Numerical Algorithms for ODEs/DAEs (Transient Analysis)

Solving Differential Equation Systems

$$\frac{d}{dt}\vec{q}(\vec{x}(t)) + \vec{f}(\vec{x}(t)) + \vec{b}(t) = \vec{0}$$

- DAEs: many types of solutions useful
 - ✓ DC steady state: no time variations
 - transient: ckt. waveforms changing with time
 - periodic steady state: changes periodic w time
 - → linear(ized): all sinusoidal waveforms: AC analysis
 - → nonlinear steady state: shooting, harmonic balance
 - noise analysis: random/stochastic waveforms
 - sensitivity analysis: effects of changes in circuit parameters

Transient Analysis

$$\frac{d}{dt}\vec{q}(\vec{x}(t)) + \vec{f}(\vec{x}(t)) + \vec{b}(t) = \vec{0}$$

- What
 - inputs b(t) changing with time
 - → find waveforms of x(t) as they change with time
- Why
 - most general analysis typically needed
 - sine wave, pulse, etc. inputs typical in many applications
- How
 - solve DAE using numerical methods
 - → "discretize time": replace d/dt term
 - convert DAE to nonlinear algebraic equation at each discrete time point
 - → solve this using NR

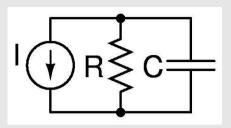
Solving DAEs: Preliminaries

- Given a DAE: does it have a solution?
 - Depends. Various conditions need to hold.
- Easier to analyze if DAEs are really ODEs
 - Ordinary Differential Equations: i.e., $\vec{q}(\vec{x}) \equiv \vec{x}$

$$\frac{d}{dt}\vec{x}(t) + \vec{f}(\vec{x}(t)) + \vec{b}(t) = \vec{0} \Rightarrow \frac{d}{dt}\vec{x}(t) = \vec{g}(\vec{x}, t)$$

- Existence/Uniqueness conditions well known for ODEs
 - → f(x) needs to be Lipschitz
 - device models must be smooth / bounded / "physically reasonable"
 - watch those if conditions, 1/x, log(x), sqrt(x), etc. terms!
- DAEs? Much more involved
 - in practice: modifications of ODE methods + heuristics

Analytical Exemplar (Model Problem)



$$A \stackrel{k_1}{\rightleftharpoons} B$$

$$C\frac{de}{dt} = -\frac{e}{R} - I(t)$$

$$\frac{d[A]}{dt} = -(k_1 + k_2)[A] + k_2$$

$$\dot{x} = \lambda x + b(t)$$

- Useful because
 - has analytical solution
 - vector linear systems reducible to this form
 - locally approximates nonlinear systems
- Prototype for
 - stability analysis of ODE solution methods

Existence and Uniqueness

- Does a solution exist? Examples:
 - $\dot{x} = \lambda x + b(t)$

$$\Rightarrow \text{ yes: } x(t) = x(t_0)e^{\lambda(t-t_0)} + \int_0^t e^{\lambda(t-\tau)}b(\tau)\,d\tau$$

→ Solution is unique, given an initial condition

•
$$\dot{x} = -\frac{1}{2x}, \quad x(0) = x_0$$

$$\Rightarrow x(t) = \sqrt{x_0^2 - t}$$
: no solution for $t > x_0^2$

$$\bullet \ \dot{x} = \frac{3}{2}x^{\frac{1}{3}}$$

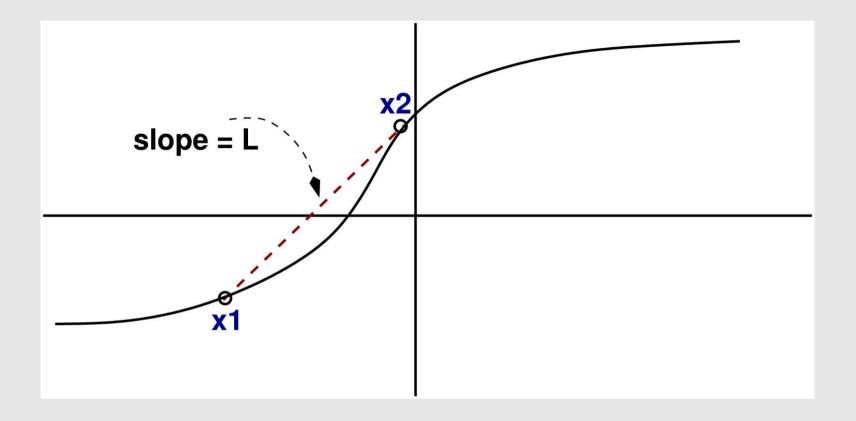
$$x(t) = \begin{cases} 0, & 0 <= t <= k \\ (t-k)^{\frac{3}{2}}, & t > k \end{cases}, \text{ for ANY } \frac{k > 0}{k > 0}!$$

→ Infinite number of solutions!

Existence/Uniqueness Theorem

- Picard-Lindelöf Theorem (roughly)
 - If $\vec{g}(\vec{x},t)$ is
 - defined over all t, x
 - → Lipschitz for all x
 - then $\dot{\vec{x}}(t) = \vec{g}(\vec{x},t)$, $\vec{x}(t_0) = \vec{x}_0$ has a unique global solution
- $\vec{g}(\vec{x},t)$ is Lipschitz if
 - there exists some <u>finite</u> L such that:
 - $||\vec{g}(\vec{x}_1,t) \vec{g}(\vec{x}_2,t)|| < L||\vec{x}_1 \vec{x}_2||, \quad \forall \vec{x}_1, \vec{x}_2$

The Lipschitz Condition



- Linear systems: Lipschitz
- $\sqrt[3]{x}$, $\frac{1}{x-x_0}$: not Lipschitz

Is this globally Lipschitz?

$$i_{E}(t) \downarrow = \underbrace{\sum_{i=d(v)}^{e_{1}(t)} e_{2}(t)}_{i_{E}(t)}$$

n1 KCL:
$$i_E + d(e_1 - e_2) = 0$$

n2 KCL: $-d(e_1 - e_2) + \frac{e_2}{R} + \frac{d}{dt}(Ce_2) = 0$
E BCR: $e_1 - E = 0$

n2 KCL:
$$\frac{d}{dt}e_2 + \frac{e_2}{RC} - \frac{d(E(t) - e_2)}{C} = 0$$

Solving ODEs: Overview of Strategy

- Discretize the time axis (starting from, e.g., t=0)
 - $t_0 = 0, t_1, t_2, \cdots, t_N$
- Approximate d/dt term by finite difference

$$wo$$
 e.g., $rac{dec{x}}{dt} \simeq rac{ec{x}_i - ec{x}_{i-1}}{t_i - t_{i-1}}$

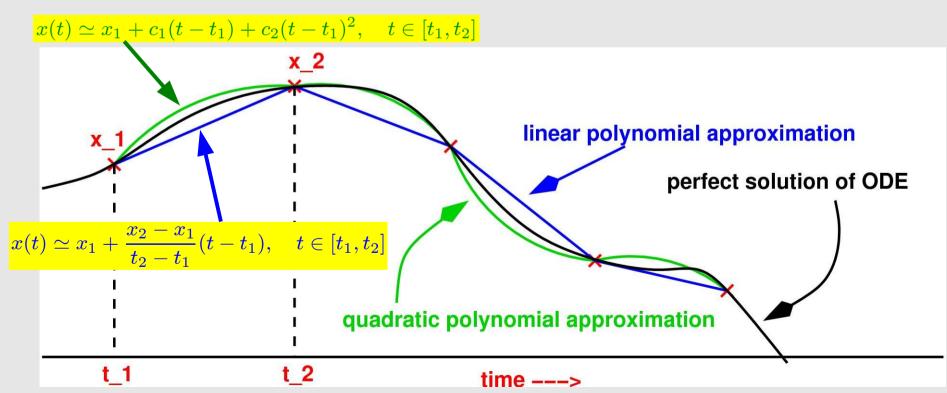
ODE becomes algebraic nonlinear equation

$$\underbrace{\frac{\vec{x}_i - \vec{x}_{i-1}}{t_i - t_{i-1}} + \vec{f}(\vec{x}_i) + \vec{b}(t_i)}_{\vec{h}_i(\vec{x}_i)} = \vec{0}$$

- Solve using Newton-Raphson
 - → repeat for i=1,2,3,...,N
 - → using previous solution (or initial condition) at t_{i-1}
- Key issues:
 - → how to discretize? errors/accuracy? computation? coding?

Piecewise Polynomial Approximation

- Using locally polynomial bases
 - assume: ODE solution is locally polynomial
 - characterize polynomial with a few numbers
 - → e.g., samples at different points
 - find those numbers so that the local polynomial satisfies the ODE



Linear Polynomials: FE and BE

Use of locally linear approximations for x(t)

•
$$\vec{x}(t) \simeq x_1 + \frac{\vec{x}_2 - \vec{x}_1}{t_2 - t_1} (t - t_1), \quad t \in [t_1, t_2]$$

- Knowns: t₁, t₂, x₁
 - → (x₁ known from, e.g., initial condition)
- Unknown: X,
- Use linear approx. to express $\dot{\vec{x}}(t)$: $\dot{\vec{x}}(t) \simeq \frac{\vec{x}_2 \vec{x}_1}{t_2 t_1}$
- Enforce ODE $\dot{\vec{x}}(t) = \vec{g}(\vec{x},t)$ at t_1 : Forward Euler (FE)

$$\dot{\vec{x}}(t_1) = \vec{g}(\vec{x}_1, t_1) : \boxed{\frac{\vec{x}_2 - \vec{x}_1}{t_2 - t_1} = \vec{g}(\vec{x}_1, t_1)}$$

• Enforce ODE $\dot{\vec{x}}(t) = \vec{g}(\vec{x},t)$ at t_2 : Backward Euler (BE)

$$\dot{\vec{x}}(t_1) = \vec{g}(\vec{x}_2, t_2) : \left[\frac{\vec{x}_2 - \vec{x}_1}{t_2 - t_1} = \vec{g}(\vec{x}_2, t_2) \right]$$

The Forward Euler Method

$$\dot{\vec{x}}(t_{i-1}) = \vec{g}(\vec{x}_{i-1}, t_{i-1}) : \frac{\vec{x}_i - \vec{x}_{i-1}}{t_i - t_{i-1}} = \vec{g}(\vec{x}_{i-1}, t_{i-1})$$

$$\Rightarrow \vec{x}_i = \vec{x}_{i-1} + (t_i - t_{i-1}) \vec{g}(\vec{x}_{i-1}, t_{i-1})$$

- Explicit integration method
 - x₂ available explicitly in terms of knowns
 - → nonlinear solve not needed (single eval of g(.,.) suffices)
 - → (we'll see later) not numerically stable
 - → (we'll see later) does not work for DAEs
- "Time stepping" ODE solution process
 - 1) start with initial condition: $x_0 = x(t_0)$, i=1
 - 2) find x_i using FE equation, above
 - 3) increment i; stop if t_i>stoptime, else goto 2

The Backward Euler Method

$$\dot{\vec{x}}(t_i) = \vec{g}(\vec{x}_i, t_i) : \frac{\vec{x}_i - \vec{x}_{i-1}}{t_i - t_{i-1}} = \vec{g}(\vec{x}_i, t_i) \Rightarrow \boxed{\frac{\vec{x}_i - \vec{x}_{i-1}}{t_i - t_{i-1}} - \vec{g}(\vec{x}_i, t_i) = \vec{0}}$$

- Implicit integration method
 - Finding x₂ requires nonlinear solution (N-R)
 - → (we'll see later) numerically (over)stable
 - → (we'll see later) does work for DAEs
- "Time stepping" ODE solution process
 - 1) start with initial condition: $x_0 = x(t_0)$, i=1
 - 2) find x_i using BE equation, above
 - solve it using Newton-Raphson
 - 3) increment i; stop if t_i>stoptime, else goto 2

The Trapezoidal Method

"Average" FE and BE:

$$\dot{\vec{x}}(t_1) = \vec{g}(\vec{x}_1, t_1) + \dot{\vec{x}}(t_2) = \vec{g}(\vec{x}_2, t_2)$$

$$\frac{\dot{\vec{x}}(t_1) + \dot{\vec{x}}(t_2)}{2} = \frac{\vec{g}(\vec{x}_1, t_1) + \vec{g}(\vec{x}_2, t_2)}{2}$$

- PWL assumption $\Rightarrow \dot{\vec{x}}(t_1) = \dot{\vec{x}}(t_2) = \frac{\vec{x}_2 \vec{x}_1}{t_2 t_1}$
- Trapezoidal Method: $\frac{\vec{x}_2 \vec{x}_1}{t_2 t_1} = \frac{\vec{g}(\vec{x}_1, t_1) + \vec{g}(\vec{x}_2, t_2)}{2}$
 - at ith timestep:

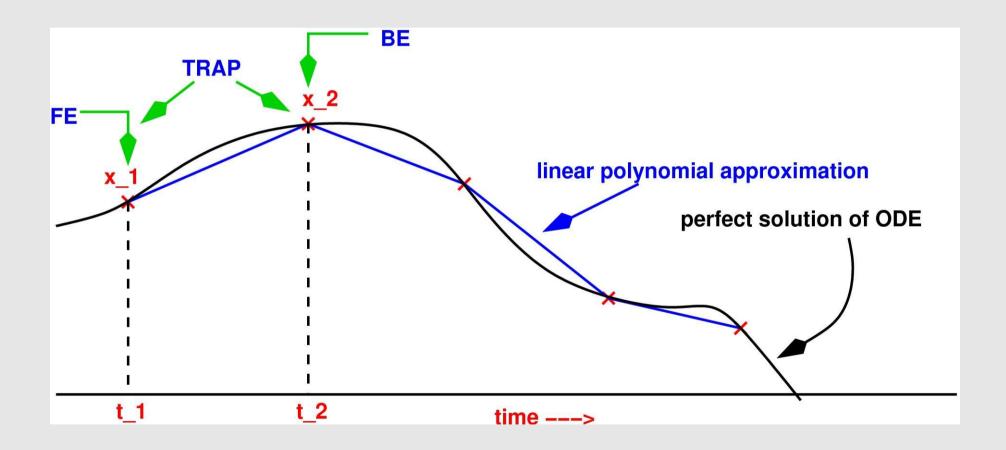
Implicit (N-R needed)

$$\frac{\vec{x}_i - \vec{x}_{i-1}}{t_i - t_{i-1}} = \frac{\vec{g}(\vec{x}_{i-1}, t_{i-1}) + \vec{g}(\vec{x}_i, t_i)}{2}$$

Perfectly stable numerically (we'll see later)

FE, BE, TRAP: Pictorial Summary

Difference: where/how ODE is "enforced"



Linear Multi Step (LMS) Methods

- FE: $\vec{x}_i \vec{x}_{i-1} = h_i \ \vec{g} (\vec{x}_{i-1}, t_{i-1}), \quad h_i \equiv t_i t_{i-1}.$
- BE: $\vec{x}_i \vec{x}_{i-1} = h_i \ \vec{g} \ (\vec{x}_i, t_i)$
- TRAP: $\vec{x}_i \vec{x}_{i-1} = h_i \frac{\vec{g}(\vec{x}_i, t_i) + \vec{g}(\vec{x}_{i-1}, t_{i-1})}{2}$
- Generic: $\alpha_0 \vec{x}_i + \alpha_1 \vec{x}_{i-1} = \beta_0 \vec{g}(\vec{x}_i, t_i) + \beta_1 \vec{g}(\vec{x}_{i-1}, t_{i-1})$
 - FE: $\alpha_0 = 1, \alpha_1 = -1, \beta_0 = 0, \beta_1 = h_i$
 - BE: $\alpha_0 = 1, \alpha_1 = -1, \beta_0 = h_i, \beta_1 = 0$
 - TRAP: $\alpha_0 = 1, \alpha_1 = -1, \beta_0 = \frac{h_i}{2}, \beta_1 = \frac{h_i}{2}$
- pth-order Linear Multi Step integration formula

$$\sum_{i=0}^{p} \alpha_i \, \vec{x}_{n-i} = \sum_{i=0}^{p} \beta_i \, \vec{g} \left(\vec{x}_{n-i}, t_{n-i} \right)$$

Use of ODE LMS Methods for DAEs

$$\frac{d}{dt}\vec{q}(\vec{x}(t)) + \vec{f}(\vec{x}(t)) + \vec{b}(t) = \vec{0} \Rightarrow \frac{d}{dt}\vec{q}(\vec{x}(t)) = \vec{g}(\vec{x},t)$$

- The (simple) idea: re-do all derivations with $\vec{q}(t) \equiv \vec{q}(\vec{x}(t))$ on LHS, instead of $\vec{x}(t)$.
 - Generic LMS for p=1 (1-step LMS):

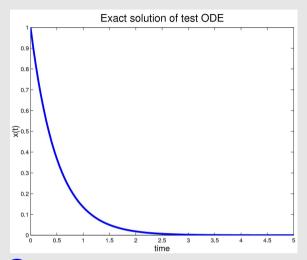
$$\alpha_0 \vec{q}(\vec{x}_i) + \alpha_1 \vec{q}(\vec{x}_{i-1}) = \beta_0 \vec{g}(\vec{x}_i, t_i) + \beta_1 \vec{g}(\vec{x}_{i-1}, t_{i-1})$$

- BE: $\vec{q}(\vec{x}_i) \vec{q}(\vec{x}_{i-1}) = h_i \ \vec{g}(\vec{x}_i, t_i)$ Implicit (N-R needed)
- TRAP: $\vec{q}(\vec{x}_i) \vec{q}(\vec{x}_{i-1}) = h_i \frac{\vec{g}(\vec{x}_i, t_i) + \vec{g}(\vec{x}_{i-1}, t_{i-1})}{2}$
- FE: $\vec{q}(\vec{x}_i) \vec{q}(\vec{x}_{i-1}) = h_i \ \vec{g}(\vec{x}_{i-1}, t_{i-1})$.

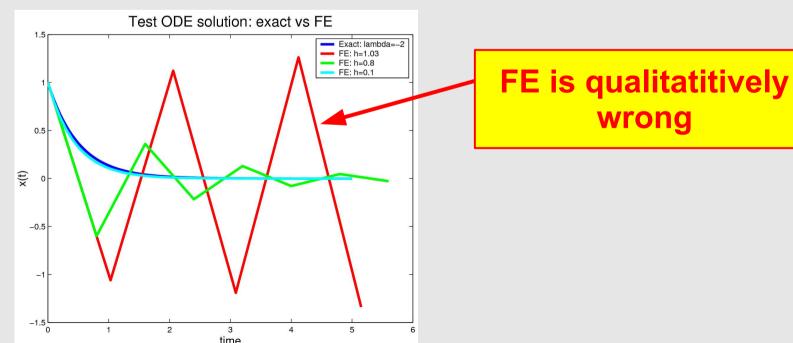
Implicit?

Stability of LMS Methods

- Test ODE: $\dot{x}(t) = \lambda x(t), \quad x(0) = x_0.$
 - exact solution: $x(t) = x_0 e^{\lambda t}$.
 - \rightarrow solution decays if $\lambda < 0$.



What do FE/BE/TRAP produce?



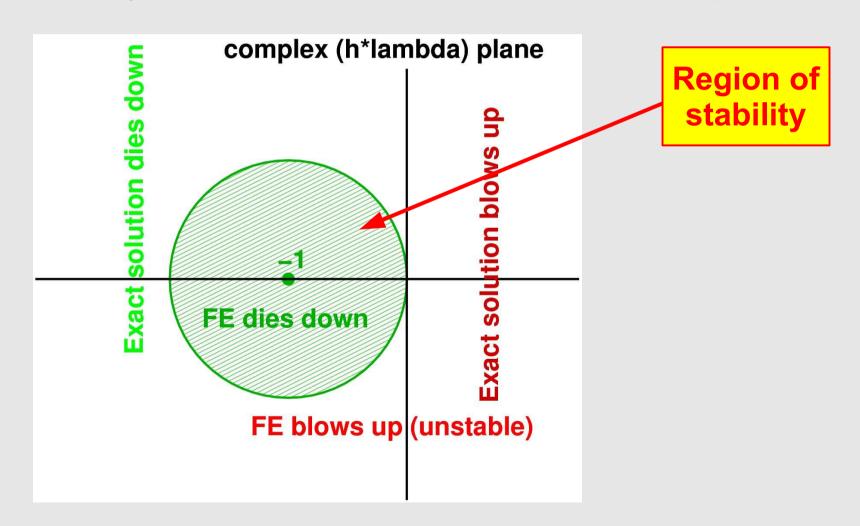
J. Roychowdhury, University of California at Berkeley

Why FE Explodes

- FE (with constant timestep h)
 - $x_n = (1+h\lambda)x_{n-1} \Rightarrow x_n = x_0(1+h\lambda)^n$.
 - if $|1+h\lambda|>1$, solution blows up w.r.t n.
 - \rightarrow solution is **qualitatively wrong** for $\lambda < 0$.
- Basic requirement for FE
 - h should be small enough s.t. $|1 + h\lambda| < 1$.
 - Example: $\lambda = 10^9 \Rightarrow h < 2 \times 10^{-9}$.
 - → FE limited to small timesteps
 - for even "qualitatitive" accuracy
- FE said to be (numerically) unstable
 - if $|1 + h\lambda| > 1$

FE: Stability Picture for Complex λ

- In general: eigenvalues can be complex
 - stability condition $\frac{|1+h\lambda|<1}{}$: circle in $h\lambda$ plane

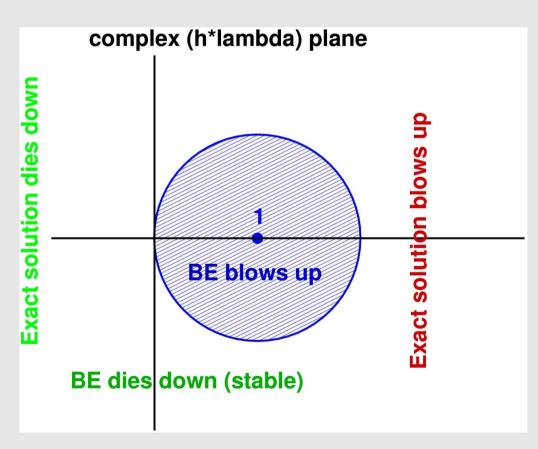


Stability of BE

BE with constant timestep h

•
$$x_n(1-h\lambda) = x_{n-1} \Rightarrow x_n = x_0 \frac{1}{(1-h\lambda)^n}$$
.

- solution will die down if $|1-h\lambda|>1$.
 - → i.e., all of left half plane.
- but also
 - much of right half plane
 - → BE is overstable
 - OK only within circle
- Applications
 - λ <0 more important
 - much better than FE



Stability of TRAP

TRAP with constant timestep h

•
$$x_n\left(1-\frac{h\lambda}{2}\right)=x_{n-1}\left(1+\frac{h\lambda}{2}\right)\Rightarrow x_n=x_0\left(\frac{2+h\lambda}{2-h\lambda}\right)^n$$
.

- solution will die down if Re(λ) < 0
- solution will blow up if Re(λ) > 0
 - → i.e, perfect stability
- Also very accurate
 - (we'll see later)
- Only concern
 - DAE initial consistency
 - → (we'll see later)
 - → (easy practical workaround)

complex (h*lambda) plane			
dies down	TRAP dies down (stable)	TRAP blows up	
Exact solution dies down	TRAP dies	TRAP Exact solution	

Accuracy and Truncation Error

- Even if stable: is the method accurate?
- Test problem exact solution: $x(nh) = x_0 e^{\lambda nh}$.
- How do numerical method solutions compare?
 - FE: $\frac{x(nh) \simeq x_0(1+h\lambda)^n}{n!}$
 - BE: $\frac{x(nh) \simeq x_0 \frac{1}{(1-h\lambda)^n}}$.
 - TRAP: $\frac{x(nh) \simeq x_0 \left(\frac{2+h\lambda}{2-h\lambda}\right)^n}{2}$.
- None are identical, but how different?

Taylor Expansion Error

• Exact:
$$x(nh) = x_0 e^{\lambda nh} = x_0 \left(1 + nh\lambda + \frac{n^2 h^2 \lambda^2}{2!} + \cdots \right)$$

• FE:
$$x(nh) \simeq x_0 \left(1 + nh\lambda + \frac{n(n-1)}{2} h^2 \lambda^2 + \cdots \right)$$

→ first-order accurate

• BE:
$$x(nh) \simeq x_0 \left(1 + nh\lambda + \frac{n(n+1)}{2}h^2\lambda^2 + \cdots \right)$$

→ also first-order accurate

• TRAP:
$$x(nh) \simeq x_0 \left(1 + nh\lambda + \frac{n^2}{2}h^2\lambda^2 + \cdots \right)$$

→ second-order accurate

Transient: Timestep Control

- Choosing the next timestep dynamically
- LTE based control
 - apply LTE formulae to estimate error
 - change timestep to meet some specified error
 - error specification: like reltol-abstol (reltol=percentage)
 - make sure these are looser than NR tolerances!
 - element-by-element vs vector norm based
 - → 2-norm vs max norm; DAE issues
 - (change integration method based on timestep)
- NR convergence based control
 - cut timestep if NR does not converge
 - → also: increase maxiter
 - increase timestep if NR converges "too easily"
 - → also: decrease maxiter

ODE/DAE Packages Out There

- MATLAB has various ODE/DAE integrators
 - ode23, ode45, ...; ode23t, ode15s, ...
- DASSL/DASPK: general purpose DAE packages
 - Linda Petzold, UCSB
 - Fortran
 - some tweaking helpful for circuit applications
- Easy (and often worthwhile) to roll your own
 - tweaking, special heuristics, debugging, ...

Transient: Other Important Issues

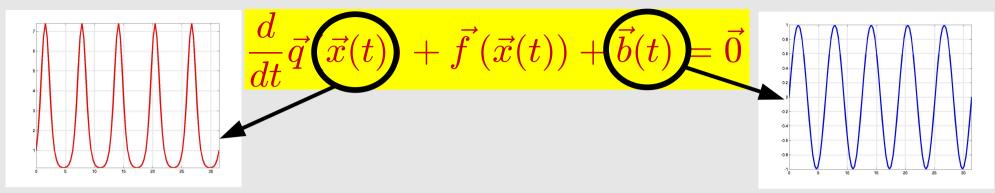
- What integration method to use?
 - stability?
 - nonlinear solution required? (implicit vs explicit)
 - accuracy loss due to discretization?
 - → Local Truncation Error (LTE)
 - higher order methods (more than 1 previous timepoint)
- Vast body of work on ODE integration
 - linear multi-step methods (LMS), Runge-Kutta, symmetric, symplectic, "energy-conserving", etc.
- Stiff differential equations
 - different variables have very different time constants
 - stiffly stable methods allow you to take larger time steps
- DAE issues
 - initial condition consistency; stability; index; ...
- Dynamic timestep control, NR heuristics, ...

Solving the Circuit's Equations

$$\frac{d}{dt}\vec{q}(\vec{x}(t)) + \vec{f}(\vec{x}(t)) + \vec{b}(t) = \vec{0}$$

- Ckt. DAEs: many types of solutions useful
 - ✓ DC steady state: no time variations
 - ✓ <u>transient</u>: ckt. waveforms changing with time
 - periodic steady state: changes periodic w time
 - → linear(ized): all sinusoidal waveforms: AC analysis
 - nonlinear steady state: <u>shooting</u>, harmonic balance
 - noise analysis: random/stochastic waveforms
 - sensitivity analysis: effects of changes in circuit parameters

The Periodic Steady State Problem



- What
 - inputs b(t) are periodic e.g., sinusoidal
 - suppose "outputs" x(t) also become periodic
 - → (happens for "stable" circuits and systems ...
 - ... eventually can take a long time)
 - want to find this <u>periodic steady state</u> directly
 - without using general/expensive transient analysis
- Why
 - → audio amps, RF amps, mixers, oscillators, clocks, ...
 - ckt nonlinear => sinusoids will in general be distorted
 - → linear circuits: frequency-domain analysis important
 - much easier and more insightful than transient
- How (for linear circuits): AC analysis

