

Signals and Systems Homework 5

Due Time: 23:59 April 20, 2018

1. (15) Suppose $g(t) = x(t)\cos(t)$ and the Fourier transform of the $g(t)$ is

$$G(jw) = \begin{cases} 1, & |w| \leq 2 \\ 0, & \text{else} \end{cases}$$

- (a) (5) Determine $x(t)$ Draw the frequency domain.
 (b) (10) Specify the Fourier transform $X_1(jw)$ of a signal $x_1(t)$,

$$g(t) = x_1(t)\cos\left(\frac{2}{3}t\right)$$

Solution:

- (a) We have that

$$w(t) = \cos(t) \Leftrightarrow_{FT} W(jw) = \pi[\delta(w-1) + \delta(w+1)]$$

$$g(t) = x(t)\cos t \Leftrightarrow G(jw) = \frac{1}{2\pi}[X(jw) * W(jw)]$$

Therefore $G(jw) = \frac{1}{2}X(j(w-1)) + \frac{1}{2}X(j(w+1))$
 is the first picture and (a) is the second.

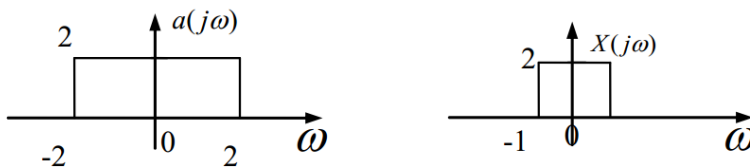


Figure 1: solution

$$(b) G(jw) = \frac{1}{2\pi} X_1(jw) * \pi[\delta(w + \frac{2}{3}) + \delta(w - \frac{2}{3})] = \frac{1}{2} [X_1(j(w + \frac{2}{3})) + X_1(j(w - \frac{2}{3}))]$$

Do a shift on frequency domain:

$$1/2(X_1(jw) + X_1(j(w + \frac{4}{3}))) = G(j(2 + \frac{2}{3})) \text{ and}$$

$$1/2(X_1(j(w + \frac{4}{3})) + X_1(j(w + \frac{8}{3}))) = G(j(2 + \frac{2}{3} + \frac{4}{3}))$$

In this way we can get the solution:

$$\begin{aligned} X_1(jw) &= 2 \times [G(j(w + \frac{2}{3})) - G(j(w + \frac{2}{3} + \frac{4}{3})) + G(j(w + \frac{2}{3} + \frac{4}{3} \times 2)) \dots] \\ &= \sum_{n=0}^{\infty} 2 \times (-1)^{n+1} G(j(w + \frac{2}{3} + \frac{4}{3} \times n)) \end{aligned}$$

2. (15) Consider a LTI system whose response to the input $x(t) = [e^{-t} + e^{-3t}]u(t)$ is $y(t) = [2e^{-t} - 2e^{-4t}]u(t)$

- (a) (5) Find the frequency response of this system.
- (b) (5) Determine the impulse response of the system.
- (c) (5) Find the differential equation of the system.

Solution:

(a) $H(jw) = \frac{Y(jw)}{X(jw)} = \frac{3(3+jw)}{(4+jw)(2+jw)}$

(b) The inverse of (a) is the solution $h(t) = \frac{3}{2}[e^{-4t} + e^{-2t}]u(t)$

(c) We have $\frac{Y(jw)}{X(jw)} = \frac{9+3jw}{8+6jw-w^2}$

The inverse of it is the solution $-\frac{d^2y(t)}{dt^2} + 6\frac{dy(t)}{dt} + 8y(t) = 3\frac{dx(t)}{dt} + 9x(t)$

3. (10) Consider a causal LTI system with frequency response

$$H(jw) = \frac{1}{jw + 3}$$

For an input

$$y(t) = [e^{-3t} - e^{-4t}]u(t)$$

determine $x(t)$ solution:

$H(jw) = \frac{Y(jw)}{X(jw)}$, as we know $y(t) = e^{-3t} - e^{-4t}u(t)$, we can get $Y(jw) = \frac{1}{3+jw} - \frac{1}{4+jw} = \frac{1}{(3+jw)(4+jw)}$,
and $H(jw) = \frac{1}{3+jw}$, we can get $x(jw) = \frac{1}{4+jw}$
so $x(t) = e^{-4t}u(t)$

4. (20) Ideal low pass filter frequency response is shown. Draw the spectrum of the output signal when input is the following function.

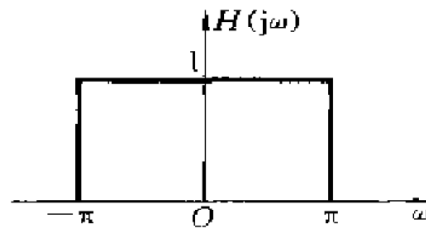


Figure 2: Lowpass Fliter

(a) $f(t) = \frac{\sin(\pi t)}{\pi t}$

(b)

$$f(t) = \begin{cases} 1, & |t| \leq 1 \\ 0, & |t| > 1 \end{cases}$$

solution:

(a) If $f(t) = \frac{\sin(\pi t)}{\pi t}$, we have $F(jw) = g_{2\pi}(w)$ so the spectrum of output signal $Y(jw) = H(jw)F(jw) = g_{2\pi}(w) \times g_{2\pi}(w) = g_{2\pi}(w)$

(b) $F(jw) = 2Sa(w)$, so $Y(jw) = H(jw)F(jw) = 2Sa(w) \times g_{2\pi}(w)$

The spectrum

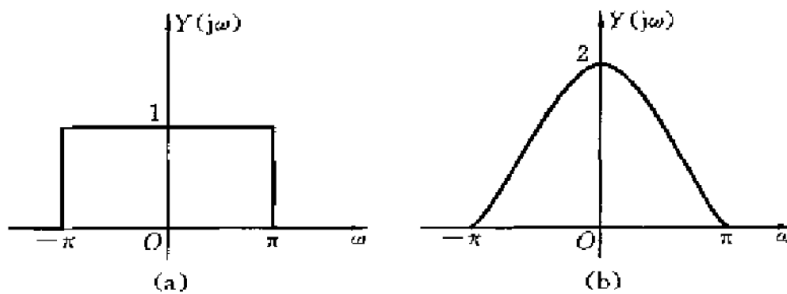


Figure 3: spectrum

5. (20) The spectrum of input band-limited signals is shown in figure a. The highest angular frequency is w_m and $w_b > w_m$, the cutoff frequency of figure b(HP) is w_b ,

$$H_1(jw) = \begin{cases} K_1, |w| > w_b \\ 0, |w| < w_b \end{cases}$$

LP is

$$H_2(jw) = \begin{cases} K_2, |w| < w_b \\ 0, |w| > w_b \end{cases}$$

draw the spectrum of $x(t)$ and $y(t)$.

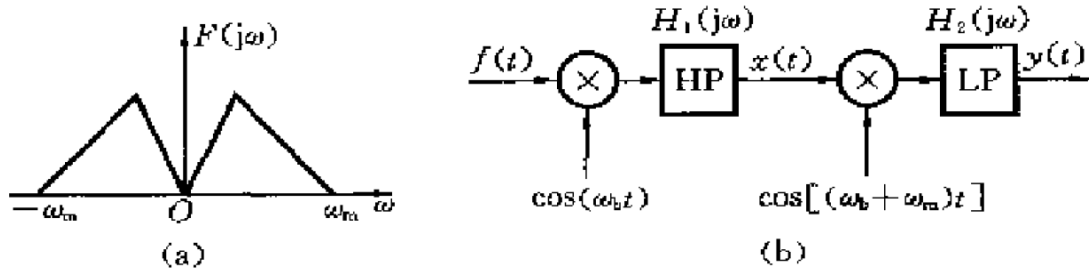


Figure 4: Signal and System

solution :

Let $p(t) = f(t)\cos(w_bt)$, from the frequency domain convolution theorem $P(jw) = \frac{1}{2\pi}F(jw) * \pi[\delta(w + w_b) + \delta(w - w_b)] = \frac{1}{2}[F(w + w_b) + F(w - w_b)]$

Because $w_b > w_m$, $P(jw)$ is shown in (a). $p(t)$ after the filter of w at the Angle of w_b is filtered out, and the component that is higher than that is multiplied by the K_1 weight.

$x(t)$ is shown in (b)

$Q(jw)$ and $Y(jw)$ is shown:

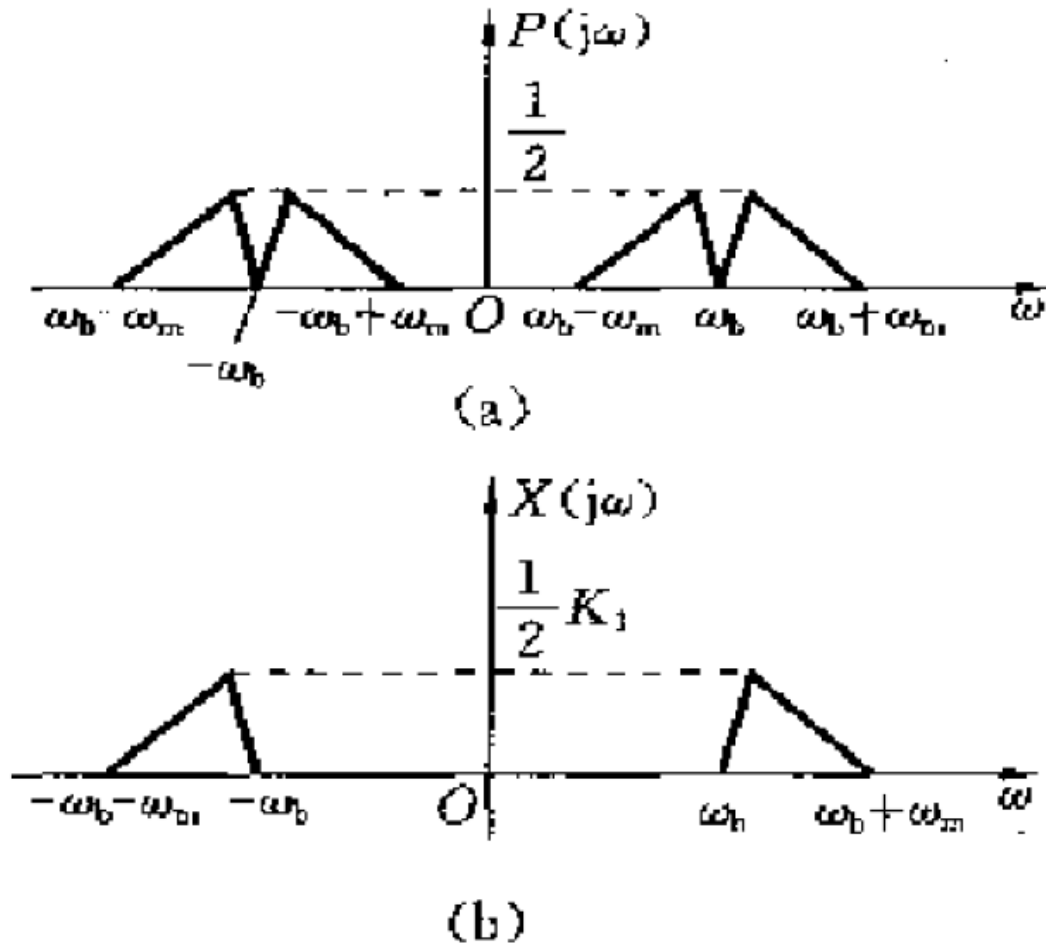


Figure 5: $P(j\omega)$ and $X(j\omega)$

- (c) Let $q(t) = x(t)\cos[(\omega_b + \omega_m)t]$, we have $Q(j\omega) = \frac{1}{2}[X(\omega + \omega_b + \omega_m) + X(\omega - \omega_b - \omega_m)]$
 $q(t)$ after the low-pass filter at the angular frequency ω_m is filtered out, and the component of this frequency is multiplied by the K_2 weight.
(d) $Q(j\omega)$ and $Y(j\omega)$ is shown:

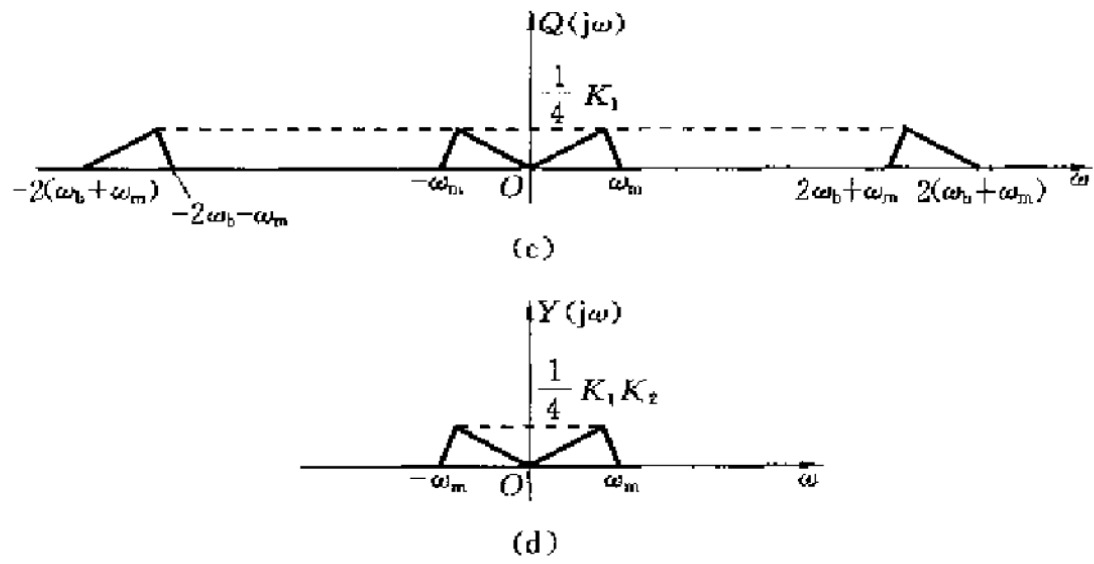


Figure 6: $Q(j\omega)$ and $Y(j\omega)$

6. (20) The bandpass filter responds to the figure.

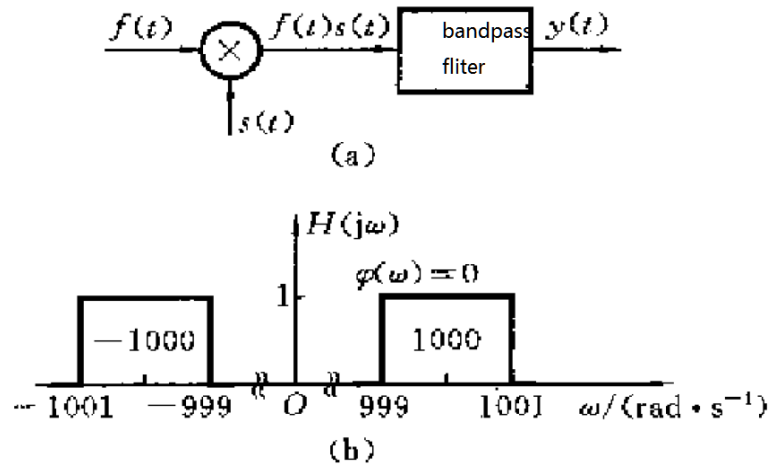


Figure 7: Signal and System

The inputs are $f(t) = \frac{\sin(2\pi t)}{2\pi t}$, $s(t) = \cos(1000t)$. Determine the output signal $y(t)$

Solution: $f(t) = \frac{\sin(2\pi t)}{2\pi t} \Leftrightarrow F(j\omega) = \frac{1}{2}g_{4\pi}(\omega)$

$$s(t) = \cos(1000t) \Leftrightarrow S(j\omega) = \pi[\delta(\omega + 1000) + \delta(\omega - 1000)]$$

so

$$f(t)s(t) \Leftrightarrow \frac{1}{2\pi}F(j\omega) * S(j\omega) = \frac{1}{4}[g_{4\pi}(\omega + 1000) + g_{4\pi}(\omega - 1000)]$$

let $x(t) = f(t)s(t)$, then

$$X(j\omega) = \frac{1}{4}[g_{4\pi}(\omega + 1000) + g_{4\pi}(\omega - 1000)]$$

the output signal

$$Y(j\omega) = X(j\omega)H(j\omega) = \frac{1}{4}[g_2(\omega + 1000) + g_2(\omega - 1000)]$$

because $\frac{1}{\pi}Sa(t) \Leftrightarrow g_2(\omega)$, in conclusion

$$\begin{aligned} y(t) &= \frac{1}{4\pi}Sa(t)e^{-j1000t} + \frac{1}{4\pi}Sa(t)e^{j1000t} = \frac{1}{2\pi}Sa(t)\cos(1000t) \\ &= \frac{\sin t}{2\pi t}\cos(1000t) \end{aligned}$$