Lecture 11

- Magnetically Coupled Circuits



Outline

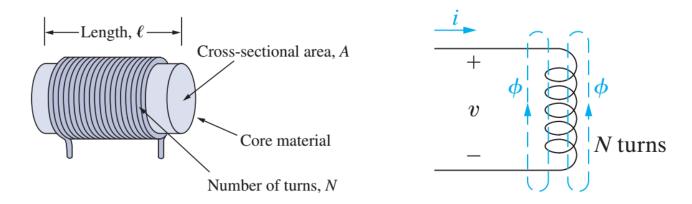
- Mutual inductance
- Transformers



Review: Self Inductance

Self inductance:

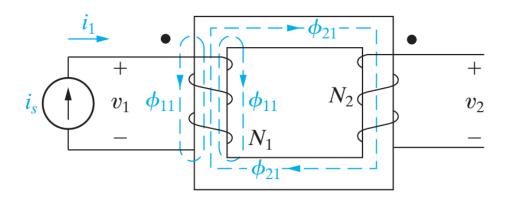
reaction of the inductor to the change in current through itself.



$$v = L \frac{di}{dt}$$

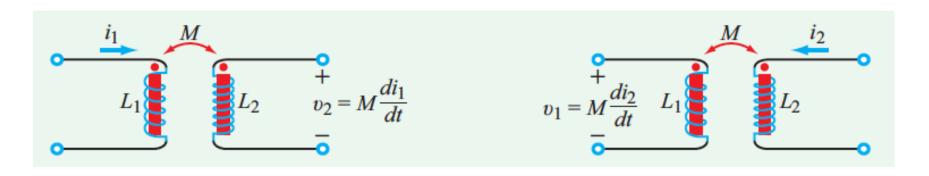
Mutual Inductance

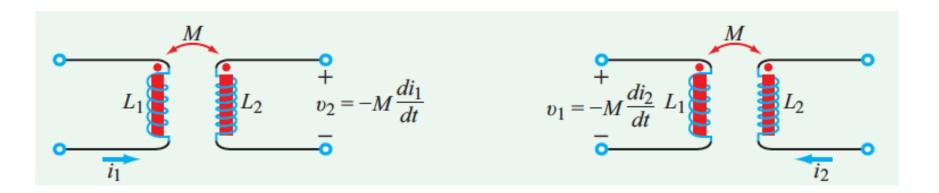
 Mutual inductance: reaction of one inductor to the change in current through another inductor.



$$v_2 = \frac{d(N_2\phi_{21})}{dt} = N_2 \frac{d}{dt} (\mathcal{P}_{21}N_1i_1)$$
$$= N_2N_1\mathcal{P}_{21}\frac{di_1}{dt}$$
$$= M_{21}\frac{di_1}{dt}$$

Dot Convention: Defines Directions of Windings





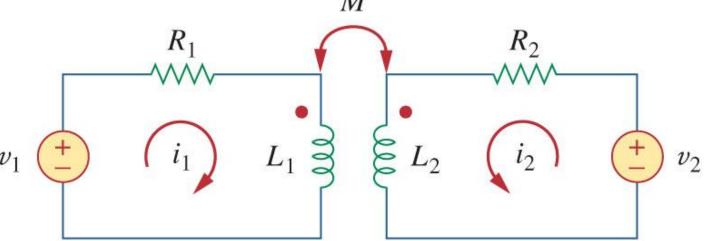
If a current enters the dotted terminal of one coil, the reference polarity of mutual voltage in the 2nd coil is the positive at the dotted terminal of the 2nd coil.



Magnetically Coupled Circuits

- L_1, L_2 : self-inductances
- *M*: mutual inductance
- Dots: indicating polarity of mutually induced voltages.

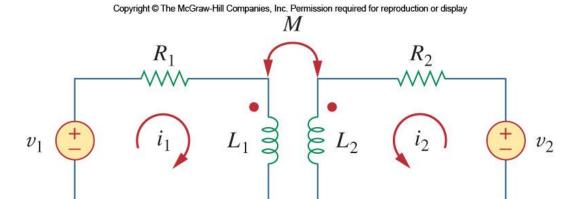
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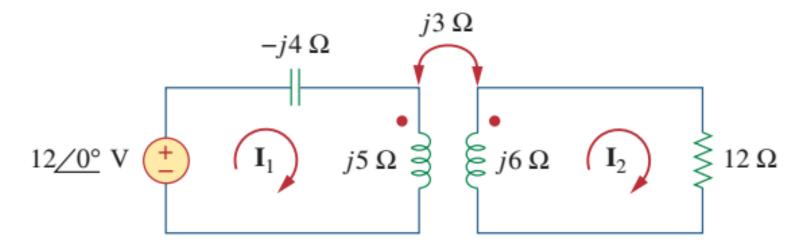
Analysis

- Find i_1 and i_2 .
 - In time domain
 - In phasor domain





- Calculate the phasor currents I₁, and I₂
- Calculate the phasor voltages V₁, and V₂ across the inductors

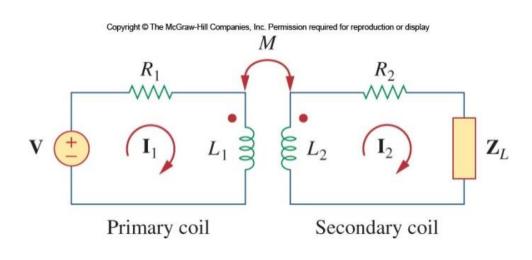






Transformers

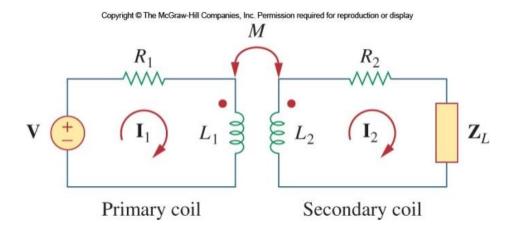
 A transformer is a magnetic device that takes advantage of mutual inductance.





Transformer Impedance

• An important parameter to know for a transformer is how the input impedance Z_{in} is seen from the source.



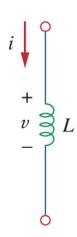
$$\mathbf{Z}_{\text{in}} = \frac{\mathbf{V}}{\mathbf{I}_1} = R_1 + j\omega L_1 + \frac{\omega^2 M^2}{R_2 + j\omega L_2 + \mathbf{Z}_L}$$

Reflected impedance from secondary to primary



Energy in a Coupled Circuit

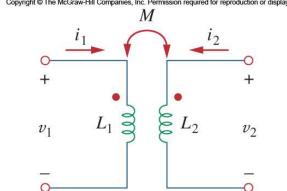
The energy stored in an inductor is



 For coupled inductors, the total energy stored is

$$w = \frac{1}{2} L_1 i_1^2 + \frac{1}{2} L_2 i_2^2 \pm M i_1 i_2$$

 The positive sign is selected when the currents both enter or leave the dotted terminals.



Coupling Coefficient k

The system cannot have negative energy

$$\frac{1}{2}L_{1}i_{1}^{2} + \frac{1}{2}L_{2}i_{2}^{2} - Mi_{1}i_{2} \ge 0 \qquad \Longrightarrow \qquad M \le \sqrt{L_{1}L_{2}}$$

 Define a parameter describes how closely M approaches upper limit.

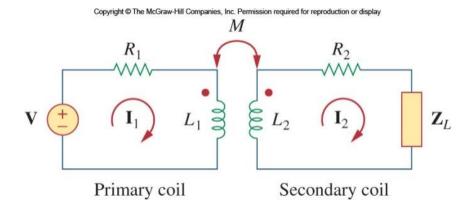
$$k = \frac{M}{\sqrt{L_1 L_2}}$$

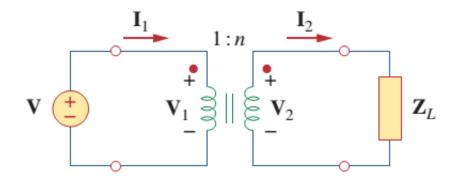
• Coupling coefficient, $0 \le k \le 1$.

- The ideal transformer has:
 - Coils with very large reactance

$$(L_1, L_2, M \rightarrow \infty)$$

- Coupling coefficient k=1.
- Primary and secondary coils are lossless, $R_1 = R_2 = 0$.



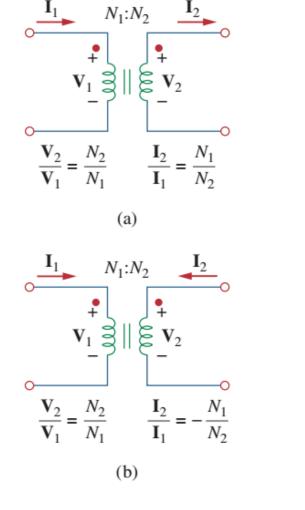


The voltage is related as:

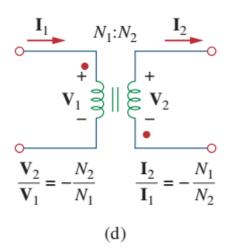
$$\frac{\mathbf{V_2}}{\mathbf{V_1}} = \frac{N_2}{N_1} = n$$

The current is related as:

- 1. If V_1 and V_2 are *both* positive or both negative at the dotted terminals, use +n in Eq. (13.52). Otherwise, use -n.
- 2. If I_1 and I_2 both enter into or both leave the dotted terminals, use -n in Eq. (13.55). Otherwise, use +n.

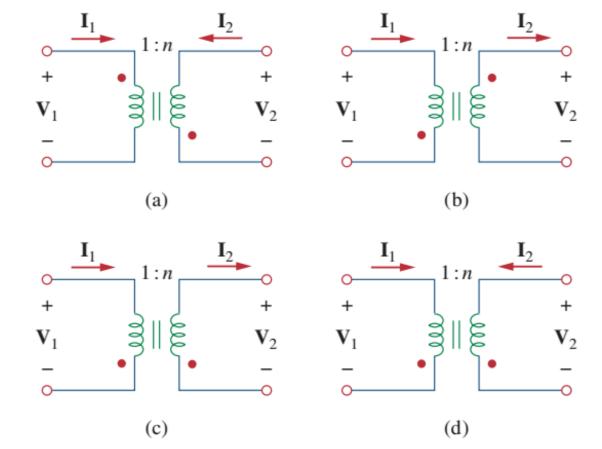


$$\begin{array}{c|c}
\mathbf{I}_1 & N_1:N_2 & \mathbf{I}_2 \\
\mathbf{V}_1 & \mathbf{V}_2 & \mathbf{V}_2 \\
\hline
\mathbf{V}_2 & \mathbf{V}_1 & \mathbf{I}_2 & \mathbf{V}_2 \\
\hline
\mathbf{V}_1 & \mathbf{V}_2 & \mathbf{I}_1 & \mathbf{V}_2
\end{array}$$
(c)

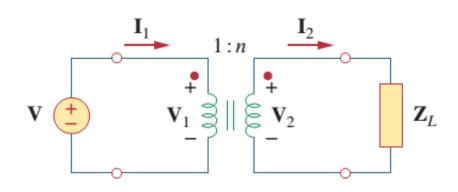


Practice

13.36 As done in Fig. 13.32, obtain the relationships between terminal voltages and currents for each of the ideal transformers in Fig. 13.105.



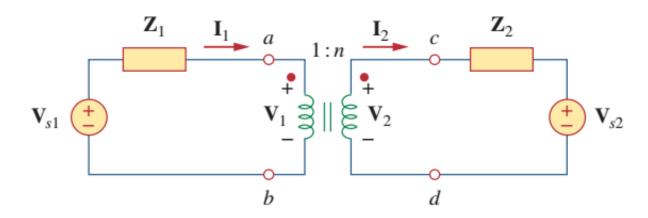




$$\frac{\mathbf{V_2}}{\mathbf{V_1}} = \frac{N_2}{N_1} = n$$

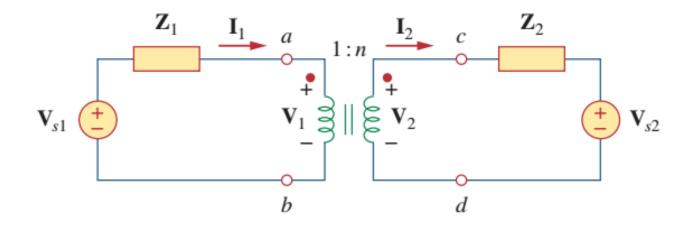
Reflected impedance

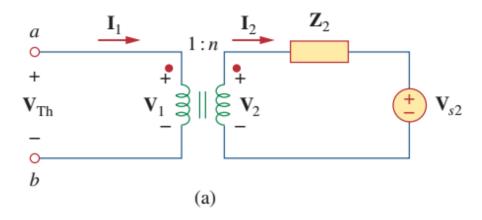
$$\mathbf{Z}_{\mathrm{in}} = \frac{\mathbf{V}_1}{\mathbf{I}_1} =$$

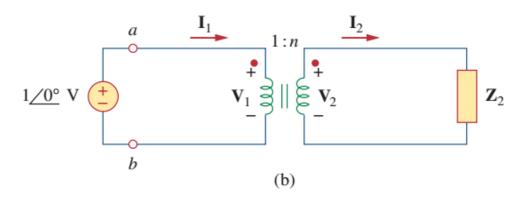


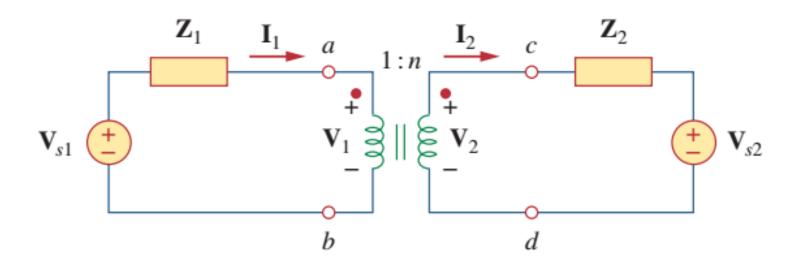
$$\frac{\mathbf{V_2}}{\mathbf{V_1}} = \frac{N_2}{N_1} = n$$

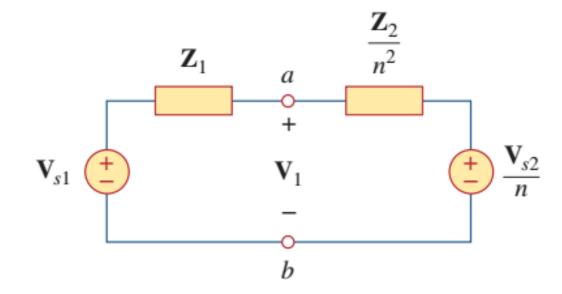
Reflected impedance and source





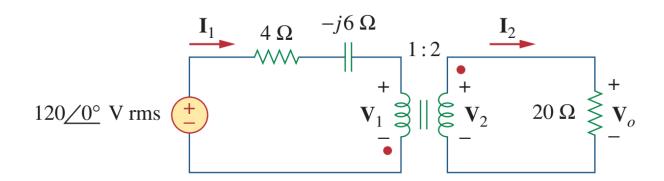


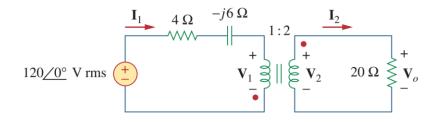




Example

For the ideal transformer circuit of Fig. 13.37, find: (a) the source current I_1 , (b) the output voltage V_o , and (c) the complex power supplied by the source.





Solution:

(a) The 20- Ω impedance can be reflected to the primary side and we get

$$\mathbf{Z}_R = \frac{20}{n^2} = \frac{20}{4} = 5 \,\Omega$$

Thus,

$$\mathbf{Z}_{\text{in}} = 4 - j6 + \mathbf{Z}_{R} = 9 - j6 = 10.82 / -33.69^{\circ} \Omega$$

$$\mathbf{I}_{1} = \frac{120 / 0^{\circ}}{\mathbf{Z}_{\text{in}}} = \frac{120 / 0^{\circ}}{10.82 / -33.69^{\circ}} = 11.09 / 33.69^{\circ} A$$

(b) Since both I_1 and I_2 leave the dotted terminals,

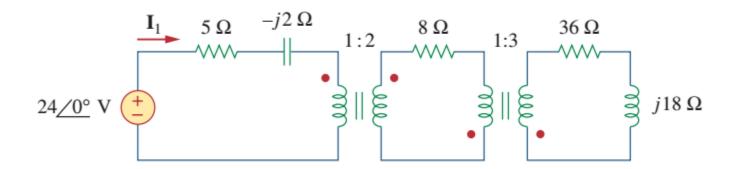
$$\mathbf{I}_2 = -\frac{1}{n}\mathbf{I}_1 = -5.545 / 33.69^{\circ} \text{ A}$$
 $\mathbf{V}_o = 20\mathbf{I}_2 = 110.9 / 213.69^{\circ} \text{ V}$

(c) The complex power supplied is

$$\mathbf{S} = \mathbf{V}_s \mathbf{I}_1^* = (120/0^\circ)(11.09/-33.69^\circ) = 1,330.8/-33.69^\circ \text{ VA}$$

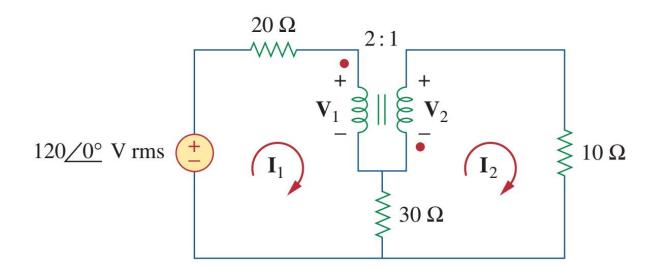
Practice

Find reflected impedance and I₁



Example

Calculate the power supplied to the $10-\Omega$ resistor in the ideal transformer circuit of Fig. 13.39.



$$-120 + (20 + 30)\mathbf{I}_1 - 30\mathbf{I}_2 + \mathbf{V}_1 = 0$$

or

$$50\mathbf{I}_1 - 30\mathbf{I}_2 + \mathbf{V}_1 = 120 \tag{13.9.1}$$

For mesh 2,

$$-\mathbf{V}_2 + (10 + 30)\mathbf{I}_2 - 30\mathbf{I}_1 = 0$$

or

$$-30\mathbf{I}_1 + 40\mathbf{I}_2 - \mathbf{V}_2 = 0 ag{13.9.2}$$

At the transformer terminals,

$$\mathbf{V}_2 = -\frac{1}{2}\mathbf{V}_1 \tag{13.9.3}$$

$$I_2 = -2I_1 (13.9.4)$$

(Note that n = 1/2.) We now have four equations and four unknowns, but our goal is to get I_2 . So we substitute for V_1 and I_1 in terms of V_2 and I_2 in Eqs. (13.9.1) and (13.9.2). Equation (13.9.1) becomes

$$-55I_2 - 2V_2 = 120 (13.9.5)$$

and Eq. (13.9.2) becomes

$$15I_2 + 40I_2 - V_2 = 0 \implies V_2 = 55I_2$$
 (13.9.6)

Substituting Eq. (13.9.6) in Eq. (13.9.5),

$$-165\mathbf{I}_2 = 120$$
 \Rightarrow $\mathbf{I}_2 = -\frac{120}{165} = -0.7272 \,\mathrm{A}$

The power absorbed by the $10-\Omega$ resistor is

$$P = (-0.7272)^2(10) = 5.3 \text{ W}$$