

ECE 65: Components & Circuits Lab

Lecture 18

MOSFET Amplifier small signal model

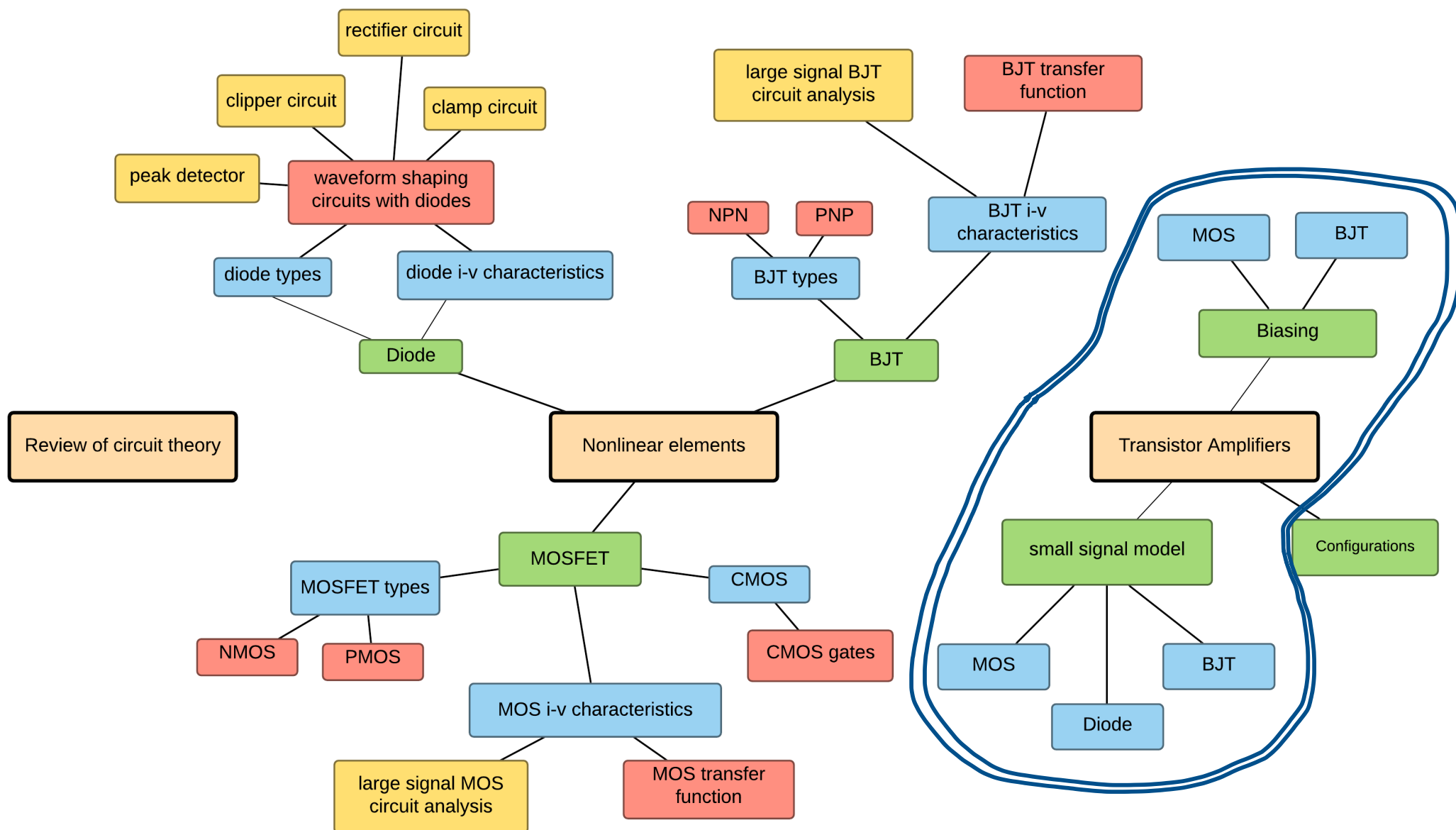
Reference notes: sections 5.1, 5.2

Sedra & Smith (7th Ed): sections 7.1

Saharnaz Baghdadchi

Course map

5. Transistor Amplifiers – Bias and small signal



Review of definitions

Signal: We want the response of the circuit to this input.

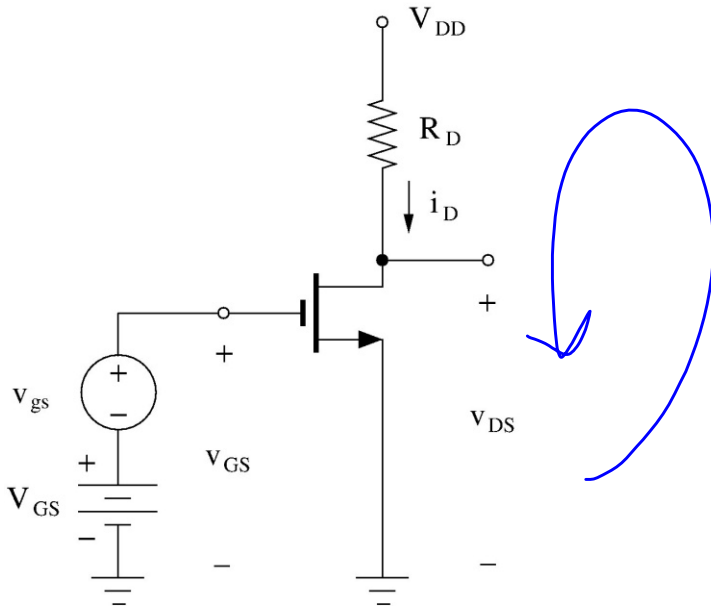
Bias: State of the system when there is no signal.

- Bias is constant in time (may vary extremely slowly compared to signal).
- Purpose of the bias is to ensure that MOS is in saturation at all times.

Derivation of MOS small signal model

$$\lambda = 0$$

The DC Bias point: $v_{gs} = 0$



$$I_D = \frac{1}{2} k_n (V_{GS} - V_t)^2 = \frac{1}{2} k_n V_{OV}^2$$

$$k_n = \mu_n C_{ox} \left(\frac{W}{L} \right)_n$$

$$V_{DS} = V_{DD} - R_D I_D$$

In the presence of the input signal, v_{gs} :

The total instantaneous gate-to-source voltage is: v_{GS} = V_{GS} + v_{gs}

$$\textcircled{i_D} = \frac{1}{2} k_n (V_{GS} + v_{gs} - V_t)^2$$

$$i_D = \frac{1}{2} k_n (V_{GS} + v_{gs} - V_t)^2$$

$$= \frac{1}{2} k_n (V_{GS} - V_t)^2 + k_n (V_{GS} - V_t) v_{gs} + \frac{1}{2} k_n (v_{gs})^2$$

Consider the case:

$$\frac{1}{2} k_n (v_{gs})^2 \ll k_n (V_{GS} - V_t) v_{gs}$$

or

$$\underline{v_{gs}} \ll \underline{2 (V_{GS} - V_t)}$$

In this case, v_{gs} is small enough and the **Small Signal Condition** is satisfied.
we can neglect the $\frac{1}{2} k_n (v_{gs})^2$ term in the i_D equation and

$$i_D \simeq \underline{I_D} + \underline{i_d}$$

$$\left\{ \begin{array}{l} i_D \simeq \frac{1}{2} k_n (V_{GS} - V_t)^2 + k_n (V_{GS} - V_t) v_{gs} \\ i_D \simeq I_D + i_d \end{array} \right.$$

$$I_D = \frac{1}{2} k_n (V_{GS} - V_t)^2$$

$$i_d = k_n (V_{GS} - V_t) v_{gs}$$

The MOSFET transconductance g_m is defined as

$$g_m \equiv \frac{i_d}{v_{gs}} = k_n (V_{GS} - V_t)$$

$$g_m = k_n V_{OV} = \frac{2 I_D}{V_{OV}}$$

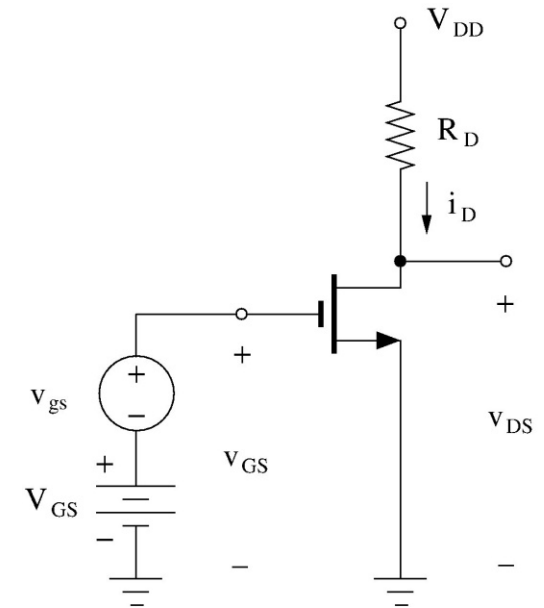
$$i_d = g_m v_{gs}$$

$$v_{DS} = V_{DD} - R_D \underline{i_D}$$

$$= V_{DD} - R_D \underline{(I_D + i_d)}$$

$$\left\{ \begin{array}{l} v_{DS} = V_{DD} - R_D I_D - R_D i_d \\ \underline{v_{DS} = V_{DS} + v_{ds}} \end{array} \right.$$

$$V_{DS} = V_{DD} - R_D I_D$$



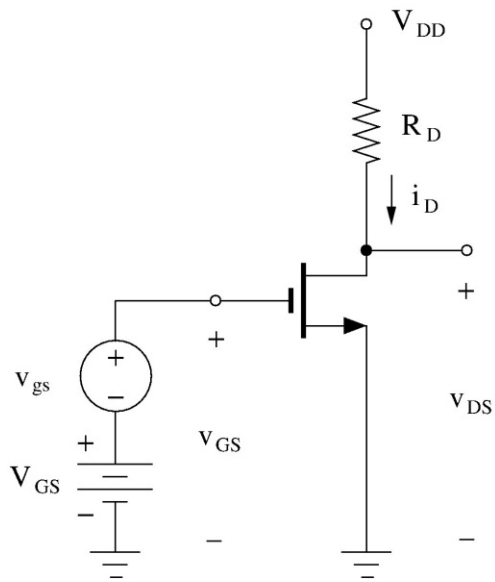
$$v_{ds} = \underline{-i_d R_D} = \underline{-g_m v_{gs} R_D}$$

Amplifier voltage gain is

$$A_v = \frac{v_{ds}}{v_{gs}} = -g_m R_D$$

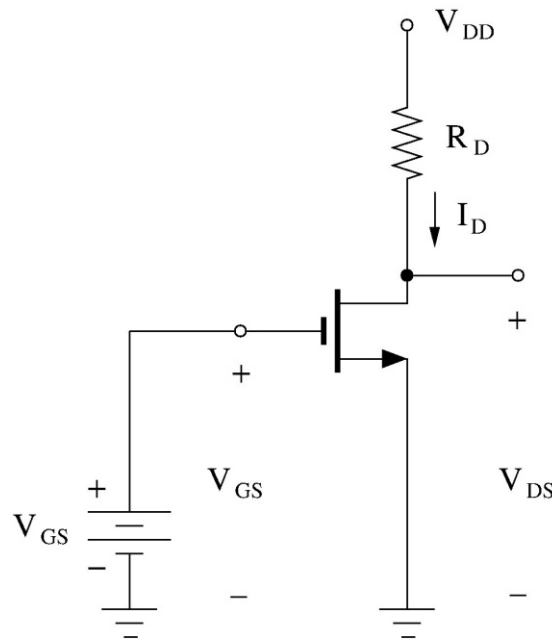
Bias and Signal circuits under small signal approximation

Bias & Signal



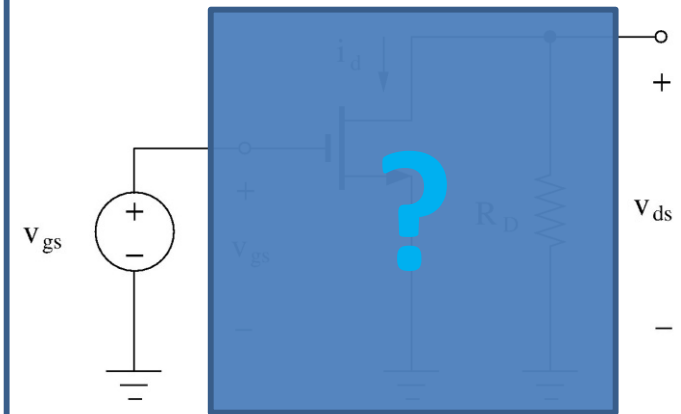
$$v_{GS}, i_D, v_{DS}, \dots$$

Bias



$$V_{GS}, I_D, V_{DS}, \dots$$

Signal only
= (Bias + Signal) - Bias

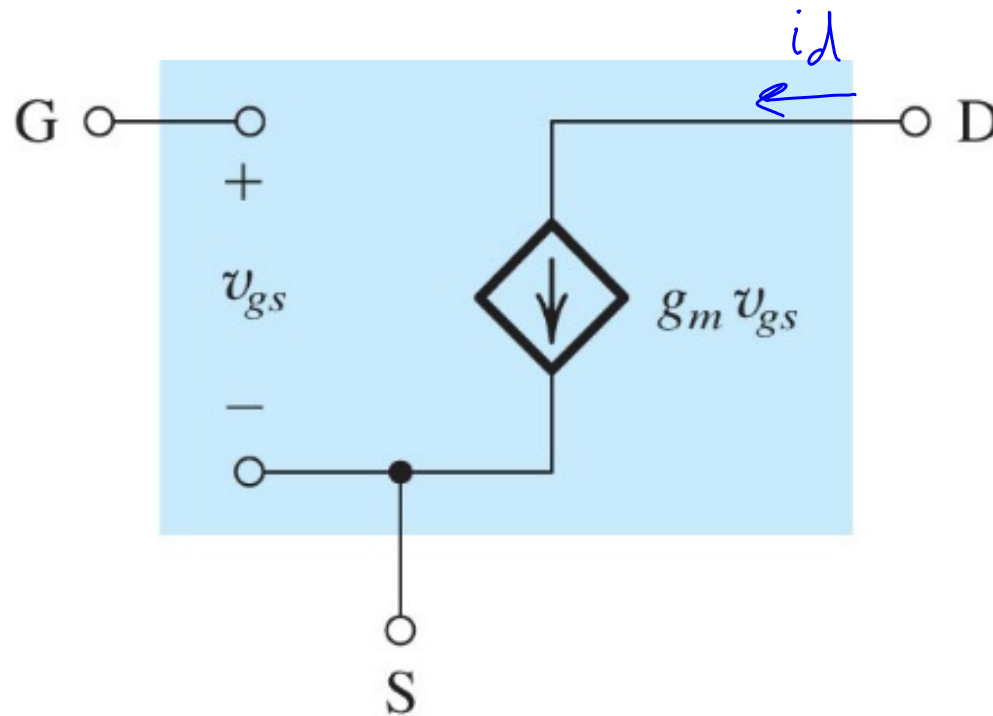


$$v_{gs}, i_d, v_{ds}, \dots$$

MOS small signal model

$$i_g = 0$$

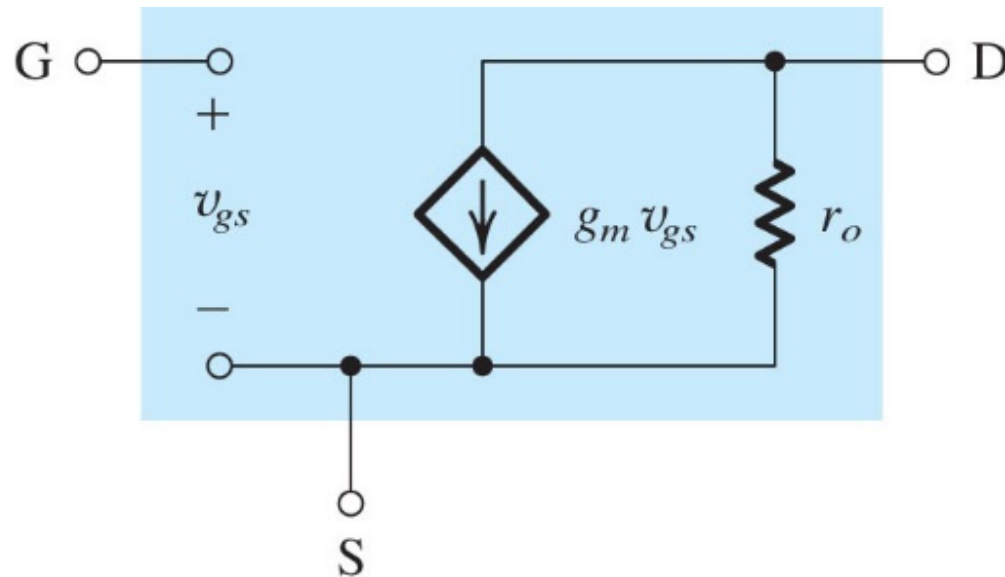
$$i_d = g_m v_{gs}$$



MOS small signal model

$$i_g = 0$$

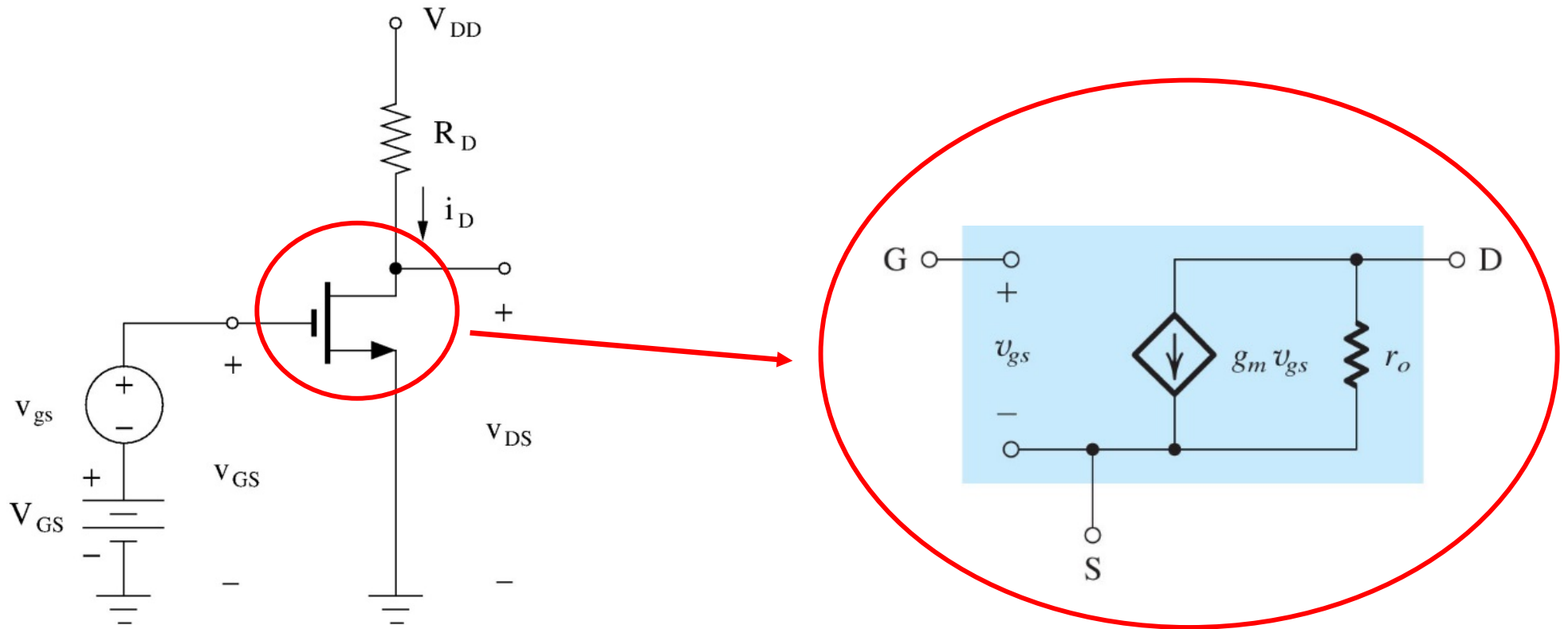
$$i_d = g_m v_{gs} + \frac{v_{ds}}{r_o}$$



$$g_m = \frac{2 I_D}{V_{OV}}$$

$$r_o \approx \frac{1}{\lambda \cdot I_D}$$

MOS Small Signal Model



What about the rest of the circuit element?

Other circuit elements in the small signal equivalent circuit

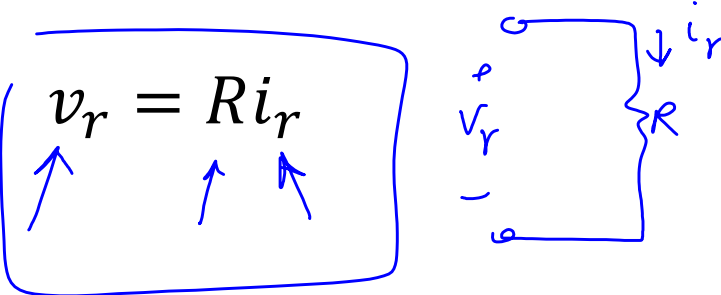
A resistor remains as a resistor in the signal circuit.

$$v_R = \underbrace{V_R + v_r}$$

Bias + Signal: $\underline{v_R} = R \underline{i_R}$

Bias: $V_R = \underline{R I_R}$

Signal: $v_r = \underline{v_R} - \underline{V_R} = \underline{R i_r} \Rightarrow$



The diagram illustrates the signal equivalent circuit for a resistor. It shows a box containing the equation $v_r = R i_r$, with three arrows pointing to v_r , R , and i_r respectively. To the right of the box is a circuit diagram of a resistor R connected between a signal source v_r and ground, with current i_r flowing into the resistor.

Other circuit elements in the small signal equivalent circuit

A capacitor remains as a capacitor in the signal circuit.

Unless we perform the frequency analysis of amplifier circuits, the capacitors will be short in the signal equivalent circuit.

Reminder:

A capacitor acts as an open circuit in the bias circuit.

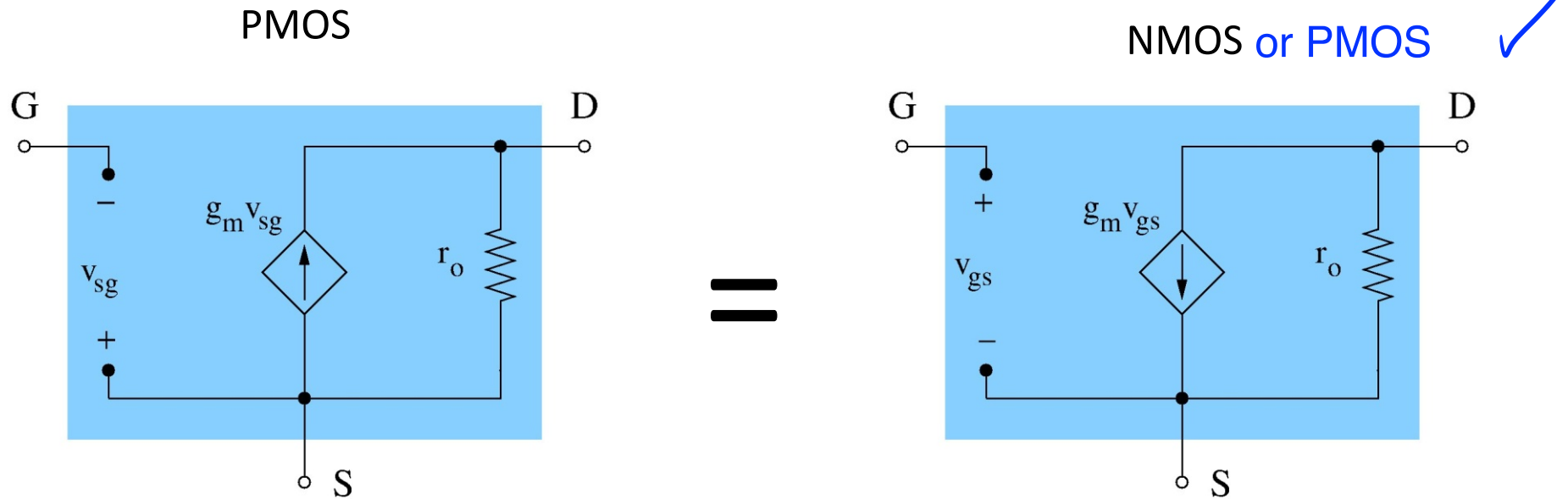
Other circuit elements in the small signal equivalent circuit

A DC voltage source becomes a short circuit.

For example, V_{DD} , V_{SS} will be effectively grounded.

A DC current source becomes an open circuit.

PMOS small signal model is identical to NMOS



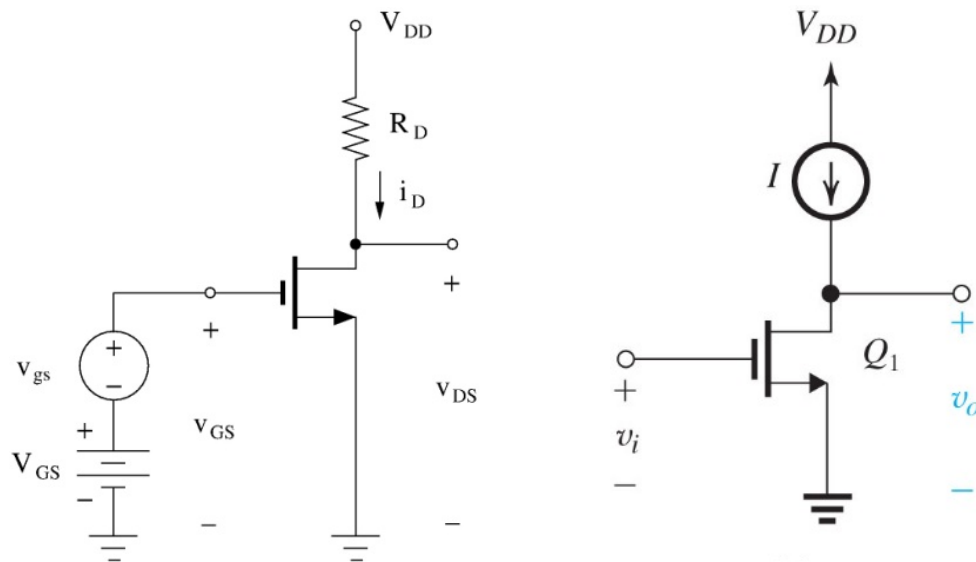
PMOS small-signal circuit model is identical to NMOS

We will use NMOS circuit model for both!

How to add signal to the bias

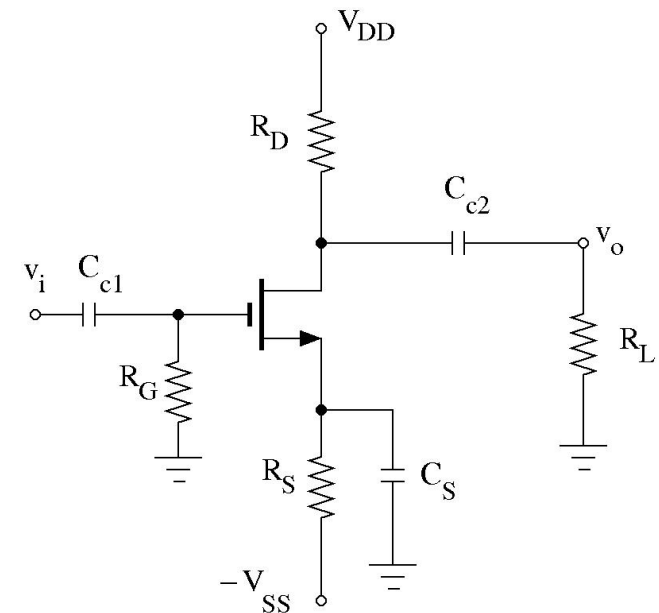
1. Direct Coupling

The signal is directly applied to the transistor.



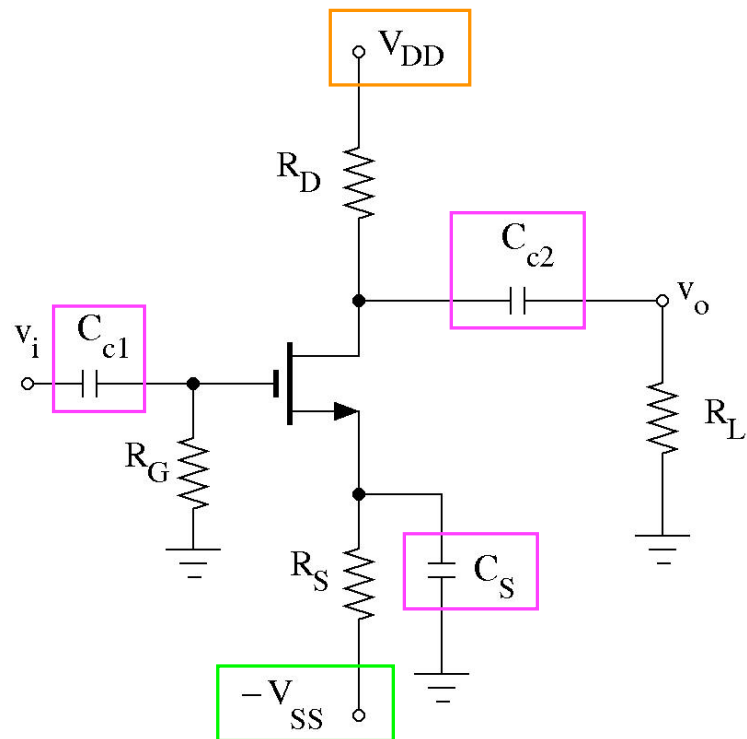
2. Capacitive Coupling

A capacitor is used to couple the signal to the transistor.

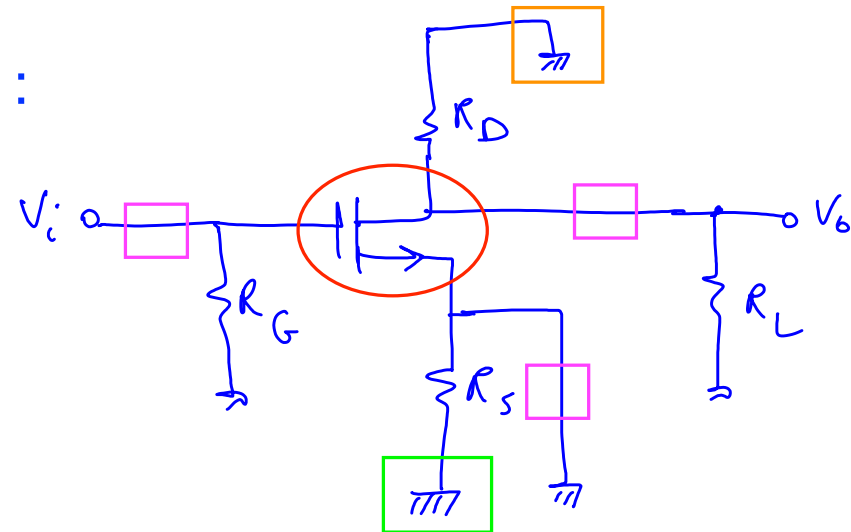


Example 1:

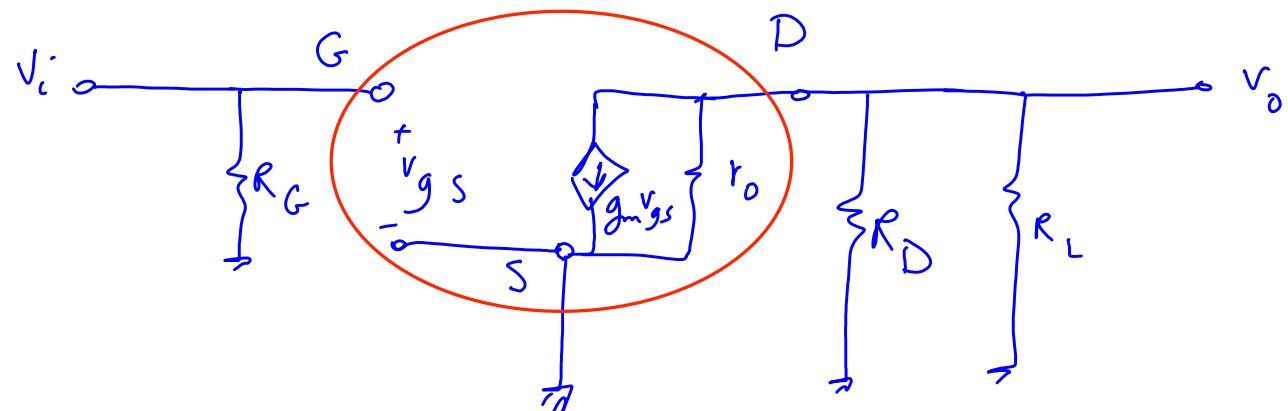
Draw the signal circuit (assume capacitors are short for signal).



step 1:

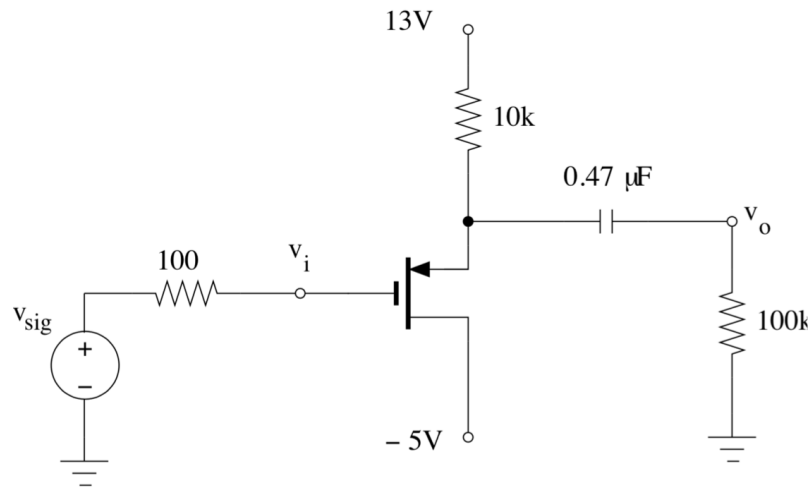


step 2:



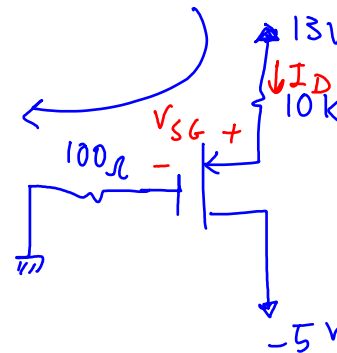
Lecture 18 reading quiz.

Find the transconductance, g_m , in this circuit ($V_{tn} = |V_{tp}| = 4\text{ V}$, $k_n = k_p = 0.4\text{ mA/V}^2$, $\lambda = 0.01\text{ V}^{-1}$. Assume $\lambda = 0$ for the bias circuit.



$$g_m = \frac{2 I_D}{V_{ov}} = \frac{2 I_D}{(V_{SG} - |V_{tp}|)}$$

Bias circuit to find I_D and V_{GS} .



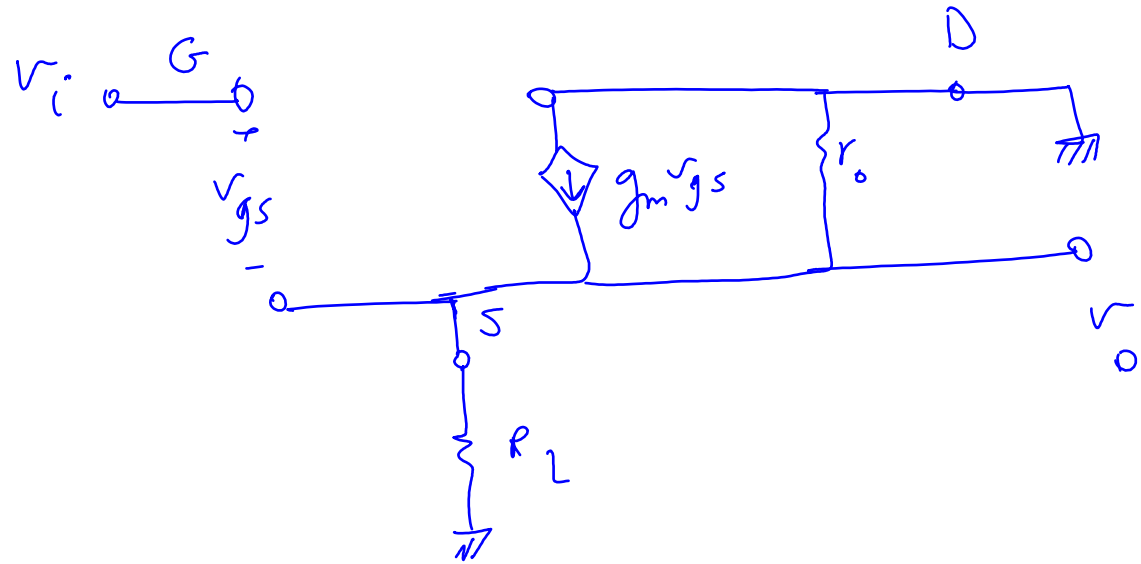
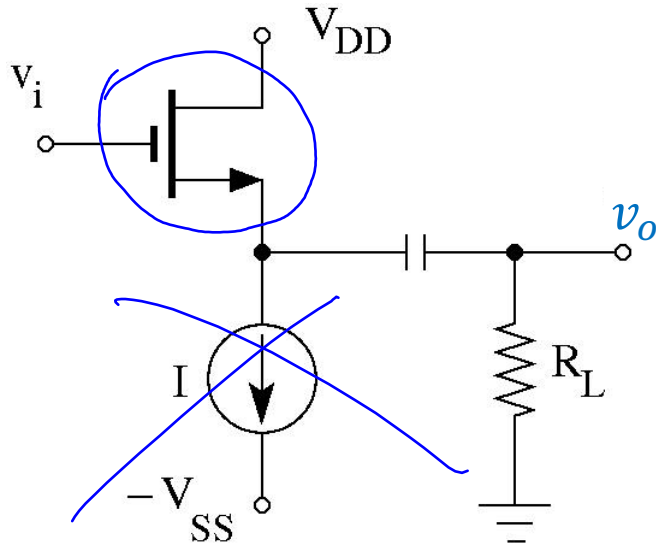
$$I_D = \frac{1}{2} \times k_p \times V_{ov}^2 \Rightarrow I_D = 0.2 (V_{SG} - 4)^2$$

$$\text{SG KVL: } 13 = 10\text{k} \times I_D + V_{SG} \rightarrow 13 = 10 \times 0.2 (V_{SG} - 4)^2 + V_{SG} \rightarrow V_{SG} = \begin{cases} 1.61 \times \\ 5.89 \sqrt{} \end{cases}$$

$$V_{SG} = 5.89\text{ V and } I_D = 0.71\text{ mA} \rightarrow g_m = \frac{2 \times 0.71\text{ mA}}{5.89 - 4} = 0.751\text{ mA/V}$$

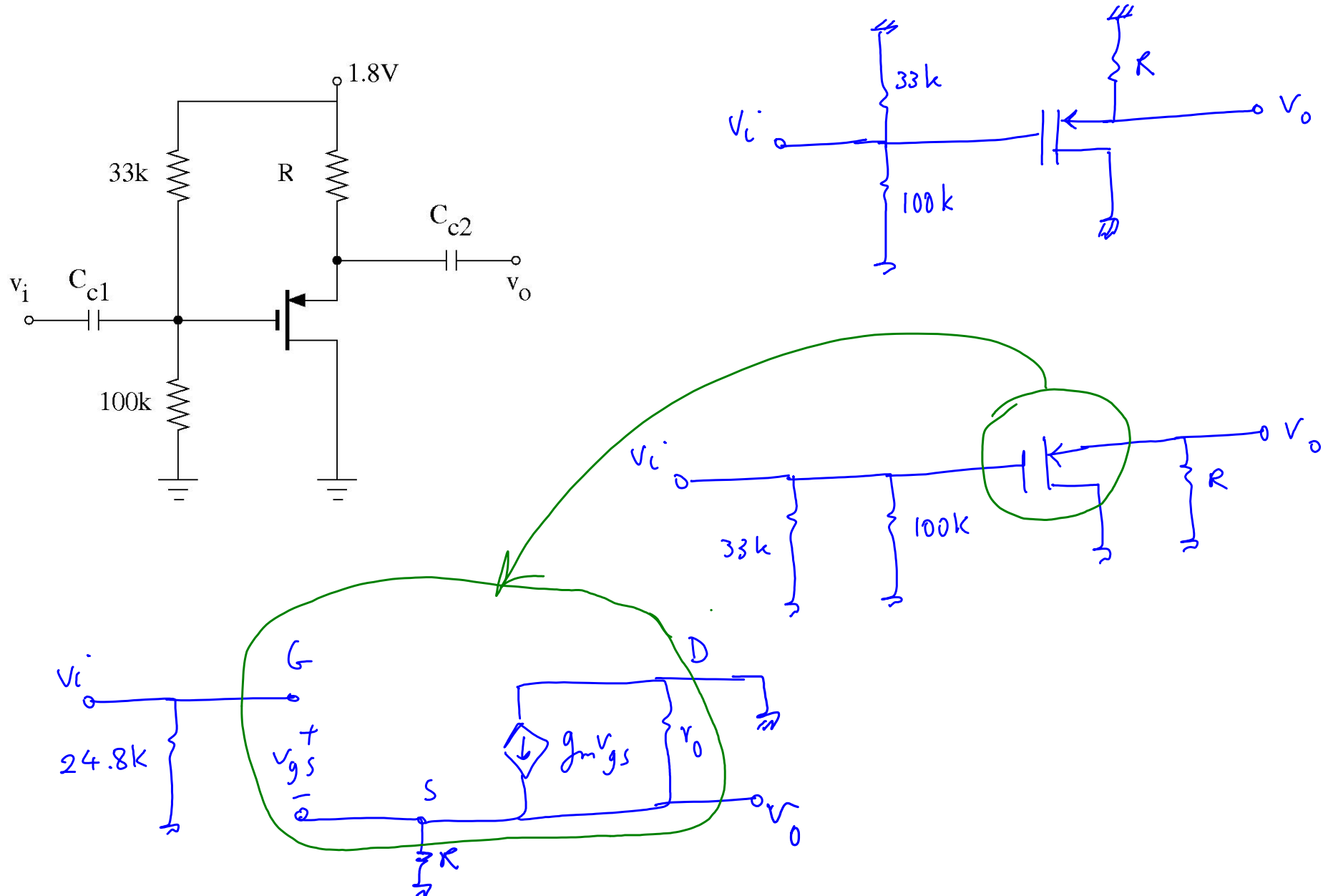
Discussion question 1.

Draw the signal circuit (assume capacitors are short for signal).



Discussion question 2.

Draw the signal circuit (assume capacitors are short for signal).

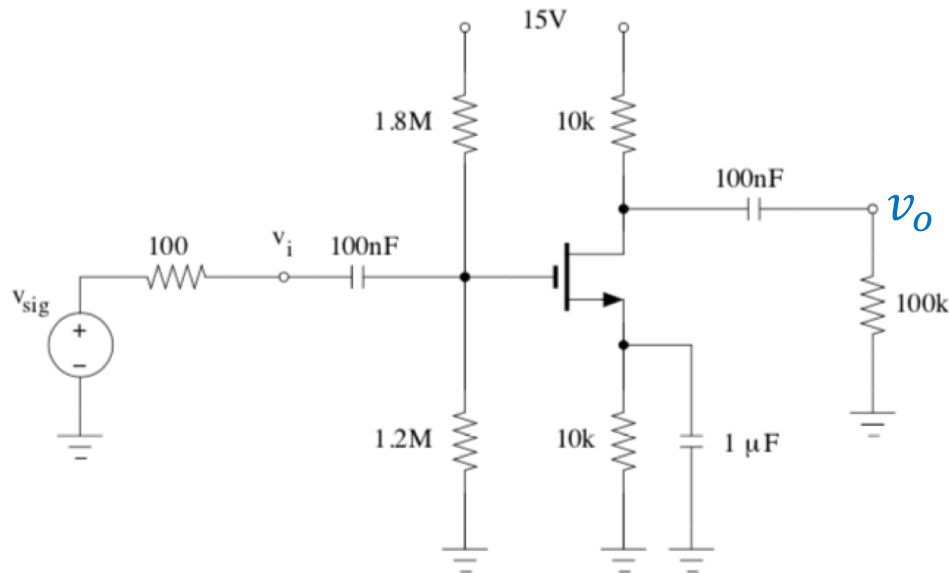


Discussion question 3.

Find the small signal parameters and draw the signal circuit (assume capacitors are short for signal).

$$V_{tn} = 1\text{ V}, k_n = 0.8\text{ mA/V}^2, \lambda = 0.01\text{ V}^{-1}.$$

Ignore the channel-width modulation effect in biasing calculations.

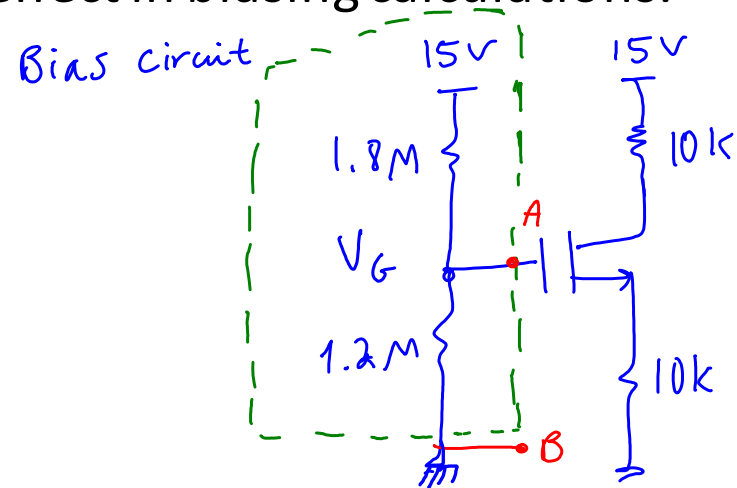
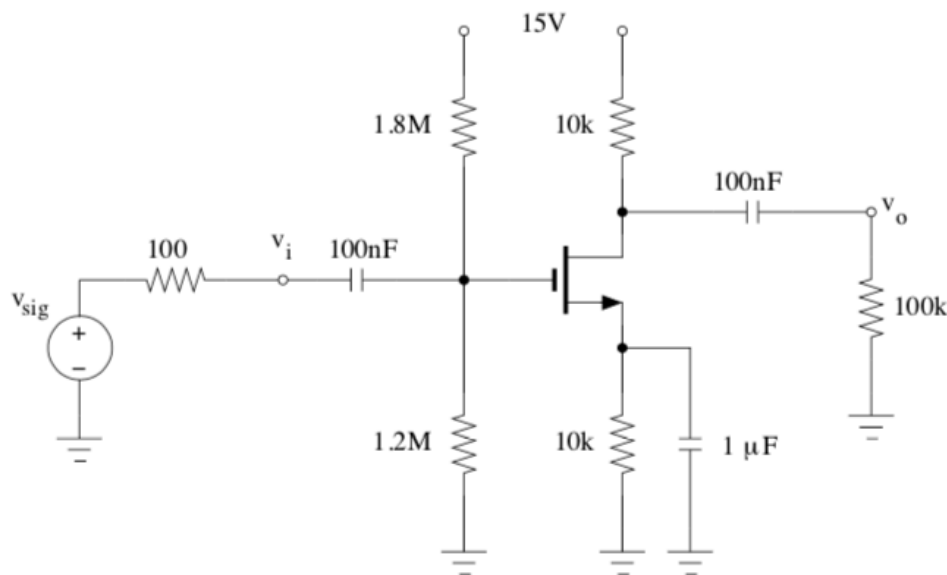


Discussion question 3.

Find the small signal parameters and draw the signal circuit (assume capacitors are short for signal).

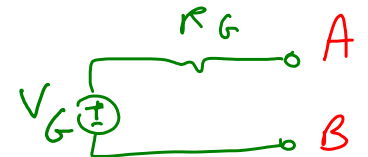
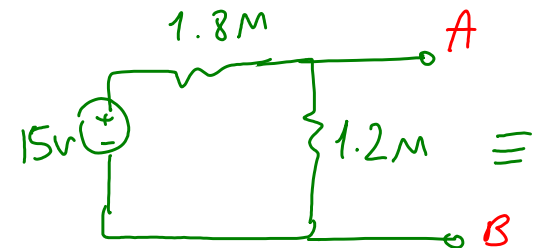
$$V_{tn} = 1\text{ V}, k_n = 0.8\text{ mA/V}^2, \lambda = 0.01\text{ V}^{-1}.$$

Ignore the channel-width modulation effect in biasing calculations.



$$V_G = \frac{1.2\text{ M}}{1.2\text{ M} + 1.8\text{ M}} \times 15\text{ V} = 6\text{ V}$$

$$R_G = 1.2\text{ M} \parallel 1.8\text{ M} = 720\text{ k}$$

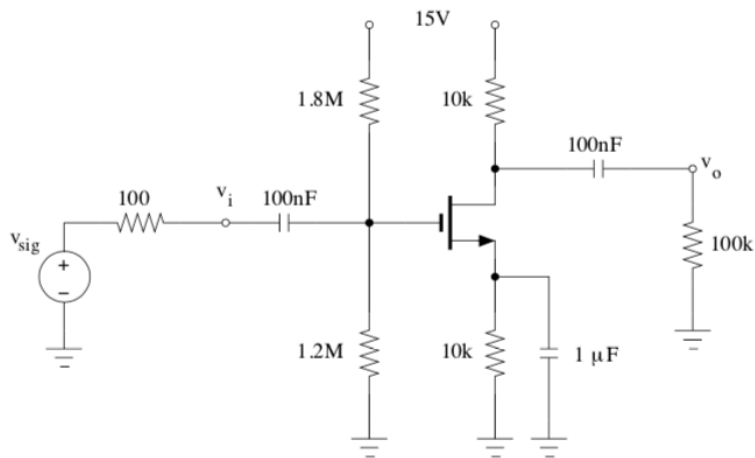


Discussion question 3.

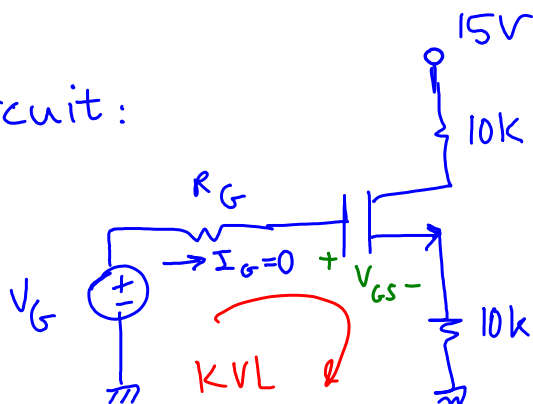
Find the small signal parameters and draw the signal circuit (assume capacitors are short for signal).

$$V_{tn} = 1\text{ V}, k_n = 0.8\text{ mA/V}^2, \lambda = 0.01\text{ V}^{-1}.$$

Ignore the channel-width modulation effect in biasing calculations.



Bias circuit:



$$I_D = \frac{1}{2} k_n V_{ov}^2$$

$$\text{KVL: } V_G = V_{GS} + 10k I_D$$

$$6 = V_{ov} + V_t + 10k I_D$$

$$\rightarrow 10k I_D + V_{ov} - 6 + 1 = 0$$

$$10 \times 10^3 \times \frac{1}{2} \times k_n V_{ov}^2 + V_{ov} - 5 = 0$$

$$\rightarrow 4 V_{ov}^2 + V_{ov} - 5 = 0$$

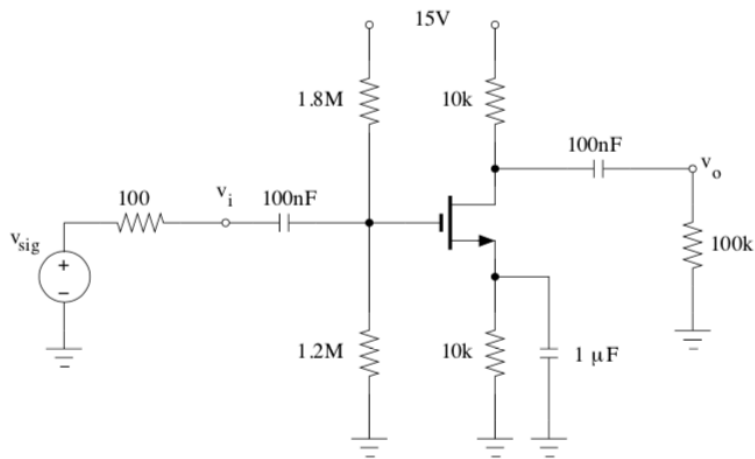
$$\rightarrow V_{ov} = 1\text{ V} \text{ and } V_{GS} = V_{ov} + V_t = 2\text{ V}$$

Discussion question 3.

Find the small signal parameters and draw the signal circuit (assume capacitors are short for signal).

$$V_{tn} = 1 \text{ V}, k_n = 0.8 \text{ mA/V}^2, \lambda = 0.01 \text{ V}^{-1}.$$

Ignore the channel-width modulation effect in biasing calculations.



$$I_D = 0.5 \times 0.8 \times 10^{-3} \times V_{ov}^2 = 0.4 \text{ mA}$$

$$\text{DS KVL: } 15 = 10 \text{ k } I_D + V_{DS} + 10^4 I_D$$

$$\rightarrow V_{DS} = 7 \text{ V} > V_{ov} \quad \text{MOS is in saturation}$$

Small signal parameters :

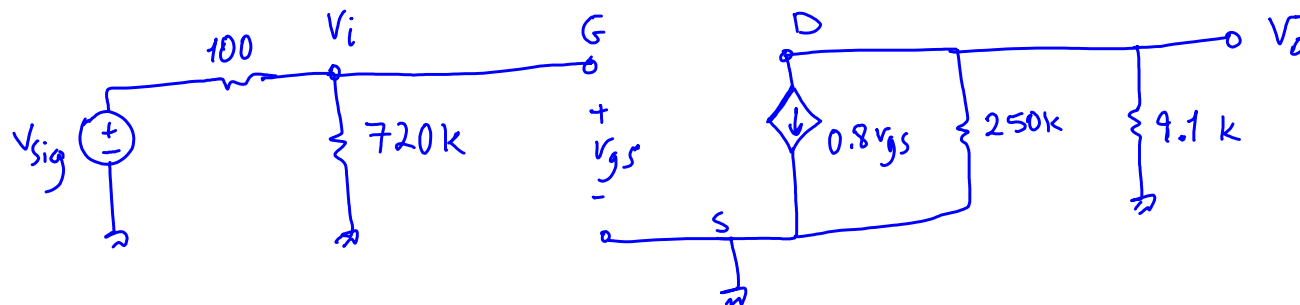
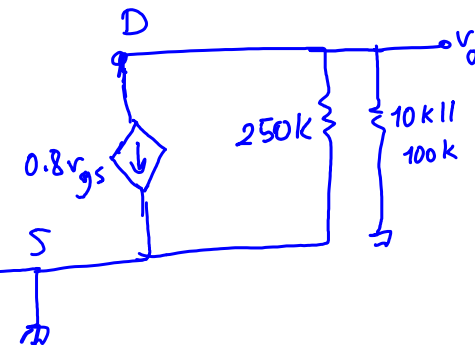
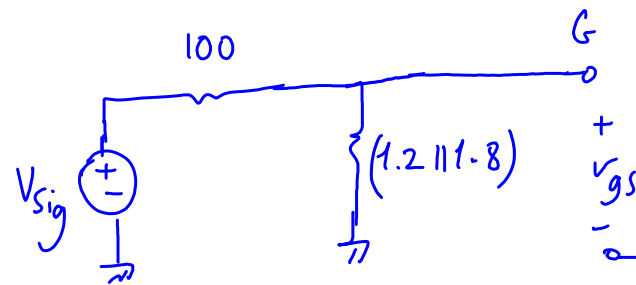
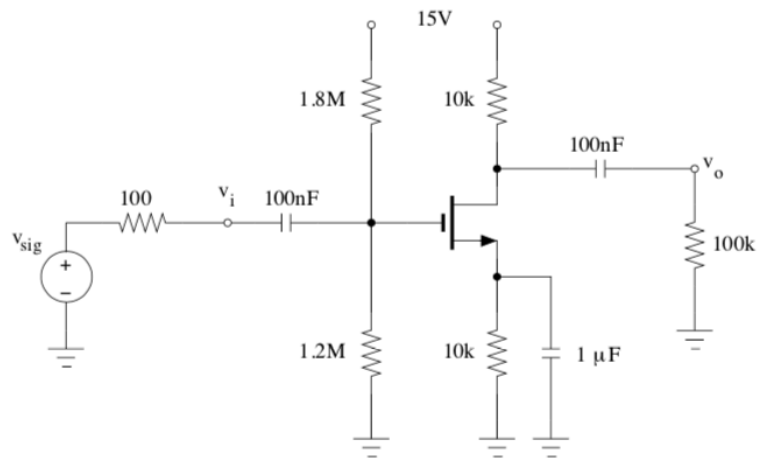
$$\begin{cases} g_m = \frac{2 I_D}{V_{ov}} = \frac{2 \times 0.4 \times 10^{-3}}{1} = 0.8 \text{ mA/V} \\ r_o = \frac{1}{\lambda I_D} = \frac{1}{0.01 \times 0.4 \times 10^{-3}} = 250 \text{ k} \end{cases}$$

Discussion question 3.

Find the small signal parameters and draw the signal circuit (assume capacitors are short for signal).

$$V_{tn} = 1\text{ V}, k_n = 0.8\text{ mA/V}^2, \lambda = 0.01\text{ V}^{-1}.$$

Ignore the channel-width modulation effect in biasing calculations.



v_{gs} is the same as v_i