EE 111 Homework 7

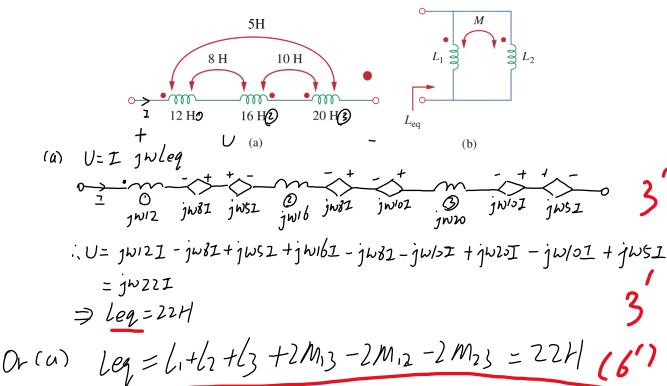
Due date: May. 17th,2019
Turn in your homework in class

Rules:

- Work on your own. Discussion is permissible, but similar submissions will be judged as plagiarism.
- Please show all intermediate steps: a correct solution without an explanation will get zero credit.
- Please submit on time. No late submission will be accepted.
- Please prepare your submission in English only. No Chinese submission will be accepted.

- 1. (a) For the three coupled coils in Fig. (a), calculate the total inductance.
 - (b) For the coupled coils in Fig. (b), show that

$$L_{\rm eq} = \frac{L_1 L_2 - M^2}{L_1 + L_2 - 2M}$$



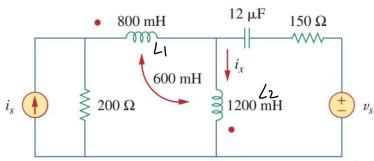
$$= jwl_{1}I_{1}+jwmI_{2}$$

$$= jw\left(\frac{l_{1}(l_{2}-m)}{l_{1}+l_{1}-lm}+\frac{m(l_{1}-m)}{l_{1}+l_{2}-lm}\right)I = jw\frac{l_{1}l_{2}-m^{2}}{l_{1}+l_{2}-lm}I$$

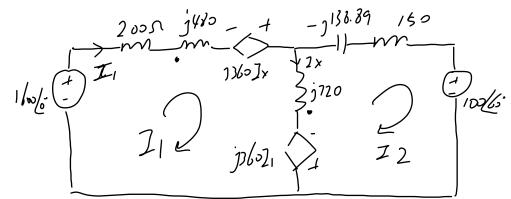
$$= jw\left(\frac{l_{1}(l_{2}-m)}{l_{1}+l_{2}-lm}+\frac{m(l_{1}-m)}{l_{1}+l_{2}-lm}\right)I = jw\frac{l_{1}l_{2}-m^{2}}{l_{1}+l_{2}-lm}I$$

$$= jwl_{1}I_{2}-m^{2}I_{1}I_{2}-m^{2}I_{2}I_{2}-m^{2}I_{2}I_{2}-m^{2}I_{2}-$$

- 2. With $\dot{t}_s = 8\cos(600t)$ A and $V_s = 100\cos(600t + 60^\circ)V$,
 - (a) find the coupling coefficient,
 - (b) use mesh analysis to find i_x ,
 - (c) determine the energy stored in the coupled inductors at t = 2 s.



(a)
$$K = \frac{M}{\sqrt{11/12}} = \frac{\sqrt{6}}{4} = 0.6124 \approx 0.61$$
 2



$$= \begin{cases} -1600 + (200 + j480)21 - j36021 = 0 \\ (j720 - j18889 + 150)21 - j36021 + 10068 = 0 \end{cases} = \begin{cases} 21 = 2.80 - 3.22j = 4.27 (-48.99)^{\circ} \\ 22 = 1.95 - 1.40j = 2.40 (-35.68)^{\circ} \end{cases}$$

$$1.1x = 1-12 = 0.85 - 1.829 = 2.01 / -64.97° A$$

$$1.1x = 2.01 \cos(600t - 64.97°) A$$

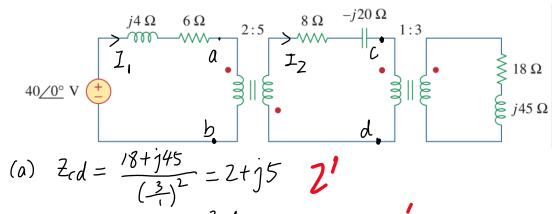
$$i_{1} = 2.01 \ cos(600t - 64.97)A$$

$$(c) \ t=25. \ i_{x} = 2.01 (105600 2 - 64.97) = 0.69A, \ i_{1} = 4.17 (05(600.2 - 48.99) = 2.51A$$

$$i_{x} = \frac{1}{2} l_{1} \ i_{1}^{2} + \frac{1}{2} l_{1} i_{x}^{2} - M \ i_{1} \cdot i_{x} = \frac{1}{2} 0.82.51^{2} + \frac{1}{2} 1.20.69^{2} - 0.62.51 \ 0.69$$

$$= 1.77J$$

- 3. For the network in the figure, find
- (a) the complex power supplied by the source,
- (b) the average power delivered to the 8- Ω resistor.



(a)
$$Z_{cd} = \frac{78+j45}{(\frac{3}{4})^2} = 2+j5$$
 2'
 $Z_{ab} = \frac{8-j20+2cd}{(\frac{5}{2})^2} = 1.6-2.4j$ 2'

$$Z_{ab} = \frac{8-j20+2cd}{(\frac{5}{2})^2} = 1.6-2.4j 2$$

$$II_{1} = \frac{4000}{j4+6+2ab} = \frac{4000}{7.6+1.6j} = 5.04 - 1.06j = 5.15 [-11.89] A$$

$$II_{2} = \frac{1}{2} 4000 \cdot I_{1} = \frac{1}{2} 4000 \cdot 5.15 [11.89] = 103 (11.89] VA$$

(b)
$$\frac{I_2}{I_1} = -\frac{2}{5}$$

 $\frac{I_2}{I_1} = -0.4I_1 = -0.4 \cdot 5.15 \angle 11.89^\circ = -2.06 \angle -11.89^\circ A$ 2
 $\frac{I_2}{I_1} = -\frac{2}{5}$
 $\frac{I_2}{I_2} = -\frac{2}{5}$

4. Calculate the average power dissipated by the 40Ω resistor.

Node A:
$$\frac{V_1-200}{10} + \frac{V_1-V_4}{40} + I_1=0$$

Node B: $\frac{V_4-V_1}{40} + \frac{V_4}{60} - I_3=0$
 $V_2 \in V_3$: $I_2 = \frac{V_2-V_3}{50}$
 $\frac{V_2}{V_1} = -\frac{1}{1}$ $\frac{I_2}{I_1} = -\frac{1}{2}$
 $\frac{V_4}{V_3} = -\frac{3}{1}$ $\frac{I_3}{I_2} = \frac{1}{3}$

$$\frac{1}{1000} = \frac{(V_1 - V_4)^2}{40} = \frac{14.76^2}{100} = 5.45W 2$$

Mesh Analysis.

Mesh 2:
$$-V_2 + 50(J_2 - J_4) + V_3 = 0$$

Mesh 2: $-V_2 + 50(J_2 - J_4) + V_3 = 0$

Mesh 3: $-V_4 + 60J_3 = 0$

Mesh 4: $40J_4 + V_4 - V_3 + 50(J_4 - J_4) + V_2 - V_1 = 0$

Then howers: V_2
 $V_4 = -\frac{3}{1}$
 $V_4 = -\frac{3}{12 - 24}$
 $V_4 = -\frac{3}{12 - 24}$
 $V_4 = -\frac{1}{12 - 24}$

$$P_{420} = I_4^2 40 = 0.37^2 40 = 5.48W$$
or $P_{900} = I_4^2 40 = 0.369^2 40 = 5.45W$

5. Sketch the Bode plots for

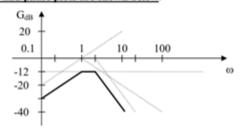
$$G(s) = \frac{s}{(s+2)^2(s+1)}, \qquad s = j\omega$$

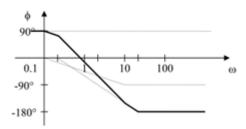
5. Solution:

$$\begin{split} \mathbf{G}(\omega) &= \frac{(1/4)\,j\omega}{(1+j\omega)(1+j\omega/2)^2} \\ \mathbf{G}_{dB} &= -20log_{10}\,4 + 20\,log_{10}\Big|\,j\omega\Big| - 20\,log_{10}\Big|\,1 + j\omega\Big| - 40\,log_{10}\Big|\,1 + j\omega/2\,\Big| \end{split}$$

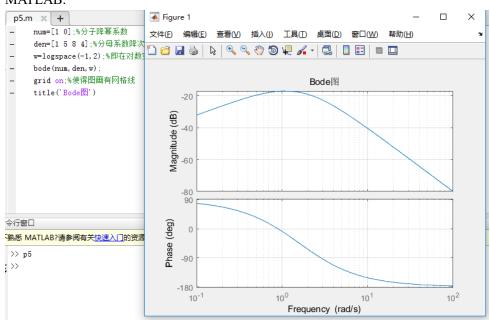
The magnitude and phase plots are shown below.

 $\phi = 90^{\circ} - \tan^{-1}\omega - 2 \tan^{-1}\omega/2$

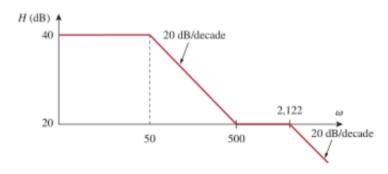




MATLAB:



6. The magnitude plot below represents the transfer function of a preamplifier. Find H(s).



6.Solution

$$40 = 20 \log_{10} K \longrightarrow K = 100$$

There is a pole at ω =50 giving $1/(1+j\omega/50)$

There is a zero at ω =500 giving $(1 + j\omega/500)$.

There is another pole at ω =2122 giving 1/(1 + j ω /2122).

Thus,

$$H(\omega) = \frac{100(1+j\omega/500)}{(1+j\omega/50)(1+j\omega/2122)} = \frac{100x\frac{1}{500}(s+500)}{\frac{1}{50}x\frac{1}{2122}(s+50)(s+2122)}$$

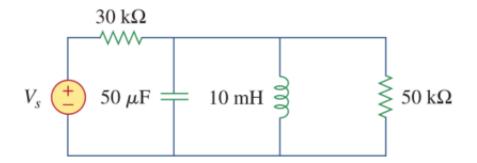
or

$$H(s) = \frac{21220(s+500)}{(s+50)(s+2122)}$$

> Hint:

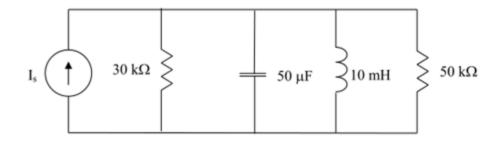
- $H(S) = \frac{-21220(S+500)}{(S+50)(S+2122)}$ is also correct.
- One solution gets full marks.

7. For the circuit shown, find ω_0 , B, and Q, as seen by the voltage across the inductor.



7. Solution:

Convert the voltage source to a current source as shown below.



$$R = 30/\!/50 = 30x50/\!/80 = 18.75 \; k\Omega$$
 This is a parallel resonant circuit.

parallel resonant circuit.
$$\omega_o = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{10x10^{-3}x50x10^{-6}}} = \frac{1414.21 \text{ rad/s}}{\sqrt{10x10^{-3}x50x10^{-6}}} = \frac{1414.21 \text{ rad/s}}{18.75x10^{3}x50x10^{-6}} = \frac{1.067 \text{ rad/s}}{1.067}$$

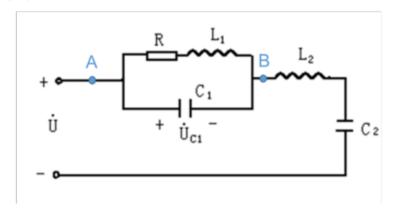
$$Q = \frac{\omega_o}{B} = \frac{447.21}{1.067} = \frac{1325.41}{1.067}$$

8. For the circuit below, R=50Ω, L₁=5mH, L₂=20mH, C₂=1μF, when the frequency

of voltage source $f = \frac{10^4}{2\pi}$ Hz, R, L₁, C₁ is in resonance as observed between Points

A and B.

At this moment, the voltage Uc_1 of capacitor C_1 is 10 V (Uc_1 =10 V). Please find C_1 and U(rms).



8.Solution: when R_{Σ} $L_{1\Sigma}$ C_{1} is resonance, the

resonance angular frequency is: $\omega_1 = 2\pi f = 2\pi \times \frac{10^4}{2\pi} = 10^4 \text{ Hz}$

$$\begin{split} Y_1 &= \frac{1}{R + j\omega_1 L_1} + j\omega_1 C_1 = \left[\frac{R}{R^2 + (\omega_1 L_1)^2} \right] + j \left[\omega_1 C_1 - \frac{\omega_1 L_1}{R^2 + (\omega_1 L_1)^2} \right] \\ &I_m \left[Y_1 \right] = 0 \,, \qquad \omega_1 C_1 - \frac{\omega_1 L_1}{R^2 + (\omega_1 L_1)^2} = 0 \\ & \therefore \qquad C_1 = \frac{L_1}{R^2 + (\omega_1 L_1)^2} = 10^{-6} \, F = 1 \mu \, F \\ & Z &= \frac{1}{Re[Y_1]} + j\omega_1 L_2 + \frac{1}{j\omega_1 C_2} = \frac{R^2 + (\omega_1 L_1)^2}{R} + j \left(\omega_1 L_2 - \frac{1}{\omega_1 C_2} \right) = 100(1 + j) \\ &R_1 &= \frac{1}{Re[Y_1]} = 100\Omega \qquad \qquad I &= \frac{U_{CI}}{R_1} = \frac{10 \, V}{100\Omega} = 0.1 \, A \\ &U &= I \cdot Z = 0.1 \times 100(1 + j) = 10(1 + j) \, V = 10\sqrt{2} \angle 45^{\circ} \, V \\ &\dot{U} &= 20 \cos(\omega t + 45^{\circ}) \, V \qquad \qquad \therefore \qquad U &= \frac{20}{\sqrt{2}} = \sqrt{2} \times 10 = 14.14 \, V \end{split}$$