

***pn* Junction**

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2. Open-Circuit Conditions
3. Reverse-Bias Conditions
4. Breakdown Region
5. Forward-Bias Conditions
6. Terminal Current-Voltage Characteristics
7. Depletion Capacitance and Diffusion Capacitance
8. Modeling the Diode
9. The *pn* Junction Circuit: Rectifier

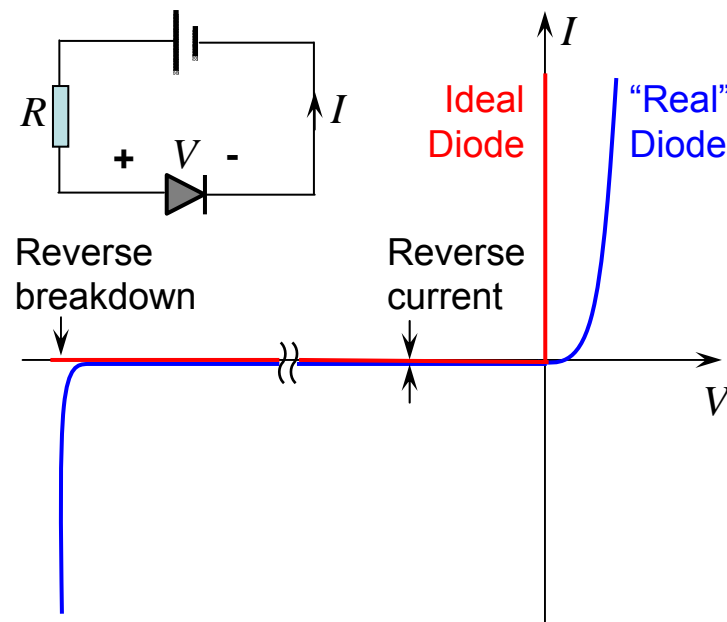
Reference

- A.D. Sedra & K.C. Smith, “Microelectronic Circuits – Theory and Application”, 5th Edition (International Version), Oxford University Press, Chapter 2.

pn Junction - Introduction

Why study *pn* junctions?

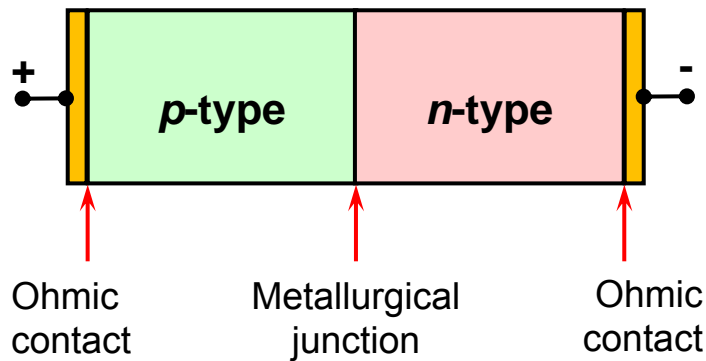
- The semiconductor diode is basically a *pn* junction.
- Diode is the simplest and most fundamental *nonlinear* circuit element. It allows a current to flow through it easily in one direction, but not in the opposite direction (unlike a resistor).
 - As such, it can be used in a rectifier circuit, to convert ac into dc.



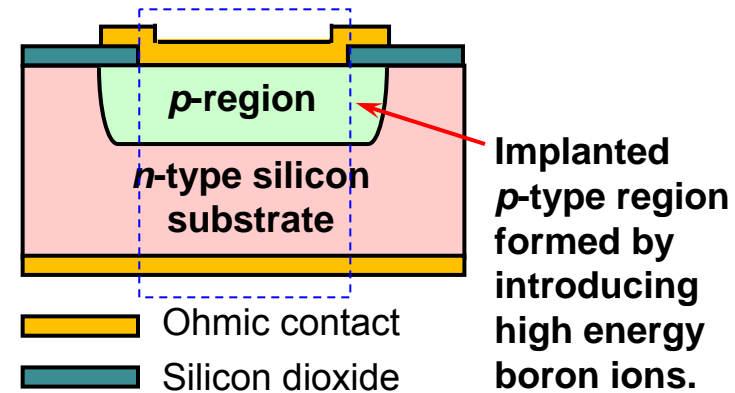
- The *pn* junction is the basic element of bipolar junction transistors (BJTs) and field-effect transistors (FETs), which are widely used in electronic circuits.

pn Junction - Introduction & Definitions

Junction – the joining together of two dissimilar materials, e.g., metal-semiconductor junction, *pn* junction (semiconductor diode).



Simplified physical structure of a pn junction



Realistic cross-section of fabricated silicon *pn* junction diode

pn junction - a *p*-type semiconductor region in close contact with an *n*-type semiconductor region on the same single-crystal semiconductor material (e.g., Si).

Metallurgical junction - the interface between the *p*- and *n*-doped regions.

Ohmic contact – a metal-semiconductor junction which allows electrical current to pass freely through the semiconductor.

In this course, we will deal with a simplified structure for analysis purposes.

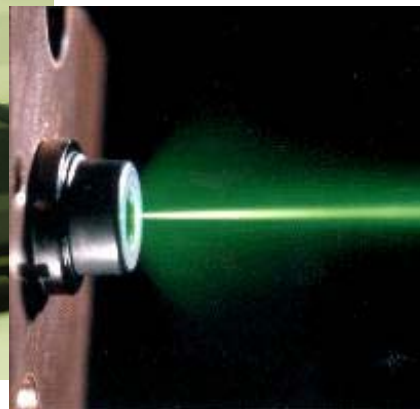
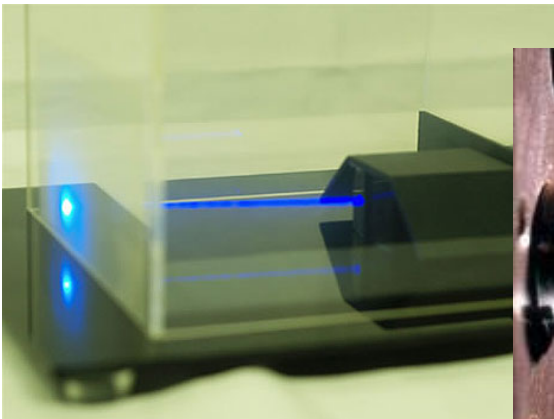
***pn* Junction - Introduction**

Some special semiconductor diodes -

Light Emitting Diodes (LEDs)



Laser Diodes



Solar Cells



***pn* Junction**

***pn* Junction**

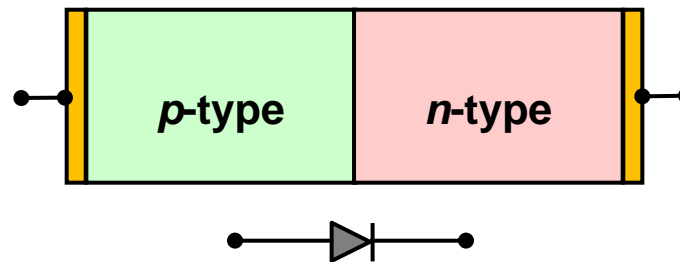
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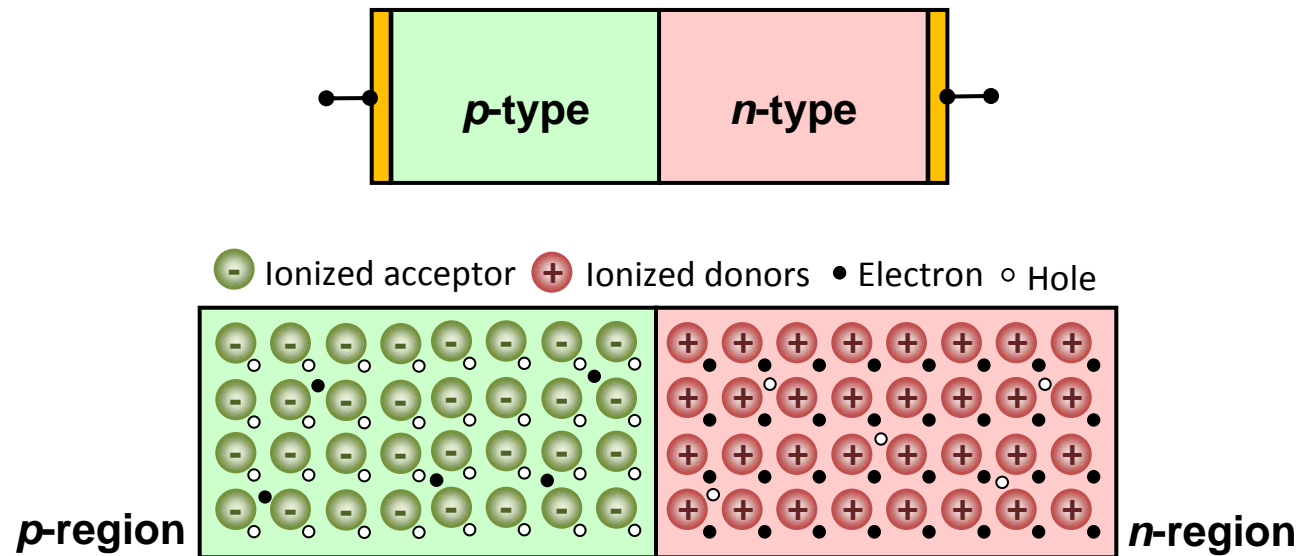
pn Junction – Open-Circuit Conditions

- We begin with a *pn* junction under **open-circuit conditions**, i.e., external terminals are left open or not connected to a voltage source
 - Hence, there is **no net current flow** through it



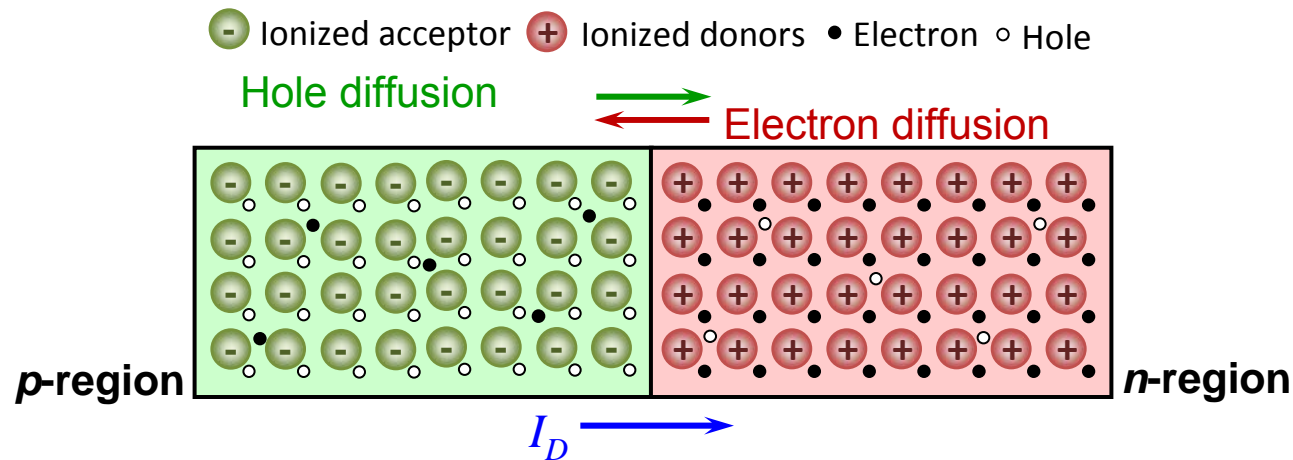
- Assumptions:
 - Room temperature operation ($T = 300 \text{ K}$ or $27 \text{ }^{\circ}\text{C}$)
 - Steady state conditions
 - *p*-type region is doped with acceptor concentration of N_A ($\gg n_i$)
 - *n*-type region is doped with donor concentration of N_D ($\gg n_i$)
- Consider what happens when a *p*-type semiconductor first makes contact with an *n*-type material. This is a thought process as a *pn* junction is not formed this way. However, this approach is useful to help us understand some of the important principles in *pn* junctions.

pn Junction – Open-Circuit Conditions



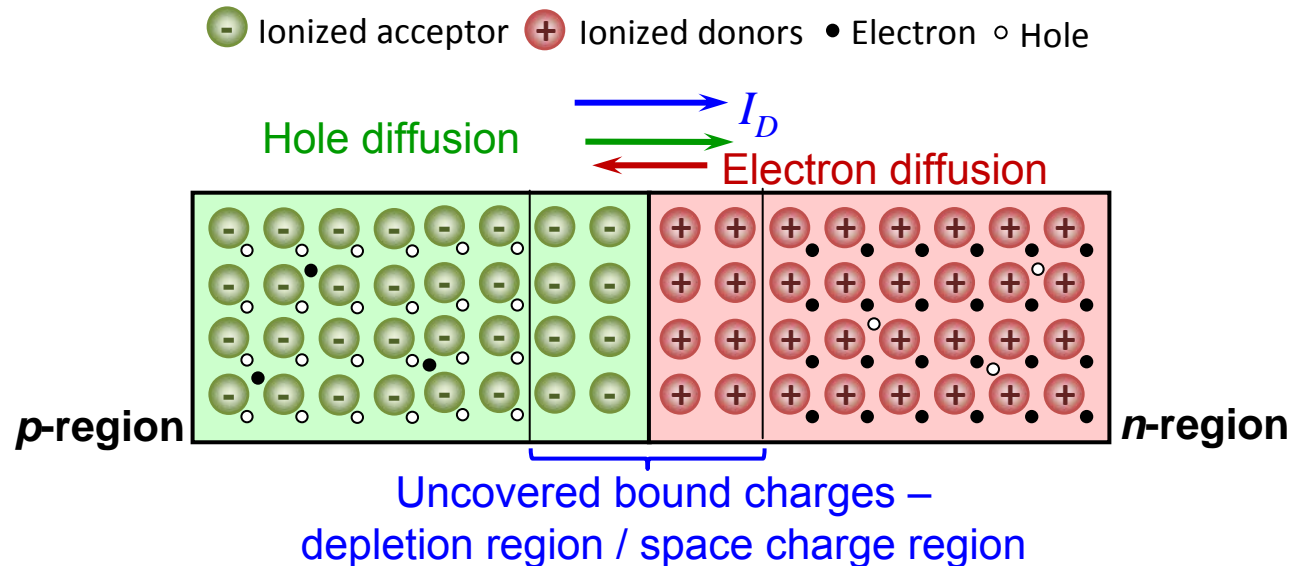
- The *p*-region has
 - N_A negative bound ionized acceptors
 - $p_{p0} \approx N_A$ positive holes (many - majority carriers)
 - $n_{p0} \approx n_i^2/N_A$ negative electrons (few - minority carriers)
- Hence, electrically neutral
- The *n*-region has
 - N_D positive bound ionized donors
 - $n_{n0} \approx N_D$ negative electrons (many - majority carriers)
 - $p_{n0} \approx n_i^2/N_D$ positive holes (few - minority carriers)
- Hence, electrically neutral
- Clearly, there are equivalent phenomena for electrons and holes in *pn* junction. Lectures will only focus on one carrier. Explanation for both carriers are included in the appendix for your reference.

pn Junction – Open-Circuit Conditions



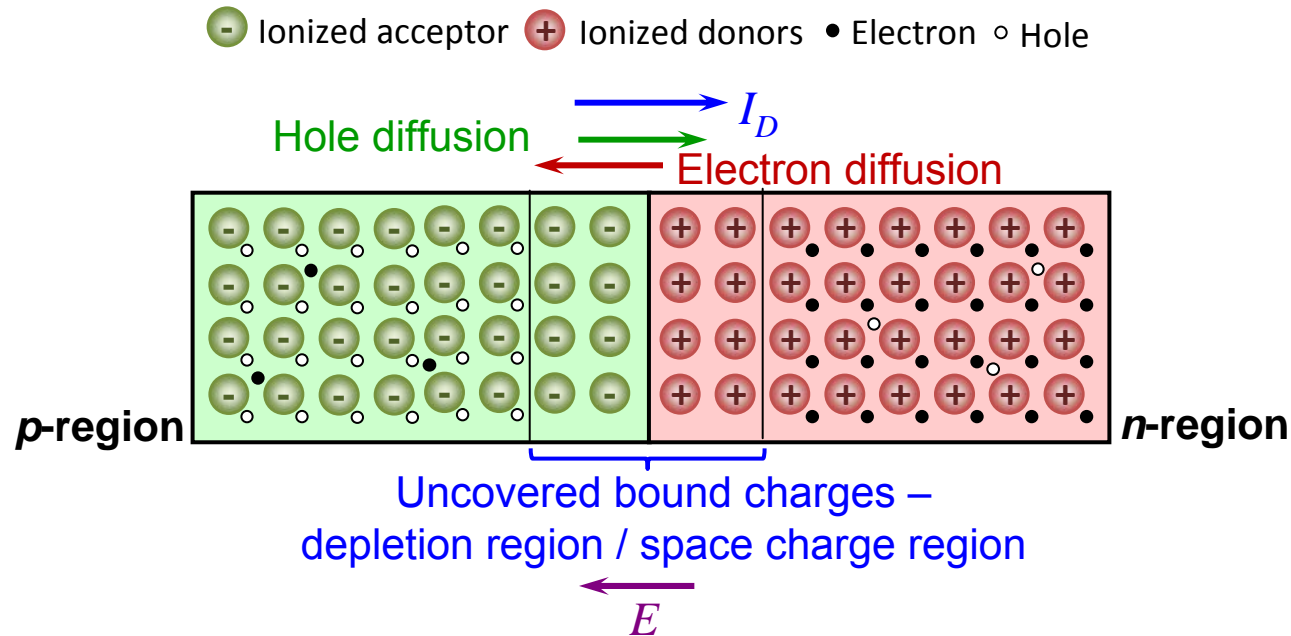
- The n -region has many more electrons ($n_{n0} \approx N_D$) than the p -region ($n_{p0} \approx n_i^2/N_A$), meaning electron concentration gradient exists across the metallurgical junction.
- Hence, electrons diffuse from the n -region to the p -region
- Similarly, holes diffuse from the p -region to the n -region
- The hole and electron diffusions add to form the diffusion current I_D (from the p -region to n -region). Note that I_D is due to majority carriers.

pn Junction – Open-Circuit Conditions



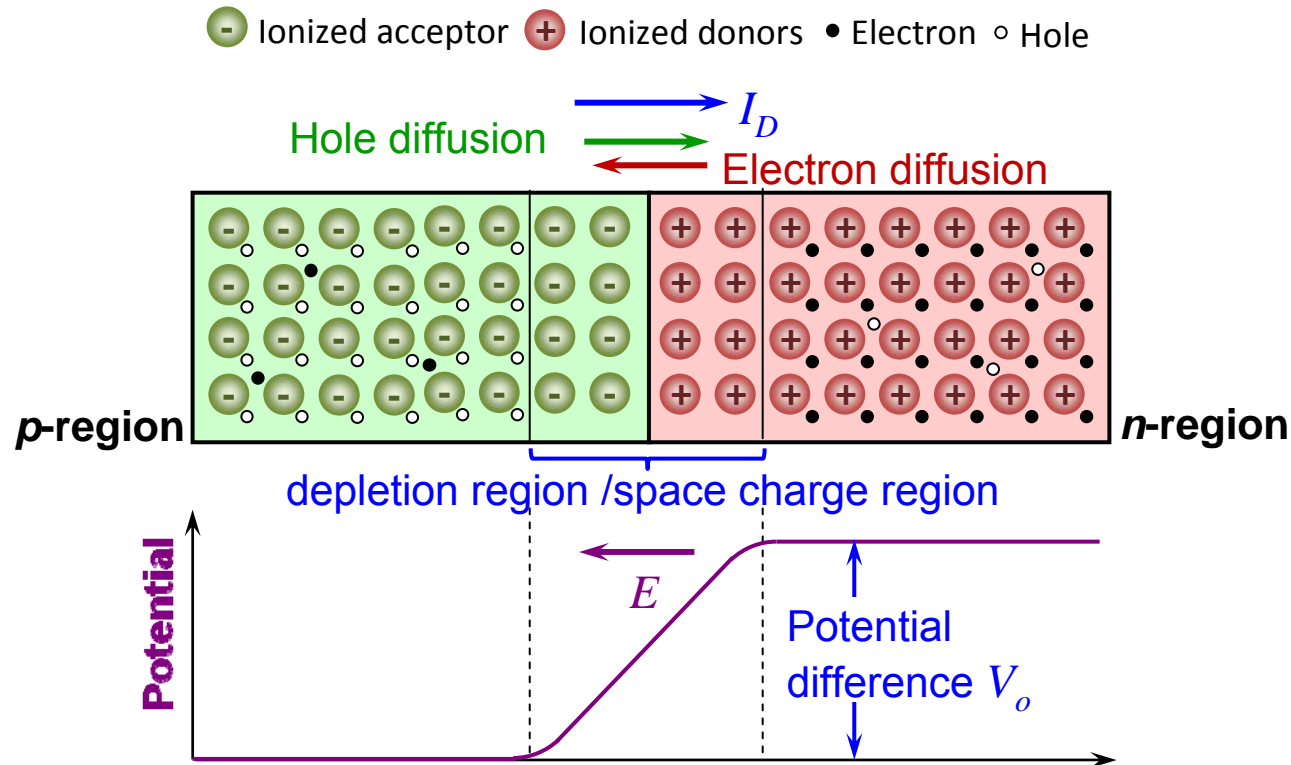
- Holes (from the *p*-region) that diffuse across the (metallurgical) junction into the *n*-region recombine with some of the majority electrons there
- Hence, some electrons in the *n*-region near the (metallurgical) junction disappear, resulting in the bound positive ionized donors being no longer neutralized by electrons (i.e., uncovered)
- Considering both electrons and holes, the region near the junction comprises uncovered bound positive ionized donors (in the *n*-region) and negative ionized acceptors (in the *p*-region) and is depleted of electrons and holes.

pn Junction – Open-Circuit Conditions



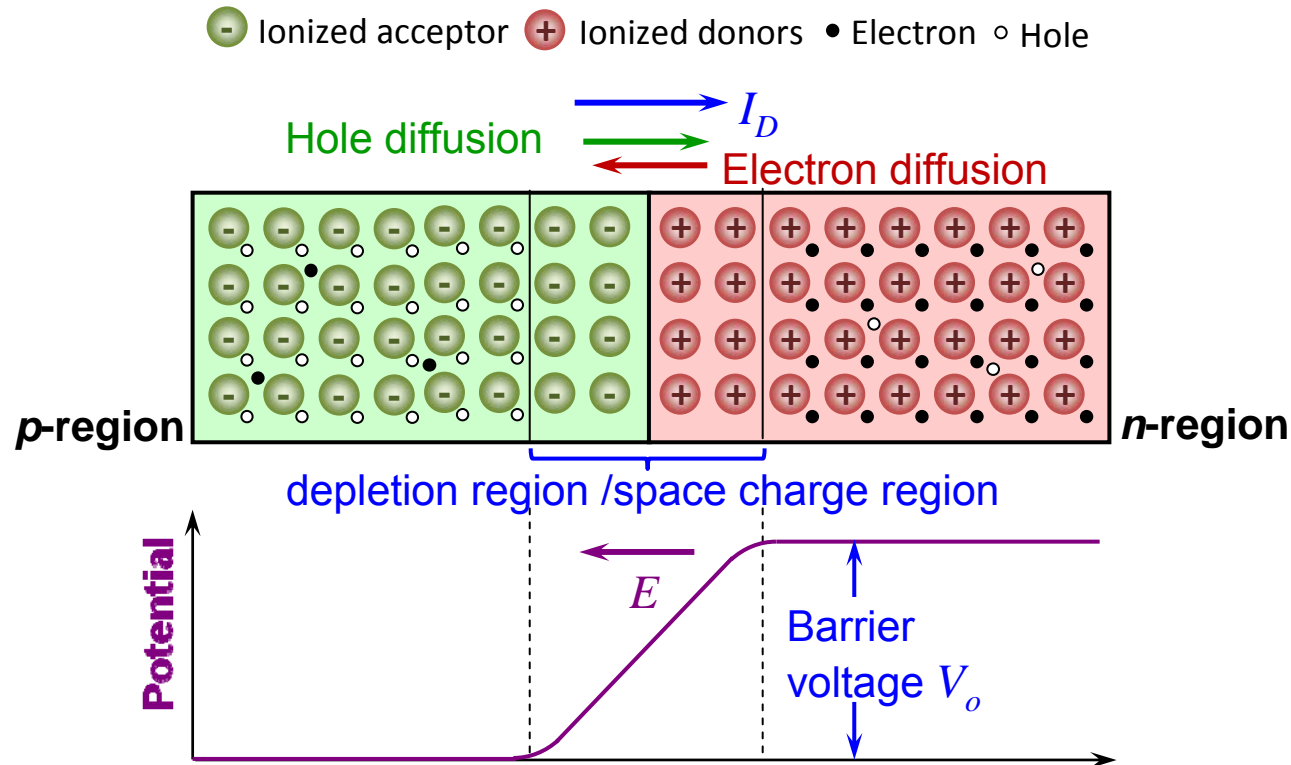
- The region near the junction is depleted of electrons and holes, hence is known as the **depletion region**. It is also called the **space charge region** as it has fixed donor/acceptor charges.
- The depletion region has 2 sides with different types of charges (negative in the *p*-side and positive in the *n*-side) and they cause an **electric field, E** , to establish across the depletion region, in the direction from the *n*-side to the *p*-side.

pn Junction – Open-Circuit Conditions



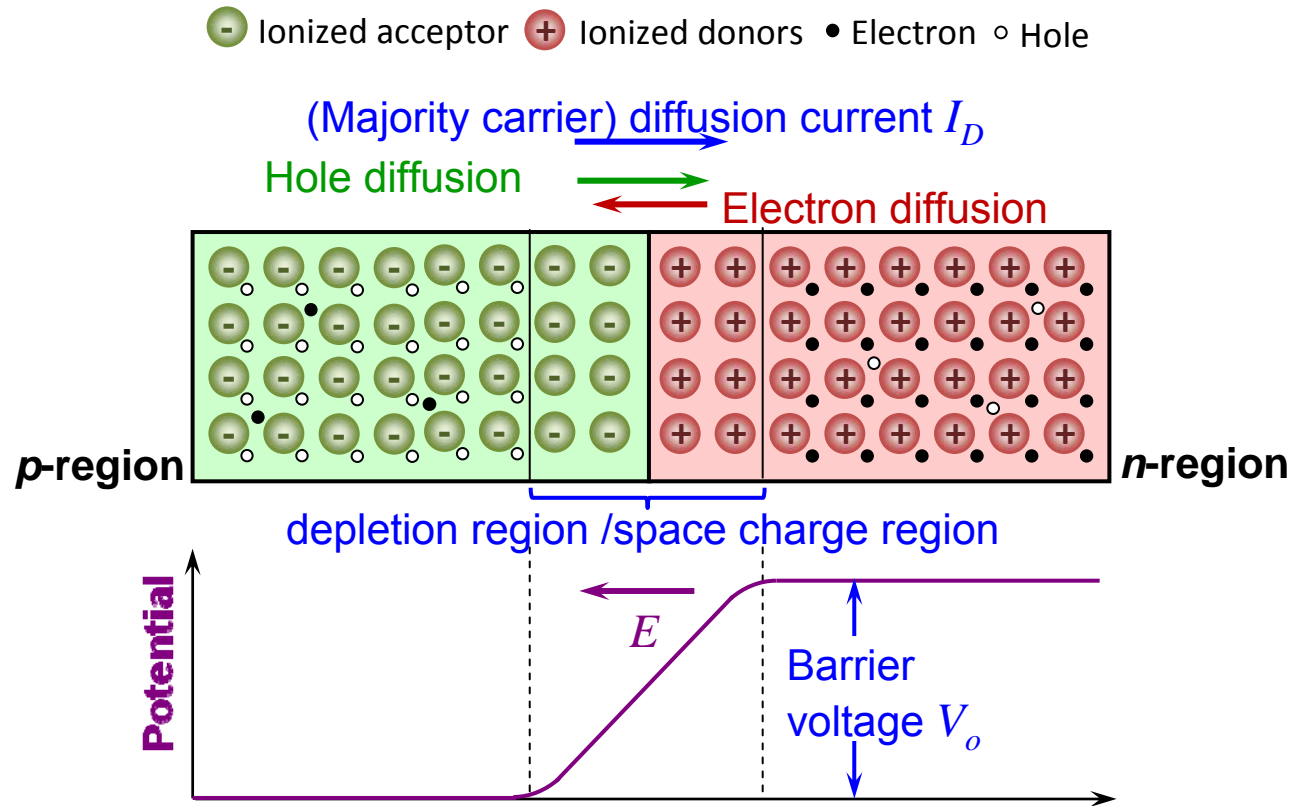
- With an established electric field, E , across the depletion region in the direction from the n -side to the p -side, a potential difference (voltage), V_o , results across the depletion region, with the n -side at a higher potential than the p -side.

pn Junction – Open-Circuit Conditions



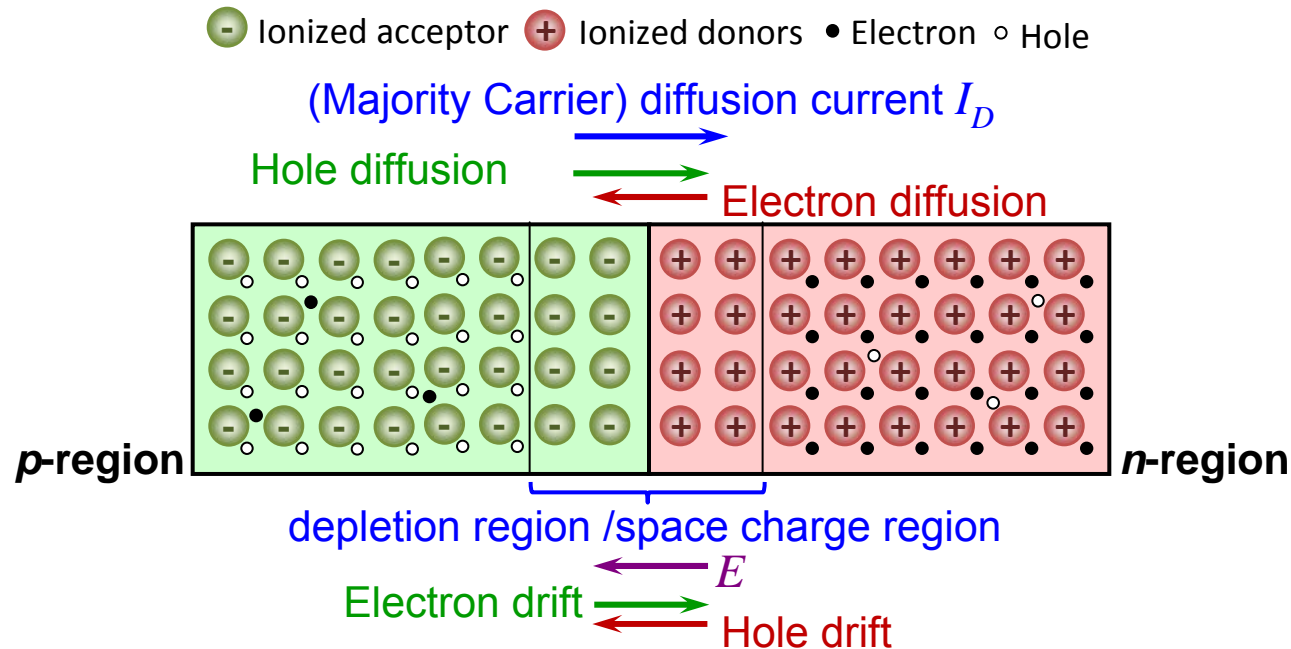
- The established **electric field**, E , across the depletion region **opposes the diffusion** of holes into the n -region, and electrons into the p -region.
- The potential difference (voltage) across the depletion region (V_o) acts a barrier that must be **overcome** by holes diffusing into the n -region and by electrons diffusing into the p -region. Hence, V_o is known as the **barrier voltage**.

pn Junction – Open-Circuit Conditions



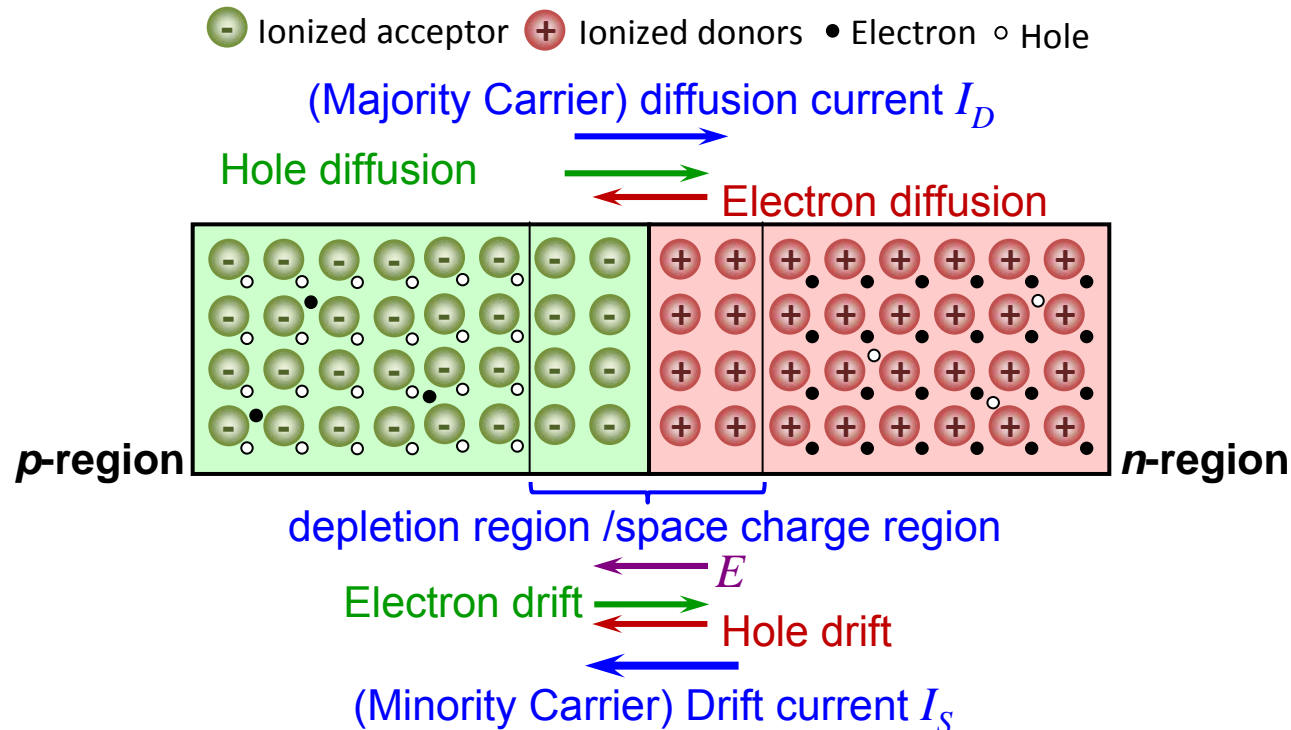
- Larger barrier voltage V_o across the depletion region, corresponding to a higher electric field E , opposes (majority carrier) diffusion, hence a smaller diffusion current I_D resulted. This means a smaller number of carriers are able to overcome the barrier.
- (Majority carrier) diffusion current I_D depends strongly on barrier voltage V_o .

pn Junction – Open-Circuit Conditions



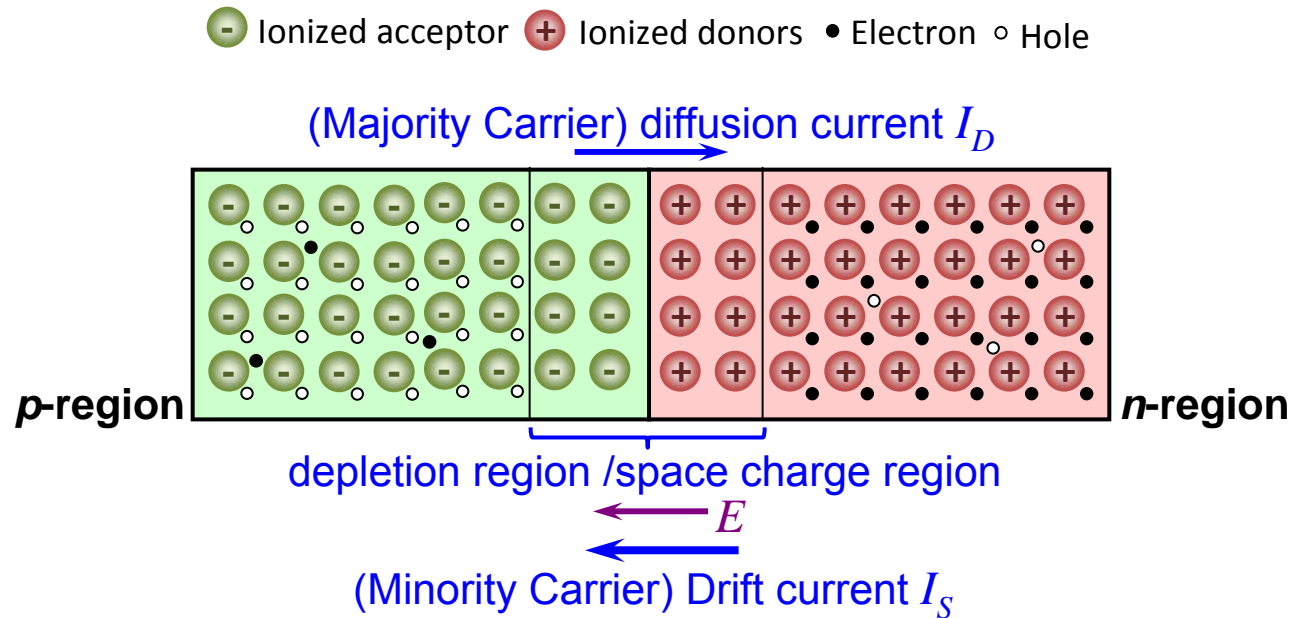
- As an **electric field** E is present across the depletion region, **drift** of carriers can happen.
- Some thermally generated minority holes in the n -region diffuse through the n -region to the edge of the depletion region, where holes experience the **electric field** E and are swept across the depletion region into the p -region (**drift of holes**).
- Similarly, **drift of minority electrons** from the p -region into n -region also occurs, as shown by the **green** arrow.

pn Junction – Open-Circuit Conditions



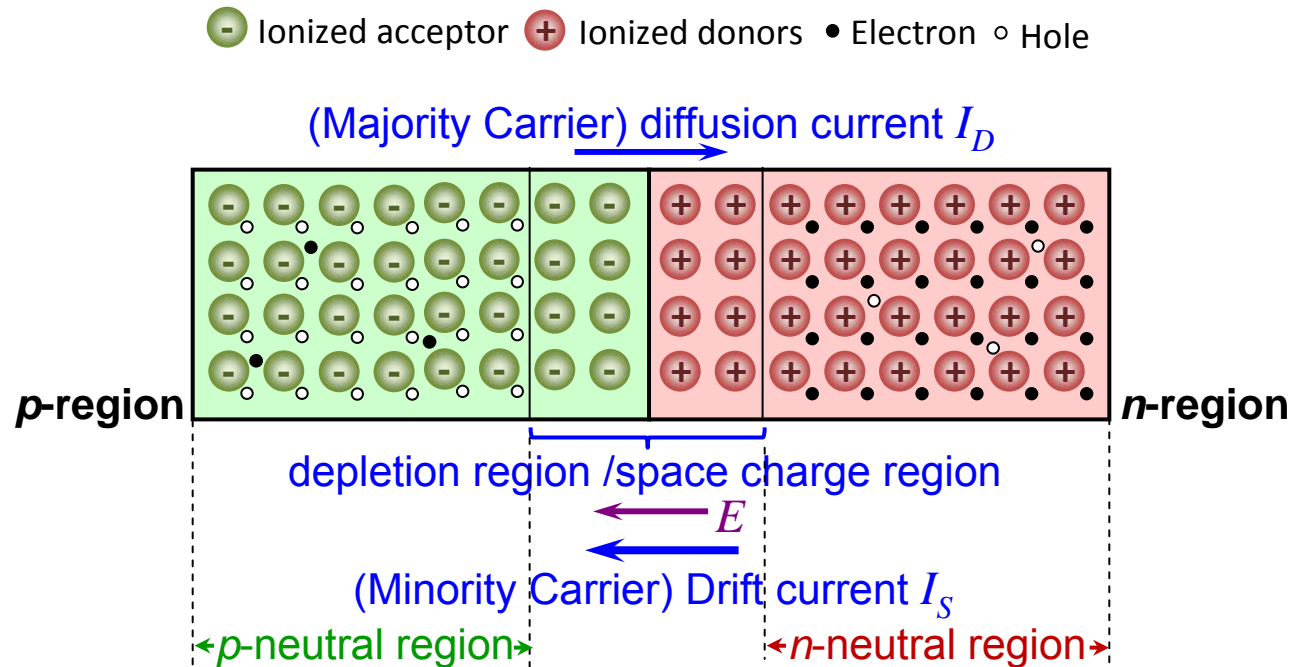
- Electrons moved by drift from p -region to n -region and holes moved by drift from n -region to p -region add to form the drift current I_S (from the n -region to p -region).
- The drift current I_S is carried by thermally generated minority carriers that are swept across the depletion region by the electric field E . Hence, its value is strongly dependent on temperature
 - higher temperature, more thermally generated minority carriers, higher I_S .
- The drift current I_S is independent on the value of the barrier voltage V_o .

pn Junction – Open-Circuit Conditions



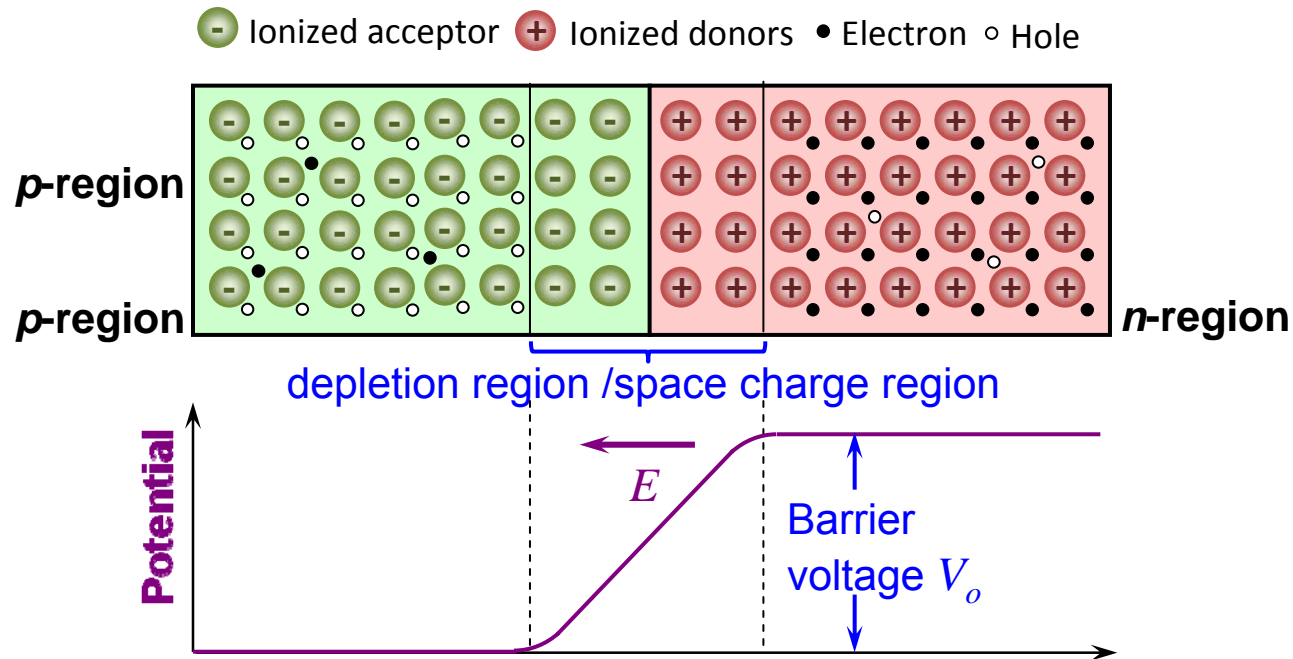
- Under open-circuit conditions, **no external current** exists. Hence, the two opposite currents across the junction are equal in magnitude: $I_D = I_S$
- Above equilibrium condition is maintained by the barrier voltage V_o . If I_D exceeds I_S , more bound charges are uncovered on both sides of junction, leading to a wider depletion region and higher voltage V_o across it. This in turn causes I_D to decrease until equilibrium is achieved with $I_D = I_S$.

pn Junction – Open-Circuit Conditions



- Similarly, if I_S exceeds I_D , amount of uncovered bound charges will decrease, the depletion region will narrow and the barrier voltage V_o will decrease. This in turn causes I_D to increase until equilibrium is achieved with $I_D = I_S$.
- The width of the depletion region (or the amount of bound charge therein) is just enough to give $I_D = I_S$. Beyond the depletion region, the p -region and n -region remain **electrically neutral**.

pn Junction – Open-Circuit Conditions



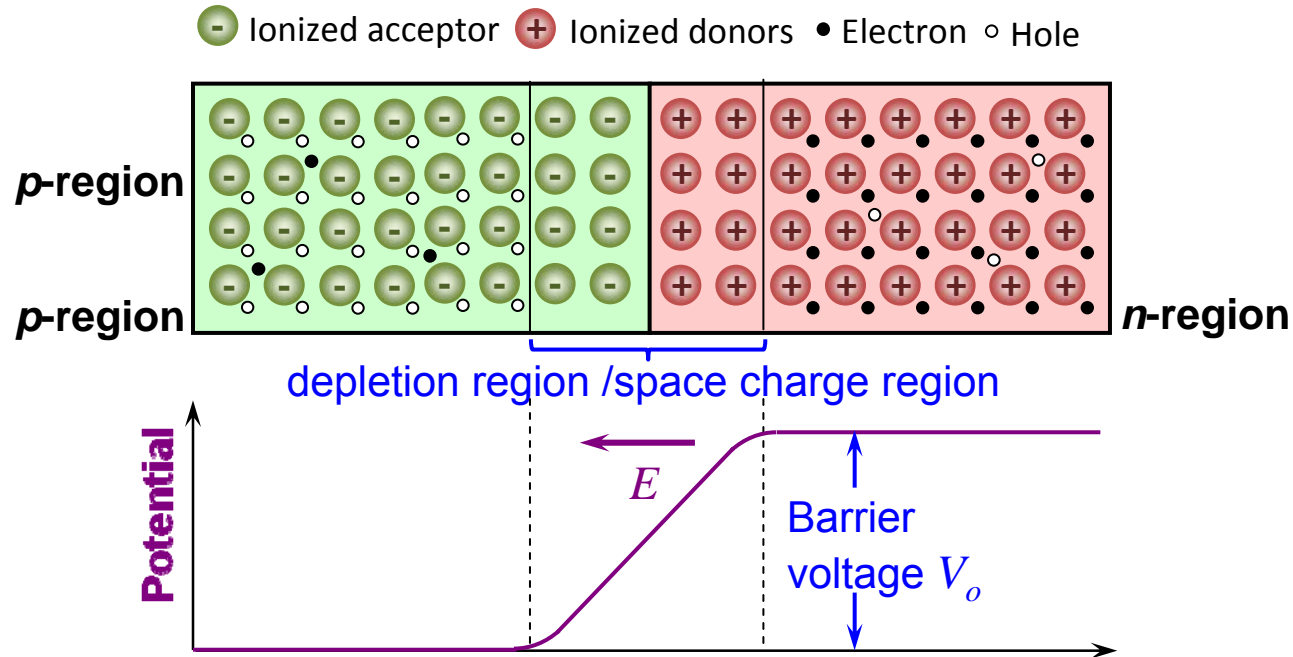
Junction Built-in Voltage (Barrier Voltage)

- With no external voltage applied, the **barrier voltage** V_o across the *pn* junction can be shown to be given by

$$V_o = \frac{kT}{q} \ln \left(\frac{N_A N_D}{n_i^2} \right) = V_T \ln \left(\frac{N_A N_D}{n_i^2} \right) \quad (2.1)$$

- $V_T = \frac{kT}{q} = 0.0259 \text{ V} \approx 0.025 \text{ V}$ at $T = 300 \text{ K}$ is the **thermal voltage**.

pn Junction – Open-Circuit Conditions

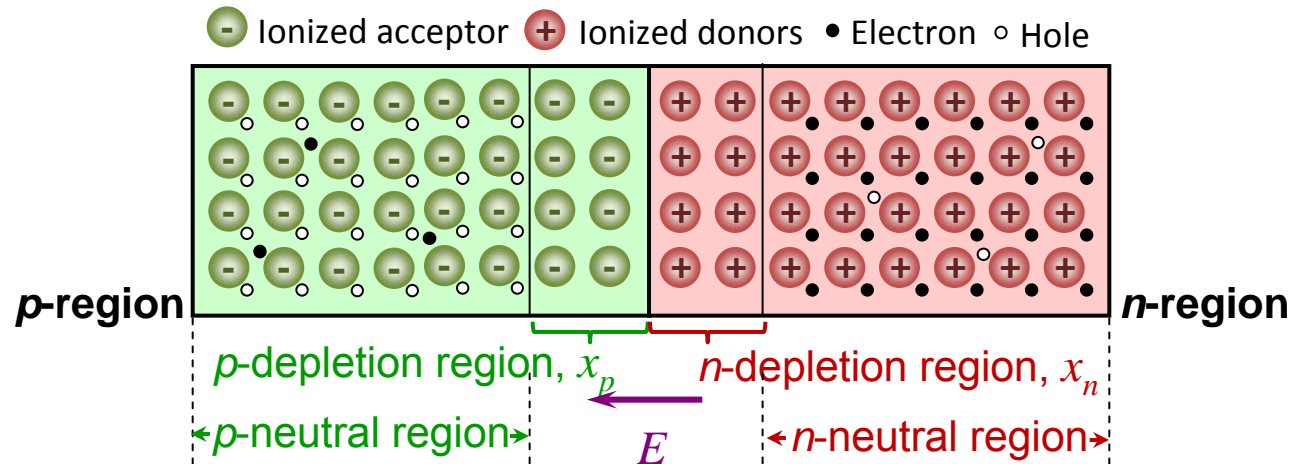


Junction Built-in Voltage (Barrier Voltage)

$$V_o = \frac{kT}{q} \ln \left(\frac{N_A N_D}{n_i^2} \right) = V_T \ln \left(\frac{N_A N_D}{n_i^2} \right)$$

- V_o is also known as the **junction built-in voltage** (an **internal** voltage). It depends on the doping concentrations of the *pn* junction (N_A , N_D) and is typically between 0.6 to 0.8 V for silicon at room temperature.

pn Junction – Open-Circuit Conditions



Width of the depletion region

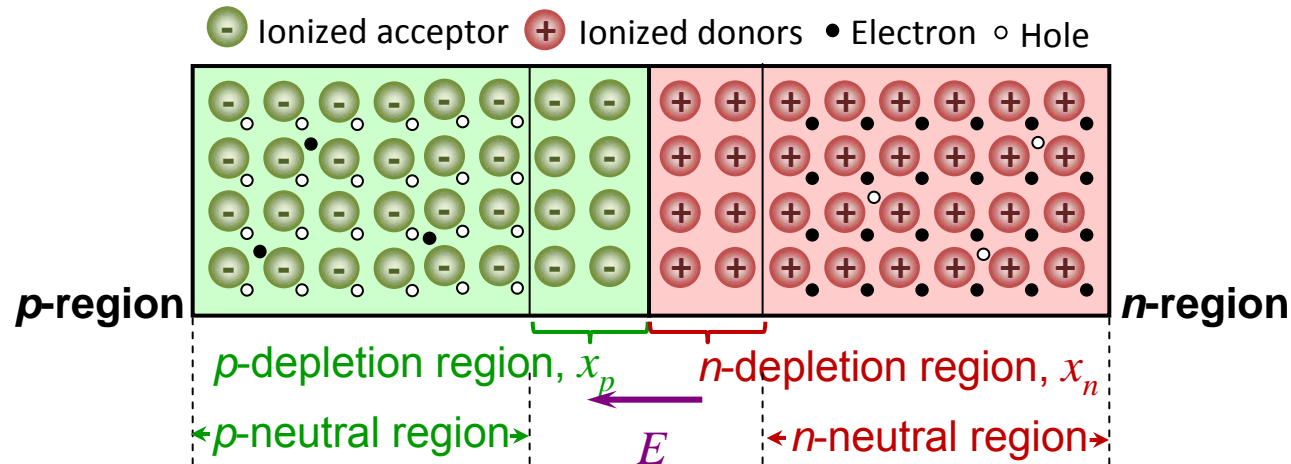
- Depletion region exists in **both** the p -region (of width x_p with negative ionized acceptors) and n -region (of width x_n with positive ionized donors), and the two sides have equal amounts of charge:

$$q_j = qx_p AN_A = qx_n AN_D \quad \left(\text{or} \quad \frac{x_p}{x_n} = \frac{N_D}{N_A} \right), \quad (2.2)$$

where A is the cross-sectional area of the pn junction.

- Higher N_A than N_D means $x_p < x_n$ and vice versa.

pn Junction – Open-Circuit Conditions



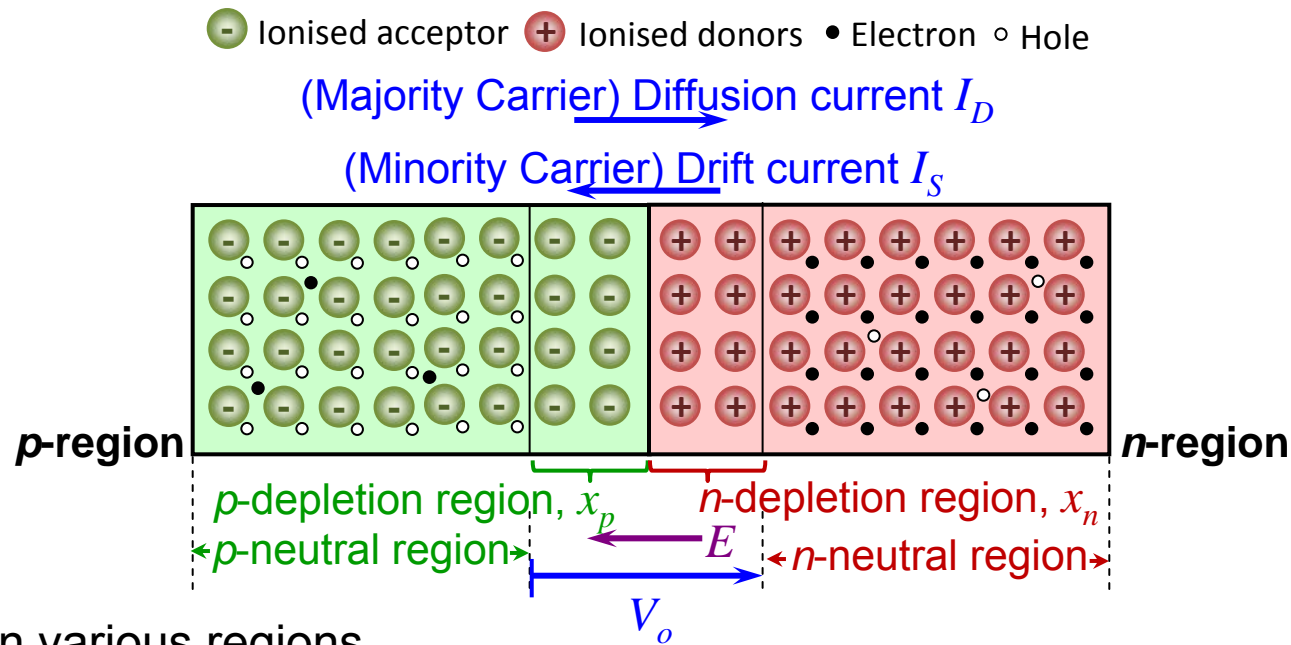
Width of the depletion region

- The (total) width of the depletion region W_{dep} of an open-circuited *pn* junction can be shown to be

$$W_{dep} = x_p + x_n = \sqrt{\frac{2\epsilon_s}{q} \left[\frac{1}{N_A} + \frac{1}{N_D} \right] V_o} \quad (2.3)$$

- ϵ_s is the permittivity of semiconductor = $11.7 \epsilon_o = 1.04 \times 10^{-12} \text{ F/cm}$ for silicon. ϵ_o is the permittivity of free space = $8.854 \times 10^{-14} \text{ F/cm}$.
- Higher N_A and N_D lead to smaller W_{dep} .

pn Junction – Open-Circuit Conditions



Charges in various regions

	Region			
Charge type	p -neutral	p -depletion	n -depletion	n -neutral
Acceptors ($-q$)	N_A	N_A	0	0
Donors ($+q$)	0	0	N_D	N_D
Holes ($+q$)	N_A	~ 0	~ 0	$\lll N_D$
Electrons ($-q$)	$\lll N_A$	~ 0	~ 0	N_D
Charge (C)	0	$-qAx_p N_A$	$+qAx_n N_D$	0

pn Junction – Open-Circuit Conditions

Exercise

For a Si *pn* junction with $N_A = 10^{17} \text{ cm}^{-3}$ and $N_D = 10^{16} \text{ cm}^{-3}$, find at $T = 300 \text{ K}$, the built-in voltage, and the width of the depletion region, including the distance it extends into the *p*-side and *n*-side of the junction under open-circuit conditions. Use $n_i = 1.5 \times 10^{10} \text{ cm}^{-3}$.

$$V_o = V_T \ln \left(\frac{N_A N_D}{n_i^2} \right) = 0.025 \cdot \ln \left(\frac{10^{17} \cdot 10^{16}}{[1.5 \times 10^{10}]^2} \right) = 0.728 \text{ V}$$

$$\begin{aligned} W_{dep} &= \sqrt{\frac{2\epsilon_s}{q} \left[\frac{1}{N_A} + \frac{1}{N_D} \right] V_o} \\ &= \sqrt{\frac{2(11.7)(8.854 \times 10^{-14})}{1.602 \times 10^{-19}} \left[\frac{1}{10^{17}} + \frac{1}{10^{16}} \right] (0.728)} \\ &= 0.32 \times 10^{-4} \text{ cm} = 0.32 \mu\text{m} \end{aligned}$$

pn Junction – Open-Circuit Conditions

$$W_{dep} = x_p + x_n \quad \text{and} \quad \frac{x_p}{x_n} = \frac{N_D}{N_A}$$

$$\Rightarrow W_{dep} = x_p + \frac{N_A}{N_D} x_p = x_p \left(\frac{N_D + N_A}{N_D} \right)$$

$$\therefore x_p = \left(\frac{N_D}{N_A + N_D} \right) W_{dep} \quad , \quad x_n = \left(\frac{N_A}{N_A + N_D} \right) W_{dep}$$

$$x_p = \left(\frac{10^{16}}{10^{17} + 10^{16}} \right) \times 0.32 \, \mu\text{m} = 0.029 \, \mu\text{m}$$

$$x_n = \left(\frac{10^{17}}{10^{17} + 10^{16}} \right) \times 0.32 \, \mu\text{m} = 0.29 \, \mu\text{m}$$

***pn* Junction**

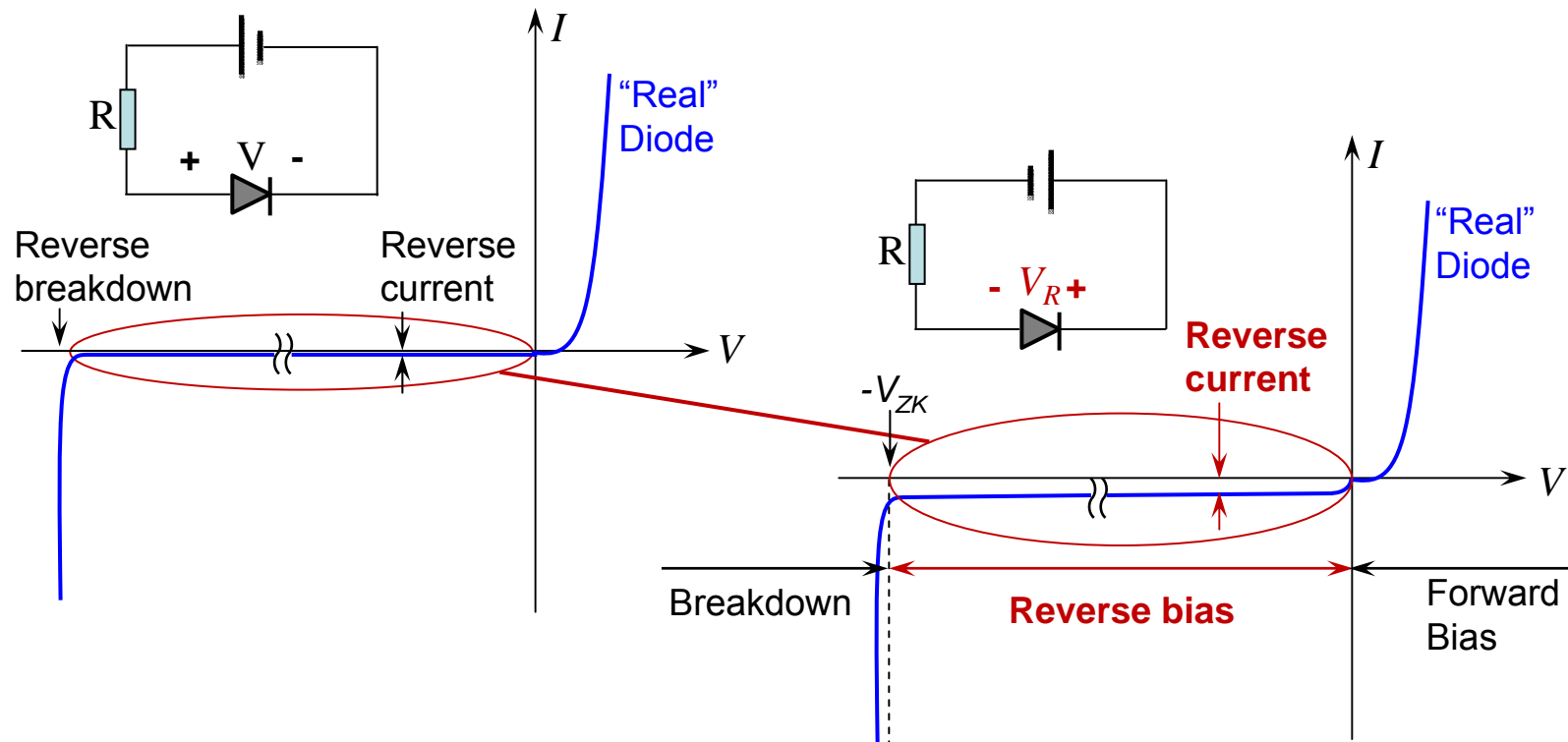
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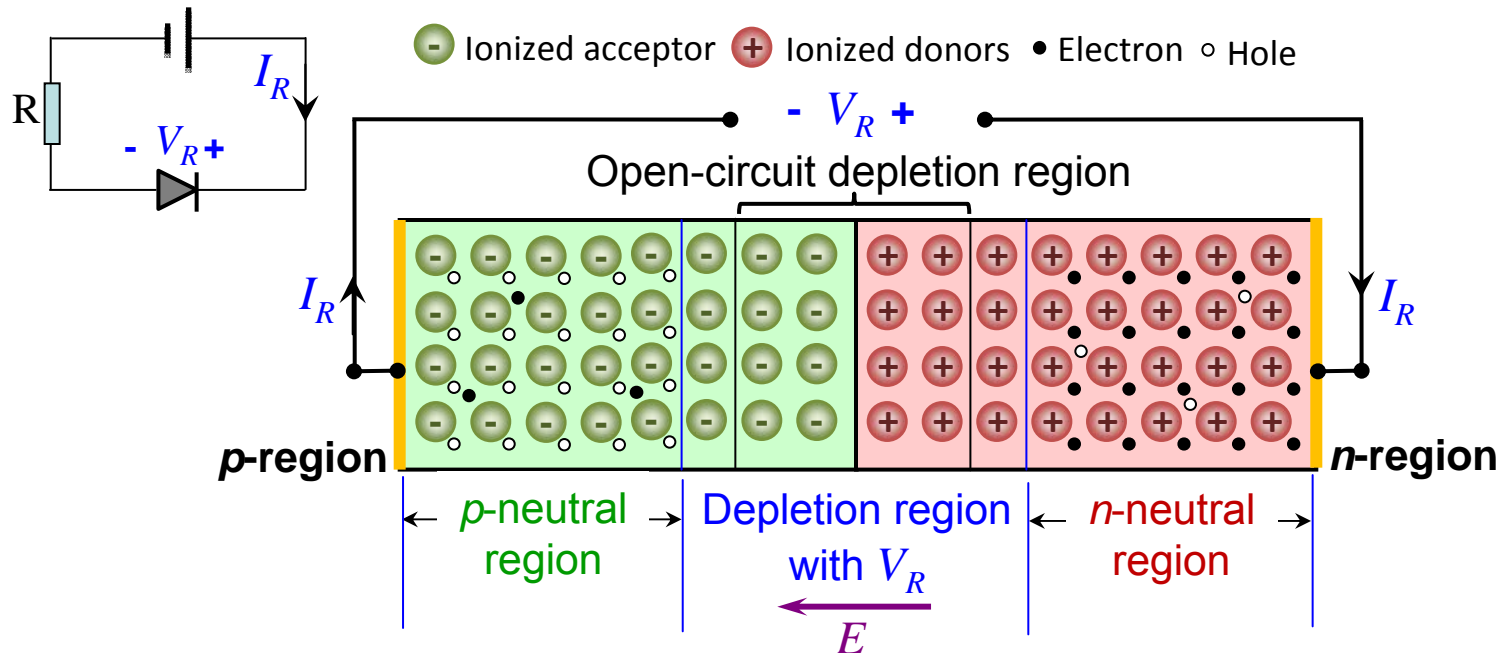
- A.D. Sedra & K.C. Smith, “Microelectronic Circuits – Theory and Application”, 5th Edition (International Version), Oxford University Press, Section 2.7.3.

pn Junction - Reverse-Bias Conditions



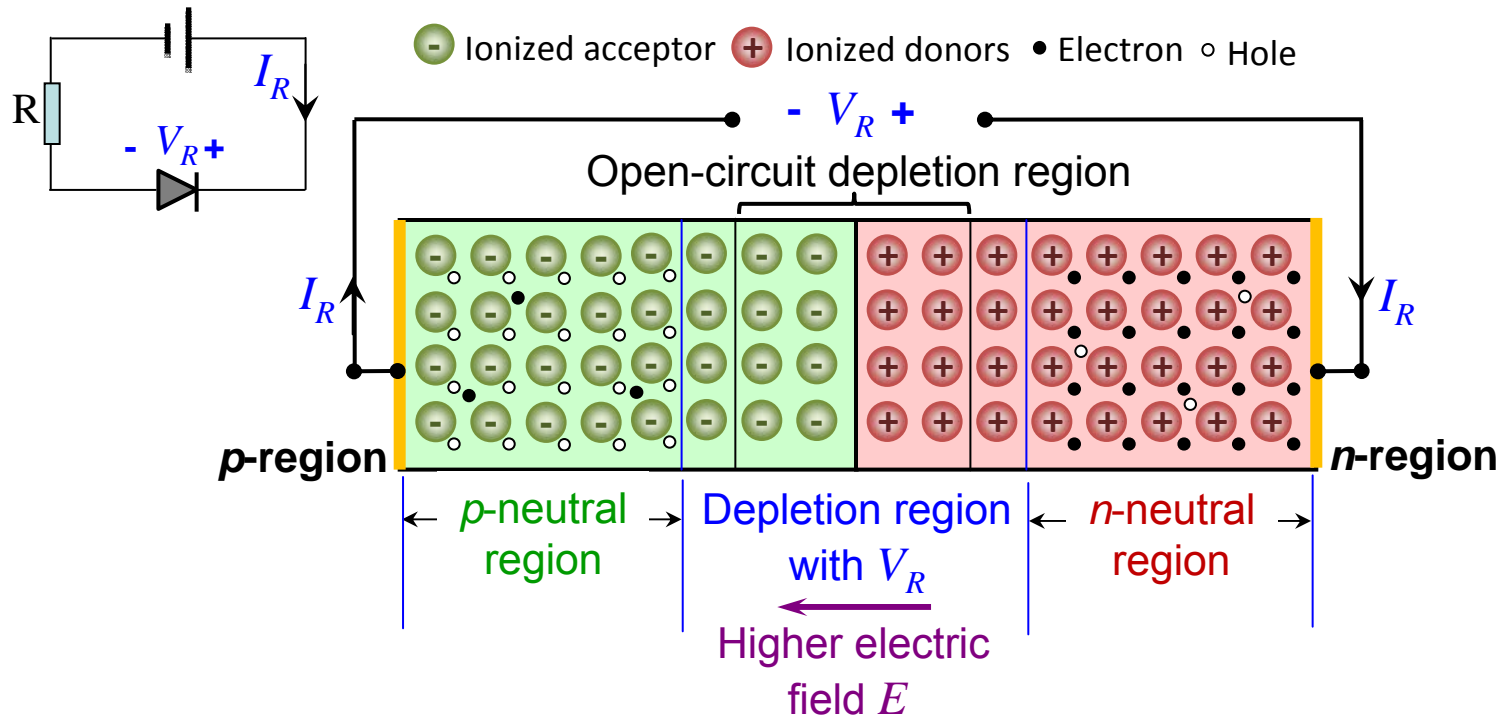
- Under reverse-bias, an external voltage supply is applied such that the voltage across the *pn* junction V_R is lower (or negative) at the *p*-region with respect to the *n*-region.
- We will consider the reverse bias region where V_R is less than the breakdown (knee) voltage V_{ZK} . For $V_R < V_{ZK}$, the (reverse) current flowing through the *pn* junction is **very small**, but **not zero**.

pn Junction – Reverse-Bias Conditions



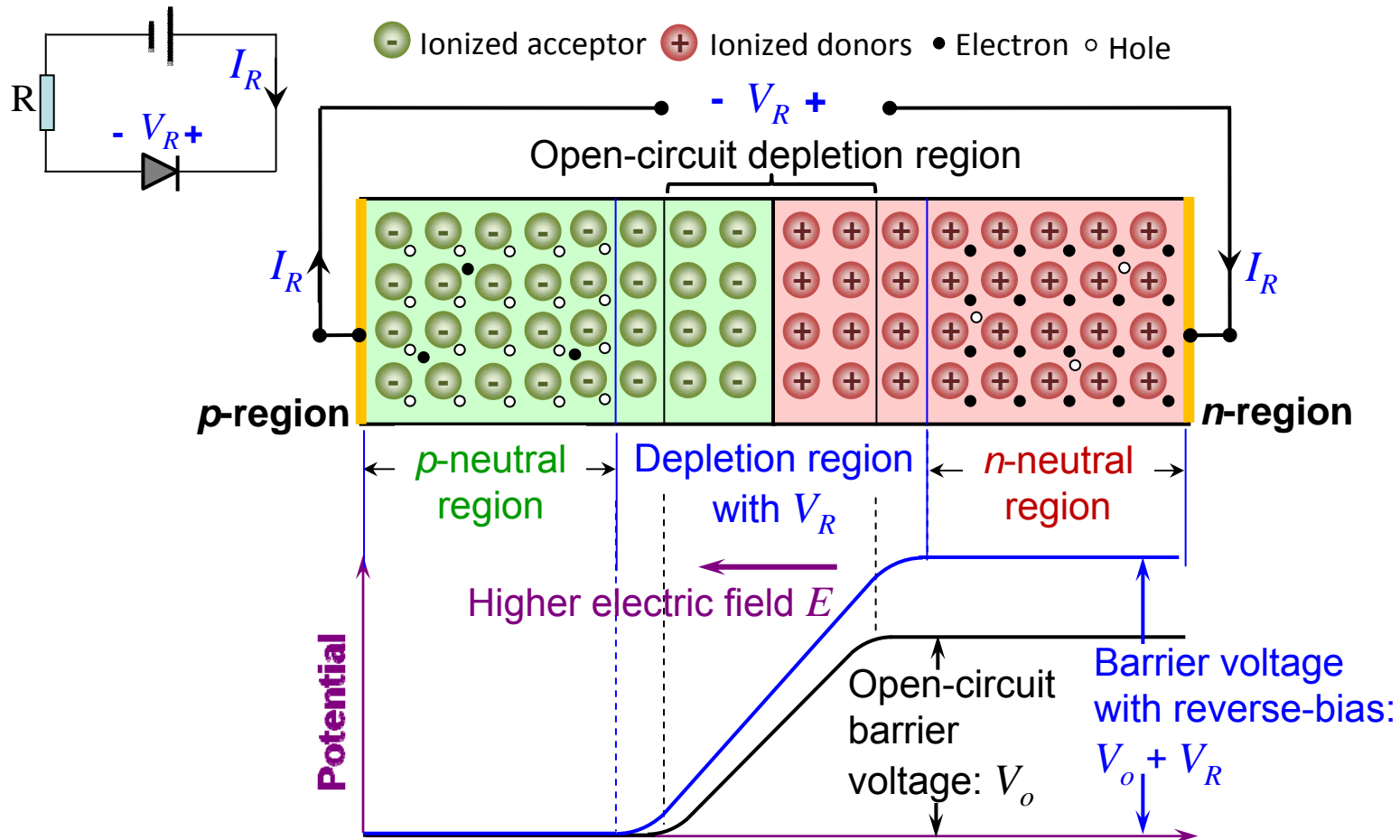
- Holes in the (neutral) p -region will be attracted by the **negative** voltage of V_R to leave the p -region and flow through the external circuit.
- Above causes the uncovered negative bound charge (ionized acceptors) to increase.
- Electrons in the (neutral) n -region will be attracted by the **positive** voltage of V_R to leave the n -region and flow through the external circuit.
- Above causes the uncovered positive bound charge (ionized donors) to increase.

pn Junction – Reverse-Bias Conditions



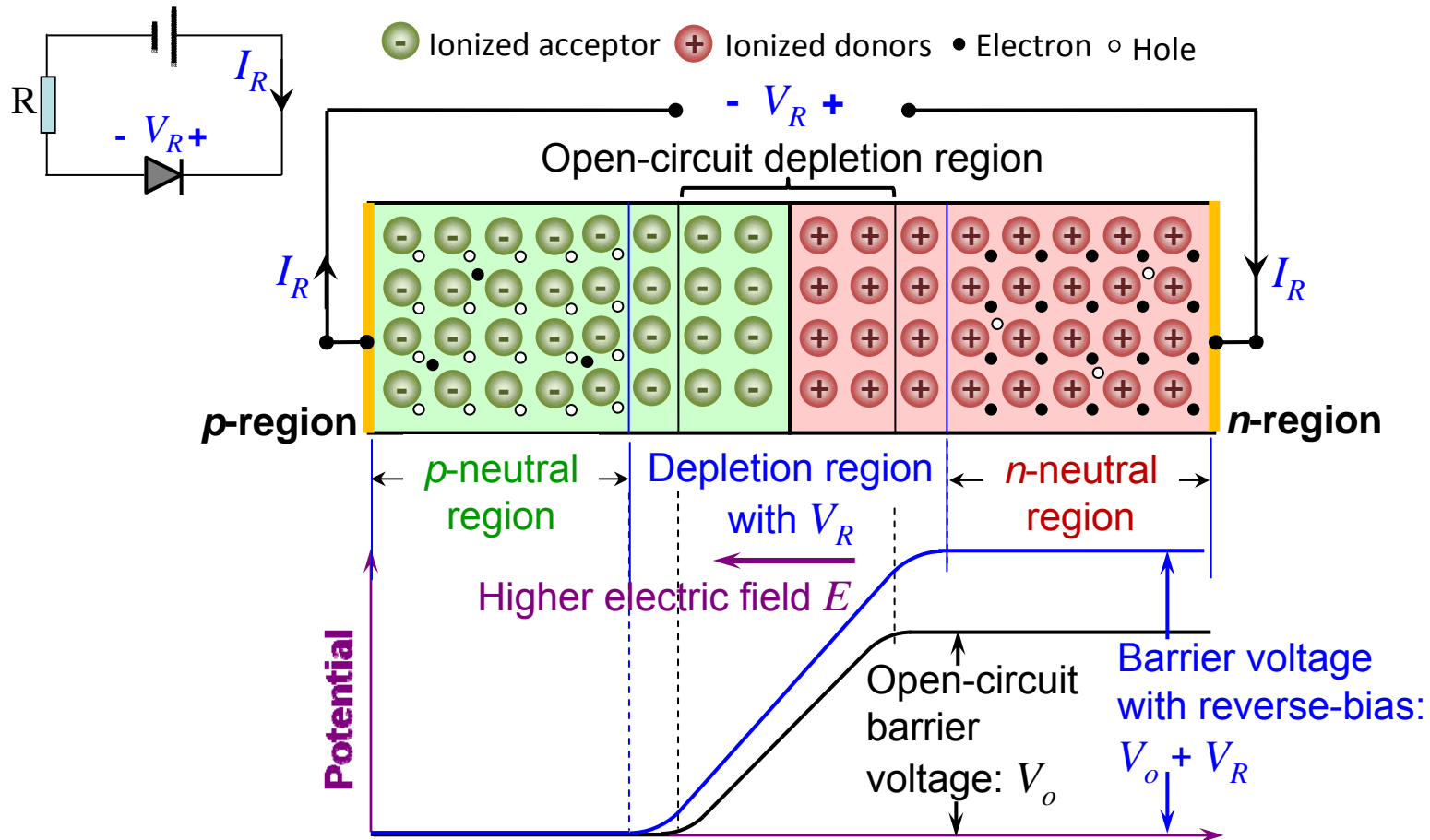
- A current flows from the positive terminal of V_R , through the pn junction (from n -region to p -region), to the negative terminal of V_R .
- As a result of increased bound charge, depletion region widens (with respect to that under open-circuit conditions)
 - Higher electric field and higher (barrier) voltage across the wider depletion region
 - Barrier voltage of the wider depletion region is given by $V_o + V_R$

pn Junction – Reverse-Bias Conditions



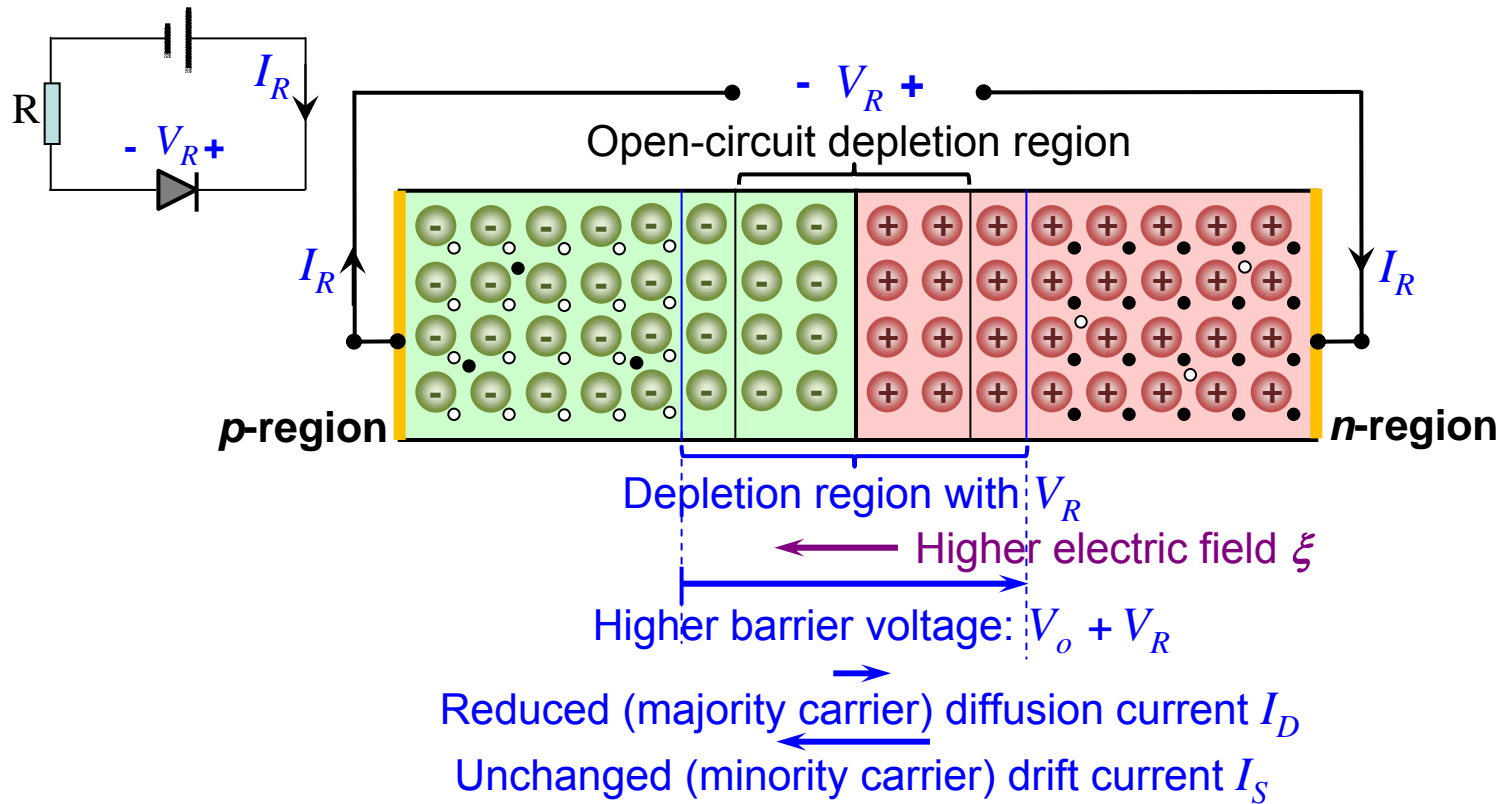
- Higher barrier voltage ($V_o + V_R$) and higher electric field opposing electron/hole diffusion leads to a **lower** (majority carrier) diffusion current I_D , which becomes negligible small with sufficient large magnitude of V_R .

pn Junction – Reverse-Bias Conditions



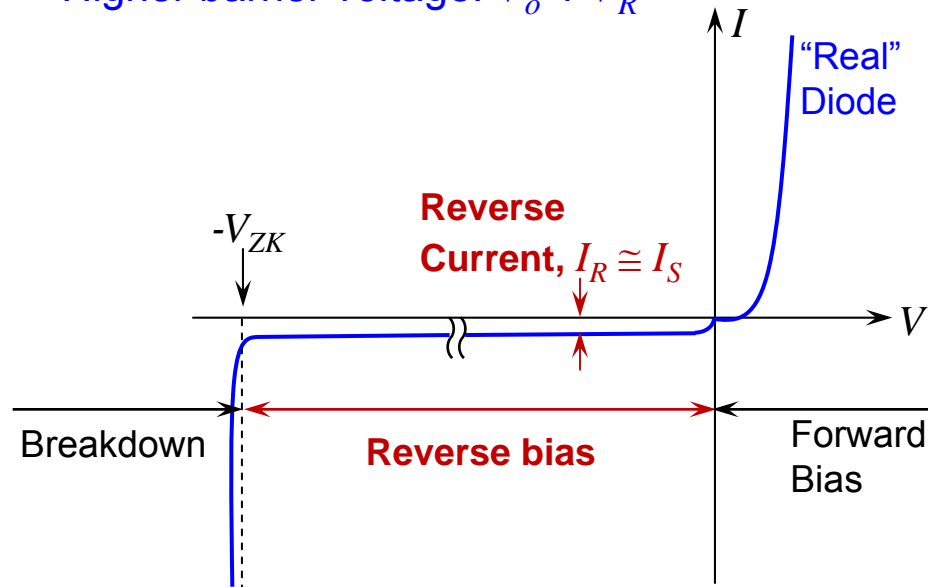
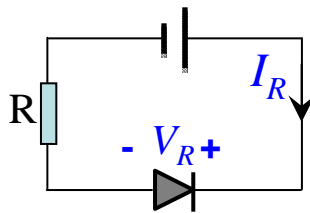
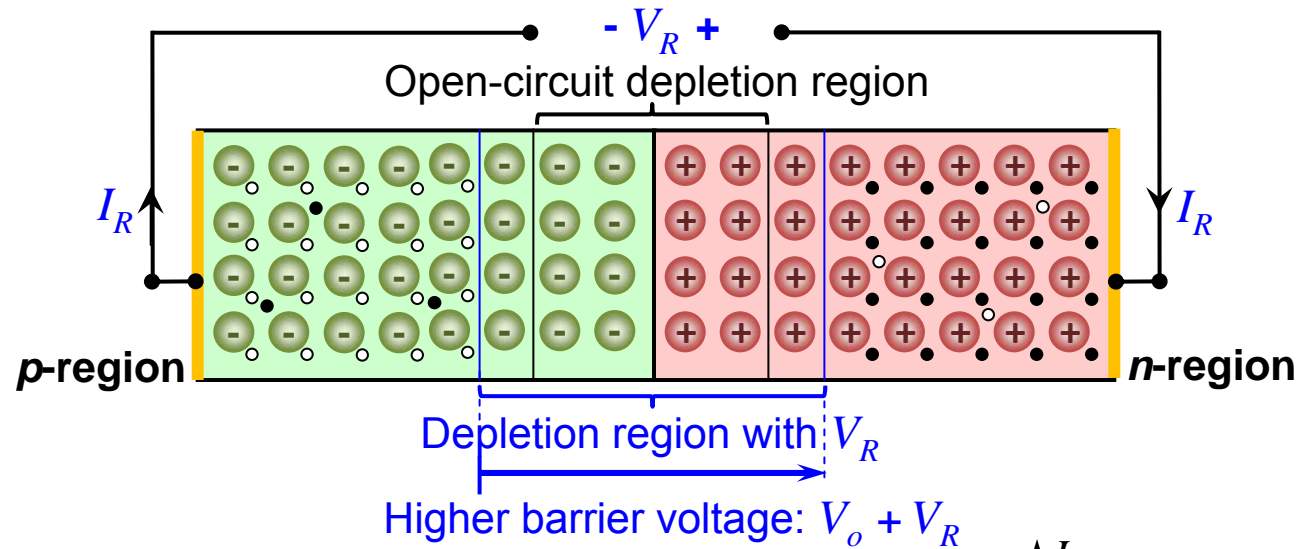
- No change in the (minority carrier) drift current I_S , since it is independent on barrier voltage across the depletion region.

pn Junction – Reverse-Bias Conditions

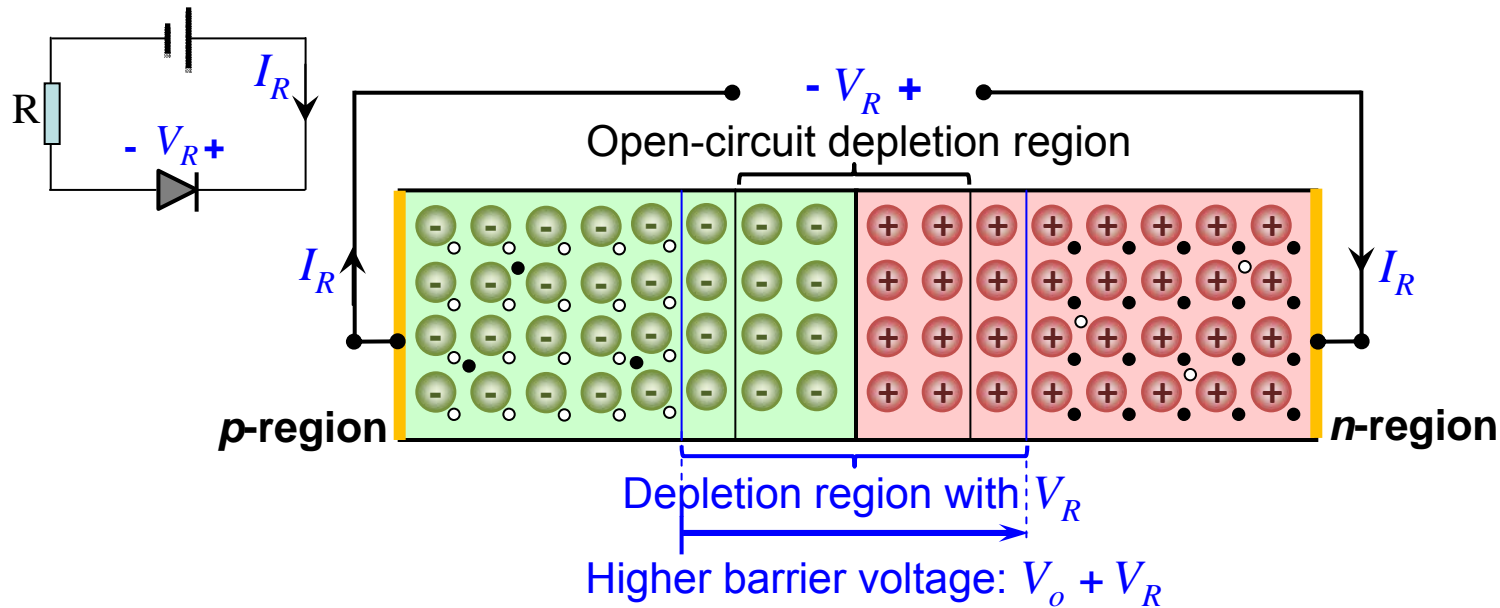


- (Net) current flow through a reverse-bias *pn* junction: $I_R = I_S - I_D \approx I_S$, a small current since I_S is due to drift of minority carriers.
- $I_R \approx I_S$ is also independent on V_R since I_S does not depend on the barrier voltage across the depletion region.

pn Junction – Reverse-Bias Conditions



pn Junction – Reverse-Bias Conditions



- (Total) width of the depletion region under reverse-bias is given by equation (2.3) by replacing V_o , the barrier voltage across the depletion region under open-circuit conditions, by $(V_o + V_R)$ –

$$W_{dep} = x_p + x_n = \sqrt{\frac{2\epsilon_s}{q} \left[\frac{1}{N_A} + \frac{1}{N_D} \right] (V_o + V_R)} \quad (2.4)$$

- Equation (2.2) that relates x_n and x_p is still valid.

pn Junction

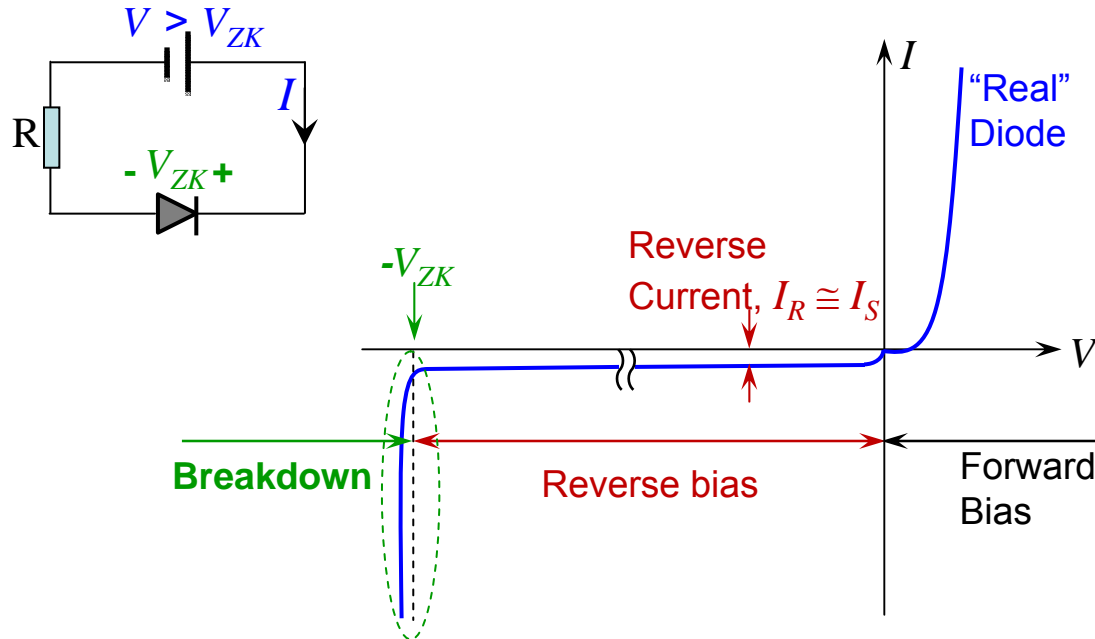
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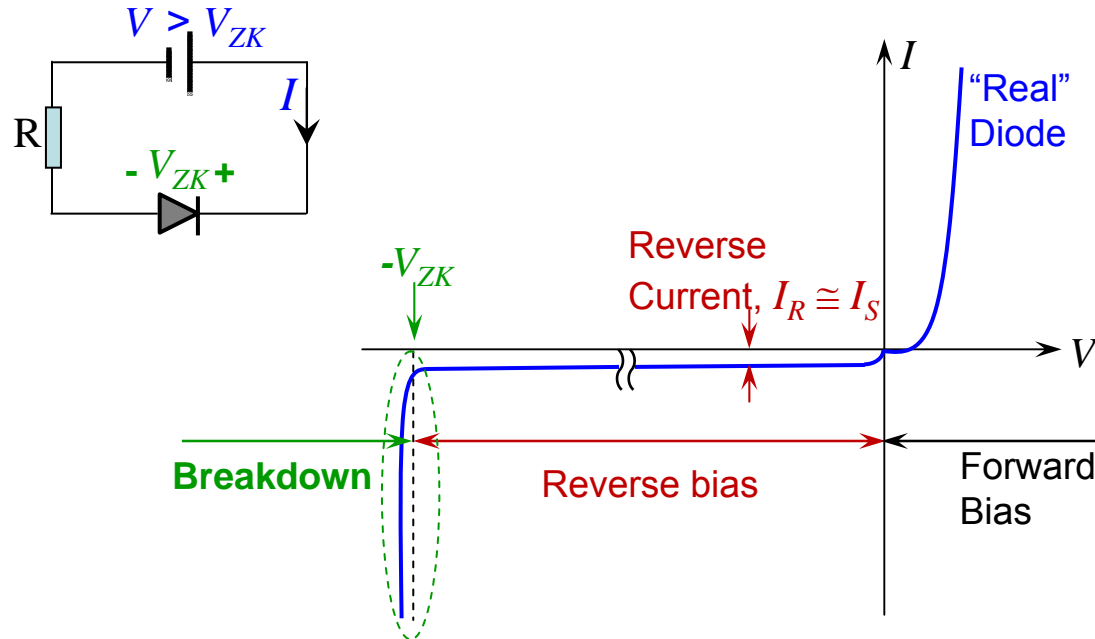
- A.D. Sedra & K.C. Smith, “Microelectronic Circuits – Theory and Application”, 5th Edition (International Version), Oxford University Press, Sections 2.7.4. & 2.4.

pn Junction – Breakdown Region



- When an external voltage supply in reverse bias $V > V_{ZK}$ (breakdown voltage) is applied across the *pn* junction, the (reverse) current increases rapidly with very small increase in the *pn* junction voltage. This condition is known as **breakdown**.
- Breakdown **does not destroy the junction** provided the current is kept below a certain level, to keep the power dissipation ($V \times I$) below what the device can handle. Current can be limited by connecting a resistor of suitable value in series with the *pn* junction.

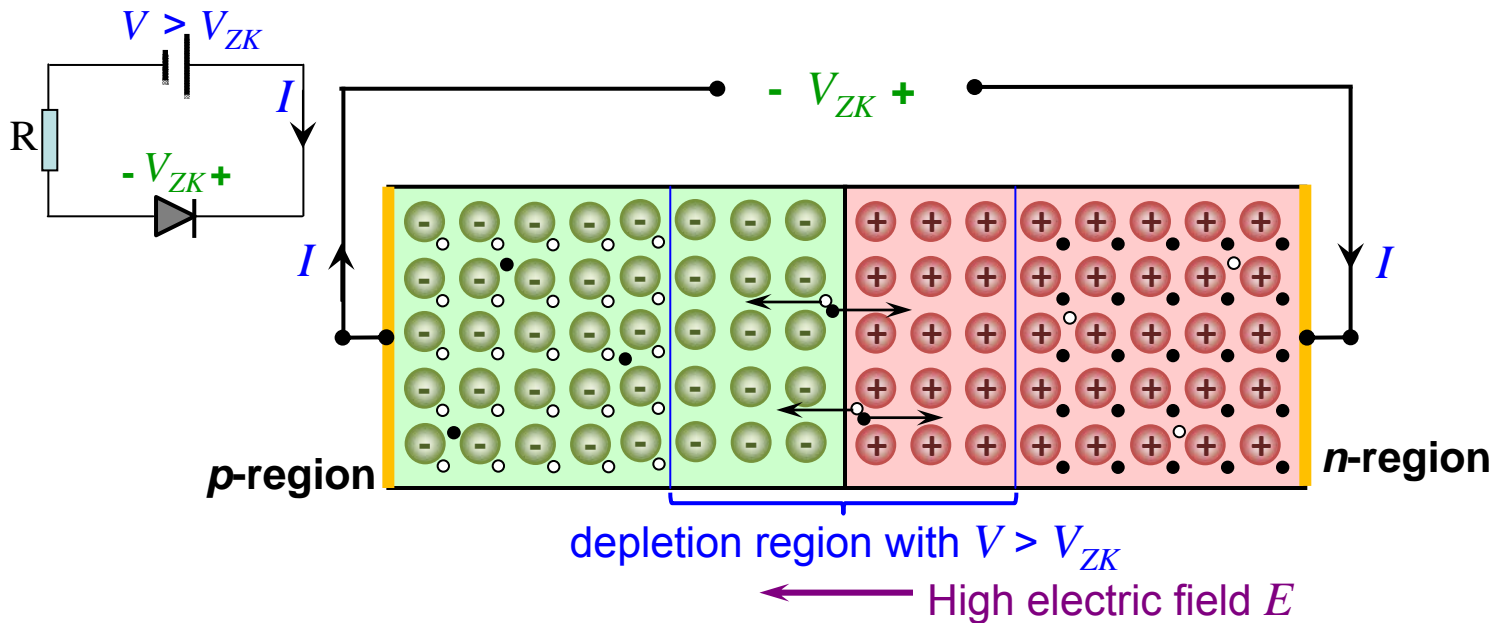
pn Junction – Breakdown Region



Breakdown mechanisms

- There are 2 possible breakdown mechanisms –
 - **Zener breakdown** - V_{ZK} is typically less than ~ 5 V
 - **Avalanche breakdown** - V_{ZK} is greater than approximately ~ 7 V
- For $5 \text{ V} < V_{ZK} < 7 \text{ V}$, the breakdown mechanism can be one of the above two mechanisms or a combination of the two.

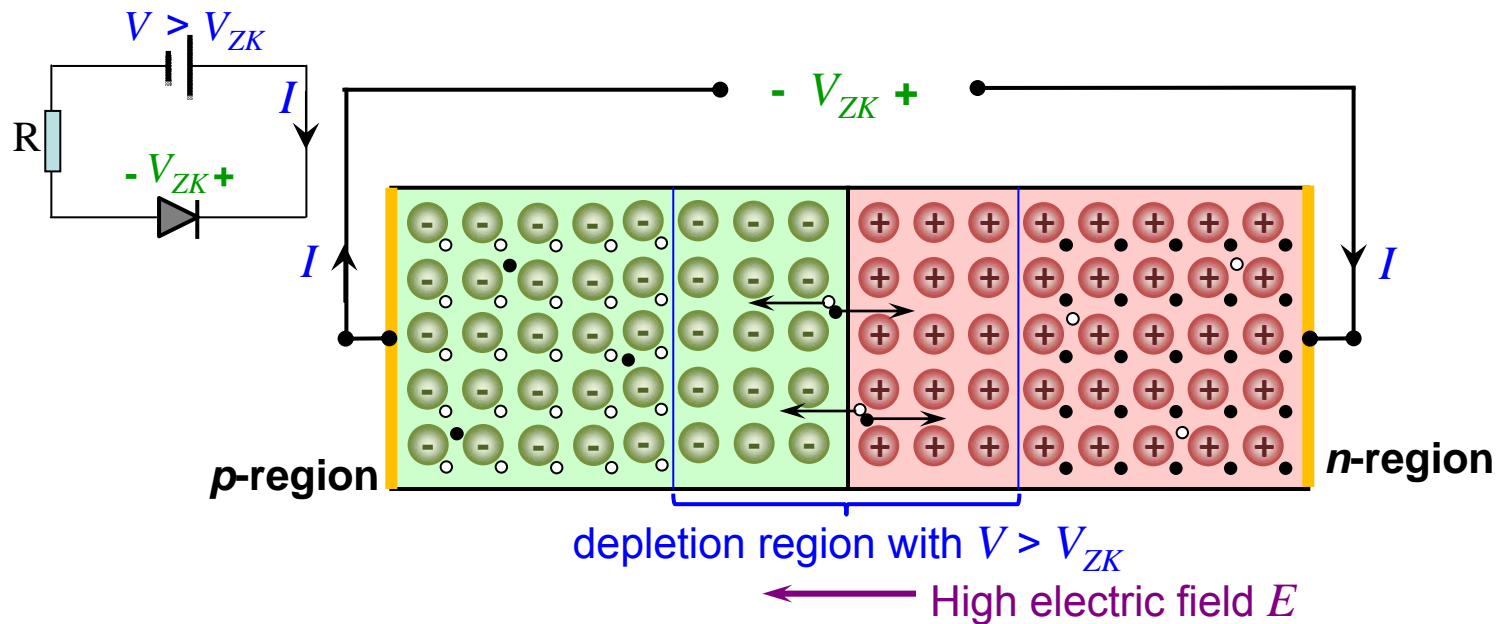
pn Junction – Breakdown Region



Zener breakdown

- Zener breakdown occurs when the electric field E in the depletion region has increased to a magnitude that it can **break** the covalent bonds and generate electron-hole pairs (EHPs).
- Electrons generated will be swept by the electric field E into the n -side and holes generated will be swept into the p -side. These electron and hole flows constitute an additional reverse current component.

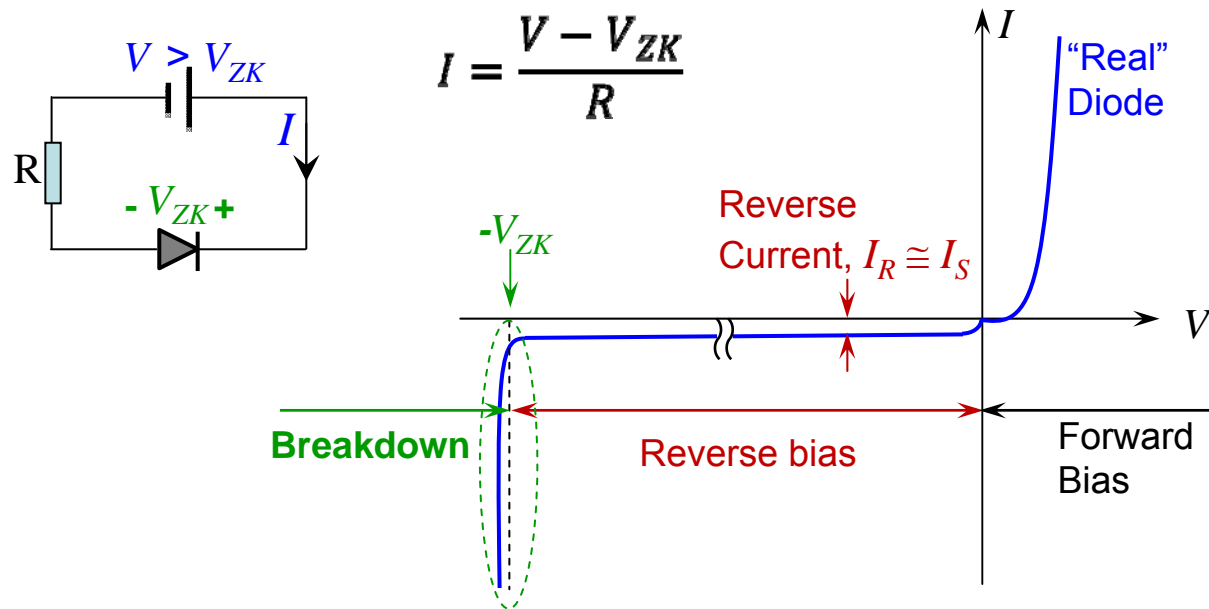
pn Junction – Breakdown Region



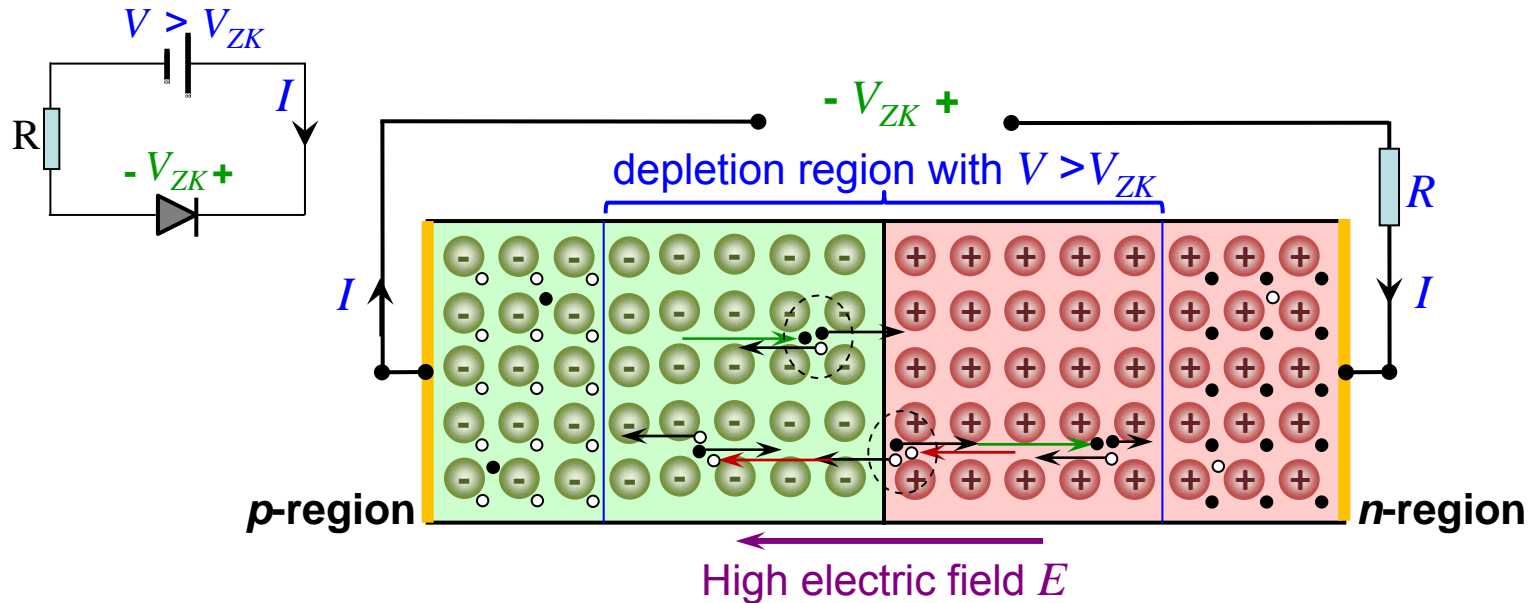
Zener breakdown

- Once Zener breakdown occurs ($V_R > V_{ZK}$), a large number of EHPs can be generated within the depletion region with negligible increase in the *pn* junction voltage. Hence, leading to large reverse current.
- Reverse current in the breakdown region will be determined by the external circuit, while reverse voltage across the *pn* junction will remain close to V_{ZK}

pn Junction – Breakdown Region



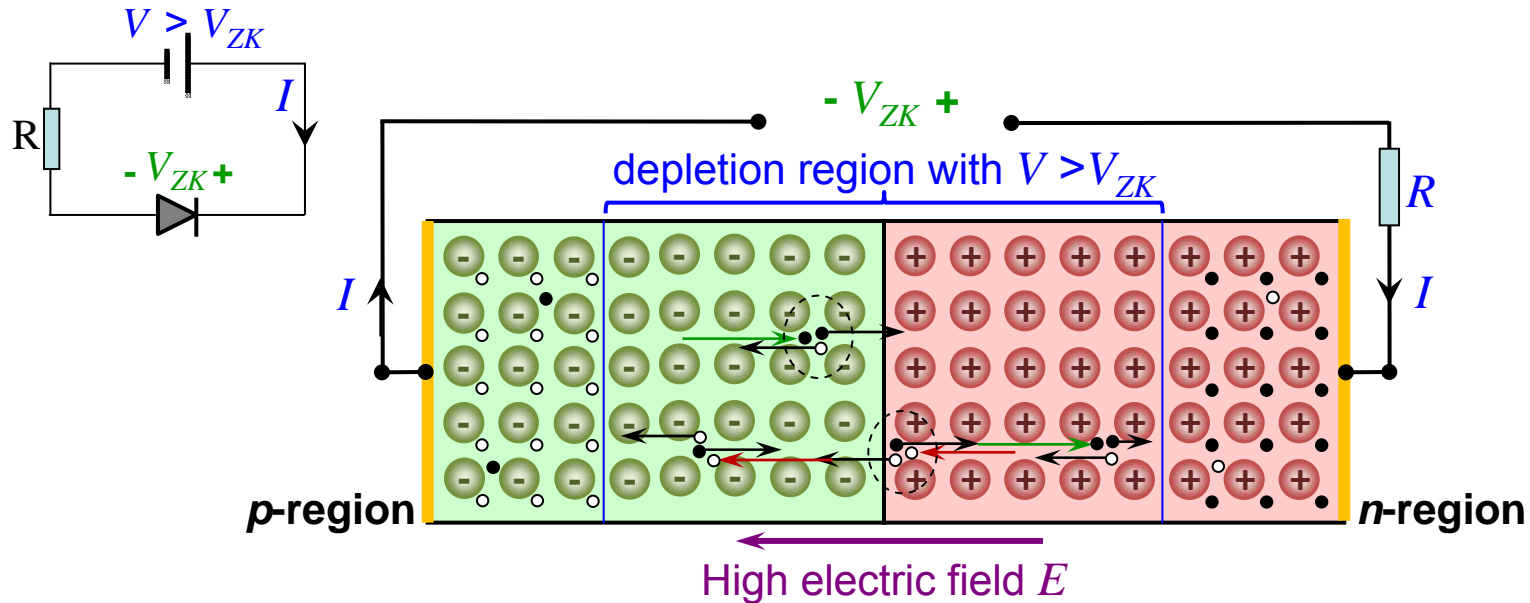
pn Junction – Breakdown Region



Avalanche breakdown

- Avalanche breakdown occurs when the minority carriers that cross the depletion region under the influence of the electric field (E) gain **sufficient kinetic energy** to be able to break the covalent bonds of atoms with which they collide. This results in EHPs generation and the process is called **impact ionization**.
- EHPs generated in turn may gain sufficient kinetic energy to cause additional impact ionizations and more EHPs generated, in an avalanche fashion. Hence, **avalanche multiplication**.

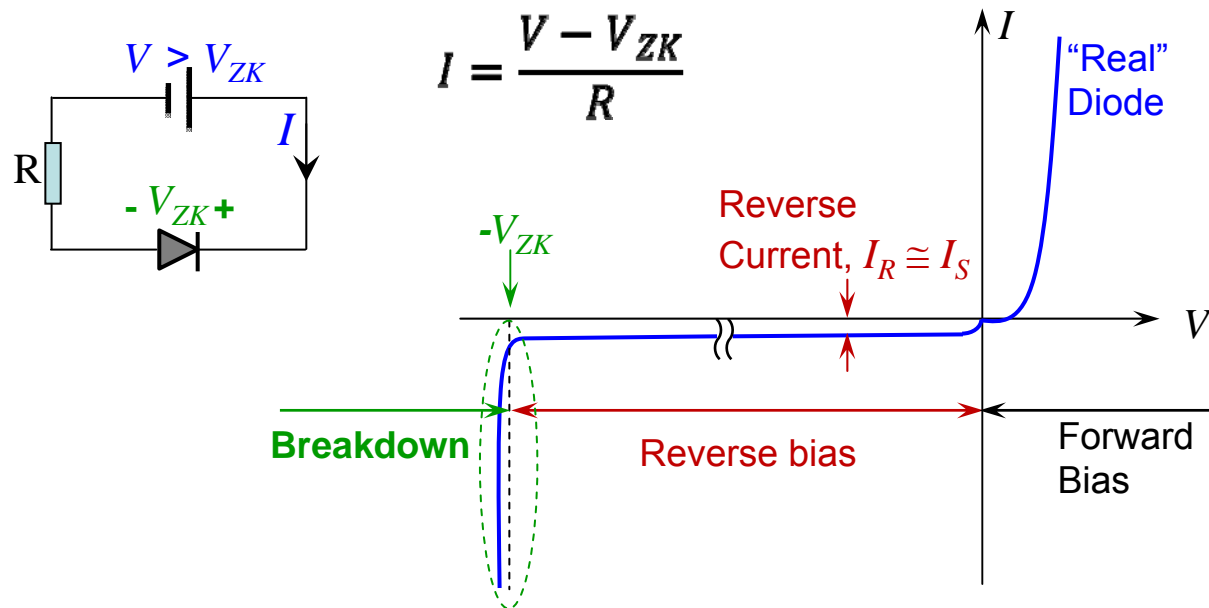
pn Junction – Breakdown Region



Avalanche breakdown

- Avalanche multiplication results in many EHPs generated in the depletion region, hence contributing to a large reverse current.
- The reverse current will be determined by the external circuit, with negligible increase in the *pn* junction voltage that remains close to V_{ZK}

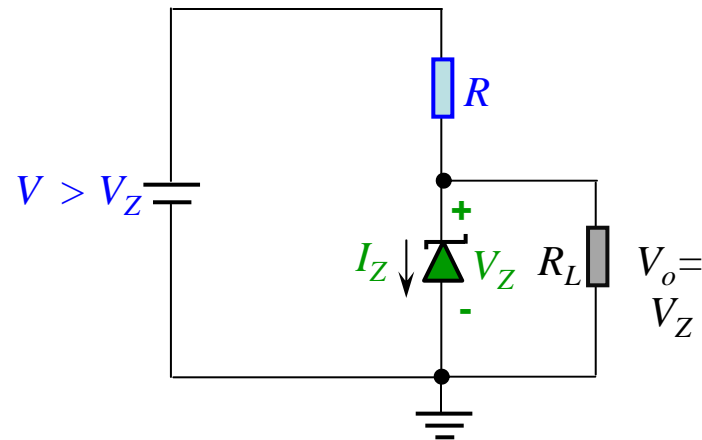
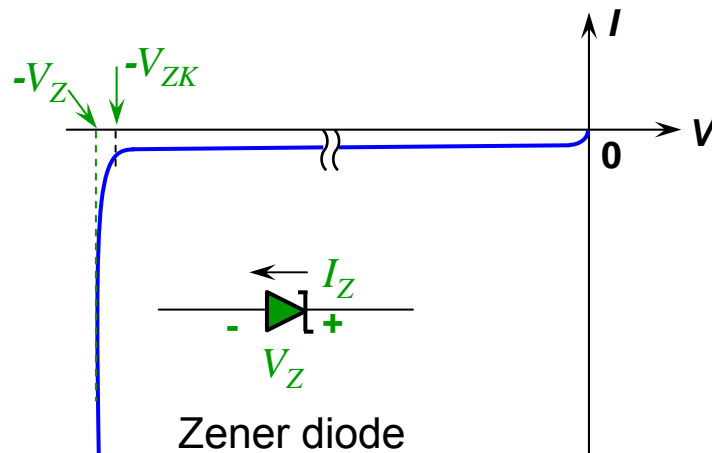
pn Junction – Breakdown Region



Avalanche versus Zener breakdown

- Avalanche breakdown requires **high** electric field **and** **wide** enough depletion region to allow sufficient travel distance for carriers to be accelerated to adequately high kinetic energy for impact ionization and multiplication.
- In contrast, Zener breakdown occurs in sufficiently **narrow** depletion region with **high** electric field.

pn Junction – Breakdown Region



Zener diode as a voltage reference

- **Zener diodes** are special diodes designed to operate in the breakdown region and they can be used in the design of **voltage regulator** (a circuit that provides a constant dc voltage between its terminals). Zener diodes are specified with V_Z , the breakdown voltage.
- In the above circuit, as long as $V > V_Z$ (meaning Zener diode operates in the breakdown region), the voltage across the load R_L is **kept constant** (or **regulated**) at $V_o = V_Z$ by the Zener diode.
- Virtually replaced by specially designed ICs that perform voltage regulation much more effectively and greater flexibility.

pn Junction

Appendix

Has full equivalent explanations on both p and n side of pn Junction

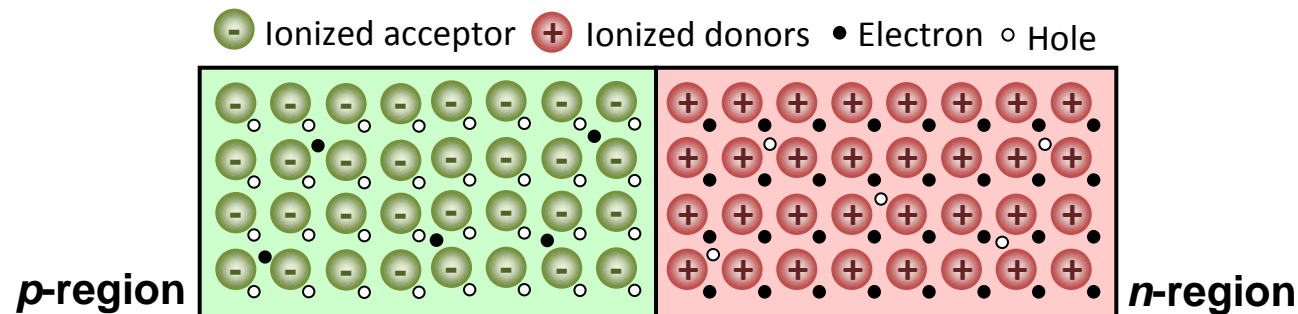
These slides also contain corresponding slide numbers in the above notes.
This table contains consolidated information on all slides in the appendix.

Appendix slide	Corresponding slide/s in the notes above
pn-1.45	pn-1.7
pn-1.46	pn-1.8
pn-1.47	pn-1.9
pn-1.48	pn-1.9,10
pn-1.49	pn-1.14

pn Junction – Open-Circuit Conditions

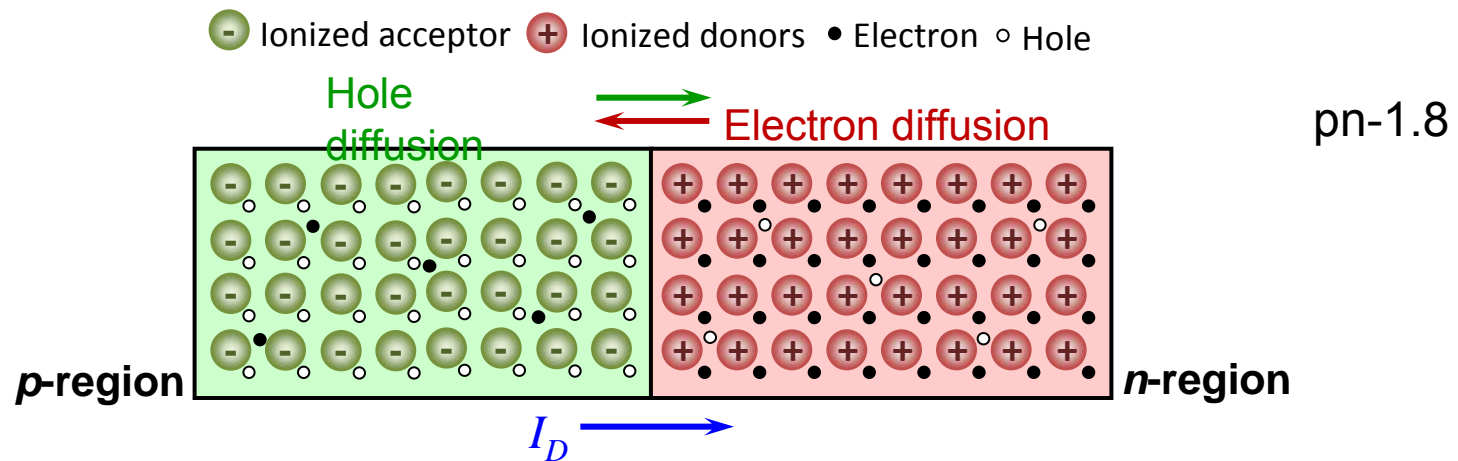


pn-1.7



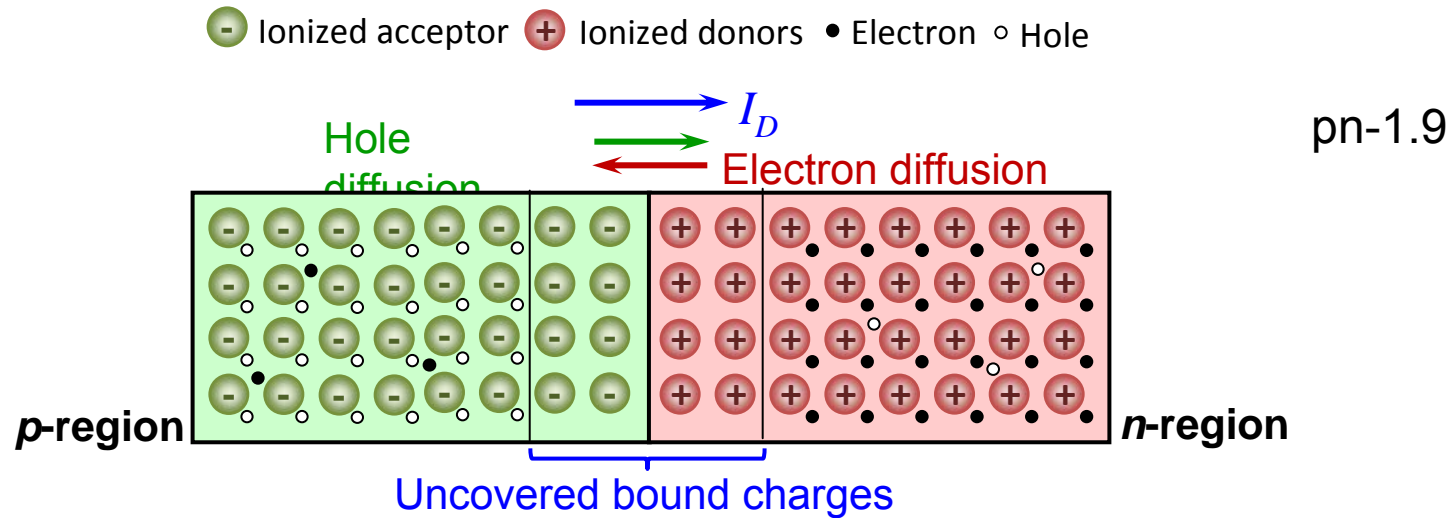
- | | |
|--|--|
| <ul style="list-style-type: none"> • The <i>p</i>-region has <ul style="list-style-type: none"> – N_A negative bound ionized acceptors – $p_{p0} \approx N_A$ positive holes (many - majority carriers) – $n_{p0} \approx n_i^2/N_A$ negative electrons (few - minority carriers) • Hence, electrically neutral | <ul style="list-style-type: none"> • The <i>n</i>-region has <ul style="list-style-type: none"> – N_D positive bound ionized donors – $n_{n0} \approx N_D$ negative electrons (many - majority carriers) – $p_{n0} \approx n_i^2/N_D$ positive holes (few - minority carriers) • Hence, electrically neutral |
|--|--|

pn Junction – Open-Circuit Conditions



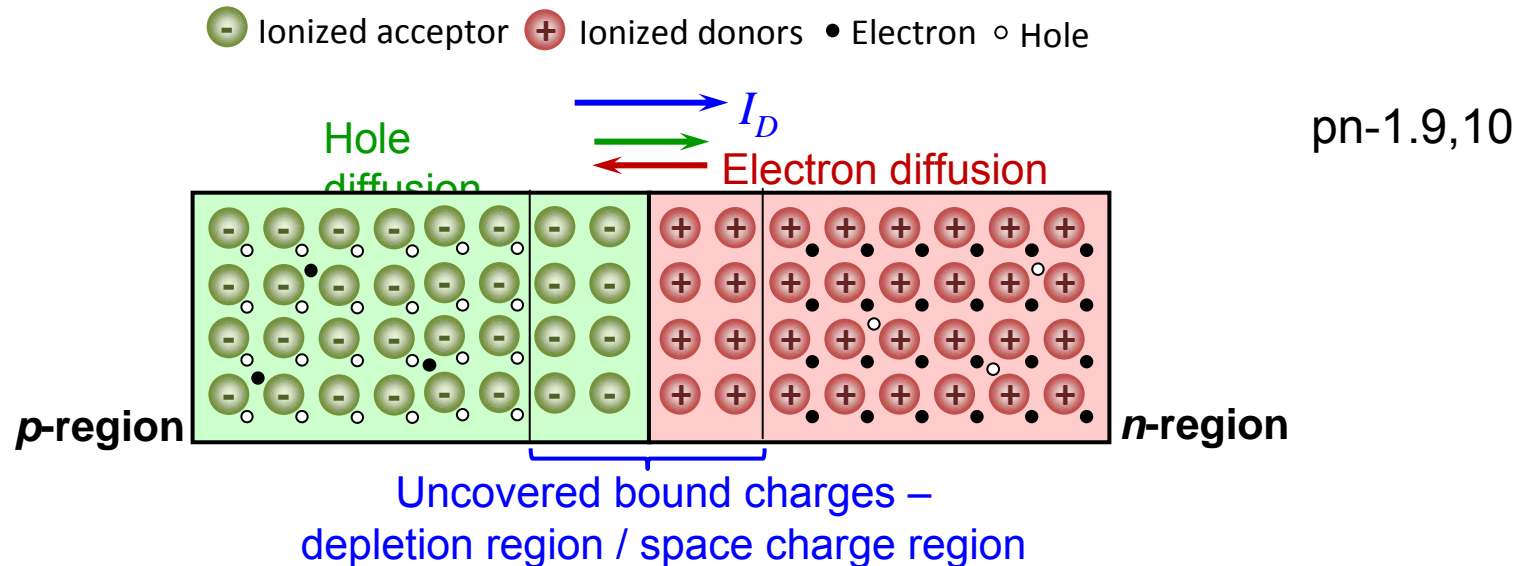
- The p -region has many more holes than the n -region (hole concentration gradient exists across the metallurgical junction).
- Hence holes diffuse from the p -region to the n -region
- The n -region has many more electrons than the p -region (electron concentration gradient exists across the metallurgical junction).
- Hence electrons diffuse from the n -region to the p -region
- The hole and electron diffusions add to form the diffusion current I_D (from the p -region to n -region). Note that I_D is due to majority carriers.

pn Junction – Open-Circuit Conditions



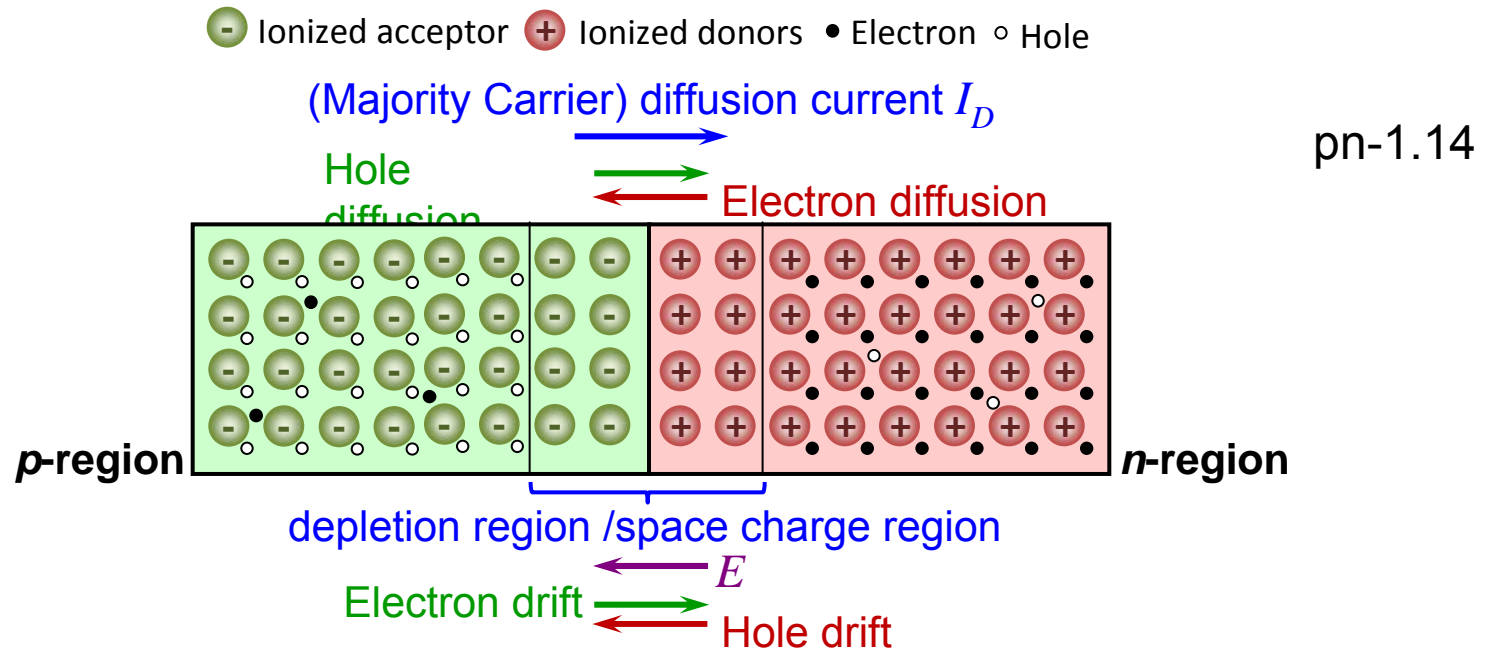
- Electrons that diffuse across the junction into the p -region **recombine** with some of the majority holes there
- Hence, some holes in the p -region near the junction disappear, resulting in the bound negative ionized **acceptors** being **no longer neutralized** by holes (i.e., uncovered)
- Holes that diffuse across the junction into the n -region **recombine** with some of the majority electrons there
- Hence, some electrons in the n -region near the junction disappear, resulting in the bound positive ionized donors being **no longer neutralized** by electrons (i.e., uncovered)

pn Junction – Open-Circuit Conditions



- Region near the junction comprises **uncovered bound negative ionized acceptors** and is depleted of holes.
- Hence, region is known as the **depletion region**. It is also called the **space charge region** as it has negative charges (ionized acceptors).
- Region near the junction comprises **uncovered bound positive ionized donors** and is depleted of electrons.
- Hence, region is known as the **depletion region**. It is also called the **space charge region** as it has positive charges (ionized donors).

pn Junction – Open-Circuit Conditions



- Some thermally generated minority electrons in the p -region diffuse through the p -region to the edge of the depletion region, where electrons experience the electric field E and are swept across the depletion region into the n -region (**drift of electrons**).
- Some thermally generated minority holes in the n -region diffuse through the n -region to the edge of the depletion region, where holes experience the electric field E and are swept across the depletion region into the p -region (**drift of holes**).