

# 034IN - FONDAMENTI DI AUTOMATICA FUNDAMENTALS OF AUTOMATIC CONTROL A.Y. 2023-2024

Part VII: Step-Response Analysis

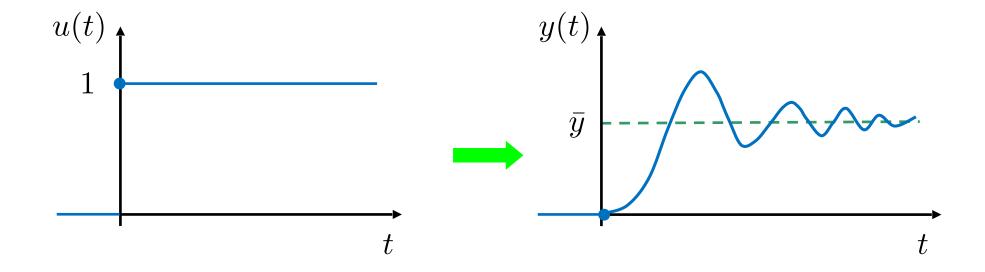
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#### **Step Response**



$$x(0) = 0;$$
  $u(t) = 1(t)$ 



- ◆ For asymptotically stable systems, the step response describes the way the systems "moves" from an equilibrium to another
- ★ The characteristics of the step response are a key element in the requirements for a control systems

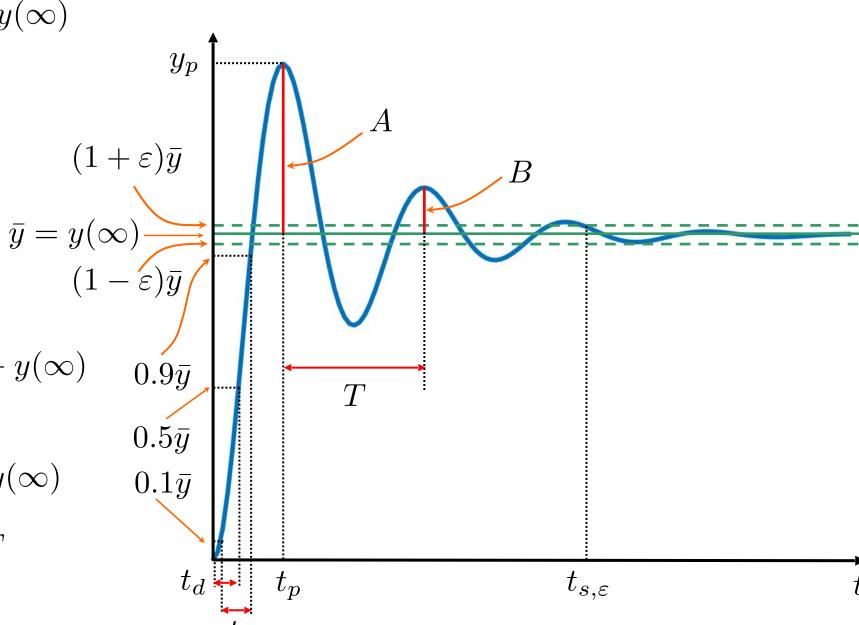
# **Characteristic Parameters of the Step Response**



- Steady-state value:  $\bar{y} = y(\infty)$
- Settling time:  $t_{s,\varepsilon}$
- Rise time:  $t_r$
- Delay time:  $t_d$
- Peak time:  $t_p$
- Peak value:  $y_p$
- Max. overshoot:  $A = y_p y(\infty)$
- Max. % overshoot:

$$\Delta\% = 100 \cdot A/y(\infty)$$

- "Period" of oscillations: T
- Damping factor: B/A



#### **Step Response: First Order Systems**



• Case A)

$$G(s) = \frac{\mu}{1+s\tau} \,; \quad \mu > 0; \, \tau > 0 \qquad \text{strictly proper first-order system}$$
 asymptotic stability

Case B)

$$G(s) = \frac{\mu(1+sT)}{1+s\tau}\,; \quad \mu > 0; \, \tau > 0 \quad \text{non strictly proper first-order system}$$
 asymptotic stability

#### **Step Response: First Order Systems** (contd.)

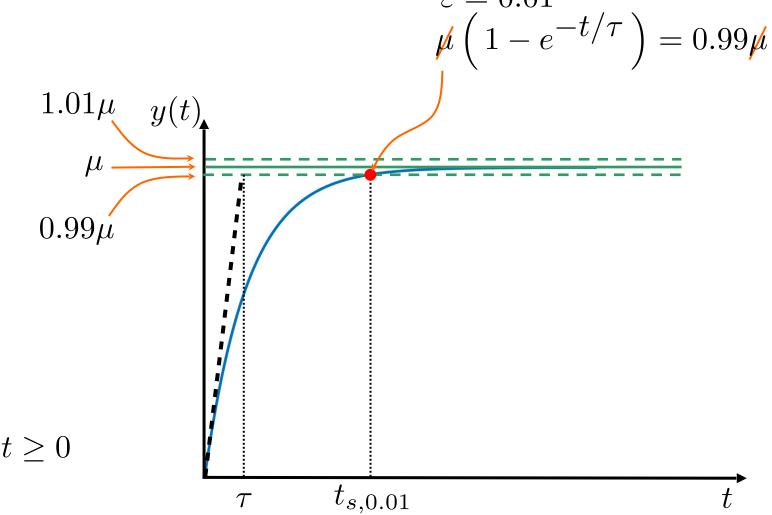


Case A)

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$
$$= \mathcal{L}^{-1} \left[ \frac{\mu}{s(1+s\tau)} \right]$$

$$= \mathcal{L}^{-1} \left[ \frac{\mu}{s} - \frac{\mu \tau}{1 + s\tau} \right]$$

$$= \mu \left( 1 - e^{-t/\tau} \right), \quad t \ge 0$$



#### **Settling-Time Calculation**



For example, the settling time for  $\varepsilon = 0.01$  can be characterised as follows:

$$1 - e^{-t/\tau} = 0.99 \implies e^{-t/\tau} = 0.01 \implies e^{t/\tau} = 100$$

$$t_{s,0.01} = \tau \ln 100 \simeq 4.6\tau$$

The calculation of the rising time  $t_r$  and the delay time  $t_d$  follows similar lines.

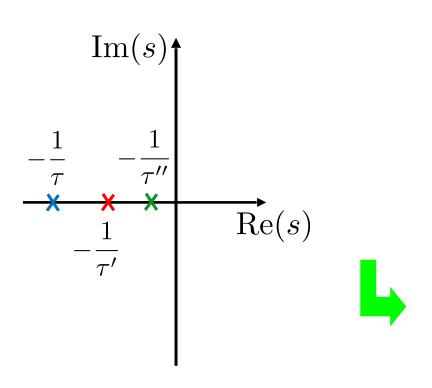
The following approximations are useful:

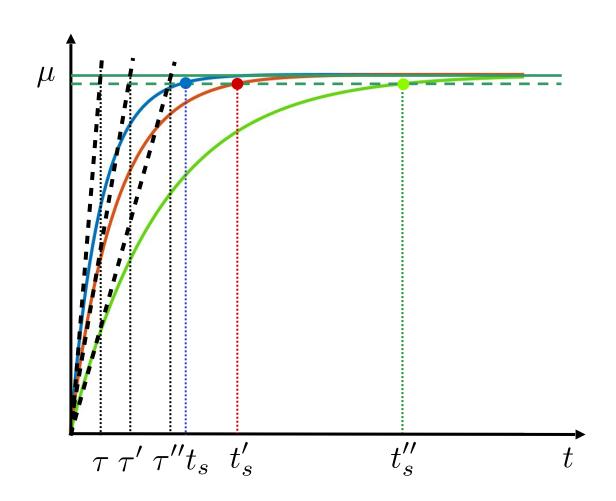
$$t_r \simeq 2.2\tau$$
  $t_d \simeq 0.7\tau$   $t_{s,0.05} \simeq 3\tau$   $t_{s,0.01} \simeq 4.6\tau$ 

**Remark**: without loss of generality, from now on we shall use  $t_s$  as a shorthand for  $t_{s,0.01}$ 

#### **Qualitative Analysis of the Step Response**







# **Step Response: First Order Systems** (contd.)



#### Case B)

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

$$= \mathcal{L}^{-1} \left[ \frac{\mu(1+sT)}{s(1+s\tau)} \right]$$

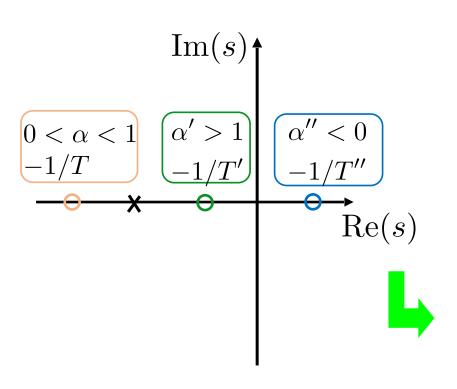
$$= \mathcal{L}^{-1} \left[ \frac{\mu}{s} + \frac{\mu(T-\tau)}{1+s\tau} \right]$$

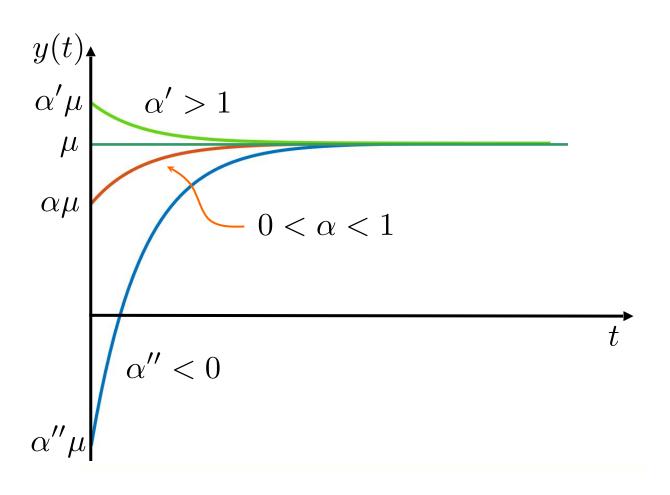
$$= \mu \left( 1 + (\alpha - 1)e^{-t/\tau} \right), \quad t \ge 0 \text{ with } T = \alpha \tau$$

Note that (the system is not strictly proper):  $\lim_{t\to 0^+}y(t)=\mu\frac{T}{\tau}\neq 0$ 

#### **Qualitative Analysis of the Step Response**







#### **Step Response: Second Order Systems**



Case A)

$$G(s) = \frac{\mu}{(1+s\tau_1)(1+s\tau_2)}$$
 real poles, no zeros

Case B)

$$G(s) = \frac{\mu(1+sT)}{(1+s\tau_1)(1+s\tau_2)}$$
 real poles, one zero

• Case C)

$$G(s) = \frac{\varrho}{(s + \sigma + i\omega)(s + \sigma - i\omega)}$$
 complex poles, no zeros

Case D)

$$G(s) = rac{arrho(1+sT)}{(s+\sigma+j\omega)(s+\sigma-j\omega)}$$
 complex poles, one zero



Case A)

$$G(s) = \frac{\mu}{(1+s\tau_1)(1+s\tau_2)}; \quad \mu > 0; \quad \tau_1 \neq \tau_2$$

$$\left. egin{array}{l} au_1 > 0 \\ au_2 > 0 \end{array} 
ight\} \, ext{asymptotic stability}$$

Without loss of generality, assume  $\tau_1 > \tau_2$ 



$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[ \frac{\mu}{s(1+s\tau_1)(1+s\tau_2)} \right]$$
$$= \mathcal{L}^{-1} \left[ \frac{A}{s} + \frac{B}{1+s\tau_1} + \frac{C}{1+s\tau_2} \right]$$

where

$$A = \frac{\mu}{(1+s\tau_1)(1+s\tau_2)}\Big|_{s=0} = \mu$$

$$B = \frac{\mu}{s(1+s\tau_2)}\Big|_{s=-1/\tau_1} = \frac{\mu}{-\frac{1}{\tau_1}(1-\frac{\tau_2}{\tau_1})} = \frac{\mu\tau_1^2}{\tau_2-\tau_1}$$

$$C = \frac{\mu}{s(1+s\tau_1)}\Big|_{s=-1/\tau_2} = \frac{\mu}{-\frac{1}{\tau_2}(1-\frac{\tau_1}{\tau_2})} = \frac{\mu\tau_2^2}{\tau_1-\tau_2}$$



Hence:

$$y(t) = \mathcal{L}^{-1} \left[ \frac{\mu}{s} + \frac{\frac{\mu \tau_1^2}{\tau_2 - \tau_1}}{1 + s\tau_1} + \frac{\frac{\mu \tau_2^2}{\tau_1 - \tau_2}}{1 + s\tau_2} \right]$$

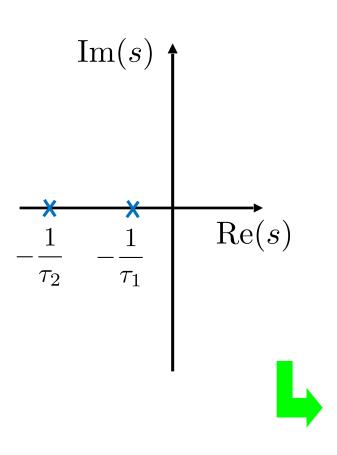
$$= \mu \left( 1 - \frac{\tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \ge 0$$

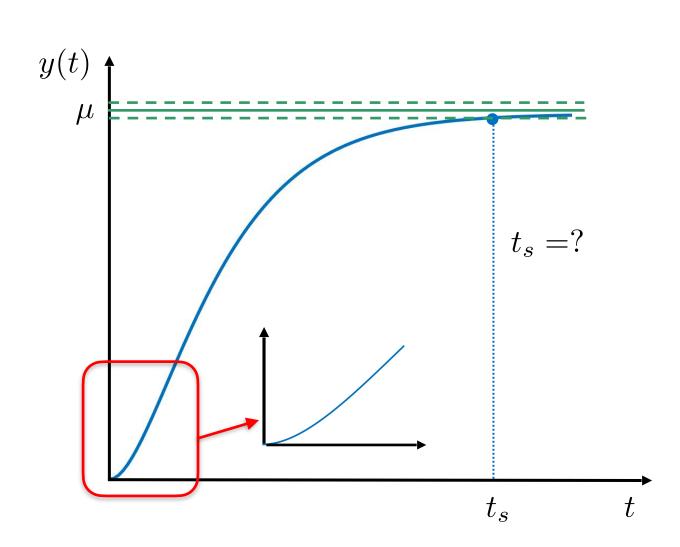
#### Characteristics:

- $y(\infty) = \mu > 0$
- y(0) = 0
- $\dot{y}(0) = 0$   $\ddot{y}(0) = \frac{\mu}{\tau_1 \tau_2} > 0$

#### **Qualitative Analysis of the Step Response**







# **Approximate Calculation of the Settling Time**



If  $\tau_1 \gg \tau_2$ :

$$y(t) = \mu \left( 1 - \frac{\tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \ge 0$$

$$\simeq \mu \left( 1 - e^{-t/\tau_1} \right), \quad t \ge 0$$

$$t_s \simeq 4.6\tau_1$$

In general, in the absence of zeros, the most influential poles on the qualitative behaviour of the step response are the ones closer to the imaginary axis.

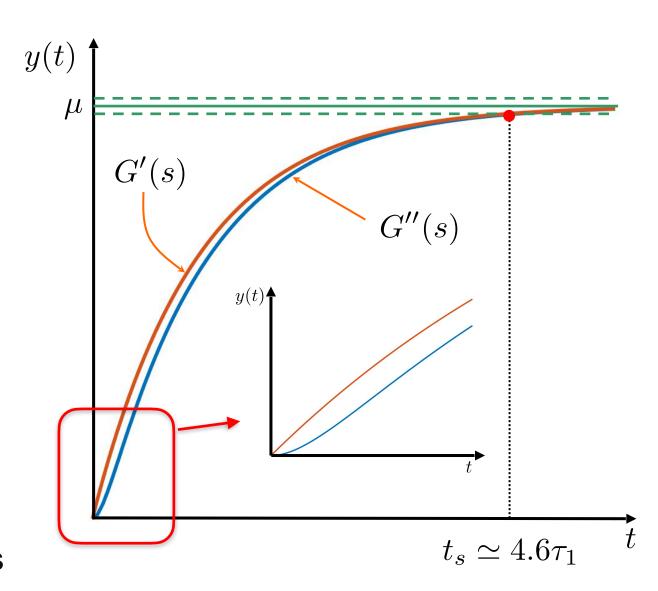
# Qualitative Analysis: Comparison Between First and Second Order DITRIESTE



$$G'(s) = \frac{\mu}{1 + s\tau_1}$$

$$G''(s) = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2)}; \, \tau_1 \gg \tau_2$$

- The main difference lies in the initial transient behaviour
- For a given settling time, the stepresponse in the second-order case without zeros has a "slower" dynamics





Case B)

$$G(s) = \frac{\mu(1+sT)}{(1+s\tau_1)(1+s\tau_2)}; \quad \mu > 0; \quad \tau_1 \neq \tau_2$$

$$\left. \begin{array}{c} \tau_1 > 0 \\ \tau_2 > 0 \end{array} \right\} \qquad \qquad \text{asymptotic stability}$$

Without loss of generality, assume  $\tau_1 > \tau_2$ 



$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[ \frac{\mu}{s(1+s\tau_1)(1+s\tau_2)} \right]$$
$$= \mathcal{L}^{-1} \left[ \frac{A}{s} + \frac{B}{1+s\tau_1} + \frac{C}{1+s\tau_2} \right]$$

where

$$A = \frac{\mu(1+sT)}{(1+s\tau_1)(1+s\tau_2)} \Big|_{s=0} = \mu$$

$$B = \frac{\mu(1+sT)}{s(1+s\tau_2)} \Big|_{s=-1/\tau_1} = \frac{\mu(1-T/\tau_1)}{-\frac{1}{\tau_1}(1-\frac{\tau_2}{\tau_1})} = \frac{\mu\tau_1(\tau_1-T)}{\tau_2-\tau_1}$$

$$C = \frac{\mu(1+sT)}{s(1+s\tau_1)} \Big|_{s=-1/\tau_2} = \frac{\mu(1-T/\tau_2)}{-\frac{1}{\tau_2}(1-\frac{\tau_1}{\tau_2})} = \frac{\mu\tau_2(\tau_2-T)}{\tau_1-\tau_2}$$



Hence:

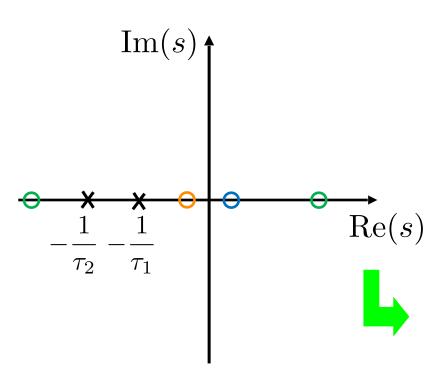
$$y(t) = \mu \left( 1 - \frac{\tau_1 - T}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2 - T}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \ge 0$$

#### **Characteristics:**

• 
$$y(\infty) = \mu > 0$$
  
•  $y(0) = 0$   
•  $\dot{y}(0) = \frac{\mu T}{\tau_1 \tau_2} \begin{cases} > 0, & \text{if } T > 0 \\ < 0, & \text{if } T < 0 \end{cases}$ 

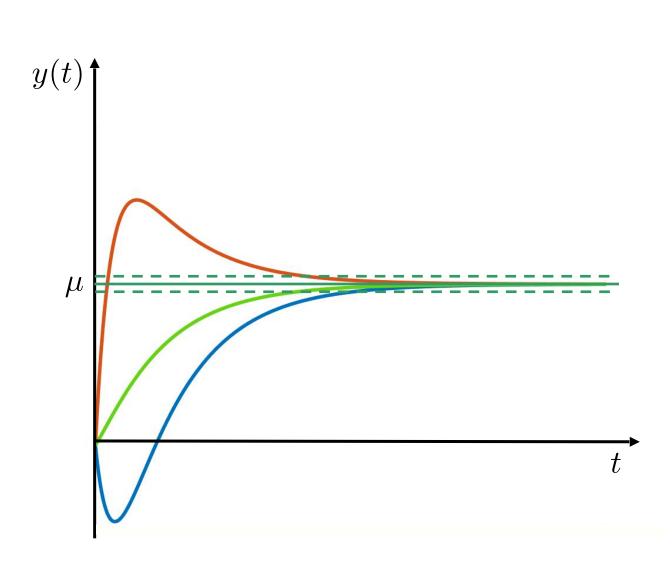
#### **Qualitative Analysis of the Step Response**







- overshoot
- undershoot





#### Case C)

$$G(s) = \frac{\varrho}{(s + \sigma + j\omega)(s + \sigma - j\omega)}$$

$$\mu = G(0) = \frac{\varrho}{\sigma^2 + \omega^2}$$

poles: 
$$-\sigma \pm j\omega$$

$$\sigma > 0$$
 asymptotic stability

$$\omega > 0$$

$$\varrho > 0$$



$$Y(s) = \frac{G(s)}{s} = \frac{A}{s} + \frac{Bs + C}{s^2 + 2\sigma s + \sigma^2 + \omega^2}$$



$$As^2 + 2A\sigma s + A\sigma^2 + A\omega^2 + Bs^2 + Cs = \varrho$$



$$\begin{cases} A + B = 0 \\ 2A\sigma + C = 0 \\ A(\sigma^2 + \omega^2) = \varrho \end{cases}$$



$$\begin{cases} A + B = 0 \\ 2A\sigma + C = 0 \\ A(\sigma^2 + \omega^2) = \varrho \end{cases} \longrightarrow \begin{cases} A = \frac{\varrho}{\sigma^2 + \omega^2} = \mu \\ B = -\mu \\ C = -2\sigma\mu \end{cases}$$

$$Y(s) = \mu \left[ \frac{1}{s} - \frac{s + 2\sigma}{s^2 + 2\sigma s + \sigma^2 + \omega^2} \right] = \mu \left[ \frac{1}{s} - \frac{s + \sigma + \sigma}{(s + \sigma)^2 + \omega^2} \right]$$

$$= \mu \left[ \frac{1}{s} - \frac{s+\sigma}{(s+\sigma)^2 + \omega^2} - \frac{\sigma}{\omega} \frac{\omega}{(s+\sigma)^2 + \omega^2} \right]$$



Hence: 
$$y(t) = \mu \left[ 1 - e^{-\sigma t} \cos(\omega t) - \frac{\sigma}{\omega} e^{-\sigma t} \sin(\omega t) \right], \quad t \ge 0$$

$$= \mu \left[ 1 - e^{-\sigma t} \left( \cos(\omega t) + \frac{\sigma}{\omega} \sin(\omega t) \right) \right], \quad t \ge 0$$

$$= \mu \left[ 1 - \frac{\sqrt{\sigma^2 + \omega^2}}{\omega} e^{-\sigma t} \sin(\omega t + \varphi) \right], \quad t \ge 0$$

damped oscillations

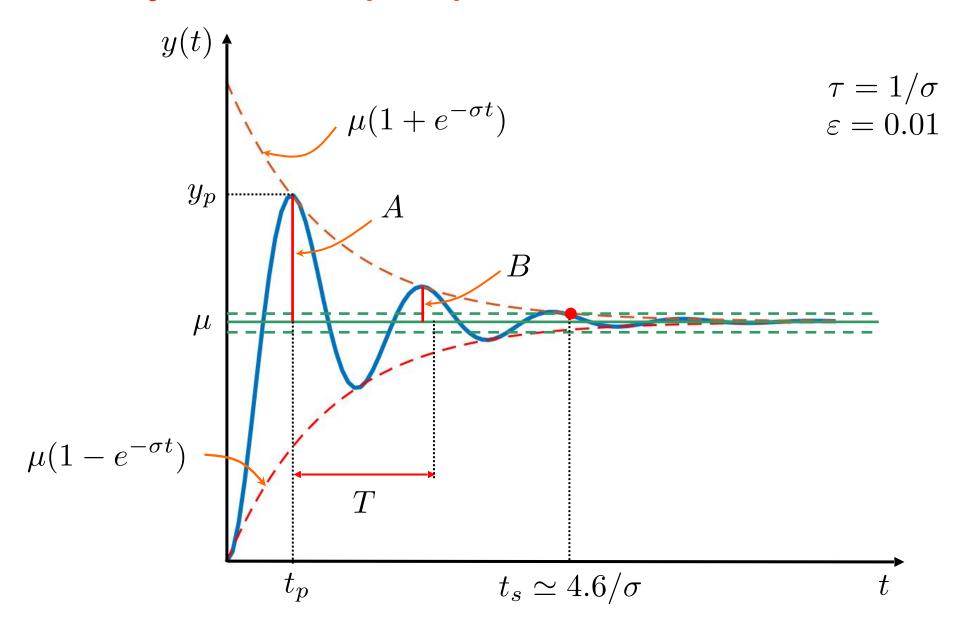
where 
$$\varphi = \arccos\left(\frac{\sigma}{\sqrt{\sigma^2 + \omega^2}}\right)$$

Characteristics: 
$$\bullet$$
  $y(\infty) = \mu > 0$ 

- y(0) = 0
- $\bullet \quad \dot{y}(0) = 0$
- $\bullet \quad \ddot{y}(0) = \varrho > 0$

#### **Qualitative Analysis of the Step Response**

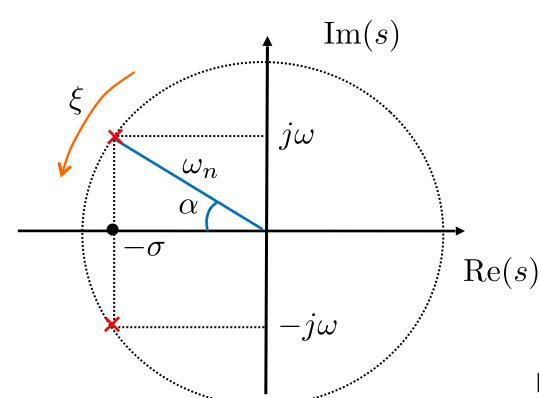




#### **Characteristic Parameters of the Step Response**



#### Recall from Part 4:



$$G(s) = \frac{\varrho}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

$$\omega_n^2 = \sigma^2 + \omega^2$$

$$\omega_n \xi = \sigma$$

$$\omega_n \sqrt{1 - \xi^2} = \omega$$

#### **Parameters**:

 $\omega_n$  natural angular frequency:

 $\xi = \cos(\alpha)$  damping ratio

#### **Characteristic Parameters of the Step Response** (contd.)



and:

$$G(s) = \frac{\varrho}{(s+\sigma+j\omega)(s+\sigma-j\omega)} = \frac{\varrho}{(s+\sigma)^2 + \omega^2}$$
$$= \frac{\varrho}{s^2 + 2\sigma s + \sigma^2 + \omega^2} = \frac{\varrho}{s^2 + 2\xi\omega_n s + \omega_n^2}$$
$$2\xi\omega_n \qquad \omega_n^2$$

$$G(s) = \frac{\varrho/\omega_n^2}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2} = \frac{\mu}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

where: 
$$\mu:=rac{arrho}{\omega_n^2}$$

# Characteristic Parameters of the Step Response (contd.)



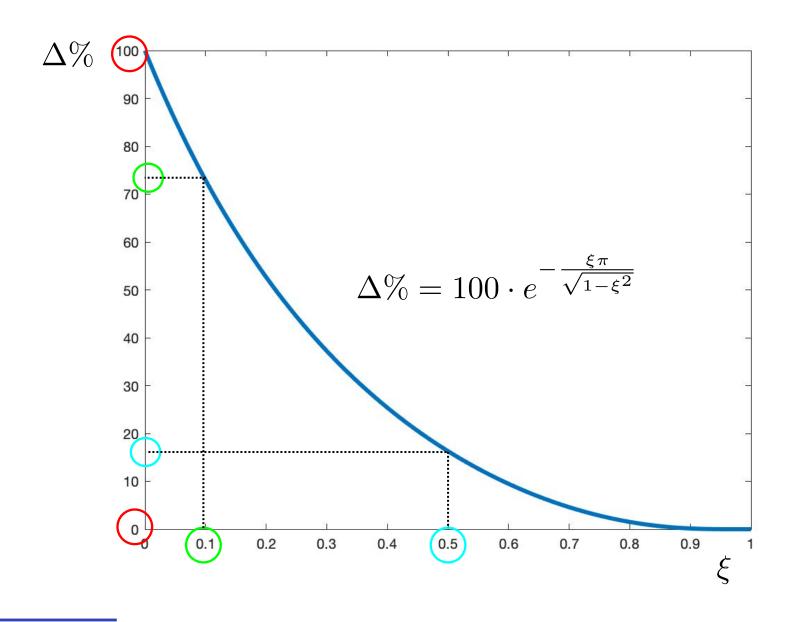
#### Hence:

- Settling time:  $t_s \simeq \frac{4.6}{\sigma} = \frac{4.6}{\xi \omega_n}$
- Peak time:  $t_p = \frac{\pi}{\omega} = \frac{\pi}{\omega_n \sqrt{1 \xi^2}}$
- Peak value:  $y_p = \mu \left[1 + e^{-\frac{\sigma\pi}{\omega}}\right] = \mu \left[1 + e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}}\right]$
- Maximum percentage overshoot:  $\Delta\%=100\cdot\frac{A}{n}=e^{-\sigma\pi/\omega}=100\cdot e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}}$
- "Period" of oscillations:  $T=\frac{2\pi}{\omega}=\frac{2\pi}{\omega_n\sqrt{1-\xi^2}}$  Damping factor:  $\frac{B}{A}=\cdots=\Delta^2=e^{-2\sigma\pi/\omega}=e^{-\frac{2\xi\pi}{\sqrt{1-\xi^2}}}$

only depend on  $\xi$ but **not** on  $\omega_n$ 

#### **Maximum Percentage Overshoot**





#### **Limit Cases**



• No damping:  $\xi = 0$ 

$$G(s) = \frac{\varrho}{s^2 + \omega_n^2}$$
 poles:  $\pm j\omega_n$ 



Undamped oscillations

• Full damping:  $\xi = 1$ 

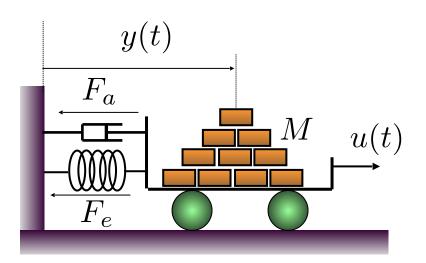
$$G(s) = \frac{\varrho}{(s + \omega_n)^2}$$
 poles:  $-\omega_n$ ;  $-\omega_n$ 



No oscillations at all

#### **Example 1**



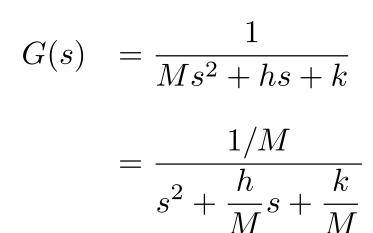


Hence:

$$\mu = G(0) = \frac{1}{k}$$

$$2\xi\omega_n = \frac{h}{M}$$

$$\omega_n^2 = \frac{k}{M}$$





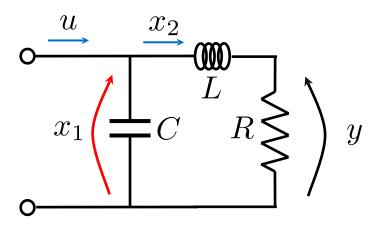
$$s^{2} + 2\xi\omega_{n}s + \omega_{n}^{2} = s^{2} + \frac{h}{M}s + \frac{k}{M}$$

$$\omega_n = \sqrt{\frac{k}{M}}$$

$$\xi = \frac{h}{2\sqrt{kM}}$$

#### Example 2





$$\begin{cases}
C\dot{x}_1 = u - x_2 \\
L\dot{x}_2 = x_1 - Rx_2 \\
y = Rx_2
\end{cases}$$

$$A = \begin{bmatrix} 0 & -1/C \\ 1/L & -R/L \end{bmatrix}$$

$$A = \begin{bmatrix} 0 & -1/C \\ 1/L & -R/L \end{bmatrix} \qquad B = \begin{bmatrix} 1/C \\ 0 \end{bmatrix} \qquad C = \begin{bmatrix} 0 & R \end{bmatrix}$$

$$G(s) = \begin{bmatrix} 0 & R \end{bmatrix} \begin{bmatrix} s & 1/C \\ -1/L & s+R/L \end{bmatrix}^{-1} \begin{bmatrix} 1/C \\ 0 \end{bmatrix} = \dots = \frac{R/(LC)}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

$$\omega_n = \frac{1}{\sqrt{LC}}; \quad \xi = \frac{R}{2} \sqrt{\frac{C}{L}}; \quad \mu = R$$



#### Case D)

$$G(s) = \frac{\mu(1+sT)}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}; \quad 0 < \xi < 1; \ \omega_n > 0; \ \mu > 0$$

#### Characteristics of the step response:

$$\bullet \quad y(\infty) = \mu > 0$$

• 
$$y(0) = 0$$

$$y(\infty) = \mu > 0$$

$$y(0) = 0$$

$$\dot{y}(0) = \mu T \omega_n^2$$

$$< 0, \quad \text{if } T < 0$$

# **Qualitative Analysis:**

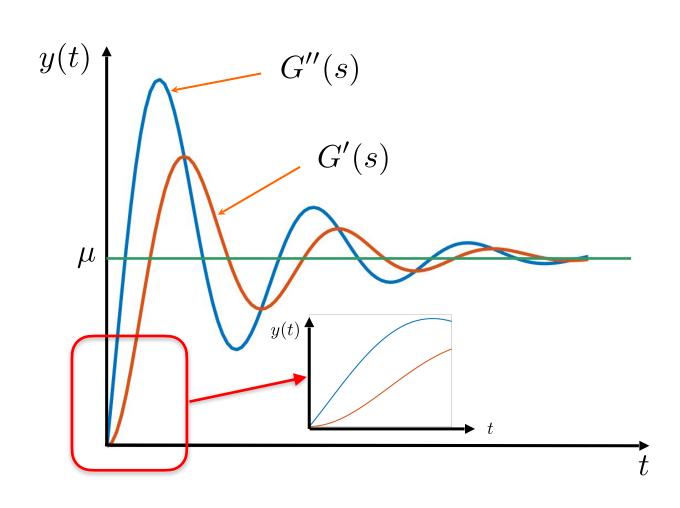


#### Comparison between Case C) (no zeros) and Case D) (one zero)

$$G'(s) = \frac{\mu}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

$$G''(s) = \frac{\mu(1+sT)}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

- Again, the main difference lies in the initial transient behaviour
- For a given settling time, the stepresponse in Case C) without zeros has a "slower" dynamics



# **Step Response for Systems of Order > 2**



For simplicity, consider the case of real poles only:

$$G(s) = \frac{\mu}{s^g} \frac{\prod_{i=1}^{n} (1 + sT_i)}{\prod_{i=1}^{n} (1 + s\tau_i)}$$

Recall (in the absence of common factors in G(s)):

Asymptotic Stability Re(poles) < 0  $g \le 0$ 

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

# **Step Response for Systems of Order > 2 (contd.)**



#### Initial Value Theorem

$$\lim_{t \to 0^+} y(t) = \lim_{s \to \infty} \int_{-s}^{s} \frac{1}{s} G(s) \begin{cases} = 0, & \text{if } m < n \\ \neq 0, & \text{if } m = n \end{cases}$$

#### Final Value Theorem

$$\lim_{t \to \infty} y(t) = \lim_{s \to 0} \int_{-s}^{s} \frac{1}{G(s)} \begin{cases} = \mu, & \text{if } g = 0 \\ = 0, & \text{if } g < 0 \end{cases}$$

#### **Dominant Poles Approximation**



Again, for simplicity, consider the case of real poles:

$$Y(s) = G(s)\frac{1}{s} = \frac{\alpha_0}{s} + \frac{\alpha_1}{1 + s\tau_1} + \dots + \frac{\alpha_n}{1 + s\tau_n}$$

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

$$= \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} + \dots + \frac{\alpha_n}{\tau_n} e^{-t/\tau_n}$$

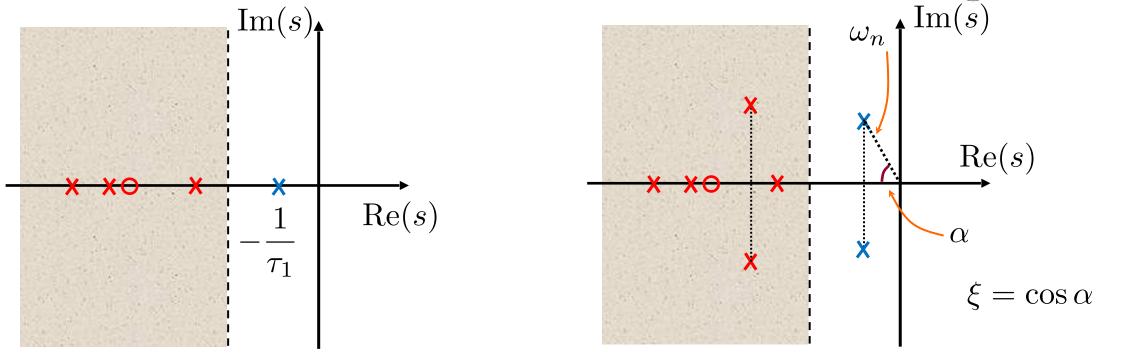
Assuming:  $\tau_1 > \tau_2 > \cdots > \tau_n$ 

$$y(t) = \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} + \dots + \frac{\alpha_n}{\tau_n} e^{-t/\tau_n}$$

$$\simeq \alpha_0 + rac{lpha_1}{ au_1} e^{-t/ au_1}$$
 dominant component, hence:  $t_s \simeq 4.6 au_1$ 

#### **Dominant Poles Approximation: Real Poles**





- When using the dominant poles approximation:
  - It is important to "preserve" the gain
  - Zeros located close to the imaginary axis have to be properly taken into account
- This approximation is useful in qualitative analysis and the for initial and rough controller's design steps

#### **Example**



$$G(s) = \frac{400(1+s)}{(1+0.2s)(1+0.1s)(s^2+2s+4)}$$

$$\psi_n = 2$$

$$\xi = 1/2$$

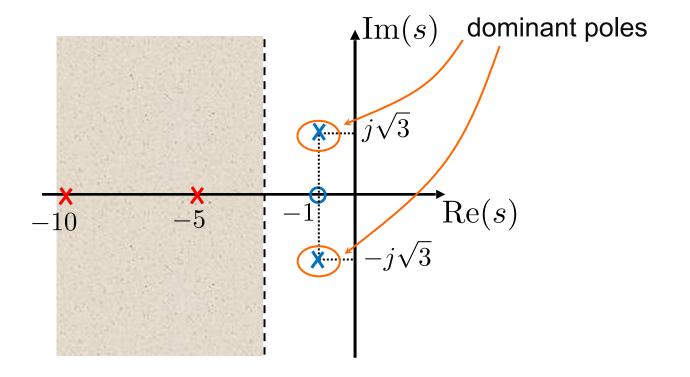
$$\mu = G(0) = 100$$

poles: 
$$-5$$

$$-10$$

$$-1 \pm j\sqrt{3}$$

zero: -1



#### Example (contd.)



