- Part 1: Overview

February 15, 2022

### Introduction to Signals

#### Signal:

a function of one or more independent variables (e.g., time and spatial variables); typically contains information about the behaviour or nature of some physical phenomena.

e.g. Voice (audio), TV (audio + video), voltage, current, stock price, etc.

e.g. Picture (brightness), air pressure, temperature, and wind speeds (altitude), etc.

This course focuses on signals involving a single independent variable, i.e., time.

### Introduction to Systems

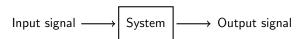
#### System:

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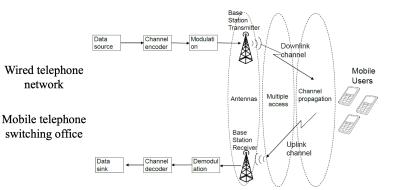
responds to a particular signal input by producing another signal (output).

e.g. Biological sensory system, electronic circuits, automobile, etc.



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### Example: Cellular Communication Systems



# Objectives

### System characterization

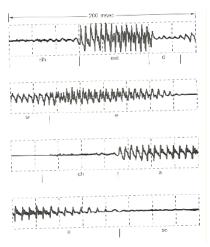
how it responds to input signal (e.g. human auditory system)

#### System design

to process signal in a particular way (e.g. signal restoration, signal identification, image processing)

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Audio (intensity vs. time) characteristics: volume, rhythm, pitch

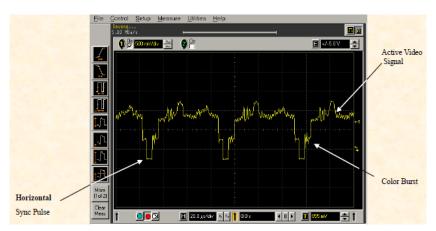


# Examples of Signals cont.

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TV signal (voltage vs. time) modulated picture signal + audio signal; carrier signal; system involves: antenna, tuner, CRT



# Examples of Signals cont.

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Biomedical signal (voltage vs. time) e.g. Electrocardiogram
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Traffic flow (quantity vs. time) volume, composition, pattern, mobility; system involved: traffic lights, roads, junctions
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Network throughput # of packets/sec., pattern

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### Stock price (index vs. time; \$ vs. time)



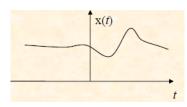
# Continuous vs. Discrete Time Signal

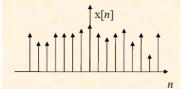
Continuous-time Signal (independent variable: t)

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Discrete-time Signal (independent variable: n)





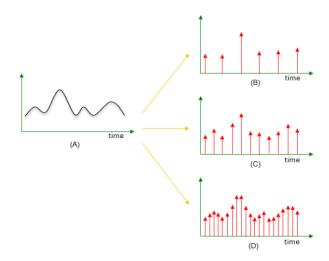
x[n] is also referred as a sequence. Any particular one in x[n] is called a sample.

Discrete-time signals are inherently discrete OR sampling of continuous-time signals

# Sampling

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Lecture 01



### Transformation of Independent Variable

Lecture 01

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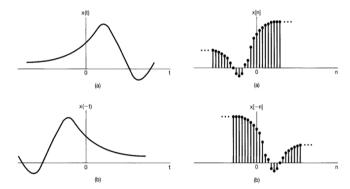
- ① Time reflection:  $x(t) \longleftrightarrow x(-t), x[n] \longleftrightarrow x[-n]$
- 2 Time scaling:  $x(t) \longleftrightarrow x(ct)$
- 3 Time shift:  $x(t) \longleftrightarrow x(t-t_0), x[n] \longleftrightarrow x[n-n_0]$

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### Transformation of Independent Variable cont.

Time reflection: reflect the signal at t = 0 or n = 0

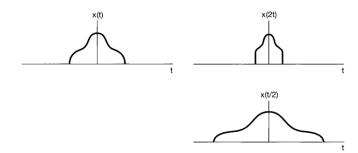
$$x(t) \longleftrightarrow x(-t)$$
  
 $x[n] \longleftrightarrow x[-n]$ 



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Time scaling: stretched (|c| < 1) or compressed (|c| > 1) version of x(t)

$$x(t) \longleftrightarrow x(ct)$$



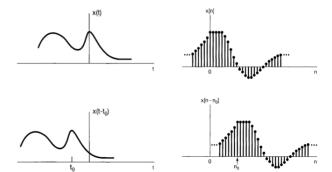
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### Transformation of Independent Variable cont.

Time shift: delayed  $(t_0 > 0)$  or advanced  $(t_0 < 0)$  version of x(t)

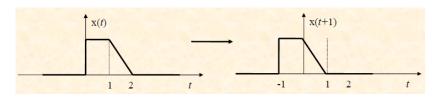
$$x(t)\longleftrightarrow x(t-t_0)$$

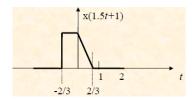
$$x[n]\longleftrightarrow x[n-n_0]$$



### Transformation of Independent Variable cont.

To perform transformation  $x(t) \rightarrow x(\alpha t + \beta)$ , You have to do time-shifting and then scaling.





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$$y(t) = x(1.5t + 1)$$

Work out a few points:

$$y(0)=x(1)$$

$$y(1) = x(2.5)$$

$$y(2) = x(4)$$

To get from y to x, we first scale t, then shift.

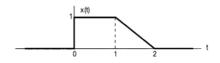
Therefore, to get from x to y, we first shift and then scale.

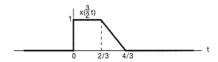
Q: What if we first scale and then shift?

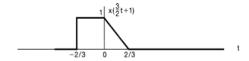
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### Transformation of Independent Variable cont.

You can also first scale and then shift







## Examples

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Problem 1.5: Let x(t) be a signal with x(t) = 0 for t < 3. For each signal given below, determine the values of t for which it is guaranteed to be zero

- (a) x(1-t)
- (b) x(1-t) + x(2-t)
- (c) x(3t)

Problem 1.22: Describe how to obtain the following signals:

- (a) x[3 n]
- (b) x[3n+1]

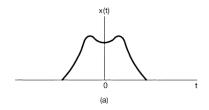
### Even and Odd Functions

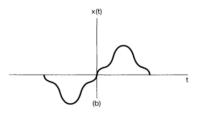
A signal is called an even signal (function) if

$$x(t) = x(-t), \quad x[n] = x[-n].$$

A signal is called an odd signal (function) if

$$x(t) = -x(-t), \quad x[n] = -x[-n].$$





### Even and Odd Functions cont.

Any signal can be broken into sum of one even and one odd signal.

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$$x(t) = Evenx(t) + Oddx(t)$$

How?

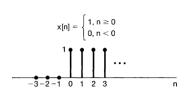
$$x(t) = Evenx(t) + Oddx(t),$$
  
 $x(-t) = Evenx(-t) + Oddx(-t) = Evenx(t) - Oddx(t),$ 

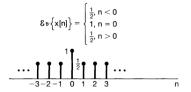
$$\implies Evenx(t) = \frac{1}{2}(x(t) + x(-t)),$$

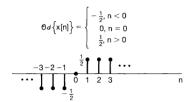
$$Oddx(t) = \frac{1}{2}(x(t) - x(-t)).$$

### Even and Odd Functions cont.

#### Example:







# Periodic and Aperiodic Signal

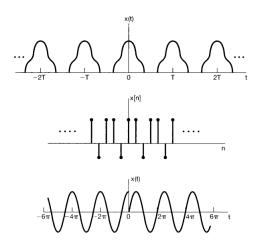
```
IF x(t) is periodic with period T, then x(t) = x(t + mT), \forall t; m \text{ can be any integer}
IF x[n] is periodic with period N, then x[n] = x[n + mN], \forall n; m \text{ can be any integer}
```

Fundamental period ( $T_0$  or  $N_0$ ): the <u>smallest positive</u> value of (T or N) for which the above equation holds.

Aperiodic is also called Non-periodic

### Periodic and Aperiodic Signal cont.

Q: What is the fundamental period of a constant function? Examples:



# **Exponential Signal**

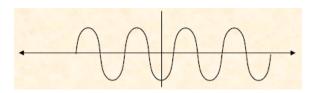
Real exponential:  $x(t) = ce^{at}$ 

positive a:

negative a:



Imaginary exponential:  $x(t) = e^{j(\omega_0 t + \Phi)}$ 



## Periodic and Sinusoidal Signal

#### Euler's Formula:

$$e^{j\omega_0t}=\cos(\omega_0t)+j\cdot\sin(\omega_0t)$$

Is  $x(t) = e^{j\omega_0 t}$  periodic? Suppose the period is T, then

$$x(t) = x(t+T), \implies e^{j\omega_0 t} = e^{j\omega_0(t+T)}.$$

Hence  $e^{j\omega_0T}=1$ .

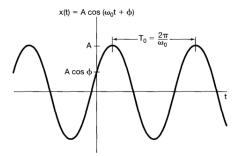
Fundamental period  $(T_0)$  is inversely proportional to fundamental frequency  $(|\omega_0|)$ :

$$T_0=2\pi/|\omega_0|.$$

## Periodic and Sinusoidal Signal cont.

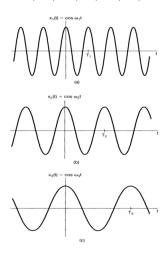
Sinusoidal signal:  $x(t) = A\cos(\omega_0 t + \phi)$  closely related to the periodic  $(T_0)$  complex exponential unit  $\omega_0$ : radians/sec;  $\phi$ : radians

phase:  $\omega_0 t + \phi$ 



### Periodic and Sinusoidal Signal cont.

Fundamental frequency:  $|\omega_1| > |\omega_2| > |\omega_3|$ 



### Periodic and Sinusoidal Signal cont.

Complex exponential can be written in terms of sinusoidal signals with the same fundamental frequency

$$e^{j\omega_0t}=\cos\omega_0t+j\sin\omega_0t$$

Sinusoidal signal can be written in terms of periodic complex exponentials with the same fundamental frequency

$$A\cos(\omega_0 t + \phi) = \frac{A}{2}e^{j\phi}e^{j\omega_0 t} + \frac{A}{2}e^{-j\phi}e^{-j\omega_0 t}$$

Periodic signal:  $x(t) = e^{j\omega_0 t}$ Total energy over a period

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$$E_{
m period} = \int_0^{T_0} |e^{j\omega_0 t}|^2 dt = \int_0^{T_0} 1 dt = T_0.$$

Average power over a period

$$P_{
m period} = rac{1}{T_0} E_{
m period} = 1.$$

Complex periodic exponential signal has finite average power Average power over a period

$$P_{\infty} = \lim_{T o \infty} rac{1}{2T} \int_{-T}^{T} |e^{i\omega_0 t}|^2 \mathrm{d}t = 1.$$

# Harmonically related Complex Exponential

A set of periodic exponentials with fundamental frequencies that are all multiples of a single positive frequency  $\omega_0=\frac{2\pi}{T_0}$ 

$$\phi_k(t) = e^{jk\omega_0 t}, \quad k = 0, \pm 1, \pm 2, \dots,$$

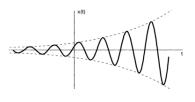
where  $\phi_k(t)$  is periodic with fundamental frequency  $|k|\omega_0$  and fundamental period

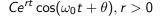
$$\frac{2\pi}{|k|\omega_0} = \frac{T_0}{|k|}.$$

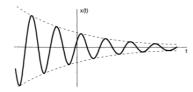
### General Complex Exponential

$$egin{aligned} x(t) &= C \cdot e^{(r+j\omega_0)t}, \ & C &= |C|e^{j heta}, \quad r, \omega_0 \in \mathbb{R} \end{aligned}$$

$$\implies x(t) = |C|e^{rt}e^{j(\omega_0 t + \theta)}$$



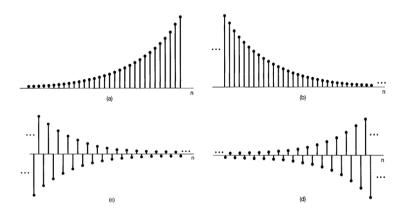




$$Ce^{rt}\cos(\omega_0 t + \theta), r < 0$$

# Discrete-Time Complex Exponential

Real exponential signals:  $x[n] = C\alpha^n$ , C and  $\alpha$  are real.

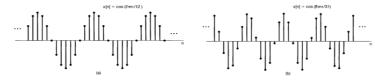


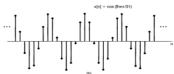
(a) 
$$\alpha > 1$$
, (b)  $0 < a < 1$ , (c)  $-1 < a < 0$ , (d)  $a < -1$ 

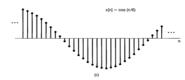
### Discrete-Time Sinusoidal Signals

Sinusoidal signal: 
$$x[n] = A\cos(\omega_0 n + \phi)$$

As 
$$e^{j\omega_0 n} = \cos(\omega_0 n) + j\sin(\omega_0 n)$$
, we have  $A\cos(\omega_0 n + \phi) = \frac{A}{2}e^{j\phi}e^{j\omega_0 n} + \frac{A}{2}e^{-j\phi}e^{-j\omega_0 n}$ 



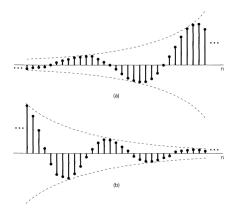




### Discrete-Time General Complex Exponential

$$C = |C|e^{j\theta}, \alpha = |\alpha|e^{j\omega_0}$$

$$C\alpha^n = |C||\alpha|^n \cos(\omega_0 n + \theta) + j|C||\alpha|^n \sin(\omega_0 n + \theta)$$



# Periodic Properties of Discrete-time Complex Exponential

For continuous-time complex exponential  $x(t) = e^{j\omega_0 t}$ 

- (1) the larger the  $\omega_0$ , the higher the rate of oscillation
- (2)  $e^{j\omega_0t}$  is periodic for any value of  $\omega_0$

Are the above two statements still valid for the discrete case

$$x[n] = e^{j\omega_0 n}$$
?

(1) oscillation: 
$$e^{j(\omega_0+2\pi)n}=e^{j2\pi n}e^{j\omega_0n}=e^{j\omega_0n}$$

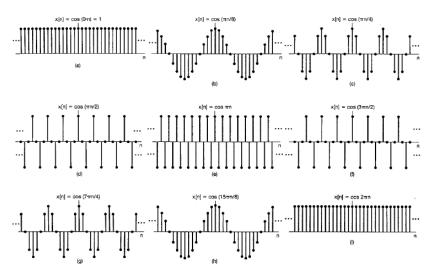


Figure 1.27 Discrete-time sinusoidal sequences for several different frequencies.

## Periodic Properties of Discrete-time Complex Exponential

- (1) For  $\omega_0$  within an interval  $0 \le \omega_0 \le 2\pi$ , the frequency  $\uparrow$  as  $\omega_0 \uparrow$  for  $0 \le \omega_0 \le \pi$ , the frequency  $\downarrow$  as  $\omega_0 \uparrow$  for  $\pi \le \omega_0 \le 2\pi$ .
- (2)  $e^{j\omega_0 n}$  might be non-periodic:

$$e^{j\omega_0(n+N)}=e^{j\omega_0n}$$

implies  $e^{j\omega_0N}=1$ , hence  $\omega_0N$  needs to be a multiple of  $2\pi$ :

$$\frac{\omega_0}{2\pi} = \frac{m}{N}$$
 is rational.

Fundamental frequency of the periodic signal  $e^{j\omega_0 n}$  is  $\frac{2\pi}{N}=\frac{\omega_0}{m}$  Fundamental period is  $N=m\left(\frac{2\pi}{\omega_0}\right)$ 

#### Example - Fundamental Period

What is the fundamental period of the discrete-time signal

$$x[n] = e^{j(2\pi/3)n} + e^{j(3\pi/4)n}$$

The fundamental period of the first term is 3;

The fundamental period of the second term is 8;

The fundamental period of the entire signal is 24.

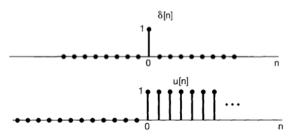
$e^{j\omega_0 t}$	$e^{j\omega_0 n}$
Distinct signals for distinct values of $\omega_0$	Identical signals for values of $\omega_0$ separated by multiples of $2\pi$
Periodic for any choice of $\omega_0$	Periodic only if $\omega_0 = 2\pi m/N$ for some integers $N > 0$ and $m$ .
Fundamental frequency $\omega_0$	Fundamental frequency* $\omega_0/m$
Fundamental period $\omega_0 = 0$ : undefined $\omega_0 \neq 0$ : $\frac{2\pi}{\omega_0}$	Fundamental period* $\omega_0 = 0$ : undefined $\omega_0 \neq 0$ : $m\left(\frac{2\pi}{\omega_0}\right)$

<sup>\*</sup>Assumes that m and N do not have any factors in common.

## Discrete Time Unit Step and Unit Impulse Sequence

Unit Impulse function  $\delta[n]$ , and Unit Step function u[n]:

$$\delta[n] = \begin{cases} 0, & n \neq 0 \\ 1, & n = 0 \end{cases}, \qquad u[n] = \begin{cases} 0, & n < 0 \\ 1, & n \ge 0 \end{cases}$$

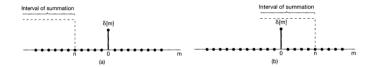


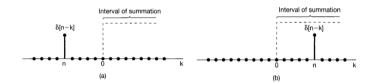
Note: u[n] at n = 0 is defined.

#### Discrete Time Unit Step and Unit Impulse Sequence cont.

Unit impulse: first difference of unit step  $\delta[n] = u[n] - u[n-1]$ Unit step: running sum of unit impulse

$$u[n] = \sum_{m=-\infty}^{n} \delta[m] = \sum_{k=0}^{+\infty} \delta[n-k]$$



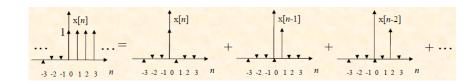


### Discrete Time Unit Step and Unit Impulse Sequence cont.

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Running sum of unit sample: superposition of delayed impulses

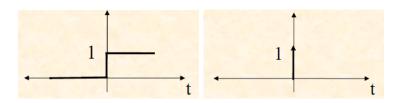
$$u[n] = \sum_{k=0}^{+\infty} \delta[n-k]$$



#### Unit Step and Unit Impulse Function

Unit Step function u(t), and Unit Impulse function  $\delta(t)$ :

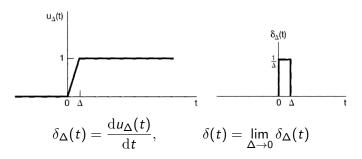
$$u(t) = \begin{cases} 0, & t < 0 \\ 1, & t > 0 \end{cases}, \qquad \delta(t) = \frac{d}{dt}u(t).$$



Running integral : 
$$u(t) = \int_{-\infty}^{t} \delta(\tau) d\tau$$
.

A system can be characterized by its unit step response or unit impulse response.

Unit step is discontinuous at t = 0. Unit step can be approximated as  $u_{\Delta}(t)$ 



Note that  $\delta_{\Lambda}(t)$  is a short pulse, with duration  $\Delta$  and with unit area for any value of  $\Delta$ .

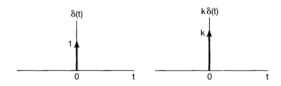
#### Unit Step and Unit Impulse Function cont.

 $\delta(t)$  has no duration but unit area.

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The arrow at t = 0 indicates that the area of the pulse is concentrated at t=0.

The height of the arrow and the "1" next to the arrow are used to represent the area of the impulse.

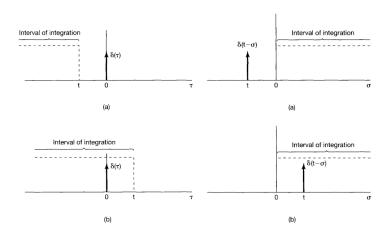


 $k\delta(t)$  has an area of k

$$\int_{-\infty}^{t} k\delta(\tau) d\tau = ku(t)$$

#### Unit Step and Unit Impulse Function cont.

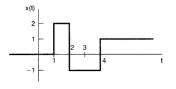
Running integral 
$$u(t) = \int_{-\infty}^{t} \delta(\tau) d\tau = \int_{0}^{\infty} \delta(t - \sigma) d\sigma$$

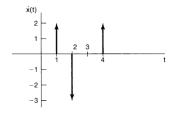


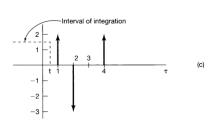
#### Unit Step and Unit Impulse Function cont.

(a)

(b)







#### Overview of System

System: any process that results in the transformation of signal



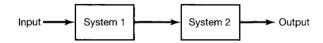
A system is continuous-time if both input and output are continuous-time.

A system is discrete-time if both input and output are discrete-time.

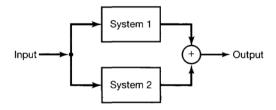
#### Interconnection of Systems

Lecture 01

(1) Cascade (Series): the output of System 1 is the input of System 2.

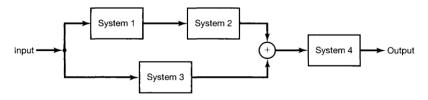


(2) Parallel: the same input is applied to Systems 1 and 2; the final output is the sum of the outputs of Systems 1 and 2.

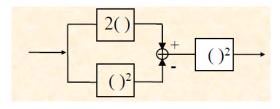


## Interconnection of Systems cont.

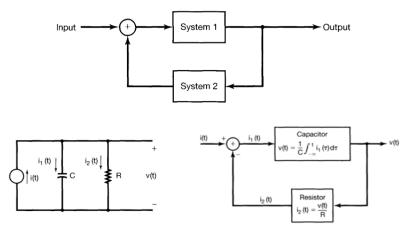
#### (3) Series/Parallel



Ex. 
$$y[n] = (2x[n] - x[n]^2)^2$$



(4) Feedback: Output of System 1 is the input to System 2; Output of System 2 is fed back and added to the external input to produce the actual input to System 1.



#### Properties of System

- Memory/Memoryless
- Invertibility and Inverse System
- Causality
- Stability
- Time-invariance
- Linearity

## (1). Memory and Memoryless

Lecture 01

A system is memoryless if the output only depends on input at the same time.

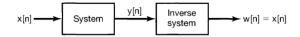
Ex. (1) 
$$y[n] = (2x[n] - x^2[n])^2$$
.

- (2) Resister is a memoryless component: v(t) = Ri(t).
- (3) With memory: e.g.  $y[n] = \sum_{k=-\infty}^{n} x[k]$ .
- (4) Delay system: y[n] = x[n-1].
- (5) Capacitor:  $y(t) = \frac{1}{C} \int_{-\infty}^{t} x(\tau) d\tau$ .

## (2). Invertibility and Inverse System

Lecture 01

A system is invertible if distinct inputs lead to distinct outputs.



If w[n] = x[n], then system 2 is the inverse system of system 1.

E.g. 
$$y(t) = 2x(t)$$
, then  $w(t) = 0.5y(t)$ 

$$y[n] = \sum_{k=-\infty}^{n} x[k], \text{ then } w[n] = y[n] - y[n-1]$$

$$x(t) \longrightarrow y(t) = 2x(t) \longrightarrow w(t) = \frac{1}{2}y(t) \longrightarrow w[t] = x(t)$$

$$x[n] \longrightarrow y[n] = \sum_{k=-\infty}^{n} x[k] \longrightarrow w[n] = y[n] - y[n-1] \longrightarrow w[n] = x[n]$$

A system is causal if the output at any time only depends on the input at the present time and before.

E.g. 
$$y[n] = x[n] - x[n-1]$$
: causal  $y(t) = x(t+1)$ : non-causal

Note: All memoryless are causal

Causal property is more important for real-time processing.

But for some applications, such as image-processing, no need to process the data causally.

$$y[n] = \frac{1}{2M+1} \sum_{k=-M}^{M} x[n-k].$$

#### Examples

1. 
$$y[n] = x[-n]$$

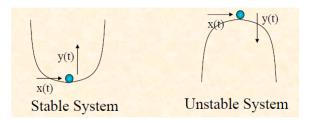
2. 
$$y(t) = x(t)\cos(t+1)$$

## (4). Stability

A system is stable if bounded input gives bounded output.

#### BIBO stable

E.g. x(t): the horizontal force; y(t): vertical displacement



#### Examples

1. 
$$y(t) = tx(t)$$

2. 
$$y(t) = e^{x(t)}$$

## (4). Stability cont.

Lecture 01

$$y(t) = \frac{d}{dt}x(t)$$
 is not stable

$$\mathsf{Let}\ \mathsf{x}(t) = \begin{cases} \sqrt{t}, & t \in [0,1] \\ -\sqrt{-t}, & t \in [-1,0] \\ \mathsf{bounded}\ \mathsf{and}\ \mathsf{with}\ \mathsf{proper}\ \mathsf{derivatives}, & \mathsf{otherwise} \end{cases}$$

Then  $y(t) \to \infty$  as  $t \to 0$ .

A system is time-invariant if a time shift in the input only causes a time shift in the output.

i.e. If 
$$x[n] o y[n]$$
, then  $x[n-n_0] o y[n-n_0]$ 

Ex. 
$$y(t) = \sin(x(t))$$

Let 
$$y_1(t) = sin(x_1(t))$$
,  $x_2(t) = x_1(t - t_0)$   
Then  $y_2(t) = sin(x_2(t)) = sin(x_1(t - t_0)) = y_1(t - t_0)$ 

Hence time-invariant (T.I.)

#### (5). Time-invariance cont.

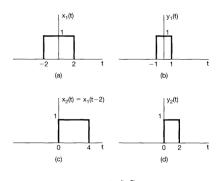
Ex. 
$$y[n] = nx[n]$$

Lecture 01

Let 
$$y_1[n] = nx_1[n]$$
,  $x_2[n] = x_1[n - n_0]$   
Then  $y_2[n] = nx_2[n] = nx_1[n - n_0]$   
However  $y_1[n - n_0] = (n - n_0)x_1[n - n_0] \neq y_2[n]$ 

Hence, y[n] is not time-invariant (T.I.)

# Consider the system y(t) = x(2t)



A system is linear if

- 1. the response to  $x_1(t) + x_2(t)$  is  $y_1(t) + y_2(t)$ 
  - additivity
- 2. the response to  $a \cdot x_1(t)$  is  $a \cdot y_1(t)$ , where a is any complex constant.
  - scaling

Combine the above two properties, we can conclude

$$ax_1(t) + bx_2(t) \implies ay_1(t) + by_2(t)$$

superposition property

For discrete-time:  $ax_1[n] + bx_2[n] \implies ay_1[n] + by_2[n]$ 

## (6). Linearity cont.

If linear, zero input gives zero output.

Q: Is 
$$y[n] = 2x[n] + 3$$
 linear?

A: No, because it violates zero-in zero-out property.

However, this system is an "incremental linear system": difference of output is a linear function of difference of input.

$$y_1[n] - y_2[n] = 2x_1[n] + 3 - (2x_2[n] + 3) = 2(x_1[n] - x_2[n])$$

#### Exercise

Lecture 01

- (1).  $y[n] = x^2[n]$ non-linear; time-invariant
- (2). y[n] = nx[n]memoryless; causal; linear; not time-invariant; not stable

(3). 
$$y(t) = x(\sin(t))$$

(4). 
$$y(t) = \frac{d}{dt}x(t)$$

(5). 
$$y(t) = \int_{-\infty}^{2t} x(\tau) d\tau$$

(6). 
$$y(t) = x(t-2) + x(2-t)$$

(7). 
$$y(t) = x(t)\cos(8t)$$

(8). 
$$y[n] = x[3n]$$

(9). 
$$y(t) = x(\frac{t}{3})$$

(10) 
$$y[n] = x[n-2] - 2x[n-8]$$

$$(11) y[n] = x[4n+1]$$

Lecture 01

- Ontinuous-time and discrete-time signals
- 2 Transformation of independent variables
- Omplex exponential and sinusoidal signals
- Unit impulse and unit step functions
- Interconnection of systems