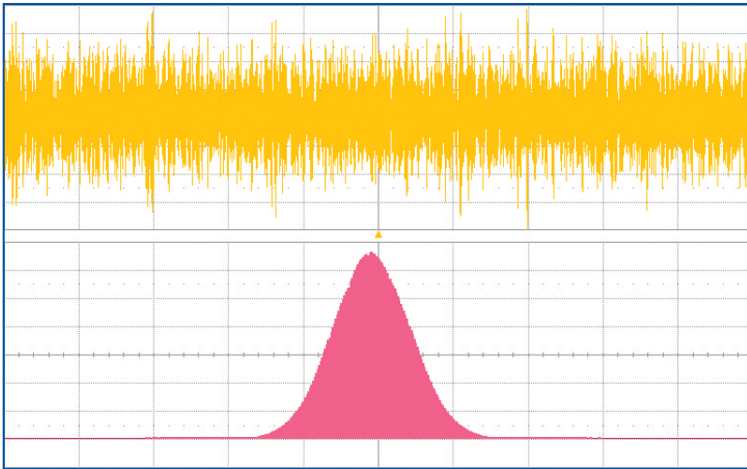


# **Noise Terminology:**

## **An Overview of Noise Terminology and Applications**

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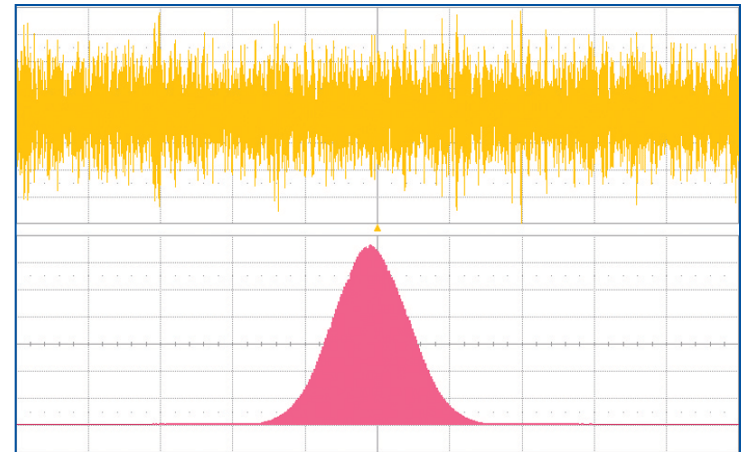
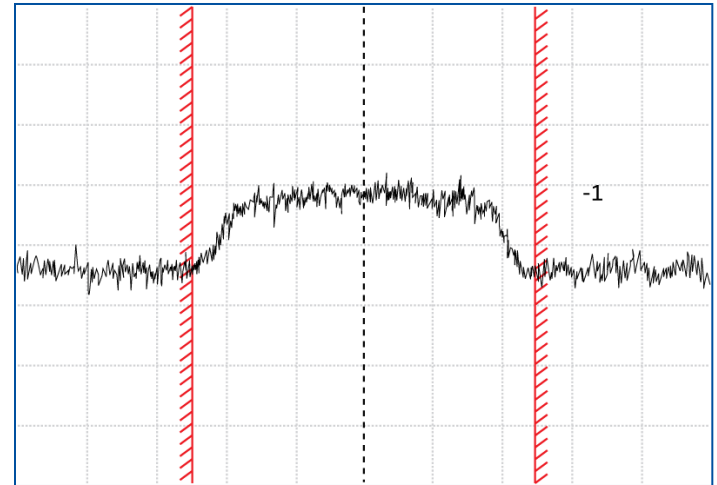
# Today's Webinar

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- Important noise characteristics
- Technologies effected by noise
- Noise applications and terms for this webinar
  - $N_p$  – noise power
  - NF – noise figure
  - $E_b/N_o$  - the ratio of bit energy to noise density
  - NPR - noise power ratio
- Example calculations

# Important Noise Characteristics

- Frequency Domain
  - Flat spectrum with uniform power spectral density
  - White noise contains all frequencies for a given BW
- Time Domain
  - Amplitude has a Gaussian distribution
  - Signal should have high crest factor (Pk/avg)



# Technologies Affected by Noise

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## Wireless Communications

- Satellite links
- Mobile devices
- HDTV services

## Radio Astronomy

- Reference source
- Calibration tool

## Military & Commercial Radar Systems

- Calibration tool
- Reliability testing

# Noise Terms & Applications

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- $N_p$  – Noise power
- NF – noise figure
- $E_b/N_o$  - the ratio of bit energy to noise density
- NPR - noise power ratio

# Calculated Noise Power I

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- Noise power,  $N_p = kTB$ 
  - $K$  = Boltzmann's constant
  - $T$  = temperature in Kelvin
  - $B$  = noise bandwidth
  - Relationship between noise temperature and power
- An Example calculation at  $290^\circ \text{ K}$  ( $62.6^\circ \text{ F}$ )( $17^\circ \text{ C}$ )
  - The average temperature of the surface of the Earth
  - $N_p = (1.38065\text{e-}23) * (290^\circ \text{ K}) * (1 \text{ Hz})$
  - $N_p = 4.004\text{e-}21 \text{ W/Hz}$
  - The BW selected depends on what you are calculating
  - Noise figure or receiver sensitivity
  - Watts per Hertz is not a common measure for communication engineers

# Calculated Noise Power II

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- More practical if normalized to 1mW in a 50Ω system
- $N_p \text{ (dBm/Hz)} = 10\log(4.004\text{e-}21/.001\text{W}) = -174 \text{ dBm/Hz}$
- Using this equation we can now relate the  $N_p$  to a 50 Ω test system at 1 mW
- This is common standard for communication test equipment
- Therefore: -174 dBm/Hz is the output power of a 50 Ω resistor at the average temperature of the earth in a 1 Hz bandwidth
- This value cannot be measured with conventional communication equipment and is only used for comparison

# Calculated Noise Power III

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- -174 dBm/Hz, or thermal noise is used as the ultimate noise floor
- The value of -174 dBm/Hz is sometimes expressed as excess noise ratio, or ENR
- Simple equation to calculate ENR
  - $\text{ENR (dB)} = 174 \text{ dBm / Hz} - \text{PSD (dBm / Hz)}$
- These values are used when measuring the noise figure of an amplifier
- The noise figure of an amplifier is necessary when calculating gain for a receiver system



# Noise Terms & Applications

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- $N_p$  – noise power
- NF – noise figure
- $E_b/N_o$  - the ratio of bit energy to noise density
- NPR - noise power ratio

# Example noise figure calculation

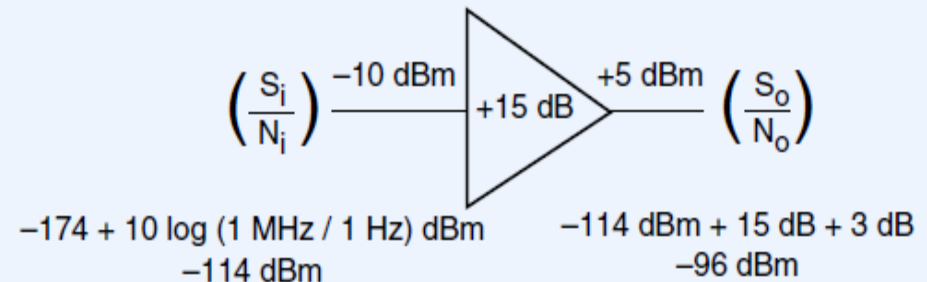
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- Noise Factor (linear) and Noise Figure (dB)
- Noise Power vs. Noise Temperature
- Scalar Noise Figure Measurement I
- Scalar Noise Figure Measurement II

# Noise Factor and Figure Equations

- The difference between SNR & SNR out excluding gain is noise figure
- Linear F is noise factor
- $10 \log(F)$  is Noise Figure
- Example amplifier with 3 dB noise figure
- Noise Figure is typically unknown and must be measured
- Noise figure is measured with
  - Noise Figure Analyzer
  - Spectrum Analyzer
  - Dedicated receiver

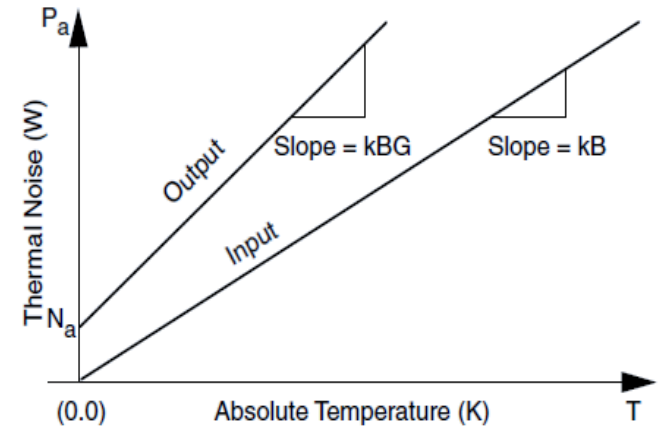
Amplifier Gain = 15 dB  
 Amplifier Noise Figure = 3 dB  
 Input Signal Power = -10 dBm  
 Transmission Bandwidth = 1 MHz



	Noise Figure Calculations in Power (dBm)
Terms	$F \text{ (dB)} = \left(\frac{S_i}{N_i}\right)_{\text{dB}} - \left(\frac{S_o}{N_o}\right)_{\text{dB}}$
$S_i / N_i$	$-10 \text{ dBm} - (-114 \text{ dBm})$ 104 dB
$S_o / N_o$	$+5 \text{ dBm} - (-96 \text{ dBm})$ 101 dB
Noise Figure	$104 \text{ dB} - 101 \text{ dB}$ 3 dB

# Noise Power vs. Noise Temperature

- The relationship between noise power and temperature is linear
- Graph of SNR without any signal power and zero return loss
- The slope  $m = kB$ ; when  $b = 0^\circ \text{ K}$

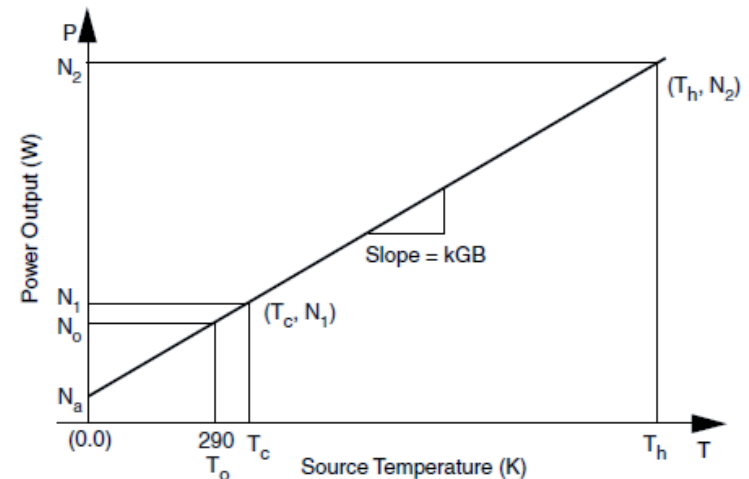


- Real amplifiers have gain and work above absolute zero
- The slope  $m = kGB$  above  $0^\circ \text{ K}$
- $N_a$  is intrinsic thermal noise generated by the amplifier
- $b$  or the y-intercept ( $N_a$ ) is the noise figure of the amplifier
- The slope, or  $kGB$  is equal to the gain for a finite BW
- This is referred to as the "Y"- factor method

# Scalar Noise Figure Measurement I

- The example uses a 15 dB ENR noise source
- 15dB ENR =  $T_{\text{hot}} = 9461^{\circ} \text{ K}$
- Assume  $T_{\text{cold}} = T_{\text{off}} = 290^{\circ} \text{ K}$ 
  - Simplify the equation
- $F (\text{dB}) = \text{ENR}(\text{dB}) - 10\log(Y-1)$
- $F (\text{dB}) = 15 - 10\log[(N_2/N_1) - 1]$
- $F \text{ dB} = 15\text{dB} - 12.91\text{dB}$
- $F \text{ dB} = 2.1\text{dB}$

Conversion	Equation
$T_h$ to ENR	$\text{ENR} = 10 \log (T_h - T_o)/T_o$
ENR to $T_h$	$T_h = T_o 10^{(\text{ENR}/10)} + T_o$



# Scalar Noise Figure Measurement II

Conversion	Equation
$T_h$ to ENR	$ENR = 10 \log (T_h - T_o)/T_o$
ENR to $T_h$	$T_h = T_o 10^{(ENR/10)} + T_o$

Term	Measured Noise Power
$N_2$	$8.51 \times 10^{-15}$ Watt
$N_1$	$4.14 \times 10^{-16}$ Watt

- If the  $T_c$  value does not equal  $T_o$ , or  $290^\circ$  K then must compute F using this method
- Assuming  $T_{cold} = T_{off}$  will be dependent on the estimated noise figure
- The larger the estimated value, the less critical the  $T_{off}$  value

$$F = \frac{\left(\frac{T_h}{T_o} - 1\right) - Y \left(\frac{T_c}{T_o} - 1\right)}{Y - 1}$$

$$F = \frac{(31.62) - 20.56 (0)}{20.56 - 1}$$

$$F = 1.62$$

$$F = 10 \log (1.62) = 2.1 \text{ dB}$$

$$Y = (N_2 / N_1)$$

$$Y = 8.51 \times 10^{-15} / 4.14 \times 10^{-16}$$

$$Y = 20.56$$

# Noise Terms & Applications

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- NP – noise power
- NF – noise figure
- $E_b/N_0$  - the ratio of bit energy to noise density
- NPR - noise power ratio

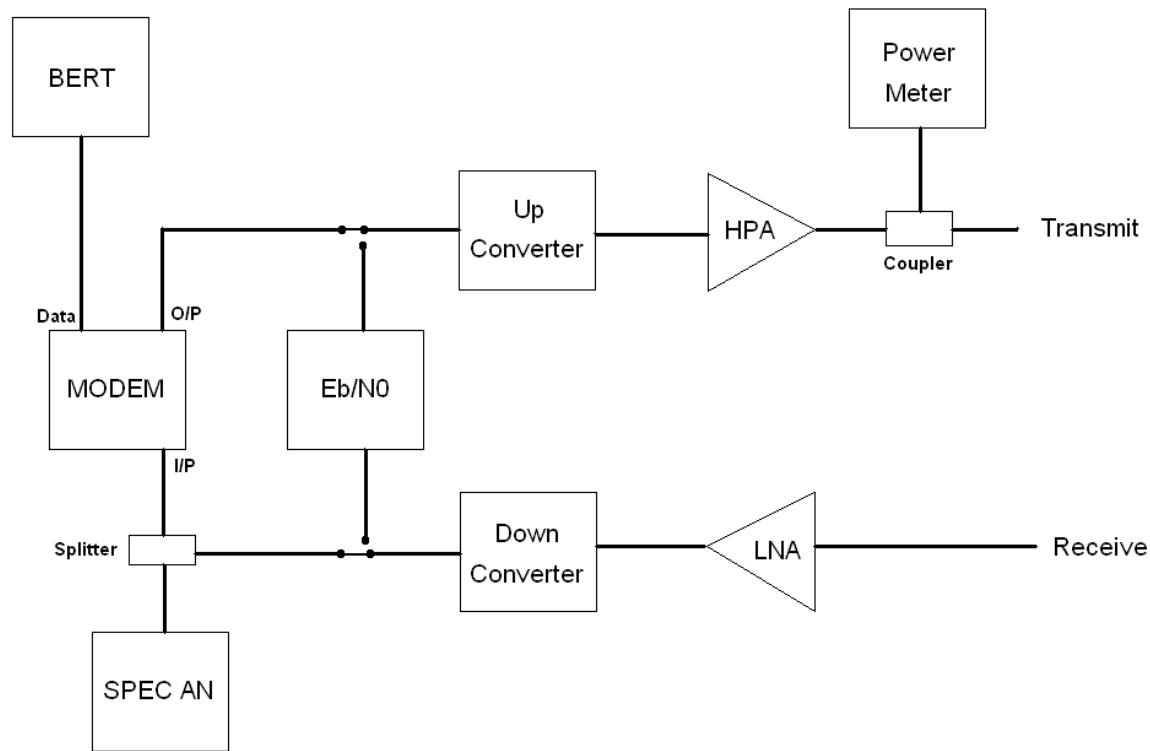
# Example SLE Calculation

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- Satellite Link emulation block diagram
- Waterfall diagram explanation
  - Example graph of different modulation schemes
- Calculate the amplifier noise floor
- Convert  $E_b/N_0$  value to C/N
- Calculate the required receiver sensitivity
  - The calculation does not include FEC, multi-path, or fading effects for simplicity

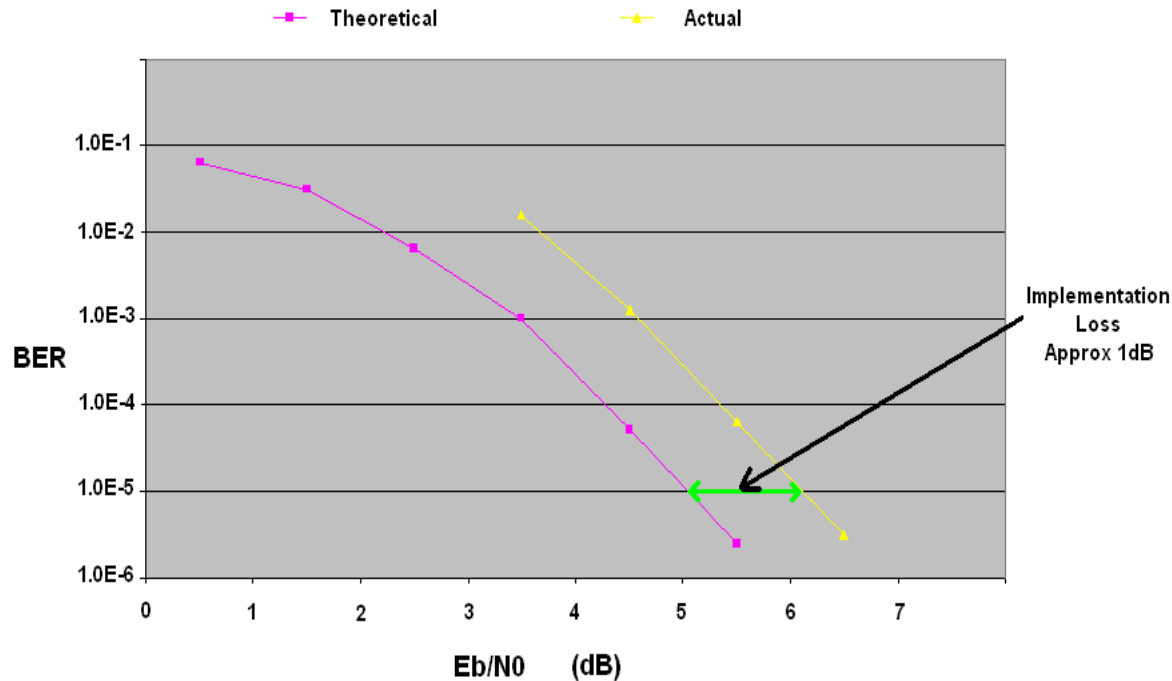


# Satellite Link Emulation (SLE) Diagram



- Base band SNR measurement block diagram
- $E_b/N_0$  can be measured on the up, or down link
- This example shows  $E_b/N_0$  at base band

# Waterfall Diagram



- Waterfall diagram compares  $E_b/N_0$  (SNR) to BER
- Used to measure implementation losses
- Important for modulation comparisons

# Amplifier Noise Floor

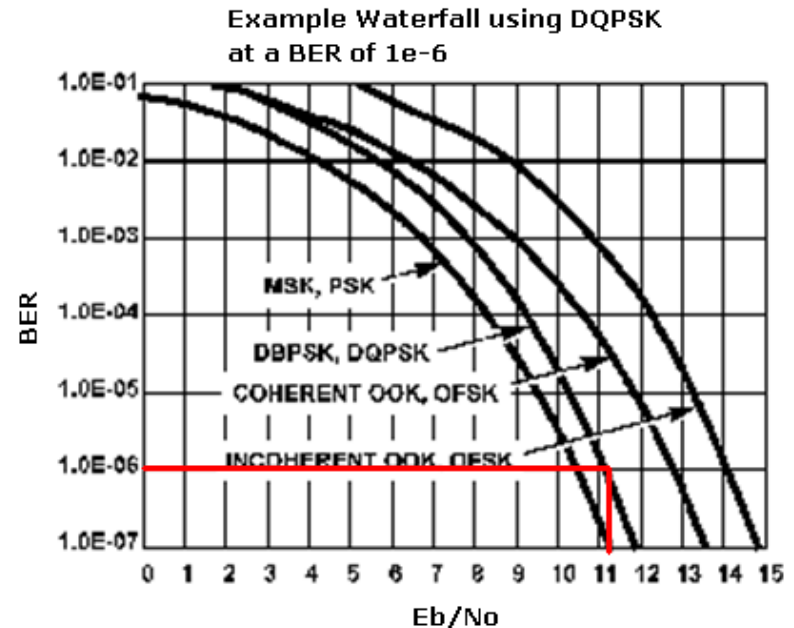
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- Theoretical amplifier noise floor
- An amplifier with 1MHz BW at 290° K (17° C)
- Noise power =  $kTB$
- $= 1.38 \times 10^{-23} \text{ J/K} \times 290\text{K} \times 1\text{e6 Hz s}^{-1}$
- $= 4 \times 10^{-15}\text{W}$
- $= -114\text{dBm}$  per 1 MHz
- NF of the amplifier in this example is 15dB
- $= -114\text{dBm} + 15\text{dB}$
- Real Receiver Noise Floor  $\approx -99\text{dBm}$ 
  - Will be slightly higher at room temperature 25° C

# Convert Eb/No to SNR (C/N)

$$C/N = (E_b/N_o) * (B_R/B_T)$$

- $E_b$  = Energy per bit
- $N_o$  = noise density (per 1Hz)
- $B_R$  = system data rate
- $B_T$  = system bandwidth



# Receiver Sensitivity

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- Or, the required signal strength at the receiver input above the noise level
- The waterfall diagram from the previous slide requires an 11.1dB Eb/No for a BER of 10<sup>-6</sup> using DQPSK modulation

$$C/N = (E_b/N_o) * (B_R/B_T)$$

- $C/N = 11.1 * 10\log(2\text{Mbps} / 1\text{MHz})$
- $C/N = 11.1\text{dB} + 3\text{dB} = 14.1 \text{ dB}$

Receiver sensitivity (Prs)

- $Prs = \text{Receiver Noise Floor} + C/N \text{ (SNR)}$
- $= -99\text{dBm} + 14.1\text{dB}$
- $= -84.9 \text{ dBm}$

# Noise Terms & Applications

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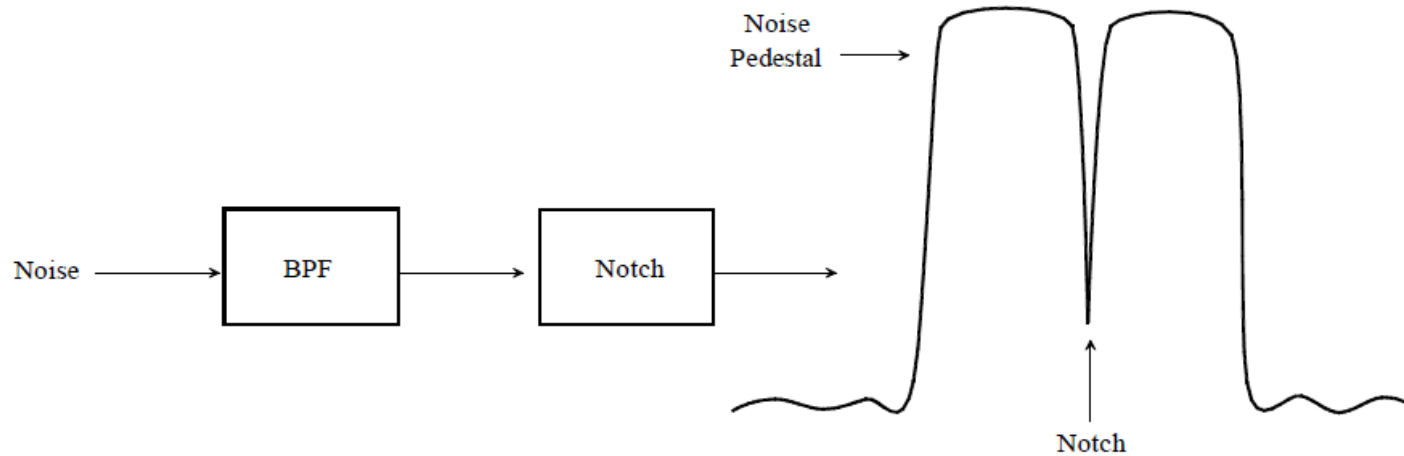
- NP – noise power
- NF – noise figure
- $E_b/N_o$  - the ratio of bit energy to noise density
- NPR - noise power ratio

# NPR Noise Power Ratio Basics

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- White noise is used to simulate the presence of multiple carriers with random amplitude and phase.
- A finite BW notch is removed from the noise pedestal
- The notch is created using a band stop filter
- The power measurements are taken from inside the notch using a narrow band receiver
- The input signal power level is increased until the DUT is saturated
- The change in NPR is caused by spectral re-growth in the notch due to nonlinearities in the DUT

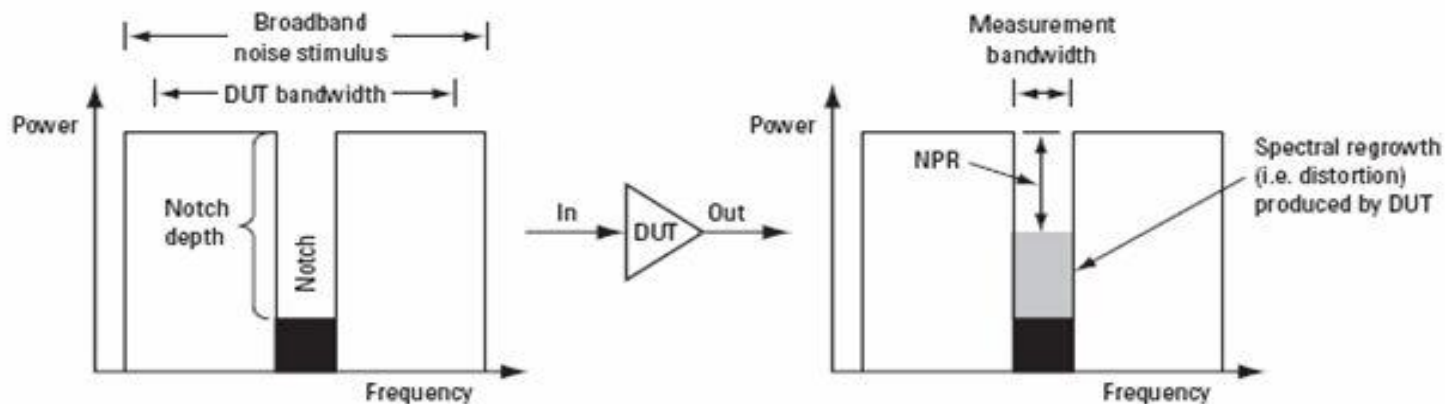
# Noise Pedestal Generation



- NPR measurement used for high power amplifiers with  $> 10$  carriers
- Pedestal is normally the width of a channel
- The notch is  $\leq 1\%$  of the pedestal
- The notch is typically  $\geq 50\text{dB}$  in depth



# NPR Notch Diagram



## No Load Notch

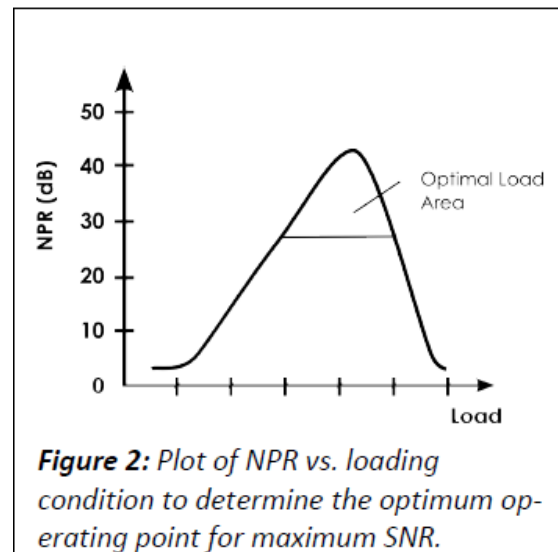
- Without DUT
- NPR of test equipment

## Loaded Notch

- With DUT

## NPR vs Load

- Loading DUT



# NPR Summary

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- NPR is a convenient method for evaluating the linear performance of amplifiers using multi-carriers ( $\geq 10$ )
- Care must be taken in selecting the proper NPR test equipment.
  - Noise pedestal  $\geq$  DUT BW
- NPR provides an accurate and repeatable measure of amplifier linear performance
- NPR represents the ratio of total output power to uncorrelated in-band distortion power

# How Can You Use Our Instruments?

- From the bench top to a rack system we have several form factors
- Bench top instruments plug into a common lab outlet and require minimal operating instructions
- Computer controlled instruments are the solution for complex ATE system
- Instruments with power meters and filters provide advanced capability



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Any Questions?



## WTG Regional Technical Contacts for Additional Questions

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