











ISO7760, ISO7761 ISO7762, ISO7763

SLLSER1E -AUGUST 2017-REVISED MARCH 2019

# ISO776x High-speed, robust EMC, reinforced six-channel digital isolators

#### **Features**

- 100 Mbps data rate
- Robust isolation barrier:
  - >100-Year projected lifetime
  - Up to 5000 V<sub>RMS</sub> isolation rating
  - Up to 12.8 kV surge capability
  - ±100 kV/μs Typical CMTI
- Wide supply range: 2.25 V to 5.5 V
- 2.25-V to 5.5-V Level translation
- Default output high (ISO776x) and low (ISO776xF) Options
- Wide temperature range: -55°C to +125°C
- Low power consumption, typical 1.4 mA per channel at 1 Mbps
- Low propagation delay: 11 ns typical at 5 V
- Robust Electromagnetic Compatibility (EMC):
  - System-level ESD, EFT, and surge immunity
  - ±8 kV IEC 61000-4-2 Contact discharge protection across isolation barrier
  - Low emissions
- Wide-SOIC (DW-16) and SSOP (DBQ-16) package options
- Automotive version available: ISO776x-Q1
- Safety-related certifications:
  - Reinforced insulation per DIN V VDE V 0884-11:2017-01
  - UL 1577 component recognition program
  - CSA Certification per IEC 60950-1, IEC 62368-1, and IEC 60601-1
  - CQC Certification per GB4943.1-2011
  - TUV Certification according to EN 60950-1 and EN 61010-1

# **Applications**

- Industrial automation
- Motor control
- Power supplies
- Solar inverters
- Medical equipment

#### Description

The ISO776x devices are high-performance, sixchannel digital isolators with 5000-V<sub>RMS</sub> package) and 3000-V<sub>RMS</sub> (DBQ package) isolation ratings per UL 1577. This family of devices is also certified according to VDE, CSA, TUV and CQC.

The ISO776x family of devices provides highelectromagnetic immunity and low emissions at lowpower consumption, while isolating CMOS or LVCMOS digital I/Os. Each isolation channel has a logic-input and logic-output buffer separated by a double capacitive silicon dioxide (SiO2) insulation barrier. The ISO776x family of devices is available in all possible pin configurations such that all six channels are in the same direction, or one, two, or three channels are in reverse direction while the remaining channels are in forward direction. If the input power or signal is lost, the default output is high for devices without suffix F and low for devices with suffix F. See the *Device Functional Modes* section for further details.

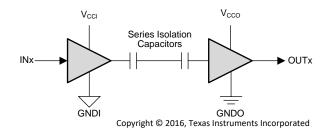
Used in conjunction with isolated power supplies, this family of devices helps prevent noise currents on data buses, such as RS-485, RS-232, and CAN, or other circuits from entering the local ground and interfering with or damaging sensitive circuitry. chip design Through innovative and layout techniques, electromagnetic compatibility of the ISO776x family of devices has been significantly enhanced to ease system-level ESD, EFT, surge, and emissions compliance. The ISO776x family of devices is available in 16-pin SOIC and SSOP packages.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
ISO7760	SOIC (16)	10.30 mm × 7.50 mm
ISO7761 ISO7762 IOS7763	SSOP (16)	4.90 mm × 3.90 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

#### Simplified Schematic



V<sub>CCI</sub>=Input V<sub>CC</sub>, V<sub>CCO</sub>=Output V<sub>CC</sub> GNDI=Input ground, GNDO=Output ground



#### **Table of Contents**

1	Features 1		6.19 Typical Characteristics	19
2	Applications 1	7	Parameter Measurement Information	22
3	Description 1	8	Detailed Description	23
4	Revision History2		8.1 Overview	23
5	Pin Configuration and Functions4		8.2 Functional Block Diagram	23
6	Specifications6		8.3 Feature Description	<mark>2</mark> 4
-	6.1 Absolute Maximum Ratings 6		8.4 Device Functional Modes	25
	6.2 ESD Ratings	9	Application and Implementation	26
	6.3 Recommended Operating Conditions		9.1 Application Information	26
	6.4 Thermal Information		9.2 Typical Application	26
	6.5 Power Ratings	10	Power Supply Recommendations	30
	6.6 Insulation Specifications8	11	Layout	31
	6.7 Safety-Related Certifications9		11.1 Layout Guidelines	<mark>3</mark> 1
	6.8 Safety Limiting Values9		11.2 Layout Example	31
	6.9 Electrical Characteristics—5-V Supply 10	12	Device and Documentation Support	32
	6.10 Supply Current Characteristics—5-V Supply 11		12.1 Documentation Support	32
	6.11 Electrical Characteristics—3.3-V Supply 12		12.2 Related Links	32
	6.12 Supply Current Characteristics—3.3-V Supply 13		12.3 Receiving Notification of Documentation Up	dates 32
	6.13 Electrical Characteristics—2.5-V Supply 14		12.4 Community Resources	32
	6.14 Supply Current Characteristics—2.5-V Supply 15		12.5 Trademarks	32
	6.15 Switching Characteristics—5-V Supply 16		12.6 Electrostatic Discharge Caution	33
	6.16 Switching Characteristics—3.3-V Supply 16		12.7 Glossary	33
	6.17 Switching Characteristics—2.5-V Supply 17	13	Mechanical, Packaging, and Orderable	
	6.18 Insulation Characteristics Curves		Information	33

#### 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

#### Changes from Revision D (November 2018) to Revision E

**Page** 

#### Changes from Revision C (January 2018) to Revision D

Page

•	Changed From: "Isolation Barrier Life: >40 Years" To:">100-Year Projected Lifetime" in Features
•	Added "Up to 5000 V <sub>RMS</sub> Isolation Rating" in Features
•	Added "Up to 12.8 kV Surge Capability" in Features
•	Added "±8 kV IEC 61000-4-2 Contact Discharge Protection across Isolation Barrier" in Features
•	Added "Automotive Version Available: ISO776x-Q1" in Features
•	Deleted "Certification Planned" statement throughout the document
•	Updated Applications list
•	Changed Simplified Schematic to show series isolation capacitors
•	Added "Contact discharge per IEC 61000-4-2" specification of ±8000 V in ESD Ratings table
•	Added table note to Data rate specification in Recommended Operating Conditions table
•	Changed V <sub>IORM</sub> Value for DW-16 package From: "1414 V <sub>PK</sub> " To: "2121 V <sub>PK</sub> " in Insulation Specifications table
•	Changed $V_{\text{IOWM}}$ Values for DW-16 package From: "1000 $V_{\text{RMS}}$ " and "1414 $V_{\text{DC}}$ " To: "1500 $V_{\text{RMS}}$ " and "2121 $V_{\text{DC}}$ " in Insulation Specifications table

Submit Documentation Feedback

Copyright © 2017–2019, Texas Instruments Incorporated





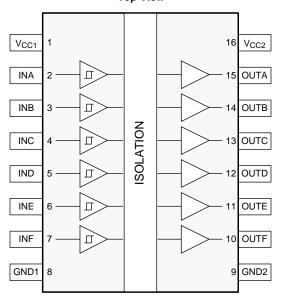
www.	tı	ഹ	m

•	Changed From: "Table 2" To: "Safety Related Certifications" in Table 1 table note	
•	Changed Figure 23	
<u>.</u>	Added Insulation Lifetime sub-section under Application Curves section	29
٥.	6	
Cr	nanges from Revision B (November 2017) to Revision C	Page
•	Changed the C <sub>IO</sub> value for the DBQ package from 1.1 to 0.9 pF in the <i>Insulation Specifications</i> table	8
Cł	nanges from Revision A (August 2017) to Revision B	Page
•	Changed the CSA certification wording in the Features and Safety-Related Certifications table	1
•	Changed the isolation voltage for the DBQ-16 package from 2500 to 3000 V <sub>RMS</sub>	1
•	Added the maximum transient isolation voltage for the DW-16 package of the ISO7761, ISO7762, and ISO7763 devices in the <i>Insulation Specifications</i> and <i>Safety-Related Certifications</i> tables. Also changed the maximum value for the DBQ-16 package from 3600 to 4242 for all devoies	
•	Changed the table note and table condition for the Safety Limiting Values	
•	Added the supply current vs data rate graphs for the ISO7761, ISO7762, and ISO7763 in the <i>Typical</i>	
	Characteristics section	19
Cł	nanges from Original (August 2017) to Revision A	Page
•	Deleted EN from the Common-Mode Transient Immunity Test Circuit figure	22
•	Changed the V <sub>CC1</sub> and V <sub>CC2</sub> signals in the <i>Typical ISO7761 Circuit Hook-up</i> figure	

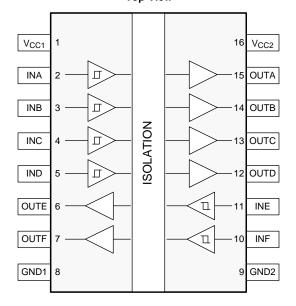


# 5 Pin Configuration and Functions

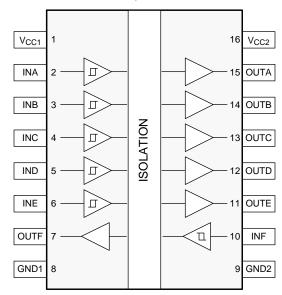
ISO7760 DW and DBQ Packages 16-Pin SOIC and SSOP Top View



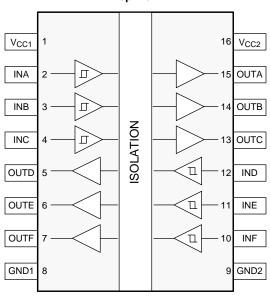
ISO7762 DW and DBQ Packages 16-Pin SOIC and SSOP Top View



#### ISO7761 DW and DBQ Packages 16-Pin SOIC and SSOP Top View



#### ISO7763 DW and DBQ Packages 16-Pin SOIC and SSOP Top View





## **Pin Functions**

		PIN				
NAME		NO.			I/O	DESCRIPTION
INAIVIE	ISO7760	ISO7761	ISO7762	ISO7763		
GND1	8	8	8	8	_	Ground connection for V <sub>CC1</sub>
GND2	9	9	9	9	_	Ground connection for V <sub>CC2</sub>
INA	2	2	2	2	I	Input, channel A
INB	3	3	3	3	I	Input, channel B
INC	4	4	4	4	I	Input, channel C
IND	5	5	5	12	I	Input, channel D
INE	6	6	11	11	I	Input, channel E
INF	7	10	10	10	I	Input, channel F
OUTA	15	15	15	15	0	Output, channel A
OUTB	14	14	14	14	0	Output, channel B
OUTC	13	13	13	13	0	Output, channel C
OUTD	12	12	12	5	0	Output, channel D
OUTE	11	11	6	6	0	Output, channel E
OUTF	10	7	7	7	0	Output, channel F
V <sub>CC1</sub>	1	1	1	1	_	Power supply, side 1
$V_{CC2}$	16	16	16	16		Power supply, side 2



## 6 Specifications

#### 6.1 Absolute Maximum Ratings

See (1)

		MIN	MAX	UNIT
V <sub>CC1</sub> , V <sub>CC2</sub>	Supply voltage <sup>(2)</sup>	-0.5	6	V
V	Voltage at INx, OUTx	-0.5	$V_{CCX} + 0.5^{(3)}$	V
Io	Output current	-15	15	mA
$T_{J}$	Junction temperature		150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
- (3) Maximum voltage must not exceed 6 V

#### 6.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±6000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101 (2)	±1500	V
		Contact discharge per IEC 61000-4-2; Isolation barrier withstand test <sup>(3)(4)</sup>	±8000	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.
- (3) IEC ESD strike is applied across the barrier with all pins on each side tied together creating a two-terminal device.
- (4) Testing is carried out in air or oil to determine the intrinsic contact discharge capability of the device.

## 6.3 Recommended Operating Conditions

			MIN	NOM	MAX	UNIT
V <sub>CC1</sub> , V <sub>CC2</sub>	Supply voltage		2.25		5.5	V
V <sub>CC(UVLO+)</sub>	UVLO threshold when supply	voltage is rising		2	2.25	V
V <sub>CC(UVLO-)</sub>	UVLO threshold when supply	voltage is falling	1.7	1.8		<b>V</b>
V <sub>HYS(UVLO)</sub>	Supply voltage UVLO hysteres	sis	100	200		mV
		$V_{CCO}^{(1)} = 5 \text{ V}$	-4			
I <sub>OH</sub>	High-level output current	V <sub>CCO</sub> = 3.3 V	-2			mA
		V <sub>CCO</sub> = 2.5 V	-1			
		V <sub>CCO</sub> = 5 V			4	
I <sub>OL</sub>	Low-level output current	V <sub>CCO</sub> = 3.3 V			2	mA
		$V_{CCO} = 2.5 \text{ V}$			1	
V <sub>IH</sub>	High-level input voltage		0.7 × V <sub>CCI</sub> <sup>(1)</sup>		V <sub>CCI</sub>	V
V <sub>IL</sub>	Low-level input voltage		0		0.3 × V <sub>CCI</sub>	V
DR (2)	Data rate		0		100	Mbps
T <sub>A</sub>	Ambient temperature		-55	25	125	°C

(1)  $V_{CCI} = Input\text{-side } V_{CC}$ ;  $V_{CCO} = Output\text{-side } V_{CC}$ .

(2) 100 Mbps is the maximum specified data rate, although higher data rates are possible.



#### 6.4 Thermal Information

		ISO776x		
	THERMAL METRIC <sup>(1)</sup>	DW (SOIC)	DBQ (SSOP)	UNIT
		16 PINS	16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	60.3	86.5	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	24.0	26.9	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	29.3	36.6	°C/W
ΨЈТ	Junction-to-top characterization parameter	3.3	1.7	°C/W
ΨЈВ	Junction-to-board characterization parameter	28.7	36.1	°C/W
$R_{\theta JC(bottom)}$	Junction-to-case(bottom) thermal resistance	n/a	n/a	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

# 6.5 Power Ratings

0.5	6.5 Power Ratings								
	PARAMETER	TEST CONDITIONS	MIN TYP	MAX	UNIT				
ISO7	760			•					
P <sub>D</sub>	Maximum power dissipation (both sides)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		292	mW				
P <sub>D1</sub>	Maximum power dissipation (side 1)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		50	mW				
P <sub>D2</sub>	Maximum power dissipation (side 2)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		242	mW				
ISO7	507761								
P <sub>D</sub>	Maximum power dissipation (both sides)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		292	mW				
P <sub>D1</sub>	Maximum power dissipation (side 1)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		83	mW				
P <sub>D2</sub>	Maximum power dissipation (side 2)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		209	mW				
ISO7	762								
$P_D$	Maximum power dissipation (both sides)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		292	mW				
P <sub>D1</sub>	Maximum power dissipation (side 1)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		116	mW				
P <sub>D2</sub>	Maximum power dissipation (side 2)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		176	mW				
ISO7	763								
$P_D$	Maximum power dissipation (both sides)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		292	mW				
P <sub>D1</sub>	Maximum power dissipation (side 1)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		146	mW				
P <sub>D2</sub>	Maximum power dissipation (side 2)	$V_{\rm CC1}$ = $V_{\rm CC2}$ = 5.5 V, $T_{\rm J}$ = 150°C, $C_{\rm L}$ = 15 pF, input a 50-MHz 50% duty cycle square wave		146	mW				

# TEXAS INSTRUMENTS

## 6.6 Insulation Specifications

						VALUE		LUE	
	PARAMETER	TEST CONDIT	IONS	DW-16	DBQ- 16	UNIT			
CLR	External clearance <sup>(1)</sup>	Shortest terminal-to-terminal distance the	nrough air	>8	>3.7	mm			
CPG	External creepage (1)	Shortest terminal-to-terminal distance a	cross the package surface	>8	>3.7	mm			
DTI	Distance through the insulation	Minimum internal gap (internal clearance	ee)	>21	>21	μm			
СТІ	Tracking resistance (comparative tracking index)	DIN EN 60112 (VDE 0303-11); IEC 601	12; UL 746A	>600	>600	٧			
	Material group	According to IEC 60664-1		1	- 1				
		Rated mains voltage ≤ 150 V <sub>RMS</sub>		I–IV	I–IV				
	Overvoltage category per IEC	Rated mains voltage ≤ 300 V <sub>RMS</sub>		I–IV	I–III				
	60664-1	Rated mains voltage ≤ 600 V <sub>RMS</sub>		I–IV	n/a				
		Rated mains voltage ≤ 1000 V <sub>RMS</sub>		I–III	n/a				
DIN V	VDE V 0884-11:2017-01 <sup>(2)</sup>								
V <sub>IORM</sub>	Maximum repetitive peak isolation voltage	AC voltage (bipolar)		2121	566	V <sub>PK</sub>			
V <sub>IOWM</sub>	Maximum working isolation	AC voltage; Time dependent dielectric breakdown (TDDB) test; see Figure 30  DC voltage		1500	400	V <sub>RMS</sub>			
· IOVVIVI	voltage	DC voltage		2121	566	$V_{DC}$			
		$V_{TEST} = V_{IOTM}$ , $t = 60$ s (qualification)	ISO7760	8000 7071 4242					
$V_{IOTM}$	Maximum transient isolation voltage	rest = 1.2 x V <sub>IOTM</sub> , t = 1 s (100% ISC ISC	ISO7761, ISO7762, ISO7763		4242	V <sub>PK</sub>			
V <sub>IOSM</sub>	Maximum surge isolation voltage <sup>(3)</sup>	Test method per IEC 62368-1, 1.2/50 µ: V <sub>TEST</sub> = 1.6 × V <sub>IOSM</sub> (qualification)	s waveform,	8000	4000	V <sub>PK</sub>			
		Method a, After Input/Output safety test subgroup 2/3, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60 \text{ s}$ ; $V_{pd(m)} = 1.2 \times V_{IORM}$ , $t_m = 10 \text{ s}$		≤5	≤5				
$q_{pd}$	Apparent charge <sup>(4)</sup>		ogroup 1,	≤5	≤5	рС			
		$\label{eq:method b1; At routine test (100% productive test)}                                    $	ction) and preconditioning	≤5	≤5				
C <sub>IO</sub>	Barrier capacitance, input to output (5)	$V_{IO} = 0.4 \times \sin(2\pi ft), f = 1 \text{ MHz}$		~1.1	~0.9	pF			
		V <sub>IO</sub> = 500 V, T <sub>A</sub> = 25°C		>10 <sup>12</sup>	>10 <sup>12</sup>				
$R_{IO}$	Isolation resistance (5)	V <sub>IO</sub> = 500 V, 100°C ≤ T <sub>A</sub> ≤ 125°C		>10 <sup>11</sup>	>10 <sup>11</sup>	Ω			
		V <sub>IO</sub> = 500 V, T <sub>S</sub> = 150°C		>10 <sup>9</sup>	>10 <sup>9</sup>				
	Pollution degree			2	2				
	Climatic category			55/125/ 21	55/125/ 21				
UL 157	7					•			
V <sub>ISO</sub>	Withstanding isolation voltage	$V_{TEST} = V_{ISO}$ , t = 60 s (qualification), $V_{TEST} = 1.2 \times V_{ISO}$ , t = 1 s (100% prod	uction)	5000	3000	V <sub>RMS</sub>			

<sup>(1)</sup> Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.

Submit Documentation Feedback

Copyright © 2017–2019, Texas Instruments Incorporated

<sup>(2)</sup> This coupler is suitable for safe *electrical insulation only* within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits

<sup>(3)</sup> Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.

<sup>(4)</sup> Apparent charge is electrical discharge caused by a partial discharge (pd).

<sup>(5)</sup> All pins on each side of the barrier tied together creating a two-terminal device.



#### 6.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Certified according to DIN V VDE V 0884-11:2017-01	Certified according to IEC 60950-1, IEC 62368-1 and IEC 60601-1	Recognized under UL 1577 Component Recognition Program	Certified according to GB 4943.1-2011	Certified according to EN 61010-1:2010 (3rd Ed) and EN 60950- 1:2006/A2:2013
Reinforced Insulation; Maximum transient isolation voltage, 8000 V <sub>PK</sub> (ISO7760 in DW-16), 7071 V <sub>PK</sub> (ISO7761, ISO7762, ISO7763 in DW-16) and 4242 V <sub>PK</sub> (DBQ-16); Maximum repetitive peak isolation voltage, 2121 V <sub>PK</sub> (DW-16) and 566 V <sub>PK</sub> (DBQ-16); Maximum surge isolation voltage, 8000 V <sub>PK</sub> (DW-16) and 4000 V <sub>PK</sub> (DW-16) and 4000 V <sub>PK</sub> (DW-16)	Reinforced insulation per CSA 60950-1-07+A1+A2, IEC 60950-1 2nd Ed.+A1+A2, CSA 62368-1-14 and IEC 62368-1:2014 800 V <sub>RMS</sub> (DW-16) and 370 V <sub>RMS</sub> (DBQ-16) maximum working voltage (pollution degree 2, material group I); <b>DW-16:</b> 2 MOPP (Means of Patient Protection) per CSA 60601-1:14 and IEC 60601-1 Ed. 3.1, 250 V <sub>RMS</sub> maximum working voltage	<b>DW-16:</b> Single protection, 5000 V <sub>RMS</sub> ; <b>DBQ-16:</b> Single protection, 3000 V <sub>RMS</sub>	DW-16: Reinforced Insulation, Altitude ≤ 5000 m, Tropical Climate, 400 V <sub>RMS</sub> maximum working voltage; DBQ-16: Basic Insulation, Altitude ≤ 5000 m, Tropical Climate, 250 V <sub>RMS</sub> maximum working voltage	5000 V <sub>RMS</sub> Reinforced insulation per EN 61010-1:2010 (3rd Ed) up to working voltage of 600 V <sub>RMS</sub> (DW-16) and 300 V <sub>RMS</sub> (DBQ-16) 5000 V <sub>RMS</sub> Reinforced insulation per EN 60950-1:2006/A2:2013 up to working voltage of 800 V <sub>RMS</sub> (DW-16) and 370 V <sub>RMS</sub> (DBQ-16)
Certificate number: 40040142	Master contract number: 220991	File number: E181974	Certificate number: CQC15001121716 (DW) Certificate number: CQC18001199097 (DBQ)	Client ID number: 77311

## 6.8 Safety Limiting Values

Safety limiting intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

	PARAMETER	TEST CONDITIONS	MIN TYP MAX	UNIT
DW-	16 PACKAGE			
		$R_{\theta JA}$ = 60.3 °C/W, $V_I$ = 5.5 V, $T_J$ = 150°C, $T_A$ = 25°C, see Figure 1	377	
Is	Safety input, output, or supply current (1)	$R_{\theta JA}$ = 60.3 °C/W, $V_I$ = 3.6 V, $T_J$ = 150°C, $T_A$ = 25°C, see Figure 1	576	mA
		$R_{\theta JA}$ = 60.3 °C/W, $V_I$ = 2.75 V, $T_J$ = 150°C, $T_A$ = 25°C, see Figure 1	754	
$P_S$	Safety input, output, or total power <sup>(1)</sup>	$R_{\theta JA} = 60.3$ °C/W, $T_J = 150$ °C, $T_A = 25$ °C, see Figure 3	2073	mW
$T_S$	Maximum safety temperature (1)		150	°C
DBQ	-16 PACKAGE			
		$R_{\theta JA}$ = 86.5 °C/W, $V_I$ = 5.5 V, $T_J$ = 150°C, $T_A$ = 25°C, see Figure 2	263	
I <sub>S</sub>	Safety input, output, or supply current <sup>(1)</sup>	$R_{\theta JA}$ = 86.5 °C/W, $V_I$ = 3.6 V, $T_J$ = 150°C, $T_A$ = 25°C, see Figure 2	401	mA
		$R_{\theta JA}$ = 86.5 °C/W, $V_I$ = 2.75 V, $T_J$ = 150°C, $T_A$ = 25°C, see Figure 2	525	
Ps	Safety input, output, or total power <sup>(1)</sup>	$R_{\theta JA} = 86.5$ °C/W, $T_J = 150$ °C, $T_A = 25$ °C, see Figure 4	1445	mW
T <sub>S</sub>	Maximum safety temperature <sup>(1)</sup>		150	°C

<sup>(1)</sup> The maximum safety temperature, T<sub>S</sub>, has the same value as the maximum junction temperature, T<sub>J</sub>, specified for the device. The I<sub>S</sub> and P<sub>S</sub> parameters represent the safety current and safety power respectively. The maximum limits of I<sub>S</sub> and P<sub>S</sub> should not be exceeded. These limits vary with the ambient temperature, T<sub>A</sub>.

The junction-to-air thermal resistance, R<sub>0JA</sub>, in the *Thermal Information* table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$ , where P is the power dissipated in the device.

 $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$ , where  $T_{J(max)}$  is the maximum allowed junction temperature.

 $P_S = I_S \times V_I$ , where  $V_I$  is the maximum input voltage.



# 6.9 Electrical Characteristics—5-V Supply

 $V_{CC1} = V_{CC2} = 5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -4 mA; see Figure 18	$V_{\rm CCO}^{(1)} - 0.4$	4.8		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 4 mA; see Figure 18		0.2	0.4	V
V <sub>IT+(IN)</sub>	Rising input threshold voltage			0.6 x V <sub>CCI</sub>	0.7 x V <sub>CCI</sub>	V
V <sub>IT-(IN)</sub>	Falling input threshold voltage		0.3 x V <sub>CCI</sub>	0.4 x V <sub>CCI</sub>		V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 × V <sub>CCI</sub>	0.2 x V <sub>CCI</sub>		V
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = V <sub>CCI</sub> <sup>(1)</sup> at INx			10	μА
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10			μΑ
CMTI	Common-mode transient immunity	$V_I = V_{CCI}$ or 0 V, $V_{CM} = 1200$ V; see Figure 20	85	100		kV/μs
Cı	Input capacitance (2)	$V_1 = V_{CC} / 2 + 0.4 \times \sin(2\pi ft), f = 1$ MHz, $V_{CC} = 5 \text{ V}$		2		pF

$$<sup>\</sup>label{eq:VCC} \begin{split} V_{CCI} &= \text{Input-side V}_{CC}; \ V_{CCO} = \text{Output-side V}_{CC}. \\ \text{Measured from input pin to ground}. \end{split}$$



# 6.10 Supply Current Characteristics—5-V Supply

 $V_{CC1} = V_{CC2} = 5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN TYP	MAX	UNIT	
ISO7760							
	$V_{I} = V_{CC1}$ (ISO7760);		I <sub>CC1</sub>	1.6	2.3		
Supply current - DC	$V_1 = 0 \text{ V (ISO7760 with F suffix)}$		I <sub>CC2</sub>	3	4.9	mA	
signal	V <sub>I</sub> = 0 V (ISO7760);		I <sub>CC1</sub>	8	11.3		
	$V_I = V_{CC1}$ (ISO7760 with F suffix)	C1 (ISO7760 with F suffix)		3.3	5.3		
		1 Mbps	I <sub>CC1</sub>	5	6.4		
		1 IVIDPS	I <sub>CC2</sub>	3.5	5.6		
Supply current - AC	All channels switching with square wave	10 Mbps	I <sub>CC1</sub>	5.2	6.7	mA	
signal	clock input; C <sub>L</sub> = 15 pF	10 IVIDPS	I <sub>CC2</sub>	6.4	9	ША	
		100 Mbps	I <sub>CC1</sub>	7	9		
		100 lvibps	I <sub>CC2</sub>	35	44		
SO7761	·	,			·		
$V_1 = V_{CCI}^{(1)}(ISO7761);$			I <sub>CC1</sub>	1.9	2.7		
Supply current - DC	$V_I = 0 \text{ V (ISO7761 with F suffix)}$		I <sub>CC2</sub>	2.9	4.7	mΛ	
signal	V <sub>I</sub> = 0 V (ISO7761);		I <sub>CC1</sub>	7.3	10.6	mA	
	$V_I = V_{CCI}$ (ISO7761 with F suffix)		I <sub>CC2</sub>	4.2	6.6	6.6	
		1 Mbps	I <sub>CC1</sub>	4.7	6.4		
Supply current - AC signal		1 IVIDPS	I <sub>CC2</sub>	3.8	5.9		
	All channels switching with square wave	40 Mb = -	I <sub>CC1</sub>	5.3	7.2	mΛ	
	Clock input, C <sub>L</sub> = 15 pr	10 Mbps	I <sub>CC2</sub>	6.3	8.8	mA	
		400 Mb = =	I <sub>CC1</sub>	11.5	15		
		100 Mbps	I <sub>CC2</sub>	30.5	38		
SO7762					,		
	V <sub>I</sub> = V <sub>CCI</sub> (ISO7762);	$V_1 = V_{CC1}$ (ISO7762):		2.1	3.2		
Supply current - DC	$V_1 = 0 \text{ V (ISO7762 with F suffix)}$		I <sub>CC2</sub>	2.6	4.2	A	
signal	V <sub>I</sub> = 0 V (ISO7762);		I <sub>CC1</sub>	6.5	9.3	mA	
	$V_I = V_{CCI}$ (ISO7762 with F suffix)		I <sub>CC2</sub>	5	7.5		
		4 1 1 1 1	I <sub>CC1</sub>	4.5	6.3		
		1 Mbps	I <sub>CC2</sub>	4	6.1		
Supply current - AC	All channels switching with square wave	40 Mbps	I <sub>CC1</sub>	5.6	7.6	A	
signal	clock input; C <sub>L</sub> = 15 pF	10 Mbps	I <sub>CC2</sub>	6	8.4	mA	
		100 Mbps	I <sub>CC1</sub>	16.5	21		
		100 Mbps	I <sub>CC2</sub>	25.7	32		
SO7763					·		
Supply current - DC	$V_I = V_{CCI}$ (ISO7763); $V_I = 0$ V (ISO7763 with F suffix)		I <sub>CC1</sub> , I <sub>CC2</sub>	2.4	3.7	A	
signal	$V_I = 0 \text{ V (ISO7763)};$ $V_I = V_{CCI} \text{ (ISO7763 with F suffix)}$	V <sub>I</sub> = 0 V (ISO7763);		5.7	8.6	mA	
		1 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>	4.2	6.1		
Supply current - AC signal	All channels switching with square wave clock input; C <sub>1</sub> = 15 pF	10 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>	5.8	8	mA	
signai	Sook input, of - 10 bi	100 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>	21	26.5	1	

<sup>(1)</sup>  $V_{CCI} = Input\text{-side } V_{CC}$ 



## 6.11 Electrical Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

to the state of th						
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OH}$	High-level output voltage	I <sub>OH</sub> = −2 mA; see Figure 18	$V_{CCO}^{(1)} - 0.3$	3.2		V
$V_{OL}$	Low-level output voltage	I <sub>OL</sub> = 2 mA; see Figure 18		0.1	0.3	V
V <sub>IT+(IN)</sub>	Rising input threshold voltage			0.6 x V <sub>CCI</sub>	0.7 x V <sub>CCI</sub>	V
V <sub>IT-(IN)</sub>	Falling input threshold voltage		0.3 x V <sub>CCI</sub>	0.4 x V <sub>CCI</sub>		V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 × V <sub>CCI</sub>	0.2 x V <sub>CCI</sub>		V
I <sub>IH</sub>	High-level input current	V <sub>CCIIH</sub> = V <sup>(1)</sup> at INx			10	μА
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10			μА
CMTI	Common-mode transient immunity	$V_I = V_{CCI}$ or 0 V, $V_{CM} = 1200$ V; see Figure 20	85	100		kV/μs

<sup>(1)</sup>  $V_{CCI} = Input\text{-side } V_{CC}$ ;  $V_{CCO} = Output\text{-side } V_{CC}$ .



# 6.12 Supply Current Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT	
ISO7760								
	V <sub>I</sub> = V <sub>CC1</sub> (ISO7760);		I <sub>CC1</sub>		1.6	2.2		
Supply current - DC	$V_1 = 0 \text{ V (ISO7760 with F suffix)}$	$V_I = 0 \text{ V (ISO7760 with F suffix)}$			3	4.8	mA	
signal	V <sub>I</sub> = 0 V (ISO7760);	V <sub>I</sub> = 0 V (ISO7760);			8	11.4	ША	
	$V_I = V_{CC1}$ (ISO7760 with F suffix)		I <sub>CC2</sub>		3.3	5.3		
		1 Mbps	I <sub>CC1</sub>		4.9	6.6		
		1 Mbp3	I <sub>CC2</sub>		3.4	5.3		
Supply current - AC	All channels switching with square wave	10 Mbps	I <sub>CC1</sub>		5	6.7	mA	
signal	clock input; C <sub>L</sub> = 15 pF	TO MIDPS	I <sub>CC2</sub>		5.5	7.8	ША	
		100 Mbps	I <sub>CC1</sub>		6.3	8.2		
		100 Mbps	I <sub>CC2</sub>		26	33		
SO7761								
	$V_{I} = V_{CCI}^{(1)}$ (ISO7761);		I <sub>CC1</sub>		1.8	2.7		
Supply current - DC	$V_I = 0 \text{ V (ISO7761 with F suffix)}$		I <sub>CC2</sub>		2.9	4.7	mA	
signal	V <sub>I</sub> = 0 V (ISO7761);		I <sub>CC1</sub>		7.2	10.3	ША	
	$V_I = V_{CCI}$ (ISO7761 with F suffix)		I <sub>CC2</sub>		4.2	6.6		
		1 Mbps	I <sub>CC1</sub>		4.6	6.5		
Supply current - AC signal		1 MDPS	$I_{CC2}$		3.7	5.7		
	All channels switching with square wave	10 Mbps	I <sub>CC1</sub>		5.1	7	mA	
	clock input; C <sub>L</sub> = 15 pF	TO MIDPS	I <sub>CC2</sub>		5.5	7.8	IIIA	
	10	100 Mbps	I <sub>CC1</sub>		9.4	12		
		100 Mibbs	I <sub>CC2</sub>		22.8	29		
SO7762								
	V <sub>I</sub> = V <sub>CCI</sub> (ISO7762);	$V_1 = V_{CC1}$ (ISO7762):			2.1	3.2	^	
Supply current - DC	V <sub>I</sub> = 0 V (ISO7762 with F suffix)		I <sub>CC2</sub>		2.5	4.2		
signal	V <sub>I</sub> = 0 V (ISO7762);		I <sub>CC1</sub>		6.5	9.4	mA	
	$V_I = V_{CCI}$ (ISO7762 with F suffix)		I <sub>CC2</sub>		5	7.5		
		4 1 1 1 1 1 1	I <sub>CC1</sub>		4.4	6.2		
		1 Mbps	I <sub>CC2</sub>		3.9	5.8		
Supply current - AC	All channels switching with square wave	40 Mb = -	I <sub>CC1</sub>		5.2	7.1	A	
signal	clock input; C <sub>L</sub> = 15 pF	10 Mbps	I <sub>CC2</sub>		5.4	7.5	mA	
		400 Mb = =	I <sub>CC1</sub>	12	12.9	16.5		
100		100 Mbps	I <sub>CC2</sub>		19.5	25	1	
SO7763						·		
$V_{I} = V_{CCI} \text{ (ISO7763);}$ $V_{I} = 0 \text{ V (ISO7763 with F suffix)}$			I <sub>CC1</sub> , I <sub>CC2</sub>		2.4	3.7	A	
signal	$V_{I} = 0 \text{ V (ISO7763)};$ $V_{I} = V_{CCI} \text{ (ISO7763 with F suffix)}$		I <sub>CC1</sub> , I <sub>CC2</sub>		5.7	8.4	mA	
		1 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>		4.2	6.2		
Supply current - AC signal	All channels switching with square wave clock input; C <sub>1</sub> = 15 pF	10 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>		5.2	7.5	mA	
gnal	0.00K IIIput, OL = 10 pi	100 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>		16	20.5	ı	

<sup>(1)</sup>  $V_{CCI} = Input\text{-side } V_{CC}$ 



## 6.13 Electrical Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OH}$	High-level output voltage	I <sub>OH</sub> = −1 mA; see Figure 18	$V_{CCO}^{(1)} - 0.2$	2.45		V
$V_{OL}$	Low-level output voltage	I <sub>OL</sub> = 1 mA; see Figure 18		0.05	0.2	V
V <sub>IT+(IN)</sub>	Rising input threshold voltage			0.6 x V <sub>CCI</sub>	0.7 x V <sub>CCI</sub>	V
$V_{IT-(IN)}$	Falling input threshold voltage		0.3 x V <sub>CCI</sub>	0.4 x V <sub>CCI</sub>		V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 × V <sub>CCI</sub>	0.2 x V <sub>CCI</sub>		V
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = V <sub>CCI</sub> <sup>(1)</sup> at INx			10	μΑ
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10			μΑ
СМТІ	Common-mode transient immunity	V <sub>I</sub> = V <sub>CCI</sub> or 0 V, V <sub>CM</sub> = 1200 V; see Figure 20	85	100		kV/μs

<sup>(1)</sup>  $V_{CCI} = Input\text{-side } V_{CC}$ ;  $V_{CCO} = Output\text{-side } V_{CC}$ .



# 6.14 Supply Current Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN TYP	MAX	UNIT
ISO7760						
	V <sub>I</sub> = V <sub>CC1</sub> (ISO7760); V <sub>I</sub> = 0 V (ISO7760 with F suffix)		I <sub>CC1</sub>	1.6	2.2	
Supply current - DC	$V_I = 0 \text{ V (ISO7760 with F suffix)}$		I <sub>CC2</sub>	3	4.8	mA
signal	V <sub>I</sub> = 0 V (ISO7760);	V <sub>I</sub> = 0 V (ISO7760);		8	11.6	
	$V_I = V_{CC1}$ (ISO7760 with F suffix)		I <sub>CC2</sub>	3.3	5.3	
	1 Mbs		I <sub>CC1</sub>	4.9	6.8	
		1 Mbps	I <sub>CC2</sub>	3.4	5.3	
Supply current - AC	All channels switching with square wave	10 Mbps	I <sub>CC1</sub>	5	7	mA
signal	clock input; C <sub>L</sub> = 15 pF	10 Mbps	I <sub>CC2</sub>	4.9	7.2	ША
		100 Mbps	I <sub>CC1</sub>	6	8	
	Too wibps		I <sub>CC2</sub>	20.3	26	
ISO7761						
	$V_I = V_{CCI}^{(1)}$ (ISO7761); $V_I = 0 \text{ V (ISO7761 with F suffix)}$		I <sub>CC1</sub>	1.8	2.7	
Supply current - DC	$V_I = 0 \text{ V (ISO7761 with F suffix)}$		I <sub>CC2</sub>	2.9	4.6	mA
signal	V <sub>I</sub> = 0 V (ISO7761);		I <sub>CC1</sub>	7.2	10.3	IIIA
	$V_I = V_{CCI}$ (ISO7761 with F suffix)		I <sub>CC2</sub>	4.2	6.5	
Supply current - AC signal		1 Mbps	I <sub>CC1</sub>	4.6	6.7	
		1 IVIDPS	I <sub>CC2</sub>	3.7	5.8	
	All channels switching with square wave	10 Mbps	I <sub>CC1</sub>	4.9	7.1	mA
	clock input; C <sub>L</sub> = 15 pF	10 Mbps	I <sub>CC2</sub>	5	7.3	ША
	100 M	100 Mbps	I <sub>CC1</sub>	8.3	10.7	
		100 Mbps	I <sub>CC2</sub>	18.1	24	
ISO7762						
	$V_I = V_{CCI}$ (ISO7762);	$V_1 = V_{CCI}$ (ISO7762);		2.1	3.2	
Supply current - DC	$V_I = 0 \text{ V (ISO7762 with F suffix)}$		I <sub>CC2</sub>	2.6	4.1	mΛ
signal	V <sub>I</sub> = 0 V (ISO7762);		I <sub>CC1</sub>	6.5	9.6	mA
	$V_I = V_{CCI}$ (ISO7762 with F suffix)		I <sub>CC2</sub>	4.9	7.5	
		1 Mhno	I <sub>CC1</sub>	4.4	6.4	
		1 Mbps	I <sub>CC2</sub>	3.9	5.8	
Supply current - AC	All channels switching with square wave	10 Mbps	I <sub>CC1</sub>	5	7.1	mΛ
signal	clock input; C <sub>L</sub> = 15 pF	TO IVIDES	I <sub>CC2</sub>	5	7.1	mA
		100 Mbps	I <sub>CC1</sub>	10.9	14.1	
		100 Mbps	I <sub>CC2</sub>	15.6	20.1	
ISO7763						
Supply current - DC	$V_I = V_{CCI}$ (ISO7763); $V_I = 0$ V (ISO7763 with F suffix)		I <sub>CC1</sub> , I <sub>CC2</sub>	2.3	3.7	m ^
signal	$V_I = 0 \text{ V (ISO7763)};$ $V_I = V_{CCI} \text{ (ISO7763 with F suffix)}$		I <sub>CC1</sub> , I <sub>CC2</sub>	5.7	8.4	mA
		1 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>	4.1	6.1	
Supply current - AC signal	All channels switching with square wave clock input; C <sub>L</sub> = 15 pF	10 Mbps	I <sub>CC1</sub> , I <sub>CC2</sub>	4.9	7.1	mA
orginal	clock input; $C_L = 15 \text{ pF}$ $100 \text{ Mbps}$		I <sub>CC1</sub> , I <sub>CC2</sub>	13	17	

<sup>(1)</sup>  $V_{CCI} = Input\text{-side } V_{CC}$ 



## 6.15 Switching Characteristics—5-V Supply

V<sub>CC1</sub> = V<sub>CC2</sub> = 5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time	See Figure 40	6	11	16	ns
PWD	Pulse width distortion <sup>(1)</sup>  t <sub>PHL</sub> - t <sub>PLH</sub>	See Figure 18		0.4	4.9	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time (2)	Same-direction channels			4	ns
t <sub>sk(pp)</sub>	Part-to-part skew time <sup>(3)</sup>				4.5	ns
t <sub>r</sub>	Output signal rise time	See Figure 40		1.1	3.9	ns
t <sub>f</sub>	Output signal fall time	See Figure 18		1.4	3.9	ns
t <sub>DO</sub>	Default output delay time from input power loss	Measured from the time V <sub>CC</sub> goes below 1.7 V. See Figure 19		0.2	0.3	μS
t <sub>ie</sub>	Time interval error	2 <sup>16</sup> – 1 PRBS data at 100 Mbps		1.3		ns

<sup>(1)</sup> Also known as pulse skew.

## 6.16 Switching Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time	Con Figure 40	6	12	16	ns
PWD	Pulse width distortion (1)  t <sub>PHL</sub> - t <sub>PLH</sub>	See Figure 18		0.5	5	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time (2)	Same-direction channels			4.1	ns
t <sub>sk(pp)</sub>	Part-to-part skew time (3)				4.5	ns
t <sub>r</sub>	Output signal rise time	Con Figure 40		1	3	ns
t <sub>f</sub>	Output signal fall time	See Figure 18		1	3	ns
t <sub>DO</sub>	Default output delay time from input power loss	Measured from the time V <sub>CC</sub> goes below 1.7 V. See Figure 19		0.2	0.3	μS
t <sub>ie</sub>	Time interval error	2 <sup>16</sup> – 1 PRBS data at 100 Mbps		1.3		ns

<sup>(1)</sup> Also known as pulse skew.

<sup>(2)</sup> t<sub>sk(0)</sub> is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

<sup>(3)</sup> t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

<sup>(2)</sup> t<sub>sk(o)</sub> is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

<sup>(3)</sup> t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



# 6.17 Switching Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time	Coo Figure 10	7.5	13	18.5	ns
PWD	Pulse width distortion (1)  t <sub>PHL</sub> - t <sub>PLH</sub>	See Figure 18		0.6	5.1	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time (2)	Same-direction channels			4.1	ns
t <sub>sk(pp)</sub>	Part-to-part skew time (3)				4.6	ns
t <sub>r</sub>	Output signal rise time	Coo Figure 10		1	3.5	ns
t <sub>f</sub>	Output signal fall time	See Figure 18		1	3.5	ns
t <sub>DO</sub>	Default output delay time from input power loss	Measured from the time V <sub>CC</sub> goes below 1.7 V. See Figure 19		0.1	0.3	μS
t <sub>ie</sub>	Time interval error	2 <sup>16</sup> – 1 PRBS data at 100 Mbps		1.3		ns

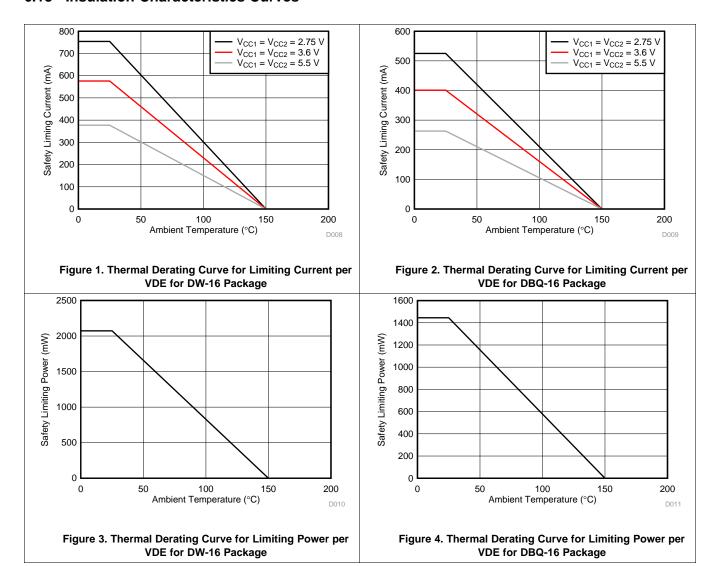
<sup>(1)</sup> Also known as pulse skew.

<sup>2)</sup> t<sub>sk(o)</sub> is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

<sup>(3)</sup> t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

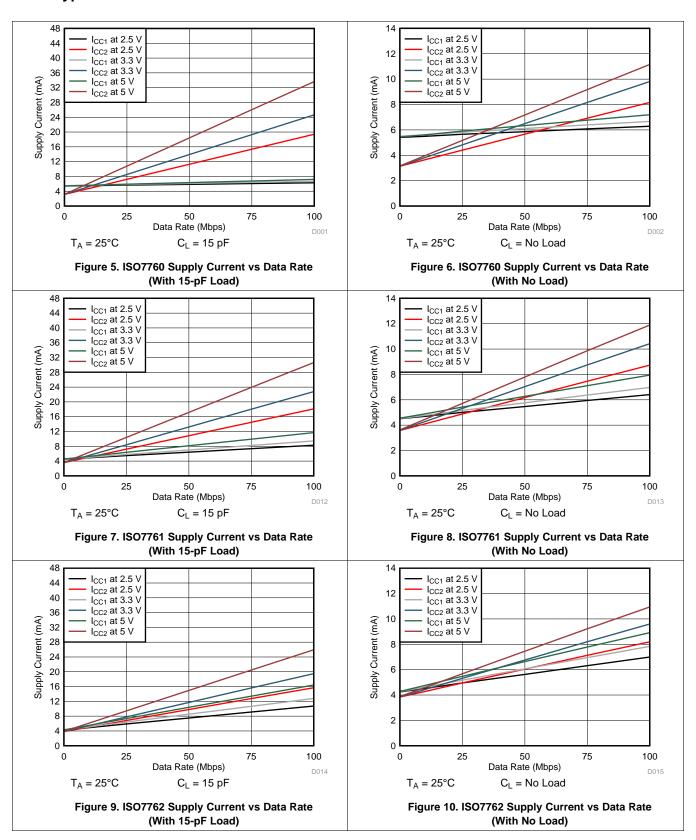


#### 6.18 Insulation Characteristics Curves



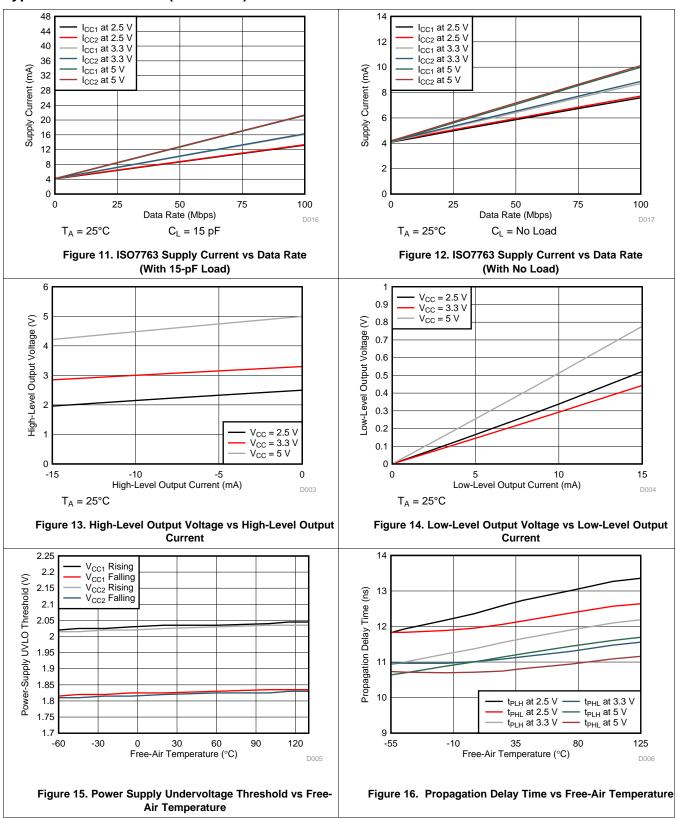


# 6.19 Typical Characteristics



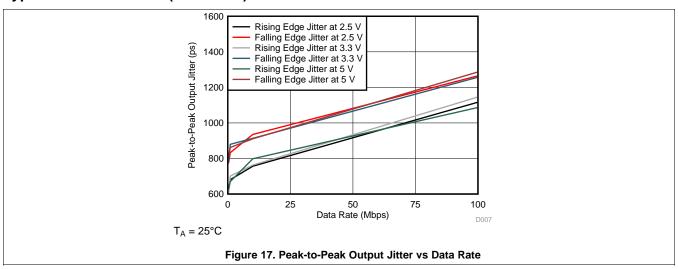


#### **Typical Characteristics (continued)**



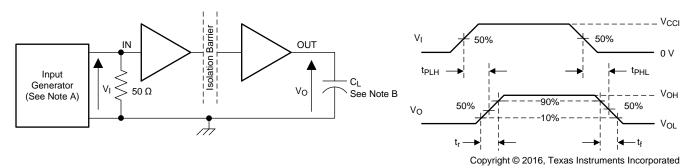


# **Typical Characteristics (continued)**



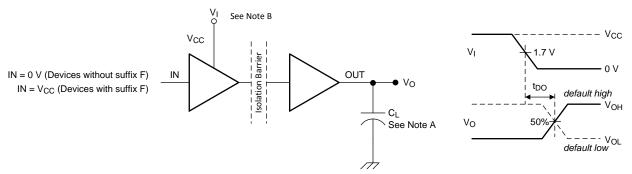
# TEXAS INSTRUMENTS

#### 7 Parameter Measurement Information



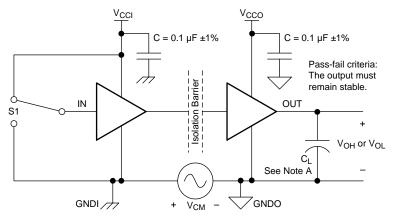
- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  50 kHz, 50% duty cycle,  $t_r \leq$  3 ns,  $t_f \leq$  3 ns,  $Z_O =$  50  $\Omega$ . At the input, a 50- $\Omega$  resistor is required to terminate Input Generator signal. It is not needed in actual application.
- B. C<sub>L</sub> = 15 pF and includes instrumentation and fixture capacitance within ±20%.

Figure 18. Switching Characteristics Test Circuit and Voltage Waveforms



- A.  $C_L = 15 \text{ pF}$  and includes instrumentation and fixture capacitance within  $\pm 20\%$ .
- B. Power-supply ramp rate = 10 mV/ns

Figure 19. Default Output Delay Time Test Circuit and Voltage Waveforms



A.  $C_L = 15 \text{ pF}$  and includes instrumentation and fixture capacitance within  $\pm 20\%$ .

Figure 20. Common-Mode Transient Immunity Test Circuit

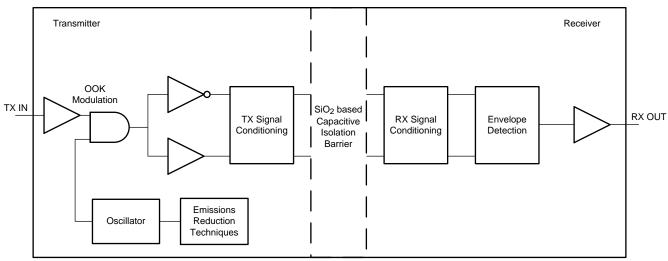


#### 8 Detailed Description

#### 8.1 Overview

The ISO776x family of devices uses an ON-OFF keying (OOK) modulation scheme to transmit the digital data across a silicon-dioxide based isolation barrier. The transmitter sends a high-frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output through a buffer stage. The ISO776x family of devices also incorporates advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions because of the high-frequency carrier and IO buffer switching. The conceptual block diagram of a digital capacitive isolator, Figure 21, shows a functional block diagram of a typical channel. Figure 22 shows a conceptual detail of how the ON-OFF keying scheme works.

#### 8.2 Functional Block Diagram



Copyright © 2016, Texas Instruments Incorporated

Figure 21. Conceptual Block Diagram of a Digital Capacitive Isolator

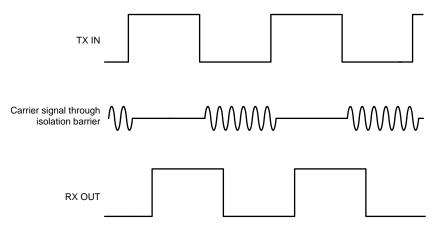


Figure 22. ON-OFF Keying (OOK) Based Modulation Scheme



#### 8.3 Feature Description

Table 1 lists the device features.

**Table 1. Device Features** 

PART NUMBER	CHANNEL DIRECTION	MAXIMUM DATA RATE	DEFAULT OUTPUT	PACKAGE	RATED ISOLATION <sup>(1)</sup>
ISO7760	6 Forward,	100 Mbps	High	DW-16	5000 V <sub>RMS</sub> / 8000 V <sub>PK</sub>
1307700	0 Reverse		DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>	
ISO7760 with F suffix	6 Forward,	100 Mbps	Low	DW-16	5000 V <sub>RMS</sub> / 8000 V <sub>PK</sub>
1307760 WILLI F SUIIX	0 Reverse	Too wibps	LOW	DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>
ISO7761	5 Forward,	100 Mbps	100 Mbps High -	DW-16	5000 V <sub>RMS</sub> / 7071 V <sub>PK</sub>
1307761	1 Reverse	Too wibps		DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>
ISO7761 with F suffix	5 Forward,		s Low	DW-16	5000 V <sub>RMS</sub> / 7071 V <sub>PK</sub>
1307761 WILLI F SUIIX	1 Reverse			DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>
ISO7762	4 Forward,	100 Mbps	High	DW-16	5000 V <sub>RMS</sub> / 7071 V <sub>PK</sub>
1507762	2 Reverse	Too wibps	High	DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>
ISO7762 with F suffix	4 Forward,	100 Mbps	Low	DW-16	5000 V <sub>RMS</sub> / 7071 V <sub>PK</sub>
1307762 WILLI F SULLX	2 Reverse	Too wibps	LOW	DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>
ISO7763	3 Forward,	100 Mbps	High	DW-16	5000 V <sub>RMS</sub> / 7071 V <sub>PK</sub>
1301103	3 Reverse	squivi oor	High	DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>
ISO7763 with F suffix	3 Forward,	100 Mhna	Law	DW-16	5000 V <sub>RMS</sub> / 7071 V <sub>PK</sub>
ISO1103 WILLI F SULLX	3 Reverse	100 Mbps	Low	DBQ-16	3000 V <sub>RMS</sub> / 4242 V <sub>PK</sub>

<sup>(1)</sup> See Safety-Related Certifications for detailed isolation ratings.

## 8.3.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 22. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO776x family of devices incorporates many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection for input and output signal pins and inter-chip bond pads.
- Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- Reduced common mode currents across the isolation barrier by ensuring purely differential internal operation.



#### 8.4 Device Functional Modes

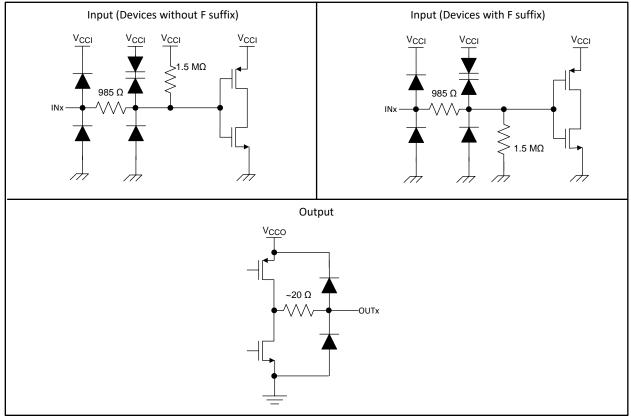
Table 2 lists the functional modes for the ISO776x.

Table 2. Function Table<sup>(1)</sup>

V <sub>CCI</sub>	V <sub>cco</sub>	INPUT (INx) <sup>(2)</sup>	OUTPUT (OUTx)	COMMENTS	
PU	PU	Н	Н	Normal Operation: A channel output assumes the logic state of the input.	
		L	L		
		Open	Default	Default mode: When INx is open, the corresponding channel output goes to its default logic state. Default is <i>High</i> for ISO776x and <i>Low</i> for ISO776x with F suffix.	
PD	PU	x	Default	Default mode: When $V_{\text{CCI}}$ is unpowered, a channel output assumes the logic state based on the selected default option. Default is $\textit{High}$ for ISO776x and $\textit{Low}$ for ISO776x with F suffix. When $V_{\text{CCI}}$ transitions from unpowered to powered-up, a channel output assumes the logic state of its input. When $V_{\text{CCI}}$ transitions from powered-up to unpowered, channel output assumes the selected default state.	
х	PD	Х	Undetermined	When V <sub>CCO</sub> is unpowered, a channel output is undetermined <sup>(3)</sup> . When V <sub>CCO</sub> transitions from unpowered to powered-up, a channel output assumes the logic state of the input	

- (1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ ; PU = Powered up ( $V_{CC} \ge 2.25$  V); PD = Powered down ( $V_{CC} \le 1.7$  V); X = Irrelevant; H = High level; L = Low level
- (2) A strongly driven input signal can weakly power the floating V<sub>CC</sub> via an internal protection diode and cause undetermined output.
- 3) The outputs are in undetermined state when 1.7 V < V<sub>CCI</sub>, V<sub>CCO</sub> < 2.25 V.

#### 8.4.1 Device I/O Schematics



Copyright © 2016, Texas Instruments Incorporated

Figure 23. Device I/O Schematics



## 9 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

#### 9.1 Application Information

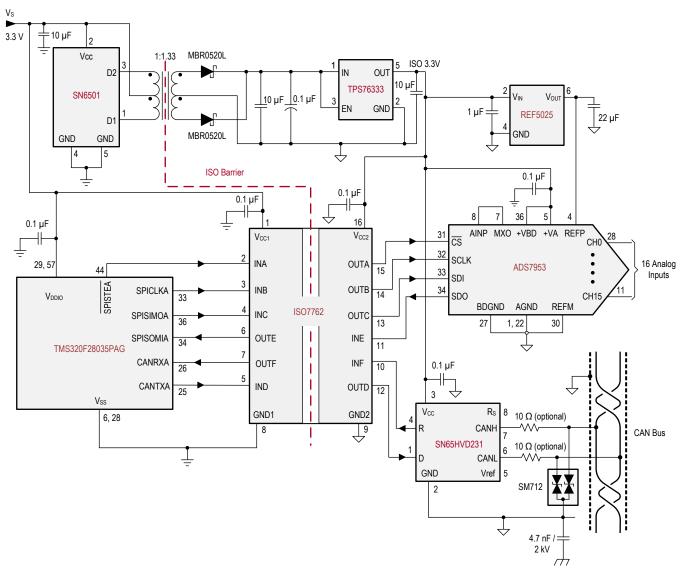
The ISO776x family of devices is a high-performance, six-channel digital isolators. The ISO776x family of devices uses single-ended CMOS-logic switching technology. The voltage range is from 2.25 V to 5.5 V for both supplies,  $V_{CC1}$  and  $V_{CC2}$ . When designing with digital isolators, keep in mind that because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is,  $\mu$ C or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

#### 9.2 Typical Application

Figure 24 shows the isolated serial-peripheral interface (SPI) and controller-area network (CAN) interface implementation.



## **Typical Application (continued)**



NOTE: Multiple pins and discrete components omitted for clarity purpose.

Figure 24. Isolated SPI and CAN Interface

#### 9.2.1 Design Requirements

For this design example, use the parameters listed in Table 3.

**Table 3. Design Parameters** 

PARAMETER	VALUE	
Supply voltage, $V_{CC1}$ and $V_{CC2}$	2.25 to 5.5 V	
Decoupling capacitor between V <sub>CC1</sub> and GND1	0.1 μF	
Decoupling capacitor from V <sub>CC2</sub> and GND2	0.1 μF	



#### 9.2.2 Detailed Design Procedure

Unlike optocouplers, which require external components to improve performance, provide bias, or limit current, the ISO776x family of devices only requires two external bypass capacitors to operate.

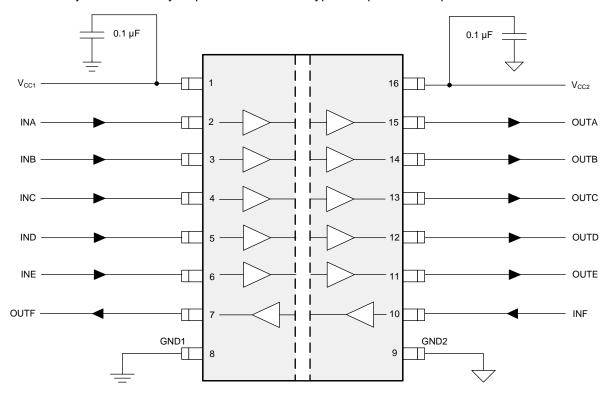
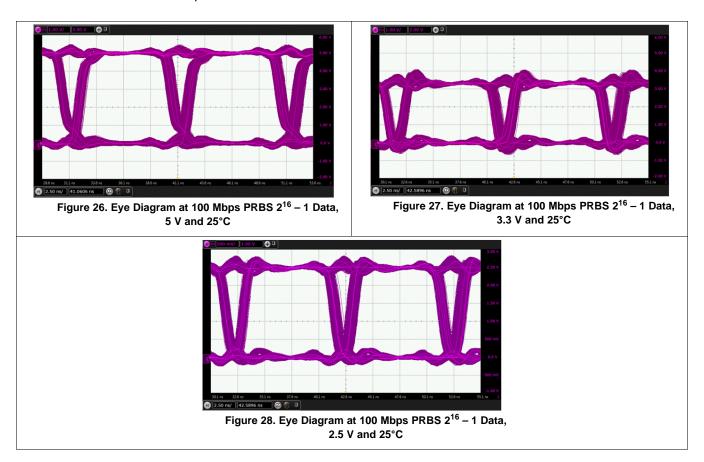


Figure 25. Typical ISO7761 Circuit Hook-up



#### 9.2.3 Application Curves

The typical eye diagram of the ISO776x family of devices indicates low jitter and a wide open eye at the maximum data rate of 100 Mbps.



#### 9.2.3.1 Insulation Lifetime

Insulation lifetime projection data is collected by using industry-standard Time Dependent Dielectric Breakdown (TDDB) test method. In this test, all pins on each side of the barrier are tied together creating a two-terminal device and high voltage applied between the two sides; See Figure 29 for TDDB test setup. The insulation breakdown data is collected at various high voltages switching at 60 Hz over temperature. For reinforced insulation, VDE standard requires the use of TDDB projection line with failure rate of less than 1 part per million (ppm). Even though the expected minimum insulation lifetime is 20 years at the specified working isolation voltage, VDE reinforced certification requires additional safety margin of 20% for working voltage and 87.5% for lifetime which translates into minimum required insulation lifetime of 37.5 years at a working voltage that's 20% higher than the specified value.

Figure 30 shows the intrinsic capability of the isolation barrier to withstand high voltage stress over its lifetime. Based on the TDDB data, the intrinsic capability of the insulation is 1500  $V_{RMS}$  with a lifetime of 135 years. Other factors, such as package size, pollution degree, material group, etc. can further limit the working voltage of the component. The working voltage of DW-16 package is specified upto 1500  $V_{RMS}$  and DBQ-16 package up to 400  $V_{RMS}$ . At the lower working voltages, the corresponding insulation lifetime is much longer than 135 years.



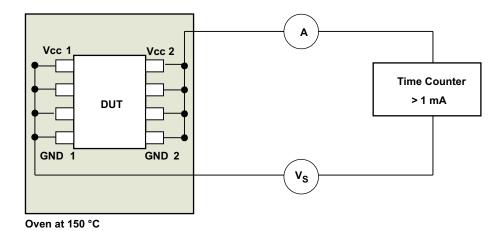


Figure 29. Test Setup for Insulation Lifetime Measurement

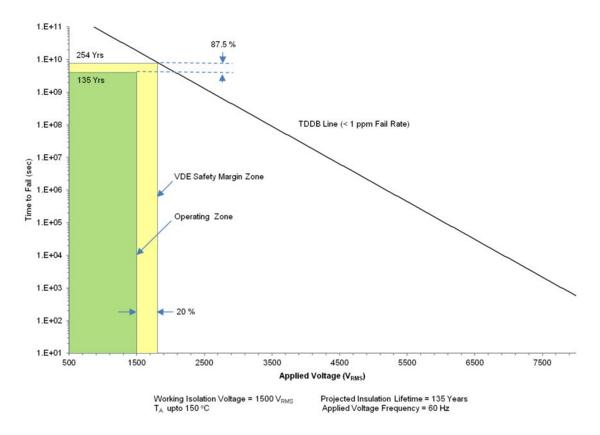


Figure 30. Insulation Lifetime Projection Data

## 10 Power Supply Recommendations

To help ensure reliable operation at data rates and supply voltages, a 0.1- $\mu$ F bypass capacitor is recommended at input and output supply pins ( $V_{CC1}$  and  $V_{CC2}$ ). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments' SN6501 or SN6505. For such applications, detailed power supply design and transformer selection recommendations are available in the SN6501 Transformer Driver for Isolated Power Supplies data sheet or the SN6505 Low-Noise 1-A Transformer Drivers for Isolated Power Supplies data sheet.



## 11 Layout

#### 11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see Figure 31). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/inch<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links
  usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, see the Digital Isolator Design Guide application report.

#### 11.1.1 PCB Material

For digital circuit boards operating at less than 150 Mbps, (or rise and fall times greater than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit board. This PCB is preferred over cheaper alternatives because of lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and the self-extinguishing flammability-characteristics.

#### 11.2 Layout Example

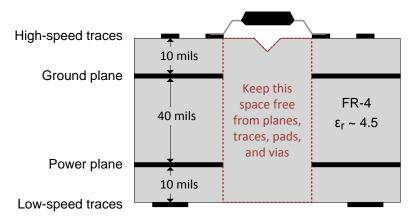


Figure 31. Layout Example Schematic



## 12 Device and Documentation Support

## 12.1 Documentation Support

#### 12.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, Digital Isolator Design Guide application report
- Texas Instruments, How to use isolation to improve ESD, EFT and Surge immunity in industrial systems
  application report
- Texas Instruments, Isolation Glossary
- Texas Instruments, TMS320F2803x Piccolo™ Microcontrollers data sheet
- Texas Instruments, ADS79xx 12/10/8-Bit, 1 MSPS, 16/12/8/4-Channel, Single-Ended, MicroPower, Serial Interface ADCs data sheet
- Texas Instruments, REF50xx Low-Noise, Very Low Drift, Precision Voltage Reference data sheet
- Texas Instruments, SN6501 Transformer Driver for Isolated Power Supplies data sheet
- Texas Instruments, SN65HVD23x 3.3-V CAN Bus Transceivers data sheet
- Texas Instruments, TPS76333 Low-Power 150-mA Low-Dropout Linear Regulators data sheet

#### 12.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
ISO7760	Click here	Click here	Click here	Click here	Click here
ISO7761	Click here	Click here	Click here	Click here	Click here
ISO7762	Click here	Click here	Click here	Click here	Click here
ISO7763	Click here	Click here	Click here	Click here	Click here

## 12.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 12.4 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community T's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 12.5 Trademarks

Piccolo, E2E are trademarks of Texas Instruments.

All other trademarks are the property of their respective owners.



#### 12.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## 12.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

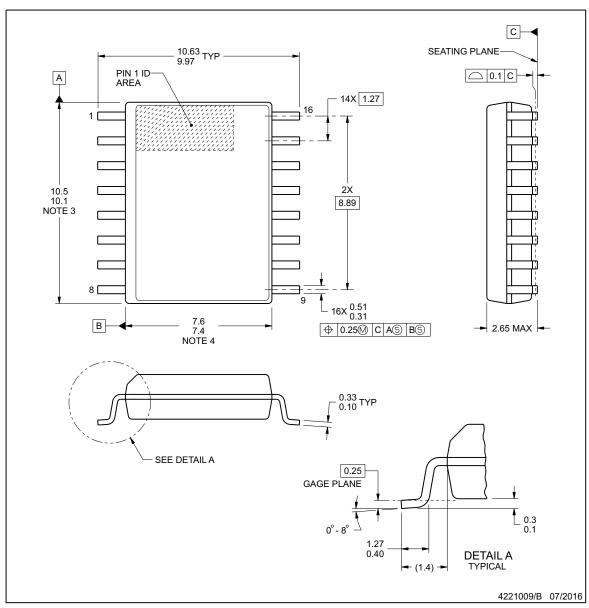
**DW0016B** 





## **PACKAGE OUTLINE**

## SOIC - 2.65 mm max height



#### NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.

www.ti.com

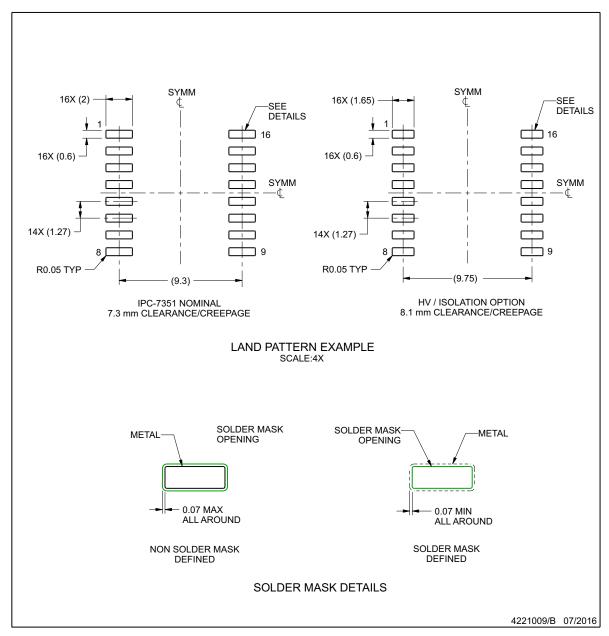


## **EXAMPLE BOARD LAYOUT**

# **DW0016B**

## SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

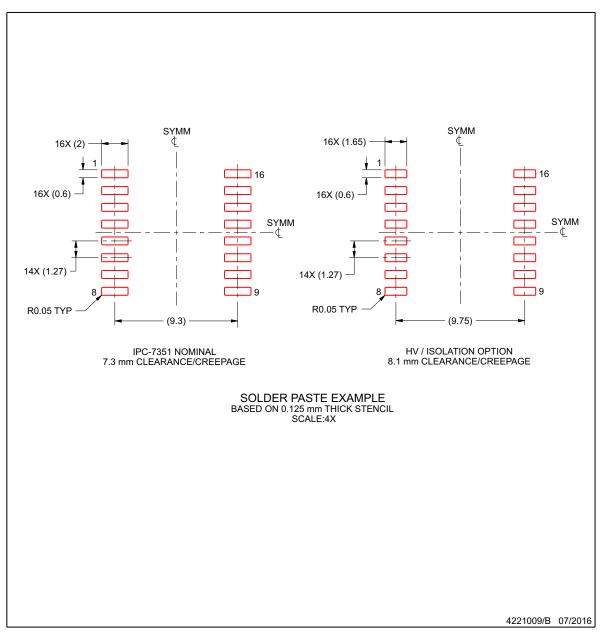


### **EXAMPLE STENCIL DESIGN**

# **DW0016B**

## SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

- Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- Board assembly site may have different recommendations for stencil design.

www.ti.com





5-Feb-2019

## **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
ISO7760DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7760	Samples
ISO7760DBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7760	Samples
ISO7760DW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7760	Samples
ISO7760DWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7760	Samples
ISO7760FDBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7760F	Samples
ISO7760FDBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7760F	Samples
ISO7760FDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7760F	Samples
ISO7760FDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7760F	Samples
ISO7761DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7761	Samples
ISO7761DBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7761	Samples
ISO7761DW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7761	Samples
ISO7761DWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7761	Samples
ISO7761FDBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7761F	Samples
ISO7761FDBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7761F	Samples
ISO7761FDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7761F	Samples
ISO7761FDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7761F	Samples
ISO7762DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7762	Samples





www.ti.com

5-Feb-2019

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Sample
ISO7762DBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7762	Sample
ISO7762DW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7762	Sample
ISO7762DWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7762	Sample
ISO7762FDBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7762F	Sample
ISO7762FDBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7762F	Sample
ISO7762FDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7762F	Sample
ISO7762FDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7762F	Sampl
ISO7763DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7763	Sampl
ISO7763DBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7763	Sampl
ISO7763DW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7763	Sampl
ISO7763DWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7763	Sampl
ISO7763FDBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7763F	Sampl
ISO7763FDBQR	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7763F	Sampl
ISO7763FDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7763F	Sampl
ISO7763FDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7763F	Sampl

<sup>(1)</sup> The marketing status values are defined as follows: **ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.





5-Feb-2019

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

#### OTHER QUALIFIED VERSIONS OF ISO7760, ISO7761, ISO7762, ISO7763:

Automotive: ISO7760-Q1, ISO7761-Q1, ISO7762-Q1, ISO7763-Q1

NOTE: Qualified Version Definitions:

Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

www.ti.com 26-Feb-2019

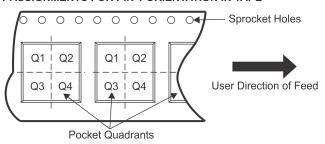
# TAPE AND REEL INFORMATION



# TAPE DIMENSIONS KO P1 BO W Cavity A0

A0	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO7760DBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7760DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7760FDBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7760FDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7761DBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7761DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7761FDBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7761FDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7762DBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7762DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7762FDBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7762FDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7763DBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7763DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO7763FDBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO7763FDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

www.ti.com 26-Feb-2019



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO7760DBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7760DWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7760FDBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7760FDWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7761DBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7761DWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7761FDBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7761FDWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7762DBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7762DWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7762FDBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7762FDWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7763DBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7763DWR	SOIC	DW	16	2000	350.0	350.0	43.0
ISO7763FDBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0
ISO7763FDWR	SOIC	DW	16	2000	535.4	167.6	48.3

7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





SOIC



## NOTES:

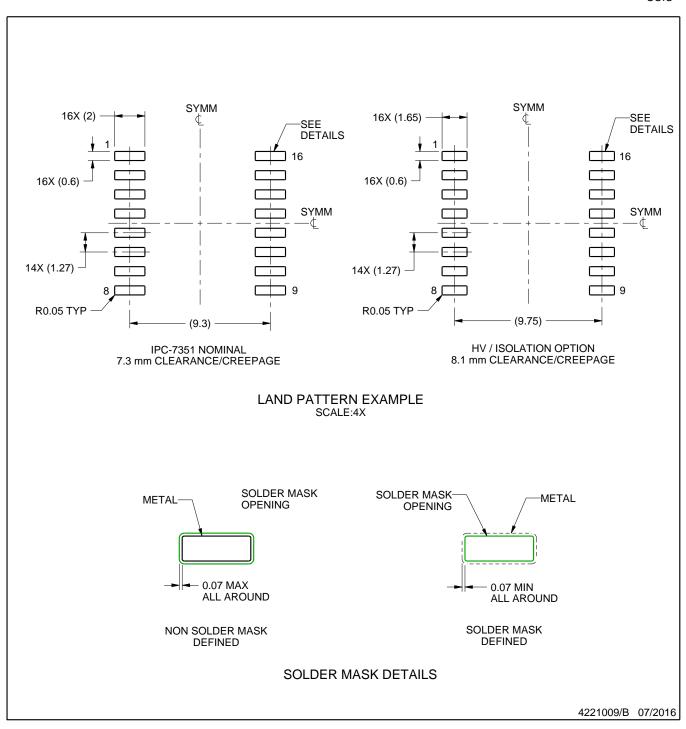
- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.



SOIC



## NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOIC



### NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.





Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.







## NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 inch, per side.
- 4. This dimension does not include interlead flash.5. Reference JEDEC registration MO-137, variation AB.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



#### IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale (<a href="www.ti.com/legal/termsofsale.html">www.ti.com/legal/termsofsale.html</a>) or other applicable terms available either on ti.com or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265 Copyright © 2019, Texas Instruments Incorporated