

$$\text{Impuls bei Kronecker} \\ J(K) = \begin{cases} 1 & K=0 \\ 0 & K \neq 0 \end{cases}$$

$$\text{graduierter Anteil} \\ 1(K) = \begin{cases} 1 & K \geq 0 \\ 0 & K < 0 \end{cases}$$

$$z\{1(K)\} = \frac{2}{z-1}$$

$$n(t) = \begin{cases} + & t \geq 0 \\ 0 & t < 0 \end{cases}$$

$$z\{n(t)\} = T \frac{\frac{2}{z}}{(z-1)^2}$$

$$x(K) = \begin{cases} a^K & K \geq 0 \\ 0 & K < 0 \end{cases}$$

$$z\{a^K\} = \frac{2}{z-a}$$

$$\frac{2}{z-a}$$

$$\frac{2}{z-a}$$

$$z\{c_1 X_1(K) + c_2 X_2(K)\} = (c_1 X_1(z)) + (c_2 X_2(z))$$

$$z\{\sin(Kt)\} = z \frac{\sin(wt)}{z^2 - 2\cos(wt)z + 1}$$

$$z\{(\cos(Kt))\} = \frac{z(z - \cos(wt))}{z^2 - 2z\cos(wt) + 1}$$

$$X(K) = a^K x(K)$$

$$z\{a^K x(K)\} = x\left(\frac{z}{a}\right)$$

$$z\left\{ e^{at} \sin(wt) \right\} = \frac{\frac{2}{z-a} \sin(wt)}{\frac{z^2}{a^2} - \frac{2z}{a} \cos(wt) + 1}$$

$$\begin{array}{c} x(K) \\ \downarrow \\ \frac{2}{z-a} \\ \downarrow \\ \frac{1}{z-a} \\ \downarrow \\ -1+1+2+3+4+5 \end{array} \rightarrow K$$

$$x(K) = -\delta(K) + 2\delta(K-1) + \delta(K-2) + \delta(K-3)$$

$$X(z) = -1 + 2z^{-1} + z^{-2} + 4z^{-3}$$

$$z\{Kx(K)\} = -z \frac{dx(z)}{dz}$$

$$z\{Kt\delta(K)\} = -z \frac{d}{dt} \left[ KT \frac{z}{z-1} \right] = +2KT$$

$$2 \left\{ \frac{(K+1-1)!}{K!(n-x)!} a^K \right\} = \left\{ \binom{K+1-1}{K} a^K \right\} = \frac{z^n}{(z-a)^n} \quad z\left\{ \frac{10}{z^2(z+1)} \right\} = 10 \frac{z^{-1}}{z^2} \left[ \frac{1}{z^2} + \frac{1}{z} + \frac{1}{z+1} \right] = 2 \left\{ 10Kz - 10\delta(K) + 10e^{Kz} \right\}$$

$$z\{Ka^{K-1}\} = z\left(\frac{ka^K}{a}\right) = \frac{2}{(z-a)^2}$$

$$f(K) = \left| \cos\left(\frac{2\pi}{3}(1+K)\right) \right|$$

$$\begin{matrix} T=9s & K & f(K) \\ 0 & 1 & \\ 1 & 1/2 & \\ 2 & 1/2 & \\ 3 & 1 & \end{matrix}$$

$$z\{ka^K\}$$

$$\frac{2}{(z-a)^2}$$

$$\frac{10z}{(z-1)^2} - 10 \frac{2}{z-1} + 10 \frac{2}{z-2}$$

$$\begin{array}{c} x(z) \\ \frac{z^2}{z^2-3z+1} = \frac{z^2}{z^2-2z-1} \\ // \quad z+1 \\ \quad 2-z-1 \\ // \quad 2-z \\ \quad 2-z^2 \\ // \quad 3z^2+2z^2 \end{array}$$

$$\frac{z^2-1}{1+z-\frac{1}{z}+z^2-\frac{1}{z^2}}$$

$$\frac{E(z)=\delta(K)}{(1+\frac{z}{z})(1+\frac{1}{z^2})} \quad G(z) = \frac{z}{(z+1)(z+2)} = \frac{z^{-1}}{1+z^{-1}+2z^{-2}} = \frac{z^{-1}}{E(z)}$$

$$M(K) + 3M(K-1) + 2M(K-2) = \delta(K-1)$$

$$K$$

$$0 \quad M(0) = 0$$

$$1 \quad M(1) = 1 - 3 \cdot 0 + 2 \cdot 0 = 1$$

$$2 \quad M(2) = 0 - 3 \cdot 1 - 2 \cdot 0 = -3$$

$$X(z) = \frac{2}{(z+1)(z+2)} = \frac{A}{z+1} + \frac{B}{z+2} = -\frac{\bar{z}^{-1}z}{z+1} + \frac{\bar{z}^{-1}z}{z+2} = -(-1)^{K-1} + (-2)^{K-1} = -(-1)^{K-1} \underbrace{\delta_{1(K-1)}}_{\text{graduierter Anteil}} + 2(-2)^{K-1} \underbrace{\delta_{2(K-1)}}_{\text{modulante}}$$

$$\begin{matrix} A = -1 \\ B = 2 \end{matrix}$$

$$\delta_{ab} = \frac{e^{izx} - e^{-izx}}{2i}$$

$$X(z) = \frac{z^{-1}}{(1+\frac{z}{z})(1+\frac{1}{z^2})} = \frac{1}{z+1} + \frac{-1}{z+2} = (-1)^K - (-2)^K \quad M(x) = \frac{e^{izx} - e^{-izx}}{2i} \quad X(z) = \frac{10(z+2)}{(z^2+1)z} = \frac{A}{z} + \frac{B}{z-1} + \frac{C}{z+1} = \frac{-20 + (-10 \cdot 5)z^{-1}}{z} + \frac{(10+5)z^{-2}}{z+1}$$

$$\begin{matrix} z = e^{it/2} \\ -j = e^{-it/2} \end{matrix}$$

$$= (-10 \cdot 5)z^K + (-10 \cdot 5)z^K + 20\delta(K-1) + [-20 \cdot 0 \cdot \delta(K-1)] \frac{1}{z} + 10 \operatorname{Im}(\delta(K-1)) \frac{1}{z+1}$$

$$X(z) = \frac{2}{(z+1)(z+2)} \quad X(K) = (-1)^K - (-2)^K$$

$$\frac{z^2}{z+1} \Big|_{z=-1} = (-1)^K$$

$$\frac{2^K}{z+1} \Big|_{z=-2} = -(-2)^K$$

$$X(z) = \frac{10(z+2)}{z(z+1)(z-1)} z^{K-1} = (R_1 + R_2 + R_3) \delta_{1(K-2)}$$

$$R_1 = \sum \operatorname{Res} \left( \frac{10(z+2)}{2(z+1)(z-1)} \right)_{z=1}$$

$$R_2 = \frac{10(z+2)}{(z-1)} z^{K-2}$$

$$R_3 = \frac{10(z+2)z^{K-2}}{z+1} \Big|_{z=1} = 15$$

$$K=1$$

$$-S(-1)^{K-2}$$

$$X(z) = \frac{S_0 z^3}{(z+1)(z^2+4)}$$

$$R_1 = S_0 \frac{z^3 z^{k-1}}{z^2+4} \Big|_{z=-1} = S_0 \frac{(-1)^{k+2}}{3}$$

$$R_2 + R_3 = S_0 \frac{z^3 z^{k+2}}{(z+1)(z^2+4)} + S_0 \frac{(-1)^{k+2}}{(z+1)(-4)}$$

$$S_0 \frac{z^{k+2} \sin((k+2)\pi/2)}{2} - 10 z^{k+2} \cos((k+2)\pi/2)$$

PROVA DI ESAME DEL 25/02/2021

#### QUESITO A

Si calcolino la z-transformata e l'anttrasformata (in forma esplicita) riportate di seguito, fissando un passo di campionamento  $T = 0.5$  secondi:

- $X(z) = Z\left(\frac{4}{z^2 + 4}\right)$
- $x(k) = Z^{-1}\left(\frac{z}{(z-1)^2(z+3)}\right)$

$$\frac{1}{z^2} + \frac{-14}{z} + \frac{14}{z^2+4} = \frac{3T}{(z-1)^2} - \frac{3}{(z-1)4} + \frac{z}{4(z-4)}$$

$$\frac{A}{(z-1)^2} + \frac{B}{z-1} + \frac{C}{z+1} = \frac{1}{2T} u(k) - \frac{1}{4} \delta(k) + \frac{(-1)^k}{4}$$

TURNO 1 - PROVA DI ESAME DEL 15/06/2021

#### QUESITO A

Sia assegnato il sistema tempo-discreto descritto dalla seguente equazione alle differenze:  
 $y(k) = -4y(k-2) + u(k-3) - u(k-4)$

con  $y(0) = 0$  per  $k \leq 0$ .

Supponendo di applicare in ingresso al sistema il seguente segnale  $u(k)$ :

$$\begin{aligned} u(k) &= 0 & \text{per } k < 0 \text{ e } k > 2 \\ u(0) &= 0 \\ u(1) &= 1 \\ u(2) &= 1.5 \end{aligned}$$

- Si determini la Z-transformata di  $y(k)$ ;
- Si determini in forma esplicita il segnale di uscita  $y(k)$  mediante anttrasformata-Z;
- Se ha senso, si determini il valore finale dell'uscita  $y(k)$  del sistema.

PROVA DI ESAME DEL 19/09/2023

#### QUESITO A

Un sistema digitale è descritto dalla seguente funzione di trasferimento:

$$G(z) = \frac{4}{(z+0.5)(z^2+1)}$$

Si calcoli, in forma esplicita, l'espressione temporale della risposta  $y(k)$  del sistema al seguente ingresso (avendo indicato con  $1(k)$  il gradino unitario discreto e con  $r(k)$  la rampa unitaria discreta):

$$u(k) = r(k-1) - 2 \cdot 1(k-2)$$

riproduzione (zero order hold)

$$H_S = \frac{1-e^{-Ts}}{S}$$

$$X_H(t) = X(KT) + \left( \frac{X(KT) - X((K-1)T)}{T} \right) (t-KT)$$

PROVA DI ESAME DEL 18/07/2023

#### QUESITO A

- Si determini l'espressione del segnale  $f(k) = Z^{-1}\left(\frac{z^2}{(z+1)(z^2+4)}\right)$  utilizzando esclusivamente il metodo dell'integrale di inversione; sfruttando l'espressione così ottenuta, si calcolino i primi due campioni del segnale.
- Si verifichi che i campioni calcolati al punto precedente corrispondono a quelli ottenuti con il metodo della lunga divisione.

$$\frac{z^{k+2}}{(z+1)(z^2+4)} = R_1 + R_2 + R_3$$

$$R_2 + R_3 = \frac{z^{k+2}}{(z+1)(z^2+3)} \Big|_{z=2} + \frac{z^{k+2}}{(z+1)(z-2)} \Big|_{z=2}$$

$$R_1 = \frac{z^{k+2}}{(z+1)(z^2+4)} \Big|_{z=-1} = \frac{(-1)^{k+2}}{5}$$

$$\frac{z^3 + z^2 + 4z + 4}{1-z^2} = \frac{10}{z^3+2z^2} - \frac{2}{z^2+4} \Rightarrow$$

$$\frac{1}{z^3+2z^2} \sin((k+2)\pi/2) - \frac{2}{z^2+4} \cos((k+2)\pi/2)$$

$$f(k) \frac{z^3}{z^3} = \frac{1}{1+z^{-1}+4z^{-2}+4z^{-3}} = \frac{N}{E} \Rightarrow$$

$$\begin{aligned} u(k) &= f_k \\ u(k) &= e(k) - u(k-1) \\ u(k) &= 4u(k-2) \\ u(k) &= 4u(k-3) \\ u(0) &= 1 \\ u(1) &= 0-1 = -1 \\ u(2) &= 0+1 = 1 \\ u(3) &= 0-1 = -1 \end{aligned}$$

$$\begin{aligned} G(z) &= \frac{y(z)}{u(z)} = \frac{z^{-3}(1-z^{-1})}{1+4z^{-2}} \\ M(z) &= z^3 + 15z^{-2} \end{aligned}$$

$$y = G \cdot u \Rightarrow 9z \cdot 2^{k-5} \cos((k-5)\pi/2) - 27z^2 2^{k-5} \sin((k-5)\pi/2)$$

$$N = \frac{-(2-2)}{2(2-1)^2}$$

$$Y = \frac{z^{-4}}{z} \{G \cdot u\} = \left( \frac{32}{9} (-0.5)^{k-2} + 4 \sin((k-2)\pi/2) + \frac{4}{3} u(k-2) - \frac{3}{2} u(k-3) \right) \cdot 1(k-2)$$

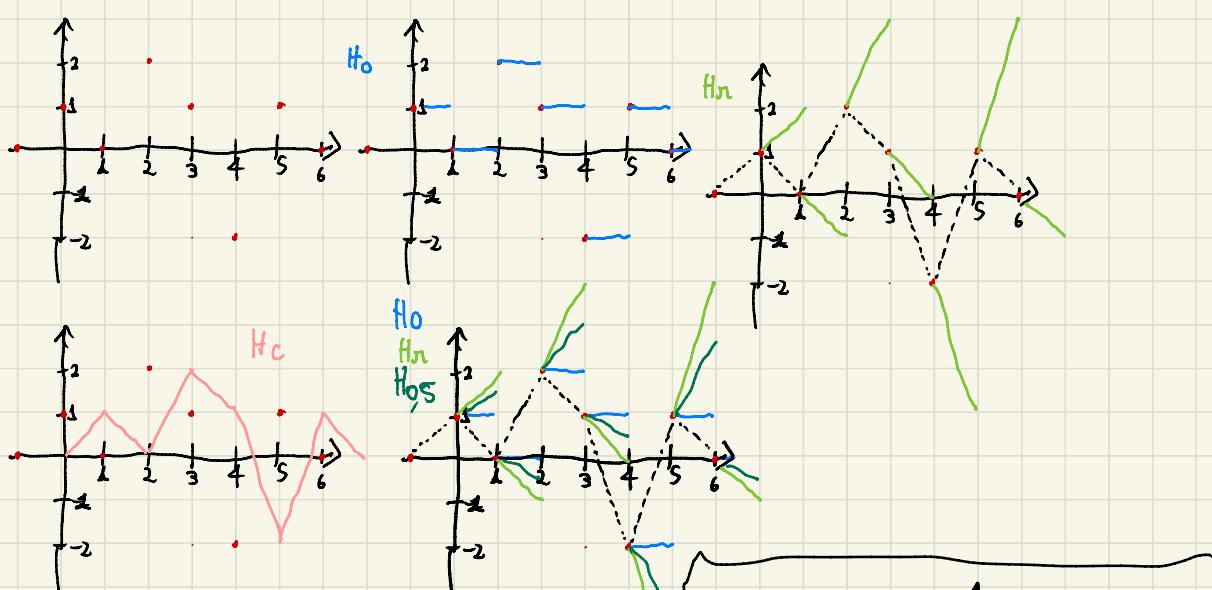
$$\text{ordine } 1^{\circ} \quad H_1(s) = \frac{1+Ts}{T} \left( \frac{1-e^{-Ts}}{s} \right)^2$$

ordine 0° XLSL

$$H_2(s) = \frac{1+Ts}{T} \left( \frac{1-e^{-Ts}}{s} \right) + (1-K) e^{-Ts} \left( \frac{1-e^{-Ts}}{s} \right)^2$$

$$\begin{cases} H_0, H_1, H_2 \rightarrow \text{discontinuità} \\ x_{\pi} = X((K-1)T) \\ (1-K) \frac{X(KT) - X((K-1)T)}{T} \end{cases}$$

$$\begin{cases} K \leq t \leq (K+1)T \\ \text{interpolazione continua} \\ H_C = \frac{1}{T} \left( \frac{1-e^{-Ts}}{s} \right)^2 \end{cases}$$



### QUESITO B

Si rappresenti graficamente (con precisione) l'andamento nei primi tre intervalli di campionamento del segnale tempo-continuo  $y(t)$



$$\textcircled{1} \quad \frac{1}{1+2z^{-1}} = \frac{z+2}{z^2-2z^2}$$

$$\textcircled{2} \quad u(k) = (-2)^{\frac{k-1}{2}} u(k-1)$$

$$k \quad u(k)$$

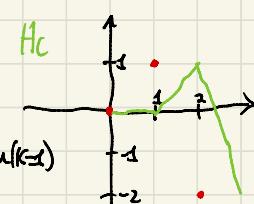
0	0
1	-1
2	-2
3	0
4	1
5	2

$$\textcircled{3} \quad g(z) = \frac{z^{-1}}{1+2z^{-1}}$$

$$u(k) = g(k-1) - 2u(k-2)$$

$$k \quad u(k)$$

0	0
1	-1
2	-2
3	0
4	1
5	2



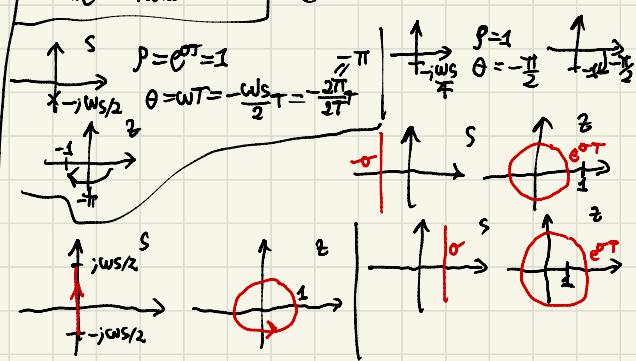
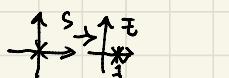
$$X(s) = \frac{2}{s} + \frac{2}{Ts^2} - \frac{2}{s} e^{-Ts} - \frac{2}{Ts^2} e^{-Ts}$$

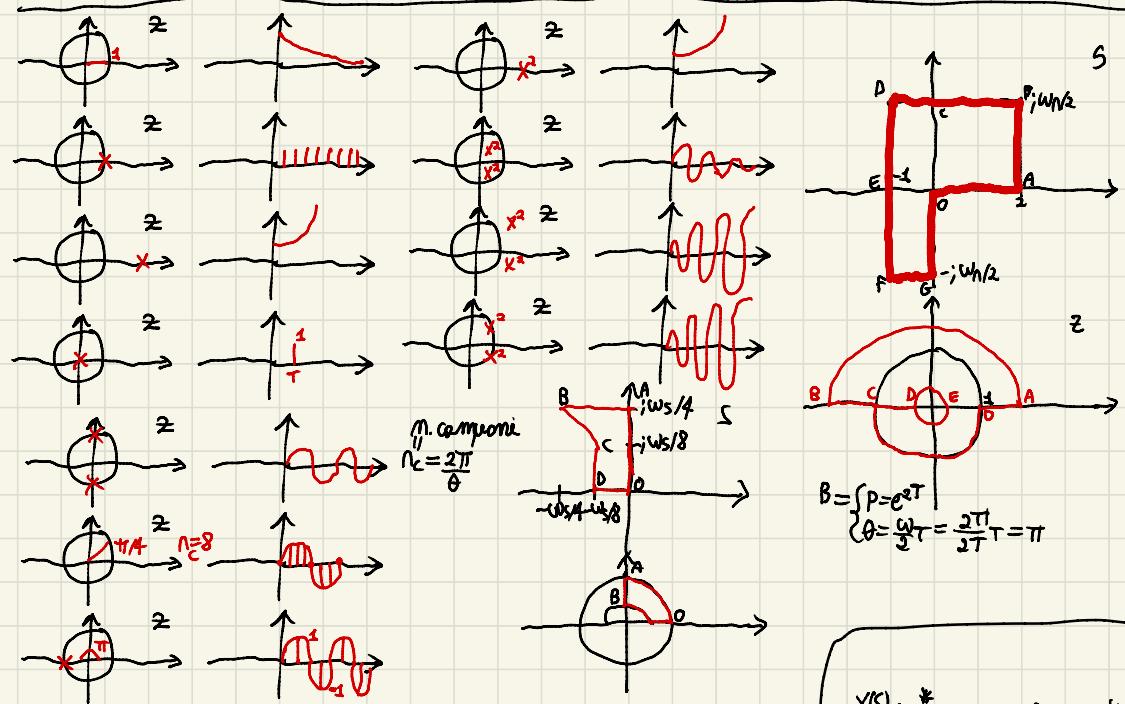
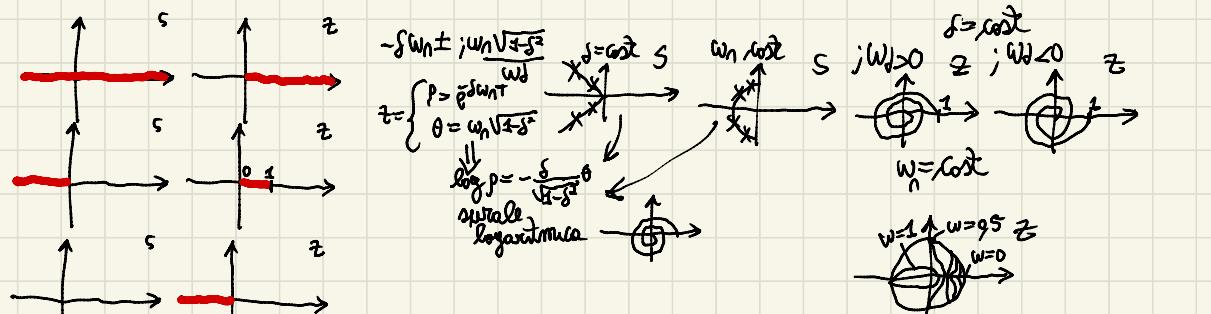
$$= \frac{(1+Ts)}{Ts^2} (1-e^{-Ts})$$

$$X_{CS} = 2 \left\{ 1(k) - \frac{2}{z-1} \Big|_{z=e^{Ts}} \right\} \frac{e^{-Ts}}{e^{-Ts}} = \frac{1}{1-e^{-Ts}}$$

$$\frac{X(s)}{X_{CS}} \Rightarrow 1^{\text{o}} \text{ ordine MR}$$

$$\begin{cases} H_0 \rightarrow \text{ritardo } T/2 \\ H_1 \rightarrow \text{ritardo } T \\ H_C \rightarrow \text{ritardo } T \end{cases} \quad \begin{cases} p = e^{j\omega T} \\ \theta = \omega T \end{cases}$$





$$\begin{aligned} C &= e^{-j\pi/4} \\ \theta &= \frac{3\pi}{8}T = \frac{\pi}{4} \\ C &= e^{-j\pi/4} \cdot e^{j\pi/4} \end{aligned}$$

Block diagram of a system with two parallel branches:

$$X(S) \xrightarrow{G_1(S)} X_1(S) \quad X_1(S) \xrightarrow{G_2(S)} Y(S)$$

$$X(S) \xrightarrow{G_2(S)} X_2(S) \quad X_2(S) \xrightarrow{G_1(S)} Y(S)$$

Properties of the system:

$$Y(S) = G(S) \cdot Y(S)$$

$$Y^*(S) = [G(S) \cdot Y(S)]^* \neq G^*(S) \cdot X^*(S)$$

$$\begin{aligned} Y(S) &= X_1^*(S) G_2(S) \\ X_1(S) &= G_1(S) X^*(S) \\ Y^*(S) &= G_2^*(S) G_1^*(S) X^*(S) \\ X(2) &= G_1(2) G_2(2) Y(2) \end{aligned}$$

$$\frac{X(S)}{Y(S)} = \frac{G_1(S) - G_2(S)}{Y^*(S)}$$

$$Y(S) = G_2(S) G_3(S) X^*(S)$$

$$Y^*(S) = [G_3(S) G_2(S)]^* X^*(S) = G_3 G_2(z) X(z)$$

$$G_1 = \frac{1}{z-1} \quad G_2 = \frac{1}{z-2}$$

$$G_3 = \frac{2}{z-3} \quad G_2 = \frac{2}{z-2}$$

$$G_3 G_2 = \frac{2}{z-3} \left\{ \frac{1}{z-1} + \frac{1}{z-2} \right\} = \frac{2}{z-2} + \frac{-2}{z-3}$$

$$\frac{H(S)}{Y(S)} = \frac{G_1(S)}{Y^*(S)}$$

$$H_0 G_p = \frac{1}{z} \left\{ \frac{G_1(S)}{S} \right\} - z^{-1} \left\{ \frac{G_1(S)}{S} \right\}$$

$$\frac{Y(S)}{Y^*(S)} = G_2(S) G_3(S) \frac{Y_1(S)}{X^*(S)}$$

$$\begin{cases} Y(S) = G_2(S) G_3(S) \\ X_1(S) = G_3(S) X^*(S) \end{cases}$$

$$Y^*(S) = G_2^*(S) G_3^*(S) X^*(S)$$

$$Y(z) = G_2(z) G_3(z) X(z)$$

$$X(z) = G_3(z) G_2(z) Y(z)$$

$$\frac{R(S)}{E(S)} = \frac{E^*(S)}{G(S)} Y(S) \quad Y^*(S)$$

$$\begin{cases} E(S) = R(S) - Y(S) \\ Y(S) = G(S) E^*(S) \end{cases} \quad \begin{cases} E(z) = R(z) - G(z) E(z) \\ Y(z) = G(z) E^*(z) \end{cases}$$

$$\frac{R(z)}{E(z)} = G(z) \rightarrow Y(z)$$

$$\frac{R(S)}{E(S)} = \frac{E^*(S)}{G_1(S)} Y(S) \quad Y^*(S)$$

$$\begin{cases} E(S) = R(S) - G_2(S) Y^*(S) \\ Y(S) = G_1(S) E^*(S) \end{cases} \Rightarrow \begin{cases} E^*(S) = R^*(S) - G_2^*(S) Y^*(S) \\ Y^*(S) = G_1^*(S) E^*(S) \end{cases}$$

$$E(z) = R(z) - G_2(z) G_1(z) E(z) \quad Y(z) = G_1(z) E(z)$$

$$\frac{R(z)}{E(z)} = G_2(z) \rightarrow G_1(z)$$

$$\frac{R(S)}{E(S)} = \frac{E_1(S)}{G_2(S)} Y(S) \quad Y^*(S)$$

$$\begin{cases} E_1(S) = G_3(S) [R(S) - G_2(S) G_3(S) E_1^*(S)] \\ Y(S) = G_2(S) E_1^*(S) \end{cases}$$

$$\begin{cases} E_1(z) = G_3(z) [R(z) - G_2(z) G_3(z) E_1(z)] \\ Y(z) = G_2(z) E_1(z) \end{cases}$$

$$\frac{G_1 R(z)}{E_1(z)} = G_2(z) \rightarrow G_3(z)$$

$$\frac{X(S)}{Y(S)} = \frac{G_1(S)}{Y^*(S)}$$

$$Y(S) = G_2(S) X^*(S)$$

$$X_1(S) = G_3(S) X(S)$$

$$Y(z) = G_2(z) X(z)$$

$$\begin{cases} E(S) = R(S) - Y^*(S) \\ Y(S) = G(S) E^*(S) \end{cases}$$

$$E(z) = R(z) - G(z) E(z)$$

$$\frac{R(z)}{E(z)} = G(z) \rightarrow Y(z)$$

$$\frac{R(S)}{E(S)} = \frac{E^*(S)}{G_1(S)} Y(S) \quad Y^*(S)$$

$$\begin{cases} E(S) = R(S) - G_1(S) G_2(S) E^*(S) \\ Y(S) = G_2(S) E^*(S) \end{cases}$$

$$\begin{cases} E(z) = R(z) - G_1(z) G_2(z) E(z) \\ Y(z) = G_2(z) E(z) \end{cases}$$

$$\frac{R(z)}{E(z)} = G_1(z) \rightarrow G_2(z)$$

$$\frac{R(S)}{E(S)} = \frac{E^*(S)}{G_1(S)} Y(S) \quad Y^*(S)$$

$$\begin{cases} E(S) = R(S) - G_2(S) G_3(S) E^*(S) \\ Y(S) = G_3(S) E^*(S) \\ E_1(S) = G_2(S) E^*(S) \end{cases}$$

$$\begin{cases} E(z) = R(z) - G_2(z) G_3(z) E(z) \\ Y(z) = G_3(z) E(z) \\ E_1(z) = G_2(z) E(z) \end{cases}$$

$$\frac{R(z)}{E(z)} = G_2(z) \rightarrow G_3(z)$$

$$\frac{R(S)}{E(S)} = \frac{H(S)}{H_0(S)} \frac{G_p(S)}{G_p(S)} \frac{Y(S)}{Y^*(S)}$$

$$E(S) = R(S) - H(S) H_0(S) G_p(S) M^*(S)$$

$$Y(S) = H_0(S) G_p(S) M^*(S)$$

$$E(z) = R(z) - H_0 G_p(z) M(z)$$

$$Y(z) = H_0 G_p(z) M(z)$$

$$M(z) = D(z) E(z)$$

$$\frac{R(S)}{E(S)} = \frac{G_1(S)}{Y(S)} \quad Y^*(S)$$

$$\begin{cases} Y(S) = E(S) G_1(S) \\ E(S) = R(S) - G_2(S) Y^*(S) \end{cases}$$

$$\frac{R(S)}{E(S)} = \frac{G_1(S)}{G_2(S) Y(S)}$$





$$K = \left| \frac{1}{G(z)} \right| = 178$$

$$178 = \frac{1}{|G(z)|}$$

$$\left( P(z) = z^3 + 0.4z^2 + 0.25z \right) (z+2) + K(z+0.5)$$

$$K = 178$$

$$P(z) = z^3 + 2.4z^2 + 2.87z + 1.97$$

$$\begin{aligned} & z^3 + 2.4z^2 + 2.87z + 1.97 \\ & z^3 + z^2 \\ & // 1.4z^2 + 2.87z + 1.97 \\ & 1.4z^2 + 1.4z \\ & // 1.4 + 1.47 \end{aligned}$$

$$\begin{aligned} & 1.4z + 1.47 \\ & // \end{aligned}$$

$$G(z) = \frac{2}{z^2(z^2 - 1.4z - 1)}$$

$$\begin{aligned} & x_1^2 = 0 \\ & x_2 = \pm \sqrt{1.4} \pm j0.5 \end{aligned}$$

$$|x_{12}| = 1$$

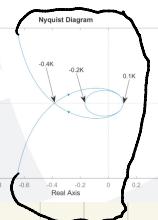
Contorno di Nyquist

$$N_{\text{Nyq}} = P - Z$$

$$N \leq 0$$

6. Sia assegnato il diagramma polare in figura, corrispondente ad un sistema tempo-discreto privo di zeri e con 2 poli di modulo inferiore a uno, un polo di modulo unitario, un polo di modulo superiore a uno. Supponendo di chiudere il sistema in retroazione unitaria ed assumendo un guadagno  $K$  variabile reale, si indichino le due affermazioni vere.

- A Per  $0 < K < 1$  il sistema è asintoticamente stabile
- B Per  $K=0$  il sistema è asintoticamente stabile
- C Il sistema è instabile per qualsiasi valore di  $K$
- D Il sistema è asintoticamente stabile per qualsiasi valore di  $K$
- E Per  $0 < K < 2.5$  il sistema è asintoticamente stabile
- F Per  $2.5 < K < 5$  il sistema presenta 3 poli fuori dal cerchio unitario
- G Per  $K=5$  il sistema presenta 3 poli fuori dal cerchio unitario
- H Per  $10 < K < 25$  il sistema è asintoticamente stabile
- I Solo una delle precedenti affermazioni è vera



instabile 11/06

$$0 < K < 1/04$$

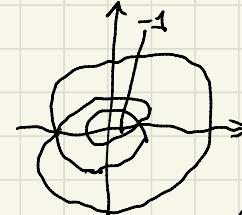
$$N = P - Z \quad 8 = 1 \quad \text{poli 3 stabili} \\ 0 \quad 1 \quad \text{1 instabile}$$

$$-0.4K \leq 1 \leq -0.2K$$

$$\begin{aligned} & N = P - Z \quad Z = 3 \quad \text{poli} \\ & -2 \quad f_1 \quad \text{instabili: 3} \\ & \text{stabili: 1} \end{aligned}$$

$$-0.2K \leq 1$$

$$\begin{aligned} & N = P - Z \Rightarrow Z = 4 \quad \text{poli} \\ & -3 \quad \text{stabili: 3} \\ & \text{instabili: 1} \end{aligned}$$



$$0 < -1 < +0.1K$$

$$N = P - Z \quad Z = 4 \\ -3 \quad 1$$

$$-1 > 0 > 1K$$

$$N = P - Z \\ -1 \quad 1 \\ Z = 2$$

$$M_F = \pi + \angle G(e^{j\omega T})$$

$$M_A = \frac{1}{|G(j\omega T)|}$$

$$S\% = \frac{Y(t_p) - Y(0)}{Y(0)} \cdot 100 = 100 e^{-\frac{\pi}{\sqrt{1-s^2}}}$$

$$\delta = \omega_0 \alpha \\ -s(\omega_0 \pm j\omega_0 \sqrt{1-s^2})$$

$$n(K)=1 \quad R(z) = \frac{3}{z-1} \quad \text{errore di posizionamento}$$

$$n(K)=KT \quad R(z) = \frac{Tz}{(z-1)^2} \quad // \text{di rimbombo}$$

$$n(K) = \frac{K^2 T^2}{2} \quad R(z) = \frac{T^2 z(z+1)}{2(z-1)^3} \quad // \text{di accelerazione}$$

2 tipi ep lv la

$$0 \quad \frac{1}{z-1} \quad \infty \quad \infty$$

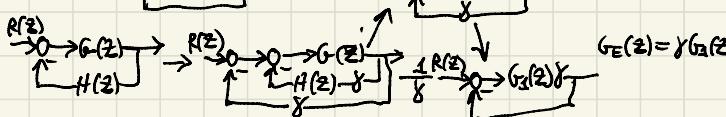
$$1 \quad 0 \quad 1/KV \quad \infty$$

$$2 \quad 0 \quad 0 \quad 1/Ka$$

$$K_P = \lim G(z)$$

$$K_V = \lim_{z \rightarrow 1} \frac{z-1}{Tz} G(z)$$

$$K_a = \lim_{z \rightarrow 1} \frac{(z-1)^2}{T^2} G(z)$$



$$E_P = \frac{1}{z-1} K_P \quad K_P = \lim G(z)$$

$$E_V = \frac{1}{KV} \quad K_V = \lim_{z \rightarrow 1} \frac{z-1}{Tz} G(z)$$

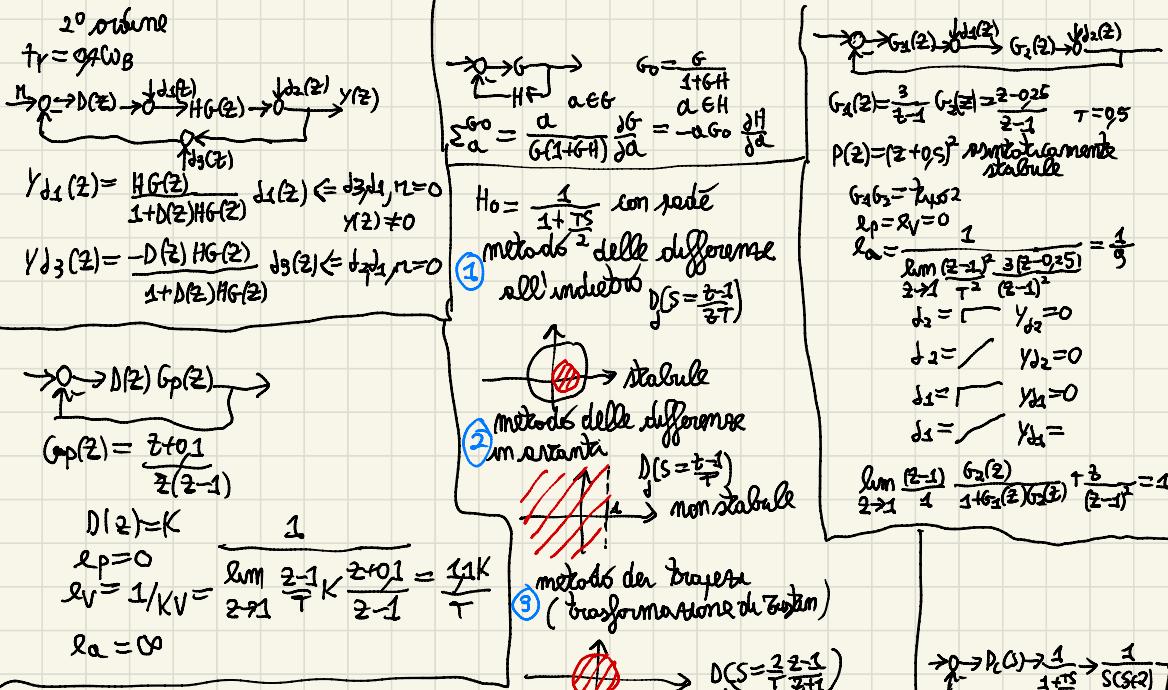
$$E_a = \frac{1}{Ka} \quad K_a = \lim_{z \rightarrow 1} \frac{(z-1)^2}{T^2} G(z)$$

Risposta → 1° ordine

$$T_a \tau_{10} = 42, \quad T_a \tau_{50} = 37, \quad T_a \tau_{100} = 4.62$$

$$T_a \sigma_{10} = \frac{3}{J\omega_n}$$

$$\begin{aligned} & \tau_{10} = \frac{\pi}{\omega_n} = \frac{4.6}{J\omega_n} \\ & \tau_{50} = \frac{\pi}{\omega_n} = \frac{3}{J\omega_n} \\ & \tau_{100} = \frac{\pi}{\omega_n} = \frac{1}{J\omega_n} \end{aligned}$$



risposta al impulso

 $Z(D_C(S)) = Y(z)$ 

risposta al gradino

 $D_f(z) = (1-z^{-1}) z \left\{ \frac{D_C(S)}{S} \right\}$

trasformate polari

 $D_C(S) = K(S-p_1)(S^2 + 2\omega_n S + \omega_n^2)$ 
 $S^2 + 2\omega_n S + \omega_n^2 = (S-p_1)(S-p_2)$ 
 $D_f(z) = K \frac{1}{(z-p_1)(z-p_2)} (z^2 - 2e^{-j\omega_n T} z + e^{-2j\omega_n T})$ 
 $\sum_{S \rightarrow 0} D_C(S) = \lim_{z \rightarrow 1} (z-1)^B D_f(z)$

$\rightarrow 0 \rightarrow D(z) \rightarrow \frac{1}{1+\frac{T}{T}} \rightarrow \frac{1}{S(S+T)}$ 
 $\epsilon_p = 0 \rightarrow D(z) \rightarrow 0 \rightarrow \frac{1}{S(S+T)} \rightarrow$ 
 $S_0 = 1.79$ 
 $T = 100 \text{ e } \frac{-17}{\sqrt{82}}$ 
 $\zeta = 0.5$ 
 $\frac{4}{\sqrt{82}} = 2 \delta \omega_n = 2$ 
 $-2 \pm 2\sqrt{3}$

1)  $w_b$  grande permanente

 $w_5 = 4 \div 10 \div 20 w_6$ 
 $w_6 \approx 25 \approx 14,44 \text{ rad/s}$ 
 $T_n (0-100\%) = \frac{\pi - \alpha}{\omega_d} = 0,605$ 
 $\delta = \omega_0 \alpha \text{ de a de a}$ 
 $T = \frac{2\pi}{\omega_5} = (0,072; 0,152; 0,38)$

$K = \frac{1}{G''(S)} = S^3 + (10+\alpha)^2 + 100\alpha S + K'$

$S^3 + 4S^2 + 16S$

$\text{---} (6+\alpha)S^2 + (100-16)S + K'$

$(6+\alpha)S^2 + (84+4\alpha)S + 16(6+\alpha)$ 
 $\text{---} (6\alpha-40)S - 96-16\alpha + K'$

2)  $\zeta = 0.5$

 $T = \frac{9\pi}{2 \cdot 10} = (0.05 \div 0.25)$ 
 $T = \frac{5\pi}{2\sqrt{3}} = (0.28 \div 0, 25)$ 

3)  $T = 0.2$

 $G(S) = 10K$ 
 $D_C(S) = 1$ 
 $D_C = \frac{(S+2)}{S+2} \quad G''(S) = \frac{K^2}{S(S+2)(S+10)}$ 
 $K^2 = 10K \quad K' = 10K$

$\begin{cases} s^2 + 4S + 16 \\ S + 6 + \alpha \end{cases} \quad \begin{cases} 6\alpha - 40 = 0 \\ -96 - 16\alpha + K' = 0 \end{cases}$ 
 $K' = 202$ 
 $\alpha = 6.67$

$$D_C(S) = \frac{S+2}{S+667} \quad T=0,2$$

$$\rightarrow D_C(S) \xrightarrow{\frac{1}{z}} G_P(z) \rightarrow$$

① differenza in esercizio

$$D_d(z) = D_C(S) \Big|_{S=\frac{z-1}{z}} = \frac{1-0,6z^{-1}}{1+0,334z^{-1}} - \frac{u(z)}{e(z)}$$

$$u(K) = -0,334u(K-1) + e(K) - 0,66e(K-1)$$

② differenza in industria

$$D_d(z) = D_C(S) \Big|_{S=\frac{z-1}{z}} = \frac{0,599z-0,428}{z-0,479}$$

$$u(K) = 0,428u(K-1) + 0,599e(K) - 0,428e(K-1)$$

③ metodo di tangent

$$D_d(z) = D_C(S) \Big|_{S=\frac{2}{T} \frac{z-1}{z-1}} = \frac{0,719z-0,4799}{z-0,2998}$$

$$u(K) = 0,4958u(K-1) + 0,7199e(K) - 0,4799e(K-1)$$

④ buoni presupposti

$$S+2 = K(z+2s)$$

$$S+667 = \frac{1}{1+0,6z}$$

$$w = \frac{1}{2\pi j\omega} = 3,65 \text{ rad/s}$$

$$\zeta = 0,5 \quad \alpha = 0,3$$

$$D_d(z) = D_C(S) \Big|_{S=\frac{w}{\zeta g w^2 z} \frac{z-1}{z-1}} =$$

$$= \frac{0,912z-0,465}{z-0,477}$$

$$u(K) = 0,477u(K-1) + 0,912e(K) - 0,465e(K-1)$$

⑤ inviare una risposta all'imulo

$$u(z) = z \left\{ D_c(S) \right\} = z \left\{ \frac{0,599}{z-0,479} \right\} - 0,599e^{-667z}$$

⑥ inviare una risposta a gradins

$$D_d(z) = (1-z^{-1})z \left\{ \frac{D_c(S)}{S} \right\} = \frac{z-0,779}{z-0,163}$$

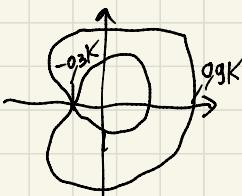
⑦ pole / zero

$$N(z) = K_d(z-e^{-\frac{1}{T}})$$

$$K_d = \lim_{S \rightarrow 0} \frac{S+2}{S+667} = \lim_{z \rightarrow 1} \left( \frac{z-e^{-\frac{1}{T}}}{z-e^{-\frac{1}{T}}-0,667} \right) \cdot 0,66$$

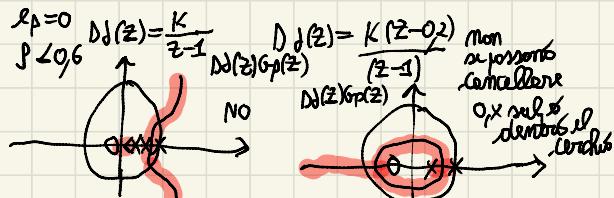
$$D_d(z) = 0,669 \frac{z-0,67}{z-0,265}$$

$$u(K) = 0,263u(K-1) + 0,669e(K) - 0,449e(K-1)$$



- 1 < 0,3K inst. stabile
- 1 > -0,3K instabile
- 1 = -0,3K semplic. stabile

$$G_p(z) = \frac{z+0,5}{(z-0,7)(z-0,2)} \rightarrow D(z)G_p(z) \rightarrow$$



$$D_d(z) = \frac{K(z-0,2)}{(z-1)(z+0,5)}$$

$$K(z) = \frac{1}{(G_p(z))D_d(z)} = 99$$

$$\frac{\partial K(z)}{\partial z} = 0$$

$$z^2 - 0,3z + 1,70 - 0,7 = 0$$

$$z_{1,2} = \alpha \pm \sqrt{\alpha^2 - 1,70 + 0,7}$$

$$\alpha = 0,7 \sqrt{0,21}$$

$$D_d = \frac{0,9(z-0,2)(z-0,6)}{(z+0,5)(z-1)}$$

$$\rightarrow 0 \rightarrow D(z) \rightarrow G_p(z) \rightarrow \text{dead beat } \#_1$$

$$G_p(z) = \frac{z+0,8}{z^2 + 0,7z + 0,1} \quad \begin{matrix} \text{dentro il} \\ \text{cerchio} \end{matrix}$$

$$D(z) = \frac{1}{G_p(z)} \frac{z^n - 1}{z^n - 1} = \frac{z^2 + 0,7z + 0,1}{(z+0,8)(z-1)}$$

$$H_0 G_p(z)$$

$$G_p(z) = \frac{z^2(0,143 + 0,209z^{-1})}{1-0,774z^{-1}} \quad \text{dead beat } \#_2$$

$$\sum p_i = 0,143 + 0,209 = 0,352$$

$$G_p(z) = \frac{(z^2(0,143 + 0,209z^{-1})) / 0,258}{(1-0,774z^{-1}) / 0,258} = \frac{z^2(0,577 + 0,943z^{-1})}{3,86 - 2,86z^{-1}}$$

$$D(z) = \frac{3,86 - 2,86z^{-1}}{1-z^2(0,577 + 0,943z^{-1})}$$

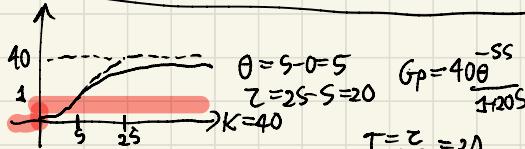
$$G(s) = K_p + \frac{K_I}{s} + K_D s = K_p \left( 1 + \frac{1}{T_I s} + T_D s \right)$$

differenze in esercit

$$G(s) \Big|_{S=\frac{2}{T}} \Rightarrow D(z) = \frac{q_0 + q_1 z^{-1} + q_2 z^{-2}}{1-z^{-1}} = \frac{u}{E}$$

differenze all'indietro

$$G(S) \Big|_{S=\frac{T(2-t)}{2}} = K_P \left( 1 + \frac{\alpha}{2-t} + \beta \frac{(2-t)}{2-t} \right) = \frac{M}{E} = \frac{q_0 + q_1 t^{-1} + q_2 t^{-2}}{(1-t^{-1})(2-t^{-1})}$$

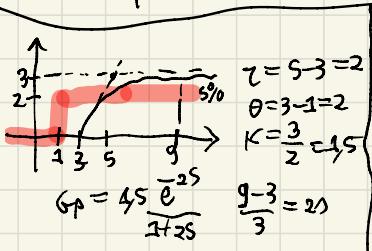


$$N_2 = 9 + 9 \cdot 2^{-1} / (1 - 2^{-1})$$

$$q_0 = \left(1 - \frac{T^0}{T}\right) k_F = 9075$$

$$q_1 = \left(-1 + \frac{1}{T} - 2\frac{T}{T}\right)k_p = -9067$$

$$q_2 = \frac{K D D}{\pi} = 0$$



三

$$y_p = A \left(\frac{\theta}{\tau}\right)^B = 435$$

$$K_p = \frac{y^p}{x} = 0,209$$

$$Y_3 = 0.878 \left( \frac{6}{\cdot} \right)^{-0.749}$$

7-9-45

$$Y_0 = 0.982 \left( \frac{6}{20} \right)^{4.137} Y = \frac{T_D}{T_0} = 0.11$$

$$T_D = \mathcal{C} Y_D$$

$$G_p = 40 \frac{\bar{e}^{-5S}}{1+20S}$$

$$\tau = 20$$

1

$$D(z) = Kp\left(1 + \frac{\alpha}{z-1} + \frac{\beta z-1}{z-\gamma}\right)$$

$$\alpha = \frac{I}{T_S} = \frac{2}{9.45} = 0.216$$

$$\beta = \frac{NTD}{NT-ATA} = \frac{10.2452}{307.1952} = 1.1$$

$$\begin{aligned}
 &= 0,228 \mu(K) \\
 &- 0,335 \mu(K-1) \\
 &+ 0,128 \mu(K-2) \\
 &+ 0,109 \mu(K-3) \\
 &- 0,109 \mu(K-4)
 \end{aligned}$$

distrubbi  $\Rightarrow$  da corris  
sull'uscita  
ottimizzazione  $\Rightarrow$  da  
nella risposta set point

nelle risposte setpoint

$$\mu(k) = 0.228 \varrho(k)$$

$$E = -933.52(K-1) + 0.128 \cdot 8(K-2)$$

$$+ \zeta_{109} \alpha(k-1)$$

$$-0.109 \mu(k-2)$$

$$g_{11110}(\text{controlls}) > 0$$

## Integratore accumula

## regione lineare

solutions  
↳ closure  
integrator  
( $\exists$  in saturation)