Circuits and Transforms

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Abstract—This manual provides a simple introduction to Transforms

1 Definitions

1. The unit step function is

$$u(t) = \begin{cases} 1 & t > 0 \\ \frac{1}{2} & t = 0 \\ 0 & t < 0 \end{cases}$$
 (1.1)

2. The Laplace transform of g(t) is defined as

$$G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt$$
 (1.2)

2 Laplace Transform

1. In the circuit, the switch S is connected to position P for a long time so that the charge on the capacitor becomes $q_1 \mu C$. Then S is switched to position Q. After a long time, the charge on the capacitor is $q_2 \mu C$.

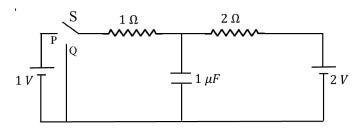


Fig. 2.1

2. Find q_1 .

Solution: The equivalent circuit at steady-state when the switch is at P is shown alongside.

Assuming the circuit to be grounded at G and the relative potential at point X to be V, we use KCL at X and get

$$\frac{V-1}{1} + \frac{V-2}{2} = 0 \tag{2.1}$$

$$\implies V = \frac{4}{3} V \tag{2.2}$$

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Hence,

$$q_1 = CV = \frac{4}{3} \mu C$$
 (2.3)

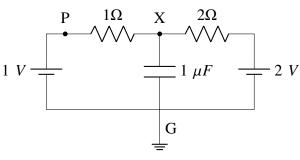


Fig. 2.2

3. Show that the Laplace transform of u(t) is $\frac{1}{s}$ and find the ROC.

Solution: We have,

$$u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} \int_0^\infty u(t)e^{-st}dt$$
 (2.4)

$$= \int_0^0 \frac{1}{2} e^{-st} dt + \int_0^\infty e^{-st} dt$$
 (2.5)

$$=\frac{1}{s}, \quad \Re(s) > 0 \tag{2.6}$$

4. Show that

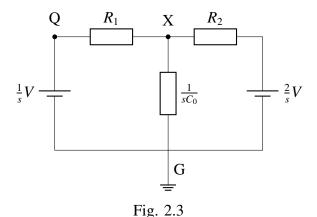
$$e^{-at}u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} \frac{1}{s+a}, \quad a > 0$$
 (2.7)

and find the ROC.

Solution: Note that by substituting s := s + a in (2.6), and considering $a \in \mathbb{R}$,

$$e^{-at}u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} \int_0^\infty u(t)e^{-(s+a)t}dt \qquad (2.8)$$
$$= \frac{1}{s+a}, \quad \Re(s) > -a \qquad (2.9)$$

5. Now consider the following resistive circuit transformed from Fig. 2.1 where



$$u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} V_1(s)$$
 (2.10)

$$2u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} V_2(s) \tag{2.11}$$

Find the voltage across the capacitor $V_{C_0}(s)$. **Solution:** We see that

$$V_1(s) = \frac{1}{s}V_2(s) = \frac{2}{s}$$
 (2.12)

Now, labelling points G and X as in Fig. 2.2, we use KCL at X.

$$\frac{V - \frac{1}{s}}{R_1} + \frac{V - \frac{2}{s}}{R_2} + sC_0V = 0 \tag{2.13}$$

$$V\left(\frac{1}{R_1} + \frac{1}{R_2} + sC_0\right) = \frac{1}{s}\left(\frac{1}{R_1} + \frac{2}{R_2}\right)$$
 (2.14)

$$V(s) = \frac{\frac{1}{R_1} + \frac{2}{R_2}}{s\left(\frac{1}{R_1} + \frac{1}{R_2} + sC_0\right)}$$
(2.15)

$$= \frac{\frac{1}{R_1} + \frac{2}{R_2}}{\frac{1}{R_1} + \frac{1}{R_2}} \left(\frac{1}{s} - \frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right)$$
(2.16)

6. Find $v_{C_0}(t)$. Plot using python. **Solution:** Taking the inverse Laplace transform

in (2.16),

$$V(s) \stackrel{\mathcal{L}}{\longleftrightarrow} \frac{2R_1 + R_2}{R_1 + R_2} u(t) \left(1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}} \right)$$

$$= \frac{4}{3} \left(1 - e^{-\left(1.5 \times 10^6\right)t} \right) u(t)$$
(2.18)

The python code plots the graph below.

```
import numpy as np
from matplotlib import pyplot as plt
def unitStep(t):
    if (t < 0): return 0
    elif (t == 0):return 0.5
    else: return 1
def v1(t):
    if (t \ge 0): return 4/3*(1 - np.exp(-t))
         *1.5e6))*unitStep(t)
    else: return 0
vc0 = np.vectorize(v1, otypes=['double'])
# print(vc0)
\# vc1 = np.loadtxt('v1.txt')
t = \text{np.linspace}(0, 1e-5, 1000000)
plt.plot(t, vc0(t))
# plt.plot(vc1[:,0], vc1[:,1], '.')
plt.xlabel('t_(s)')
plt.ylabel('v \{C 0\}(t)\_(V)')
plt.grid()
plt.legend(['Simulation', 'Analysis'])
plt.show()
```

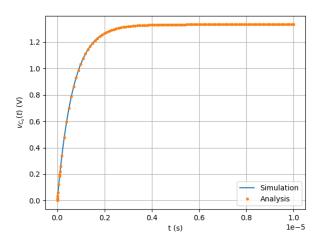


Fig. 2.4: $v_{C_0}(t)$ before the switch is flipped

7. Verify your result using ngspice.

Solution: The ngspice script simulates the given circuit and the generated output is depicted in Fig. (2.4)

Output Before Switching
V1 1 0 dc 1V
R1 X 1 1
C1 X 0 1u ic=0
R2 X 3 2
V2 3 0 dc 2V
.tran 100n 10u uic

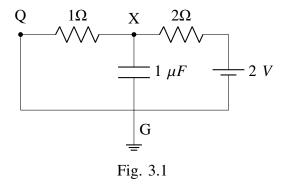
.control
run
wrdata v1.txt x(X)
.endc

.end

3 Initial Conditions

1. Find q_2 in Fig. 2.1.

Solution: The equivalent circuit at steady state when the switch is at Q is shown below.



Since capacitor behaves as an open circuit, we use KCL at X.

$$\frac{V-0}{1} + \frac{V-2}{2} = 0 \implies V = \frac{2}{3} V$$
 (3.1)

and hence, $q_2 = \frac{2}{3} \mu C$.

- 2. Draw the equivalent *s*-domain resistive circuit when S is switched to position Q. Use variables R_1, R_2, C_0 for the passive elements.
- 3. $V_{C_0}(s) = ?$

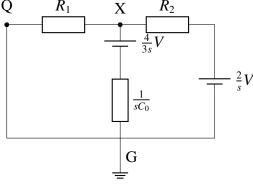


Fig. 3.2

Solution: Using KCL at node X in Fig. 3.2

$$\frac{V-0}{R_1} + \frac{V-\frac{2}{s}}{R_2} + sC_0\left(V - \frac{4}{3s}\right) = 0$$
 (3.2)

$$\Longrightarrow V_{C_0}(s) = \frac{\frac{2}{sR_2} + \frac{4C_0}{3}}{\frac{1}{R_1} + \frac{2}{R_2} + sC_0}$$
 (3.3)

4. $v_{C_0}(t) = ?$ Plot using python.

Solution: From (3.3),

$$V_{C_0}(s) = \frac{4}{3} \left(\frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right) + \frac{2}{R_2 \left(\frac{1}{R_1} + \frac{1}{R_2} \right)} \left(\frac{1}{s} - \frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right)$$
(3.4)

Taking an inverse Laplace Transform,

$$v_{C_0}(t) = \frac{4}{3}e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}u(t) + \frac{2}{R_2\left(\frac{1}{R_1} + \frac{1}{R_2}\right)}\left(1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}\right)u(t)$$
(3.5)

Substituting values gives

$$v_{C_0}(t) = \frac{2}{3} \left(1 + e^{-(1.5 \times 10^6)t} \right) u(t)$$
 (3.6)

The Python code plots the graph below.

import numpy as np
from matplotlib import pyplot as plt

def unitstep(t):
 if (t < 0): return 0
 elif (t == 0):return 0.5
 else: return 1</pre>

def v1(t):

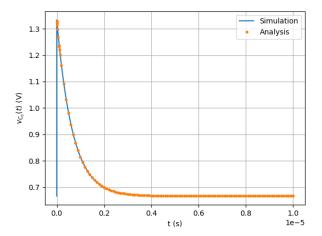


Fig. 3.3: $v_{C_0}(t)$ after the switch is flipped

5. Verify your result using ngspice.

Solution: The ngspice script

Output After Switching
R1 X 0 1
C1 X 0 1u ic=1.33
R2 X 3 2
V2 3 0 dc 2V
.tran 100n 10u uic

.control
run
wrdata v2.txt v(X)
.endc

.end

simulates the given circuit and the generated output is depicted in Fig. (3.3).

6. Find $v_{C_0}(0-)$, $v_{C_0}(0+)$ and $v_{C_0}(\infty)$.

Solution: From the initial conditions,

$$v_{C_0}(0-) = \frac{q_1}{C} = \frac{4}{3} V$$
 (3.7)

Using (3.6),

$$v_{C_0}(0+) = \lim_{t \to 0+} v_{C_0}(t) = \frac{4}{3} V$$
 (3.8)

$$v_{C_0}(\infty) = \lim_{t \to \infty} v_{C_0}(t) = \frac{2}{3} V$$
 (3.9)

7. Obtain the Fig. in problem 3.2 using the equivalent differential equations.

Solution: The equivalent circuit in the *t*-domain is shown below.

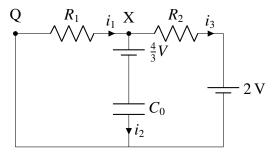


Fig. 3.4

From KCL and KVL,

$$i_1 = i_2 + i_3 \tag{3.10}$$

$$i_1 R_1 + \frac{4}{3} + \frac{1}{C_0} \int_0^t i_2 dt = 0$$
 (3.11)

$$\frac{4}{3} + \frac{1}{C_0} \int_0^t i_2 dt - i_3 R_2 - 2 = 0$$
 (3.12)

(3.13)

Taking Laplace Transforms on both sides and using the properties of Laplace Transforms,

$$I_1 = I_2 + I_3 \tag{3.14}$$

$$I_1 R_1 + \frac{4}{3} + \frac{1}{sC_0} I_2 = 0 {(3.15)}$$

$$\frac{4}{3} + \frac{1}{sC_0}I_2 - I_3R_2 - 2 = 0 \tag{3.16}$$

(3.17)

where $i(t) \stackrel{\mathcal{L}}{\longleftrightarrow} I(s)$. Note that the capacitor is equivalent to a resistive element of resistance $R_C = \frac{1}{sC_0}$ in the *s*-domain. Equations (3.14) - (3.16) precisely describe Fig. 3.2.