

# Final Semester Examination \*

Signals and Systems (ELC 321-2)  
Department of Electrical and Computer Engineering  
The College of New Jersey.

Spring 2014

**Last Name:**  
**Instructions:**

**First Name:**

1. This is a closed-book examination
2. Attempt all questions. Total score obtainable is 100%

**Problem 1** (25 Marks). *Consider the system simulation diagram of Figure 1. This figure shows a simulation diagram form used in the area of automatic control*

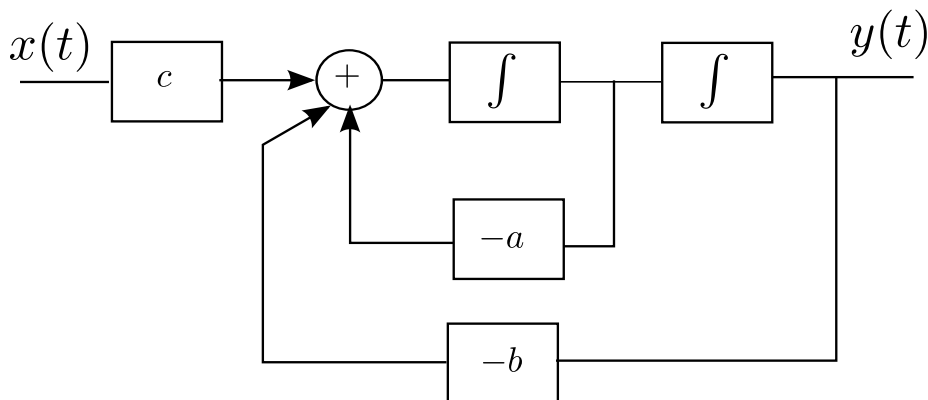


Figure 1: Simulation Diagram for Problem 1

- a) Find the differential equation of the system.
- b) Find the system transfer function  $H(s)$ . Assuming zero initial conditions.
- c) Suppose that  $a = 5$ ,  $b = 6$  and  $c = 10$ , determine the impulse response  $h(t)$  of the system.
- d) Determine the stability and causality of the system for the values of  $a, b$  and  $c$  in (c)

□

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**Solution 1** (25 Marks).

**Problem 2** (25 Marks). Consider the series  $RC$  circuit of Fig.2. The input is the applied voltage  $v(t)$  and the output is the voltage  $v_c(t)$  across the capacitor.

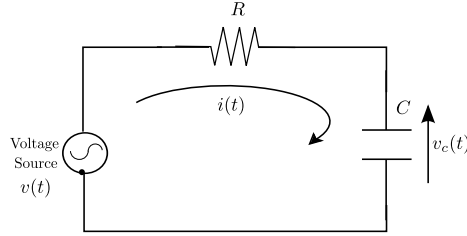


Figure 2: Sequence for Problem 1

a) Show that the differential equation describing the  $RC$  circuit is given by;

$$RC \frac{dv_c(t)}{dt} + v_c(t) = v(t). \quad (1)$$

b) Determine if the system is a causal linear time-invariant.

c) Suppose that  $R = 0.5\Omega$  and  $C = 0.25F$ , determine an output expression for the voltage across the capacitor  $v_c(t)$  for a unit step input ( $v(t) = \mu(t)$ ) and the initial condition  $v_c(0) = 0$ .

d) Suppose that we would like to recover the original signal  $v(t)$  from the output  $v_c(t)$  by using a system with transfer function  $W(s)$  (with input  $v_c(t)$  and output  $v(t)$ ), determine  $W(s)$ .

□

**Solution 2** (25 Marks).

**Problem 3** (25 Marks). *A simple way to smooth data is to take a weighted average of a number of samples. Consider the following moving average filter:*

$$y[n] = (1 - \alpha)x[n] + \alpha x[n - 1] \quad (2)$$

*where  $x[n]$  and  $y[n]$  represent the input and output sequences respectively. The coefficient  $\alpha$  is the smoothing factor.*

- a) Draw the simulation diagram for the filter.*
- b) Determine the impulse response  $h[n]$  for the filter.*
- c) Find the Discrete-time Fourier transform  $H(\Omega)$  of  $h[n]$*
- d) Determine the system transfer function  $H(z)$ . Is it possible for the filter to be unstable?*

□

**Solution 3** (25 Marks).

**Problem 4** (25 Marks). The block diagram of Figure 4 is an electronic oscillator for generating pure sinusoidal signal of a particular frequency, say  $\omega_o$ . The block comprises of a square wave generator and a filter.

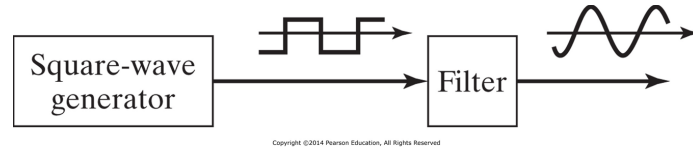


Figure 3: Electronic Oscillator

Let the output of the square wave generator be as shown in Fig. 2 and the final sinusoidal output be  $V_o(t) = A\sin(\omega_o t)$ .

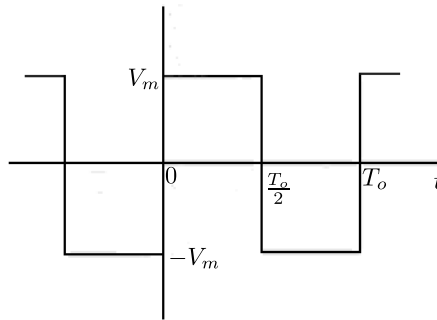


Figure 4: Square Wave

- Express the square wave as an exponential Fourier series.
- Calculate the average value of the square wave signal.
- What frequencies or harmonics must be filtered out by the filter circuit to obtain the final sinusoidal output  $V_o(t)$  and what type of filter would you deploy for this purpose?

**Solution 4** (25 Marks).



# 1 Reference

The Fourier series of a continuous-time signal  $x(t)$  is defined as:

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_o t} \quad (3)$$

$$c_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_o t} dt \quad (4)$$

The continuous-time Fourier transform (inverse Fourier transform) of  $x(t)$  is defined as

$$X(\omega) = \int_{t=-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad (5)$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega. \quad (6)$$

The magnitude and phase spectra of  $X(\omega)$  are given by  $|X(\omega)|$  and  $\angle X(\omega)$  respectively.

The Laplace transform (two-sided or bilateral) of signal  $x(t)$  is defined as

$$X(s) = \int_{t=-\infty}^{\infty} x(t) e^{-st} dt \quad (7)$$

For example  $\mathcal{L}[\mu(t)] = \frac{1}{s}$  and  $\mathcal{L}[e^{-at}\mu(t)] = \frac{1}{s+a}$

Given a discrete-time signal  $x[n]$ , its discrete-time Fourier transform (DTFT) (inverse DTFT)

$$X(\Omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\Omega n} \quad (8)$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\Omega) e^{j\Omega n} d\Omega \quad (9)$$

$X(\Omega)$  is said to be periodic with respect  $\Omega$  if  $X(\Omega + kT) = X(\Omega)$  where  $T$  is the period and  $k$  is any integer.

The z-transform (two-sided or bilateral) of signal  $x[n]$  is defined as

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n} \quad (10)$$

For example  $\mathcal{Z}(\mu[n]) = \frac{z}{z-1}$ ,  $\mathcal{Z}(a^n \mu[n]) = \frac{z}{z-a}$  and  $\mathcal{Z}(x[n-1]) = z^{-1} X(z)$ .

Given that an LTI system has an impulse response  $h[n]$ , the output response of the system  $y[n]$  for an input  $x[n]$  is given by

$$y[n] = h[n] * x[n] = \sum_{k=-\infty}^{k=\infty} h[n-k] x[k] = \sum_{k=-\infty}^{k=\infty} h[k] x[n-k] \quad (11)$$

The system function is given by

$$H(z) = \frac{Y(z)}{X(z)} = \sum_{k=-\infty}^{k=\infty} h[n] z^{-n} \quad (12)$$