

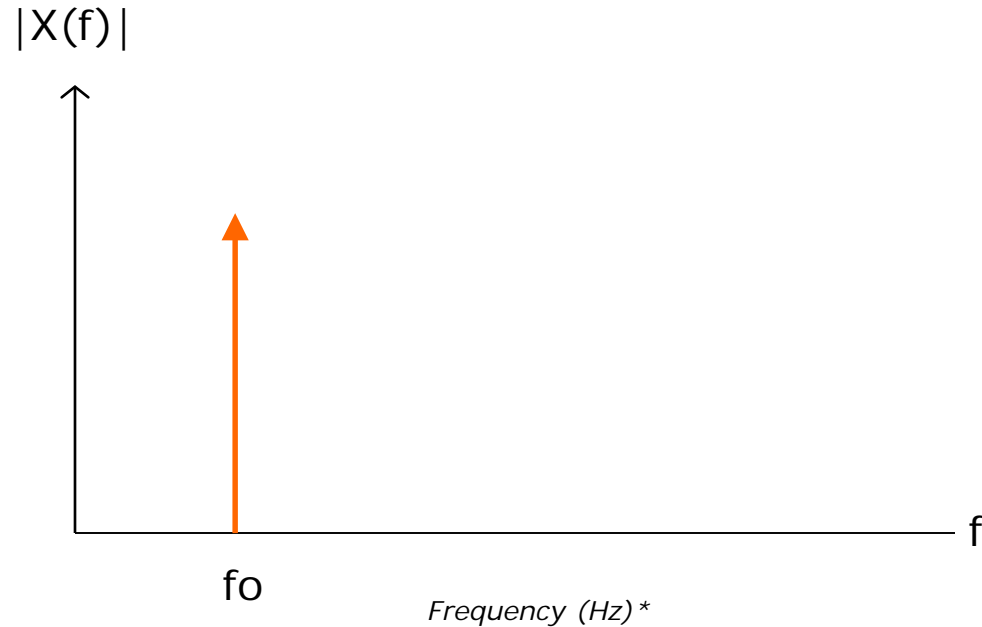
03 Speech Signal Fundamentals II

Deviation from Ideal Spectra

$$x(t) = \cos(2\pi f_0 t)$$



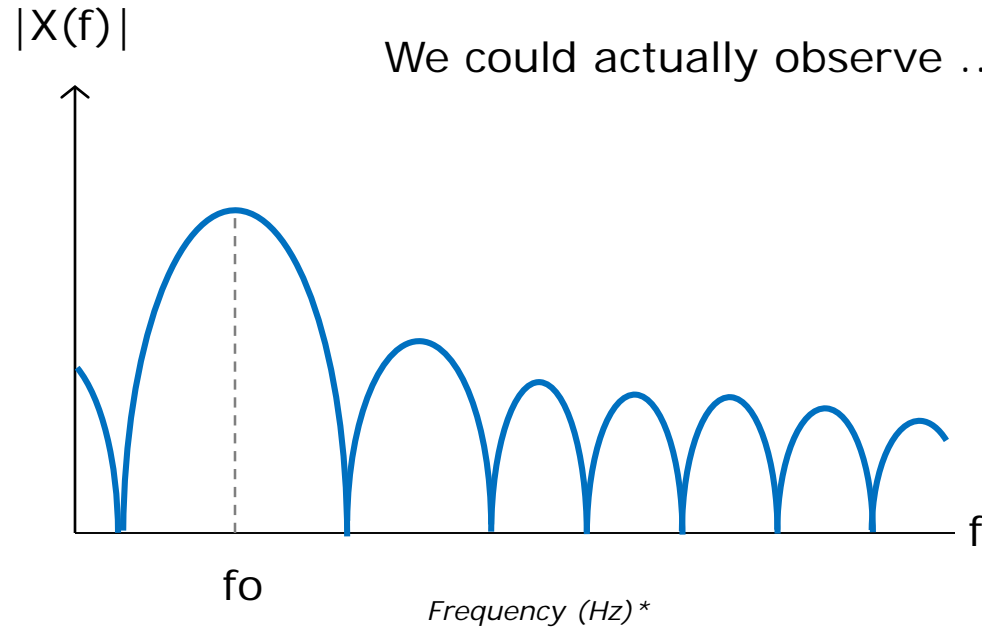
$x(t)$ has infinite length.



* Negative frequencies omitted

Deviation from Ideal Spectra

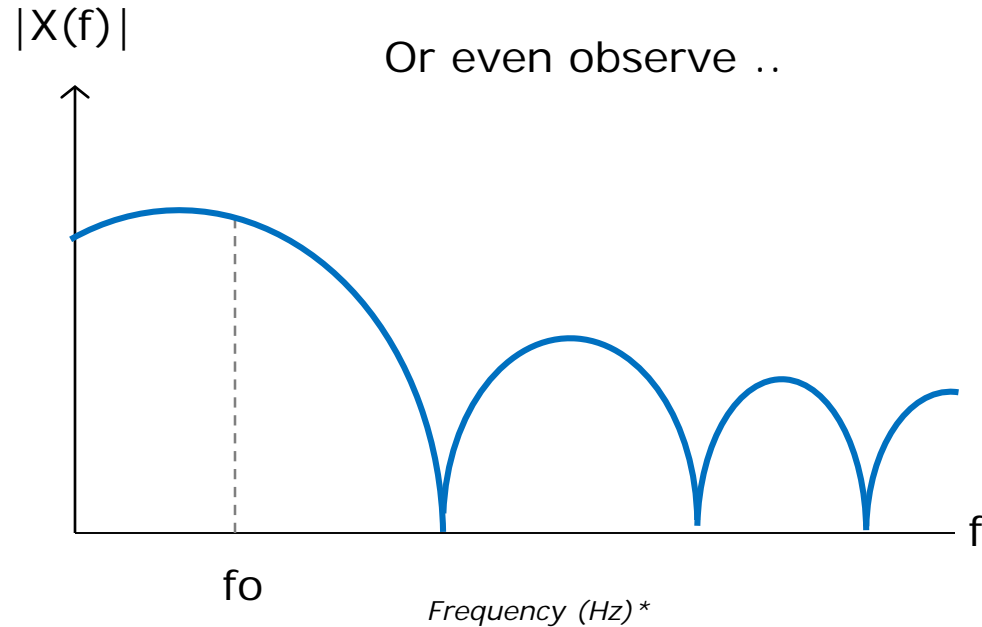
$$x(t) = \cos(2\pi f_0 t)$$



* Negative frequencies omitted

Deviation from Ideal Spectra

$$x(t) = \cos(2\pi f_0 t)$$



* Negative frequencies omitted



Photo by Paul on Unsplash

Finite-length = Time-windowed

$$x(t) = \cos(2\pi f_0 t) \quad \longrightarrow \quad x(t) \text{ has infinite length.}$$

But in real life, we can never have a signal of infinite length.

A portion of sine wave with finite length can be considered as a windowed version of the infinite length sine wave.

E.g.: A sine wave from t_1 to t_2 , $x_w(t)$ can be written as:

$$x_w(t) = \cos(2\pi f_0 t) w(t)$$

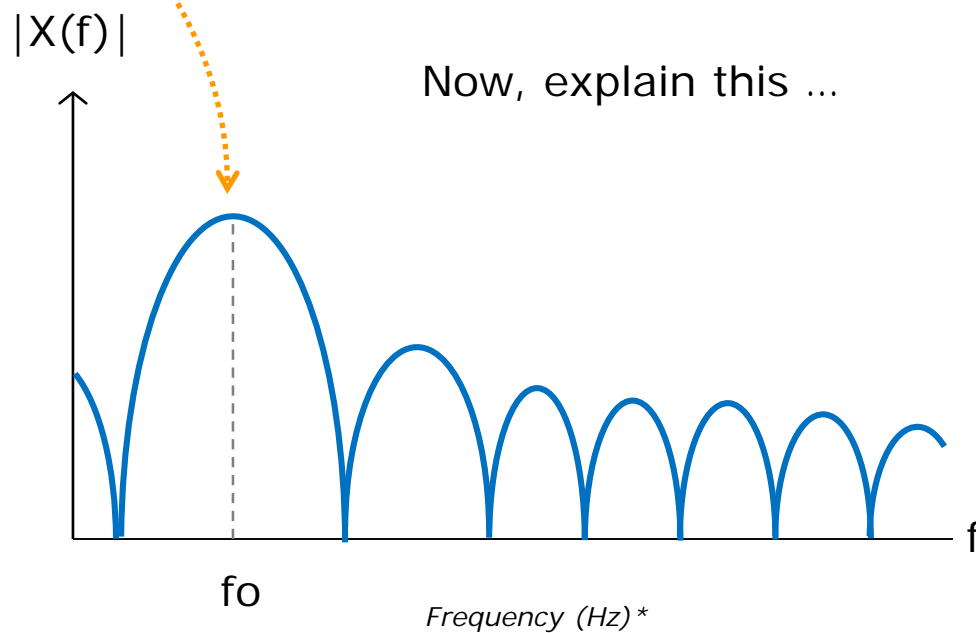
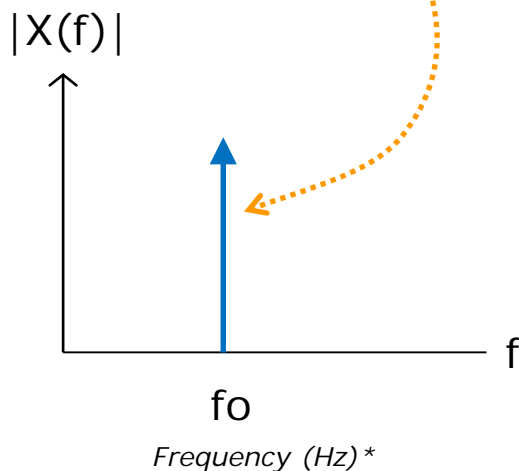
Where:

$$w(t) = \begin{cases} 1 & ; t_1 \leq t < t_2 \\ 0 & ; otherwise \end{cases}$$

Recall that ...

$$x(t)y(t) \longleftrightarrow X(f) * Y(f)$$

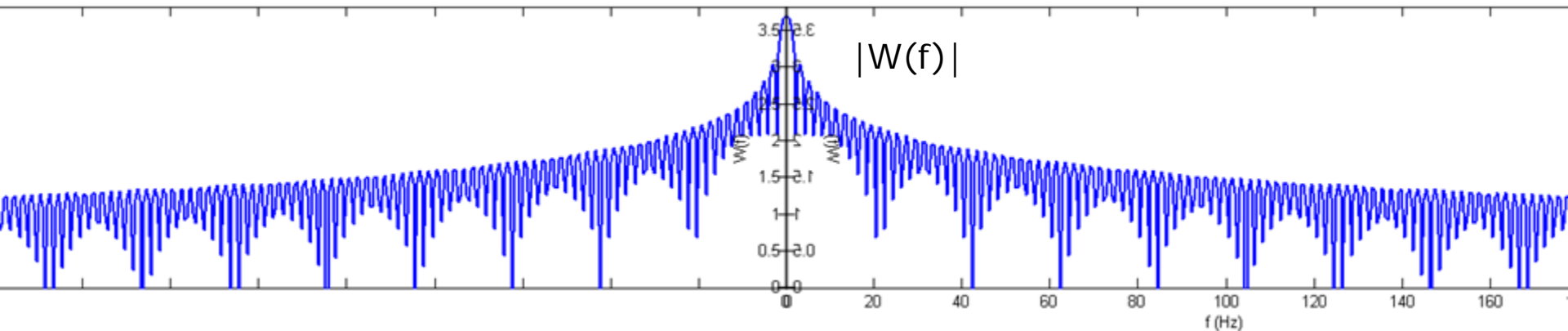
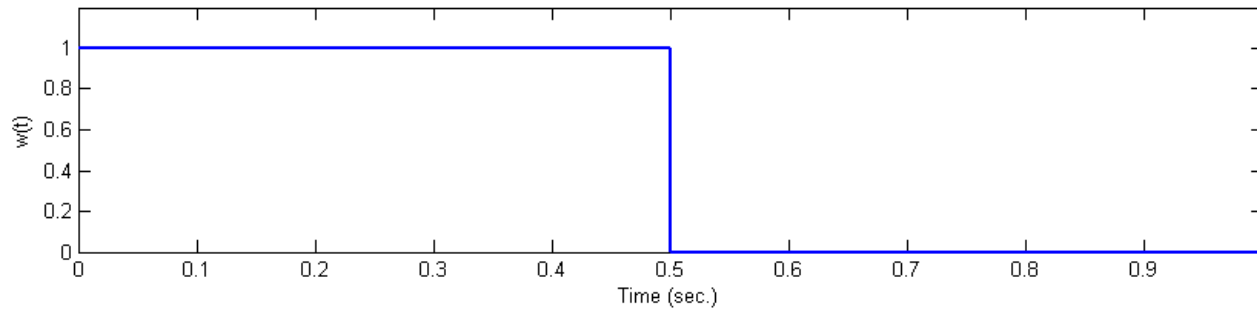
$$x_w(t) = \underbrace{\cos(2\pi f_0 t)}_{\text{carrier}} w(t)$$



Now, explain this ...

Square Window

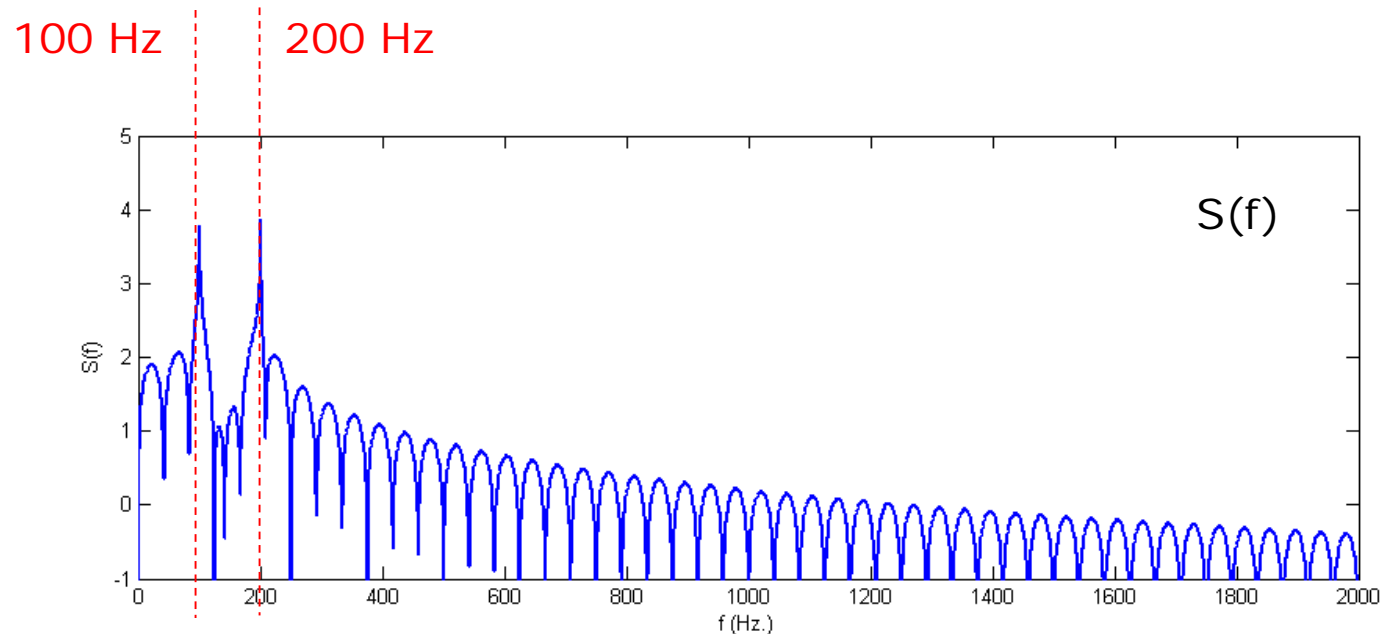
$$w(t) = \begin{cases} 1 & ; 0 \leq t < 0.5 \text{ sec.} \\ 0 & ; \text{otherwise} \end{cases}$$



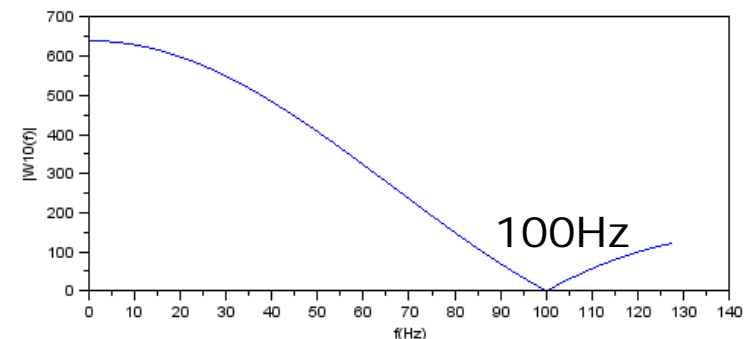
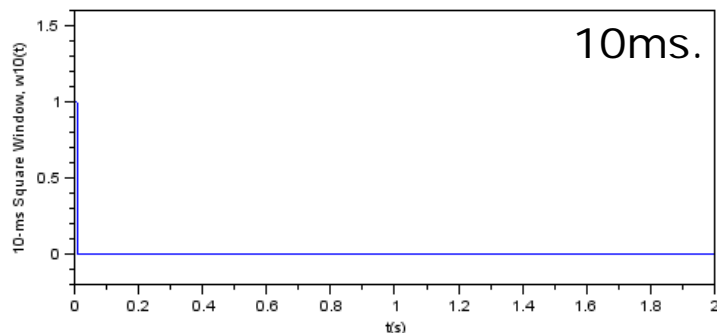
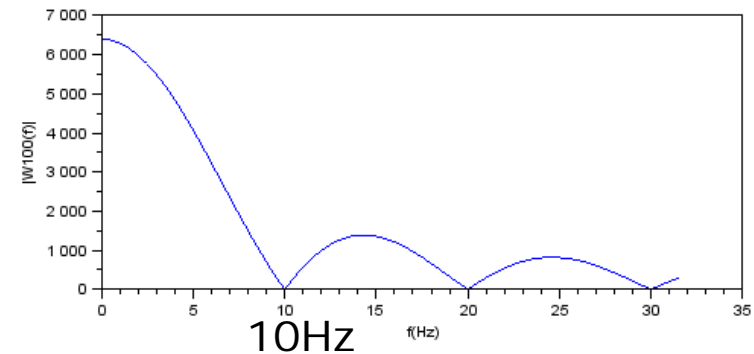
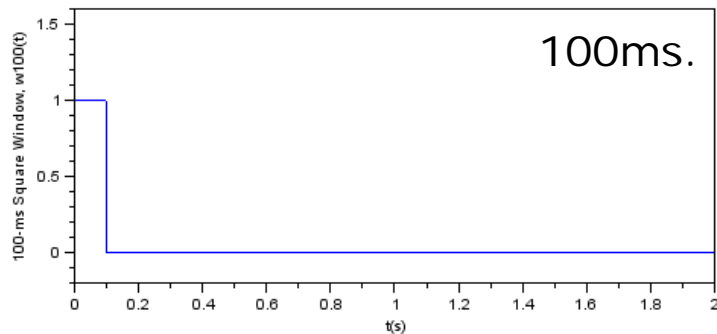
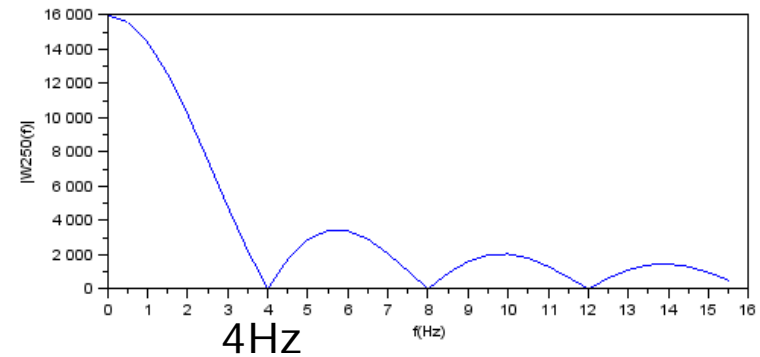
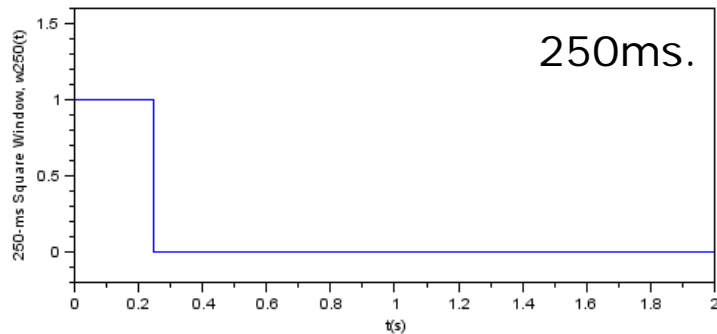
Effect of Time-windowing

$$s(t) = (\sin(2\pi(100)t) + \sin(2\pi(200)t))w(t)$$

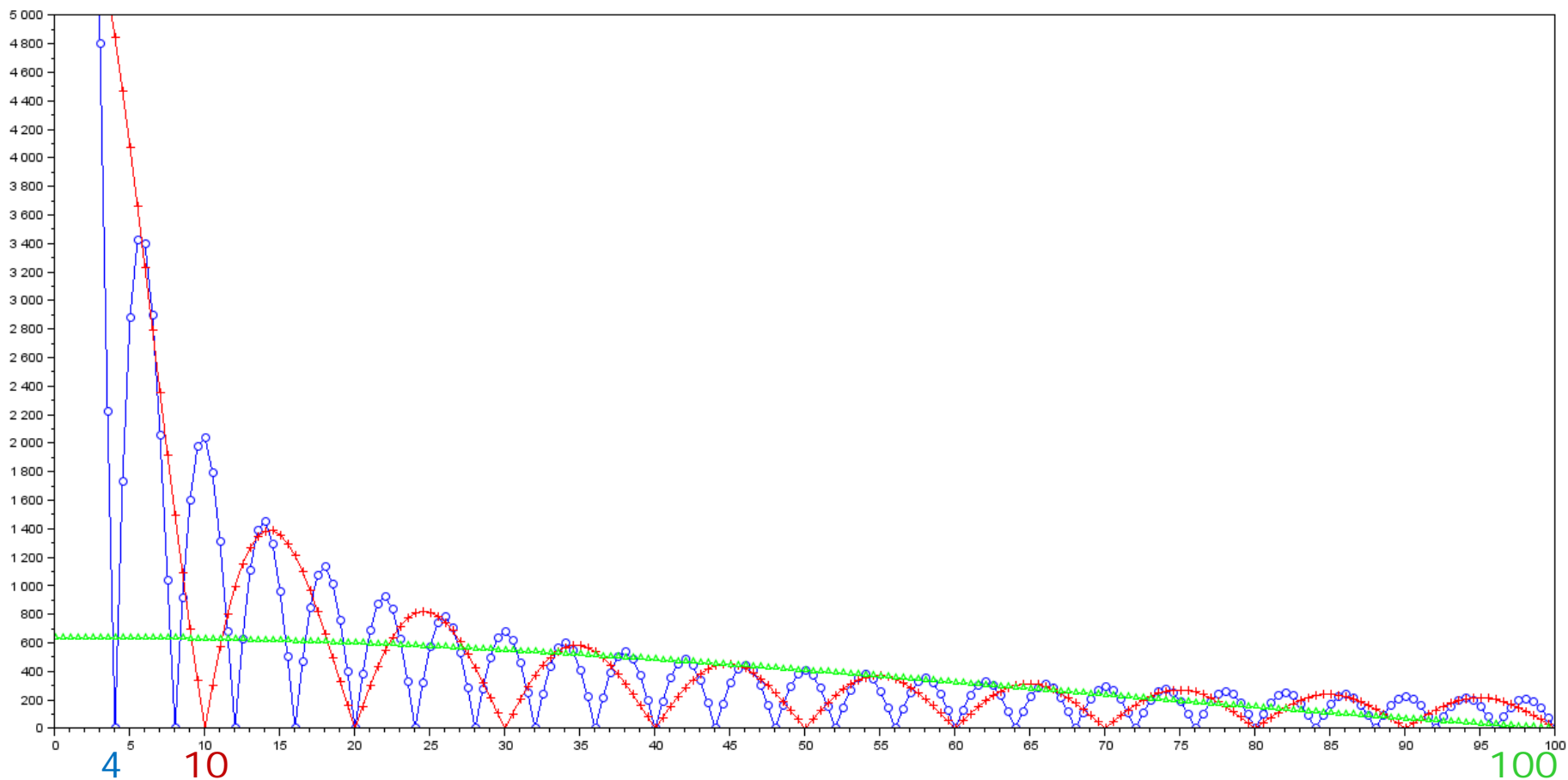
$w(t)$ is a 1-second-long square window



Window Length Vs. Frequency Resolution

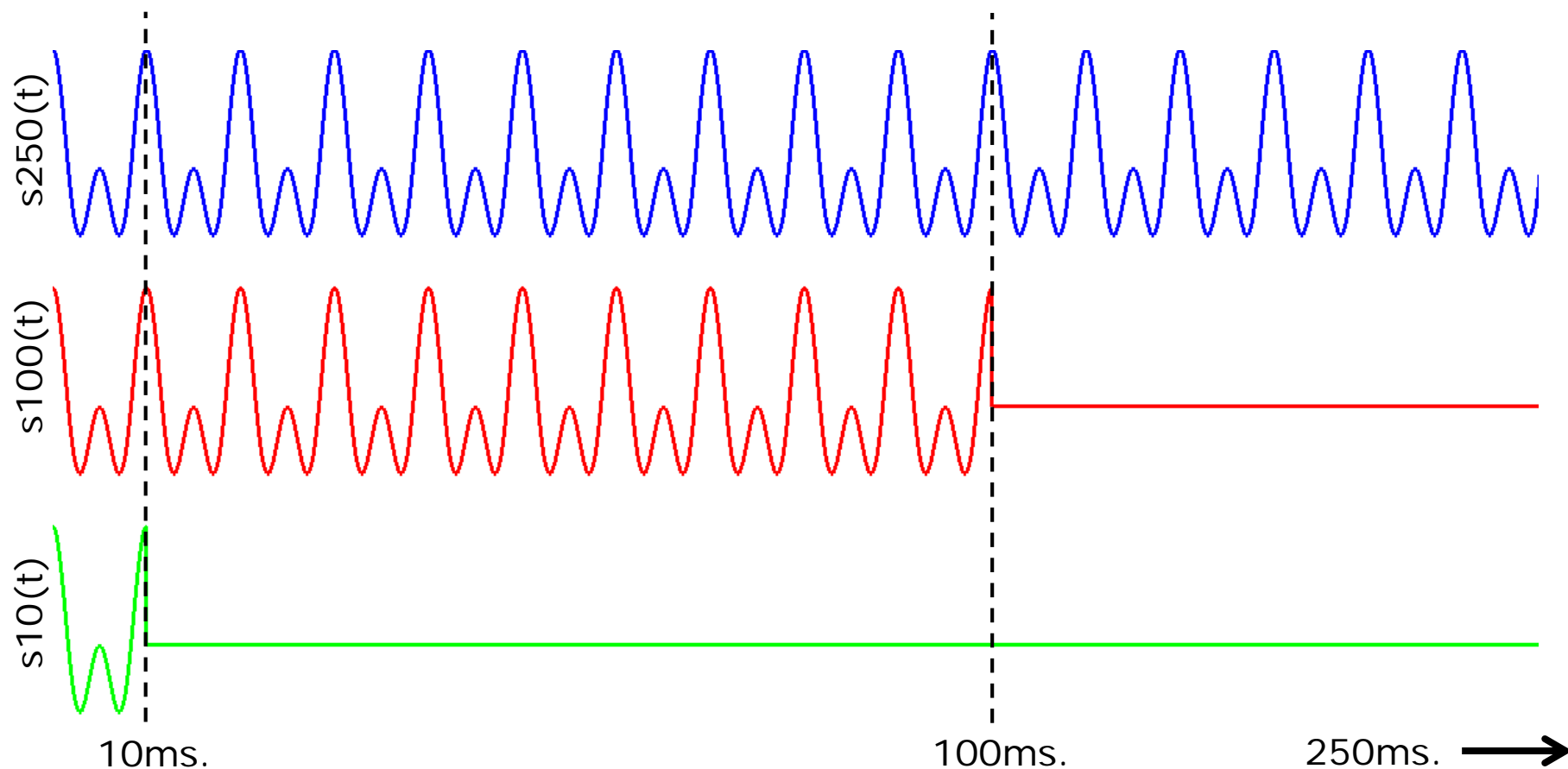


Window Length Vs. Frequency Resolution

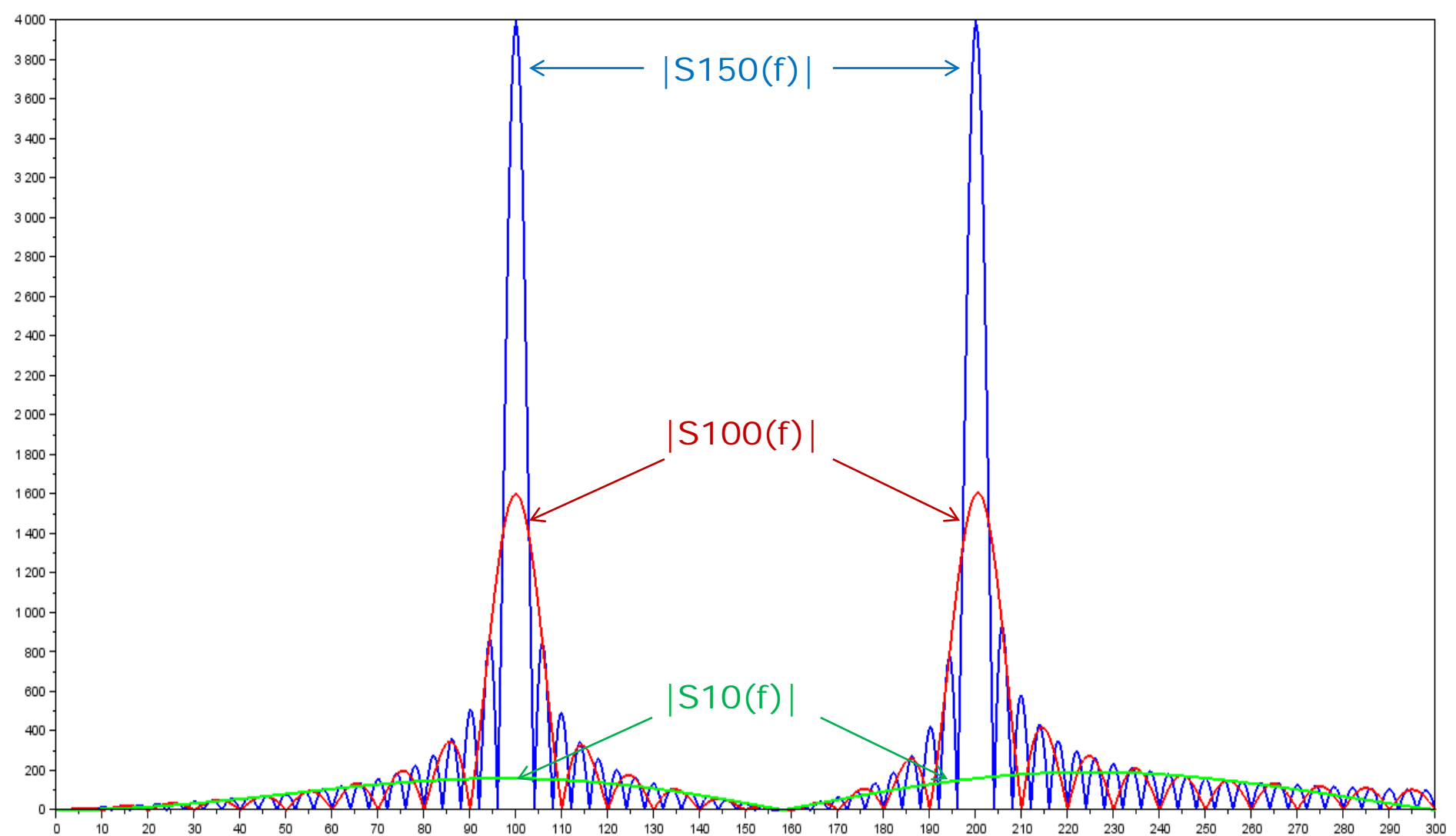


Window Length Vs. Frequency Resolution

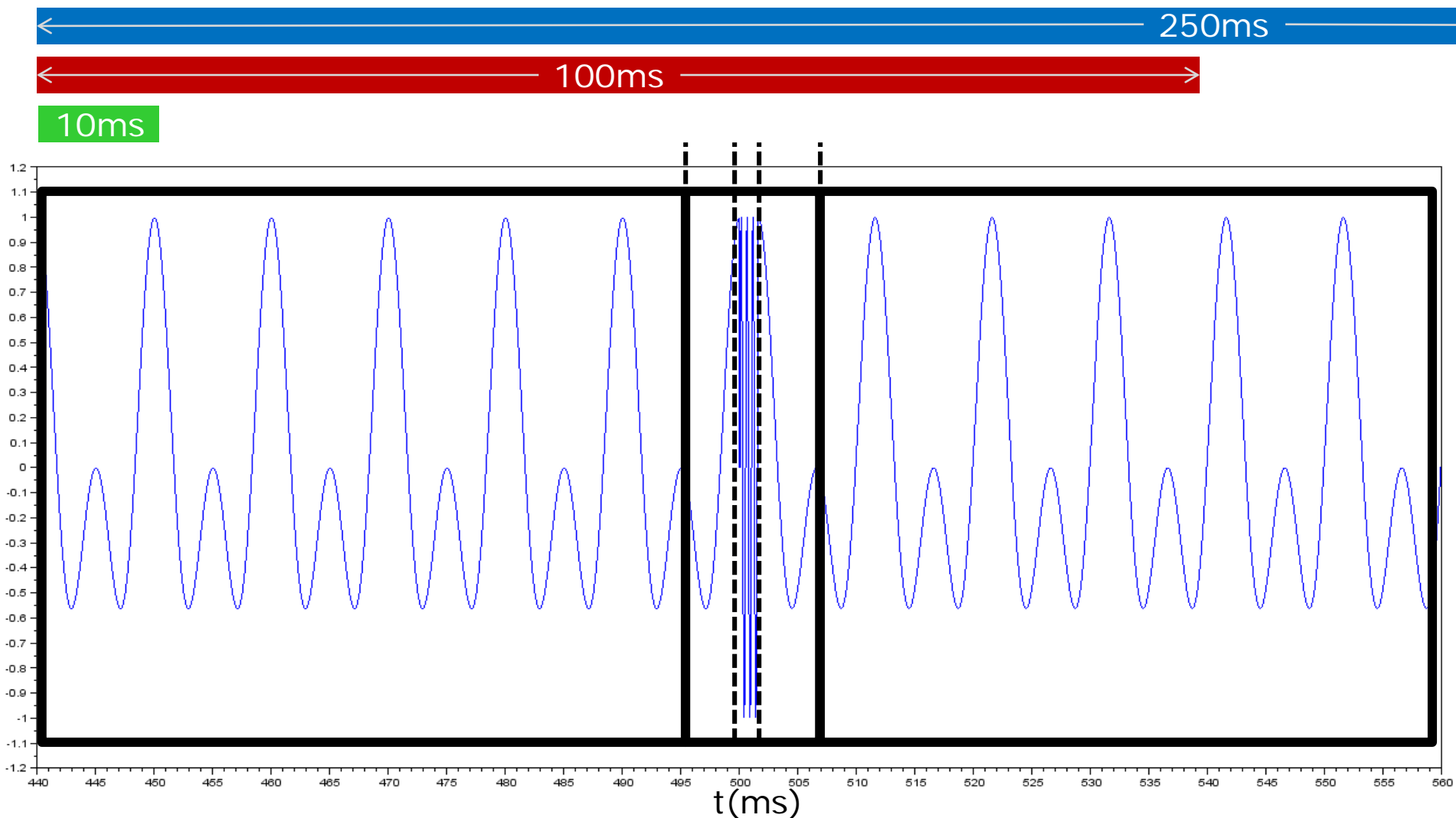
$$s(t) = (\sin(2\pi(100)t) + \sin(2\pi(200)t))w(t)$$



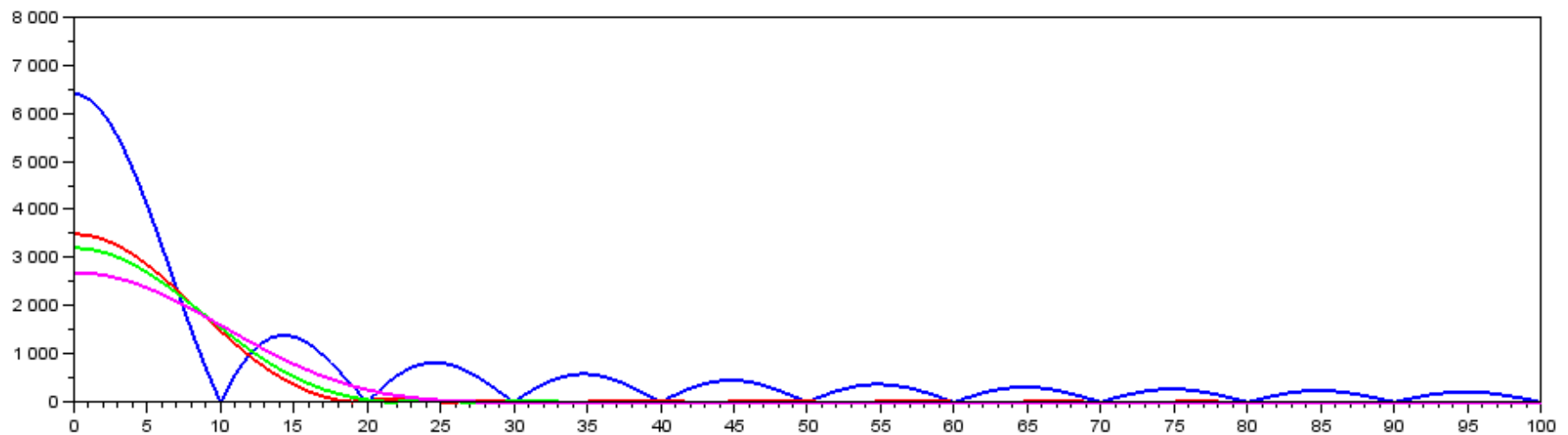
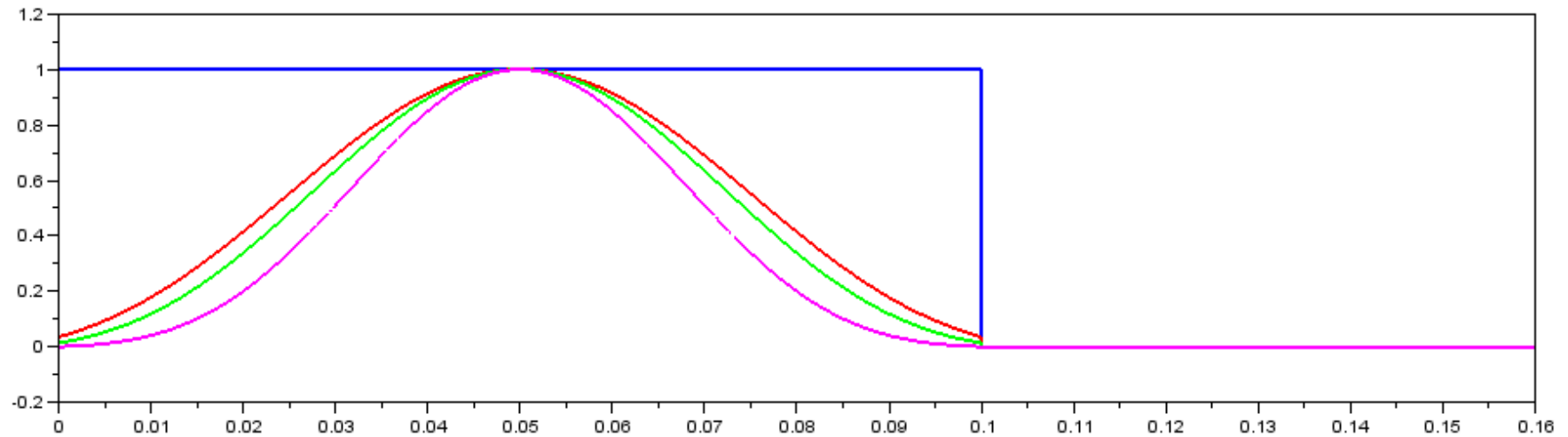
Window Length Vs. Freq. Resolution



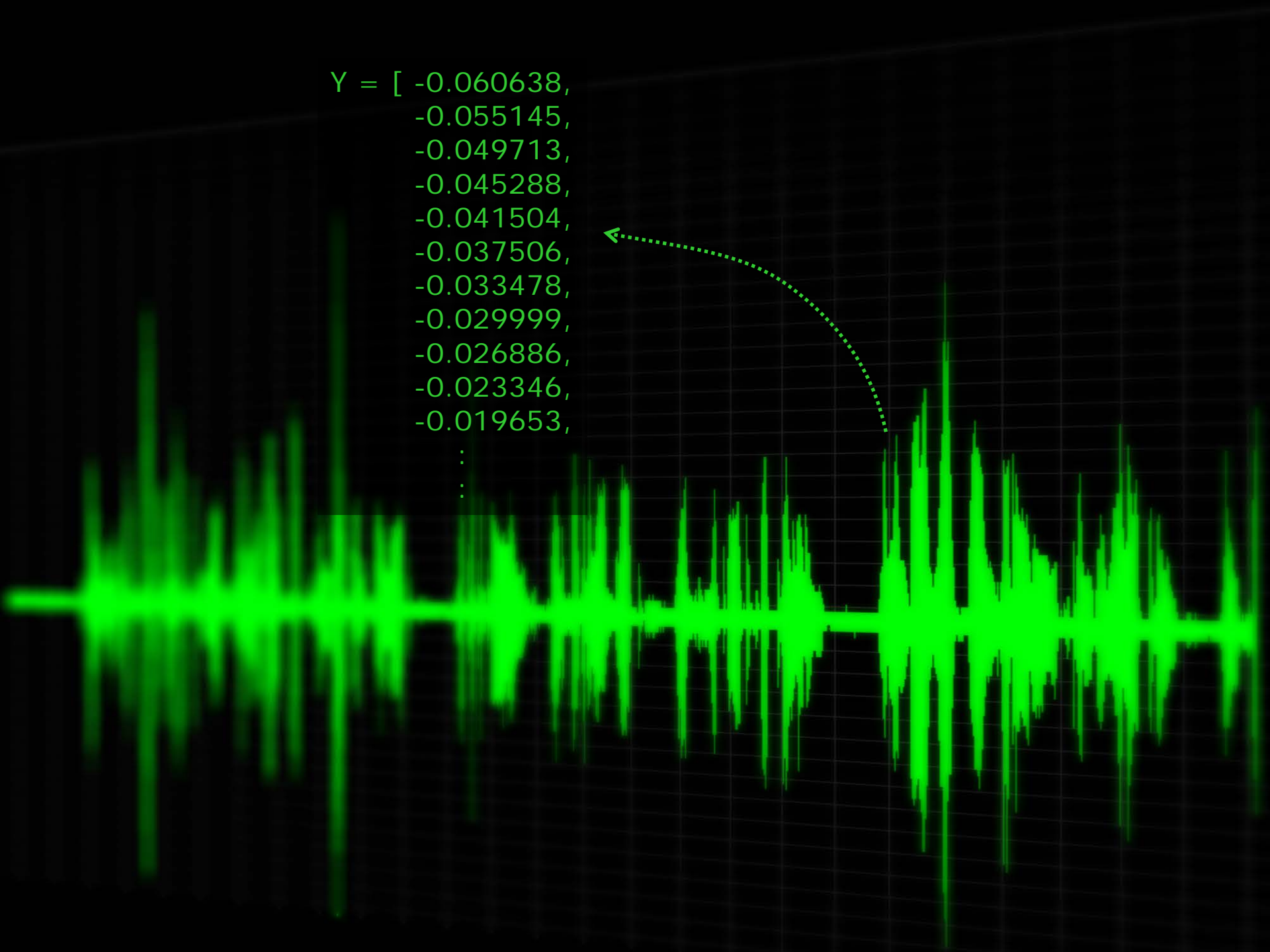
Time Resolution Vs. Frequency Resolution



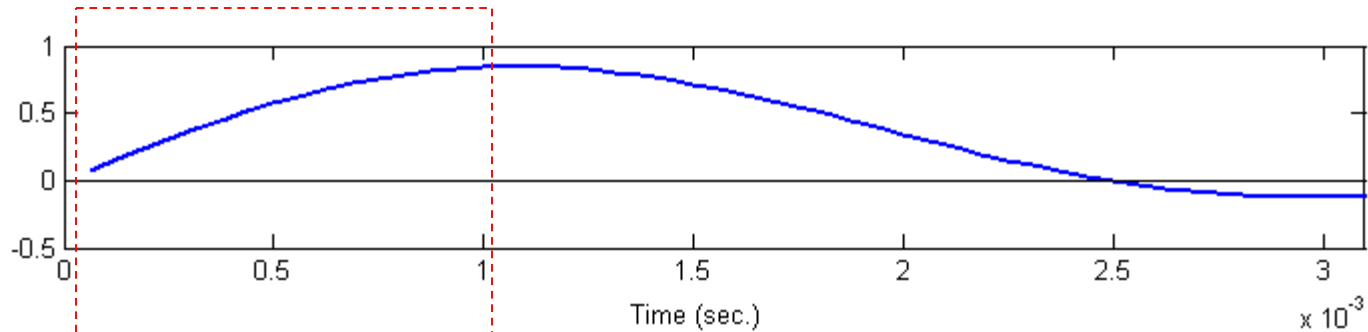
Different Types of Window



Y = [-0.060638,
-0.055145,
-0.049713,
-0.045288,
-0.041504,
-0.037506,
-0.033478,
-0.029999,
-0.026886,
-0.023346,
-0.019653,
:
:
:]



Sampling

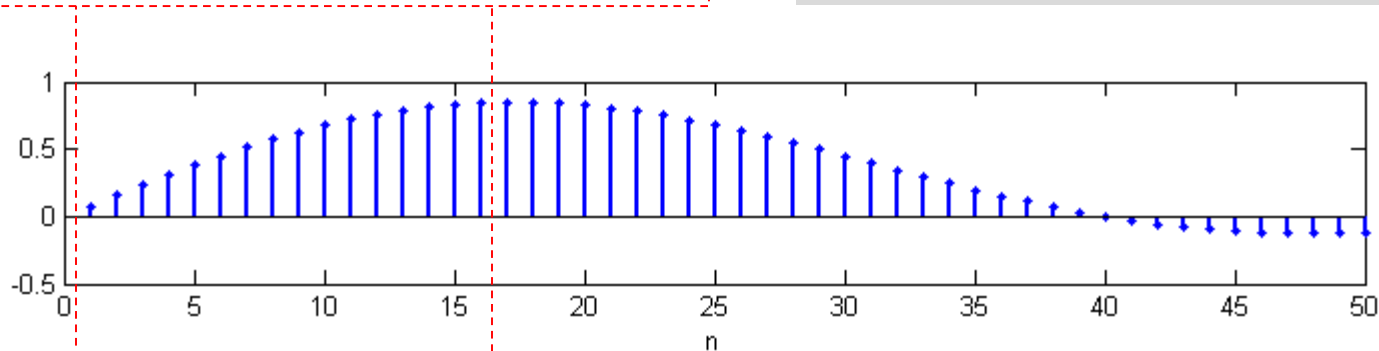


In 1×10^{-3} sec.
16 samples are picked

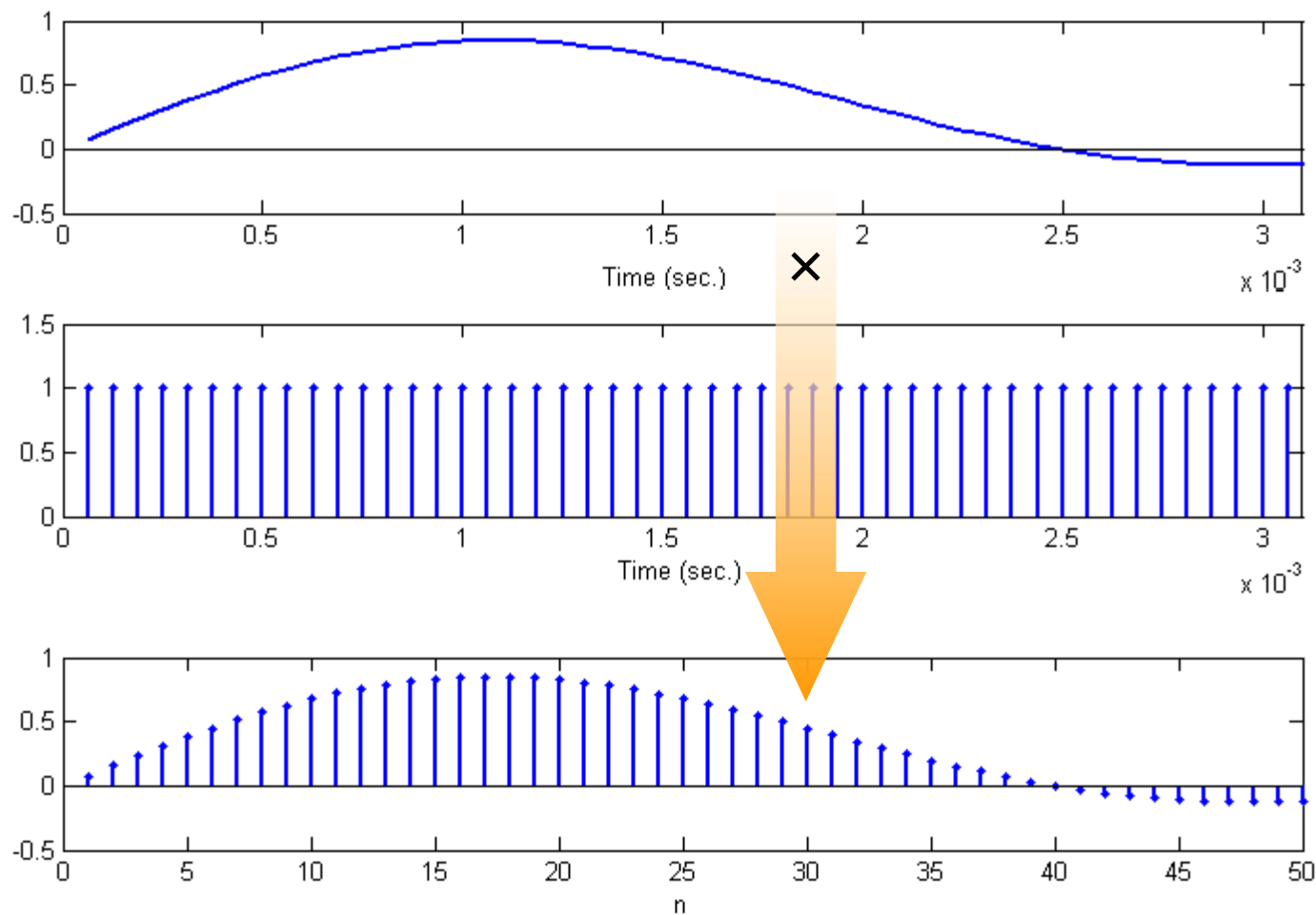
$\therefore F_s = 16,000$ samples/sec.
 $= 16$ kHz.

Sampling Frequency, F_s

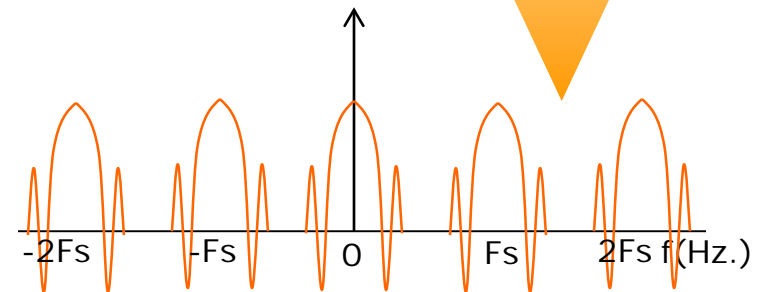
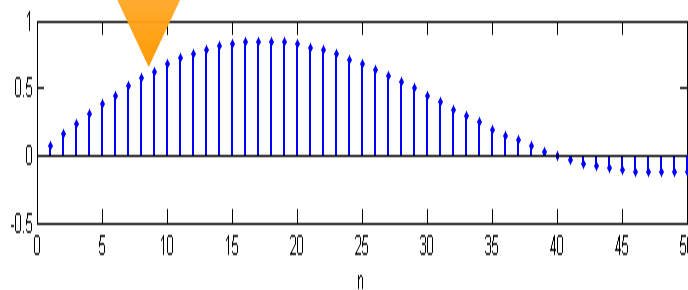
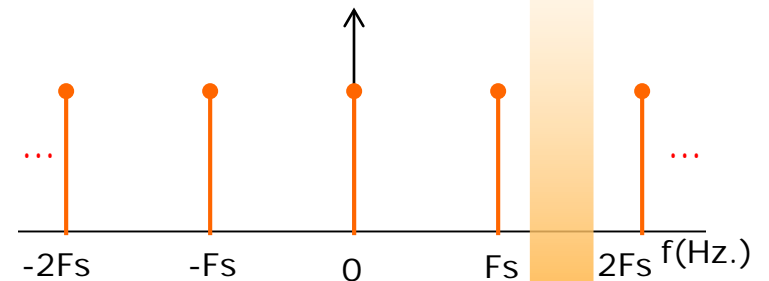
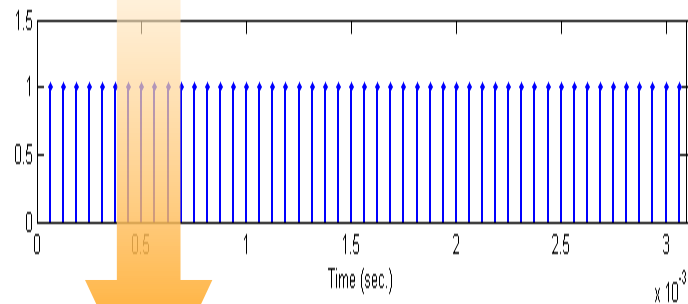
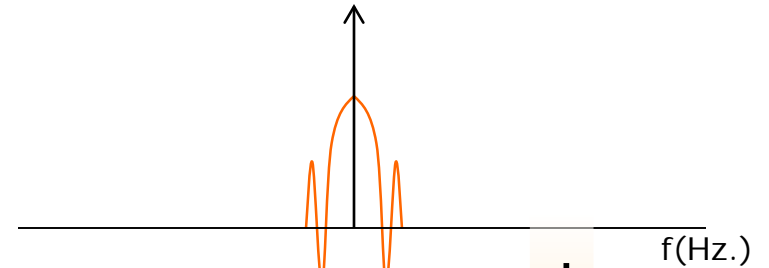
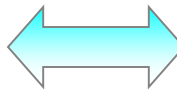
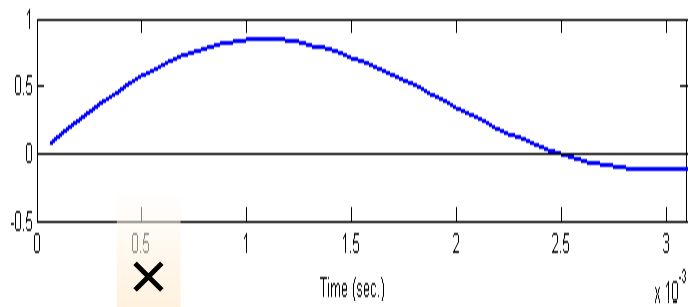
= how many samples are
picked in the interval of 1
second.



Sampling

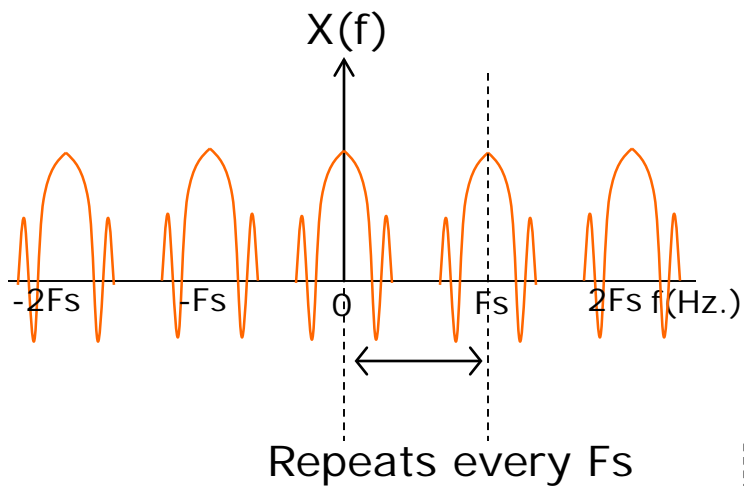


Sampling



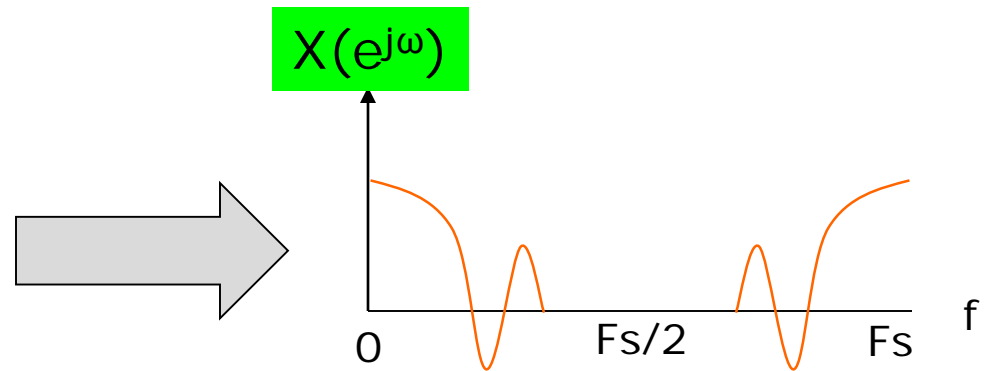
Normalized Frequency

Fourier Transform

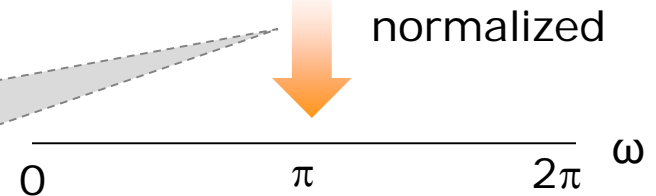


f : frequency (Hz.)

Discrete-time Fourier Transform



$$\omega = 2\pi \frac{f}{F_s}$$



ω : Normalized frequency
(Radian per sample)

Discrete-time Fourier Transform

$$x[n] \Leftrightarrow X(e^{j\omega})$$

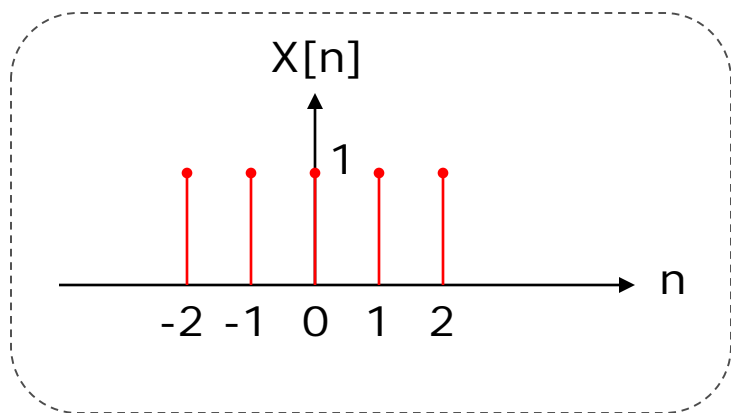
Discrete-time
Fourier
Transform

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

Inverse
Discrete-time
Fourier
Transform

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})e^{j\omega n} d\omega$$

Example

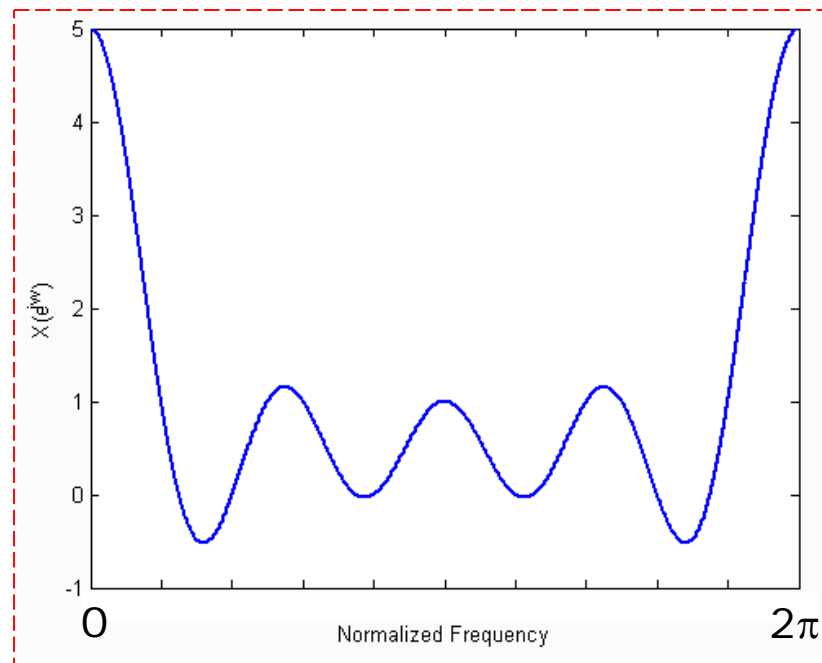


$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

$$= x[-2]e^{-j\omega(-2)} + x[-1]e^{-j\omega(-1)} + x[0]e^{-j\omega(0)} + x[1]e^{-j\omega(1)} + x[2]e^{-j\omega(2)}$$

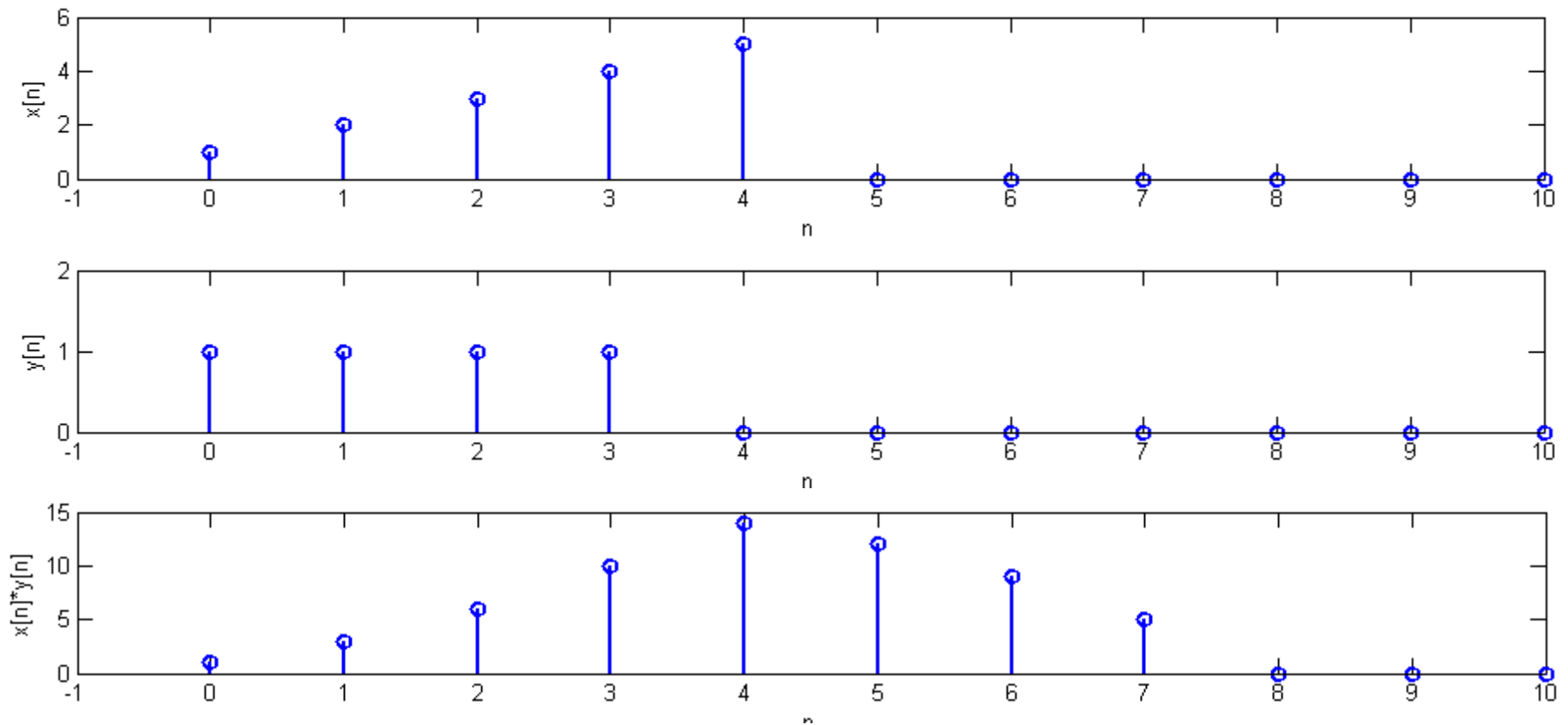
$$= e^{j\omega(2)} + e^{j\omega} + 1 + e^{-j\omega} + e^{-j\omega(2)}$$

$$= 1 + 2\cos(\omega) + 2\cos(2\omega)$$

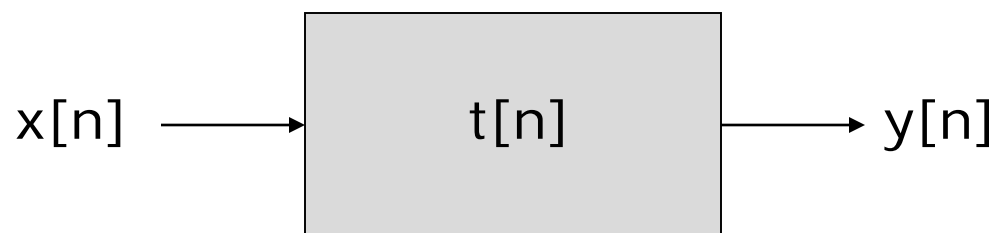


Discrete-time Convolution

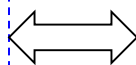
$$x[n] * y[n] = \sum_{m=-\infty}^{\infty} x[m]y[n-m]$$



Discrete-time Filter



$$Y(e^{j\omega}) = X(e^{j\omega})T(e^{j\omega})$$



$$y[n] = x[n] * t[n]$$

Family of Fourier Transform

- “Discreteness” in one domain implies “Periodicity” in the other domain.
- “Continuity” in one domain implies “Aperiodicity” in the other domain.

Transform	Time	Freq.
Cont. Fourier Trans.	Cont. & aperiodic	Cont. & aperiodic
Fourier Series	Cont. & periodic	Disc. & aperiodic
DTFT	Disc. & aperiodic	Cont. & periodic
DFT	Disc. & periodic	Disc & periodic

Discrete Fourier Transform

$$x[n] \Leftrightarrow X[k]$$

DFT

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi kn}{N}} \quad k = 0, 1, \dots, N-1$$

IDFT

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j \frac{2\pi kn}{N}} \quad n = 0, 1, \dots, N-1$$

Discrete Fourier Transform

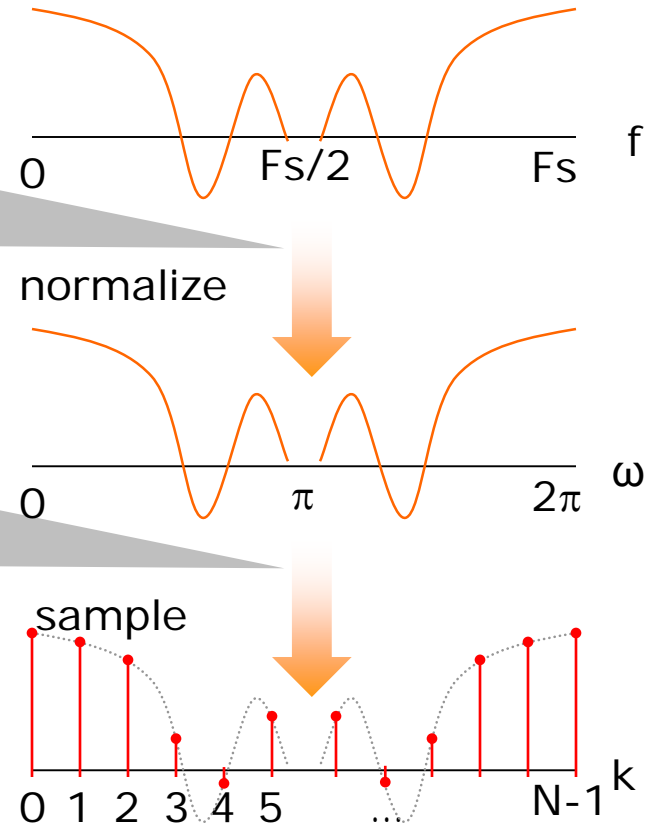
$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

$$\omega = 2\pi \frac{f}{F_s}$$

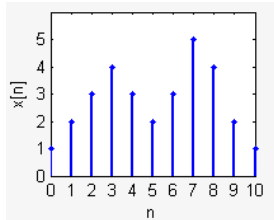
$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$k = \frac{\omega}{2\pi} N$$

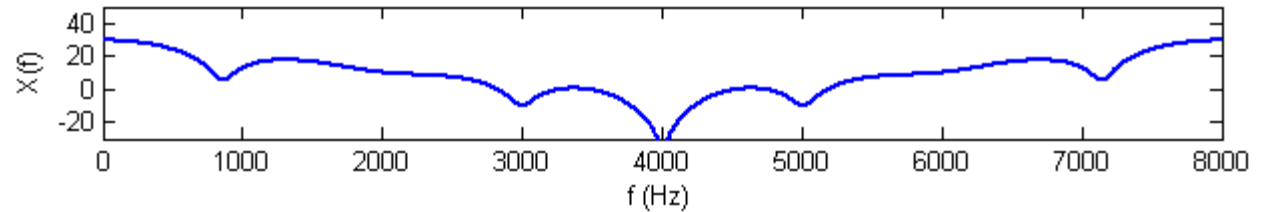
$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}}$$



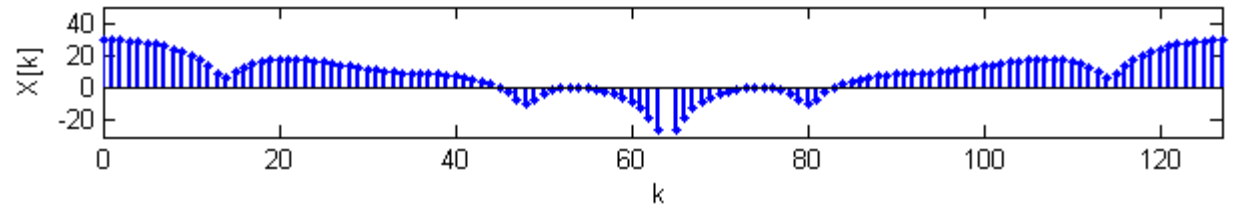
N DFT



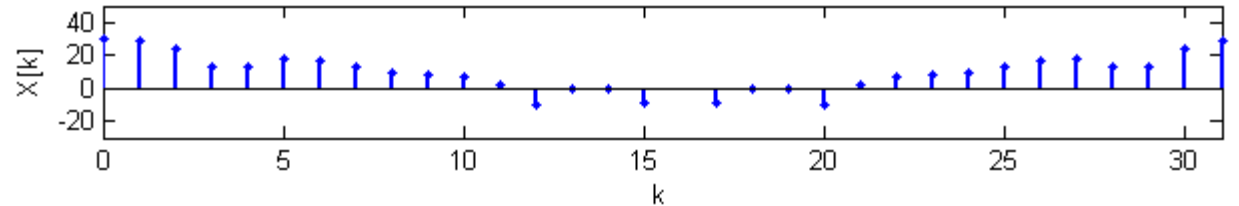
DTFT
↔



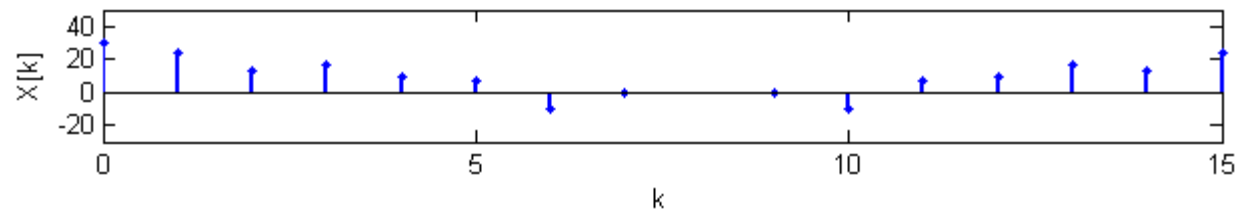
DFT
↔
 $N=128$



DFT
↔
 $N=32$



DFT
↔
 $N=16$



Numerical Computing Environment



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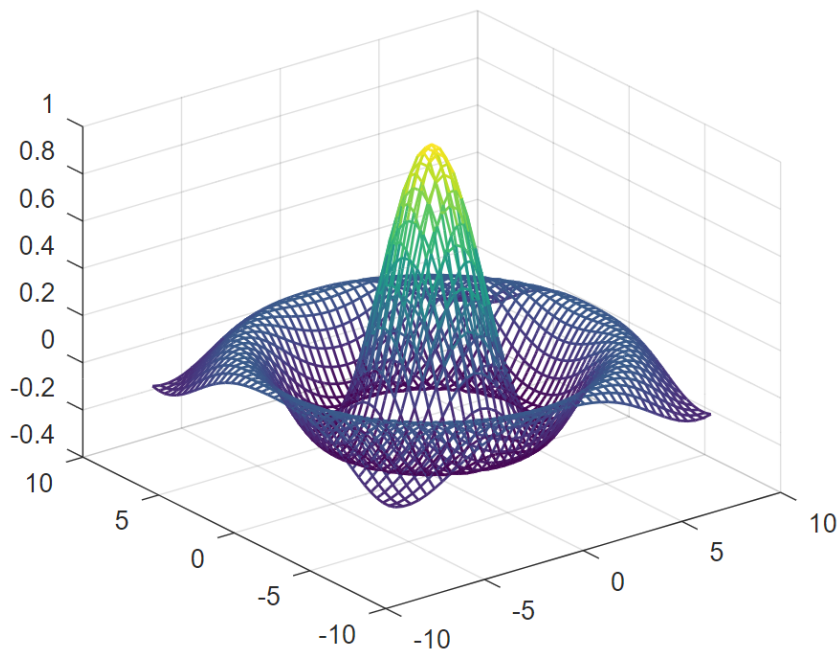
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GNU Octave



Scientific Programming Language

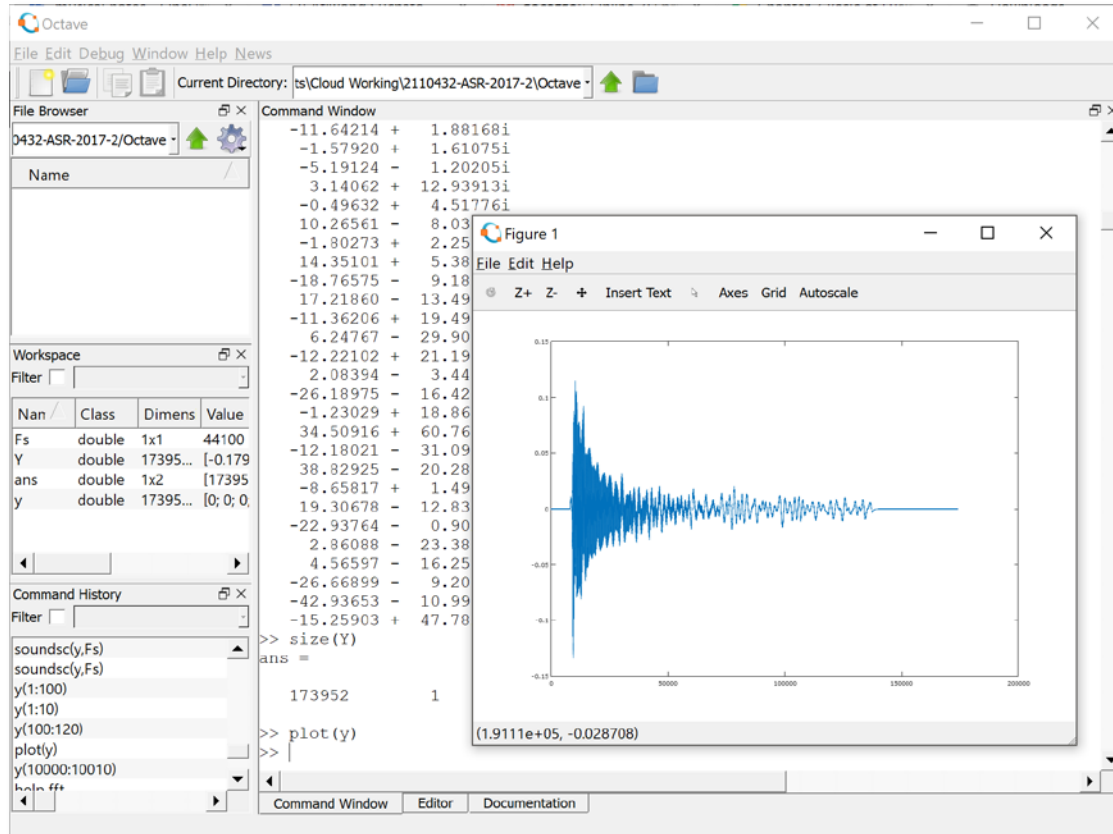
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- Free software, runs on GNU/Linux, macOS, BSD, and Windows
- Drop-in compatible with many Matlab scripts

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Octave Demo



- Basic operations
 - Variables
 - Colon operator
 - Matrix manipulation
 - Convolution
- Arrays
- Plotting
- Loading/Saving audio
- Help
- Control structures