

**UNIVERSITY OF CALIFORNIA, SAN DIEGO**

Electrical and Computer Engineering Department

ECE 65 – Fall 2022

*Components and Circuits lab*

Final Exam *Solutions*

- Closed books, two one-sided cheat sheets, and calculators are allowed
- Electronic devices are not allowed.
- Please put all answers in the provided sheets.
- You can use the back of every page as a scratch paper.
- Please scan your answers and cheat sheets and submit them to Gradescope by 11:40 am.

**Please do not begin until you are told to do so.**

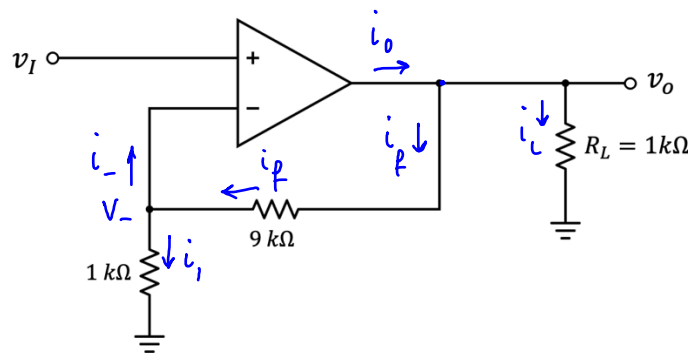
Show your work and good luck!

**Problem 1.**

Consider the following op-amp circuit. Assume an ideal op-amp.

Choose the op-amp saturation voltage, maximum output current, and an AC input voltage such that the output voltage would be distorted because of op-amp's maximum output current.

- What are your selected saturation voltage, maximum output current, and AC input voltage?
- Analyze the circuit and show why the output voltage is distorted.
- Draw the input and output waveforms.



There is more than one correct answer to this problem.

**Show your work.**

op-amp is ideal:  $i_+ = i_- = 0$

there is negative feedback:  $V_+ = V_- = V_I$

KCL at the inverting input terminal:  $i_f = i_1 + i_- = i_1$

$$\frac{V_o - V_-}{9\text{ k}} = \frac{V_-}{1\text{ k}} \rightarrow V_o = 10 V_-$$

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$$V_o = 10 V_- \quad \text{and} \quad V_+ = V_-$$

$$V_o = 10 V_I \quad \text{when we have linear amplification}$$

$$\text{KCL at the output node:} \quad i_o = i_f + i_L$$

$$i_f = \frac{V_o - V_-}{9k} \quad \text{and} \quad V_o = 10 V_- \quad \Rightarrow \quad i_f = \frac{V_o - 0.1 V_o}{9k} = \frac{0.9 V_o}{9k} = \frac{V_o}{10} \text{ (mA)}$$

$$i_L = \frac{V_o}{1} \text{ (mA)}$$

$$i_o = \frac{V_o}{10} + \frac{V_o}{1} \text{ (mA)} = 1.1 V_o \text{ (mA)}$$

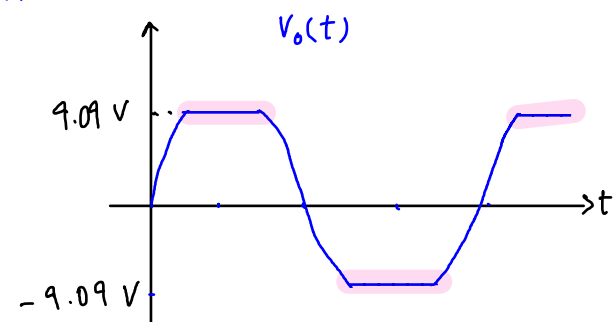
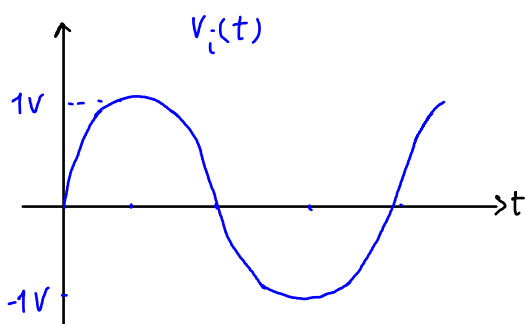
choose some values for  $V_{sat}$ ,  $I_{o,max}$  and choose  $V_I$ :

$V_{sat} = \pm 12V$ ,  $V_I = 1(V) \sin \omega t \rightarrow$  linear amplification will not be limited by the saturation voltage

if  $I_{o,max} = 10 \text{ mA} \rightarrow$  to have  $\hat{V}_o = 10V$ , we need  $\hat{i}_o = 11 \text{ mA}$ , which can not supplied

The output voltage will be distorted because of the op-amp max output current.

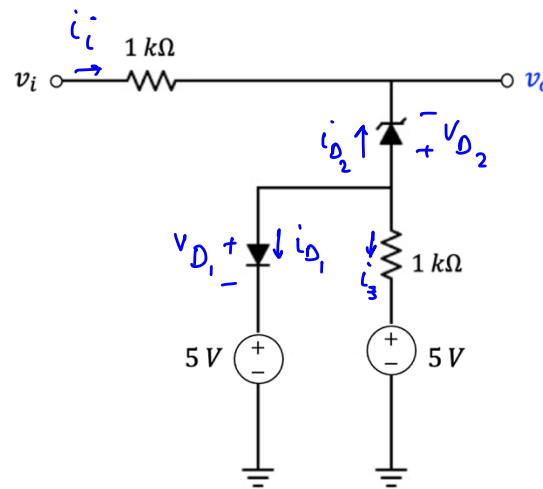
$$i_{o,max} = 1.1 V_{o,max} = 10 \text{ mA} \rightarrow V_{o,max} = \frac{10}{1.1} = 9.09 \text{ (V)}$$



**Problem 3.**

The diodes in the below circuit have  $V_{D0} = 0.7\text{ V}$ , and  $V_Z = 8\text{ V}$ .

- Write the possible cases of the operation of the diodes.
- For each case, include the calculation of finding the relationship between  $v_o$  and  $v_i$  and the range of  $v_i$ .
- Sketch the output signal when  $v_i$  is a sinusoidal signal with a peak amplitude of  $1\text{ V}$ . You do not need to label the time axis.



**Show your work.**

Case 1:  $D_2$  in Zener &  $D_1$  is off

Case 2:  $D_2$  in Zener &  $D_1$  is on

Case 3:  $D_2$  is on &  $D_1$  is off

Case 4:  $D_1$  &  $D_2$  off

$$i_i = -i_{D_2} \quad \text{and} \quad i_{D_1} + i_{D_2} + i_3 = 0$$

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Case 1 analysis:

$D_2$  in Zener &  $D_1$  is off

$$V_{D_2} = -V_Z, \quad i_{D_2} \leq 0, \quad i_{D_1} = 0, \quad V_{O_1} < V_{O_0}$$

KVL 1:

$$V_i = 1k\Omega \times i_i - V_{D_2} + 1k\Omega \times i_3 + 5$$

$$i_i = -i_{D_2} \quad \text{and} \quad i_{D_1} + i_{D_2} + i_3 = 0 \rightarrow i_3 = -i_{D_2}$$

$$V_i = 1k\Omega \times (-i_{D_2}) + 8 + 1k\Omega \times (-i_{D_2}) + 5$$

$$\rightarrow i_{D_2} = \frac{-V_i + 13}{2k\Omega} \quad i_{D_2} \leq 0 \Rightarrow V_i \geq 13V$$

$V_{D_1}$  must also be less than 0.7V

KVL 2:

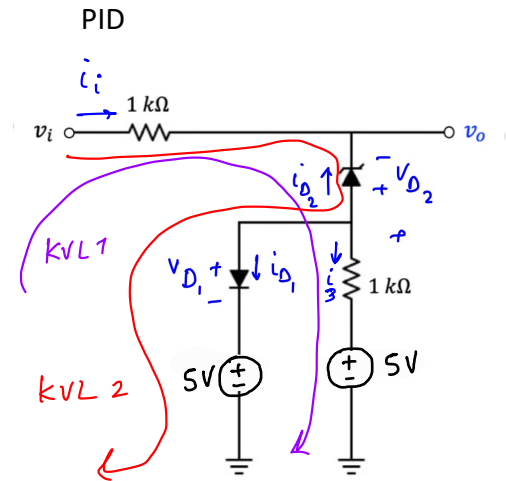
$$V_i = 1k\Omega \times i_i - V_{D_2} + V_{D_1} + 5 = 1k\Omega \times (-i_{D_2}) + 13 + V_{D_1}$$

$$V_i = 1k\Omega \times \left( \frac{-V_i + 13}{2k\Omega} \right) + 13 + V_{D_1} \rightarrow \frac{1}{2} V_i - \frac{13}{2} = V_{D_1}$$

$$V_{D_1} < V_{O_0} \rightarrow \frac{1}{2} V_i - 6.5 < 0.7V \rightarrow V_i < 14.4V$$

the range of  $V_i$  for this case is  $13 \leq V_i < 14.4$

$$V_0 = -V_{D_2} + 1k\Omega \times i_3 + 5 = 8V + 1k\Omega \times (-i_{D_2}) + 5 = 13V + 1k\Omega \times \frac{V_i - 13}{2k\Omega}$$



$$V_0 = \frac{1}{2} V_i + 6.5$$

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Case 2 analysis:

$D_2$  in Zener &  $D_1$  is ON

$$V_{D_2} = -V_Z, \quad i_{D_2} \leq 0, \quad i_{D_1} \geq 0, \quad V_{D_1} = V_{D_0}$$

KVL 1:

$$V_i = 1k\Omega \times i_i - V_{D_2} + 1k\Omega \times i_3 + 5$$

$$i_i = -i_{D_2} = i_{D_1} + i_3$$

KVL 3:

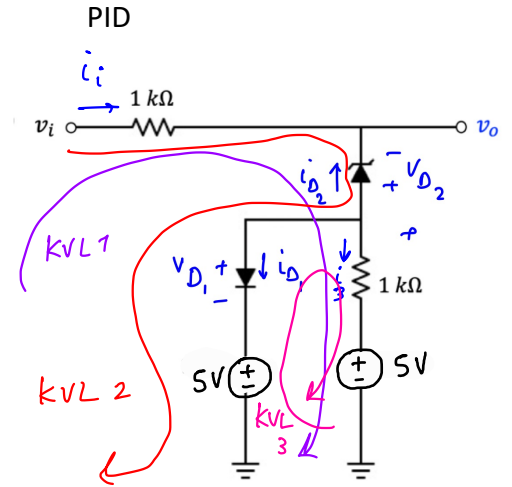
$$-5 - V_{D_1} + 1k\Omega \times i_3 + 5 = 0 \Rightarrow 1k\Omega \times i_3 = V_{D_0}$$

$$V_i = 1k\Omega \times (i_{D_1} + \frac{0.7}{1k\Omega}) + 8 + 0.7 + 5 \rightarrow i_{D_1} = \frac{V_i - 14.4}{1k\Omega}$$

$$i_{D_1} \geq 0 \rightarrow V_i \geq 14.4 \text{ V}$$

the range of  $V_i$  for this case is  $V_i \geq 14.4 \text{ V}$

$$V_o = -V_{D_2} + V_{D_1} + 5 = 13.7 \text{ V} \rightarrow V_o = 13.7 \text{ V}$$



Case 3 analysis:

$D_2$  is ON &  $D_1$  is off

$$V_{D_2} = V_o, \quad i_{D_2} \geq 0, \quad i_{D_1} = 0, \quad V_{D_1} < V_{D_0}$$

KVL 1:

$$V_i = 1k\Omega \times i_i - V_{D_2} + 1k\Omega \times i_3 + 5$$

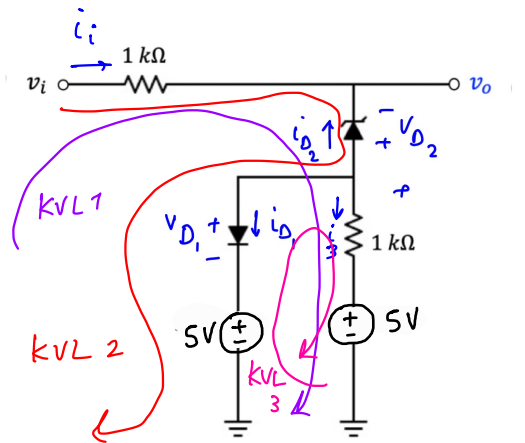
$$i_i = -i_{D_2}, \quad i_3 = -i_{D_2}$$

$$V_i = 1k\Omega \times (-i_{D_2}) - 0.7 + 1k\Omega \times (-i_{D_2}) + 5$$

$$\rightarrow i_{D_2} = \frac{-V_i + 4.3}{2k\Omega} \quad i_{D_2} \geq 0 \Rightarrow V_i \leq 4.3 \text{ V}$$

$$V_o = 1k\Omega \times (-i_i) + V_i = 1k\Omega \times (i_{D_2}) + V_i = -\frac{V_i}{2} + \frac{4.3}{2} + V_i = \frac{1}{2} V_i + \frac{4.3}{2}$$

$$V_o = \frac{1}{2} V_i + \frac{4.3}{2} \text{ V}$$



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Case 4 analysis:

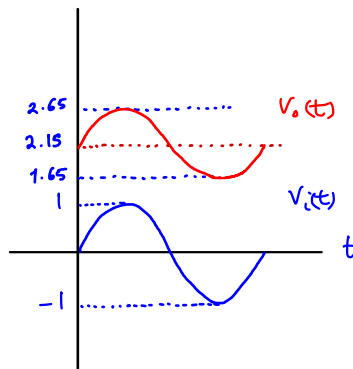
$D_2$  &  $D_1$  are off

$$-V_2 < V_{D_2} < V_{D_0}, \quad i_{D_2} = 0, \quad i_{D_1} = 0, \quad V_{D_1} < V_{D_0}$$

$$4.3 < V_i < 13$$

$$V_o = 1k\Omega \times (-i_i) + V_i = V_i \rightarrow V_o = V_i$$

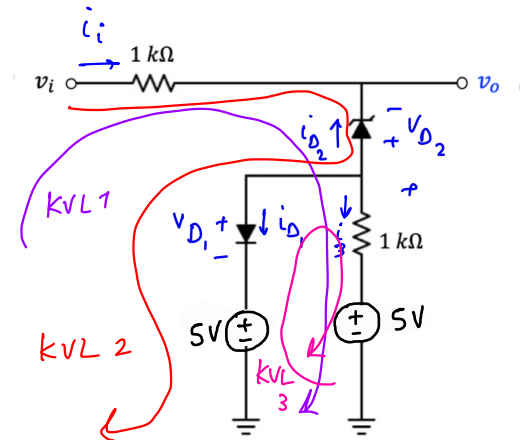
c)  $V_i(t) = \sin \omega t \rightarrow$  The amplitude of  $V_i$  will change between  $-1V$  and  $1V$ .  $D_1$  will be off and  $D_2$  will be ON.



$$V_o = \frac{1}{2} V_i + \frac{4.3}{2} V$$

$$V_i = 1 \rightarrow V_o = 0.5 + 2.15 = 2.65$$

$$V_i = -1 \rightarrow V_o = -0.5 + 2.15 = 1.65$$

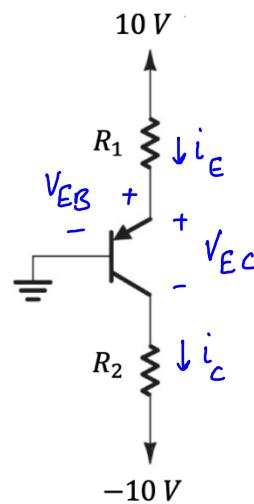




**Problem 4.**

Design the following circuit to have the BJT operate in the active mode.

Assume  $\beta = 100$ ,  $V_{sat} = 0.2 \text{ V}$ , and  $V_{D0} = 0.7 \text{ V}$



BJT ON and in active mode:

$$V_{EB} = V_{D0} = 0.7 \text{ V}$$

$$i_c = \beta i_b \quad \text{and} \quad V_{EC} \geq V_{D0}$$

$$i_E = i_c + i_b$$

$$i_E = \frac{1+\beta}{\beta} i_c$$

**Show your work.**

$$V_{EB} = 0.7 \text{ V} \quad \text{and} \quad V_B = 0 \rightarrow V_E = 0.7 \text{ V}$$

$$\text{Choose } V_{EC} = 4 \text{ V} \rightarrow V_E - V_C = 4 \text{ V} \rightarrow 0.7 - V_C = 4 \text{ V} \Rightarrow V_C = -3.3 \text{ V}$$

$$\text{Choose } R_2 = 1 \text{ k}\Omega \rightarrow i_c = \frac{V_C - (-10)}{R_2} = \frac{-3.3 + 10}{1 \text{ k}\Omega} = 6.7 \text{ mA}$$

$$i_E = \frac{1+100}{100} i_c = 1.01 \times 6.7 \text{ mA} \approx 6.77 \text{ mA}$$

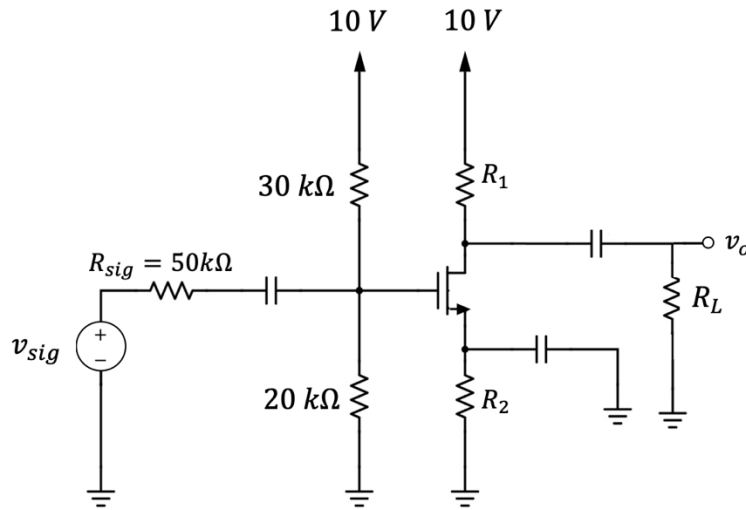
$$R_1 = \frac{10 - V_E}{i_E} = \frac{10 - 0.7}{6.77} \approx 1.37 \text{ k}\Omega$$

**Problem 5**

Design the following amplifier circuit for  $I_D = 2 \text{ mA}$ , and show the effect of the load resistor value on the total circuit voltage gain.

- Design the circuit and find the Bias point.
- Find the small signal parameters and draw the signal circuit.
- Analyze the circuit and show why and how the total circuit voltage gain changes when the load resistor value changes.
- Assume  $R_L = 10 \text{ k}\Omega$  and  $V_{sig}$  is a sine wave with the peak to peak amplitude of  $28 \text{ mV}$  and DC voltage of zero. Find and draw the instantaneous (total) drain to source voltage.

Assume  $V_t = 1 \text{ V}$ ,  $\mu C_{ox} \frac{W}{L} = 1 \text{ mA/V}^2$ , and  $\lambda = 0$ .



There is more than one correct answer to this problem.

**Show your work.**

$$V_G = \frac{2}{5} \times 10 = 4V$$

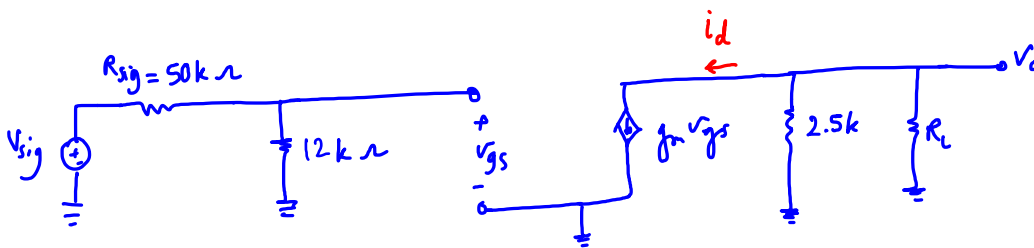
$$I_D = \frac{1}{2} \times 1 \times V_{ov}^2 = 2mA \rightarrow V_{ov} = 2V$$

$$V_{GS} - V_t = 2 \rightarrow V_{GS} = 3V \quad V_G - V_S = 3 \rightarrow V_S = 1V$$

$$R_2 = \frac{1-0}{2mA} = 0.5k\Omega$$

$$V_{ov} = 2V, \text{ choose } V_{DS} = 4V \rightarrow V_D = 5V \rightarrow R_1 = \frac{10-5}{2} = 2.5k\Omega$$

$$g_m = \frac{2I_D}{V_{ov}} = \frac{4mA}{2V} = 2mA/V, \quad r_o = \infty, \quad R_S = 20k\Omega \parallel 30k\Omega = 12k\Omega$$



$$\text{total circuit gain} = A = \frac{V_o}{V_{sig}} = \frac{V_o}{V_{gs}} \times \frac{V_{gs}}{V_{sig}}$$

$$V_o = -g_m(2.5k\Omega \parallel R_L) V_{gs} \rightarrow \frac{V_o}{V_{gs}} = -2mA/V \times (2.5k\Omega \parallel R_L)$$

$$V_{gs} = \frac{12k\Omega}{12k\Omega + 50k\Omega} \times V_{sig} \rightarrow \frac{V_{gs}}{V_{sig}} = \frac{6}{31}$$

$$A = \frac{-12}{31} (mA/V) \times (2.5k\Omega \parallel R_L) \Rightarrow \text{As } R_L \text{ decreases the total circuit gain will decrease}$$

$$i_d = g_m v_{gs} \quad , \quad v_{gs} = \frac{6}{31} v_{sig} \quad \rightarrow \quad \hat{v}_{gs} = \frac{6}{31} \hat{v}_{sig}$$

$$\hat{v}_{sig} = 14 \text{ mV} \quad \rightarrow \quad \hat{v}_{gs} = 2.71 \text{ mV}$$

$$\hat{i}_d = g_m \hat{v}_{gs} \quad \rightarrow \quad \hat{i}_d = \frac{2 \text{ mA}}{\text{V}} \times 2.71 \text{ mV} = 5.42 \text{ } \mu\text{A}$$

$$i_D = I_D + i_d = 2 \text{ mA} + 5.42 \text{ } \mu\text{A} \sin(\omega t)$$

$$\begin{aligned} \hat{v}_d = \hat{v}_o &= -(2.5 \text{ k}\Omega \parallel 10 \text{ k}\Omega) \times \hat{i}_d = -2 \text{ k}\Omega \times 5.42 \text{ } \mu\text{A} \sin(\omega t) \\ &= -10.84 \text{ (mV)} \sin(\omega t) \end{aligned}$$

$$V_D = V_D + v_d = 5 \text{ V} - 10.84 \text{ (mV)} \sin(\omega t)$$

$$v_s = 0 \text{ V} \quad , \quad V_S = 1 \text{ V} \quad \rightarrow \quad v_s = 1 \text{ V} + 0 \text{ V} = 1 \text{ V}$$

$$v_{Ds} = v_D - v_s = 5 \text{ V} - 10.84 \text{ (mV)} \sin(\omega t) - 1 \text{ V} = 4 \text{ V} - 10.84 \text{ (mV)} \sin(\omega t)$$

