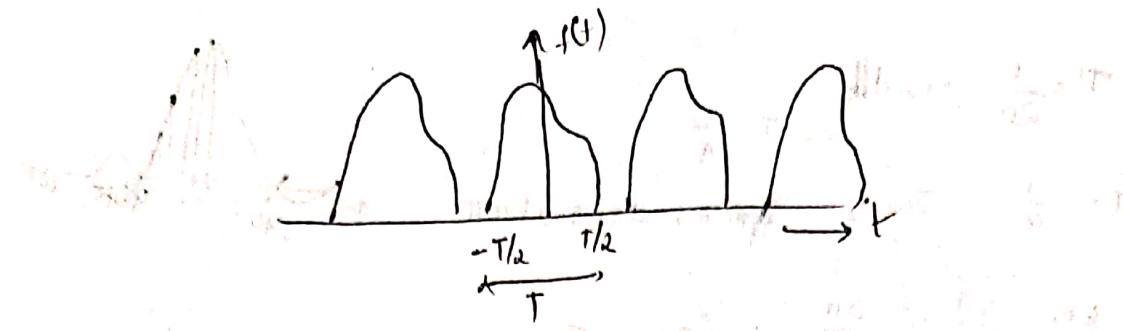


$$F.S \Rightarrow f(t) = \sum_{n=1}^{\infty} f_n e^{j\omega_n t} \xrightarrow{\text{discrete sum of complex}} \text{Exponential}$$

freq = finite
Spectrum \rightarrow discrete in nature

$F.T \Rightarrow$ freq = infinite
Spectrum \rightarrow continuous sum of exponentials

Repeating f_n over the entire interval $(-\infty, \infty)$
 ~~\sum~~ replaced with \int



at T is very large

$f_r(t)$ is periodic $\Rightarrow F_s$

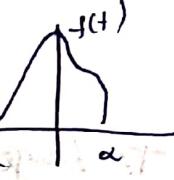
$$f_r(t) = \sum_{n=-\infty}^{\infty} f_n e^{jn\omega_0 t}$$

$T \uparrow \rightarrow$ periodic becomes non-per

$$f_n = \frac{1}{T} \int_{-T/2}^{T/2} f_r(t) e^{-jn\omega_0 t} dt$$

as periodic $\Rightarrow T=d$ means

$$\frac{LT}{T-d} f_r(t) = f(t)$$



$$f_r(t) = \sum_n f_n e^{jn\omega_0 t}$$

$$f_n = \frac{1}{T} \int_{-T/2}^{T/2} f_r(t) e^{-jn\omega_0 t} dt$$

$$f_n = \int_{-T/2}^{T/2} f_r(t) e^{-jn\omega_0 t} dt$$

$$\boxed{\text{det } F(\omega) = \int_{-d}^d f(t) e^{-j\omega t} dt}$$

$$\text{if } f_n = F(n\omega_0) \Rightarrow$$

$$f_n = \frac{1}{T} F(n\omega_0)$$

$$f_r(t) = \sum_{n=-\infty}^{\infty} \frac{1}{T} F(n\omega_0) e^{jn\omega_0 t}$$

$$\boxed{T = \frac{2\pi}{\omega_0} \rightarrow T=d}$$

$$\text{det } f_r(t) = \frac{1}{2\pi} \sum_{n=-\infty}^{\infty} F(n\omega_0) e^{jn\omega_0 t \omega_0}$$

$$T \rightarrow d$$

as $T \rightarrow \infty \neq \omega = 0$ small $= \Delta\omega$

freq spacing = 0 close to each other in the range \Rightarrow more freq

$$f(t) = \frac{1}{2\pi}$$

$$\text{let } f_1(t) = \frac{1}{2\pi} \sum_{n=-\infty}^{\infty} F(n\Delta\omega_0) e^{jn(\pi\Delta\omega_0)t}. \Delta\omega_0$$

non periodic $f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt \rightarrow FT[f(t)] = F(\omega)$$

FS \rightarrow periodic

FS \rightarrow non periodic

the $FT[f(t)]$ denoted $F(\omega)$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt \rightarrow 0$$

$$IFT[F(\omega)] = f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega \rightarrow ②$$

$$f(t) \xrightarrow{FT} F(\omega)$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$TD \quad FD$$
$$f(t) \longleftrightarrow F(\omega)$$

$$IFT[f(t)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$$