The z-Transform
(ch.10)

Properties of the z-transform

Linearity

$$x_1[n] \xleftrightarrow{\mathcal{Z}} X_1(z) \quad \mathrm{ROC} = R1$$

$$\Longrightarrow \quad \boxed{a x_1[n] + b x_2[n] \xleftrightarrow{\mathcal{Z}} a X_1(z) + b X_2(z)}$$

with ROC containing $R_1 \cap R_2$

$$x_2[n] \xleftrightarrow{\mathcal{Z}} X_2(z) \quad \mathrm{ROC} = R2$$

Time shifting

$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \quad \mathrm{ROC} = R$$

$$\Downarrow$$

$$x[n-n_0] \xleftrightarrow{\mathcal{Z}} z^{-n_0} X(z)$$

$\mathrm{ROC} = R$ except for the possible addition or deletion of the origin or infinity

Properties of the z-transform

Scaling in the z-domain

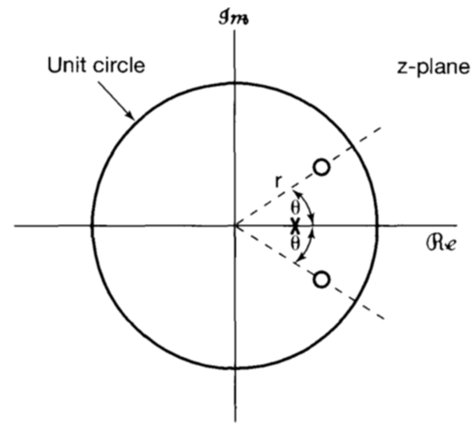
$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \qquad \text{ROC} = R$$

$$\Downarrow$$

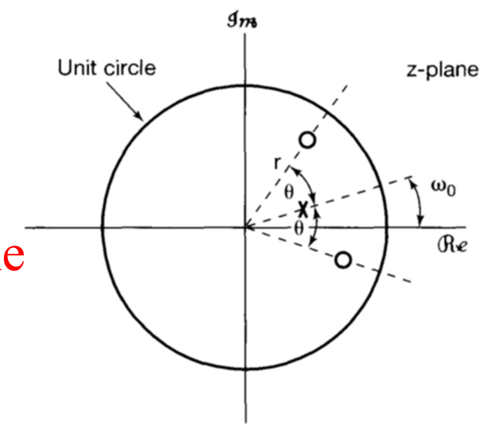
$$z_0^n x[n] \xleftrightarrow{\mathcal{Z}} X(z/z_0) \qquad \text{ROC} = |z_0|R$$

$$\Downarrow \; z_0 = e^{j\omega_0}$$

$$e^{j\omega_0 n} x[n] \xleftrightarrow{\mathcal{Z}} X\left(e^{-j\omega_0} z\right) \qquad \text{ROC} = R$$



$\Downarrow$



Multiplication by $e^{j\omega_0 n}$ $\Longleftrightarrow$ Rotation by ω_0 in the Z-plane

Properties of the z-transform

Time reversal

$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \qquad \text{ROC} = R$$

$$\Downarrow$$

$$x[-n] \xleftrightarrow{\mathcal{Z}} X\left(\frac{1}{z}\right) \qquad \text{ROC} = \frac{1}{R}$$

Time expansion

$$x_{(k)}[n] = \begin{cases} x[n/k] & \text{if } n \text{ is a multiple of } k \\ 0 & \text{if } n \text{ is not a multiple of } k \end{cases}$$

$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \qquad \text{ROC} = R$$

$$\Downarrow$$

$$x_{(k)}[n] \xleftrightarrow{\mathcal{Z}} X(z^k) \qquad \text{ROC} = R^{1/k}$$

$$X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$$

$$\Downarrow$$

$$X(z^k) = \sum_{n=-\infty}^{+\infty} x[n]z^{-kn}$$

Properties of the z-transform

Conjugation

$$x[n] \overset{\mathcal{Z}}{\longleftrightarrow} X(z) \qquad \mathrm{ROC} = R$$

$$\Downarrow$$

$$x^*[n] \overset{\mathcal{Z}}{\longleftrightarrow} X^*(z^*) \qquad \mathrm{ROC} = R$$

Convolution

$$x_1[n] \overset{\mathcal{Z}}{\longleftrightarrow} X_1(z) \qquad \mathrm{ROC} = R_1$$

$$x_2[n] \overset{\mathcal{Z}}{\longleftrightarrow} X_2(z) \qquad \mathrm{ROC} = R_2$$

$$\Longrightarrow \quad x_1[n]*x_2[n] \overset{\mathcal{Z}}{\longleftrightarrow} X_1(z)X_2(z)$$

with ROC containing $R_1 \cap R_2$

Properties of the z-transform

First-difference

$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \qquad \text{ROC} = R$$

$$x[n] - x[n-1] \xleftrightarrow{\mathcal{Z}} (1 - z^{-1})X(z)$$ ROC $= R$, possible deletion of $z = 0$ and/or addition of $z = 1$

Accumulation

$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \qquad \text{ROC} = R$$

$$w[n] = \sum_{k=-\infty}^{n} x[k] \xleftrightarrow{\mathcal{Z}} \frac{1}{(1 - z^{-1})} X(z)$$ ROC $= R$, possible deletion of $z = 1$ and/or addition of $z = 0$

Properties of the z-transform

Differentiation in the z-domain

$$x[n] \xleftrightarrow{\mathcal{Z}} X(z) \qquad \mathrm{ROC} = R$$

$$\Downarrow$$

$$nx[n] \xleftrightarrow{\mathcal{Z}} -\mathrm{z}\frac{dX(z)}{dz} \qquad \mathrm{ROC} = R$$

Properties of the z-transform

Examples

$$X(z) = \log(1 + az^{-1}) \qquad |z| > |a| \qquad x[n] = ?$$

Solution

$$nx[n] \overset{\mathcal{Z}}{\longleftrightarrow} -z\frac{dX(z)}{dz} = \frac{az^{-1}}{1 + az^{-1}} \qquad |z| > |a|$$

$$a(-a)^n u[n] \overset{\mathcal{Z}}{\longleftrightarrow} \frac{a}{1 + az^{-1}} \qquad |z| > |a|$$

$$a(-a)^{n-1} u[n-1] \overset{\mathcal{Z}}{\longleftrightarrow} \frac{az^{-1}}{1 + az^{-1}} \qquad |z| > |a|$$

$$x[n] = -\frac{(-a)^{n-1}}{n} u[n-1]$$

Examples

$$X(z) = \frac{az^{-1}}{(1 - az^{-1})^2} \qquad |z| > |a| \qquad x[n] = ?$$

Solution

$$a^n u[n] \overset{\mathcal{Z}}{\longleftrightarrow} \frac{1}{1 - az^{-1}} \qquad |z| > |a|$$

$$na^n u[n] \overset{\mathcal{Z}}{\longleftrightarrow} -z\frac{d}{dz}\left(\frac{1}{1 - az^{-1}}\right) = \frac{az^{-1}}{(1 - az^{-1})^2} \qquad |z| > |a|$$

Properties of the z-transform

The initial-value theorem

If

$$x[n] = 0 \text{ for } n < 0,$$

Then,

$$x(0) = \lim_{z \to \infty} X(z)$$

$$X(z) = \sum_{n=0}^{\infty} x[n] z^{-n}$$

For $n > 0$, $z \to \infty \implies z^{-n} \to 0$

For $n = 0$, $z^{-n} = 1$

❑ Examples

$$x[n] = 7\left(\frac{1}{3}\right)^n u[n] - 6\left(\frac{1}{2}\right)^n u[n]$$

$$X(z) = \frac{7}{1 - \frac{1}{3}z^{-1}} - \frac{6}{1 - \frac{1}{2}z^{-1}}$$

$\implies$

$$x(0) = 1$$

$$\lim_{z \to \infty} X(z) = 1$$

Properties of the z-transform

Summary

Section	Property	Signal	z-Transform	ROC
		$x[n]$	$X(z)$	R
		$x_1[n]$	$X_1(z)$	R_1
		$x_2[n]$	$X_2(z)$	R_2
10.5.1	Linearity	$ax_1[n] + bx_2[n]$	$aX_1(z) + bX_2(z)$	At least the intersection of R_1 and R_2
10.5.2	Time shifting	$x[n - n_0]$	$z^{-n_0}X(z)$	R, except for the possible addition or deletion of the origin
10.5.3	Scaling in the z-domain	$e^{j\omega_0 n}x[n]$	$X(e^{-j\omega_0}z)$	R
		$z_0^n x[n]$	$X\left(\frac{z}{z_0}\right)$	z_0R
		$a^n x[n]$	$X(a^{-1}z)$	Scaled version of R (i.e., $\|a\|R$ = the set of points $\{\|a\|z\}$ for z in R)
10.5.4	Time reversal	$x[-n]$	$X(z^{-1})$	Inverted R (i.e., R^{-1} = the set of points z^{-1}, where z is in R)
10.5.5	Time expansion	$x_{(k)}[n] = \begin{cases} x[r], & n = rk \\ 0, & n \neq rk \end{cases}$ for some integer r	$X(z^k)$	$R^{1/k}$ (i.e., the set of points $z^{1/k}$, where z is in R)
10.5.6	Conjugation	$x^*[n]$	$X^*(z^*)$	R
10.5.7	Convolution	$x_1[n] * x_2[n]$	$X_1(z)X_2(z)$	At least the intersection of R_1 and R_2
10.5.7	First difference	$x[n] - x[n-1]$	$(1 - z^{-1})X(z)$	At least the intersection of R and $\|z\| > 0$
10.5.7	Accumulation	$\sum_{k=-\infty}^{n} x[k]$	$\frac{1}{1 - z^{-1}}X(z)$	At least the intersection of R and $\|z\| > 1$
10.5.8	Differentiation in the z-domain	$nx[n]$	$-z\frac{dX(z)}{dz}$	R
10.5.9		Initial Value Theorem If $x[n] = 0$ for $n < 0$, then $x[0] = \lim_{z\to\infty} X(z)$		

The z-Transform
(ch.10)

- ❑ The z-transform
- ❑ The region of convergence for the z-transforms
- ❑ The inverse z-transform
- ❑ Geometric evaluation of the Fourier transform from the pole-zero plot
- ❑ Properties of the z-transform
- ❑ Some common z-transform pairs
- ❑ Analysis and characterization of LTI systems using z-transforms
- ❑ System function algebra and block diagram representations
- ❑ The unilateral z-transform

Some z-transform pairs

Signal	Transform	ROC
1. $\delta[n]$	1	All z
2. $u[n]$	$\frac{1}{1-z^{-1}}$	$\lvert z\rvert > 1$
3. $-u[-n-1]$	$\frac{1}{1-z^{-1}}$	$\lvert z\rvert < 1$
4. $\delta[n-m]$	z^{-m}	All z, except 0 (if $m > 0$) or ∞ (if $m < 0$)
5. $\alpha^n u[n]$	$\frac{1}{1-\alpha z^{-1}}$	$\lvert z\rvert > \lvert\alpha\rvert$
6. $-\alpha^n u[-n-1]$	$\frac{1}{1-\alpha z^{-1}}$	$\lvert z\rvert < \lvert\alpha\rvert$
7. $n\alpha^n u[n]$	$\frac{\alpha z^{-1}}{(1-\alpha z^{-1})^2}$	$\lvert z\rvert > \lvert\alpha\rvert$
8. $-n\alpha^n u[-n-1]$	$\frac{\alpha z^{-1}}{(1-\alpha z^{-1})^2}$	$\lvert z\rvert < \lvert\alpha\rvert$
9. $[\cos\omega_0 n]u[n]$	$\frac{1-[\cos\omega_0]z^{-1}}{1-[2\cos\omega_0]z^{-1}+z^{-2}}$	$\lvert z\rvert > 1$
10. $[\sin\omega_0 n]u[n]$	$\frac{[\sin\omega_0]z^{-1}}{1-[2\cos\omega_0]z^{-1}+z^{-2}}$	$\lvert z\rvert > 1$
11. $[r^n\cos\omega_0 n]u[n]$	$\frac{1-[r\cos\omega_0]z^{-1}}{1-[2r\cos\omega_0]z^{-1}+r^2z^{-2}}$	$\lvert z\rvert > r$
12. $[r^n\sin\omega_0 n]u[n]$	$\frac{[r\sin\omega_0]z^{-1}}{1-[2r\cos\omega_0]z^{-1}+r^2z^{-2}}$	$\lvert z\rvert > r$

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Causality

Causal $\Longleftrightarrow$ ROC of $H(z)$ is the exterior of a circle, including infinity

A system with rational $H(z)$ is causal $\Longleftrightarrow$

- ROC is the exterior of a circle outside the outermost pole;
- With $H(z)$ expressed as a ratio of polynomials in z, the order of the numerator cannot be greater than the order of the denominator.

Examples

$$H(z) = \frac{z^3 - 2z^2 + z}{z^2 + \frac{1}{4}z + \frac{1}{8}}$$ Noncausal

Examples

$$H(z) = \frac{1}{1 - \frac{1}{2}z^{-1}} + \frac{1}{1 - 2z^{-1}} \qquad |z| > 2$$

Solution 1

$|z| > 2$: ROC is the exterior of a circle outside the outermost pole.

$$H(z) = \frac{2 - \frac{5}{2}z^{-1}}{\left(1 - \frac{1}{2}z^{-1}\right)(1 - 2z^{-1})} = \frac{2z^2 - \frac{5}{2}z}{z^2 - \frac{5}{2}z + 1}$$

$\Longrightarrow$ Causal

Solution 2

$h[n] = [(1/2)^n + 2^n]u[n]$ $\Longrightarrow$ $h[n] = 0$ for $n < 0$ $\Longrightarrow$ Causal

Stability

For an LTI system,

Stable $\Leftrightarrow$ The ROC of $H(z)$ includes the unit circle, $|z| = 1$

❑ Examples

$$H(z) = \frac{1}{1 - \frac{1}{2}z^{-1}} + \frac{1}{1 - 2z^{-1}}$$

ROC	Causal	Stable
$\lvert z\rvert > 2$	Yes	No
$1/2 < \lvert z\rvert < 2$	No	Yes
$\lvert z\rvert < 1/2$	No	No

Stability

For a causal LTI system with rational system function $H(z)$,

Stable $\Leftrightarrow$ All of the poles of $H(z)$ lie inside the unit circle. (magnitude smaller than 1)

❑ Examples

$$H(z) = \frac{1}{1 - az^{-1}} \text{ is stable} \quad \Longrightarrow \quad |a| < 1$$

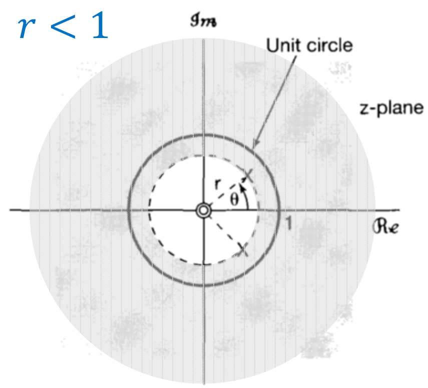
❑ Examples

$$H(z) = \frac{1}{1 - (2r\cos\theta)z^{-1} + r^2 z^{-2}}$$

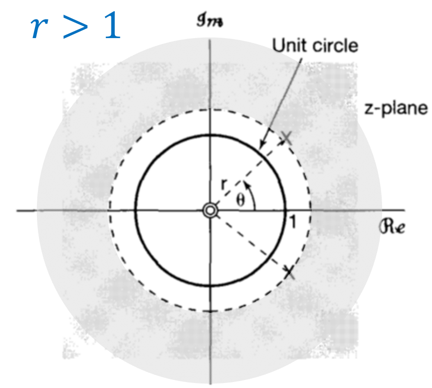
Poles: $z_1 = re^{j\theta}$ $\quad z_2 = re^{-j\theta}$

Stable $\Longrightarrow$ $r < 1$

$r < 1$



$r > 1$



LTI systems characterized by linear constant-coefficient difference equations

❑ Examples

$$y[n]-\frac{1}{2}y[n-1]=x[n]+\frac{1}{3}x[n-1]$$

$$Y(z)-\frac{1}{2}z^{-1}Y(z)=X(z)+\frac{1}{3}z^{-1}X(z)$$

$$Y(z)=X(z)\left[\frac{1+\frac{1}{3}z^{-1}}{1-\frac{1}{2}z^{-1}}\right]$$

$$H(z)=\frac{Y(z)}{X(z)}=\left[\frac{1+\frac{1}{3}z^{-1}}{1-\frac{1}{2}z^{-1}}\right]\Longrightarrow\begin{cases}|z|>\frac{1}{2} & h[n]=\left(\frac{1}{2}\right)^{n}u[n]+\frac{1}{3}\left(\frac{1}{2}\right)^{n-1}u[n-1]\\[2ex] |z|<\frac{1}{2} & h[n]=-\left(\frac{1}{2}\right)^{n}u[-n-1]-\frac{1}{3}\left(\frac{1}{2}\right)^{n-1}u[-n]\end{cases}$$

LTI systems characterized by linear constant-coefficient difference equations

❑ In general

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k]$$

$$Y(z) \sum_{k=0}^{N} a_k z^{-k} = X(s) \sum_{k=0}^{M} b_k z^{-k}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^{M} b_k z^{-k}}{\sum_{k=0}^{N} a_k z^{-k}} \quad \Longrightarrow \quad \begin{cases} \text{Poles at the solution of} & \sum_{k=0}^{N} a_k z^{-k} = 0 \\ \text{Zeros at the solution of} & \sum_{k=0}^{M} b_k z^{-k} = 0 \end{cases}$$

Examples relating system behavior to the system function

Given the following information about an LTI system, $H(z) = ?\ \ h[n] = ?$

- If $x_1[n] = (1/6)^n u[n]$, then $y_1[n] = \left[a\left(\frac{1}{2}\right)^n + 10\left(\frac{1}{3}\right)^n\right]u[n]$
- If $x_2[n] = (-1)^n$, then $y_2[n] = \frac{7}{4}(-1)^n$

Solution

$$X_1(z) = \frac{1}{1 - \frac{1}{6}z^{-1}}, |z| > \frac{1}{6}$$

$$Y_1(z) = \frac{a}{1 - \frac{1}{2}z^{-1}} + \frac{10}{1 - \frac{1}{3}z^{-1}} = \frac{(a+10) - \left(5 + \frac{a}{3}\right)z^{-1}}{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{3}z^{-1}\right)}, |z| > \frac{1}{2}$$

$$H(z) = \frac{Y_1(z)}{X_1(z)} = \frac{\left[(a+10) - \left(5 + \frac{a}{3}\right)z^{-1}\right]\left(1 - \frac{1}{6}z^{-1}\right)}{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{3}z^{-1}\right)},$$

Examples relating system behavior to the system function

Solution continue

$$\frac{7}{4} = H(-1) = \frac{\left[(a+10) + \left(5 + \frac{a}{3}\right)\right]\left(\frac{7}{6}\right)}{\left(\frac{3}{2}\right)\left(\frac{4}{3}\right)} \quad \Longrightarrow \quad a = -9$$

$$H(z) = \frac{(1-2z^{-1})\left(1 - \frac{1}{6}z^{-1}\right)}{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{3}z^{-1}\right)}$$

ROC of $X_1(z)$: $|z| > \frac{1}{6}$ $\quad \Longrightarrow \quad$ ROC of $H(z)$: $|z| > \frac{1}{2}$

Examples relating system behavior to the system function

Consider a stable and causal system with impulse response $h[n]$ and rational system function $H(z)$, which contains a pole at $z = 1/2$ and a zero somewhere on the unit circle.

- ❑ $\mathcal{F}\{(1/2)^n h(t)\}$ converges. True
- ❑ $H(e^{j\omega}) = 0$ for some ω True
- ❑ $h[n]$ has finite duration False
- ❑ $h[n]$ is real Insufficient information
- ❑ $g[n] = n\big[h[n] * h[n]\big]$ is the impulse response of a stable system True

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(ch.10)

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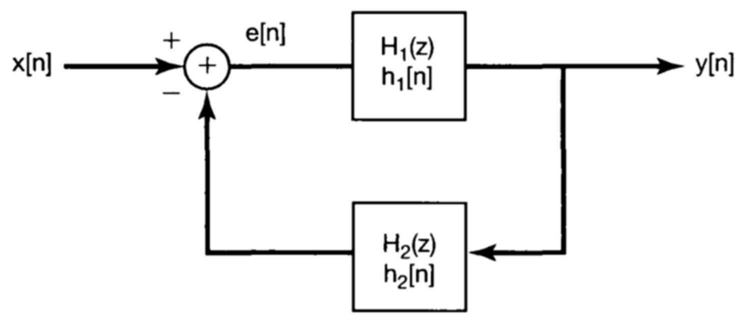
System functions for interconnections of LTI systems

$$\frac{Y(z)}{X(z)} = H(z) = \frac{H_1(z)}{1 + H_1(z)H_2(z)}$$

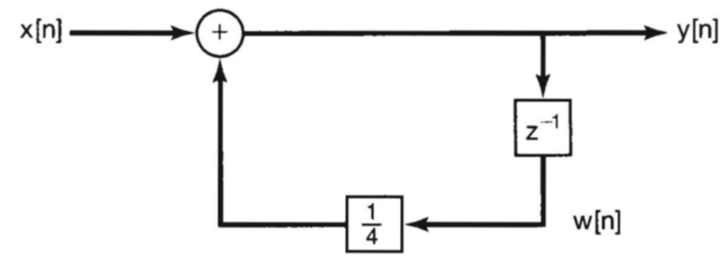


Block diagram representations for causal LTI systems

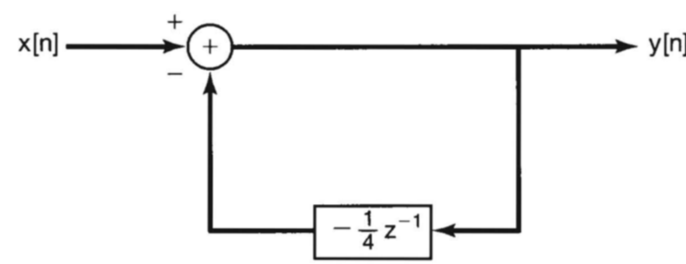
$$H(s) = \frac{1}{1 - \frac{1}{4}z^{-1}}$$

$$y[n] - \frac{1}{4}y[n-1] = x[n]$$

$$w[n] = y[n-1]$$



Or equivalently

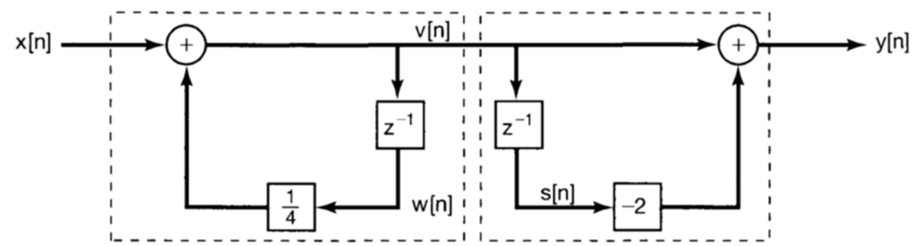


Examples: block diagram representations for causal LTI systems

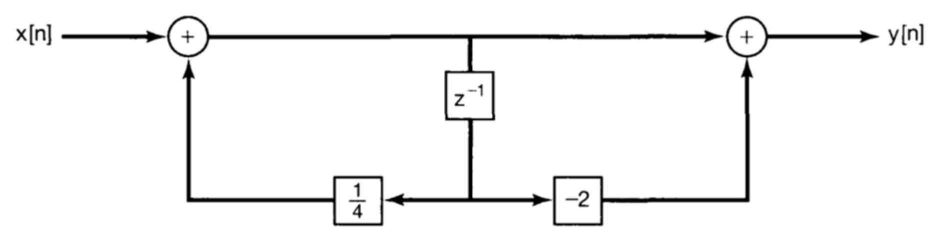
$$H(z) = \frac{1 - 2z^{-1}}{1 - \frac{1}{4}z^{-1}} = \left(\frac{1}{1 - \frac{1}{4}z^{-1}}\right)(1 - 2z^{-1})$$

$$y[n] = v[n] - 2v[n-1]$$

$$w[n] = s[n] = v[n-1]$$



Or equivalently



System function algebra and block diagram representations

Examples: block diagram representations for causal LTI systems

$$H(z) = \frac{1}{1+\frac{1}{4}z^{-1}-\frac{1}{8}z^{-2}} = \frac{1}{\left(1-\frac{1}{2}z^{-1}\right)}\frac{1}{\left(1-\frac{1}{4}z^{-1}\right)} = \frac{2/3}{\left(1-\frac{1}{2}z^{-1}\right)} + \frac{1/3}{\left(1-\frac{1}{4}z^{-1}\right)}$$

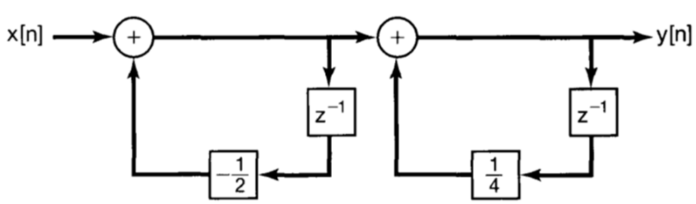
$$y[n] + \frac{1}{4}y[n-1] - \frac{1}{8}y[n-2] = x[n]$$

$$y[n] = -\frac{1}{4}y[n-1] + \frac{1}{8}y[n-2] + x[n]$$

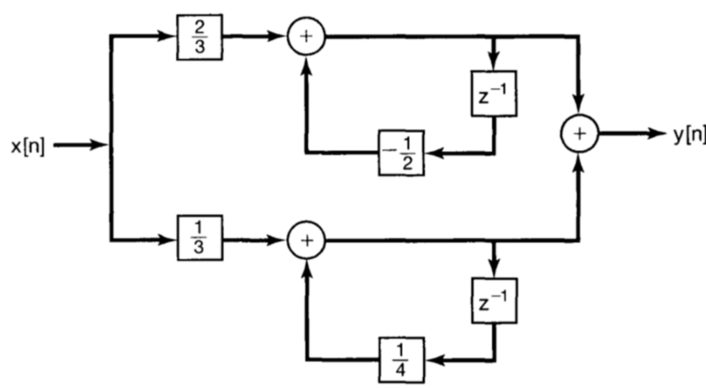
Direct form

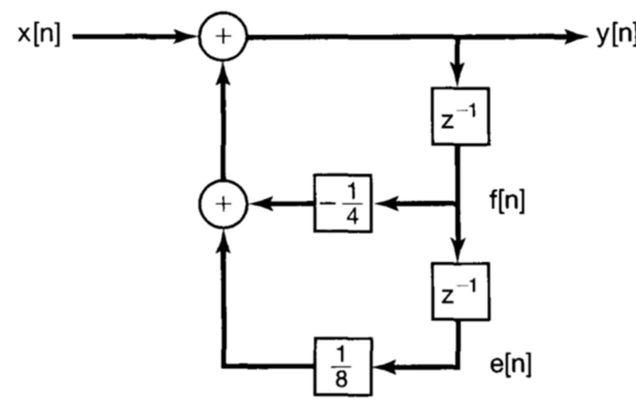
$f[n] = y[n-1]$
$e[n] = f[n-1] = y[n]$

Cascade form



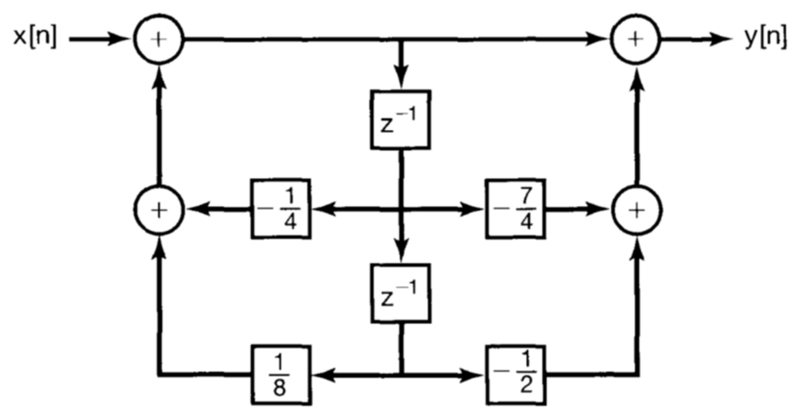
Parallel form





Examples: block diagram representations for causal LTI systems

$$H(z) = \frac{1 - \frac{7}{4}z^{-1} - \frac{1}{2}z^{-2}}{1 + \frac{1}{4}z^{-1} - \frac{1}{8}z^{-2}} = \frac{1}{1 + \frac{1}{4}z^{-1} - \frac{1}{8}z^{-2}}\left(1 - \frac{7}{4}z^{-1} - \frac{1}{2}z^{-2}\right)$$



The Laplace Transform
(ch.9)

- ❑ The Laplace transform
- ❑ The region of convergence for Laplace transforms
- ❑ The inverse Laplace transform
- ❑ Geometric evaluation of the Fourier transform from the pole-zero plot
- ❑ Properties of the Laplace transform
- ❑ Some Laplace transform pairs
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- ❑ System function algebra and block diagram representations
- ❑ The unilateral Laplace transform

The unilateral Laplace transform

$$x[n] \xleftrightarrow{\mathcal{UZ}} \mathcal{X}(z) = \mathcal{U}\mathfrak{L}\{x[n]\}$$

$$\mathcal{X}(z) \triangleq \sum_{n=0}^{\infty} x[n] z^{-n}$$

Examples

$$x[n] = a^n u[n]$$

$x[n] = 0$, for $n < 0$

$$\mathcal{X}(z) = \frac{1}{1 - az^{-1}}, \qquad |z| > |a|$$

The unilateral Laplace transform

Examples

$$x[n] = a^{n+1}u[n+1]$$

$$X(z) = \frac{z}{1 - az^{-1}}, \qquad |z| > |a|$$

$$\mathcal{X}(z) = \sum_{n=0}^{\infty} a^{n+1} z^{-n} = \frac{a}{1 - az^{-1}}, \qquad |z| > |a|$$

Not equal
$(x[-1] \neq 0)$

Examples

$$\mathcal{X}(z)=\frac{3-\frac{5}{6}z^{-1}}{\left(1-\frac{1}{4}z^{-1}\right)\left(1-\frac{1}{3}z^{-1}\right)}$$

Solution

$$\mathcal{X}(z)=\frac{1}{1-\frac{1}{4}z^{-1}}+\frac{2}{1-\frac{1}{3}z^{-1}}$$

$$x[n]=x_1[n]+x_2[n]$$

$$x_1[n]\xleftrightarrow{\mathcal{Z}}\frac{1}{1-\frac{1}{4}z^{-1}}\qquad |z|>\frac{1}{4}$$

$$x_2[n]\xleftrightarrow{\mathcal{Z}}\frac{2}{1-\frac{1}{3}z^{-1}}\qquad |z|>\frac{1}{3}$$

$$\Longrightarrow\quad x[n]=\left(\frac{1}{4}\right)^n u[n]+2\left(\frac{1}{3}\right)^n u[n],\qquad n\geq 0$$

Properties of the unilateral Laplace transform

Property	Signal	Unilateral z-Transform
—	$x[n]$	$\mathcal{X}(z)$
—	$x_1[n]$	$\mathcal{X}_1(z)$
—	$x_2[n]$	$\mathcal{X}_2(z)$
Linearity	$ax_1[n] + bx_2[n]$	$a\mathcal{X}_1(z) + b\mathcal{X}_2(z)$
Time delay	$x[n-1]$	$z^{-1}\mathcal{X}(z) + x[-1]$
Time advance	$x[n+1]$	$z\mathcal{X}(z) - zx[0]$
Scaling in the z-domain	$e^{j\omega_0 n}x[n]$ $z_0^n x[n]$ $a^n x[n]$	$\mathcal{X}(e^{-j\omega_0}z)$ $\mathcal{X}(z/z_0)$ $\mathcal{X}(a^{-1}z)$
Time expansion	$x_k[n] = \begin{cases} x[m], & n = mk \\ 0, & n \neq mk \end{cases}$ for any m	$\mathcal{X}(z^k)$
Conjugation	$x^*[n]$	$\mathcal{X}^*(z^*)$
Convolution (assuming that $x_1[n]$ and $x_2[n]$ are identically zero for $n < 0$)	$x_1[n] * x_2[n]$	$\mathcal{X}_1(z)\mathcal{X}_2(z)$
First difference	$x[n] - x[n-1]$	$(1 - z^{-1})\mathcal{X}(z) - x[-1]$
Accumulation	$\sum_{k=0}^{n} x[k]$	$\frac{1}{1 - z^{-1}}\mathcal{X}(z)$
Differentiation in the z-domain	$nx[n]$	$-z\frac{d\mathcal{X}(z)}{dz}$

Initial Value Theorem

$$x[0] = \lim_{z \to \infty} \mathcal{X}(z)$$

Convolution Examples

A causal LTI system

$$y[n] + 3y[n-1] = x[n] \qquad x[n] = \alpha u[n] \qquad y[n] =?$$

Solution

$$\mathcal{H}(z) = \frac{1}{1+3z^{-1}}$$

$$\mathcal{Y}(z) = \mathcal{H}(z)\mathcal{X}(z) = \frac{\alpha}{(1+3z^{-1})(1-z^{-1})} = \frac{(3/4)\alpha}{1+3z^{-1}} + \frac{(1/4)\alpha}{1-z^{-1}}$$

$$y[n] = \alpha\left[\frac{1}{4} + \left(\frac{3}{4}\right)(-3)^n\right]u[n]$$

Shifting property

$$x[n+1] \overset{\mathcal{UZ}}{\longleftrightarrow} z\mathcal{X}(z) - zx[0]$$

$$x[n-1] \overset{\mathcal{UZ}}{\longleftrightarrow} z^{-1}\mathcal{X}(z) + x[-1]$$

$$x[n-2] \overset{\mathcal{UZ}}{\longleftrightarrow} z^{-2}\mathcal{X}(z) + z^{-1}x[-1] + x[-2]$$

Consider $y[n] = x[n-1]$:

$$\mathcal{Y}(z) = \sum_{n=0}^{\infty} x[n-1]z^{-n}$$

$$= x[-1] + \sum_{n=1}^{\infty} x[n-1]z^{-n}$$

$$= x[-1] + \sum_{n=0}^{\infty} x[n]z^{-(n+1)}$$

$$= x[-1] + z^{-1}\mathcal{X}(z)$$

Solving differential equations using the unilateral z-transform

$y[n] + 3y[n-1] = x[n] \qquad x[n] = \alpha u[n] \qquad y[-1] = \beta$

$y[n] = ?$

Solution

$$\mathcal{Y}(z) + 3\beta + 3z^{-1}\mathcal{Y}(z) = \frac{\alpha}{1 - z^{-1}}$$

$$\mathcal{Y}(z) = \boxed{\frac{3\beta}{1 + 3z^{-1}}} + \boxed{\frac{\alpha}{(1 + 3z^{-1})(1 - z^{-1})}}$$

Zero-input response

Zero-state response

If $\alpha = 8$, $\beta = 1$, $y[n] = [3(-3)^n + 2]u[n]$, for $n \geq 0$