SGM8521 SGM8522 SGM8524

150KHz, 4.7µA, Rail-to-Rail I/O CMOS Operational Amplifier

PRODUCT DESCRIPTION

The SGM8521 (single), SGM8522 (dual) and SGM8524 (quad) are rail-to-rail input and output voltage feedback amplifiers offering low cost. They have a wide input common-mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 2.1V and the maximum recommended supply voltage is 5.5 V. All are specified over the extended –40°C to +125°C temperature range.

The SGM8521/8522/8524 provide150KHz bandwidth at a low current consumption of $4.7\mu A$ per amplifier. Very low input bias currents of 0.5pA, enable the SGM8521/8522/8524 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-Rail inputs and outputs are useful to designers buffering ASIC in single-supply systems.

Applications for these amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The SGM8521 is available in the tiny SOT23-5 and SO-8 packages. The SGM8522 comes in the miniature SO-8 and MSOP-8 packages. The SGM8524 is offered in TSSOP-16 SO-16 and SO-14 packages.

APPLICATIONS

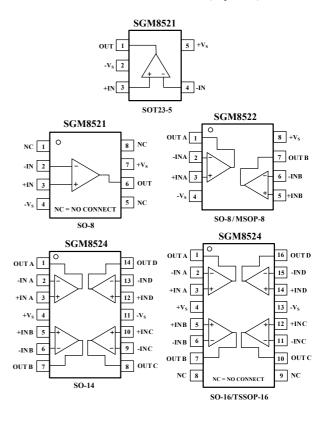
ASIC Input or Output Amplifier
Sensor Interface
Piezo Electric Transducer Amplifier
Medical Instrumentation
Mobile Communication
Audio Output
Portable Systems
Smoke Detectors
Mobile Telephone
Notebook PC
PCMCIA cards
Battery-Powered equipment

FEATURES

- Low Cost
- Rail-to-Rail Input and Output 0.8mV Typical Vos
- Unity Gain Stable
- Gain Bandwidth Product: 150KHz
- Very low input bias currents: 0.5pA
- Operates on 2.1 V to 5.5 V Supplies
- Input Voltage Range = -0.1 V to +5.6 V with Vs = 5.5 V
- Low Supply Current: 4.7µA/Amplifier
- Small Packaging

SGM8521 Available in SO-8 and SOT23-5 SGM8522 Available in SO-8 and MSOP-8 SGM8524 Available in SO-16, SO-14 and TSSOP-16

PIN CONFIGURATIONS (Top View)



ELECTRICAL CHARACTERISTICS: $V_S = +5V$

(At RL = $500k\Omega$ connected to Vs/2 and Vout = Vs/2, unless otherwise noted)

		CONDITION	SGM8521/8522/8524						
PARAMETER	Symbol		TYP	MIN/MAX OVER TEMPERATURE					
			+25℃	+25 ℃	0℃ to 70℃	- 40℃ to +85℃	- 40℃ to +125℃	UNITS	MIN/ MAX
INPUT CHARACTERISTICS									
Input Offset Voltage	Vos		±0.8	±3.5				mV	MAX
Input Bias Current	I _B		0.5					pА	TYP
Input Offset Current	los		0.5					pA	TYP
Common-Mode Voltage Range	V _{CM}	V _S = 5.5V	-0.1 to +5.6					V	TYP
Common-Mode Rejection Ratio	CMRR	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 4 V	91	72				dB	MIN
		$V_S = 5.5V$, $V_{CM} = -0.1V$ to 5.6 V	83	63				dB	MIN
Open-Loop Voltage Gain	A _{OL}	$R_L = 100K\Omega$, $Vo = 0.1V$ to 4.9V	100	84				dB	MIN
		$R_L = 500 \text{K}\Omega$, $Vo = 0.015 \text{V}$ to 4.965V	104	90				dB	MIN
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$		1.7					µV/°C	TYP
OUTPUT CHARACTERISTICS									
Output Voltage Swing from Rail		R _L = 500KΩ	0.005					V	MAX
Output Current	I _{OUT}		22					mA	MIN
Short-Circuit Current	I _{SC}							mA	TYP
POWER SUPPLY									
Operating Voltage Range				2.1				V	MIN
				5.5				V	MAX
Power Supply Rejection Ratio	PSRR	V _s = +2.5 V to + 5.5 V							
		$V_{CM} = (-V_S) + 0.5$	91					dB	MIN
Quiescent Current / Amplifier	IQ	I _{OUT} = 0	4.7					μΑ	MAX
DYNAMIC PERFORMANCE		C _L = 100pF							
Gain-Bandwidth Product	GBP		150					KHz	TYP
Slew Rate	SR	G = +1 , 2V Output Step	0.04					V/µs	TYP
Settling Time to 0.1%	ts	G = +1, 2 V Output Step	32					μs	TYP
Overload Recovery Time		V _{IN} ·Gain = Vs						μs	TYP
NOISE PERFORMANCE									
Voltage Noise Density	e _n	f = 1kHz	40					nV/\sqrt{Hz}	TYP
		f = 10kHz	12					$nV/_{\sqrt{Hz}}$	TYP

Specifications subject to change without notice.

PACKAGE/ORDERING INFORMATION

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION	
CCN 40501	SGM8521XN5/TR	SOT23-5	Tape and Reel, 3000	8521	
SGM8521	SGM8521XS/TR	SO-8	Tape and Reel, 2500	SGM8521XS	
SGM8522	SGM8522XS/TR	SO-8	Tape and Reel, 2500	SGM8522XS	
	SGM8522XMS/TR	MSOP-8	Tape and Reel, 3000	SGM8522XMS	
SGM8524	SGM8524XS/TR	SO-16	Tape and Reel, 2500	SGM8524XS	
	SGM8524XS14	SO-14	Tube	SGM8524XS14	
	SGM8524XTS/TR	TSSOP-16	Tape and Reel, 3000	SGM8524XTS	

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V+ to V
Common-Mode Input Voltage
$(-V_s) - 0.5 V$ to $(+V_s) + 0.5V$
Storage Temperature Range -65° C to $+150^{\circ}$ C
Junction Temperature
Operating Temperature Range -55° C to $+150^{\circ}$ C
Package Thermal Resistance @ $T_A = 25^{\circ}C$
SOT23-5, θ _{JA}
SO-8, θ _{JA}
MSOP-8, θ _J A
SO-16, θ _{JA}
TSSOP-16, θ_{JA}
Lead Temperature Range (Soldering 10 sec)
260℃
ESD Susceptibility
HBM4000V
MM400V

NOTES

1. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

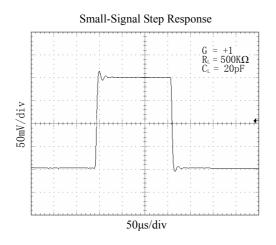
CAUTION

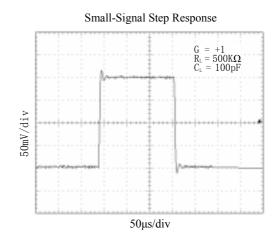
This integrated circuit can be damaged by ESD. Shengbang Micro-electronics recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

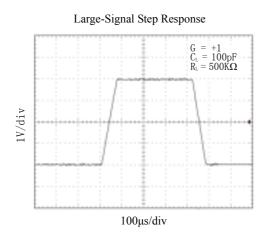
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

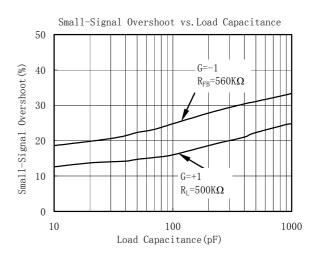
TYPICAL PERFORMANCE CHARACTERISTICS

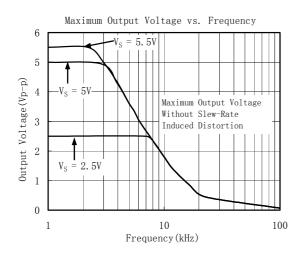
At T_A = +25°C, V_S = +5V, and R_L =500k Ω connected to Vs/2,unless otherwise noted.

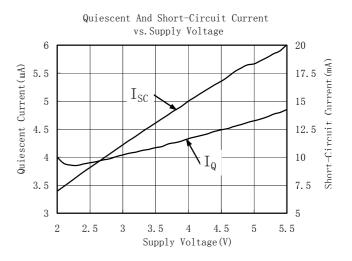






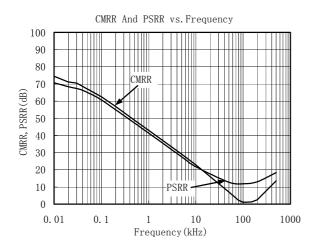


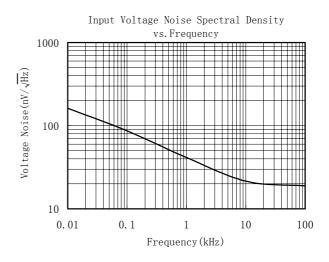


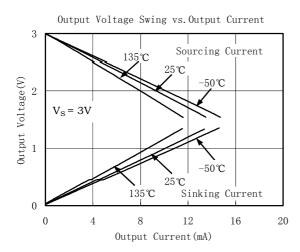


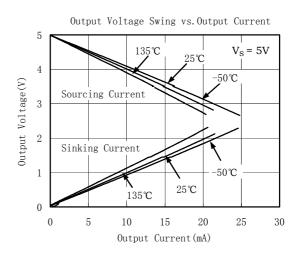
TYPICAL PERFORMANCE CHARACTERISTICS

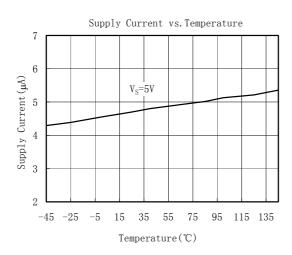
At $T_A = +25$ °C, $V_S = +5V$, and $R_L = 500$ k Ω connected to Vs/2, unless otherwise noted.











APPLICATION NOTES

Driving Capacitive Loads

The SGM852X can directly drive 250pF in unity-gain without oscillation. The unity-gain follower (buffer) is the most sensitive configuration to capacitive loading. Direct capacitive loading reduces the phase margin of amplifiers and this results in ringing or even oscillation. Applications that require greater capacitive drive capability should use an isolation resistor between the output and the capacitive load like the circuit in Figure 1. The isolation resistor $R_{\rm ISO}$ and the load capacitor $C_{\rm L}$ form a zero to increase stability. The bigger the $R_{\rm ISO}$ resistor value, the more stable $V_{\rm OUT}$ will be. Note that this method results in a loss of gain accuracy because $R_{\rm ISO}$ forms a voltage divider with the $R_{\rm LOAD}$.

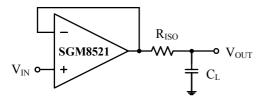


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2, It provides DC accuracy as well as AC stability. R_{F} provides the DC accuracy by connecting the inverting signal with the output, C_{F} and R_{Iso} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

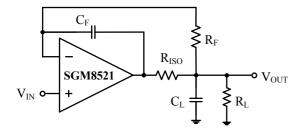


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For no-buffer configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

Power-Supply Bypassing and Layout

The SGM852X family operates from either a single +2.5V to +5.5V supply or dual $\pm 1.25V$ to $\pm 2.75V$ supplies. For single-supply operation, bypass the power supply V_{DD} with a $0.1\mu F$ ceramic capacitor which should be placed close to the V_{DD} pin. For dual-supply operation, both the V_{DD} and the V_{SS} supplies should be bypassed to ground with separate $0.1\mu F$ ceramic capacitors. $2.2\mu F$ tantalum capacitor can be added for better performance.

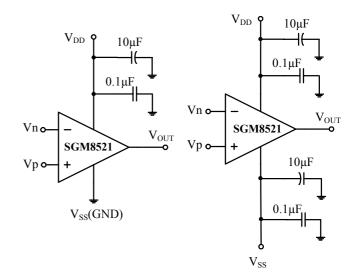


Figure 3. Amplifier with Bypass Capacitors

Typical Application Circuits

Differential Amplifier

The circuit shown in Figure 4 performs the difference function. If the resistors ratios are equal (R4 / R3 = R2 / R1), then $V_{OUT} = (Vp - Vn) \times R_2 / R_1 + Vref$.

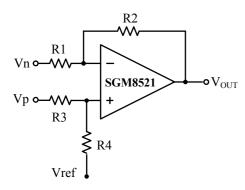


Figure 4. Differential Amplifier

Instrumentation Amplifier

The circuit in Figure 5 performs the same function as that in Figure 4 but with the high input impedance.

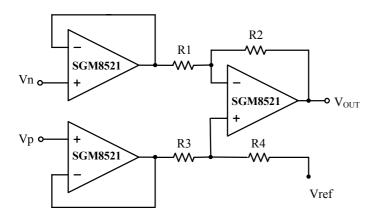


Figure 5. Instrumentation Amplifier

Low Pass Active Filter

The low pass filter shown in Figure 6 has a DC gain of (- R_2/R_1) and the –3dB corner frequency is $1/2\pi R_2 C$. Make sure the filter is within the bandwidth of the amplifier. The Large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

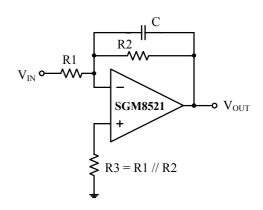
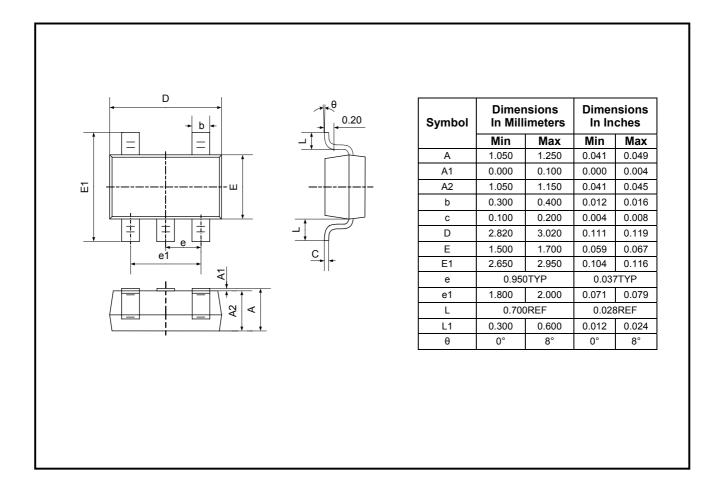
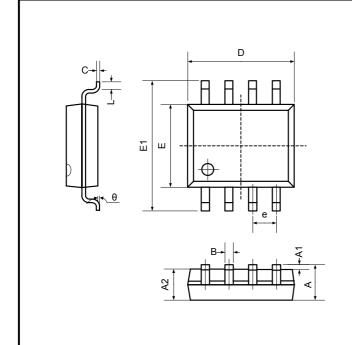


Figure 6. Low Pass Active Filter

SOT23-5

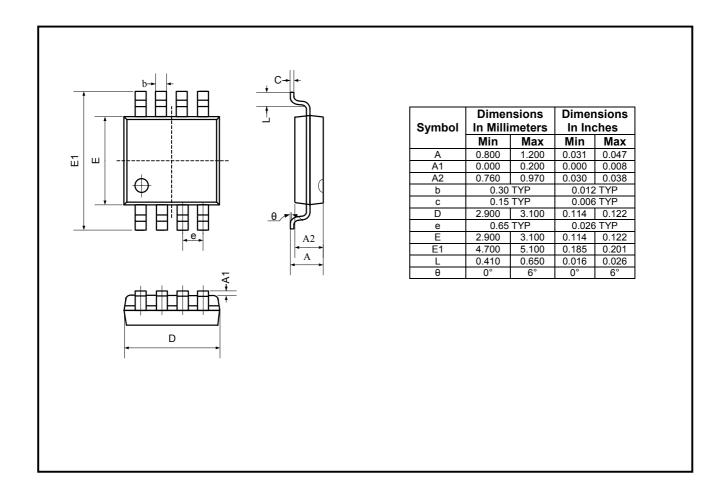


SO-8

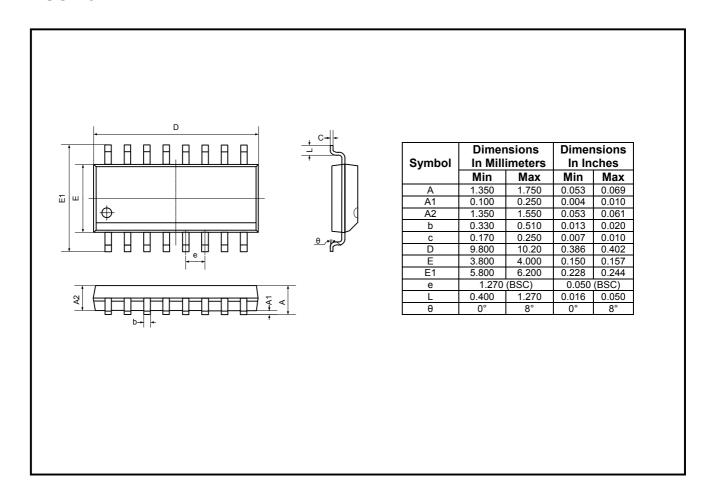


Symbol	Dimer In Milli	nsions meters	Dimensions In Inches			
	Min	Max	Min	Max		
Α	1.350	1.750	0.053	0.069		
A1	0.100	0.250	0.004	0.010		
A2	1.350	1.550	0.053	0.061		
В	0.330	0.510	0.013	0.020		
С	0.190	0.250	0.007	0.010		
D	4.780	5.000	0.188	0.197		
Е	3.800	4.000	0.150	0.157		
E1	5.800	6.300	0.228	0.248		
е	1.270)TYP	0.050TYP			
L	0.400	1.270	0.016	0.050		
θ	0°	8°	0°	8°		

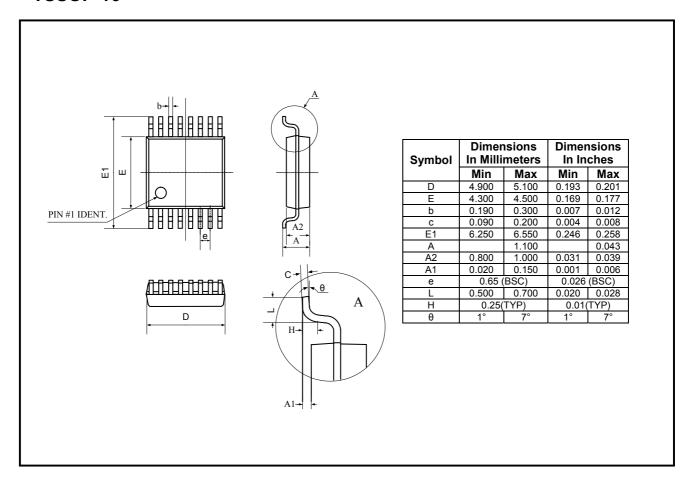
MSOP-8



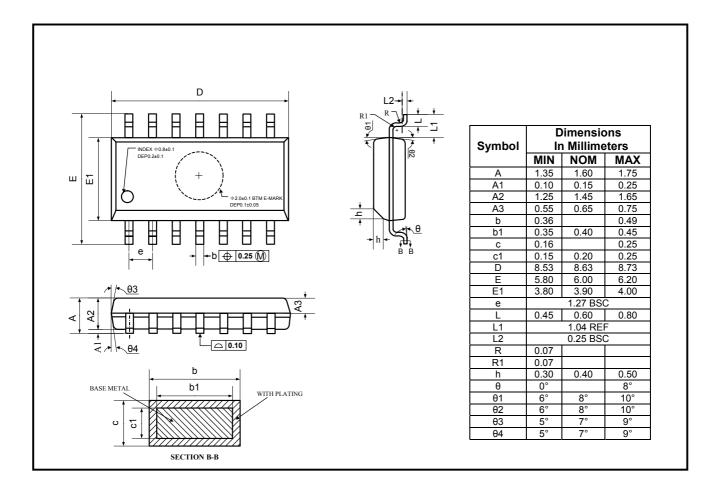
SO-16



TSSOP-16



SO-14



REVISION HISTORY

Location	Page
11/06— Data Sheet changed from Preliminary to REV. A	
Changes to ABSOLUTE MAXIMUM ATINGS	3
11/06— Data Sheet changed from REV. A to REV. B	
Adds SO-14 Package	1, 3

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