



Politecnico di Torino

Microelectronic Systems

DLX Microprocessor: Design & Development

Final Project Report

Master degree in Electronics Engineering

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Contents

1	Presentation	1
2	Datapath	2
2.1	Fetch Stage	2
2.1.1	Introduction	2
2.1.2	Main units	3
2.2	Decode Stage	3
2.2.1	Introduction	3
2.3	Execution Stage	4
2.3.1	Overview	4
2.3.2	Adder	5
2.3.3	Logic	7
2.3.4	Shifter	8
2.3.5	Logic comparison	8
2.3.6	Multiplier	9
2.3.7	PC adder	9
2.4	Memory Stage	9
2.4.1	Embedded part	10
2.4.2	External RAM	10
2.5	Write Back	11
3	Additional features	13
3.1	Data forwarding	13
3.2	Dynamic branch prediction	13
4	Control Unit	15
5	Synthesis	16
A	Shifter behavioural VHDL	17

Presentation

CHAPTER 2

Datapath

2.1 Fetch Stage

2.1.1 Introduction

The fetch stage is important to provide to our processor a new instruction at every clock cycle, if needed. It is the first stage that a function actually enters in its execution flow.

Here are presented the signals involved:

- *CLK* is the usual clock signal needed to synchronize the system
- *STALL* is a control signal, that blocks the update of the *PC* and therefore does not let any other operations to be issued
- *RST* it's the canonical reset signal
- *RST_DEC* **ALBI PLEASE ADD EXPLANATION HERE**
- *PC_SEL* is a control signal which selects the next value of the *PC*, either from the normal execution flow or from the *ALU*
- *JB_INST* is the computed new *PC* value coming from the *ALU*
- *FUNC* is a part of the instruction, for the R_TYPE operations
- *OPCODE* represents the type of operation in the instruction set to be performed by the processor
- *NPC* which is the canonical updated *PC*, pointing to the next instruction
- *INST_OUT* is a signal that comes from the *IRAM*, after the instruction has been read
- *MISS_HIT* is used to inform the system whether the prediction made was correct or not

- *FLUSH_CTL* is a control signal which allows us to clean the pipe in case of a wrong prediction

2.1.2 Main units

There are several instances we decided to use in this stage. The main ones are reported below.

Branch predictor The purpose of this unit is to give a reliable dynamic prediction regarding the behaviour of a branch before its correct target address is computed. More details can be found in a dedicated section 3.2.

IRAM It is the instruction memory, where the operations that have to be executed are stored. It is addressed by the *PC*.

Registers There are several registers that are actually used to store the values of:

- *PC* counter, current and next value
- *INSTR* which saves the instruction fetched from the IRAM
- *TMP_RST*, where the reset signal is sampled in order to synchronize the reset phase

Mux This unit is used here to selected the correct value of the next *PC*: either the one coming from the memory or the following instruction, following the normal execution flow.

Adder This very simple unit just provides the +4 value for the new *PC*.

2.2 Decode Stage

2.2.1 Introduction

The decode stage is the second one. Here the instructions stored in the *IRAM* that have been previously read are decomposed in order to correctly identify all the information regarding the operation that will be performed. In this way, we can read the correct source, target and destination address, as well as the immediate values. Moreover, also the *OPCODE*, which represents the particular instruction to be executed, is read.

The most relevant signals involved in this stage are:

- *FLUSH* is a control signal, which is necessary to clean the pipe
-
-

- **shifter**, used in order to shift and rotate the value contained in an input register
- **adder**, necessary obviously to perform addition and subtraction between values, but also used in the logic comparison operations
- **logic**, unit to perform logic bitwise operations, such as *OR*, *AND*, *XOR* and so on
- **comparator**, which purpose is to determine whether an input is greater, smaller or equal than another one
- **multiplier**, whose goal is to perform multiplications between inputs
- **PC adder**, a dedicated adder to increase by the value of the PC, useful in case of a *JMP* operation
- **registers**, to update the inputs and store the results
- **muxes**, to select the results coming from the required unit and correctly update the outputs of the stage
- **inverter**, needed to perform the *2'S complement* and perform the subtraction

For this stage, a bunch of bits of the *CW* are used. Here are reported:

- *ENEX*, a general enable for the components and especially for the registers
- *MUX1_SEL*, a selection signal for the mux that is able to select an immediate value or a register an input that as the first term for the other units
- *MUX2_SEL*, with a goal similar as the one above, but for the second input
- *UN_SEL*, 3 bits signal that selects the correct input, which is then sent to a register
- *OP_SEL*, 4 bits selection signal, used to indicate to each unit, with different encoding, the operation to be performed
- *PC_SEL*, selection signal to update the PC counter correctly, following a branch or jump.

2.3.2 Adder

For the adder architecture, we decided not to start from scratches. As long as our P4 adder, designed during the laboratories, has a general description, we resized it to our architecture on 32 bits. This means that we have included the components capable of producing the PG propagate and generate signals and the G ones used to produce only the generate and so the final carry. We consider that therefore the result, in terms of performance can satisfy us. At the input of the block, we have also a *C_IN* signal, which is used to create, in cooperation with the inverter, the *2'S complement* of the input used (either immediate or coming from a register), in order to produce as a final result the signed sum. There are few control signals used for this unit. In figure 2.1 is presented the simple schematic of the adder block.

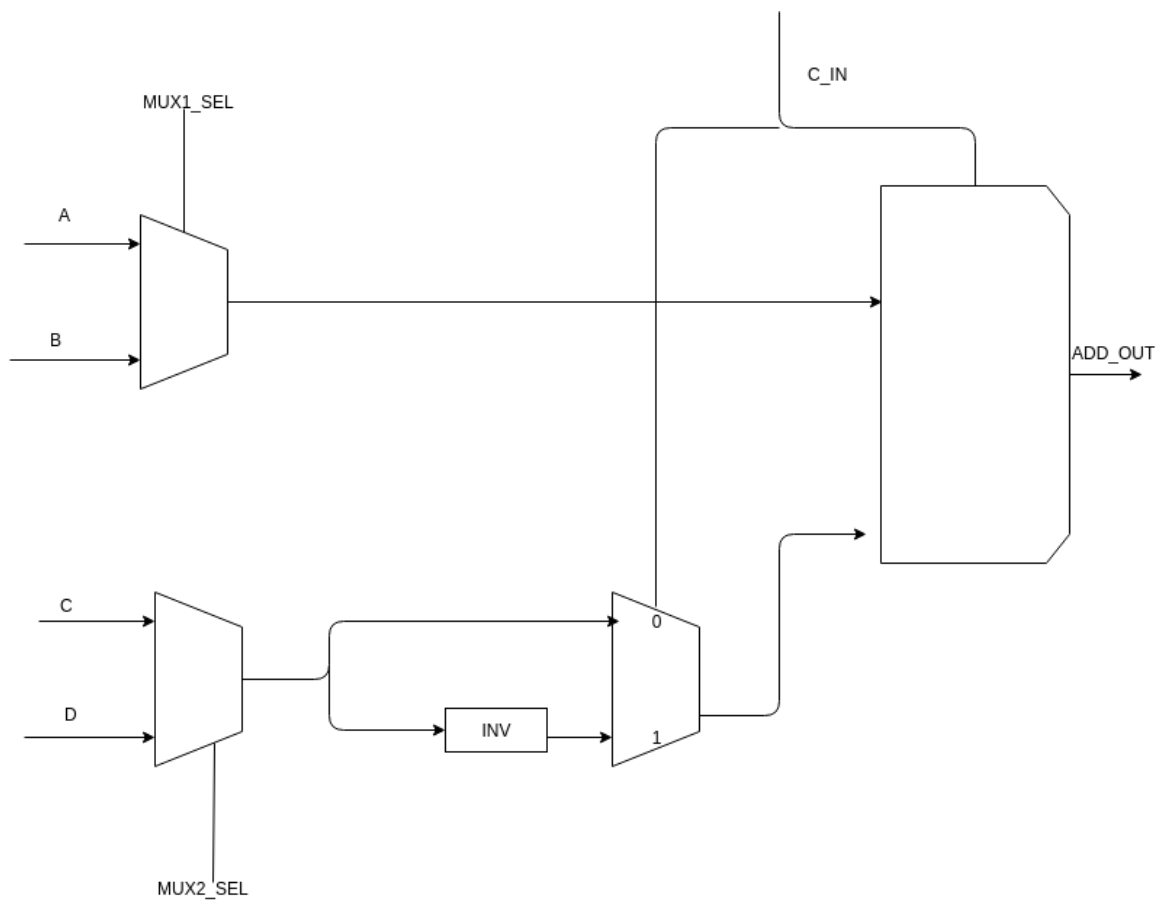


Figure 2.1: Adder with inverter and related selection signal

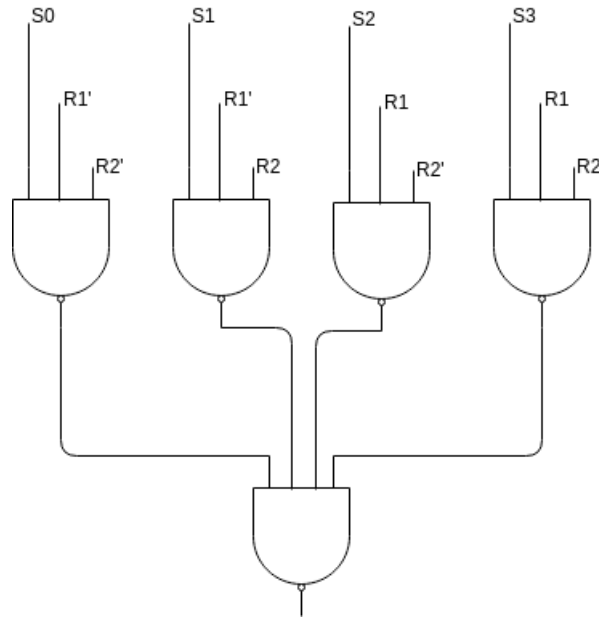


Figure 2.2: T2 logic for a single bit

2.3.3 Logic

In order to perform some logic bitwise operations, we have reproduced the microarchitecture that has been used for the T2. Here, in figure 2.2 is briefly reported the gates needed, that combined between each other, are capable of giving us the desired operations, as reported in the table 2.1.

	S0	S1	S2	S3
AND	0	0	0	1
NAND	1	1	1	0
OR	0	1	1	0
NOR	1	0	0	0
XOR	0	1	1	0
XNOR	1	0	0	1

Table 2.1: Combination of control signal for logic operations

The control signal S is required to be on 4 bits: for this reason, we have implemented in our control word 4 bits for this purpose. In particular, the signal `OP_SEL` plays this role. We could have used 3 bits on the CW and design an encoder, but the savings in term of connection would have been ruined by the need of an additional decoder. Moreover, these bits are not switched most of the time, thus providing a not so relevant switching power dissipation.

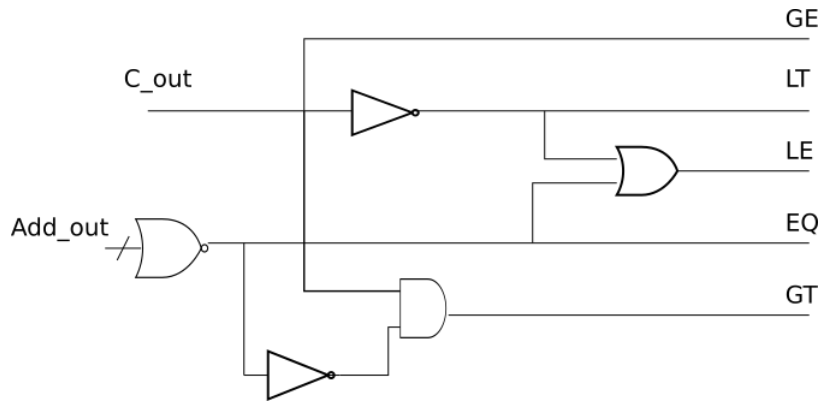


Figure 2.3: Circuit for comparison - Unsigned

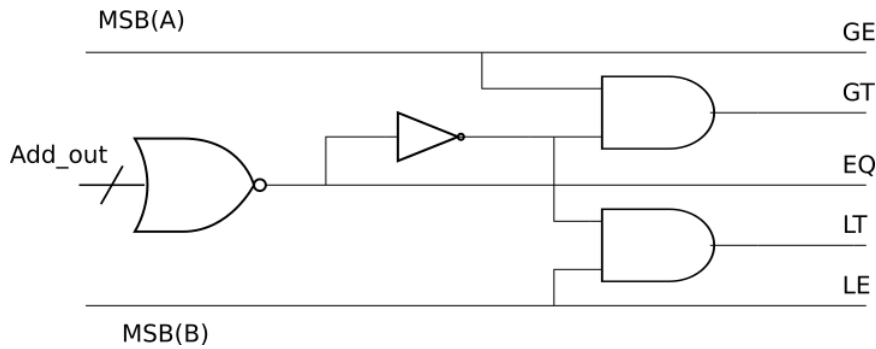


Figure 2.4: Comparison for signed case - different MSBs

2.3.4 Shifter

To define the lateral shifts, left or right, and the rotation, we have simply provided a behavioral description of the circuit, as reported in appendix A. It generates a barrel shifter, thus providing the needed operation between the ones possible, and moving the bits of the first input by the quantity that is introduced by the second input.

2.3.5 Logic comparison

The logic comparison requires the output of the subtraction between the two values to be compared. So, *ADD_OUT*, output of the adder, is here used as an input. Then, all the bits of the result are *NOR_reduced*. This allows us to see if all the bits are equal to 0, i.e. the two operands are equal. In this case, an additional signal *US* is used, to differentiate between an unsigned or a signed comparison. While in the first case the schematic is the one presented in figure 2.3, for the signed case we have to differentiate. As a matter of fact, if the first digit is equal, the two numbers have the same sign, so we can use the same HW presented before. Differently, if the two bits differ, we use what is reported in 2.4. The equivalence of the two MSBs is performed through a XNOR gate.

Followingly, the value corresponding to the logic comparison needed is chosen by a mux,

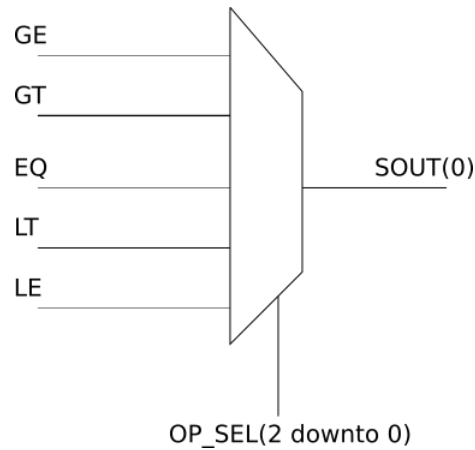


Figure 2.5: Mux to select the required condition

whose selection signal is represented by *OP_SEL*, and assigned to the lowest bit of the output signal *SOUT*, as shown in figure 2.5. All the other bits of this signal are independently set to 0.

2.3.6 Multiplier

For our multiplier we again reused the *Booth* architecture that has been designed for the laboratories. As said, the multiplication is usually supposed to be divided in more than one stage, but here it works on a single one, thus reducing the operating frequency. However, we decided to include this component in order to include the MUL in our instruction set.

2.3.7 PC adder

In order to perform correctly some jump instruction, the temporary PC needs a +4, which, due to the design of our datapath, can not be performed by the adder. An additional adder, reported in figure 2.6 is in charge of this specific operation.

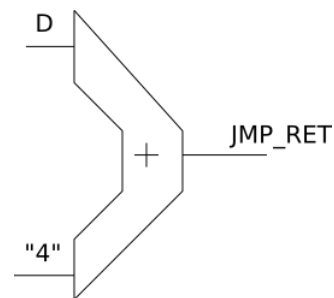


Figure 2.6: PC adder

2.4 Memory Stage

This stage can be divided in two major parts: the registers that contains the addresses and the data to be written in the memory or read from the memory and the RAM. The former is actually embedded in the processor, while the latter one is actually external, but provides an interface and some signals to communicate with the processor itself.

Reg Name	# bits	Input	Output
<code>exec_reg</code>	32	FROM_ALU	ALU_OUT
<code>mem_reg</code>	32	FROM_MEM	MEM_OUT
<code>dest_reg</code>	5	DEST_IN	DEST_OUT

Table 2.2: Table with registers of Memory Stage

2.4.1 Embedded part

The part included in the datapath is mainly made up of three different registers, which are used to store different values coming from different units:

- *dest_reg*, which is used to save the address of the commit register, to be passed to the following write back stage
- *mem_reg*, where the data coming from the external memory presented in 2.4.2 are stored
- *exec_reg*, which is used to save the address of the commit register, to be passed to the following write back stage

All these registers share the same reset, clock and enable signals. They differ for the input/output ones, which are summarized in table 2.2.

2.4.2 External RAM

The external RAM receives signals both from the datapath and from the CU. An overview can be seen in figure 2.7. This part is external since it will not be synthesized. In particular, here are presented the most relevant ones:

- *ENABLE*, *RW* are signals used to control the writing on the memory. They should be both high to complete a write
- *D_TYPE* specifies the kind of data we are reading or writing in/from memory, i.e byte, half-word, word
- *US* defines a signed/unsigned representation for the data; it is relevant to correctly extend the data on 32 bits
- *ADDRESS* obviously used to correctly point to a precise memory location
- *MEMIN* is the signal containing values to be eventually stored in memory
- *MEMOUT* is the output signal which contains the data read from the memory

While the writing is performed synchronously only if the *RW* signal is high, there is always a value in correspondence of the output signal. This means that a read always happens, and a value is written every time on the *MEMOUT* signal. This value will be eventually discarded in the following stages.

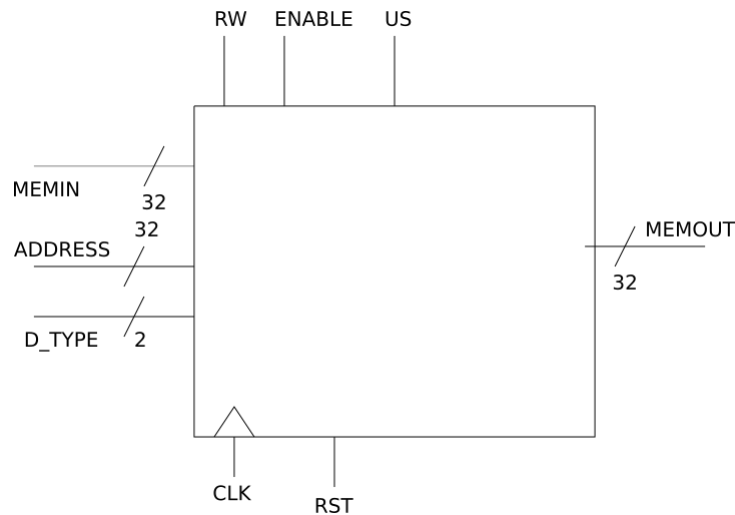


Figure 2.7: Interface for external RAM

2.5 Write Back

The *Write Back* stage is needed to write in the registers the final result obtained after the execution and the access to memory. With this goal, an destination address is needed, as well as some muxes to select the right signal to be sent back to the RF. The overall structure is presented in figure 2.8. Few signals are needed in the entire unit:

- **MEM_ALU_SEL**, selection signal coming from the control word, controlling the output mux
- **DEST_IN**, destination address from the memory stage
- **FROM_ALU**, coming directly from the execution stage
- **FROM_MEM**, that comes from the memory, after a read operation has been performed
- **DATA_OUT**, which contains the data to be written on the Rf
- **DEST_OUT**, destination address to the RF

To simply explain the behavior, one can say that the destination address is contained in the 5 bits of *DEST_IN*, which are passed to *DEST_OUT*. This output signal is then connected to the RF, where the *WR* signal enables the writing. The actual value that is written on memory is chosen between the outputs from memory and ALU, and ultimately written to close the pipeline and complete the instruction, which is finally committed.

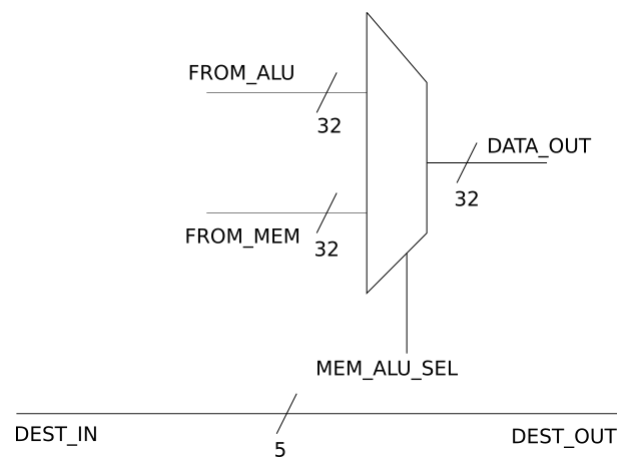


Figure 2.8: Mux to choose the required output

CHAPTER 3

Additional features

3.1 Data forwarding

In order to reduce the number of required stalls in the pipeline and thus improving the efficiency and reducing the average latency of our processor, we have implemented a simple form of data forwarding. It is important to notice that, after an initial phase of testing without this capability, some additional signals and components have been added to connect some data at different stages of the pipeline. In particular, these are the following:

- between *MEM* and *EX* stage
- between *WB* and *EX* stage

Each of these signals goes through a different multiplexer, which select either this data or the data coming from the previous stage. This selection is made by checking the destination and the source registers of two following instructions. When we have that

$$R_{dest}^{i-1} = R_{source}^i$$

then the signal coming from the subsequent stage is selected.

3.2 Dynamic branch prediction

In order not to waste too much clock cycles waiting for the correct value of the *PC* register, we have designed a unit that provides a dynamic prediction regarding a branch instruction. In particular, it is possible, with a defined value of the *PC* corresponding to a branch, to get whether is a better speculation to take or not to take the branch, while waiting for the correct *PC*, which comes after the execution stage. There are actually 16 possible entries in our branch. Each of them is composed by a part of the *PC* and a prediction state, represented on two bits, as reported in table 3.1. Every time a new instruction is issued, this entity checks whether it is a branch type or not. If so, the bits from 5 to 2 of the *PC* related to

	00	01	10	11
Encoding	Strong NT	Weak NT	Weak T	Strong T
NS when HIT	00	00	11	11
NS when MISS	01	10	01	10

Table 3.1: Possible states for predictor

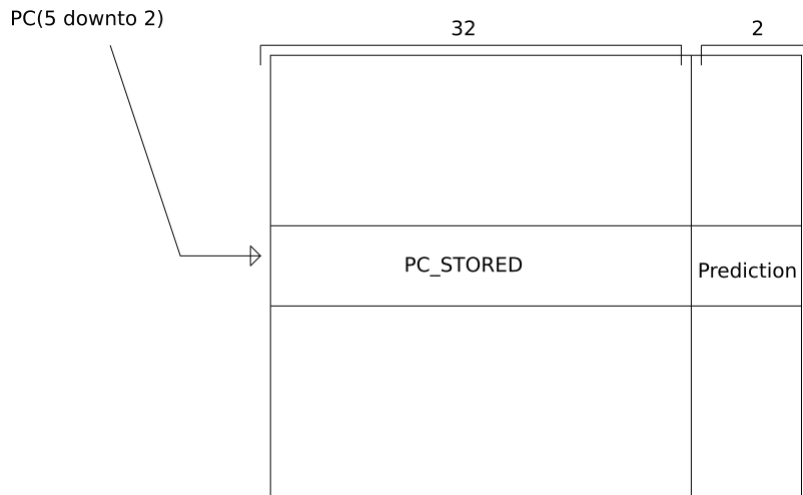


Figure 3.1: Schematic for BHT

the instruction are taken into account as address to the table, as shown in figure 3.1. At this point, the content of this location is checked:

- if the value stored corresponds to the *PC*, then the prediction contained is the one related to this current instruction, and therefore it can be freely used
- if not, the previously stored value is overwritten, as long as its prediction: the current prediction is set to a "weakly not taken case"

CHAPTER 4

Control Unit

CHAPTER 5

Synthesis

APPENDIX A

Shifter behavioural VHDL

```
library IEEE;
use IEEE.std_logic_1164.all;
use IEEE.numeric_std.all;
use IEEE.std_logic_unsigned.all;

entity SHIFTER is
  generic (NB: integer := 32;
           LS: integer:= 5
           );
  port (      FUNC:                IN std_logic_vector(1
        downto 0);
           US:                    IN std_logic;
        DATA1:                  IN std_logic_vector(NB-1
        downto 0);
        DATA2:                  IN std_logic_vector(LS-1
        downto 0);
        OUTSHFT:                 OUT std_logic_vector(NB-1
        downto 0)
        );
end SHIFTER;

architecture BEHAVIOR of SHIFTER is

begin
```

```
P_ALU: process (FUNC, US, DATA1, DATA2)
begin
```

```
    if(US='1') then
```

```
        case FUNC is
```

```
            — SLL
```

```
            when "00"
```

```
=>
```

```
OUTSHIFT <=
```

```
    std_logic_vector
    ( shift_left (
        unsigned(DATA1),
        to_integer (
            unsigned(DATA2))
        ));
```

```
            — SRL
```

```
            when "01"
```

```
=>
```

```
OUTSHIFT <=
```

```
    std_logic_vector
    ( shift_right (
        unsigned(DATA1),
        to_integer (
            unsigned(DATA2))
        ));
```

```
            — ROL
```

```
            when "10"
```

```
=>
```

```
OUTSHIFT <=
```

```
    std_logic_vector
    ( rotate_left (
        unsigned(DATA1),
        to_integer (
            unsigned(DATA2))
        ));
```

```
            — ROR
```

```
            when "11"
```

```
=>
```

```
OUTSHIFT <=
```

```
    std_logic_vector
    ( rotate_right (
        unsigned(DATA1),
        to_integer (
            unsigned(DATA2))
        ));
```

```

        when others =>
            OUTSHFT <= (others =>'0');
    end case;
else
    case FUNC is
        — SLL
        when "00"
            =>
                OUTSHFT <=
                    std_logic_vector
                    ( shift_left (
                        signed(DATA1) ,
                        to_integer (
                            unsigned(DATA2) )
                        ) );

        — SRL
        when "01"
            =>
                OUTSHFT <=
                    std_logic_vector
                    ( shift_right (
                        signed(DATA1) ,
                        to_integer (
                            unsigned(DATA2) )
                        ) );

        — ROL
        when "10"
            =>
                OUTSHFT <=
                    std_logic_vector
                    ( rotate_left (
                        signed(DATA1) ,
                        to_integer (
                            unsigned(DATA2) )
                        ) );

        — ROR
        when "11"
            =>
                OUTSHFT <=
                    std_logic_vector
                    ( rotate_right (
                        signed(DATA1) ,
                        to_integer (
                            unsigned(DATA2) )
                    ) );
    end case;
end if;

```

```
        ));  
        when others =>  
            OUTSHIFT <= (others => '0');  
        end case;  
    end if;  
end process P_ALU;  
  
end BEHAVIOR;
```