EE150: Signals and Systems, Spring 2022

Homework 2

(Due Thursday, Mar. 17 at 11:59pm (CST))

[12 points] Determine the continuous-time convolution of x(t) and h(t) for the following three cases:

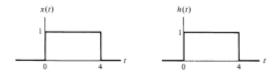


Figure 1.1: (a)

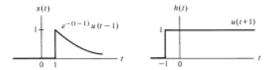


Figure 1.2: (b)

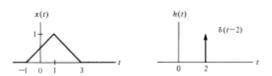


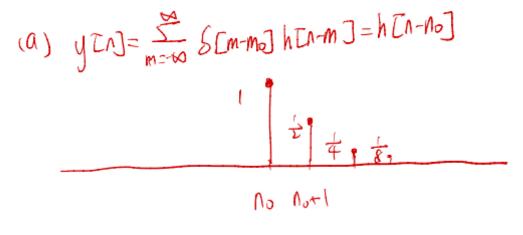
Figure 1.3: (c)

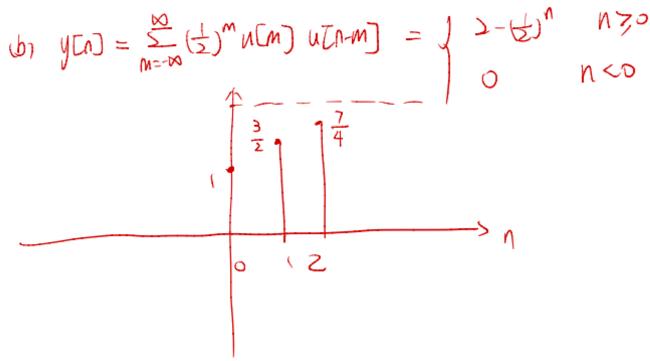
- 2. [12 points] Consider a discrete-time, linear, shift-invariant system that has unit sample response h[n] and input x[n].
 - (a) Sketch the response of this system if $x[n] = \delta[n n_0]$ for some $n_0 > 0$, and $h[n] = (\frac{1}{2})^n u[n]$.
 - (b) Evaluate and sketch the output of the system if $h[n] = (\frac{1}{2})^n u[n]$ and x[n] = u[n].
 - (c) Consider reversing the role of the input and system response in part (b). That is,

$$h[n] = u[n],$$

$$x[n] = (\frac{1}{2})^n u[n]$$

Evaluate the system output y[n] and sketch.





(C) Some as (b)

3. [5 points] Compute the convolution y[n] = x[n] * h[n] when

$$x[n] = \alpha^n u[n], 0 < \alpha < 1,$$

 $h[n] = \beta^n u[n], 0 < \beta < 1,$

Assume that α and β are not equal.

$$y[n] = x[n] * h[n]$$

$$= \sum_{m=-\infty}^{\infty} x[n-m]h[m]$$

$$= \sum_{m=-\infty}^{\infty} \alpha^{n-m} u [n-m]\beta^m u[m]$$

$$= \sum_{m=0}^{\infty} \alpha^{n-m}\beta^m, \qquad n > 0,$$

$$y[n] = \alpha^n \sum_{m=0}^n \left(\frac{\beta}{\alpha} \right)^m = \alpha^n \left[\frac{1 - (\beta/\alpha)^{n+1}}{1 - (\beta/\alpha)} \right]$$
$$= \frac{\alpha^{n+1} - \beta^{n+1}}{\alpha - \beta}, \quad n \ge 0,$$
$$y[n] = 0, \quad n < 0$$

- 4. [16 points] Determine if each of the following statements is true in general. Provide proofs for those that you think are true and counterexamples for those that you think are false.
 - (a) $x[n] * \{h[n]g[n]\} = \{x[n] * h[n]\}g[n]$
 - (b) If y(t) = x(t) * h(t), then y(2t) = 2x(2t) * h(2t).
 - (c) If x(t) and h(t) are odd signals, then y(t) = x(t) * h(t) is an even signal.
 - (d) If y(t) = x(t) * h(t), then $Ev\{y(t)\} = x(t) * Ev\{h(t)\} + Ev\{x(t)\} * h(t)$.

(Hint: It's taught that for an arbitrary signal x(t), we can have x(t) = g(t) + h(t) where g(t) is an odd signal and h(t) is an even signal, then $Ev\{x(t)\} = h(t)$.)

a) False. let
$$g[n] = S[n]$$

$$\Rightarrow \chi[n] * \{h[n] g[n]\}$$

$$= \chi[n] * h[n] \{g[n]\}$$

$$= \chi[n] * h[n] \{g[n]\}$$

$$= (\chi * h) [o]$$

$$= (\chi * h) [o]$$

b)
$$y(2t) = \int_{-\infty}^{\infty} x(\epsilon) h(2t-\epsilon) d\epsilon$$

let $\tau' = \frac{\tau}{2} \implies d\tau = 2 d\tau'$

$$= y(2t) = \int_{-\infty}^{\infty} x(2\tau') h(2t-2\tau') \cdot 2 d\tau'$$

$$= 2 x(2t) \times h(2t)$$
True

c)
$$y(t) = \chi(t) \times h(t)$$

$$y(-t) = \chi(-t) \times h(-t) = \int_{-\infty}^{\infty} \chi(-\tau) h(-t+\tau) d\tau$$

$$= \int_{-\infty}^{\infty} (-\chi(\tau)) \cdot (-h(t-\tau)) d\tau$$

$$= \int_{-\infty}^{\infty} \chi(\tau) h(t-\tau) d\tau$$

$$= y(t) = \int_{-\infty}^{\infty} \chi(\tau) h(t-\tau) d\tau$$

$$= \frac{1}{2} \left[\int_{-\infty}^{\infty} \chi(\tau) h(t-\tau) d\tau + \int_{-\infty}^{\infty} \chi(-\tau) h(-t-\tau) d\tau \right]$$

$$= \frac{1}{2} \left[\int_{-\infty}^{\infty} \chi(\tau) h(t-\tau) d\tau - \int_{-\infty}^{\infty} \chi(-\tau) h(-t+\tau) d\tau \right]$$

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$$= \chi(t) \times \text{odd} \{ h(t) \}$$

Example:
$$\chi(t) = \delta(t-1)$$

 $h(t) = \delta(t+1)$
 $\chi(t) = \delta(t)$

false.

5. [9 points] Let
$$x(t) = u(t-3) - u(t-5)$$
 and $h(t) = e^{-3t}u(t)$

- (a) Compute y(t) = x(t) * h(t).
- (b) Compute $g(t) = \frac{dx(t)}{dt} * h(t)$.
- (c) How is g(t) related to y(t)?

(a)
$$y(t) = [u(t-2) - u(t-3)] *h(t)$$

$$= u(t-3) *h(t) - u(t-3) *h(t)$$

$$= \int_{-\infty}^{\infty} u(t-3) e^{-3(t-7)} d\tau - \int_{-\infty}^{\infty} u(t-3) e^{-3(t-7)} d\tau$$

$$= \int_{3}^{4} e^{-3(t-7)} d\tau - \int_{3}^{4} e^{-3(t-7)} d\tau$$

$$= \int_{3}^{4} e^{-3(t-3)} d\tau - \int_{3}^{4} e^{-3(t-7)} d\tau$$

$$= \begin{cases} \frac{1}{3} - \frac{1}{3}e^{-3(t-3)} & 3 \le t \le J \\ -\frac{1}{3} e^{-3(t-3)} & 4 = -3(t-3) \end{cases}$$

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$$= \begin{cases} \frac{1}{3} - \frac{1}{3}e^{-3(t-3)} & 4 = -$$

 [10 points] Consider the cascade interconnection of three causal LTI systems, illustrated in Figure 6.1. The impulse response h₂[n] is

$$h_2[n] = u[n] - u[n-2],$$

and the overall impulse response is as shown in Figure 6.2.



Figure 6.1: LTI systems

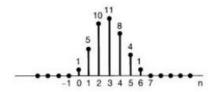


Figure 6.2: overall impulse response

- (a) Find the impulse response h₁[n].
- (b) Find the response of the overall system to the input

$$x[n] = \delta[n] - \delta[n-1].$$

$$_{(a)} h_2[n] = u[n] - u[n-2] = \delta[n] + \delta[n-1]$$

$$h_2[n] * h_2[n] = \delta[n] + 2\delta[n-1] + \delta[n-2]$$

Assume h[n] is the overall response,

$$h[n] = h_1[n] * (\delta[n] + 2\delta[n-1] + \delta[n-2]) = h_1[n] + 2h_1[n-1] + h_1[n-2]$$

$$\begin{cases} h[0] = h_1[0] = 1 \\ h[1] = h_1[1] + 2h_1[0] = 5 \\ h[2] = h_1[2] + 2h_1[1] + h_1[0] = 10 \\ h[3] = h_1[3] + 2h_1[2] + h_1[1] = 11 \\ h[4] = h_1[4] + 2h_1[3] + h_1[2] = 8 \\ h[5] = h_1[5] + 2h_1[4] + h_1[3] = 4 \\ h[6] = h_1[6] + 2h_1[5] + h_1[4] = 1 \\ , \text{solve and get} \end{cases}$$

So
$$h_1[n] = \delta[n] + 3\delta[n-1] + 3\delta[n-2] + 2\delta[n-3] + \delta[n-4]$$

(b)
$$y[n] = (\delta[n] - \delta[n-1]) * h[n] = h[n] - h[n-1]$$

$$\begin{cases} y[0] = h[0] - h[-1] = 1 \\ y[1] = h[1] - h[0] = 4 \\ y[2] = h[2] - h[1] = 5 \\ y[3] = h[3] - h[2] = 1 \\ y[4] = h[4] - h[3] = -3 \\ y[5] = h[5] - h[4] = -4 \\ y[6] = h[6] - h[5] = -3 \\ y[7] = h[7] - h[6] = -1 \\ \text{, otherwise 0} \end{cases}$$

7. [16 points] Consider the evaluation of

$$y[n] = x_1[n] * x_2[n] * x_3[n],$$

where $x_1[n] = (0.5)^n u[n]$, $x_2[n] = u[n+3]$, and $x_3[n] = \delta[n] - \delta[n-1]$.

- (a) Evaluate the convolution $x_1[n] * x_2[n]$.
- (b) Convolve the result of part (a) with $x_3[n]$ in order to evaluate y[n].
- (c) Evaluate the convolution $x_2[n] * x_3[n]$.
- (d) Convolve the result of part (c) with $x_1[n]$ in order to evaluate y[n].

Solution:

(a) We can know that

$$y_1[n] = x_1[n] * x_2[n] = \sum_{k=-\infty}^{\infty} x_1[k] x_2[n-k]$$
$$= \sum_{k=0}^{\infty} (0.5)^k u[n+3-k]$$
$$= \sum_{k=0}^{n+3} (0.5)^k$$

This evaluates to

$$y_1[n] = x_1[n] * x_2[n] = 2\{1 - \left(\frac{1}{2}\right)^{n+4}\}u[n+3]$$

Or

$$y_1[n] = \begin{cases} 2\{1 - \left(\frac{1}{2}\right)^{n+4}\}, n \ge -3\\ 0, & otherwise \end{cases}$$

(b) Now,

$$y[n] = x_3[n] * y_1[n] = y_1[n] - y_1[n-1].$$

Therefore,

$$y[n] = 2\{1 - \left(\frac{1}{2}\right)^{n+4}\}u[n+3] - 2\{1 - \left(\frac{1}{2}\right)^{n+3}\}u[n+2]$$
$$= \left(\frac{1}{2}\right)^{n+3}u[n+3]$$

Or

$$y[n] = \begin{cases} \left(\frac{1}{2}\right)^{n+3}, n \ge -2\\ 1, & n = -3\\ 0, & otherwise \end{cases}$$

Therefore, $y[n] = \left(\frac{1}{2}\right)^{n+3} u[n+3].$

(c) We have

$$y_2[n] = x_2[n] * x_3[n] = u[n+3] - u[n+2] = \delta[n+3]$$

(d) From the result of part(c), we get

$$y[n] = y_2[n] * x_1[n] = x_1[n+3] = \left(\frac{1}{2}\right)^{n+3} u[n+3]$$

8. [10 points]

(a) Consider an LTI system with input and output related through the equation

$$y(t) = \int_{-\infty}^{t} e^{-(t-\tau)}x(\tau - 2)d\tau.$$

What is the impulse response h(t) for this system?

(b) Determine the response of the system when the input x(t) is as shown in Figure 8.

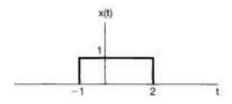


Figure 8: The Figure of x(t)

Solution:

(a) Note that

$$y(t) = \int_{-\infty}^{t} e^{-(t-\tau)} x(\tau - 2) d\tau = \int_{-\infty}^{t-2} e^{-\left(t-2-\tau\right)} x\left(\tau\right) d\tau$$

Therefore,

$$h(t) = e^{-(t-2)}u(t-2)$$

(b) We have

$$x(t) = u(t+1) - u(t-2)$$

$$y(t) = \int_{-\infty}^{\infty} h(\tau)x(t-\tau)d\tau$$

$$= \int_{2}^{\infty} e^{-(\tau-2)} [u(t-\tau+1) - u(t-\tau-2)] d\tau$$
$$= \int_{2}^{t+1} e^{-(\tau-2)} d\tau - \int_{2}^{t-2} e^{-(\tau-2)} d\tau$$

We can know that:

$$y(t) = \begin{cases} 0, & t < 1\\ 1 - e^{-(t-1)}, & 1 \le t < 4\\ e^{-(t-4)} - e^{-(t-1)}, t \ge 4 \end{cases}$$

We can know that

$$y(t) = u(t-1)(1-e^{-t+1}) - u(t-4)(1-e^{-t+4})$$

9. [10 points] Suppose that the signal

$$x(t) = u(t + 0.5) - u(t - 0.5)$$

is convolved with the signal

$$h(t) = e^{i\omega_0 t}$$

(a) Determine a value of w₀ which ensures that

$$y(0) = 0$$
,

where y(t) = x(t) * h(t).

(b) Is your answer to the previous part unique?

$$y(t) = \chi(t) + h(t) = \int_{-0.5}^{0.5} e^{i\omega_0(t-7)} d\tau$$

$$y(0) = \int_{-0.5}^{0.5} e^{-i\omega_0 \tau} d\tau = \frac{2}{\omega_0} \sin(\frac{\omega_0}{2}) = 0$$

$$\int_{-0.5}^{0.5} e^{-i\omega_0 \tau} d\tau = 2\pi$$

$$\int_{-0.5}^{0.5} e^{-i\omega_0 \tau} d\tau = 2\pi$$