

From Zero to One

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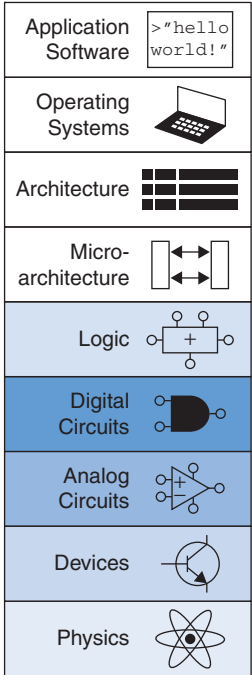
1.1 THE GAME PLAN

Microprocessors have revolutionized our world during the past three decades. A laptop computer today has far more capability than a room-sized mainframe of yesteryear. A luxury automobile contains about 100 microprocessors. Advances in microprocessors have made cell phones and the Internet possible, have vastly improved medicine, and have transformed how war is waged. Worldwide semiconductor industry sales have grown from US \$21 billion in 1985 to \$306 billion in 2013, and microprocessors are a major segment of these sales. We believe that microprocessors are not only technically, economically, and socially important, but are also an intrinsically fascinating human invention. By the time you finish reading this book, you will know how to design and build your own microprocessor. The skills you learn along the way will prepare you to design many other digital systems.

We assume that you have a basic familiarity with electricity, some prior programming experience, and a genuine interest in understanding what goes on under the hood of a computer. This book focuses on the design of digital systems, which operate on 1's and 0's. We begin with digital logic gates that accept 1's and 0's as inputs and produce 1's and 0's as outputs. We then explore how to combine logic gates into more complicated modules such as adders and memories. Then we shift gears to programming in assembly language, the native tongue of the microprocessor. Finally, we put gates together to build a microprocessor that runs these assembly language programs.

A great advantage of digital systems is that the building blocks are quite simple: just 1's and 0's. They do not require grungy mathematics or a profound knowledge of physics. Instead, the designer's challenge is to combine these simple blocks into complicated systems. A microprocessor may be the first system that you build that is too complex to fit in

- 1.1 The Game Plan
- 1.2 The Art of Managing Complexity
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your head all at once. One of the major themes weaved through this book is how to manage complexity.

1.2 THE ART OF MANAGING COMPLEXITY

One of the characteristics that separates an engineer or computer scientist from a layperson is a systematic approach to managing complexity. Modern digital systems are built from millions or billions of transistors. No human being could understand these systems by writing equations describing the movement of electrons in each transistor and solving all of the equations simultaneously. You will need to learn to manage complexity to understand how to build a microprocessor without getting mired in a morass of detail.

1.2.1 Abstraction

The critical technique for managing complexity is *abstraction*: hiding details when they are not important. A system can be viewed from many different levels of abstraction. For example, American politicians abstract the world into cities, counties, states, and countries. A county contains multiple cities and a state contains many counties. When a politician is running for president, the politician is mostly interested in how the state as a whole will vote, rather than how each county votes, so the state is the most useful level of abstraction. On the other hand, the Census Bureau measures the population of every city, so the agency must consider the details of a lower level of abstraction.

Figure 1.1 illustrates levels of abstraction for an electronic computer system along with typical building blocks at each level. At the lowest level of abstraction is the physics, the motion of electrons. The behavior of electrons is described by quantum mechanics and Maxwell’s equations. Our system is constructed from electronic *devices* such as transistors (or vacuum tubes, once upon a time). These devices have well-defined connection points called *terminals* and can be modeled by the relationship between voltage and current as measured at each terminal. By abstracting to this device level, we can ignore the individual electrons. The next level of abstraction is *analog circuits*, in which devices are assembled to create components such as amplifiers. Analog circuits input and output a continuous range of voltages. *Digital circuits* such as logic gates restrict the voltages to discrete ranges, which we will use to indicate 0 and 1. In logic design, we build more complex structures, such as adders or memories, from digital circuits.

Microarchitecture links the logic and architecture levels of abstraction. The *architecture* level of abstraction describes a computer from the programmer’s perspective. For example, the Intel x86 architecture used by microprocessors in most *personal computers* (PCs) is defined by a set of

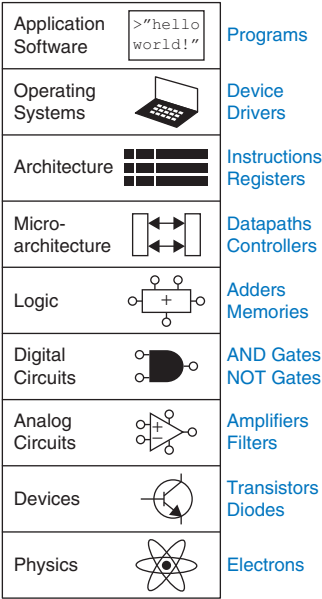


Figure 1.1 Levels of abstraction for an electronic computing system

instructions and registers (memory for temporarily storing variables) that the programmer is allowed to use. Microarchitecture involves combining logic elements to execute the instructions defined by the architecture. A particular architecture can be implemented by one of many different microarchitectures with different price/performance/power trade-offs. For example, the Intel Core i7, the Intel 80486, and the AMD Athlon all implement the x86 architecture with different microarchitectures.

Moving into the software realm, the operating system handles low-level details such as accessing a hard drive or managing memory. Finally, the application software uses these facilities provided by the operating system to solve a problem for the user. Thanks to the power of abstraction, your grandmother can surf the Web without any regard for the quantum vibrations of electrons or the organization of the memory in her computer.

This book focuses on the levels of abstraction from digital circuits through computer architecture. When you are working at one level of abstraction, it is good to know something about the levels of abstraction immediately above and below where you are working. For example, a computer scientist cannot fully optimize code without understanding the architecture for which the program is being written. A device engineer cannot make wise trade-offs in transistor design without understanding the circuits in which the transistors will be used. We hope that by the time you finish reading this book, you can pick the level of abstraction appropriate to solving your problem and evaluate the impact of your design choices on other levels of abstraction.

Each chapter in this book begins with an abstraction icon indicating the focus of the chapter in deep blue, with secondary topics shown in lighter shades of blue.

1.2.2 Discipline

Discipline is the act of intentionally restricting your design choices so that you can work more productively at a higher level of abstraction. Using interchangeable parts is a familiar application of discipline. One of the first examples of interchangeable parts was in flintlock rifle manufacturing. Until the early 19th century, rifles were individually crafted by hand. Components purchased from many different craftsmen were carefully filed and fit together by a highly skilled gunmaker. The discipline of interchangeable parts revolutionized the industry. By limiting the components to a standardized set with well-defined tolerances, rifles could be assembled and repaired much faster and with less skill. The gunmaker no longer concerned himself with lower levels of abstraction such as the specific shape of an individual barrel or gunstock.

In the context of this book, the digital discipline will be very important. Digital circuits use discrete voltages, whereas analog circuits use continuous voltages. Therefore, digital circuits are a subset of analog circuits and in some sense must be capable of less than the broader class of analog circuits. However, digital circuits are much simpler to design. By limiting

ourselves to digital circuits, we can easily combine components into sophisticated systems that ultimately outperform those built from analog components in many applications. For example, digital televisions, compact disks (CDs), and cell phones are replacing their analog predecessors.

1.2.3 The Three-Y's

In addition to abstraction and discipline, designers use the three “-y’s” to manage complexity: hierarchy, modularity, and regularity. These principles apply to both software and hardware systems.

- ▶ *Hierarchy* involves dividing a system into modules, then further subdividing each of these modules until the pieces are easy to understand.
- ▶ *Modularity* states that the modules have well-defined functions and interfaces, so that they connect together easily without unanticipated side effects.
- ▶ *Regularity* seeks uniformity among the modules. Common modules are reused many times, reducing the number of distinct modules that must be designed.

To illustrate these “-y’s” we return to the example of rifle manufacturing. A flintlock rifle was one of the most intricate objects in common use in the early 19th century. Using the principle of hierarchy, we can break it into components shown in [Figure 1.2](#): the lock, stock, and barrel.

The barrel is the long metal tube through which the bullet is fired. The lock is the firing mechanism. And the stock is the wooden body that holds the parts together and provides a secure grip for the user. In turn, the lock contains the trigger, hammer, flint, frizzen, and pan. Each of these components could be hierarchically described in further detail.

Modularity teaches that each component should have a well-defined function and interface. A function of the stock is to mount the barrel and lock. Its interface consists of its length and the location of its mounting pins. In a modular rifle design, stocks from many different manufacturers can be used with a particular barrel as long as the stock and barrel are of the correct length and have the proper mounting mechanism. A function of the barrel is to impart spin to the bullet so that it travels more accurately. Modularity dictates that there should be no side effects: the design of the stock should not impede the function of the barrel.

Regularity teaches that interchangeable parts are a good idea. With regularity, a damaged barrel can be replaced by an identical part. The barrels can be efficiently built on an assembly line, instead of being painstakingly hand-crafted.

We will return to these principles of hierarchy, modularity, and regularity throughout the book.

Captain Meriwether Lewis of the Lewis and Clark Expedition was one of the early advocates of interchangeable parts for rifles. In 1806, he explained:

The guns of Drewyer and Sergt. Pryor were both out of order. The first was repaired with a new lock, the old one having become unfit for use; the second had the cock screw broken which was replaced by a duplicate which had been prepared for the lock at Harpers Ferry where she was manufactured. But for the precaution taken in bringing on those extra locks, and parts of locks, in addition to the ingenuity of John Shields, most of our guns would at this moment be entirely unfit for use; but fortunately for us I have it in my power here to record that they are all in good order.

See Elliott Coues, ed., *The History of the Lewis and Clark Expedition...* (4 vols), New York: Harper, 1893; reprint, 3 vols, New York: Dover, 3:817.

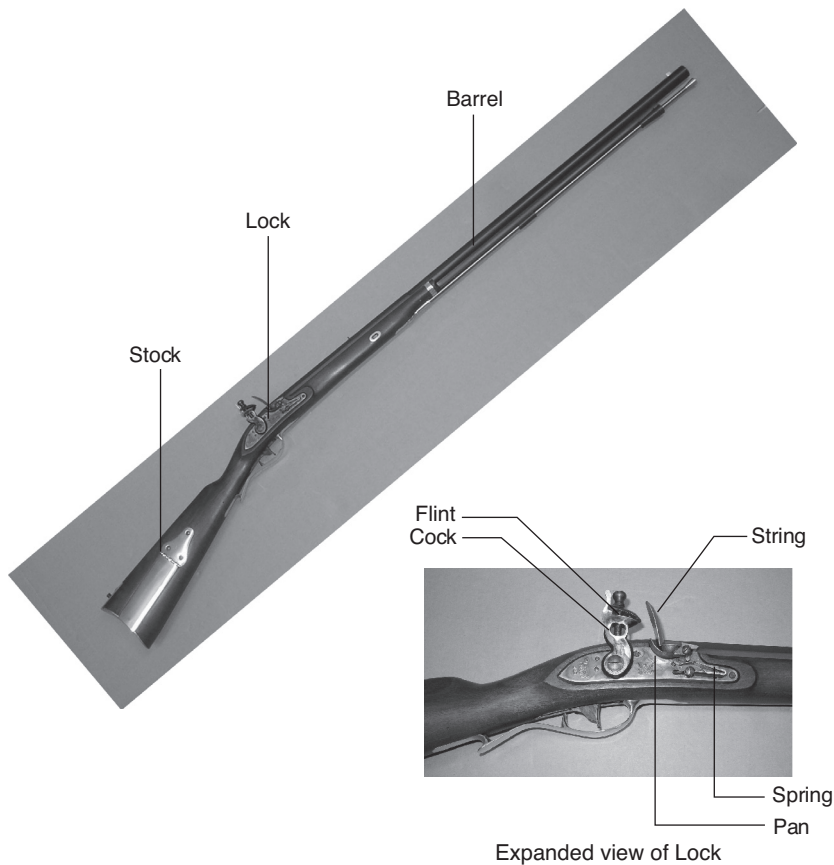
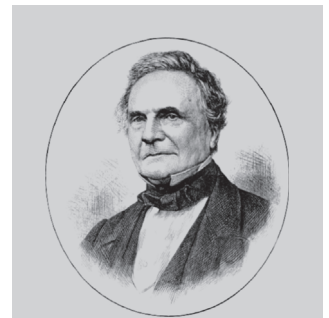


Figure 1.2 Flintlock rifle with a close-up view of the lock

(Image by Euroarms Italia. www.euroarms.net © 2006.)



Charles Babbage, 1791–1871.

Attended Cambridge University and married Georgiana Whitmore in 1814. Invented the Analytical Engine, the world's first mechanical computer. Also invented the cowcatcher and the universal postage rate. Interested in lock-picking, but abhorred street musicians (image courtesy of Fourmilab Switzerland, www.fourmilab.ch).

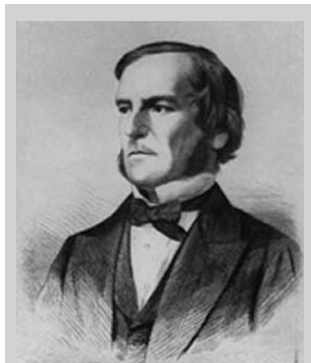
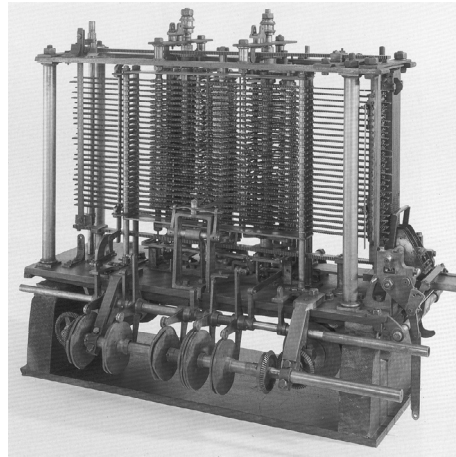
1.3 THE DIGITAL ABSTRACTION

Most physical variables are continuous. For example, the voltage on a wire, the frequency of an oscillation, or the position of a mass are all continuous quantities. Digital systems, on the other hand, represent information with *discrete-valued variables*—that is, variables with a finite number of distinct values.

An early digital system using variables with ten discrete values was Charles Babbage's Analytical Engine. Babbage labored from 1834 to 1871, designing and attempting to build this mechanical computer. The Analytical Engine used gears with ten positions labeled 0 through 9, much like a mechanical odometer in a car. Figure 1.3 shows a prototype of the Analytical Engine, in which each row processes one digit. Babbage chose 25 rows of gears, so the machine has 25-digit precision.

Figure 1.3 Babbage's Analytical Engine, under construction at the time of his death in 1871

(image courtesy of Science Museum/Science and Society Picture Library)



George Boole, 1815–1864. Born to working-class parents and unable to afford a formal education, Boole taught himself mathematics and joined the faculty of Queen's College in Ireland. He wrote *An Investigation of the Laws of Thought* (1854), which introduced binary variables and the three fundamental logic operations: AND, OR, and NOT (image courtesy of the American Institute of Physics).

Unlike Babbage's machine, most electronic computers use a binary (two-valued) representation in which a high voltage indicates a '1' and a low voltage indicates a '0', because it is easier to distinguish between two voltages than ten.

The *amount of information* D in a discrete valued variable with N distinct states is measured in units of *bits* as

$$D = \log_2 N \text{ bits} \quad (1.1)$$

A binary variable conveys $\log_2 2 = 1$ bit of information. Indeed, the word bit is short for *binary digit*. Each of Babbage's gears carried $\log_2 10 = 3.322$ bits of information because it could be in one of $2^{3.322} = 10$ unique positions. A continuous signal theoretically contains an infinite amount of information because it can take on an infinite number of values. In practice, noise and measurement error limit the information to only 10 to 16 bits for most continuous signals. If the measurement must be made rapidly, the information content is lower (e.g., 8 bits).

This book focuses on digital circuits using binary variables: 1's and 0's. George Boole developed a system of logic operating on binary variables that is now known as *Boolean logic*. Each of Boole's variables could be TRUE or FALSE. Electronic computers commonly use a positive voltage to represent '1' and zero volts to represent '0'. In this book, we will use the terms '1', TRUE, and HIGH synonymously. Similarly, we will use '0', FALSE, and LOW interchangeably.

The beauty of the *digital abstraction* is that digital designers can focus on 1's and 0's, ignoring whether the Boolean variables are physically represented with specific voltages, rotating gears, or even hydraulic fluid levels. A computer programmer can work without needing to know the intimate

details of the computer hardware. On the other hand, understanding the details of the hardware allows the programmer to optimize the software better for that specific computer.

An individual bit doesn't carry much information. In the next section, we examine how groups of bits can be used to represent numbers. In later chapters, we will also use groups of bits to represent letters and programs.

1.4 NUMBER SYSTEMS

You are accustomed to working with decimal numbers. In digital systems consisting of 1's and 0's, binary or hexadecimal numbers are often more convenient. This section introduces the various number systems that will be used throughout the rest of the book.

1.4.1 Decimal Numbers

In elementary school, you learned to count and do arithmetic in *decimal*. Just as you (probably) have ten fingers, there are ten decimal digits: 0, 1, 2, ..., 9. Decimal digits are joined together to form longer decimal numbers. Each column of a decimal number has ten times the weight of the previous column. From right to left, the column weights are 1, 10, 100, 1000, and so on. Decimal numbers are referred to as *base 10*. The base is indicated by a subscript after the number to prevent confusion when working in more than one base. For example, Figure 1.4 shows how the decimal number 9742_{10} is written as the sum of each of its digits multiplied by the weight of the corresponding column.

An N -digit decimal number represents one of 10^N possibilities: 0, 1, 2, 3, ..., $10^N - 1$. This is called the *range* of the number. For example, a three-digit decimal number represents one of 1000 possibilities in the range of 0 to 999.

1.4.2 Binary Numbers

Bits represent one of two values, 0 or 1, and are joined together to form *binary numbers*. Each column of a binary number has twice the weight of the previous column, so binary numbers are *base 2*. In binary, the

$$9742_{10} = 9 \times 10^3 + 7 \times 10^2 + 4 \times 10^1 + 2 \times 10^0$$

nine thousands
seven hundreds
four tens
two ones

Figure 1.4 Representation of a decimal number

column weights (again from right to left) are 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, 1024, 2048, 4096, 8192, 16384, 32768, 65536, and so on. If you work with binary numbers often, you'll save time if you remember these powers of two up to 2^{16} .

An N -bit binary number represents one of 2^N possibilities: 0, 1, 2, 3, ..., $2^N - 1$. Table 1.1 shows 1, 2, 3, and 4-bit binary numbers and their decimal equivalents.

Example 1.1 BINARY TO DECIMAL CONVERSION

Convert the binary number 10110_2 to decimal.

Solution: Figure 1.5 shows the conversion.

Table 1.1 Binary numbers and their decimal equivalent

1-Bit Binary Numbers	2-Bit Binary Numbers	3-Bit Binary Numbers	4-Bit Binary Numbers	Decimal Equivalents
0	00	000	0000	0
1	01	001	0001	1
	10	010	0010	2
	11	011	0011	3
		100	0100	4
		101	0101	5
		110	0110	6
		111	0111	7
			1000	8
			1001	9
			1010	10
			1011	11
			1100	12
			1101	13
			1110	14
			1111	15

1's column

2's column

4's column

8's column

16's column

$$10110_2 = 1 \times 2^4 + 0 \times 2^3 + 1 \times 2^2 + 1 \times 2^1 + 0 \times 2^0 = 22_{10}$$

one sixteen

no eight

one four

one two

no one

Figure 1.5 Conversion of a binary number to decimal

Example 1.2 DECIMAL TO BINARY CONVERSION

Convert the decimal number 84₁₀ to binary.

Solution: Determine whether each column of the binary result has a 1 or a 0. We can do this starting at either the left or the right column.

Working from the left, start with the largest power of 2 less than or equal to the number (in this case, 64). 84 ≥ 64, so there is a 1 in the 64's column, leaving 84 − 64 = 20. 20 < 32, so there is a 0 in the 32's column. 20 ≥ 16, so there is a 1 in the 16's column, leaving 20 − 16 = 4. 4 < 8, so there is a 0 in the 8's column. 4 ≥ 4, so there is a 1 in the 4's column, leaving 4 − 4 = 0. Thus there must be 0's in the 2's and 1's column. Putting this all together, 84₁₀ = 1010100₂.

Working from the right, repeatedly divide the number by 2. The remainder goes in each column. 84/2 = 42, so 0 goes in the 1's column. 42/2 = 21, so 0 goes in the 2's column. 21/2 = 10 with a remainder of 1 going in the 4's column. 10/2 = 5, so 0 goes in the 8's column. 5/2 = 2 with a remainder of 1 going in the 16's column. 2/2 = 1, so 0 goes in the 32's column. Finally 1/2 = 0 with a remainder of 1 going in the 64's column. Again, 84₁₀ = 1010100₂.

1.4.3 Hexadecimal Numbers

Writing long binary numbers becomes tedious and prone to error. A group of four bits represents one of 2⁴ = 16 possibilities. Hence, it is sometimes more convenient to work in *base 16*, called *hexadecimal*. Hexadecimal numbers use the digits 0 to 9 along with the letters A to F, as shown in Table 1.2. Columns in base 16 have weights of 1, 16, 16² (or 256), 16³ (or 4096), and so on.

“Hexadecimal,” a term coined by IBM in 1963, derives from the Greek *hexi* (six) and Latin *decem* (ten). A more proper term would use the Latin *sexa* (six), but *sexadecimal* sounded too risqué.

Example 1.3 HEXADECIMAL TO BINARY AND DECIMAL CONVERSION

Convert the hexadecimal number 2ED₁₆ to binary and to decimal.

Solution: Conversion between hexadecimal and binary is easy because each hexadecimal digit directly corresponds to four binary digits. 2₁₆ = 0010₂, E₁₆ = 1110₂ and D₁₆ = 1101₂, so 2ED₁₆ = 001011101101₂. Conversion to decimal requires the arithmetic shown in Figure 1.6.

Table 1.2 Hexadecimal number system

Hexadecimal Digit	Decimal Equivalent	Binary Equivalent
0	0	0000
1	1	0001
2	2	0010
3	3	0011
4	4	0100
5	5	0101
6	6	0110
7	7	0111
8	8	1000
9	9	1001
A	10	1010
B	11	1011
C	12	1100
D	13	1101
E	14	1110
F	15	1111

Figure 1.6 Conversion of a hexadecimal number to decimal

1's column
16's column
256's column

$$2ED_{16} = 2 \times 16^2 + E \times 16^1 + D \times 16^0 = 749_{10}$$

two
two hundred
fifty six's

fourteen
sixteens

thirteen
ones

Example 1.4 BINARY TO HEXADECIMAL CONVERSION

Convert the binary number 1111010₂ to hexadecimal.

Solution: Again, conversion is easy. Start reading from the right. The four least significant bits are 1010₂ = A₁₆. The next bits are 111₂ = 7₁₆. Hence 1111010₂ = 7A₁₆.

Example 1.5 DECIMAL TO HEXADECIMAL AND BINARY CONVERSION

Convert the decimal number 333_{10} to hexadecimal and binary.

Solution: Like decimal to binary conversion, decimal to hexadecimal conversion can be done from the left or the right.

Working from the left, start with the largest power of 16 less than or equal to the number (in this case, 256). 256 goes into 333 once, so there is a 1 in the 256's column, leaving $333 - 256 = 77$. 16 goes into 77 four times, so there is a 4 in the 16's column, leaving $77 - 16 \times 4 = 13$. $13_{10} = D_{16}$, so there is a D in the 1's column. In summary, $333_{10} = 14D_{16}$. Now it is easy to convert from hexadecimal to binary, as in Example 1.3. $14D_{16} = 101001101_2$.

Working from the right, repeatedly divide the number by 16. The remainder goes in each column. $333/16 = 20$ with a remainder of $13_{10} = D_{16}$ going in the 1's column. $20/16 = 1$ with a remainder of 4 going in the 16's column. $1/16 = 0$ with a remainder of 1 going in the 256's column. Again, the result is $14D_{16}$.

1.4.4 Bytes, Nibbles, and All That Jazz

A group of eight bits is called a *byte*. It represents one of $2^8 = 256$ possibilities. The size of objects stored in computer memories is customarily measured in bytes rather than bits.

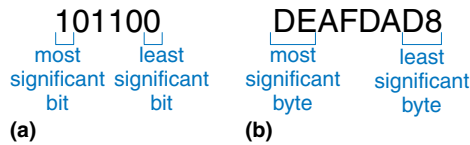
A group of four bits, or half a byte, is called a *nibble*. It represents one of $2^4 = 16$ possibilities. One hexadecimal digit stores one nibble and two hexadecimal digits store one full byte. Nibbles are no longer a commonly used unit, but the term is cute.

Microprocessors handle data in chunks called *words*. The size of a word depends on the architecture of the microprocessor. When this chapter was written in 2015, most computers had 64-bit processors, indicating that they operate on 64-bit words. At the time, older computers handling 32-bit words were also widely available. Simpler microprocessors, especially those used in gadgets such as toasters, use 8- or 16-bit words.

Within a group of bits, the bit in the 1's column is called the *least significant bit (lsb)*, and the bit at the other end is called the *most significant bit (msb)*, as shown in Figure 1.7(a) for a 6-bit binary number. Similarly, within a word, the bytes are identified as *least significant byte (LSB)* through *most significant byte (MSB)*, as shown in Figure 1.7(b) for a four-byte number written with eight hexadecimal digits.

A *microprocessor* is a processor built on a single chip. Until the 1970's, processors were too complicated to fit on one chip, so mainframe processors were built from boards containing many chips. Intel introduced the first 4-bit microprocessor, called the 4004, in 1971. Now, even the most sophisticated supercomputers are built using microprocessors. We will use the terms microprocessor and processor interchangeably throughout this book.

Figure 1.7 Least and most significant bits and bytes



By handy coincidence, $2^{10} = 1024 \approx 10^3$. Hence, the term *kilo* (Greek for thousand) indicates 2^{10} . For example, 2^{10} bytes is one kilobyte (1 KB). Similarly, *mega* (million) indicates $2^{20} \approx 10^6$, and *giga* (billion) indicates $2^{30} \approx 10^9$. If you know $2^{10} \approx 1$ thousand, $2^{20} \approx 1$ million, $2^{30} \approx 1$ billion, and remember the powers of two up to 2^9 , it is easy to estimate any power of two in your head.

Example 1.6 ESTIMATING POWERS OF TWO

Find the approximate value of 2^{24} without using a calculator.

Solution: Split the exponent into a multiple of ten and the remainder.

$2^{24} = 2^{20} \times 2^4$. $2^{20} \approx 1$ million. $2^4 = 16$. So $2^{24} \approx 16$ million. Technically, $2^{24} = 16,777,216$, but 16 million is close enough for marketing purposes.

1024 bytes is called a *kilobyte* (KB). 1024 bits is called a *kilobit* (Kb or Kbit). Similarly, MB, Mb, GB, and Gb are used for millions and billions of bytes and bits. Memory capacity is usually measured in bytes. Communication speed is usually measured in bits/sec. For example, the maximum speed of a dial-up modem is usually 56 kbits/sec.

1.4.5 Binary Addition

Binary addition is much like decimal addition, but easier, as shown in Figure 1.8. As in decimal addition, if the sum of two numbers is greater than what fits in a single digit, we *carry* a 1 into the next column. Figure 1.8 compares addition of decimal and binary numbers. In the right-most column of Figure 1.8(a), $7 + 9 = 16$, which cannot fit in a single digit because it is greater than 9. So we record the 1's digit, 6, and carry the 10's digit, 1, over to the next column. Likewise, in binary, if the sum of two numbers is greater than 1, we carry the 2's digit over to the next column. For example, in the right-most column of Figure 1.8(b),

Figure 1.8 Addition examples showing carries: (a) decimal (b) binary



the sum $1 + 1 = 2_{10} = 10_2$ cannot fit in a single binary digit. So we record the 1's digit (0) and carry the 2's digit (1) of the result to the next column. In the second column, the sum is $1 + 1 + 1 = 3_{10} = 11_2$. Again, we record the 1's digit (1) and carry the 2's digit (1) to the next column. For obvious reasons, the bit that is carried over to the neighboring column is called the *carry bit*.

Example 1.7 BINARY ADDITION

Compute $0111_2 + 0101_2$.

Solution: Figure 1.9 shows that the sum is 1100_2 . The carries are indicated in blue. We can check our work by repeating the computation in decimal. $0111_2 = 7_{10}$. $0101_2 = 5_{10}$. The sum is $12_{10} = 1100_2$.

$$\begin{array}{r} \text{111} \\ 0111 \\ + 0101 \\ \hline 1100 \end{array}$$

Figure 1.9 Binary addition example

Digital systems usually operate on a fixed number of digits. Addition is said to *overflow* if the result is too big to fit in the available digits. A 4-bit number, for example, has the range $[0, 15]$. 4-bit binary addition overflows if the result exceeds 15. The fifth bit is discarded, producing an incorrect result in the remaining four bits. Overflow can be detected by checking for a carry out of the most significant column.

Example 1.8 ADDITION WITH OVERFLOW

Compute $1101_2 + 0101_2$. Does overflow occur?

Solution: Figure 1.10 shows the sum is 10010_2 . This result overflows the range of a 4-bit binary number. If it must be stored as four bits, the most significant bit is discarded, leaving the incorrect result of 0010_2 . If the computation had been done using numbers with five or more bits, the result 10010_2 would have been correct.

$$\begin{array}{r} \text{111} \\ 1101 \\ + 0101 \\ \hline 10010 \end{array}$$

Figure 1.10 Binary addition example with overflow

1.4.6 Signed Binary Numbers

So far, we have considered only *unsigned* binary numbers that represent positive quantities. We will often want to represent both positive and negative numbers, requiring a different binary number system. Several schemes exist to represent *signed* binary numbers; the two most widely employed are called sign/magnitude and two's complement.

Sign/Magnitude Numbers

Sign/magnitude numbers are intuitively appealing because they match our custom of writing negative numbers with a minus sign followed by the magnitude. An N -bit sign/magnitude number uses the most significant

The \$7 billion Ariane 5 rocket, launched on June 4, 1996, veered off course 40 seconds after launch, broke up, and exploded. The failure was caused when the computer controlling the rocket overflowed its 16-bit range and crashed.

The code had been extensively tested on the Ariane 4 rocket. However, the Ariane 5 had a faster engine that produced larger values for the control computer, leading to the overflow.



(Photograph courtesy of ESA/CNES/ARIANESPACE-Service Optique CS6.)

bit as the sign and the remaining $N-1$ bits as the magnitude (absolute value). A sign bit of 0 indicates positive and a sign bit of 1 indicates negative.

Example 1.9 SIGN/MAGNITUDE NUMBERS

Write 5 and -5 as 4-bit sign/magnitude numbers

Solution: Both numbers have a magnitude of $5_{10} = 101_2$. Thus, $5_{10} = 0101_2$ and $-5_{10} = 1101_2$.

Unfortunately, ordinary binary addition does not work for sign/magnitude numbers. For example, using ordinary addition on $-5_{10} + 5_{10}$ gives $1101_2 + 0101_2 = 10010_2$, which is nonsense.

An N -bit sign/magnitude number spans the range $[-2^{N-1} + 1, 2^{N-1} - 1]$. Sign/magnitude numbers are slightly odd in that both $+0$ and -0 exist. Both indicate zero. As you may expect, it can be troublesome to have two different representations for the same number.

Two's Complement Numbers

Two's complement numbers are identical to unsigned binary numbers except that the most significant bit position has a weight of -2^{N-1} instead of 2^{N-1} . They overcome the shortcomings of sign/magnitude numbers: zero has a single representation, and ordinary addition works.

In two's complement representation, zero is written as all zeros: $00...000_2$. The most positive number has a 0 in the most significant position and 1's elsewhere: $01...111_2 = 2^{N-1} - 1$. The most negative number has a 1 in the most significant position and 0's elsewhere: $10...000_2 = -2^{N-1}$. And -1 is written as all ones: $11...111_2$.

Notice that positive numbers have a 0 in the most significant position and negative numbers have a 1 in this position, so the most significant bit can be viewed as the sign bit. However, the overall number is interpreted differently for two's complement numbers and sign/magnitude numbers.

The sign of a two's complement number is reversed in a process called *taking the two's complement*. The process consists of inverting all of the bits in the number, then adding 1 to the least significant bit position. This is useful to find the representation of a negative number or to determine the magnitude of a negative number.

Example 1.10 TWO'S COMPLEMENT REPRESENTATION OF A NEGATIVE NUMBER

Find the representation of -2_{10} as a 4-bit two's complement number.

Solution: Start with $+2_{10} = 0010_2$. To get -2_{10} , invert the bits and add 1. Inverting 0010_2 produces 1101_2 . $1101_2 + 1 = 1110_2$. So -2_{10} is 1110_2 .

Example 1.11 VALUE OF NEGATIVE TWO'S COMPLEMENT NUMBERS

Find the decimal value of the two's complement number 1001_2 .

Solution: 1001_2 has a leading 1, so it must be negative. To find its magnitude, invert the bits and add 1. Inverting $1001_2 = 0110_2$. $0110_2 + 1 = 0111_2 = 7_{10}$. Hence, $1001_2 = -7_{10}$.

Two's complement numbers have the compelling advantage that addition works properly for both positive and negative numbers. Recall that when adding N -bit numbers, the carry out of the N th bit (i.e., the $N + 1^{\text{th}}$ result bit) is discarded.

Example 1.12 ADDING TWO'S COMPLEMENT NUMBERS

Compute (a) $-2_{10} + 1_{10}$ and (b) $-7_{10} + 7_{10}$ using two's complement numbers.

Solution: (a) $-2_{10} + 1_{10} = 1110_2 + 0001_2 = 1111_2 = -1_{10}$. (b) $-7_{10} + 7_{10} = 1001_2 + 0111_2 = 10000_2$. The fifth bit is discarded, leaving the correct 4-bit result 0000_2 .

Subtraction is performed by taking the two's complement of the second number, then adding.

Example 1.13 SUBTRACTING TWO'S COMPLEMENT NUMBERS

Compute (a) $5_{10} - 3_{10}$ and (b) $3_{10} - 5_{10}$ using 4-bit two's complement numbers.

Solution: (a) $3_{10} = 0011_2$. Take its two's complement to obtain $-3_{10} = 1101_2$. Now add $5_{10} + (-3_{10}) = 0101_2 + 1101_2 = 0010_2 = 2_{10}$. Note that the carry out of the most significant position is discarded because the result is stored in four bits. (b) Take the two's complement of 5_{10} to obtain $-5_{10} = 1011_2$. Now add $3_{10} + (-5_{10}) = 0011_2 + 1011_2 = 1110_2 = -2_{10}$.

The two's complement of 0 is found by inverting all the bits (producing $11\dots111_2$) and adding 1, which produces all 0's, disregarding the carry out of the most significant bit position. Hence, zero is always represented with all 0's. Unlike the sign/magnitude system, the two's complement system has no separate -0 . Zero is considered positive because its sign bit is 0.

Like unsigned numbers, N -bit two's complement numbers represent one of 2^N possible values. However the values are split between positive and negative numbers. For example, a 4-bit unsigned number represents 16 values: 0 to 15. A 4-bit two's complement number also represents 16 values: -8 to 7. In general, the range of an N -bit two's complement number spans $[-2^{N-1}, 2^{N-1} - 1]$. It should make sense that there is one more negative number than positive number because there is no -0 . The most negative number $10\dots000_2 = -2^{N-1}$ is sometimes called the *weird number*. Its two's complement is found by inverting the bits (producing $01\dots111_2$) and adding 1, which produces $10\dots000_2$, the weird number, again. Hence, this negative number has no positive counterpart.

Adding two N -bit positive numbers or negative numbers may cause overflow if the result is greater than $2^{N-1} - 1$ or less than -2^{N-1} . Adding a positive number to a negative number never causes overflow. Unlike unsigned numbers, a carry out of the most significant column does not indicate overflow. Instead, overflow occurs if the two numbers being added have the same sign bit and the result has the opposite sign bit.

Example 1.14 ADDING TWO'S COMPLEMENT NUMBERS WITH OVERFLOW

Compute $4_{10} + 5_{10}$ using 4-bit two's complement numbers. Does the result overflow?

Solution: $4_{10} + 5_{10} = 0100_2 + 0101_2 = 1001_2 = -7_{10}$. The result overflows the range of 4-bit positive two's complement numbers, producing an incorrect negative result. If the computation had been done using five or more bits, the result $01001_2 = 9_{10}$ would have been correct.

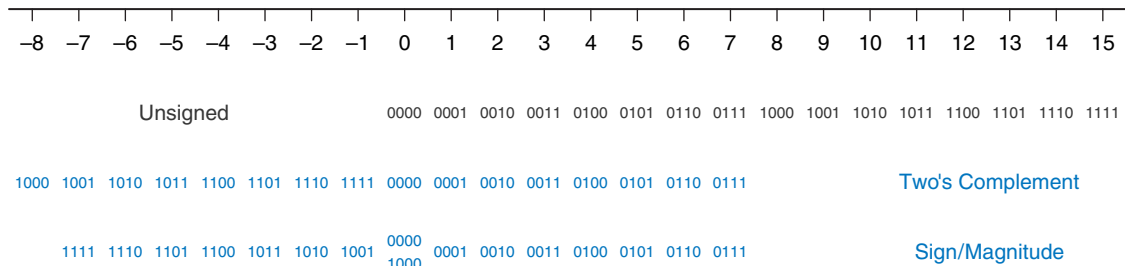
When a two's complement number is extended to more bits, the sign bit must be copied into the most significant bit positions. This process is called *sign extension*. For example, the numbers 3 and -3 are written as 4-bit two's complement numbers 0011 and 1101, respectively. They are sign-extended to seven bits by copying the sign bit into the three new upper bits to form 0000011 and 1111101, respectively.

Comparison of Number Systems

The three most commonly used binary number systems are unsigned, two's complement, and sign/magnitude. Table 1.3 compares the range of N -bit numbers in each of these three systems. Two's complement numbers are convenient because they represent both positive and negative integers and because ordinary addition works for all numbers. Subtraction is performed by negating the second number (i.e., taking the two's

Table 1.3 Range of N -bit numbers

System	Range
Unsigned	$[0, 2^N - 1]$
Sign/Magnitude	$[-2^{N-1} + 1, 2^{N-1} - 1]$
Two's Complement	$[-2^{N-1}, 2^{N-1} - 1]$

**Figure 1.11** Number line and 4-bit binary encodings

complement), and then adding. Unless stated otherwise, assume that all signed binary numbers use two's complement representation.

Figure 1.11 shows a number line indicating the values of 4-bit numbers in each system. Unsigned numbers span the range $[0, 15]$ in regular binary order. Two's complement numbers span the range $[-8, 7]$. The nonnegative numbers $[0, 7]$ share the same encodings as unsigned numbers. The negative numbers $[-8, -1]$ are encoded such that a larger unsigned binary value represents a number closer to 0. Notice that the weird number, 1000, represents -8 and has no positive counterpart. Sign/magnitude numbers span the range $[-7, 7]$. The most significant bit is the sign bit. The positive numbers $[1, 7]$ share the same encodings as unsigned numbers. The negative numbers are symmetric but have the sign bit set. 0 is represented by both 0000 and 1000. Thus, N -bit sign/magnitude numbers represent only $2^N - 1$ integers because of the two representations for 0.

1.5 LOGIC GATES

Now that we know how to use binary variables to represent information, we explore digital systems that perform operations on these binary variables. *Logic gates* are simple digital circuits that take one or more binary inputs and produce a binary output. Logic gates are drawn with a symbol showing the input (or inputs) and the output. Inputs are usually drawn on

the left (or top) and outputs on the right (or bottom). Digital designers typically use letters near the beginning of the alphabet for gate inputs and the letter Y for the gate output. The relationship between the inputs and the output can be described with a truth table or a Boolean equation. A *truth table* lists inputs on the left and the corresponding output on the right. It has one row for each possible combination of inputs. A *Boolean equation* is a mathematical expression using binary variables.

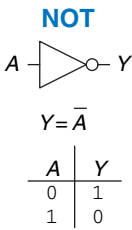


Figure 1.12 NOT gate

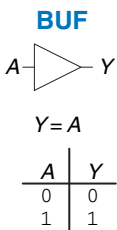


Figure 1.13 Buffer

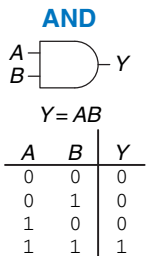


Figure 1.14 AND gate

According to Larry Wall, inventor of the Perl programming language, “the three principal virtues of a programmer are Laziness, Impatience, and Hubris.”

1.5.1 NOT Gate

A *NOT gate* has one input, A, and one output, Y, as shown in Figure 1.12. The NOT gate’s output is the inverse of its input. If A is FALSE, then Y is TRUE. If A is TRUE, then Y is FALSE. This relationship is summarized by the truth table and Boolean equation in the figure. The line over A in the Boolean equation is pronounced *NOT*, so $Y = \bar{A}$ is read “Y equals NOT A.” The NOT gate is also called an *inverter*.

Other texts use a variety of notations for NOT, including $Y = A'$, $Y = \neg A$, $Y = !A$ or $Y = \sim A$. We will use $Y = \bar{A}$ exclusively, but don’t be puzzled if you encounter another notation elsewhere.

1.5.2 Buffer

The other one-input logic gate is called a *buffer* and is shown in Figure 1.13. It simply copies the input to the output.

From the logical point of view, a buffer is no different from a wire, so it might seem useless. However, from the analog point of view, the buffer might have desirable characteristics such as the ability to deliver large amounts of current to a motor or the ability to quickly send its output to many gates. This is an example of why we need to consider multiple levels of abstraction to fully understand a system; the digital abstraction hides the real purpose of a buffer.

The triangle symbol indicates a buffer. A circle on the output is called a *bubble* and indicates inversion, as was seen in the NOT gate symbol of Figure 1.12.

1.5.3 AND Gate

Two-input logic gates are more interesting. The *AND gate* shown in Figure 1.14 produces a TRUE output, Y, if and only if both A and B are TRUE. Otherwise, the output is FALSE. By convention, the inputs are listed in the order 00, 01, 10, 11, as if you were counting in binary. The Boolean equation for an AND gate can be written in several ways: $Y = A \cdot B$, $Y = AB$, or $Y = A \cap B$. The \cap symbol is pronounced “intersection” and is preferred by logicians. We prefer $Y = AB$, read “Y equals A and B,” because we are lazy.

1.5.4 OR Gate

The *OR gate* shown in Figure 1.15 produces a TRUE output, Y , if either A or B (or both) are TRUE. The Boolean equation for an OR gate is written as $Y = A + B$ or $Y = A \cup B$. The \cup symbol is pronounced union and is preferred by logicians. Digital designers normally use the $+$ notation, $Y = A + B$ is pronounced “ Y equals A or B .”

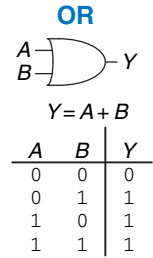


Figure 1.15 OR gate

1.5.5 Other Two-Input Gates

Figure 1.16 shows other common two-input logic gates. *XOR* (exclusive OR, pronounced “ex-OR”) is TRUE if A or B , but not both, are TRUE. The XOR operation is indicated by \oplus , a plus sign with a circle around it. Any gate can be followed by a bubble to invert its operation. The *NAND gate* performs NOT AND. Its output is TRUE unless both inputs are TRUE. The *NOR gate* performs NOT OR. Its output is TRUE if neither A nor B is TRUE. An N -input XOR gate is sometimes called a *parity gate* and produces a TRUE output if an odd number of inputs are TRUE. As with two-input gates, the input combinations in the truth table are listed in counting order.

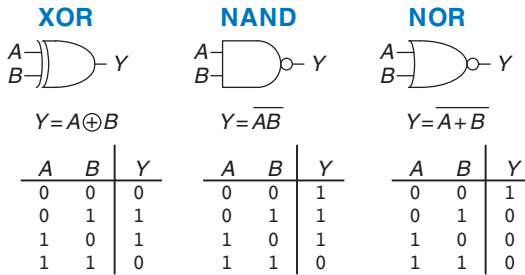
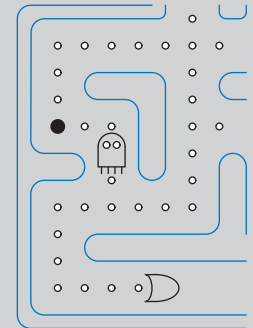


Figure 1.16 More two-input logic gates

A silly way to remember the OR symbol is that its input side is curved like Pacman’s mouth, so the gate is hungry and willing to eat any TRUE inputs it can find!



Example 1.15 XNOR GATE

Figure 1.17 shows the symbol and Boolean equation for a two-input *XNOR gate* that performs the inverse of an XOR. Complete the truth table.

Solution: Figure 1.18 shows the truth table. The XNOR output is TRUE if both inputs are FALSE or both inputs are TRUE. The two-input XNOR gate is sometimes called an *equality gate* because its output is TRUE when the inputs are equal.

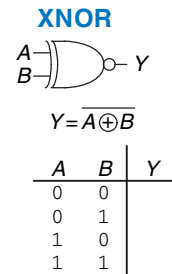


Figure 1.17 XNOR gate

1.5.6 Multiple-Input Gates

Many Boolean functions of three or more inputs exist. The most common are AND, OR, XOR, NAND, NOR, and XNOR. An N -input AND gate

A	B	Y
0	0	1
0	1	0
1	0	0
1	1	1

Figure 1.18 XNOR truth table

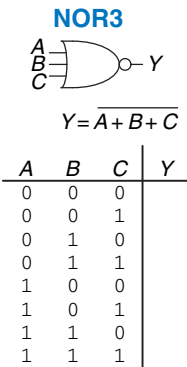


Figure 1.19 Three-input NOR gate

A	B	C	Y
0	0	0	1
0	0	1	0
0	1	0	0
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	0
1	1	1	0

Figure 1.20 Three-input NOR truth table

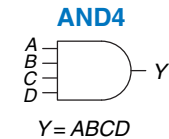


Figure 1.21 Four-input AND gate

produces a TRUE output when all N inputs are TRUE. An N -input OR gate produces a TRUE output when at least one input is TRUE.

Example 1.16 THREE-INPUT NOR GATE

Figure 1.19 shows the symbol and Boolean equation for a three-input NOR gate. Complete the truth table.

Solution: Figure 1.20 shows the truth table. The output is TRUE only if none of the inputs are TRUE.

Example 1.17 FOUR-INPUT AND GATE

Figure 1.21 shows the symbol and Boolean equation for a four-input AND gate. Create a truth table.

Solution: Figure 1.22 shows the truth table. The output is TRUE only if all of the inputs are TRUE.

1.6 BENEATH THE DIGITAL ABSTRACTION

A digital system uses discrete-valued variables. However, the variables are represented by continuous physical quantities such as the voltage on a wire, the position of a gear, or the level of fluid in a cylinder. Hence, the designer must choose a way to relate the continuous value to the discrete value.

For example, consider representing a binary signal A with a voltage on a wire. Let 0 volts (V) indicate $A = 0$ and 5 V indicate $A = 1$. Any real system must tolerate some noise, so 4.97 V probably ought to be interpreted as $A = 1$ as well. But what about 4.3 V? Or 2.8 V? Or 2.500000 V?

1.6.1 Supply Voltage

Suppose the lowest voltage in the system is 0 V, also called *ground* or GND. The highest voltage in the system comes from the power supply and is usually called V_{DD} . In 1970's and 1980's technology, V_{DD} was generally 5 V. As chips have progressed to smaller transistors, V_{DD} has dropped to 3.3 V, 2.5 V, 1.8 V, 1.5 V, 1.2 V, or even lower to save power and avoid overloading the transistors.

1.6.2 Logic Levels

The mapping of a continuous variable onto a discrete binary variable is done by defining *logic levels*, as shown in Figure 1.23. The first gate is called the *driver* and the second gate is called the *receiver*. The output of the driver is

connected to the input of the receiver. The driver produces a LOW (0) output in the range of 0 to V_{OL} or a HIGH (1) output in the range of V_{OH} to V_{DD} . If the receiver gets an input in the range of 0 to V_{IL} , it will consider the input to be LOW. If the receiver gets an input in the range of V_{IH} to V_{DD} , it will consider the input to be HIGH. If, for some reason such as noise or faulty components, the receiver's input should fall in the *forbidden zone* between V_{IL} and V_{IH} , the behavior of the gate is unpredictable. V_{OH} , V_{OL} , V_{IH} , and V_{IL} are called the output and input high and low logic levels.

1.6.3 Noise Margins

If the output of the driver is to be correctly interpreted at the input of the receiver, we must choose $V_{OL} < V_{IL}$ and $V_{OH} > V_{IH}$. Thus, even if the output of the driver is contaminated by some noise, the input of the receiver will still detect the correct logic level. The *noise margin* is the amount of noise that could be added to a worst-case output such that the signal can still be interpreted as a valid input. As can be seen in Figure 1.23, the low and high noise margins are, respectively

$$NM_L = V_{IL} - V_{OL} \quad (1.2)$$

$$NM_H = V_{OH} - V_{IH} \quad (1.3)$$

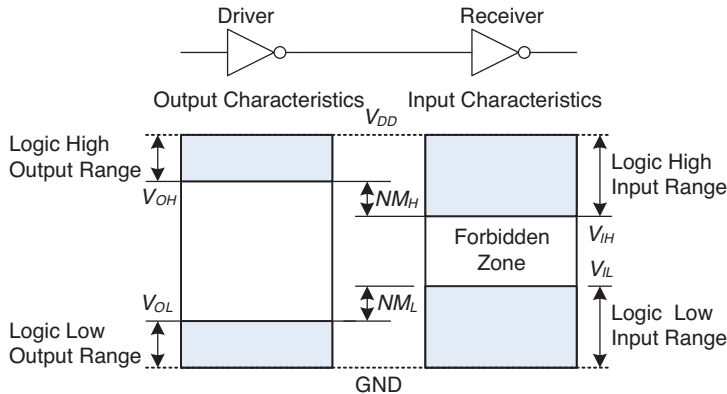


Figure 1.23 Logic levels and noise margins

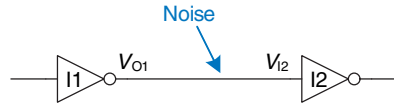
A	B	C	D	Y
0	0	0	0	0
0	0	0	1	0
0	0	1	0	0
0	0	1	1	0
0	1	0	0	0
0	1	0	1	0
0	1	1	0	0
0	1	1	1	0
1	0	0	0	0
1	0	0	1	0
1	0	1	0	0
1	0	1	1	0
1	1	0	0	0
1	1	0	1	0
1	1	1	0	0
1	1	1	1	1

Figure 1.22 Four-input AND truth table

V_{DD} stands for the voltage on the *drain* of a metal-oxide-semiconductor transistor, used to build most modern chips. The power supply voltage is also sometimes called V_{CC} , standing for the voltage on the *collector* of a bipolar junction transistor used to build chips in an older technology. Ground is sometimes called V_{SS} because it is the voltage on the *source* of a metal-oxide-semiconductor transistor. See Section 1.7 for more information on transistors.

Example 1.18 CALCULATING NOISE MARGINS

Consider the inverter circuit of Figure 1.24. V_{O1} is the output voltage of inverter I1, and V_{I2} is the input voltage of inverter I2. Both inverters have the following characteristics: $V_{DD} = 5$ V, $V_{IL} = 1.35$ V, $V_{IH} = 3.15$ V, $V_{OL} = 0.33$ V, and $V_{OH} = 3.84$ V. What are the inverter low and high noise margins? Can the circuit tolerate 1 V of noise between V_{O1} and V_{I2} ?

Figure 1.24 Inverter circuit

Solution: The inverter noise margins are: $NM_L = V_{IL} - V_{OL} = (1.35 \text{ V} - 0.33 \text{ V}) = 1.02 \text{ V}$, $NM_H = V_{OH} - V_{IH} = (3.84 \text{ V} - 3.15 \text{ V}) = 0.69 \text{ V}$. The circuit can tolerate 1 V of noise when the output is LOW ($NM_L = 1.02 \text{ V}$) but not when the output is HIGH ($NM_H = 0.69 \text{ V}$). For example, suppose the driver, I1, outputs its worst-case HIGH value, $V_{O1} = V_{OH} = 3.84 \text{ V}$. If noise causes the voltage to droop by 1 V before reaching the input of the receiver, $V_{I2} = (3.84 \text{ V} - 1 \text{ V}) = 2.84 \text{ V}$. This is less than the acceptable input HIGH value, $V_{IH} = 3.15 \text{ V}$, so the receiver may not sense a proper HIGH input.

DC indicates behavior when an input voltage is held constant or changes slowly enough for the rest of the system to keep up. The term's historical root comes from *direct current*, a method of transmitting power across a line with a constant voltage. In contrast, the *transient response* of a circuit is the behavior when an input voltage changes rapidly. Section 2.9 explores transient response further.

1.6.4 DC Transfer Characteristics

To understand the limits of the digital abstraction, we must delve into the analog behavior of a gate. The *DC transfer characteristics* of a gate describe the output voltage as a function of the input voltage when the input is changed slowly enough that the output can keep up. They are called transfer characteristics because they describe the relationship between input and output voltages.

An ideal inverter would have an abrupt switching threshold at $V_{DD}/2$, as shown in Figure 1.25(a). For $V(A) < V_{DD}/2$, $V(Y) = V_{DD}$. For $V(A) > V_{DD}/2$, $V(Y) = 0$. In such a case, $V_{IH} = V_{IL} = V_{DD}/2$. $V_{OH} = V_{DD}$ and $V_{OL} = 0$.

A real inverter changes more gradually between the extremes, as shown in Figure 1.25(b). When the input voltage $V(A)$ is 0, the output voltage $V(Y) = V_{DD}$. When $V(A) = V_{DD}$, $V(Y) = 0$. However, the transition between these endpoints is smooth and may not be centered at exactly $V_{DD}/2$. This raises the question of how to define the logic levels.

A reasonable place to choose the logic levels is where the slope of the transfer characteristic $dV(Y)/dV(A)$ is -1 . These two points are called the *unity gain points*. Choosing logic levels at the unity gain points usually maximizes the noise margins. If V_{IL} were reduced, V_{OH} would only increase by a small amount. But if V_{IL} were increased, V_{OH} would drop precipitously.

1.6.5 The Static Discipline

To avoid inputs falling into the forbidden zone, digital logic gates are designed to conform to the *static discipline*. The static discipline requires that, given logically valid inputs, every circuit element will produce logically valid outputs.

By conforming to the static discipline, digital designers sacrifice the freedom of using arbitrary analog circuit elements in return for the simplicity and robustness of digital circuits. They raise the level of abstraction

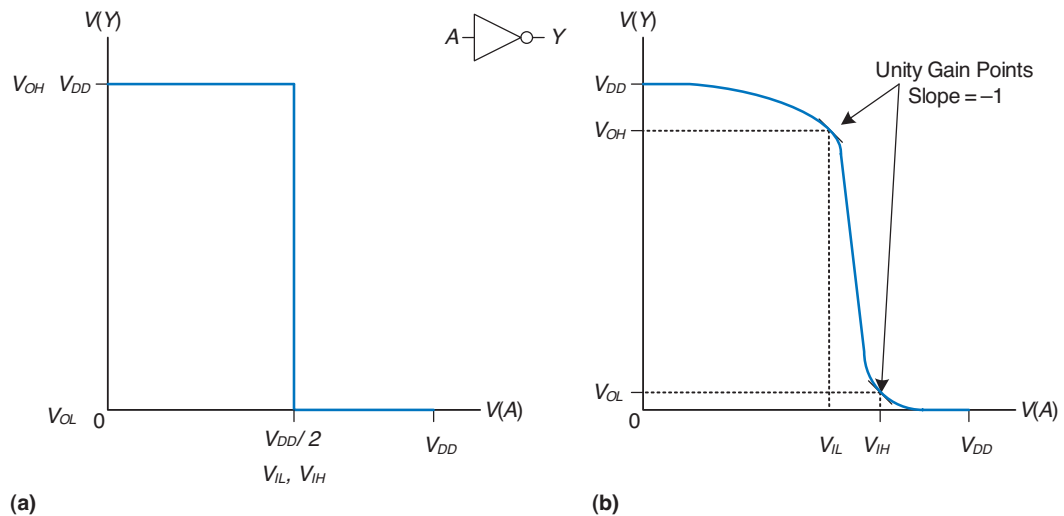


Figure 1.25 DC transfer characteristics and logic levels

from analog to digital, increasing design productivity by hiding needless detail.

The choice of V_{DD} and logic levels is arbitrary, but all gates that communicate must have compatible logic levels. Therefore, gates are grouped into *logic families* such that all gates in a logic family obey the static discipline when used with other gates in the family. Logic gates in the same logic family snap together like Legos in that they use consistent power supply voltages and logic levels.

Four major logic families that predominated from the 1970's through the 1990's are Transistor-Transistor Logic (TTL), Complementary Metal-Oxide-Semiconductor Logic (CMOS, pronounced sea-moss), Low Voltage TTL Logic (LVTTTL), and Low Voltage CMOS Logic (LVCMOS). Their logic levels are compared in Table 1.4. Since then, logic families have balkanized with a proliferation of even lower power supply voltages. Appendix A.6 revisits popular logic families in more detail.

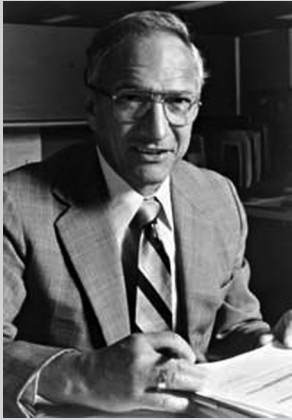
Table 1.4 Logic levels of 5 V and 3.3 V logic families

Logic Family	V_{DD}	V_{IL}	V_{IH}	V_{OL}	V_{OH}
TTL	5 (4.75–5.25)	0.8	2.0	0.4	2.4
CMOS	5 (4.5–6)	1.35	3.15	0.33	3.84
LVTTTL	3.3 (3–3.6)	0.8	2.0	0.4	2.4
LVCMOS	3.3 (3–3.6)	0.9	1.8	0.36	2.7

Table 1.5 Compatibility of logic families

		Receiver			
		TTL	CMOS	LVTTL	LVC MOS
Driver	TTL	OK	NO: $V_{OH} < V_{IH}$	MAYBE ^a	MAYBE ^a
	CMOS	OK	OK	MAYBE ^a	MAYBE ^a
	LVTTL	OK	NO: $V_{OH} < V_{IH}$	OK	OK
	LVC MOS	OK	NO: $V_{OH} < V_{IH}$	OK	OK

^a As long as a 5 V HIGH level does not damage the receiver input.



Robert Noyce, 1927–1990. Born in Burlington, Iowa. Received a B. A. in physics from Grinnell College and a Ph.D. in physics from MIT. Nicknamed “Mayor of Silicon Valley” for his profound influence on the industry. Cofounded Fairchild Semiconductor in 1957 and Intel in 1968. Coinvented the integrated circuit. Many engineers from his teams went on to found other seminal semiconductor companies (photograph © 2006, Intel Corporation. Reproduced by permission).

Example 1.19 LOGIC FAMILY COMPATIBILITY

Which of the logic families in Table 1.4 can communicate with each other reliably?

Solution: Table 1.5 lists which logic families have compatible logic levels. Note that a 5 V logic family such as TTL or CMOS may produce an output voltage as HIGH as 5 V. If this 5 V signal drives the input of a 3.3 V logic family such as LVTTL or LVC MOS, it can damage the receiver, unless the receiver is specially designed to be “5-volt compatible.”

1.7 CMOS TRANSISTORS*

This section and other sections marked with a * are optional and are not necessary to understand the main flow of the book.

Babbage’s Analytical Engine was built from gears, and early electrical computers used relays or vacuum tubes. Modern computers use transistors because they are cheap, small, and reliable. *Transistors* are electrically controlled switches that turn ON or OFF when a voltage or current is applied to a control terminal. The two main types of transistors are *bipolar junction transistors* and *metal-oxide-semiconductor field effect transistors* (MOSFETs or MOS *transistors*, pronounced “moss-fets” or “M-O-S”, respectively).

In 1958, Jack Kilby at Texas Instruments built the first integrated circuit containing two transistors. In 1959, Robert Noyce at Fairchild Semiconductor patented a method of interconnecting multiple transistors on a single silicon chip. At the time, transistors cost about \$10 each.

Thanks to more than four decades of unprecedented manufacturing advances, engineers can now pack roughly three billion MOSFETs onto a 1 cm² chip of silicon, and these transistors cost less than 1 microcent apiece. The capacity and cost continue to improve by an order of magnitude every 8 years or so. MOSFETs are now the building blocks of almost all digital

systems. In this section, we will peer beneath the digital abstraction to see how logic gates are built from MOSFETs.

1.7.1 Semiconductors

MOS transistors are built from silicon, the predominant atom in rock and sand. Silicon (Si) is a group IV atom, so it has four electrons in its valence shell and forms bonds with four adjacent atoms, resulting in a crystalline *lattice*. Figure 1.26(a) shows the lattice in two dimensions for ease of drawing, but remember that the lattice actually forms a cubic crystal. In the figure, a line represents a covalent bond. By itself, silicon is a poor conductor because all the electrons are tied up in covalent bonds. However, it becomes a better conductor when small amounts of impurities, called *dopant* atoms, are carefully added. If a group V dopant such as arsenic (As) is added, the dopant atoms have an extra electron that is not involved in the bonds. The electron can easily move about the lattice, leaving an ionized dopant atom (As^+) behind, as shown in Figure 1.26(b). The electron carries a negative charge, so we call arsenic an *n-type* dopant. On the other hand, if a group III dopant such as boron (B) is added, the dopant atoms are missing an electron, as shown in Figure 1.26(c). This missing electron is called a *hole*. An electron from a neighboring silicon atom may move over to fill the missing bond, forming an ionized dopant atom (B^-) and leaving a hole at the neighboring silicon atom. In a similar fashion, the hole can migrate around the lattice. The hole is a lack of negative charge, so it acts like a positively charged particle. Hence, we call boron a *p-type* dopant. Because the conductivity of silicon changes over many orders of magnitude depending on the concentration of dopants, silicon is called a *semiconductor*.

1.7.2 Diodes

The junction between p-type and n-type silicon is called a *diode*. The p-type region is called the *anode* and the n-type region is called the *cathode*, as illustrated in Figure 1.27. When the voltage on the anode rises above the voltage on the cathode, the diode is *forward biased*, and current

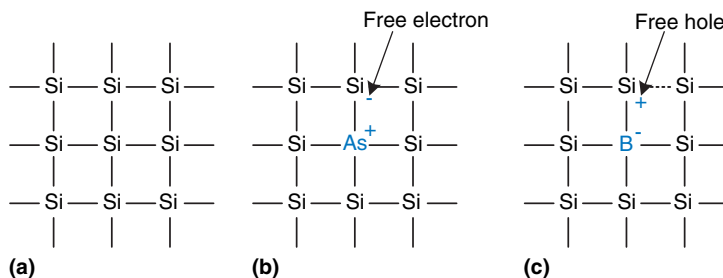


Figure 1.26 Silicon lattice and dopant atoms

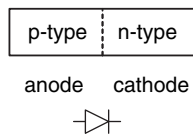


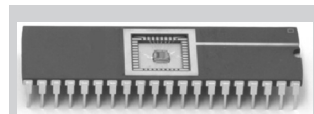
Figure 1.27 The p-n junction diode structure and symbol



Figure 1.28 Capacitor symbol



Technicians in an Intel clean room wear Gore-Tex bunny suits to prevent particulates from their hair, skin, and clothing from contaminating the microscopic transistors on silicon wafers (photograph © 2006, Intel Corporation. Reproduced by permission).



A 40-pin dual-inline package (DIP) contains a small chip (scarcely visible) in the center that is connected to 40 metal pins, 20 on a side, by gold wires thinner than a strand of hair (photograph by Kevin Mapp. © 2006 Harvey Mudd College).

flows through the diode from the anode to the cathode. But when the anode voltage is lower than the voltage on the cathode, the diode is *reverse biased*, and no current flows. The diode symbol intuitively shows that current only flows in one direction.

1.7.3 Capacitors

A *capacitor* consists of two conductors separated by an insulator. When a voltage V is applied to one of the conductors, the conductor accumulates electric *charge* Q and the other conductor accumulates the opposite charge $-Q$. The *capacitance* C of the capacitor is the ratio of charge to voltage: $C = Q/V$. The capacitance is proportional to the size of the conductors and inversely proportional to the distance between them. The symbol for a capacitor is shown in [Figure 1.28](#).

Capacitance is important because charging or discharging a conductor takes time and energy. More capacitance means that a circuit will be slower and require more energy to operate. Speed and energy will be discussed throughout this book.

1.7.4 nMOS and pMOS Transistors

A MOSFET is a sandwich of several layers of conducting and insulating materials. MOSFETs are built on thin flat *wafers* of silicon of about 15 to 30 cm in diameter. The manufacturing process begins with a bare wafer. The process involves a sequence of steps in which dopants are implanted into the silicon, thin films of silicon dioxide and silicon are grown, and metal is deposited. Between each step, the wafer is *patterned* so that the materials appear only where they are desired. Because transistors are a fraction of a micron¹ in length and the entire wafer is processed at once, it is inexpensive to manufacture billions of transistors at a time. Once processing is complete, the wafer is cut into rectangles called *chips* or *dice* that contain thousands, millions, or even billions of transistors. The chip is tested, then placed in a plastic or ceramic *package* with metal pins to connect it to a circuit board.

The MOSFET sandwich consists of a conducting layer called the *gate* on top of an insulating layer of *silicon dioxide* (SiO_2) on top of the silicon wafer, called the *substrate*. Historically, the gate was constructed from metal, hence the name metal-oxide-semiconductor. Modern manufacturing processes use polycrystalline silicon for the gate because it does not melt during subsequent high-temperature processing steps. Silicon dioxide is better known as glass and is often simply called *oxide* in the semiconductor industry. The metal-oxide-semiconductor sandwich forms a capacitor, in which a thin layer of insulating oxide called a *dielectric* separates the metal and semiconductor plates.

¹ 1 μm = 1 micron = 10^{-6} m.

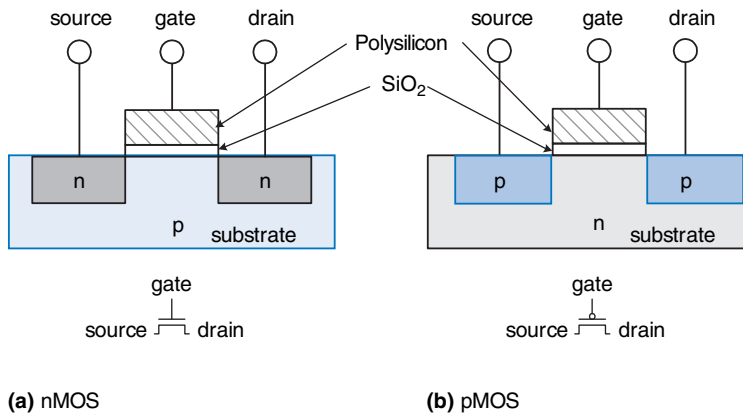


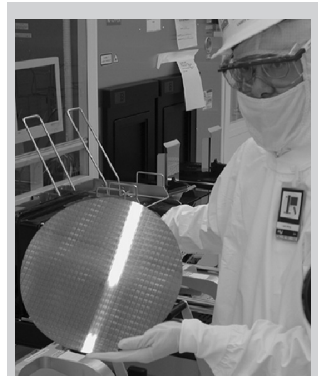
Figure 1.29 nMOS and pMOS transistors

There are two flavors of MOSFETs: nMOS and pMOS (pronounced “n-moss” and “p-moss”). Figure 1.29 shows cross-sections of each type, made by sawing through a wafer and looking at it from the side. The n-type transistors, called *nMOS*, have regions of n-type dopants adjacent to the gate called the *source* and the *drain* and are built on a p-type semiconductor substrate. The *pMOS* transistors are just the opposite, consisting of p-type source and drain regions in an n-type *substrate*.

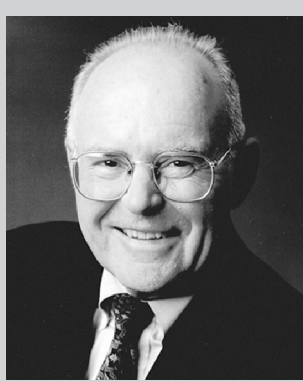
A MOSFET behaves as a voltage-controlled switch in which the gate voltage creates an electric field that turns ON or OFF a connection between the source and drain. The term *field effect transistor* comes from this principle of operation. Let us start by exploring the operation of an nMOS transistor.

The substrate of an nMOS transistor is normally tied to GND, the lowest voltage in the system. First, consider the situation when the gate is also at 0 V, as shown in Figure 1.30(a). The diodes between the source or drain and the substrate are reverse biased because the source or drain voltage is nonnegative. Hence, there is no path for current to flow between the source and drain, so the transistor is OFF. Now, consider when the gate is raised to V_{DD} , as shown in Figure 1.30(b). When a positive voltage is applied to the top plate of a capacitor, it establishes an electric field that attracts positive charge on the top plate and negative charge to the bottom plate. If the voltage is sufficiently large, so much negative charge is attracted to the underside of the gate that the region *inverts* from p-type to effectively become n-type. This inverted region is called the *channel*. Now the transistor has a continuous path from the n-type source through the n-type channel to the n-type drain, so electrons can flow from source to drain. The transistor is ON. The gate voltage required to turn on a transistor is called the *threshold voltage*, V_t , and is typically 0.3 to 0.7 V.

The source and drain terminals are physically symmetric. However, we say that charge flows from the source to the drain. In an nMOS transistor, the charge is carried by electrons, which flow from negative voltage to positive voltage. In a pMOS transistor, the charge is carried by holes, which flow from positive voltage to negative voltage. If we draw schematics with the most positive voltage at the top and the most negative at the bottom, the source of (negative) charges in an nMOS transistor is the bottom terminal and the source of (positive) charges in a pMOS transistor is the top terminal.



A technician holds a 12-inch wafer containing hundreds of microprocessor chips (photograph © 2006, Intel Corporation. Reproduced by permission).



Gordon Moore, 1929–. Born in San Francisco. Received a B.S. in chemistry from UC Berkeley and a Ph.D. in chemistry and physics from Caltech. Cofounded Intel in 1968 with Robert Noyce. Observed in 1965 that the number of transistors on a computer chip doubles every year. This trend has become known as *Moore's Law*. Since 1975, transistor counts have doubled every two years.

A corollary of Moore's Law is that microprocessor performance doubles every 18 to 24 months. Semiconductor sales have also increased exponentially.

Moore's Law has driven the incredible advances of the semiconductor industry for 50 years as the feature size of transistors has dropped from more than $10\ \mu\text{m}$ to only 28 nm. However, this progress is showing signs of slowing below the 28 nm node because building transistors much smaller than the wavelength of light is expensive. (Photograph © 2006, Intel Corporation. Reproduced by permission.)

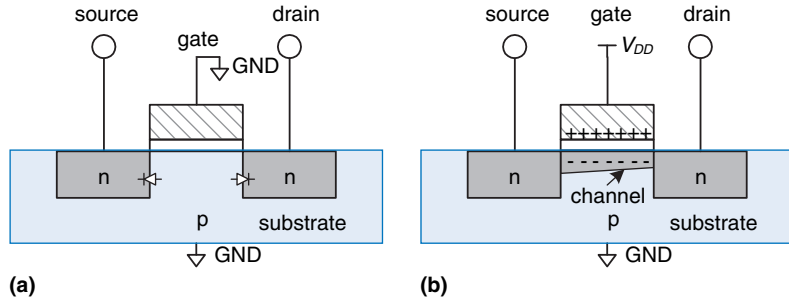


Figure 1.30 nMOS transistor operation

pMOS transistors work in just the opposite fashion, as might be guessed from the bubble on their symbol shown in Figure 1.31. The substrate is tied to V_{DD} . When the gate is also at V_{DD} , the pMOS transistor is OFF. When the gate is at GND, the channel inverts to p-type and the pMOS transistor is ON.

Unfortunately, MOSFETs are not perfect switches. In particular, nMOS transistors pass 0's well but pass 1's poorly. Specifically, when the gate of an nMOS transistor is at V_{DD} , the drain will only swing between 0 and $V_{DD} - V_t$. Similarly, pMOS transistors pass 1's well but 0's poorly. However, we will see that it is possible to build logic gates that use transistors only in their good mode.

nMOS transistors need a p-type substrate, and pMOS transistors need an n-type substrate. To build both flavors of transistors on the same chip, manufacturing processes typically start with a p-type wafer, then implant n-type regions called *wells* where the pMOS transistors should go. These processes that provide both flavors of transistors are called Complementary MOS or CMOS. CMOS processes are used to build the vast majority of all transistors fabricated today.

In summary, CMOS processes give us two types of electrically controlled switches, as shown in Figure 1.31. The voltage at the gate (g) regulates the flow of current between the source (s) and drain (d). nMOS

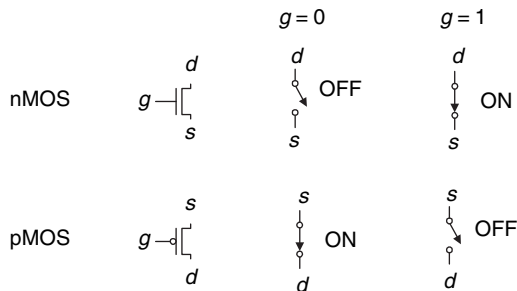


Figure 1.31 Switch models of MOSFETs

transistors are OFF when the gate is 0 and ON when the gate is 1. pMOS transistors are just the opposite: ON when the gate is 0 and OFF when the gate is 1.

1.7.5 CMOS NOT Gate

Figure 1.32 shows a schematic of a NOT gate built with CMOS transistors. The triangle indicates GND, and the flat bar indicates V_{DD} ; these labels will be omitted from future schematics. The nMOS transistor, N1, is connected between GND and the Y output. The pMOS transistor, P1, is connected between V_{DD} and the Y output. Both transistor gates are controlled by the input, A.

If $A = 0$, N1 is OFF and P1 is ON. Hence, Y is connected to V_{DD} but not to GND, and is pulled up to a logic 1. P1 passes a good 1. If $A = 1$, N1 is ON and P1 is OFF, and Y is pulled down to a logic 0. N1 passes a good 0. Checking against the truth table in Figure 1.12, we see that the circuit is indeed a NOT gate.

1.7.6 Other CMOS Logic Gates

Figure 1.33 shows a schematic of a two-input NAND gate. In schematic diagrams, wires are always joined at three-way junctions. They are joined at four-way junctions only if a dot is shown. The nMOS transistors N1 and N2 are connected in series; both nMOS transistors must be ON to pull the output down to GND. The pMOS transistors P1 and P2 are in parallel; only one pMOS transistor must be ON to pull the output up to V_{DD} . Table 1.6 lists the operation of the pull-down and pull-up networks and the state of the output, demonstrating that the gate does function as a NAND. For example, when $A = 1$ and $B = 0$, N1 is ON, but N2 is OFF, blocking the path from Y to GND. P1 is OFF, but P2 is ON, creating a path from V_{DD} to Y. Therefore, Y is pulled up to 1.

Figure 1.34 shows the general form used to construct any inverting logic gate, such as NOT, NAND, or NOR. nMOS transistors are good at passing 0's, so a pull-down network of nMOS transistors is placed between the output and GND to pull the output down to 0. pMOS transistors are

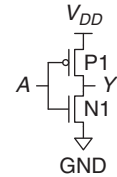


Figure 1.32 NOT gate schematic

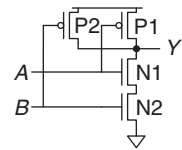


Figure 1.33 Two-input NAND gate schematic

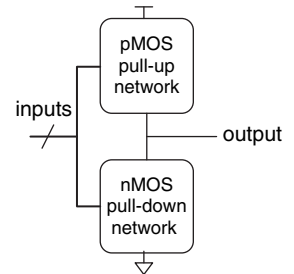


Figure 1.34 General form of an inverting logic gate

Table 1.6 NAND gate operation

A	B	Pull-Down Network	Pull-Up Network	Y
0	0	OFF	ON	1
0	1	OFF	ON	1
1	0	OFF	ON	1
1	1	ON	OFF	0

Experienced designers claim that electronic devices operate because they contain *magic smoke*. They confirm this theory with the observation that if the magic smoke is ever let out of the device, it ceases to work.

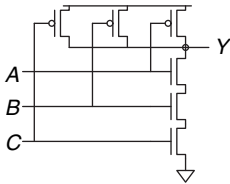


Figure 1.35 Three-input NAND gate schematic

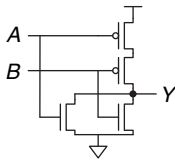


Figure 1.36 Two-input NOR gate schematic

good at passing 1's, so a pull-up network of pMOS transistors is placed between the output and V_{DD} to pull the output up to 1. The networks may consist of transistors in series or in parallel. When transistors are in parallel, the network is ON if either transistor is ON. When transistors are in series, the network is ON only if both transistors are ON. The slash across the input wire indicates that the gate may receive multiple inputs.

If both the pull-up and pull-down networks were ON simultaneously, a *short circuit* would exist between V_{DD} and GND. The output of the gate might be in the forbidden zone and the transistors would consume large amounts of power, possibly enough to burn out. On the other hand, if both the pull-up and pull-down networks were OFF simultaneously, the output would be connected to neither V_{DD} nor GND. We say that the output *floats*. Its value is again undefined. Floating outputs are usually undesirable, but in Section 2.6 we will see how they can occasionally be used to the designer's advantage.

In a properly functioning logic gate, one of the networks should be ON and the other OFF at any given time, so that the output is pulled HIGH or LOW but not shorted or floating. We can guarantee this by using the rule of *conduction complements*. When nMOS transistors are in series, the pMOS transistors must be in parallel. When nMOS transistors are in parallel, the pMOS transistors must be in series.

Example 1.20 THREE-INPUT NAND SCHEMATIC

Draw a schematic for a three-input NAND gate using CMOS transistors.

Solution: The NAND gate should produce a 0 output only when all three inputs are 1. Hence, the pull-down network should have three nMOS transistors in series. By the conduction complements rule, the pMOS transistors must be in parallel. Such a gate is shown in Figure 1.35; you can verify the function by checking that it has the correct truth table.

Example 1.21 TWO-INPUT NOR SCHEMATIC

Draw a schematic for a two-input NOR gate using CMOS transistors.

Solution: The NOR gate should produce a 0 output if either input is 1. Hence, the pull-down network should have two nMOS transistors in parallel. By the conduction complements rule, the pMOS transistors must be in series. Such a gate is shown in Figure 1.36.

Example 1.22 TWO-INPUT AND SCHEMATIC

Draw a schematic for a two-input AND gate.

Solution: It is impossible to build an AND gate with a single CMOS gate. However, building NAND and NOT gates is easy. Thus, the best way to build an AND gate using CMOS transistors is to use a NAND followed by a NOT, as shown in Figure 1.37.

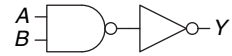


Figure 1.37 Two-input AND gate schematic

1.7.7 Transmission Gates

At times, designers find it convenient to use an ideal switch that can pass both 0 and 1 well. Recall that nMOS transistors are good at passing 0 and pMOS transistors are good at passing 1, so the parallel combination of the two passes both values well. Figure 1.38 shows such a circuit, called a *transmission gate* or *pass gate*. The two sides of the switch are called *A* and *B* because a switch is bidirectional and has no preferred input or output side. The control signals are called *enables*, EN and \overline{EN} . When $EN = 0$ and $\overline{EN} = 1$, both transistors are OFF. Hence, the transmission gate is OFF or disabled, so *A* and *B* are not connected. When $EN = 1$ and $\overline{EN} = 0$, the transmission gate is ON or enabled, and any logic value can flow between *A* and *B*.

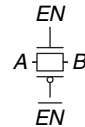


Figure 1.38 Transmission gate

1.7.8 Pseudo-nMOS Logic

An *N*-input CMOS NOR gate uses *N* nMOS transistors in parallel and *N* pMOS transistors in series. Transistors in series are slower than transistors in parallel, just as resistors in series have more resistance than resistors in parallel. Moreover, pMOS transistors are slower than nMOS transistors because holes cannot move around the silicon lattice as fast as electrons. Therefore the parallel nMOS transistors are fast and the series pMOS transistors are slow, especially when many are in series.

Pseudo-nMOS logic replaces the slow stack of pMOS transistors with a single weak pMOS transistor that is always ON, as shown in Figure 1.39. This pMOS transistor is often called a *weak pull-up*. The physical dimensions of the pMOS transistor are selected so that the pMOS transistor will pull the output *Y* HIGH weakly—that is, only if none of the nMOS transistors are ON. But if any nMOS transistor is ON, it overpowers the weak pull-up and pulls *Y* down close enough to GND to produce a logic 0.

The advantage of pseudo-nMOS logic is that it can be used to build fast NOR gates with many inputs. For example, Figure 1.40 shows a pseudo-nMOS four-input NOR. Pseudo-nMOS gates are useful for certain memory and logic arrays discussed in Chapter 5. The disadvantage is that a short circuit exists between V_{DD} and GND when the output is LOW; the weak pMOS and nMOS transistors are both ON. The short circuit draws continuous power, so pseudo-nMOS logic must be used sparingly.

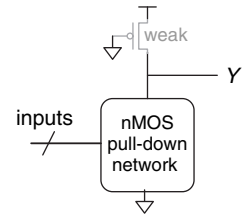


Figure 1.39 Generic pseudo-nMOS gate

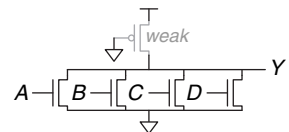


Figure 1.40 Pseudo-nMOS four-input NOR gate

Pseudo-nMOS gates got their name from the 1970's, when manufacturing processes only had nMOS transistors. A weak nMOS transistor was used to pull the output HIGH because pMOS transistors were not available.

1.8 POWER CONSUMPTION*

Power consumption is the amount of energy used per unit time. Power consumption is of great importance in digital systems. The battery life of portable systems such as cell phones and laptop computers is limited by power consumption. Power is also significant for systems that are plugged in, because electricity costs money and because the system will overheat if it draws too much power.

Digital systems draw both *dynamic* and *static* power. Dynamic power is the power used to charge capacitance as signals change between 0 and 1. Static power is the power used even when signals do not change and the system is idle.

Logic gates and the wires that connect them have capacitance. The energy drawn from the power supply to charge a capacitance C to voltage V_{DD} is CV_{DD}^2 . If the voltage on the capacitor switches at frequency f (i.e., f times per second), it charges the capacitor $f/2$ times and discharges it $f/2$ times per second. Discharging does not draw energy from the power supply, so the dynamic power consumption is

$$P_{\text{dynamic}} = \frac{1}{2} CV_{DD}^2 f \quad (1.4)$$

Electrical systems draw some current even when they are idle. When transistors are OFF, they leak a small amount of current. Some circuits, such as the pseudo-nMOS gate discussed in [Section 1.7.8](#), have a path from V_{DD} to GND through which current flows continuously. The total static current, I_{DD} , is also called the *leakage current* or the *quiescent supply current* flowing between V_{DD} and GND. The static power consumption is proportional to this static current:

$$P_{\text{static}} = I_{DD} V_{DD} \quad (1.5)$$

Example 1.23 POWER CONSUMPTION

A particular cell phone has a 6 watt-hour (W-hr) battery and operates at 1.2 V. Suppose that, when it is in use, the cell phone operates at 300 MHz and the average amount of capacitance in the chip switching at any given time is 10 nF (10^{-8} Farads). When in use, it also broadcasts 3 W of power out of its antenna. When the phone is not in use, the dynamic power drops to almost zero because the signal processing is turned off. But the phone also draws 40 mA of quiescent current whether it is in use or not. Determine the battery life of the phone (a) if it is not being used, and (b) if it is being used continuously.

Solution: The static power is $P_{\text{static}} = (0.040 \text{ A})(1.2 \text{ V}) = 48 \text{ mW}$. (a) If the phone is not being used, this is the only power consumption, so the battery life is $(6 \text{ Whr}) / (0.048 \text{ W}) = 125 \text{ hours}$ (about 5 days). (b) If the phone is being used, the dynamic power is $P_{\text{dynamic}} = (0.5)(10^{-8} \text{ F})(1.2 \text{ V})^2(3 \times 10^8 \text{ Hz}) = 2.16 \text{ W}$. Together with the static and broadcast power, the total active power is $2.16 \text{ W} + 0.048 \text{ W} + 3 \text{ W} = 5.2 \text{ W}$, so the battery life is $6 \text{ W-hr} / 5.2 \text{ W} = 1.15 \text{ hours}$. This example somewhat oversimplifies the actual operation of a cell phone, but it illustrates the key ideas of power consumption.

1.9 SUMMARY AND A LOOK AHEAD

There are 10 kinds of people in this world: those who can count in binary and those who can't.

This chapter has introduced principles for understanding and designing complex systems. Although the real world is analog, digital designers discipline themselves to use a discrete subset of possible signals. In particular, binary variables have just two states: 0 and 1, also called FALSE and TRUE or LOW and HIGH. Logic gates compute a binary output from one or more binary inputs. Some of the common logic gates are:

- ▶ **NOT:** TRUE when input is FALSE
- ▶ **AND:** TRUE when all inputs are TRUE
- ▶ **OR:** TRUE when any inputs are TRUE
- ▶ **XOR:** TRUE when an odd number of inputs are TRUE

Logic gates are commonly built from CMOS transistors, which behave as electrically controlled switches. nMOS transistors turn ON when the gate is 1. pMOS transistors turn ON when the gate is 0.

In Chapters 2 through 5, we continue the study of digital logic. Chapter 2 addresses *combinational logic*, in which the outputs depend only on the current inputs. The logic gates introduced already are examples of combinational logic. You will learn to design circuits involving multiple gates to implement a relationship between inputs and outputs specified by a truth table or Boolean equation. Chapter 3 addresses *sequential logic*, in which the outputs depend on both current and past inputs. *Registers* are common sequential elements that remember their previous input. *Finite state machines*, built from registers and combinational logic, are a powerful way to build complicated systems in a systematic fashion. We also study timing of digital systems to analyze how fast a system can operate. Chapter 4 describes hardware description languages (HDLs). HDLs are related to conventional programming languages but are used to simulate and

build hardware rather than software. Most digital systems today are designed with HDLs. SystemVerilog and VHDL are the two prevalent languages, and they are covered side-by-side in this book. Chapter 5 studies other combinational and sequential building blocks such as adders, multipliers, and memories.

Chapter 6 shifts to computer architecture. It describes the ARM processor, an industry-standard microprocessor used in almost all smart phones and tablets and many other devices, from pinball machines to cars and servers. The ARM architecture is defined by its registers and assembly language instruction set. You will learn to write programs in assembly language for the ARM processor so that you can communicate with the processor in its native language.

Chapters 7 and 8 bridge the gap between digital logic and computer architecture. Chapter 7 investigates microarchitecture, the arrangement of digital building blocks, such as adders and registers, needed to construct a processor. In that chapter, you learn to build your own ARM processor. Indeed, you learn three microarchitectures illustrating different trade-offs of performance and cost. Processor performance has increased exponentially, requiring ever more sophisticated memory systems to feed the insatiable demand for data. Chapter 8 delves into memory system architecture. Chapter 9 (available as a web supplement, see Preface) describes how computers communicate with peripheral devices such as monitors, Bluetooth radios, and motors.

Exercises

Exercise 1.1 Explain in one paragraph at least three levels of abstraction that are used by

- (a) biologists studying the operation of cells.
- (b) chemists studying the composition of matter.

Exercise 1.2 Explain in one paragraph how the techniques of hierarchy, modularity, and regularity may be used by

- (a) automobile designers.
- (b) businesses to manage their operations.

Exercise 1.3 Ben Bitdiddle is building a house. Explain how he can use the principles of hierarchy, modularity, and regularity to save time and money during construction.

Exercise 1.4 An analog voltage is in the range of 0–5 V. If it can be measured with an accuracy of ± 50 mV, at most how many bits of information does it convey?

Exercise 1.5 A classroom has an old clock on the wall whose minute hand broke off.

- (a) If you can read the hour hand to the nearest 15 minutes, how many bits of information does the clock convey about the time?
- (b) If you know whether it is before or after noon, how many additional bits of information do you know about the time?

Exercise 1.6 The Babylonians developed the *sexagesimal* (base 60) number system about 4000 years ago. How many bits of information is conveyed with one sexagesimal digit? How do you write the number 4000_{10} in sexagesimal?

Exercise 1.7 How many different numbers can be represented with 16 bits?

Exercise 1.8 What is the largest unsigned 32-bit binary number?

Exercise 1.9 What is the largest 16-bit binary number that can be represented with

- (a) unsigned numbers?
- (b) two's complement numbers?
- (c) sign/magnitude numbers?

Exercise 1.10 What is the largest 32-bit binary number that can be represented with

- (a) unsigned numbers?
- (b) two's complement numbers?
- (c) sign/magnitude numbers?

Exercise 1.11 What is the smallest (most negative) 16-bit binary number that can be represented with

- (a) unsigned numbers?
- (b) two's complement numbers?
- (c) sign/magnitude numbers?

Exercise 1.12 What is the smallest (most negative) 32-bit binary number that can be represented with

- (a) unsigned numbers?
- (b) two's complement numbers?
- (c) sign/magnitude numbers?

Exercise 1.13 Convert the following unsigned binary numbers to decimal. Show your work.

- (a) 1010_2
- (b) 110110_2
- (c) 11110000_2
- (d) 000100010100111_2

Exercise 1.14 Convert the following unsigned binary numbers to decimal. Show your work.

- (a) 1110_2
- (b) 100100_2
- (c) 11010111_2
- (d) 011101010100100_2

Exercise 1.15 Repeat Exercise 1.13, but convert to hexadecimal.

Exercise 1.16 Repeat Exercise 1.14, but convert to hexadecimal.

Exercise 1.17 Convert the following hexadecimal numbers to decimal. Show your work.

- (a) $A5_{16}$
- (b) $3B_{16}$
- (c) $FFFF_{16}$
- (d) $D0000000_{16}$

Exercise 1.18 Convert the following hexadecimal numbers to decimal. Show your work.

- (a) $4E_{16}$
- (b) $7C_{16}$
- (c) $ED3A_{16}$
- (d) $403FB001_{16}$

Exercise 1.19 Repeat Exercise 1.17, but convert to unsigned binary.

Exercise 1.20 Repeat Exercise 1.18, but convert to unsigned binary.

Exercise 1.21 Convert the following two's complement binary numbers to decimal.

- (a) 1010_2
- (b) 110110_2
- (c) 01110000_2
- (d) 10011111_2

Exercise 1.22 Convert the following two's complement binary numbers to decimal.

- (a) 1110_2
- (b) 100011_2
- (c) 01001110_2
- (d) 10110101_2

Exercise 1.23 Repeat Exercise 1.21, assuming the binary numbers are in sign/magnitude form rather than two's complement representation.

Exercise 1.24 Repeat Exercise 1.22, assuming the binary numbers are in sign/magnitude form rather than two's complement representation.

Exercise 1.25 Convert the following decimal numbers to unsigned binary numbers.

- (a) 42_{10}
- (b) 63_{10}
- (c) 229_{10}
- (d) 845_{10}

Exercise 1.26 Convert the following decimal numbers to unsigned binary numbers.

- (a) 14_{10}
- (b) 52_{10}
- (c) 339_{10}
- (d) 711_{10}

Exercise 1.27 Repeat Exercise 1.25, but convert to hexadecimal.

Exercise 1.28 Repeat Exercise 1.26, but convert to hexadecimal.

Exercise 1.29 Convert the following decimal numbers to 8-bit two's complement numbers or indicate that the decimal number would overflow the range.

- (a) 42_{10}
- (b) -63_{10}
- (c) 124_{10}
- (d) -128_{10}
- (e) 133_{10}

Exercise 1.30 Convert the following decimal numbers to 8-bit two's complement numbers or indicate that the decimal number would overflow the range.

- (a) 24_{10}
- (b) -59_{10}
- (c) 128_{10}
- (d) -150_{10}
- (e) 127_{10}

Exercise 1.31 Repeat Exercise 1.29, but convert to 8-bit sign/magnitude numbers.

Exercise 1.32 Repeat Exercise 1.30, but convert to 8-bit sign/magnitude numbers.

Exercise 1.33 Convert the following 4-bit two's complement numbers to 8-bit two's complement numbers.

- (a) 0101_2
- (b) 1010_2

Exercise 1.34 Convert the following 4-bit two's complement numbers to 8-bit two's complement numbers.

- (a) 0111_2
- (b) 1001_2

Exercise 1.35 Repeat Exercise 1.33 if the numbers are unsigned rather than two's complement.

Exercise 1.36 Repeat Exercise 1.34 if the numbers are unsigned rather than two's complement.

Exercise 1.37 Base 8 is referred to as *octal*. Convert each of the numbers from Exercise 1.25 to octal.

Exercise 1.38 Base 8 is referred to as *octal*. Convert each of the numbers from Exercise 1.26 to octal.

Exercise 1.39 Convert each of the following octal numbers to binary, hexadecimal, and decimal.

- (a) 42_8
- (b) 63_8
- (c) 255_8
- (d) 3047_8

Exercise 1.40 Convert each of the following octal numbers to binary, hexadecimal, and decimal.

- (a) 23_8
- (b) 45_8
- (c) 371_8
- (d) 2560_8

Exercise 1.41 How many 5-bit two's complement numbers are greater than 0? How many are less than 0? How would your answers differ for sign/magnitude numbers?

Exercise 1.42 How many 7-bit two's complement numbers are greater than 0? How many are less than 0? How would your answers differ for sign/magnitude numbers?

Exercise 1.43 How many bytes are in a 32-bit word? How many nibbles are in the word?

Exercise 1.44 How many bytes are in a 64-bit word?

Exercise 1.45 A particular DSL modem operates at 768 kbits/sec. How many bytes can it receive in 1 minute?

Exercise 1.46 USB 3.0 can send data at 5 Gbits/sec. How many bytes can it send in 1 minute?

Exercise 1.47 Hard disk manufacturers use the term “megabyte” to mean 10^6 bytes and “gigabyte” to mean 10^9 bytes. How many real GBs of music can you store on a 50 GB hard disk?

Exercise 1.48 Estimate the value of 2^{31} without using a calculator.

Exercise 1.49 A memory on the Pentium II microprocessor is organized as a rectangular array of bits with 2^8 rows and 2^9 columns. Estimate how many bits it has without using a calculator.

Exercise 1.50 Draw a number line analogous to Figure 1.11 for 3-bit unsigned, two's complement, and sign/magnitude numbers.

Exercise 1.51 Draw a number line analogous to Figure 1.11 for 2-bit unsigned, two's complement, and sign/magnitude numbers.

Exercise 1.52 Perform the following additions of unsigned binary numbers. Indicate whether or not the sum overflows a 4-bit result.

(a) $1001_2 + 0100_2$

(b) $1101_2 + 1011_2$

Exercise 1.53 Perform the following additions of unsigned binary numbers. Indicate whether or not the sum overflows an 8-bit result.

- (a) $10011001_2 + 01000100_2$
- (b) $11010010_2 + 10110110_2$

Exercise 1.54 Repeat Exercise 1.52, assuming that the binary numbers are in two's complement form.

Exercise 1.55 Repeat Exercise 1.53, assuming that the binary numbers are in two's complement form.

Exercise 1.56 Convert the following decimal numbers to 6-bit two's complement binary numbers and add them. Indicate whether or not the sum overflows a 6-bit result.

- (a) $16_{10} + 9_{10}$
- (b) $27_{10} + 31_{10}$
- (c) $-4_{10} + 19_{10}$
- (d) $3_{10} + -32_{10}$
- (e) $-16_{10} + -9_{10}$
- (f) $-27_{10} + -31_{10}$

Exercise 1.57 Repeat Exercise 1.56 for the following numbers.

- (a) $7_{10} + 13_{10}$
- (b) $17_{10} + 25_{10}$
- (c) $-26_{10} + 8_{10}$
- (d) $31_{10} + -14_{10}$
- (e) $-19_{10} + -22_{10}$
- (f) $-2_{10} + -29_{10}$

Exercise 1.58 Perform the following additions of unsigned hexadecimal numbers. Indicate whether or not the sum overflows an 8-bit (two hex digit) result.

- (a) $7_{16} + 9_{16}$
- (b) $13_{16} + 28_{16}$
- (c) $AB_{16} + 3E_{16}$
- (d) $8F_{16} + AD_{16}$

Exercise 1.59 Perform the following additions of unsigned hexadecimal numbers. Indicate whether or not the sum overflows an 8-bit (two hex digit) result.

- (a) $22_{16} + 8_{16}$
- (b) $73_{16} + 2C_{16}$
- (c) $7F_{16} + 7F_{16}$
- (d) $C2_{16} + A4_{16}$

Exercise 1.60 Convert the following decimal numbers to 5-bit two's complement binary numbers and subtract them. Indicate whether or not the difference overflows a 5-bit result.

- (a) $9_{10} - 7_{10}$
- (b) $12_{10} - 15_{10}$
- (c) $-6_{10} - 11_{10}$
- (d) $4_{10} - -8_{10}$

Exercise 1.61 Convert the following decimal numbers to 6-bit two's complement binary numbers and subtract them. Indicate whether or not the difference overflows a 6-bit result.

- (a) $18_{10} - 12_{10}$
- (b) $30_{10} - 9_{10}$
- (c) $-28_{10} - 3_{10}$
- (d) $-16_{10} - 21_{10}$

Exercise 1.62 In a *biased* N -bit binary number system with bias B , positive and negative numbers are represented as their value plus the bias B . For example, for 5-bit numbers with a bias of 15, the number 0 is represented as 01111, 1 as 10000, and so forth. Biased number systems are sometimes used in floating point mathematics, which will be discussed in Chapter 5. Consider a biased 8-bit binary number system with a bias of 127_{10} .

- (a) What decimal value does the binary number 10000010_2 represent?
- (b) What binary number represents the value 0?
- (c) What is the representation and value of the most negative number?
- (d) What is the representation and value of the most positive number?

Exercise 1.63 Draw a number line analogous to Figure 1.11 for 3-bit biased numbers with a bias of 3 (see Exercise 1.62 for a definition of biased numbers).

Exercise 1.64 In a *binary coded decimal* (BCD) system, 4 bits are used to represent a decimal digit from 0 to 9. For example, 37_{10} is written as 00110111_{BCD} .

- (a) Write 289_{10} in BCD
- (b) Convert $100101010001_{\text{BCD}}$ to decimal
- (c) Convert 01101001_{BCD} to binary
- (d) Explain why BCD might be a useful way to represent numbers

Exercise 1.65 Answer the following questions related to BCD systems (see Exercise 1.64 for the definition of BCD).

- (a) Write 371_{10} in BCD
- (b) Convert $000110000111_{\text{BCD}}$ to decimal
- (c) Convert 10010101_{BCD} to binary
- (d) Explain the disadvantages of BCD when compared to binary representations of numbers

Exercise 1.66 A flying saucer crashes in a Nebraska cornfield. The FBI investigates the wreckage and finds an engineering manual containing an equation in the Martian number system: $325 + 42 = 411$. If this equation is correct, how many fingers would you expect Martians to have?

Exercise 1.67 Ben Bitdiddle and Alyssa P. Hacker are having an argument. Ben says, “All integers greater than zero and exactly divisible by six have exactly two 1’s in their binary representation.” Alyssa disagrees. She says, “No, but all such numbers have an even number of 1’s in their representation.” Do you agree with Ben or Alyssa or both or neither? Explain.

Exercise 1.68 Ben Bitdiddle and Alyssa P. Hacker are having another argument. Ben says, “I can get the two’s complement of a number by subtracting 1, then inverting all the bits of the result.” Alyssa says, “No, I can do it by examining each bit of the number, starting with the least significant bit. When the first 1 is found, invert each subsequent bit.” Do you agree with Ben or Alyssa or both or neither? Explain.

Exercise 1.69 Write a program in your favorite language (e.g., C, Java, Perl) to convert numbers from binary to decimal. The user should type in an unsigned binary number. The program should print the decimal equivalent.

Exercise 1.70 Repeat Exercise 1.69 but convert from an arbitrary base b_1 to another base b_2 , as specified by the user. Support bases up to 16, using the letters of the alphabet for digits greater than 9. The user should enter b_1 , b_2 , and then the number to convert in base b_1 . The program should print the equivalent number in base b_2 .

Exercise 1.71 Draw the symbol, Boolean equation, and truth table for

- (a) a three-input OR gate
- (b) a three-input exclusive OR (XOR) gate
- (c) a four-input XNOR gate

Exercise 1.72 Draw the symbol, Boolean equation, and truth table for

- (a) a four-input OR gate
- (b) a three-input XNOR gate
- (c) a five-input NAND gate

Exercise 1.73 A *majority gate* produces a TRUE output if and only if more than half of its inputs are TRUE. Complete a truth table for the three-input majority gate shown in Figure 1.41.

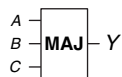


Figure 1.41 Three-input majority gate

Exercise 1.74 A three-input *AND-OR (AO) gate* shown in Figure 1.42 produces a TRUE output if both A and B are TRUE, or if C is TRUE. Complete a truth table for the gate.



Figure 1.42 Three-input AND-OR gate

Exercise 1.75 A three-input *OR-AND-INVERT (OAI) gate* shown in Figure 1.43 produces a FALSE output if C is TRUE and A or B is TRUE. Otherwise it produces a TRUE output. Complete a truth table for the gate.

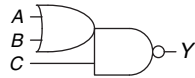


Figure 1.43 Three-input OR-AND-INVERT gate

Exercise 1.76 There are 16 different truth tables for Boolean functions of two variables. List each truth table. Give each one a short descriptive name (such as OR, NAND, and so on).

Exercise 1.77 How many different truth tables exist for Boolean functions of N variables?

Exercise 1.78 Is it possible to assign logic levels so that a device with the transfer characteristics shown in Figure 1.44 would serve as an inverter? If so, what are the input and output low and high levels (V_{IL} , V_{OL} , V_{IH} , and V_{OH}) and noise margins (NM_L and NM_H)? If not, explain why not.

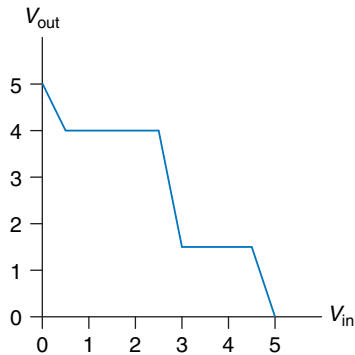


Figure 1.44 DC transfer characteristics

Exercise 1.79 Repeat Exercise 1.78 for the transfer characteristics shown in Figure 1.45.

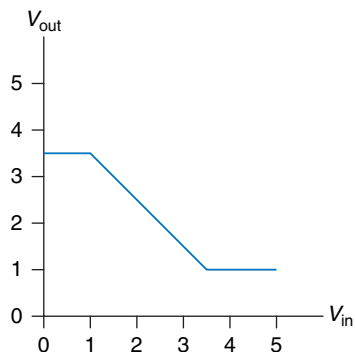


Figure 1.45 DC transfer characteristics

Exercise 1.80 Is it possible to assign logic levels so that a device with the transfer characteristics shown in Figure 1.46 would serve as a buffer? If so, what are the input and output low and high levels (V_{IL} , V_{OL} , V_{IH} , and V_{OH}) and noise margins (NM_L and NM_H)? If not, explain why not.

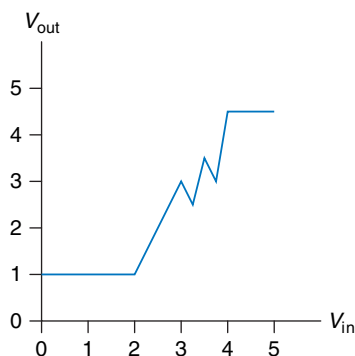


Figure 1.46 DC transfer characteristics

Exercise 1.81 Ben Bitdiddle has invented a circuit with the transfer characteristics shown in Figure 1.47 that he would like to use as a buffer. Will it work? Why or why not? He would like to advertise that it is compatible with LVC MOS and LV TTL logic. Can Ben's buffer correctly receive inputs from those logic families? Can its output properly drive those logic families? Explain.

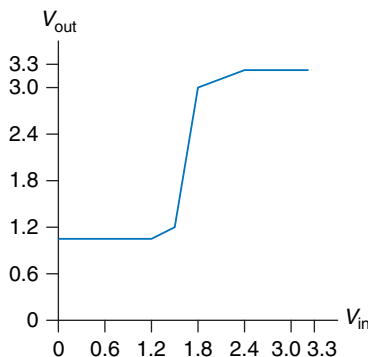


Figure 1.47 Ben's buffer DC transfer characteristics

Exercise 1.82 While walking down a dark alley, Ben Bitdiddle encounters a two-input gate with the transfer function shown in Figure 1.48. The inputs are A and B and the output is Y .

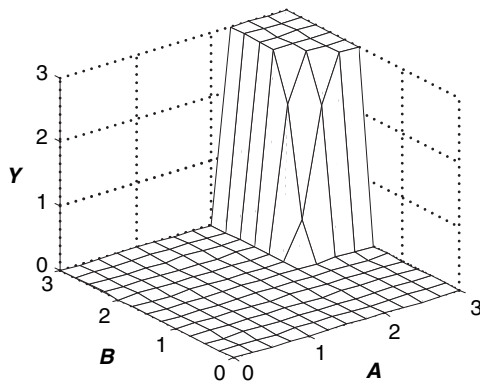


Figure 1.48 Two-input DC transfer characteristics

- (a) What kind of logic gate did he find?
- (b) What are the approximate high and low logic levels?

Exercise 1.83 Repeat Exercise 1.82 for Figure 1.49.

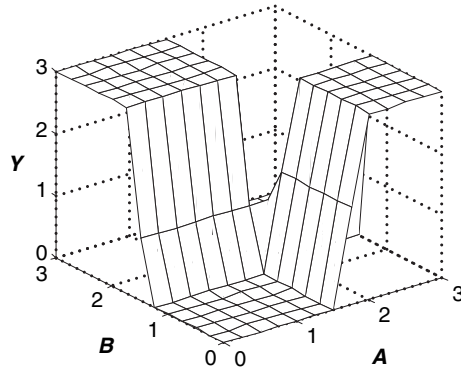


Figure 1.49 Two-input DC transfer characteristics

Exercise 1.84 Sketch a transistor-level circuit for the following CMOS gates. Use a minimum number of transistors.

- (a) four-input NAND gate
- (b) three-input OR-AND-INVERT gate (see Exercise 1.75)
- (c) three-input AND-OR gate (see Exercise 1.74)

Exercise 1.85 Sketch a transistor-level circuit for the following CMOS gates. Use a minimum number of transistors.

- (a) three-input NOR gate
- (b) three-input AND gate
- (c) two-input OR gate

Exercise 1.86 A *minority gate* produces a TRUE output if and only if fewer than half of its inputs are TRUE. Otherwise it produces a FALSE output. Sketch a transistor-level circuit for a three-input CMOS minority gate. Use a minimum number of transistors.

Exercise 1.87 Write a truth table for the function performed by the gate in Figure 1.50. The truth table should have two inputs, A and B . What is the name of this function?

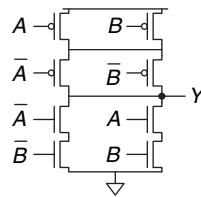


Figure 1.50 Mystery schematic

Exercise 1.88 Write a truth table for the function performed by the gate in Figure 1.51. The truth table should have three inputs, A , B , and C .

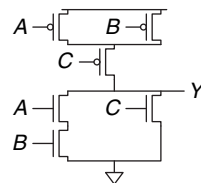


Figure 1.51 Mystery schematic

Exercise 1.89 Implement the following three-input gates using only pseudo-nMOS logic gates. Your gates receive three inputs, A , B , and C . Use a minimum number of transistors.

- (a) three-input NOR gate
- (b) three-input NAND gate
- (c) three-input AND gate

Exercise 1.90 *Resistor-Transistor Logic (RTL)* uses nMOS transistors to pull the gate output LOW and a weak resistor to pull the output HIGH when none of the paths to ground are active. A NOT gate built using RTL is shown in [Figure 1.52](#). Sketch a three-input RTL NOR gate. Use a minimum number of transistors.

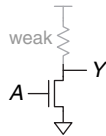


Figure 1.52 RTL NOT gate

Interview Questions

These questions have been asked at interviews for digital design jobs.

Question 1.1 Sketch a transistor-level circuit for a CMOS four-input NOR gate.

Question 1.2 The king receives 64 gold coins in taxes but has reason to believe that one is counterfeit. He summons you to identify the fake coin. You have a balance that can hold coins on each side. How many times do you need to use the balance to find the lighter, fake coin?

Question 1.3 The professor, the teaching assistant, the digital design student, and the freshman track star need to cross a rickety bridge on a dark night. The bridge is so shaky that only two people can cross at a time. They have only one flashlight among them and the span is too long to throw the flashlight, so somebody must carry it back to the other people. The freshman track star can cross the bridge in 1 minute. The digital design student can cross the bridge in 2 minutes. The teaching assistant can cross the bridge in 5 minutes. The professor always gets distracted and takes 10 minutes to cross the bridge. What is the fastest time to get everyone across the bridge?