## Digital Control Systems Final CA

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## **Equilibrium point / Transfer function (p1)**

#### **Defining Variables**

### **Equilibrium State**

```
St_e = [0;0;0];
In_e = [n*Ib*V;0];
Fe = subs(subs(F,In,In_e),St,St_e);
```

#### Linearize

```
A =

0 -4.5000 0
0 -0.0250 0.0000
0 0 -0.0923
```

#### **Transfer Function**

```
syms s
n = size(A,1);
Gp = C*inv(s*eye(n)-A)*B;
ExpFun = matlabFunction(simplifyFraction(Gp, 'Expand', true));
ExpFun = str2func(regexprep(func2str(ExpFun), '\.([/^\\*])', '$1'));
Gp = tf(ExpFun(tf('s')));
for i = 1:length(Gp)
    [num,den] = tfdata(Gp(i));
    Gp(i) = tf(num{1}/den{1}(1),den{1}/den{1}(1));
end
Gp = Gp(1)
Gp =
          -4.875e-06
  s^3 + 0.1173 s^2 + 0.002308 s
Continuous-time transfer function.
[Gp_Num, Gp_Den] = tfdata(Gp); Gp_Num = Gp_Num{1}; Gp_Den = Gp_Den{1};
```

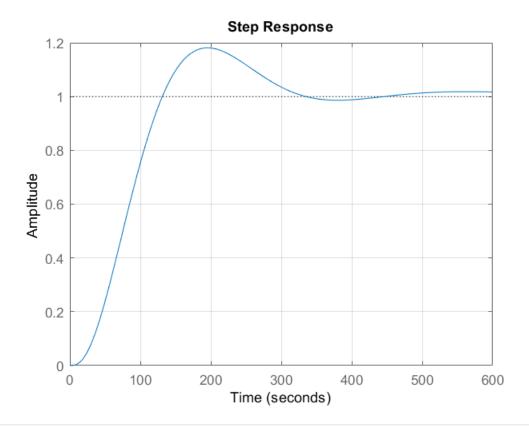
## **Analog Control (p2)**

```
%pidTuner
```

#### trans\_info = stepinfo(feedback(series(Gc,Gp),1))

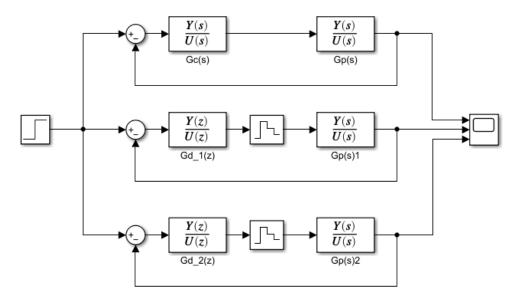
```
trans_info = struct with fields:
    RiseTime: 83.7907
SettlingTime: 312.4786
SettlingMin: 0.9142
SettlingMax: 1.1813
    Overshoot: 18.1258
Undershoot: 0
    Peak: 1.1813
PeakTime: 194.7882
```

# step(feedback(series(Gc,Gp),1)) grid on



[Gc\_Num, Gc\_Den] = tfdata(Gc); Gc\_Num = Gc\_Num{1}; Gc\_Den = Gc\_Den{1};

## **Digital Control (P3)**



```
% 2 < tr/Ts < 10
Ts = round(trans_info.RiseTime/6)</pre>
```

Ts = 14

Gd\_1 =

-7.844 z + 7.816

----z - 1

Sample time: 14 seconds
Discrete-time transfer function.

#### Gd\_2 = c2d(Gc,Ts,'matched')

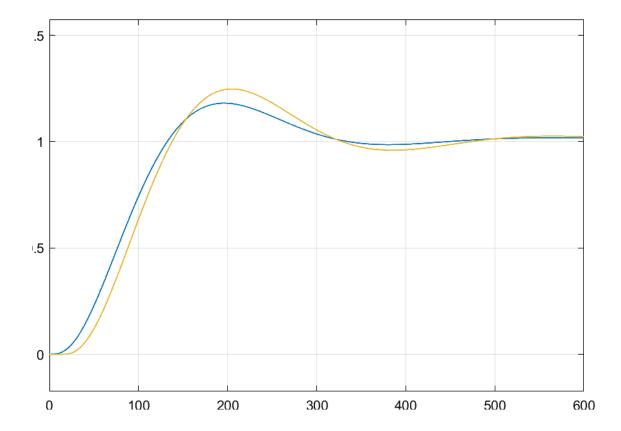
Gd\_2 =

-7.844 z + 7.816

----z - 1

Sample time: 14 seconds
Discrete-time transfer function.

[Gd\_Num\_1, Gd\_Den\_1] = tfdata(Gd\_1); Gd\_Num\_1 = Gd\_Num\_1{1}; Gd\_Den\_1 = Gd\_Den\_1{1};
[Gd\_Num\_2, Gd\_Den\_2] = tfdata(Gd\_2); Gd\_Num\_2 = Gd\_Num\_2{1}; Gd\_Den\_2 = Gd\_Den\_2{1};



Ts = 4

Ts = 4

 $Gd_1 =$ 

-7.834 z + 7.826 -----z - 1

Sample time: 4 seconds

Discrete-time transfer function.

#### Gd\_2 = c2d(Gc,Ts,'matched')

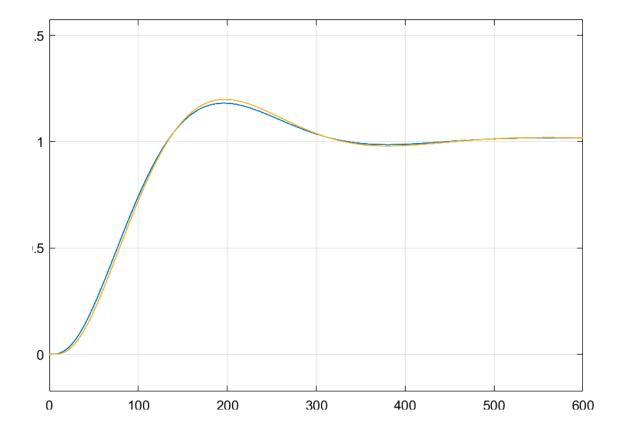
 $Gd_2 =$ 

-7.834 z + 7.826 -----z - 1

Sample time: 4 seconds

Discrete-time transfer function.

```
[Gd_Num_1, Gd_Den_1] = tfdata(Gd_1); Gd_Num_1 = Gd_Num_1{1}; Gd_Den_1 = Gd_Den_1{1};
[Gd_Num_2, Gd_Den_2] = tfdata(Gd_2); Gd_Num_2 = Gd_Num_2{1}; Gd_Den_2 = Gd_Den_2{1};
```



Ts = 40

Ts = 40

 $Gd_1 =$ 

-7.869 z + 7.791 -----z - 1

Sample time: 40 seconds

Discrete-time transfer function.

#### Gd\_2 = c2d(Gc,Ts,'matched')

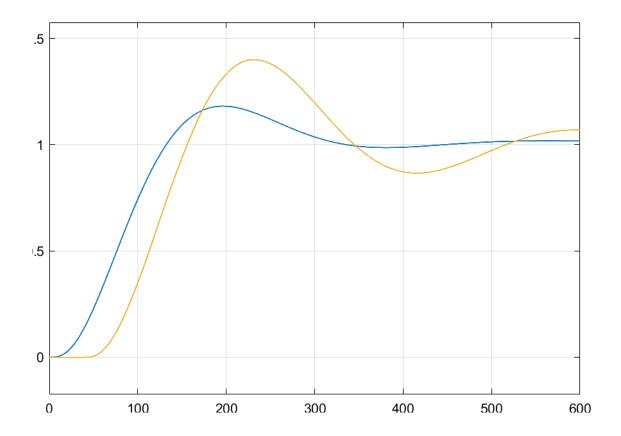
 $Gd_2 =$ 

-7.869 z + 7.791 -----z - 1

Sample time: 40 seconds

Discrete-time transfer function.

```
[Gd_Num_1, Gd_Den_1] = tfdata(Gd_1); Gd_Num_1 = Gd_Num_1{1}; Gd_Den_1 = Gd_Den_1{1};
[Gd_Num_2, Gd_Den_2] = tfdata(Gd_2); Gd_Num_2 = Gd_Num_2{1}; Gd_Den_2 = Gd_Den_2{1};
```



# **Discrete Transfer Function (p4)**

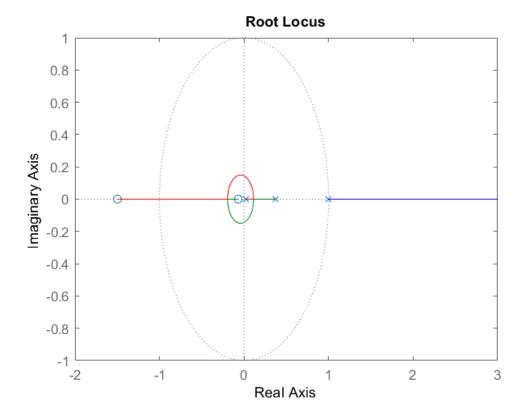
## Using c2d in matlab:

Discrete-time transfer function.

### **Root Locus**

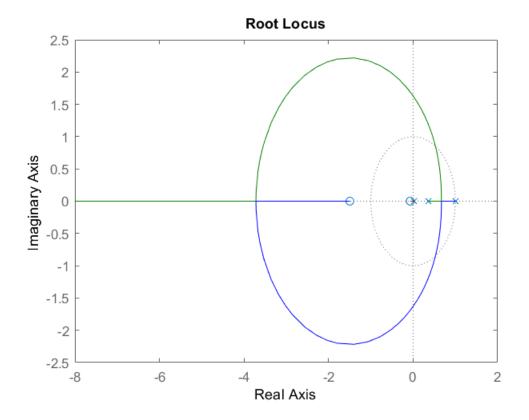
For Positive k:

rlocus(Gp\_d)

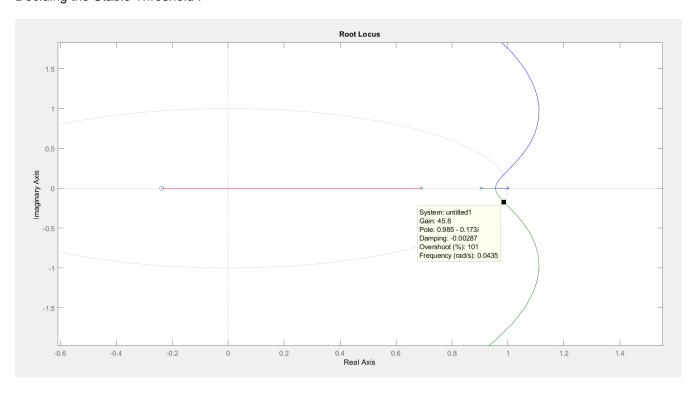


## For Negative k:

rlocus(-Gp\_d)



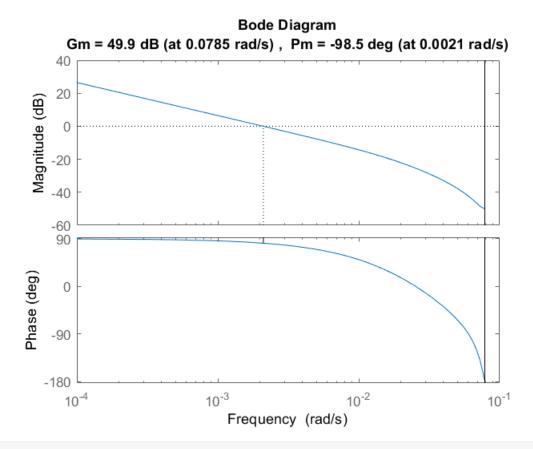
### Deciding the Stable Threshold:



Therefore => -45.8<k<0

### Bode plot

#### margin(Gp\_d)



[Gm,Pm,Wgm,Wpm] = margin(Gp\_d)

Warning: The closed-loop system is unstable.

Gm = 311.1157 Pm = -98.5242 Wgm = 0.0785Wpm = 0.0021

## Design a controler (p5)

Deciding The two major closed loop poles by the given information:

$$\begin{split} t_s &= \frac{4}{\zeta \omega_n} \text{ , } MP = \mathrm{e}^{-\frac{\zeta \omega_n}{\sqrt{1-\zeta^2}}} \to \zeta^2 = 1 - \frac{16}{(t_s \ln MP)^2} \\ t_s &= 100 \text{ , } MP = 0.2 \to \zeta = \pm 0.9994 \text{ , } \omega_n = 0.04 \text{ , } if \ T = 10 \to \frac{w_s}{w_d} \cong 453 \gg 8 \\ \to \begin{cases} |z| = e^{-\zeta \omega_n T} = 0.4493 \\ < z = T \ w_n \sqrt{1-\zeta^2} = 0.0277 \ rad = 1.587^\circ \end{cases} \end{split}$$

Continuous-time transfer function.

```
Gp_d = c2d(Gp,Ts,'zoh')
```

```
[num,den]=tfdata(Gp_d);
```

Calculate The zeros, Poles and the Gain of the Discrete Transfer Function:

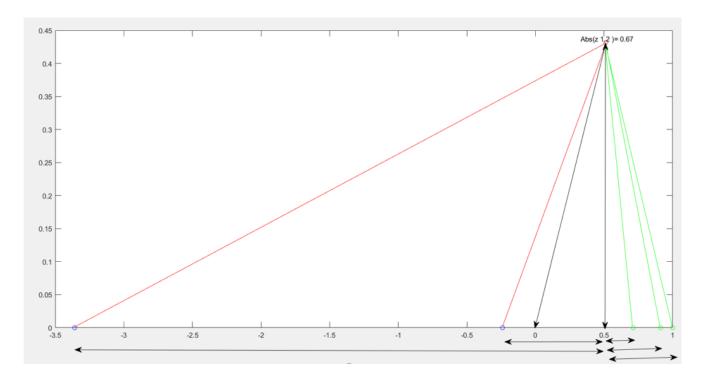
```
[z,p,k] = tf2zp(num{1},den{1})
z =
```

-0.1967 = 1.0000 0.7788 0.3972

-2.8306

k = -6.1431e-04

Designing a Lead/Lag Controller depending on the position of zeros and poles :



$$\begin{cases} \theta_1 = 180 - \tan^{-1}\left(\frac{0.4493\sin(0.0277)}{0.7788 - 0.4493\cos(0.0277)}\right) * \frac{180}{\pi} = 177.838^{\circ} \\ \theta_2 = 180 - \tan^{-1}\left(\frac{0.4493\sin(0.0277)}{1 - 0.4493\cos(0.0277)}\right) * \frac{180}{\pi} = 178.706^{\circ} \\ \theta_3 = \tan^{-1}\left(\frac{0.67\sin(0.0277)}{0.4493\cos(0.0277) - 0.3972}\right) * \frac{180}{\pi} = 19.6646^{\circ} \end{cases} , \begin{cases} \varphi_1 = \tan^{-1}\left(\frac{0.4493\sin(0.0277)}{0.1967 + 0.4493\cos(0.0277)}\right) * \frac{180}{\pi} = 1.104^{\circ} \\ \varphi_2 = \tan^{-1}\left(\frac{0.4493\sin(0.0277)}{0.4493\sin(0.0277)}\right) * \frac{180}{\pi} = 0.217^{\circ} \end{cases} \end{cases}$$

$$\varphi_1 + \varphi_2 - \theta_1 - \theta_2 - \theta_3 = -349.717 \rightarrow -374.8876 + 360 = -14.8876^{\circ} \rightarrow reduced \ phase \end{cases}$$

$$G_D(z) = \frac{k(z + \alpha)}{z + \beta} , Assume \rightarrow z + \alpha = z + 0.7123 \rightarrow \alpha = 0.7123$$

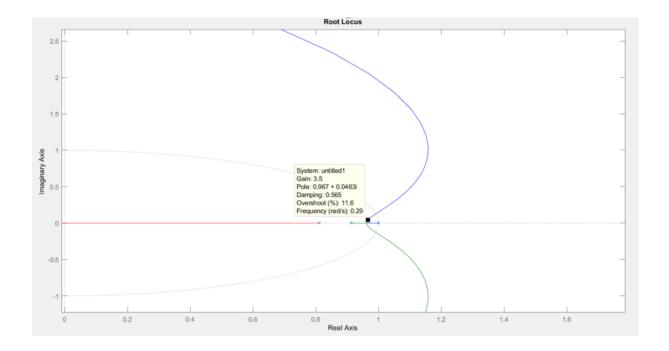
$$\Rightarrow G_D(z)G(z) = -\frac{6.1431 * 10^{-4} k (z + 0.1967)(z + 2.8306)}{(z - \beta)(z - 0.3972)(z - 0.7788)} \rightarrow \varphi_1 + \varphi_2 - \theta' - \theta_2 - \theta_3 = -360$$

$$\Rightarrow 360 + 1.104 + 0.217 - 177.838 - 19.6646 = \theta' = 163.8184^{\circ} \rightarrow \tan^{-1}\left(\frac{0.4493\sin(0.0277)}{\beta - 0.4493\cos(0.0277)}\right) * \frac{180}{\pi} = 16.1816^{\circ} \end{cases}$$

$$\Rightarrow \beta = 0.492$$

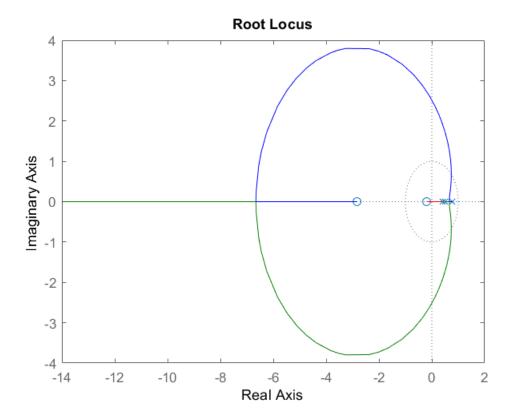
So, Reconstructing the Open loop Transfer function without the coefficient k : Gp(z)\*G(z)

Sample time: 10 seconds
Discrete-time transfer function.

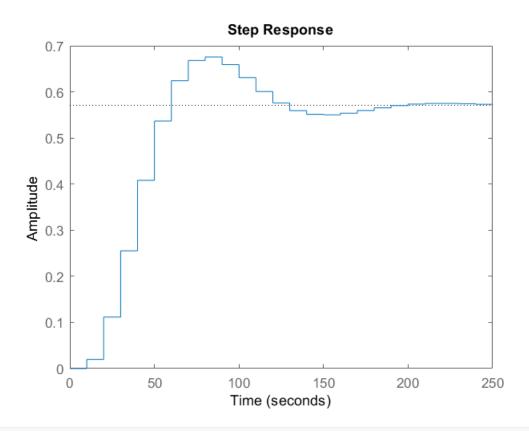


Decide to choose the controller coeffient form root locus

rlocus(-Gop\_pre)



#### step(feedback(-32\*Gop\_pre,1))



#### stepinfo(step(feedback(-32\*Gop\_pre,1)))

```
ans = struct with fields:
    RiseTime: 3.4343
    SettlingTime: 18.5526
    SettlingMin: 0.5367
    SettlingMax: 0.6758
        Overshoot: 17.6908
    Undershoot: 0
        Peak: 0.6758
        PeakTime: 9
```

### Gop = -32\*Gop\_pre

```
Gop =

0.01966 z^2 + 0.05951 z + 0.01095

-----
z^3 - 1.668 z^2 + 0.8879 z - 0.1522

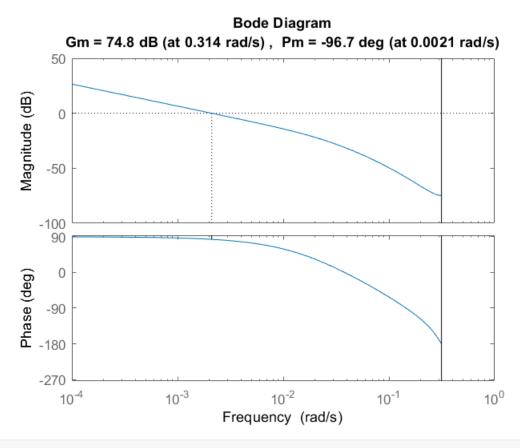
Sample time: 10 seconds
```

Discrete-time transfer function.

## Comparing the features per plot (p6)

### **Bode plot (Uncontrolled system)**

margin(Gp\_d)



#### [Gm,Pm,Wgm,Wpm] = margin(Gp\_d)

Warning: The closed-loop system is unstable.

Gm = 5.5026e+03 Pm = -96.7192 Wgm = 0.3142 Wpm = 0.0021

### Root Locus (Uncontrolled system)

```
Ts = 10;
Gp_d = c2d(Gp,Ts,'zoh')
```

```
Gp_d =

-0.0006143 z^2 - 0.00186 z - 0.000342

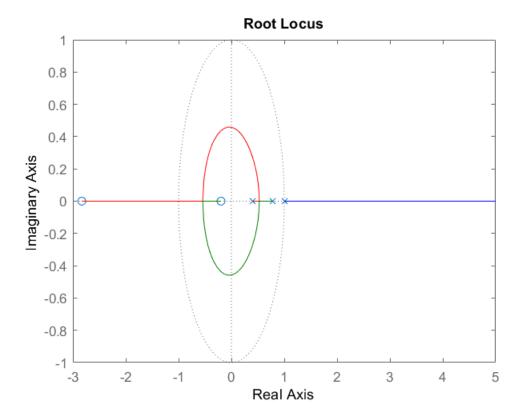
z^3 - 2.176 z^2 + 1.485 z - 0.3093
```

Sample time: 10 seconds

Discrete-time transfer function.

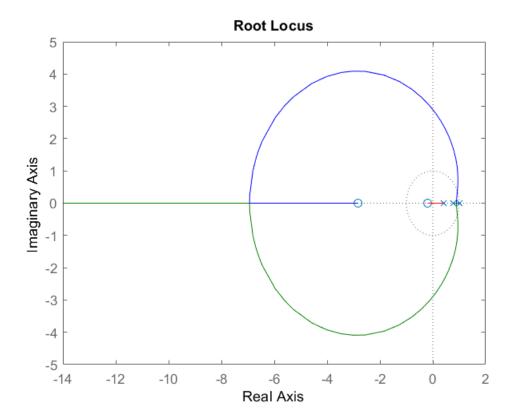
### For Positive k:

## rlocus(Gp\_d)



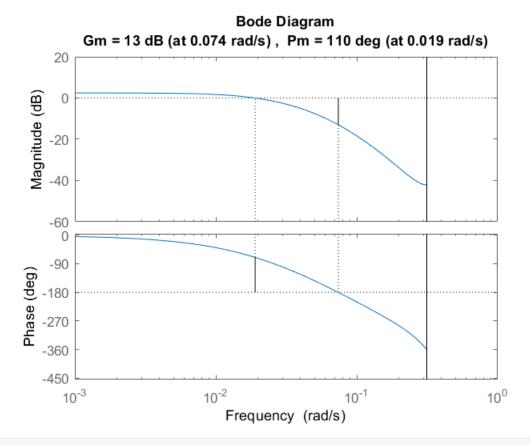
### For Negative k:

rlocus(-Gp\_d)



## **Bode Plot (Controlled system)**

margin(Gop)



[Gm,Pm,Wgm,Wpm] = margin(Gop)

Gm = 4.4548

Pm = 110.0369

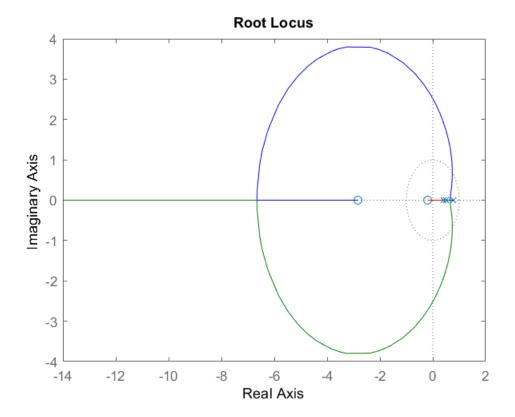
Wgm = 0.0740

Wpm = 0.0190

### **Root Locus (Controlled system)**

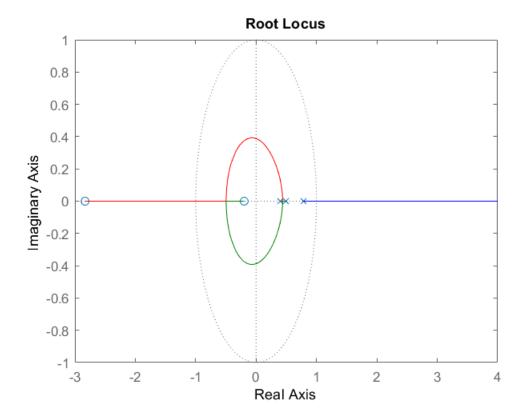
For positive k:

rlocus(Gop)



## For negative k:

rlocus(-Gop)



## **DeadBeat controller (p7)**

$$G(z) = -6.1431e - 04 * \frac{(z + 2.8306)(z + 0.1967)}{(z - 1)(z - 0.7788)(z - 0.3972)} = a * z^{-1} + \cdots$$

\* The First Sentence of G(z) starts with the term  $z^{-1} \to F(z) = f_1 z^{-1} + f_2 z^{-2}$ 

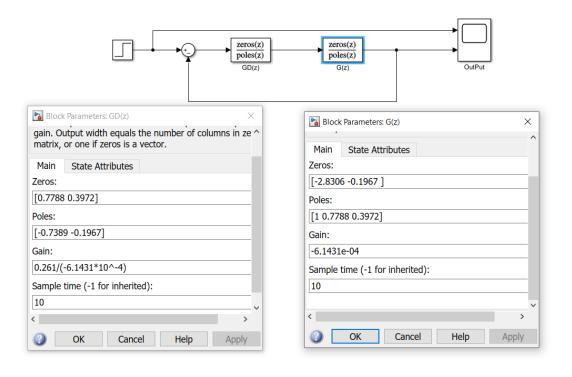
$$\begin{cases} F(z) = (1+2.8306z^{-1})(m_1z^{-1}) \\ 1 - F(z) = (1-z^{-1})(1+n_1z^{-1}) \end{cases} \rightarrow \begin{cases} f_1 + n_1 - 1 = 0 \\ f_2 - n_1 = 0 \\ f_1 - m_1 = 0 \\ f_2 - 2.8306 \ m_1 = 0 \end{cases}$$

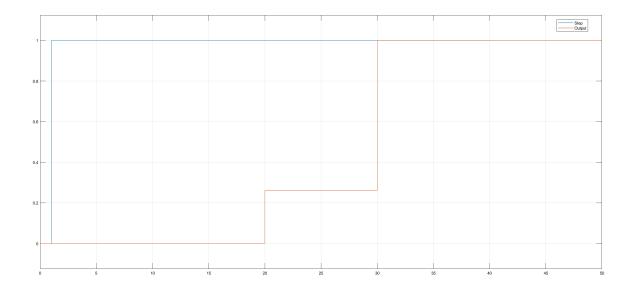
Solve[ $\{f1 + n1 - 1 = 0, f2 - n1 = 0, f1 - m1 = 0, f2 - 2.8306 * m1 = 0\}$ ,  $\{f1, n1, f2, m1\}$ ]

 $\{\{f1 \rightarrow 0.261056, n1 \rightarrow 0.738944, f2 \rightarrow 0.738944, m1 \rightarrow 0.261056\}\}$ 

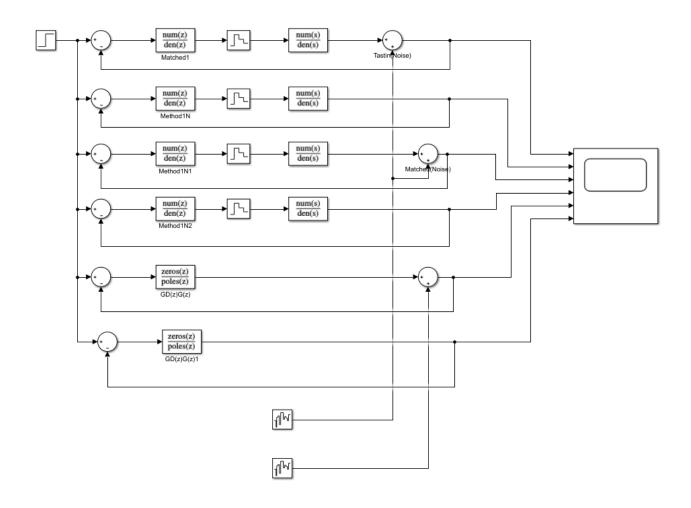
 $\rightarrow F(z) = 0.261z^{-1} + 0.7389z^{-2}$ 

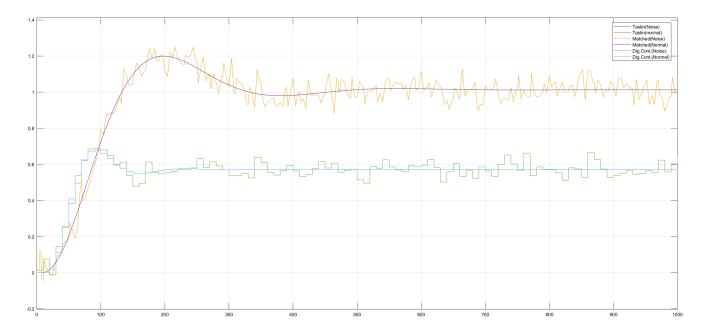
$$\rightarrow G_D(z) = \frac{F(z)}{\left(1 - F(z)\right)G(z)} = \frac{0.261(1 - 0.7788z^{-1})(1 - 0.3972z^{-1})}{-6.1431e - 04(1 + 0.7389z^{-1})(1 + 0.1967z^{-1})}$$





# **Comparing controllers (p8)**





So The Digital Controller in part 5, Outperforms in comparsion with the Matched and Tastin Methods to resist with the white Noise.

\*Tastin and Matched Methods Overlapped since the Digital controllers obtained for both of them are similar

(This is due to The fact that the Discretized version of the analog PD controller is the same for both case)

## State space (p9)

#### **Discrete State Space**

Continuous-time state-space model.

```
sys_d = c2d(sys,Ts)
sys_d =
  A =
             x1 x2 x3
1 -17.13 -0.0004014
0 0.9048 4.125e-05
0 0 0.6912
   x1
   x2
   х3
   B =
                u1
   x1 -4.635e-05
   x2 7.434e-06
            0.2787
   х3
       x1 x2 x3
   y1 1 0 0
  D =
       u1
   y1 0
Sample time: 4 seconds
Discrete-time state-space model.
G = sys_d.A; H = sys_d.B;
```

#### Controllable & Observable

```
n = size(G,1);
M = [];
N = [];
for i = 0:n-1
  M = [M G^i*H];
    N = [N ; C*G^i];
end
Μ
   -0.0000 -0.0003 -0.0007
   0.0000 0.0000 0.0000
   0.2787 0.1926 0.1332
fprintf('Rank(M):%i\n',rank(M))
Rank(M):3
if rank(M) == n
    disp('(G,H) is controllable')
end
```

(G,H) is controllable

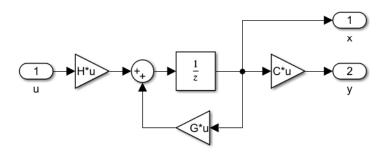
```
Ν
N =
    1.0000
    1.0000 -17.1293 -0.0004
    1.0000 -32.6285 -0.0014
fprintf('Rank(N):%i\n',rank(N))
Rank(N):3
if rank(N) == n
    disp('(G,C) is observable')
end
(G,C) is observable
% Hankel Matrix
if rank(N*M) == n
    disp('State space is minimal')
end
State space is minimal
```

## DeadBeat controller from the state space (p10)

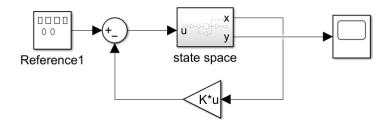
```
% we can use all of the states
tf_d = tf(sys_d)
tf_d =
  -4.635e-05 z^2 - 0.0001652 z - 3.666e-05
     z^3 - 2.596 z^2 + 2.221 z - 0.6254
Sample time: 4 seconds
Discrete-time transfer function.
[tf_Num,tf_Den] = tfdata(tf_d);
n = size(tf_Den\{1\},2);
a = zeros(1,n-1);
b = zeros(1,n);
for i = 1:n-1
    a(i) = tf_Den{1}(i+1);
    b(i) = tf_Num\{1\}(i);
end
W = [];
for i = 0:n-2
    W = [W ; flip(a(1:end-1-i)) 1 zeros(1,i)];
end
T = M*W;
K_deadbeat = -flip(a)*inv(T)
```

```
K_deadbeat =
   1.0e+05 *
   -0.0403   1.3010   0.0001
```

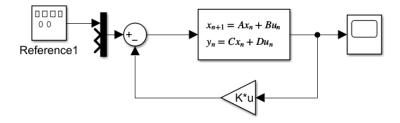
In order to achive the states, we need to implement the system with delay block:



Then simply connect the state feedback controller:



The other way is implement it as discrete state space block and connect the state feedback controller. But we need to set the C matrix equal to eye(3) because we need all the states in the output:



## Deadbeat controller with full rank observer (p11)

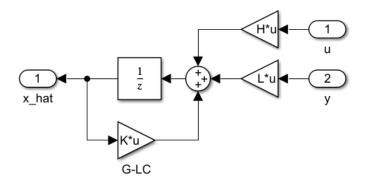
```
% x[k+1] = (G-LC)x[k] + Hu[k] + Ly[k]
G_new = transpose(G);
H_new = transpose(C);
L = transpose(acker(G_new,H_new,zeros(n-1,1)))

L =
    2.5960
    -0.0997
    -531.9399
```

$$\dot{\hat{x}} = (A - LC) * \hat{x} + Bu + Ly$$

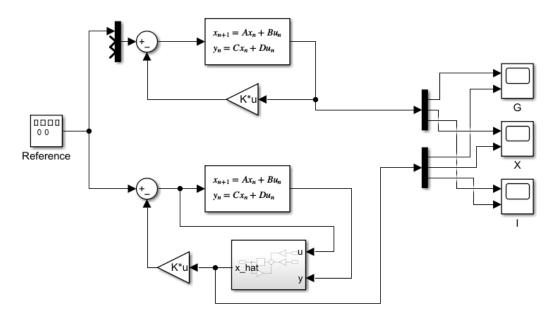
$$\begin{cases} k \to L^T \\ B \to C^T \to \det(\lambda I - (A - LC)) = \det(\lambda I - (A^T - L^T C^T)) \\ A \to A^T \end{cases}$$

So the block of observer is something like this:

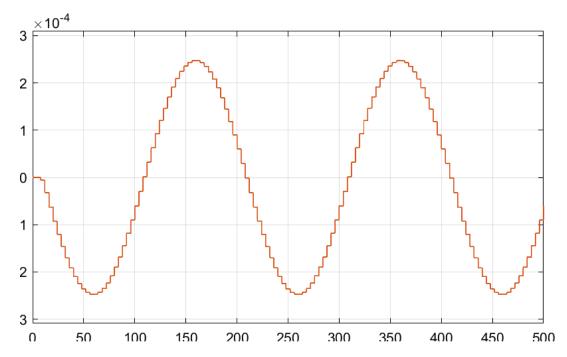


To compare states from system without observer and system with observer we impelement the circuit below.

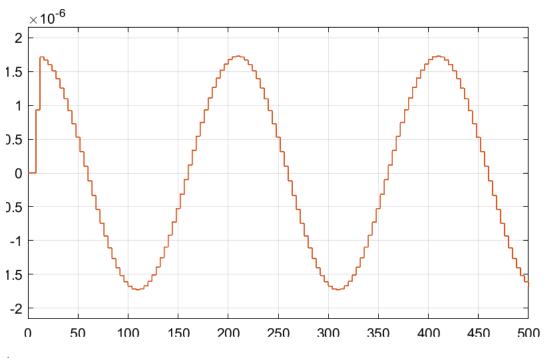
The input is a sinous signal with the frequency of 0.005Hz.



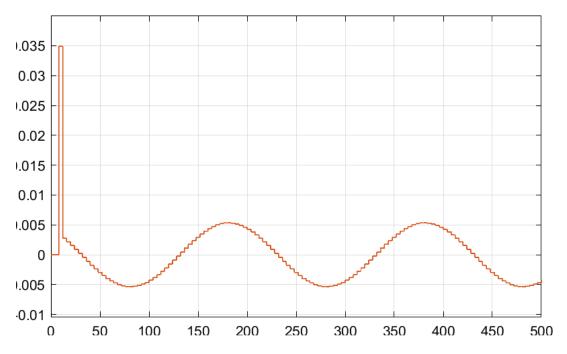
G:



X:



l:



As we can see, observer prefectly follows the states. But if we apply the initial conditions, we can see the error between them that converges to zero over the time.

initial condition = 0.1

G:

