

EMFT Project

by Harsh Singh

Submission date: 15-Nov-2019 08:30PM (UTC+0530)

Submission ID: 1214456528

File name: crdn.pdf (986.99K)

Word count: 1535

Character count: 8332

Analysing connectivity between physical dimensions and Y parameters of CRDN for wideband applications

Harsh Singh
Department of Electronics and
Communication
IIT Naya Raipur
Naya Raipur, India
harshs17101@iitr.edu.in

Abstract— a new approach for isolating two adjacently placed coupled antenna is proposed in this paper. The technique uses coupled resonator decoupling network to nullify the effect of coupling between two antennas. Placing CRDN in between the antennas gives rise to different couplings in opposite direction of what previously bare antennas were possessing. Construction and design of CRDN is presented based on the admittance parameter which has been further converted into form of coupling matrix through rational function and analytical procedure involving simple mathematics. A clear relation is depicted in elements of the coupling matrix and physical dimension of resonators. The comparative results obtained through experiments and simulation for different designs proves electrical parameters are reliable with physical dimensions of CRDN. Proposed approach presents a way to apply it for asymmetrical antennas and MIMO systems.

Keywords—CRDN, Admittance parameters, Coupling Matrix, electrical parameters, MIMO

I. INTRODUCTION

Mutual coupling between antennas is a popular unwanted effect, which reduces the performance of antenna arrays. Isolating these antennas has been a popular research topic among the researchers. Both the industry people and academicians have paid significant attention to rectify this problem. If talking in general then it can be said that coupling between antennas is unexpectedly different for n th receiver antenna and transmitter antenna. Broadband transmitting and receiving antenna arrays have become essential in a high end wireless communication system. When multi antenna structure is planned inside mobile terminal, the antennas have to be placed in very compact space which brings high mutual coupling between the elements. Additionally this mutual coupling wastes a large portion of power fed into coupling from one port to another port rather than radiating the same power in free space. Various schemes of isolation have been proposed. The most used one are described.

A. The space, polarization and angle diversity scheme

The simplest method to isolate two antennas is to increase the electrical distance between them [2] [3] [4]. When the angular speed of the channel completes the full

sphere, the additional phase difference of two antennas will optimally de-correlate the incoming signal. Although in general three dimensional angular spread is rarely implemented. It is observed in inverted-F and inverted-L which are commonly used in mobile terminals that an increment in inter resonator distance results in weak isolation [5] [6]. By exciting the electric and magnetic mode on body, polarization diversity is achieved. In spite of getting advantage of this scheme motor, camera and battery's effect on orthogonality of two antennas can't be neglected. Due to this it is practically not implementable in mobile terminals.

B. Eign-mode decomposition scheme

In the 1970's, Andersen [7] suggested an essential condition for no coupling between antennas and proved it by inserting a part of transmission-line between the coupled antenna ports. The constraint is that the antenna spacing has to be kept at fixed values. A lumped element connected with the coupled antennas can also get a fixed level of isolation [8], [9]. Such decoupling techniques can be considered as a decoupling network of zeroth-order with relatively narrow decoupling bandwidth, which results in high sensitivity to the surroundings of the antennas in touch such as human hands.

C. Coupled Resonator decoupling network(CRDN)

This technique introduces both sequential and cross coupling for a wideband decoupling performance. Compared to already existing schemes the presented scheme is having following special features: 1) it produces higher order isolation solution; 2) It provides a tradeoff between decoupling bandwidth and degree of isolation; 3) It is easily implemented in multiple antenna system[10]; 4) Currently available filter design techniques are used in realization. The synthesis and realization part of CRDN begins from finding matching and decoupling conditions. In this paper CRDN is realized using micro strip resonator. It can also be realized from low temperature co-fired ceramics and silicon based integrated passive devices.

II. SYNTHESIS AND DESIGN OF CRDN

A. Condition for decoupling and matching

For the second order system it is assumed that a pair of coupled antenna is described with admittance matrix

(Y^A). The decoupling network comprising two **1**crostrip resonators in series which basically form **CRDN is connected parallel to** this coupled antennas. **The** entries in admittance matrix for CRDN (Y^F) consist only purely imaginary values. After combining antennas with CRDN the final admittance matrix will be the sum of both the admittance matrix.

$$Y = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} = \begin{bmatrix} Y_{11}^A + Y_{11}^F & Y_{12}^A + Y_{12}^F \\ Y_{21}^A + Y_{21}^F & Y_{22}^A + Y_{22}^F \end{bmatrix} \quad (1)$$

Decoupling Condition is taken as described in the [1].

$$j.Im\{Y_{21}^A(w_B)\} + Y_{21}^F(w_B) \approx 0 \quad (2)$$

Matching Condition is obtained as described in the [1].

$$j.Im\{Y_{kk}^A(w_B)\} + Y_{kk}^F(w_B) \approx 0, \quad k=1,2 \quad (3)$$

1 B. Synthesis of decoupling network

Coupling matrix terminology **which is** actually built for designing filters, can also be applied for realizing CRDN. The complete process starts by **1**nstructing rational function for admittance values. For a second order CRDN without zeros **the** mutual admittance can be expressed as

$$Y_{21}^F(s) = \frac{j\gamma}{(s - jp_1)(s - jp_2)} \quad (4)$$

Self coupling can be represented from [1].

$$Y_{11}^F(s) = \frac{c\gamma}{(s - jp_1)(p_2 - p_1)} + \frac{d\gamma}{(s - jp_1)(p_2 - p_1)} \quad (5)$$

$$Y_{22}^F(s) = \frac{\gamma}{c(s - jp_1)(p_2 - p_1)} + \frac{\gamma}{d(s - jp_1)(p_2 - p_1)} \quad (6)$$

where c, d, γ , p_1 , p_2 are constants. As it is

$$s = \sigma + j\omega \quad (7)$$

we can always analyse the function in imaginary axis[1]. So $\sigma = 0$

$$s = j\omega \quad (8)$$

For a symmetrical network $c = d = 1$ and $-p_1 = p_2 = p$. After putting equation 8 into equation 5, 6, 4 and then putting equation 5 and 6 in equation 3 and equation 4 in equation 2 following equation are obtained by combining it with the conditions for a symmetrical network.

$$Im(Y_{21}^A(w_1)) = \frac{\gamma}{(\omega_1^2 - p^2)} \quad (9)$$

$$Im(Y_{11}^A(w_2)) = \frac{|\gamma|}{p(\omega_2 + p)} \quad (10)$$

$$Im(Y_{22}^A(w_2)) = \frac{|\gamma|}{p(\omega_2 + p)} \quad (11)$$

where w_1 and w_2 are frequencies related to wideband of antennas. After manipulating equation 9 with equation 10 or equation 11 following polynomial is obtained by which we can calculate the value of p and value of γ .

$$(Im\{Y_{21}^A(w_1)\} + Im\{Y_{11}^A(w_2)\})p^2 + (\omega_2^2)p - d\omega_1^2 = 0 \quad (12)$$

$$\gamma = cp(\omega_2 + p) \quad (13)$$

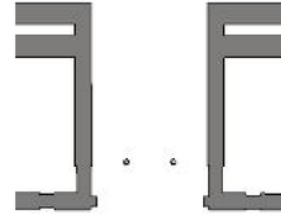
For a symmetrical network coupling elements are given from [1].

$$m_{s1} = m_{l2} = \sqrt{\frac{|\gamma|}{p}} \quad (14)$$

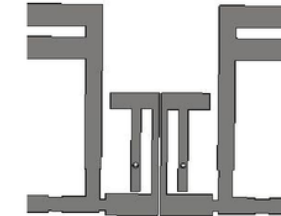
$$m_{l2} = \frac{|\gamma|p}{\gamma} \quad (15)$$

$$m_{11} = m_{22} = m_{sl} = 0 \quad (16)$$

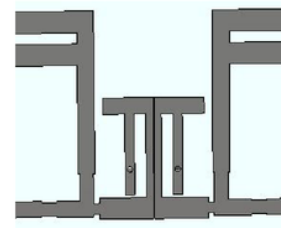
C. Designs used in simulations



(a)



(b)



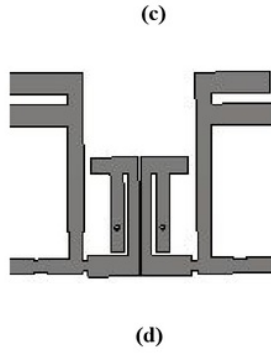


Fig. 1. Pairs of antennas for (a) Without CRDN (b)With CRDN having inter resonator distance 0.4 mm and inner resonator distance 1.64 mm (c) With CRDN having inter resonator distance 0.2 mm and inner resonator distance 1.64 mm (d) With CRDN having inter resonator distance 0.4 mm and inner resonator distance as 0.41 mm.

D. Results obtained for all the design

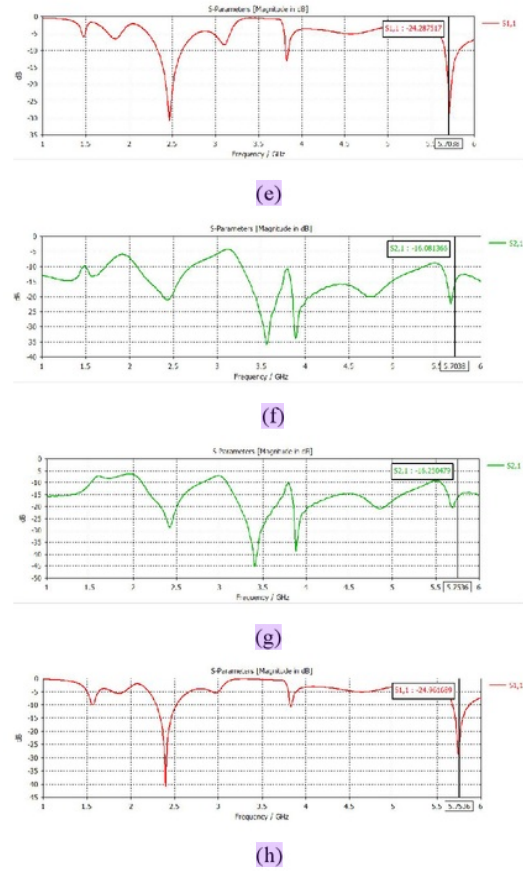
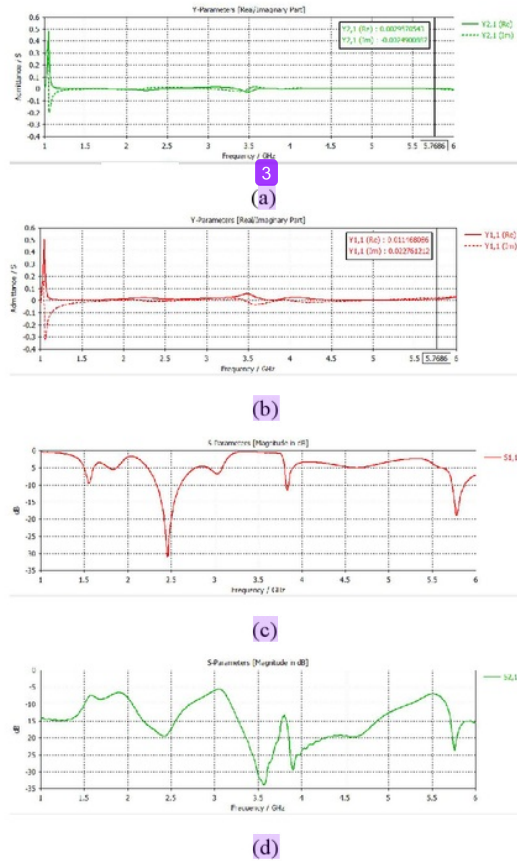


Fig. 2. (a) and (b) Y parameters variation for pair of antennas without CRDN, (c) and (d) s parameter variation for design in Fig. 1. (b), (e) and (f) s parameter variation for design in Fig. 1. (c), (g) and (h) for design in Fig. 1. (d).

F. Comparing results obtained through all designs

Using results obtained in ii (D) and analytical procedure described in ii (B) a matlab script is used with optimum value of frequencies and values of y parameters to get the values of coupling elements. The script follows the same mathematical procedure to calculate the roots of the quadratic equation 12 which by using put the value of the optimum root in equation 13 to get the value of γ . screenshot of the script is attached below with results obtained through the written script in matlab.

```

1 %Antenna_decoupling.m <| +
2 %inter resonator coupling is corresponding to m12
3 %inner resonator coupling is corresponding to m11 and m12
4 w21=[3.56 4.34 4.1];
5 w11=[2.117 2.4 3.9];
6 imy11=[0.004584 0.0025 0.0044];
7 imy21=[0.01543 0.03 0.0040];
8 xsquarecoeff=imy11-imy21;
9 xcoeff=w11.*imy11;
10 constant=(-1).*imy21.*w21.^2;
11 r = [0 0 0; 0 0 0];
12 for i=1:3
13 p = [xsquarecoeff(i) xcoeff(i) constant(i)];
14 r(i,i) = roots(p);
15 if r(i,i)>0
16 root(i)=r(i,i);
17 else
18 root(i)=r(2,i);
19 end
20 end
21 gama=imy11.*root.*(w11+root);
22 m12=root;
23 m11=sqrt(gama./root);
24 m21=m11;

```

COMMAND WINDOW

New to MATLAB? See resources for Getting Started.

```

m12 =
    2.6928    4.0785    1.9866
>> m11
m11 =
    0.1515    0.1273    0.1669

```

Results are obvious as we decrease inter resonator distance m_{12} increases and if we decrease inner resonator distance m_{11} increases.

III. CONCLUSION

This paper presented a new technique for isolating two antennas. Focus of this paper goes in relating physical dimensions of CRDN with it's electrical equivalents which is made possible through Y parameter of antennas. After fetching coupling elements such as m_{21} , m_{11} , m_{22} , m_{12} from Y matrix for all the designs physical dimensions seems relatable with electrical equivalents. As inter-resonator distance keeps decreasing the coupling element m_{21} also keeps increasing and as inner resonator distance keeps decreasing the coupling element m_{11} and m_{12} also keep increasing.

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