Lecture 16

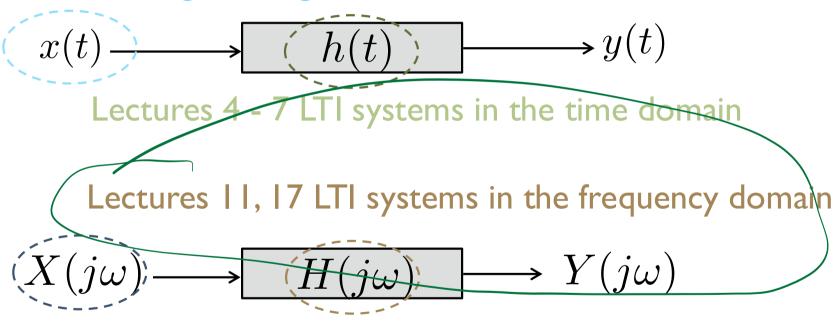
Introduction to bode plots

Preview of today's lecture

- ◆ LTI systems in the frequency domain
 - → LTI systems act as filters on signals
 - → Frequency response of a system characterizes the filter
- ◆ Introduction to Bode plots
 - → Bode plots are a way to sketch the frequency response of a system
 - Plots are constructed with careful use of logarithms
- Bode plot of an RC filter
 - → Bode plots are relevant for systems described by LCCDEs
 - → Bode plots are built from the understanding of first order systems like those described by RC circuits

Connections back to ECE 45

Lectures 2 - 3 working with signals



Lectures 8 - 10 Fourier series

Lectures II - 16 Fourier transform



LTI systems in the frequency domain

Key points

- LTI systems act as filters on signals
- o Frequency response of a system characterizes the filter

Linear and time invariant systems

◆ Fully characterized by their impulse response

$$\delta(t) \longrightarrow \text{System} \longrightarrow h(t)$$

Output computed from the impulse response for any input

$$x(t) \longrightarrow h(t) \longrightarrow y(t)$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$
 Convolution

Response to a complex sinusoid

◆ Frequency response is used to characterize LTI systems

$$e^{j\omega t} \longrightarrow h(t) \longrightarrow H(j\omega)e^{j\omega t}$$

◆ The frequency response is computed from the impulse response

$$H(j\omega) = \int_{-\infty}^{\infty} h(t)e^{-j\omega t}dt$$

◆ This is the Fourier transform of the impulse response of the system

$$h(t) \stackrel{FT}{\longleftrightarrow} H(j\omega)$$

Specialization to real signals

♦ For any complex sinusoid

$$e^{j\omega_0 t} \longrightarrow H(j\omega) \longrightarrow H(j\omega_0)e^{j\omega_0 t}$$

Response to a periodic signal

• Periodic signals can be represented as a sum of harmonics of complex sinusoids with fundamental frequency $\omega_0 = 2\pi$

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t} \qquad \text{where } x(t) \not\equiv x \not\in x$$

◆ For a periodic signal the output of an LTI system looks like

$$\sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt} \longrightarrow H(j\omega) \longrightarrow \sum_{k=-\infty}^{\infty} a_k H(j\omega_0 k) e^{j\omega_0 kt}$$

multiples of the fundamental frequency

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Response to a periodic signal



 Aperiodic signals can be represented as a linear combination of complex sinusoids via the Fourier transform

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega \qquad \text{of} \qquad \text{of}$$

◆ For a periodic signal, the output of an LTI system looks like

$$y(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(j\omega)X(j\omega)e^{j\omega t}d\omega$$

$$\mathcal{L}(j\omega) = \mathcal{L}(j\omega)X(j\omega)$$

Thinking about LTI systems as filters

◆ LTI systems in time domain: convolution with an impulse response

$$x(t) \longrightarrow h(t)$$

◆ LTI systems in frequency domain: mult. with frequency response

$$X(j\omega) \longrightarrow H(j\omega) \longrightarrow Y(j\omega)$$

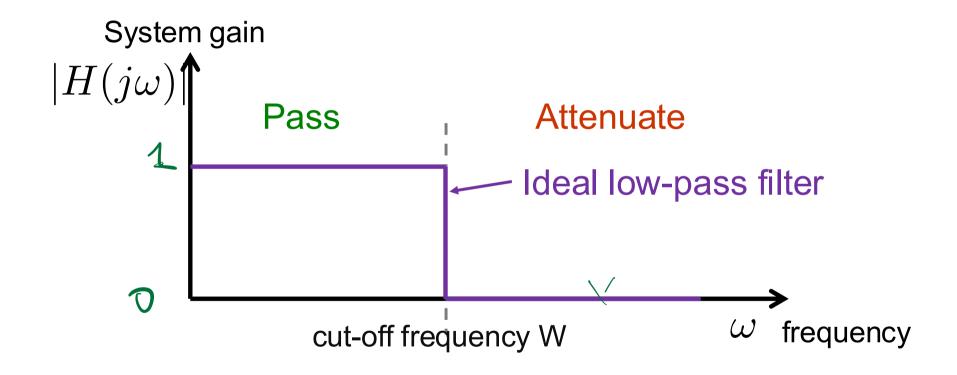
Can think of LTI systems in time or frequency domains

Introduction to Bode plots

Key points

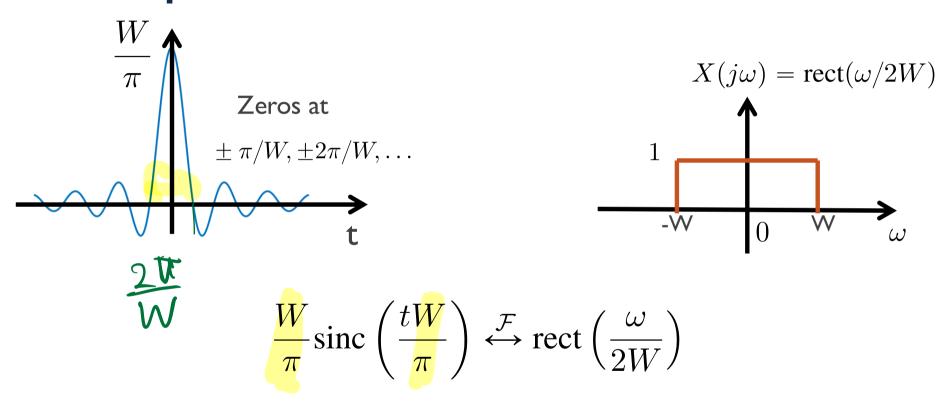
- o Bode plots are a way to sketch the frequency response of a system
- Plots are constructed with careful use of logarithms

Recall the ideal low-pass filter



Systems passes low frequencies, attenuates high frequencies

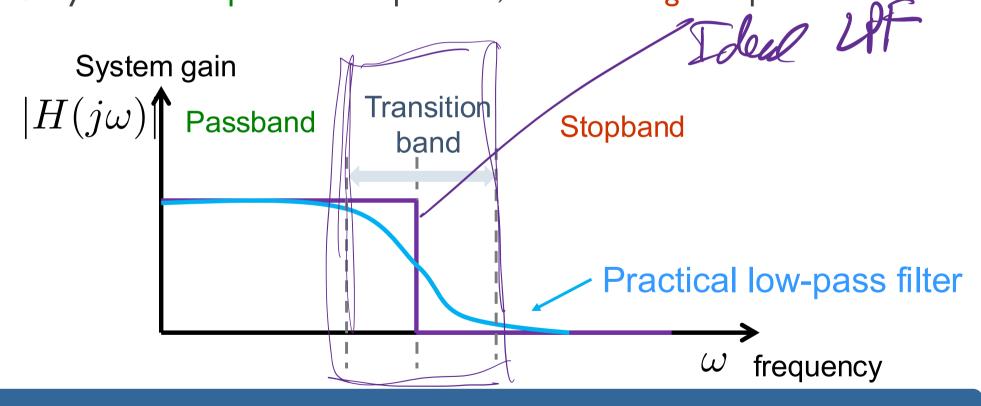
Ideal low-pass filters



Ideal low-pass filter is a sinc function

Practical low-pass filters

Systems that pass low frequencies, attenuate high frequencies



Practical filters make some compromises

Filtering using circuits

 RLC circuits have systems that are described by linear constant coefficient differential equations



◆ Such systems have a frequency response of the form

$$\sum_{k=0}^{N} a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^{M} b_k \frac{d^k x(t)}{dt^k}$$

$$\Rightarrow H(j\omega) = \frac{\sum_{k=0}^{M} b_k (j\omega)^k}{\sum_{k=0}^{N} a_k (j\omega)^k}$$

Bode plots

Bode plots are a way to sketch the frequency response

$$H(j\omega) = |H(j\omega)|e^{j\angle H(j\omega)}$$

Magnitude function

$$|H(j\omega)|$$

Phase function

$$\angle H(j\omega)$$

Used as an alternative or complement to computerbased simulation



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Digression on logs

$$c = \log_b a \iff b^c = a$$

- Most common in signals and systems
 - ullet Natural log $\ln = \log_e$ and log base 10 \log_{10}



→ Huge numbers → small positive numbers

$$\log_{10} \underbrace{10,345,034,896}_{\approx 10^{10}} = 10$$

- → Tiny numbers → small negative numbers
- + Multiplication \rightarrow addition $\log ky = \log x \log y$
- → Division → subtraction
- Used to compute amplifier gain, antenna gain, losses, etc.

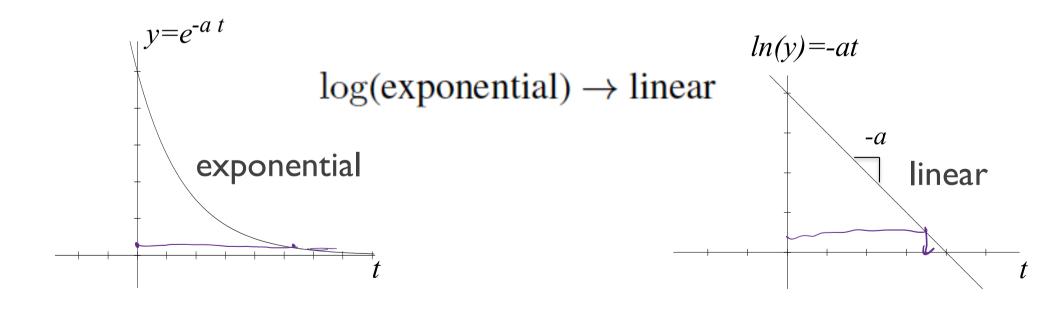


Napier

$$\log xy = \log x + \log y$$
$$\log x^y = y \log x$$
$$\log_b b = 1$$
$$\log_b 1 = 0$$
$$\log_b 0 = \text{undefined}$$

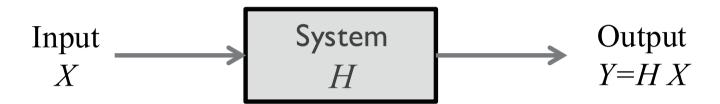
Relationship between logarithms and exponentials

$$\log a^x = x \log a$$



Use of logarithms to measure gain

◆ The Decibel (Deci = ten, bel = "Bell Labs")



- ◆ General rule
 - ullet Use $H [\mathrm{dB}] = 10 \log_{10} |H|^2$ dealing with power
 - ullet Use $\left(H \left[\mathrm{dB} \right] = 20 \log_{10} |H| \right)$ dealing with amplitude

$$Gain[dB] = 10 log_{10} \frac{P_{out}}{P_{in}}$$
 decibels are a unitless ratio of **powers**

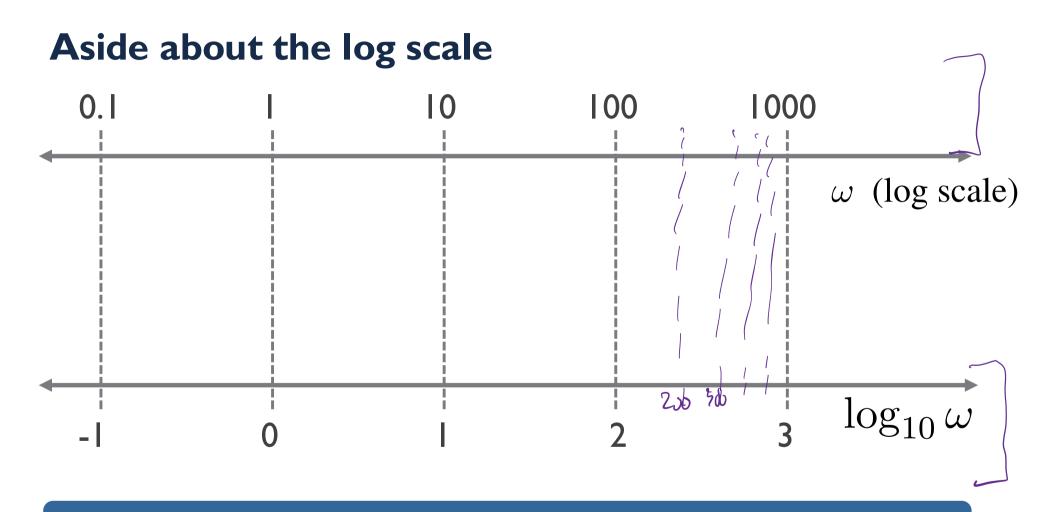
Example dB calculation

◆ Determine the gain (in dB) of the an amplifier with linear gain

$$ightharpoonup$$
 H = 20 $20 \log_{10} 20 = 20 \log_{10} 2(10)$ $= 2 \cdot 10 \log_{10} 2 + 2 \cdot 10 \log_{10} 10$ $= 26 dB$

$$\bullet$$
 H = 0.1 $20 \log_{10} 0.1 = 20(-1) = -20 dB$

♦
$$\mathbf{H} = \mathbf{50}$$
 $20 \log_{10} 50 = 20 \log_{10} 5(10)$ $= 20 \log_{10} 5 + 20 \log_{10} 10$ $= 14 dB + 20 dB$ $= 34 dB$



Equivalent but the numbers are easier to read on the log scale

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Back to Bode plots

- ◆ A Bode plot is a collection of two plots
- lacktriangle The x-axis is "semi-log x" (effectively plots $\log_{10}(\omega)$)
- ◆ The magnitude plot approximates

$$|H(j\omega)|_{\mathrm{dB}} = 20 \log_{10} |H(j\omega)|$$

◆ The phase plot approximates

$$\angle H(j\omega)$$

(recall this is implemented with arctan2)

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Summary

- ◆ Bode plots are convenient ways to illustrate the frequency response of a filter in terms of the magnitude and the phase
- ◆ Magnitude is measured using the decibel system building upon baseten logarithms with a 20 x (amplitude) or 10 x (power) factor
- ◆ X-axis is frequency in radians but plotted in a log scale (this is called semilogx in MATLAB)

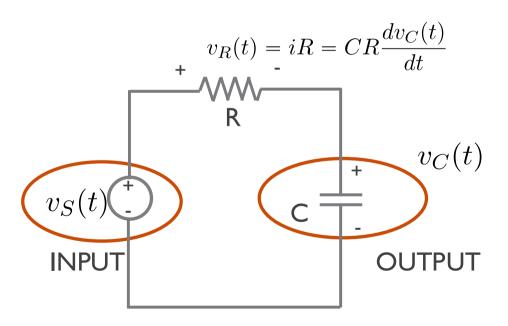
Bode plot of an RC filter

Key points

- Bode plots are relevant for systems described by LCCDEs
- Bode plots are built from the understanding of first order systems like those described by RC circuits

Recall the RC system

- ◆ Source voltage as the input
- ◆ Capacitor voltage as the output



Current through capacitor $i = C \frac{dv_C(t)}{dt}$

Resulting differential equation

$$RC\frac{dv_C(t)}{dt} + v_C(t) = v_S(t)$$

RC y'(t) +y(t)
$$= \chi(t)$$

Finding the frequency response

◆ Think of this circuit as an LTI system

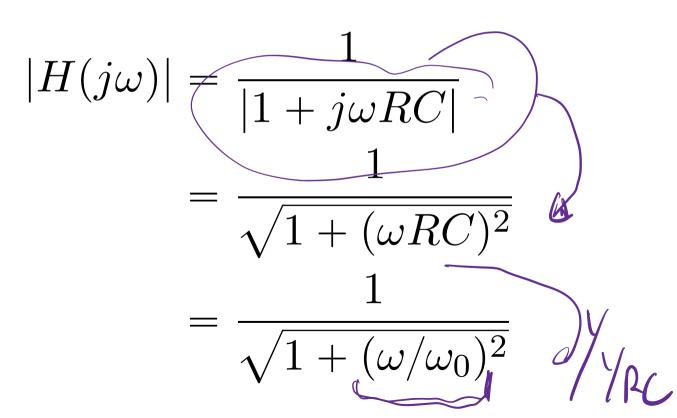
$$v_S(t) \longrightarrow H(j\omega) \longrightarrow v_C(t)$$

◆ The frequency response of this system described by an LCCDE

$$H(j\omega) = \frac{\sum_{k=0}^{M} b_k(j\omega)^k}{\sum_{k=0}^{N} a_k(j\omega)^k} \qquad H(j\omega) = \frac{1}{1 + jRC\omega}$$

Computing the magnitude

$$|H(j\omega)|_{dB} = 20 \log_{10} |H(j\omega)|$$



where the cutoff frequency is

$$\omega_0 = \frac{1}{RC}$$

Now going to decibels

Inputting into the log function

$$|H(j\omega)|_{dB} = 20 \log_{10} \frac{1}{\sqrt{1 + (\omega/\omega_0)^2}}$$

$$= 20 \log_{10} 1 - 20 \log_{10} \sqrt{1 + (\omega/\omega_0)^2}$$

$$= -20 \log_{10} \sqrt{1 + (\omega/\omega_0)^2}$$

Summarizing the final expression

$$|H(j\omega)|_{dB} = -20\log_{10}\sqrt{1 + (\omega/\omega_0)^2}$$

Asymptotic regimes for small frequencies

$$|H(j\omega)|_{dB} = -20 \log_{10} \sqrt{1 + (\omega/\omega_0)^2}$$

For small frequencies

$$\omega \ll \omega_0 \Rightarrow \left(\frac{\omega}{\omega_0}\right)^2 \approx 0$$

Incorporating this assumption

$$|H(j\omega)|_{dB} = -20\log_{10}\sqrt{1+0}$$
$$= -20\log_{10}1 = 0$$

Asymptotic regimes for large frequencies

◆ For large frequencies

$$\omega \gg \omega_0 \Rightarrow 1 + \left(\frac{\omega}{\omega_0}\right)^2 \approx \left(\frac{\omega}{\omega_0}\right)^2$$

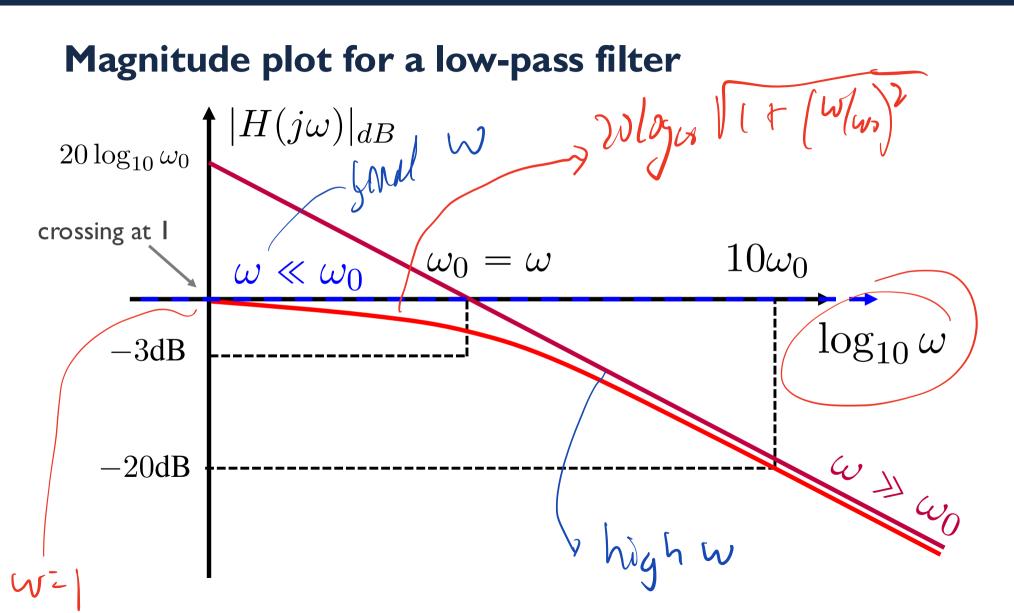
♦ Incorporating this assumption

$$|H(j\omega)|_{dB} = -20\log_{10}\sqrt{1 + \left(\frac{\omega}{\omega_0}\right)^2}$$

$$\approx -20\log_{10}\sqrt{\left(\frac{\omega}{\omega_0}\right)^2}$$

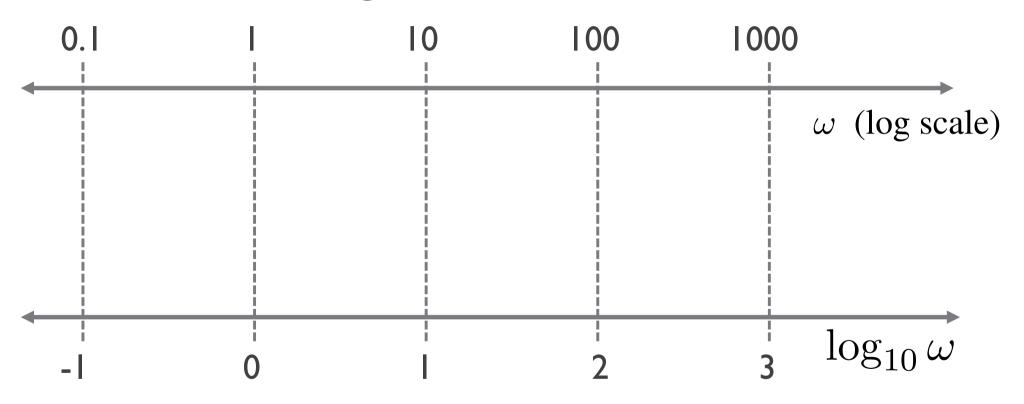
$$= -20\log_{10}\frac{\omega}{\omega_0} = -20\log_{10}\omega + 20\log_{10}\omega$$

$$|H(j\omega)|_{dB} = 20\log_{10}\omega_0 - 20\log_{10}\omega$$



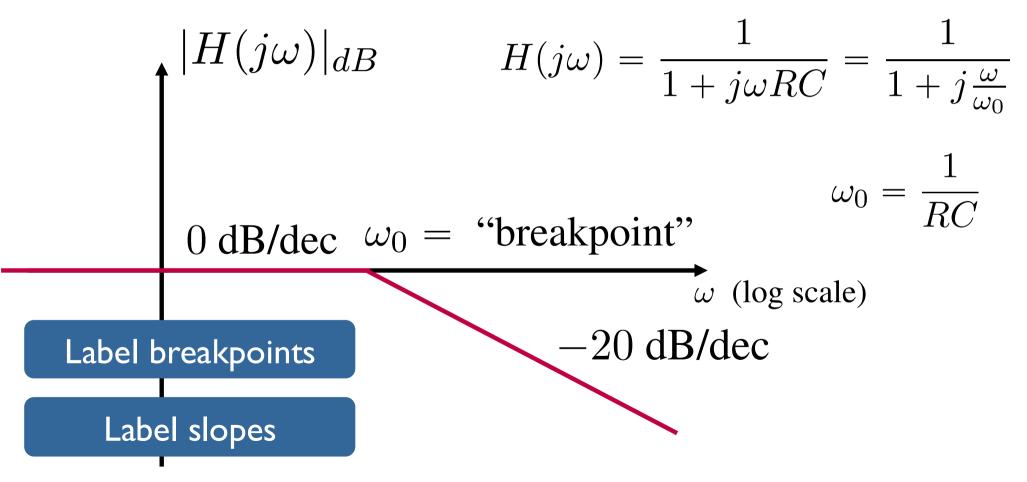
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Aside about the log scale



Equivalent but the numbers are easier to read on the log scale

Bode magnitude plot of the low-pass filter



Bode phase of the low-pass filter

Computing the phase

$$H(j\omega) = \frac{1}{1 + jRC\omega}$$

$$\angle H(j\omega) = \operatorname{atan2}(0,1) - \operatorname{atan2}(\omega RC,1)$$
$$= 0 - \tan^{-1}(\omega RC)$$

For large frequencies

$$\lim_{\omega \to \infty} -\tan^{-1}(\omega RC) = -\frac{\pi}{2}$$

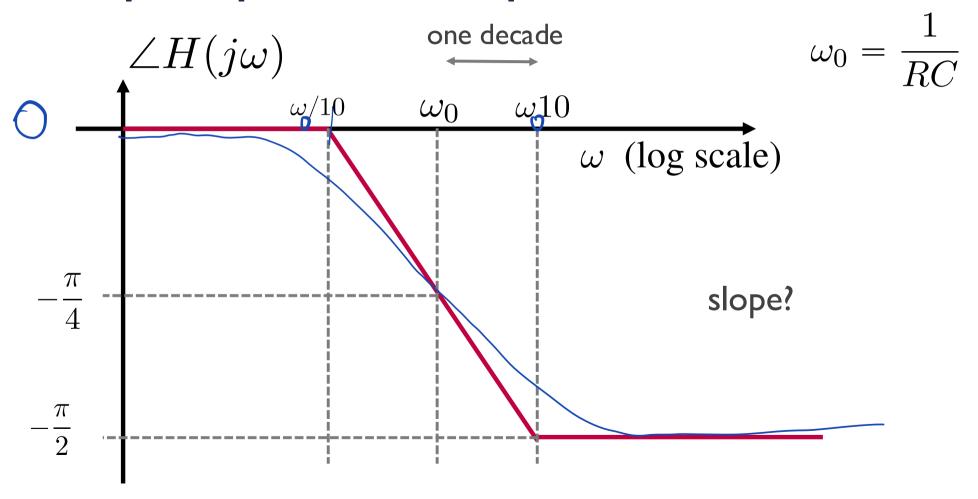
◆ At the critical frequency

$$\omega = \omega_0 \Rightarrow -\tan^{-1}(1) = -\frac{\pi}{4}$$

◆ For small frequencies

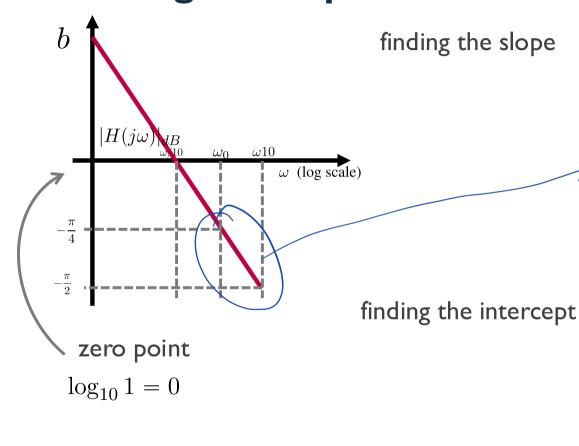
$$\omega = 0 \Rightarrow -\tan^{-1}(0) = 0$$

Bode phase plot of the low-pass filter



Finding the slope

$$y = mx + b$$



$$m = \frac{\text{rise}}{\text{run}}$$

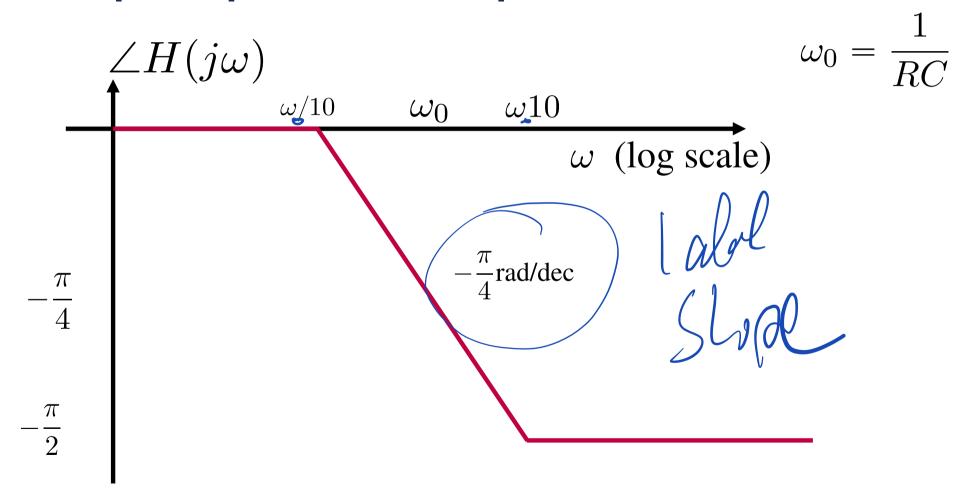
$$= \frac{\frac{\pi}{4}}{\log_{10} \omega_0 - \log_{10} 10\omega_0}$$

$$= \frac{\frac{\pi}{4}}{\log_{10} \omega_0 - \log_{10} \omega_0 - \log_{10} 10}$$

$$= -\frac{\pi}{4}$$

 $0 = m \log_{10} \frac{\omega_0}{10} + b$ $b = -m \log_{10} \frac{\omega_0}{10}$ $= \frac{\pi}{4} (\log_{10} \omega_0 - 1)$

Bode phase plot of the low-pass filter



Summary of the Bode equations for a low-pass filter

Magnitude

$$|H(j\omega)|_{dB} \approx \begin{cases} 0 \text{ dB} & \omega < \omega_0 \\ -20\log_{10}\omega + 20\log_{10}\omega_0 & \omega \geqslant \omega_0 \end{cases}$$

Phase

$$\angle H(j\omega) \approx \begin{cases} 0 & \omega < \omega_0/10 \\ -\frac{\pi}{4} \log_{10} \omega + \frac{\pi}{4} (\log_{10} \omega_0 - 1) & \frac{\omega_0}{10} < \omega < 10\omega_0 \\ -\frac{\pi}{2} & \omega > 10\omega_0 \end{cases}$$

What about a high-pass filter?

$$H(j\omega) = 1 + jRC\omega$$

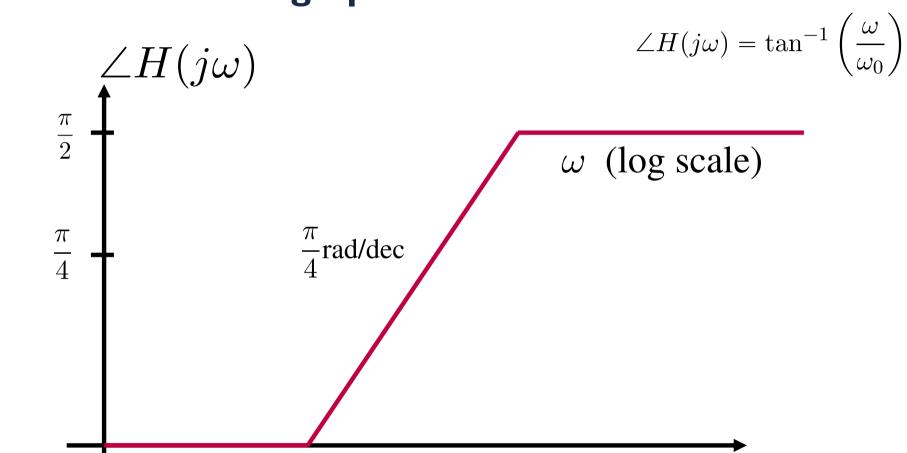
$$|H(j\omega)|_{\mathrm{dB}} = 20\log_{10}\sqrt{1+(\omega/\omega_0)^2}$$
 $|H(j\omega)|_{\mathrm{dB}} = 20\log_{10}\sqrt{1+(\omega/\omega_0)^2}$
 $\omega_0 = \frac{1}{RC}$
 $\omega_0 = \text{"breakpoint"} \quad \omega \text{ (log scale)}$

What about a high-pass filter?

 $\omega/10$

$$H(j\omega) = 1 + jRC\omega$$

 ω (log scale)



 ω_0

 $\omega 10$

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Summary

- ◆ The Bode plot of a simple RC filter is approximated using three lines
- Magnitude plot
 - → Plot frequency on the log scale vs magnitude in dB
 - → Label the breakpoint and slope
- Phase plot
 - + Plot the frequency on the log scale vs phase in radians
 - + Label the endpoints of the transition region, critical frequency and slope
- ♦ Low-pass and high-pass filters have the "opposite" response