

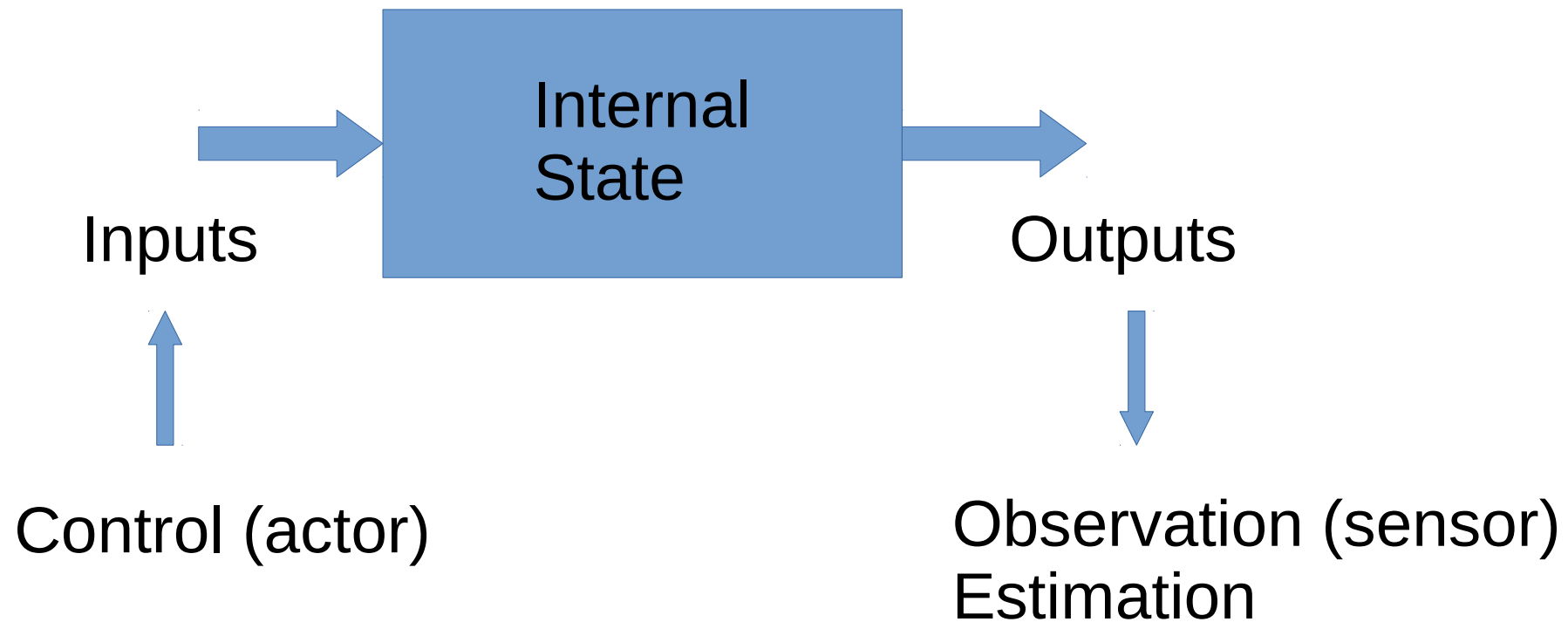
Introduction to Linear System Control & Kalman Filter

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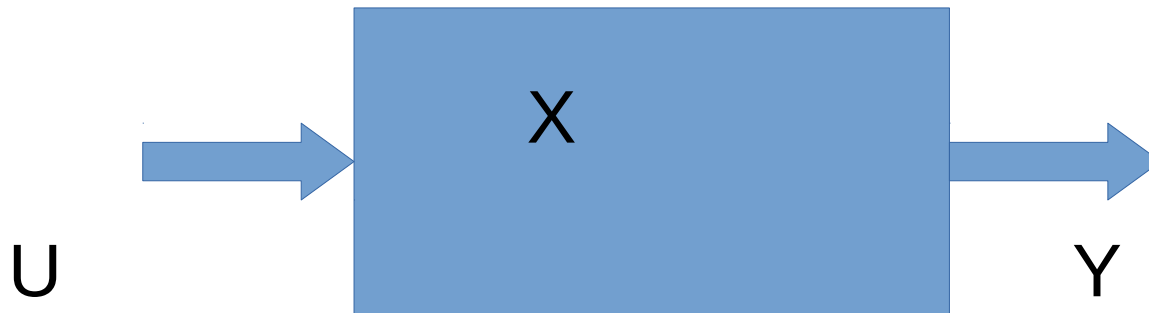
Outline

- System Model Equation
- Input = Theory of Control
(Riccati Equation – Kalman Gain)
- Output = Theory of Estimation
(Kalman Filter)

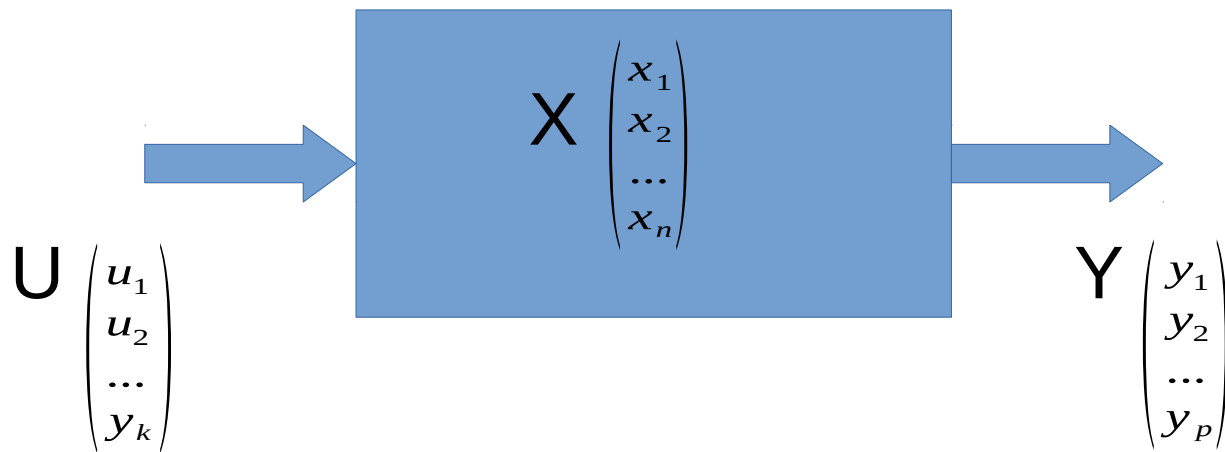
System ... Black Box



Notation

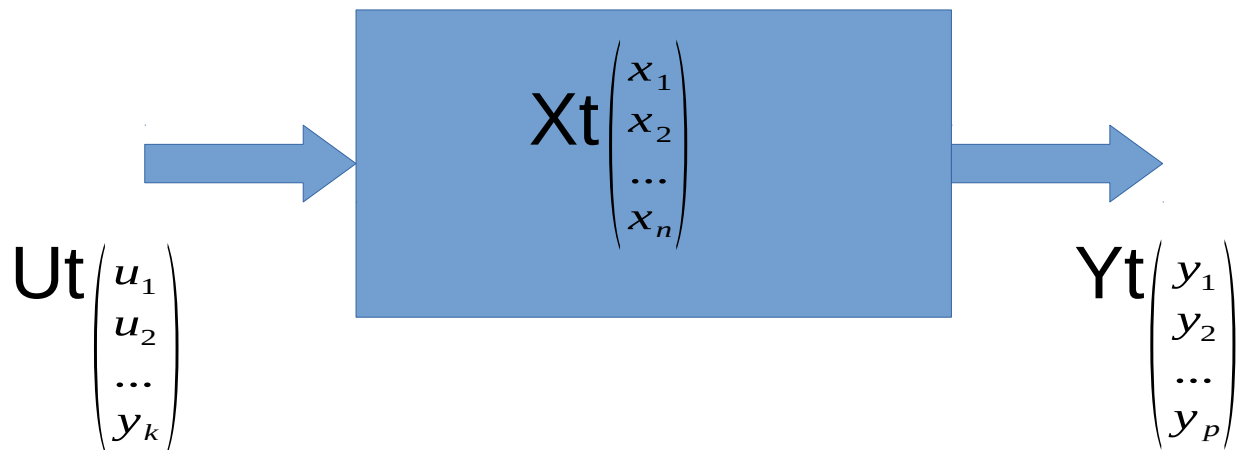


Dimension / Freedom Degrees



State Model Equation

Continuous case

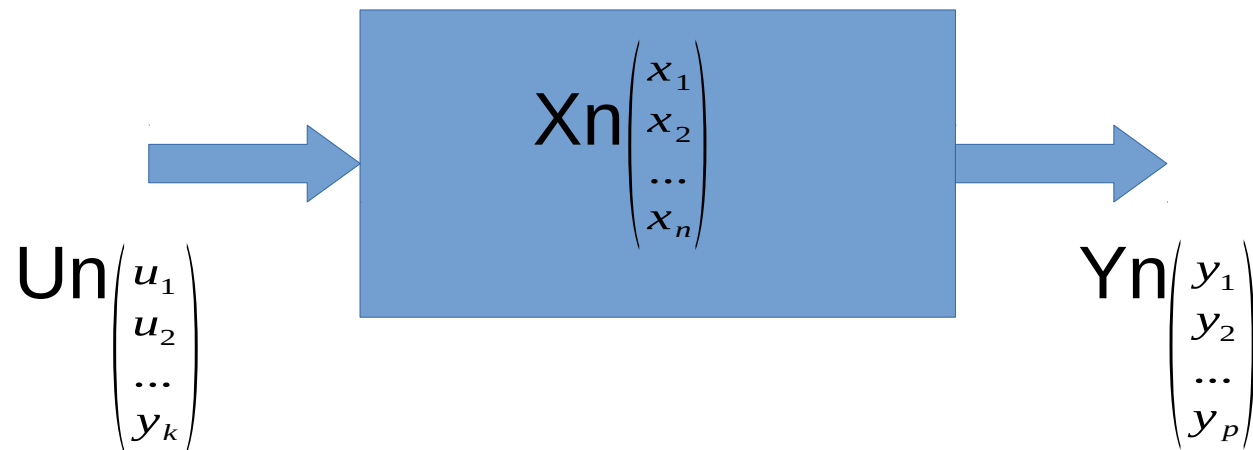


$$\dot{X}_t = \frac{\partial X_t}{\partial t} = f(X_t, U_t, W_t)$$

with $W_t \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_n \end{pmatrix}$: white noise (brownian)

State Model Equation

Discrete case



$$X_{n+1} = f(X_n, U_n, W_n)$$

with $W_n \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_n \end{pmatrix}$: white noise (normal / gaussian)

Time Discretisation

$$t_{n+1} = t_n + dt$$

$$\dots t_n = n \cdot dt + t_0$$

$$X_n = X(t_n)$$

$$\dots \dot{X}_t \cdot dx = X(t_{n+1}) - X(t_n) = X_{n+1} - X_n$$

Note on CPU Timers : for fixed delay => use Real Time linux or hardware ...

Typical loop timer: 10ms

State Model Differential Equation Order N => Order1 - Dimension N

$$\frac{\partial^n X_t}{\partial^n t} = f\left(\frac{\partial^{n-1} X_t}{\partial^{n-1} t}, \frac{\partial^{n-2} X_t}{\partial^{n-2} t}, \dots, X_t\right)$$

$$\text{let } X_t = \begin{pmatrix} X_t \\ \frac{\partial X_t}{\partial t} \\ \frac{\partial^2 X_t}{\partial^2 t} \\ \vdots \\ \frac{\partial^{n-1} X_t}{\partial^{n-1} t} \end{pmatrix} \quad \text{then } \dot{X}_t = \begin{pmatrix} \frac{\partial X_t}{\partial t} \\ \frac{\partial^2 X_t}{\partial^2 t} \\ \vdots \\ \frac{\partial^n X_t}{\partial^n t} \end{pmatrix} = \begin{pmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & & & \ddots & \\ 0 & & & 0 & 1 \end{pmatrix} \cdot \begin{pmatrix} X_t \\ \frac{\partial X_t}{\partial t} \\ \vdots \\ \frac{\partial^{n-1} X_t}{\partial^{n-1} t} \end{pmatrix} + \begin{pmatrix} 0 \\ \vdots \\ f(\dots, X_t) \end{pmatrix}$$

Which can be written as $\dot{X}_t = A \cdot X_t + B \cdot U_t$

where A=matrix(n,n) B=matrix(1,n) U_t=vector(n)

Example Order 2

Newtown Mecanic

$$m \cdot \vec{a} = \sum \vec{F}$$

Write with x,y coord:

$$\begin{pmatrix} m \ddot{x} = f_x \\ m \ddot{y} = f_y \end{pmatrix}$$

using:

$$X_t = \begin{pmatrix} x \\ y \\ \dot{x} \\ \dot{y} \end{pmatrix}$$

$$A = \begin{pmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{pmatrix}$$

$$B = \begin{pmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{pmatrix}$$

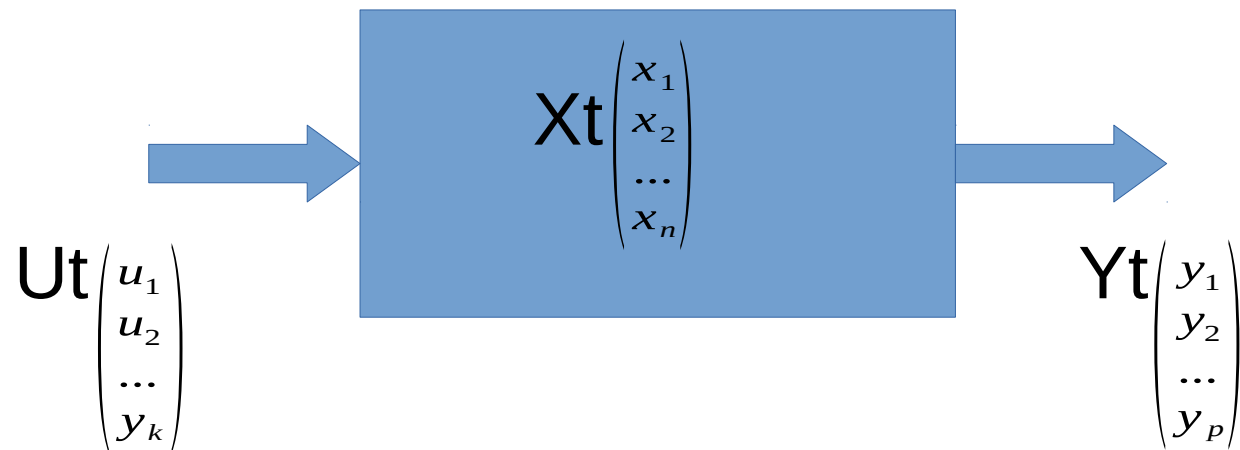
$$U = \begin{pmatrix} f_x/m \\ f_y/m \end{pmatrix}$$

Finally:

$$\dot{X}_t = A \cdot X_t + B \cdot U_t$$

Linear State Model Equation

Continuous case

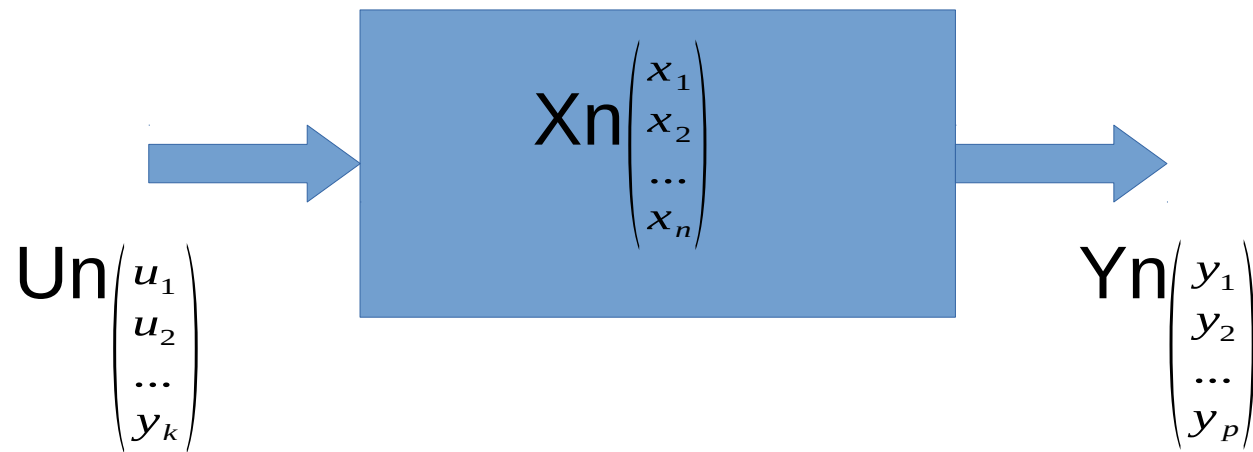


$$\dot{X}_t = A \cdot X_t + B \cdot U_t + D W_t$$

with $W_t \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_n \end{pmatrix}$: white noise (brownian)

Linear State Model Equation

Discrete case



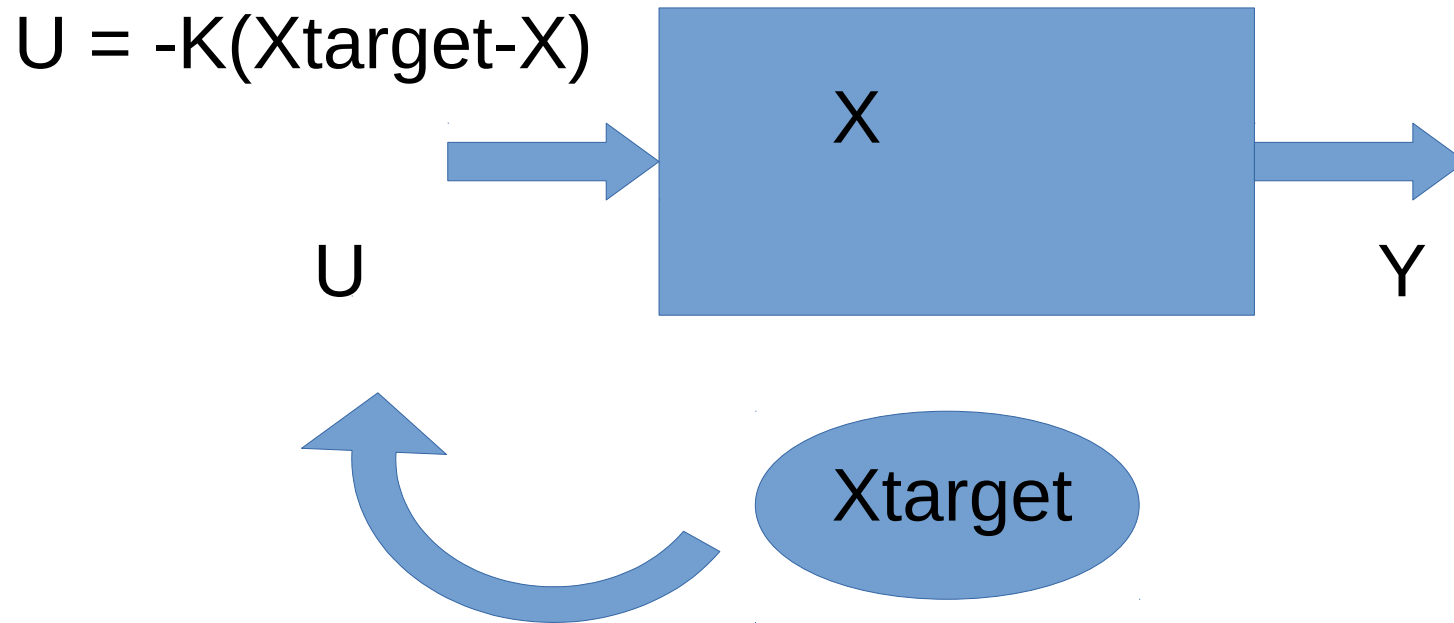
$$X_{n+1} = A \cdot X_n + B \cdot U_n + D \cdot W_n$$

With $W_n \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_n \end{pmatrix}$: white noise (normal / gaussian)

Outline

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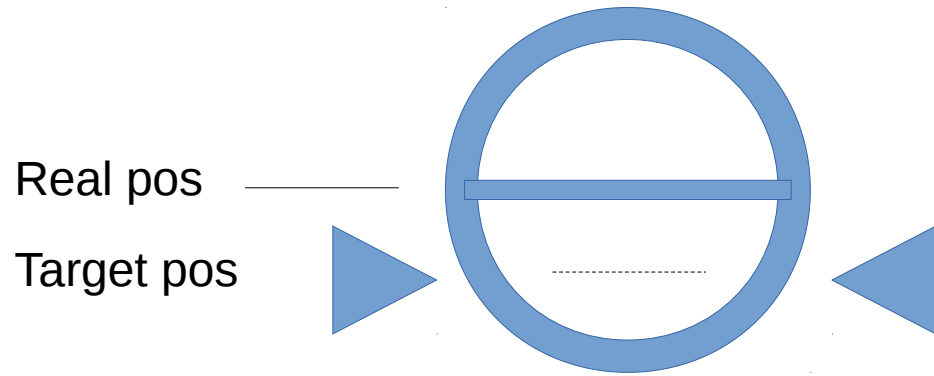
Feedback Loop : Follow Trajectory



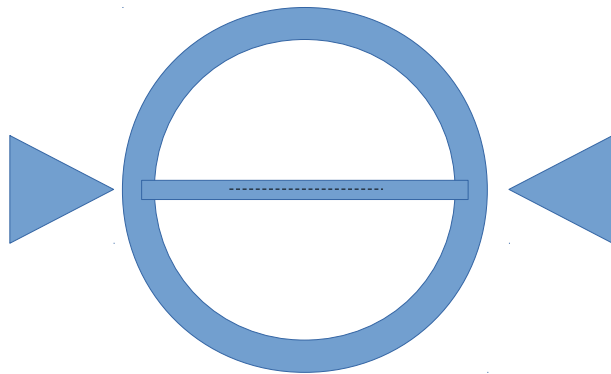
$X_{\text{target}} - X$ = target error to correct

K = feedback gain (= kalman control gain)

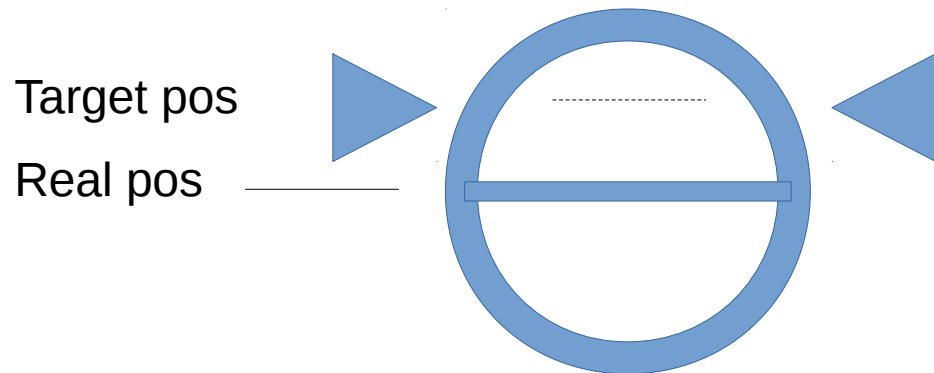
Example



Too HIGH \Rightarrow GO Down



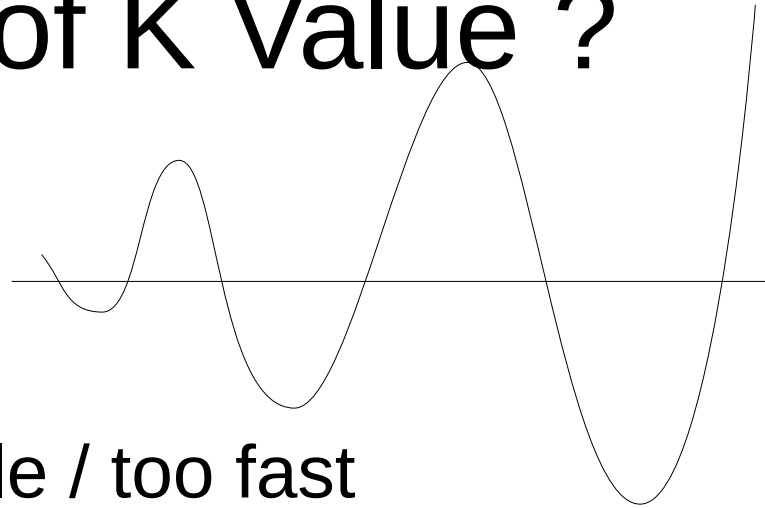
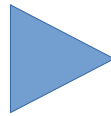
OK \Rightarrow don't move



Too LOW \Rightarrow GO Up

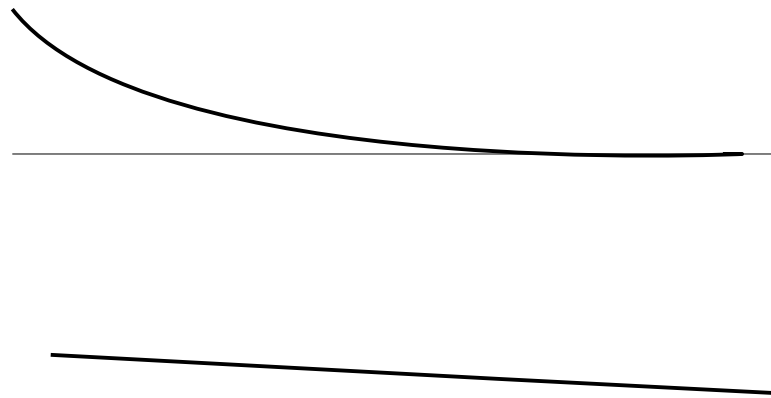
Choice of K Value ?

K Too BIG
($K > K_{max}$)

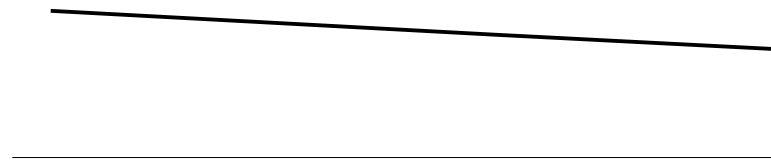


=> Divergent / unstable / too fast

K Optimal?
($K_{min} < K < K_{max}$)



K Too SMALL
($0 < K < K_{min}$)



=> Not convergent / not reactive / too slow

Response Time Shift

Optimal in Theory :

$$U_t = -K (X_{\text{target}} - X_t)$$

In practise ...

$$U_{t+\text{delayt}} = -K (X_{\text{target}} - X_t)$$

If delayt too big:

response time of computation > typical time of system

Then instability, vibration..

(example: you are drunk ... reaction > 100ms .. don't drive)

Achieve smallest delayt

Maybe split $K_1, K_2 \dots K_n$ with

$K_1 = \text{in hardware} = \text{nanos} / K_2: 10\text{ms} / K_3: 100\text{ms} \dots$

Computation for Optimal K ?

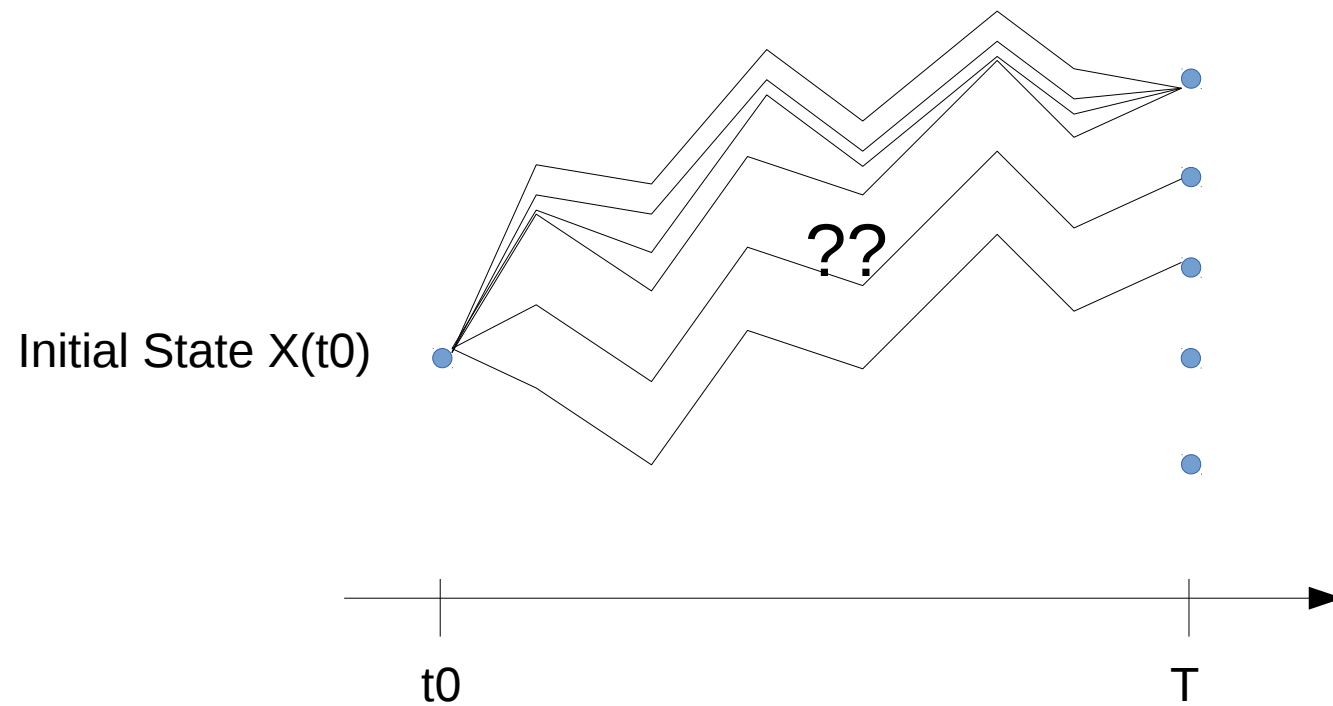
$$V(\text{trajectory}) = \int_{t_0}^T \underbrace{{}^t(X_t - X_{\text{target}_t})Q(X_t - X_{\text{target}_t})}_{\text{Term for cumulated error of trajectory following}} + \underbrace{{}^tU_t R U_t}_{\text{Term for cumulated energy of control}} dt$$

Term for cumulated
error of trajectory following

Term for cumulated
energy of control

Choose 2 symmetric matrices $Q=(n,n)$ & $R=(k,k)$

Compute for All Trajectories??



Dynamic Programming 1/3

Principle:

If $X_0 \rightarrow X_1 \rightarrow X_2 \rightarrow \dots \rightarrow X_n$ is the optimal trajectory from X_0 to any X_n

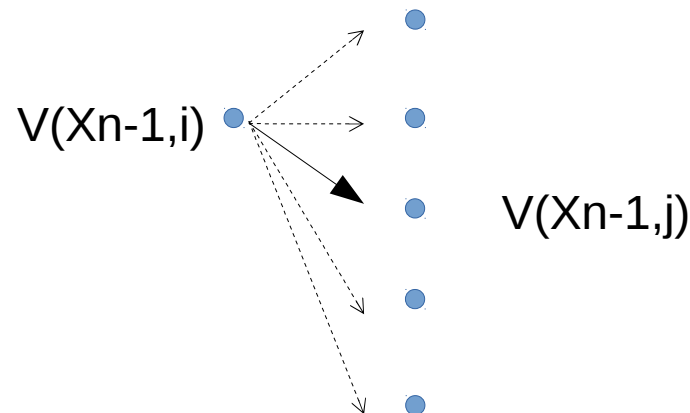
Then $X_i \rightarrow X_{i+1} \rightarrow \dots \rightarrow X_n$ is the optimal (sub) trajectory from X_i to any X_n

Compute $\Rightarrow X_{n-1} \rightarrow X_n$ the optimal last step for X_{n-1} to any X_n

Computation for last Step N:

Foreach K , compute

$$V(X_{n-1}, i) = \arg \min_j \text{cost}(X_{n-1}, i \rightarrow X_n, j) + V(X_n, j)$$

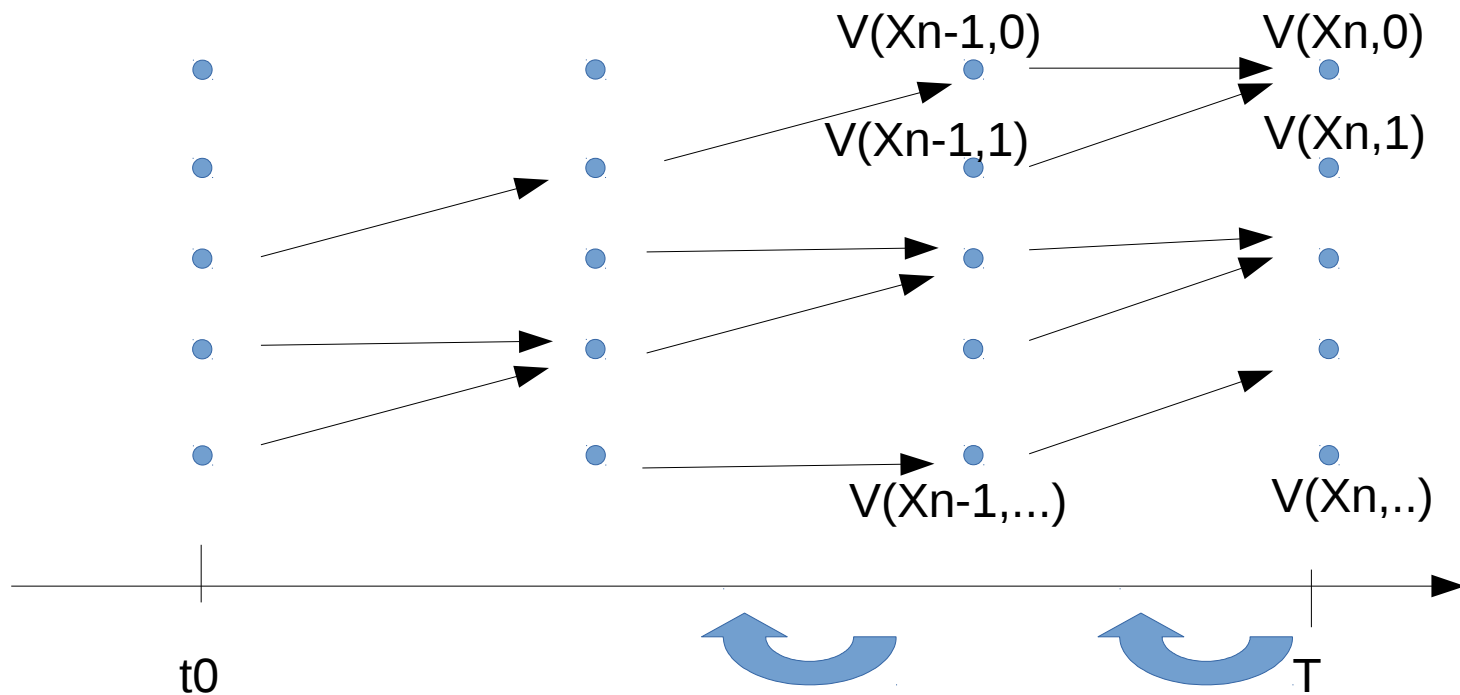


Algorithm similar to
Bellman-Kalaba “shortest path”
To any destinations

Dynamic Programming 2/3

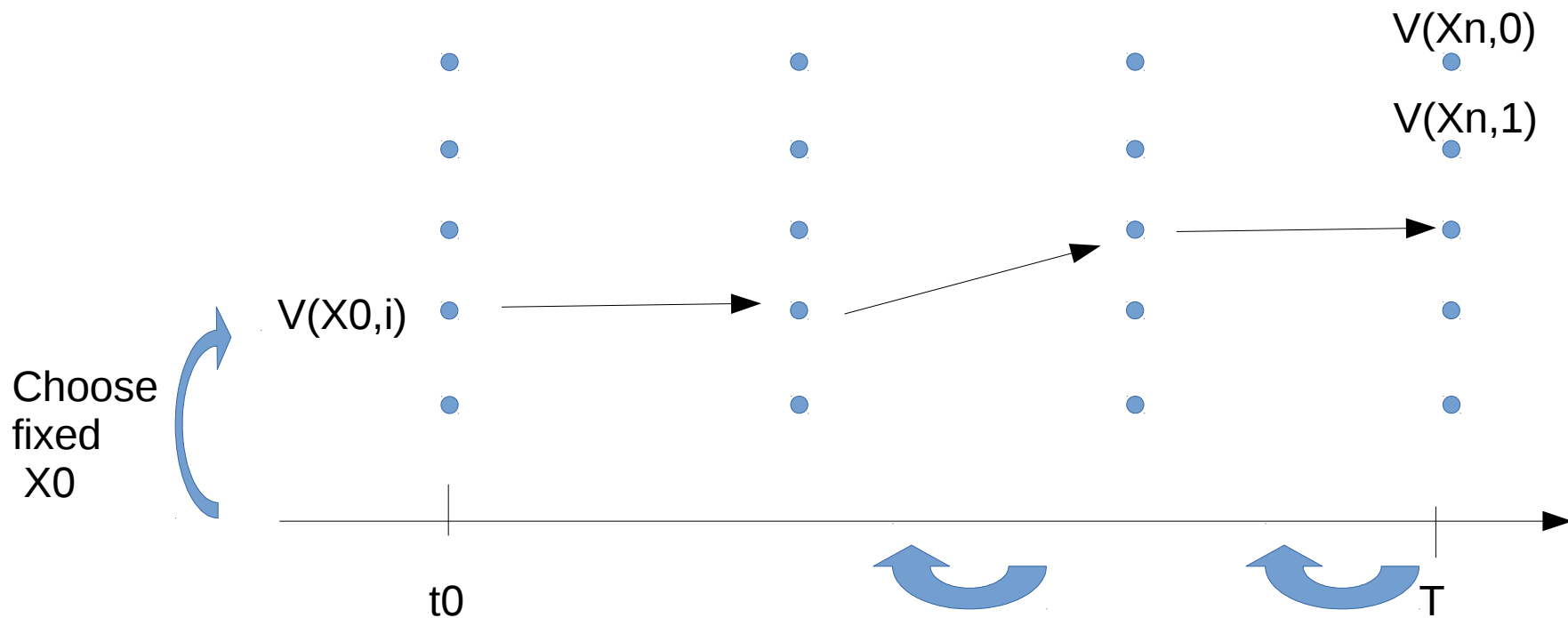
Recurse ... to get optimal Trajectories from any points:

- initialize $V(X_n, i)$
- recurse $n \rightarrow n-1$: compute $V(X_{n-1}, i)$... remember direction from $X_{n-1}, i \rightarrow X_n$



Dynamic Programming 3/3

Pick up optimal trajectory from t_0
(remembering each step directions $X_{n-1,i} \rightarrow X_n$)



Application Computation of Optimal Kalman Gain



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Algebraic Riccati equation

From Wikipedia, the free encyclopedia

An **algebraic Riccati equation** is a type of nonlinear equation that arises in the context of infinite-horizon [optimal control](#) problems in [continuous time](#) or [discrete time](#).

A typical algebraic Riccati equation is similar to one of the following:

the continuous time algebraic Riccati equation (CARE):

$$A^T X + XA - XBR^{-1}B^T X + Q = 0$$

or the discrete time algebraic Riccati equation (DARE):

$$X = A^T XA - (A^T XB)(R + B^T XB)^{-1}(B^T XA) + Q.$$

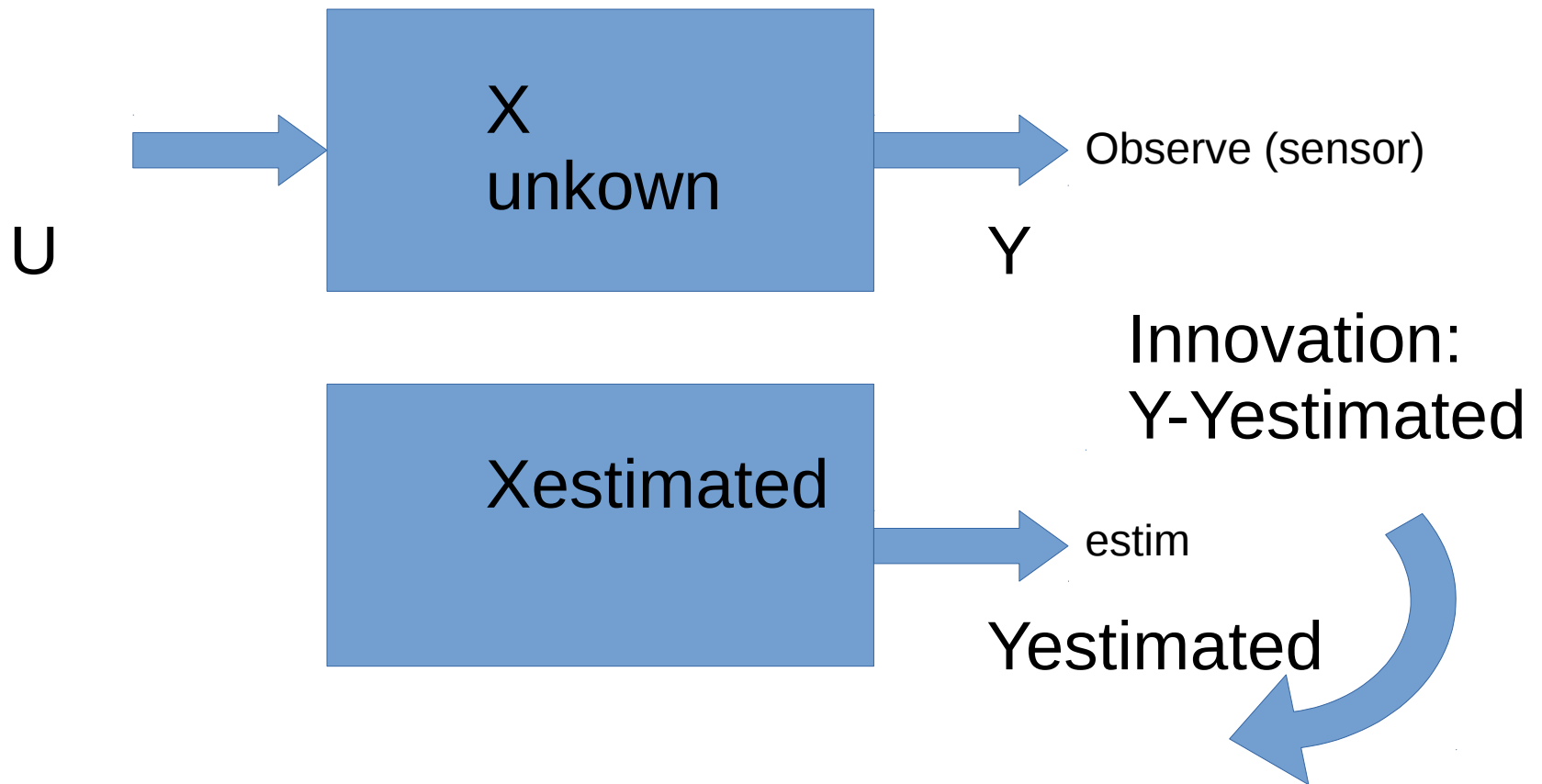
X is the unknown n by n symmetric matrix and A , B , Q , R are known [real](#) coefficient matrices.

It is called Riccati Equation (but nothing to do with Italian Mathematician)
This is from Kalman !

Outline

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Estimation Feedback Loop from Observations



Dimensions / Observation vs Degrees of Freedom

$$X_n \begin{pmatrix} x_1 \\ x_2 \\ \dots \\ x_n \end{pmatrix} \quad Y_n \begin{pmatrix} y_1 \\ y_2 \\ \dots \\ y_p \end{pmatrix}$$

In general $n > p$: not every variables are observables
 You measure only a projection

Example 1: $X_n \begin{pmatrix} x \\ y \\ z \end{pmatrix} \quad Y_n \begin{pmatrix} x \\ y \end{pmatrix}$ In 3D word ...only seeing 2D images

Example 2: $X_n \begin{pmatrix} x \\ \dot{x} \\ \ddot{x} \end{pmatrix} \quad Y_n \begin{pmatrix} \ddot{x} \end{pmatrix}$ On your cell-phone, you only have an accelerometer

Redundant Sensors Measures => merge for Accuracy

Same “measure” several times by different sensors
=> get different conflicting values for same variable !

=> use ponderations between sensors accuracy / speeds

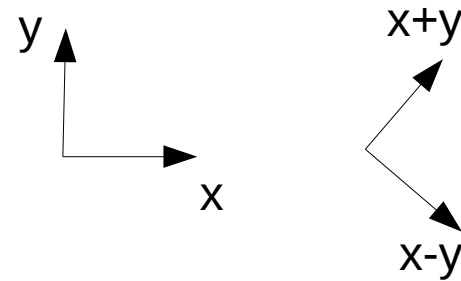
Example 1: $X_n \begin{pmatrix} x \\ y \end{pmatrix} \quad Y_n \begin{pmatrix} x \\ x+1 \\ 2x \\ 3x-1 \end{pmatrix}$

Example: Take several Pictures
(with different expositions)
=> merge pictures to remove noise

Deduce State from Linear Combination of Sensors ?

Example 1: $X_n \begin{pmatrix} x \\ y \end{pmatrix} \quad Y_n \begin{pmatrix} x+y \\ x-y \end{pmatrix}$

Measure in different basis
=> change basis (inverse basis matrix)



Example 2: $X_n \begin{pmatrix} x \\ y \end{pmatrix} \quad Y_n \begin{pmatrix} dist_{satellite\ 1} \\ dist_{satellite\ 2} \\ dist_{satellite\ 3} \\ dist_{satellite\ 4} \end{pmatrix}$

GPS :
Compute latitude/longitude position
from distances to Satellites...
in your cell-phone also

Deduce State From Current Sensor + N Past Sensor Measures

Mathematical Theorem on Linear Observable System ...

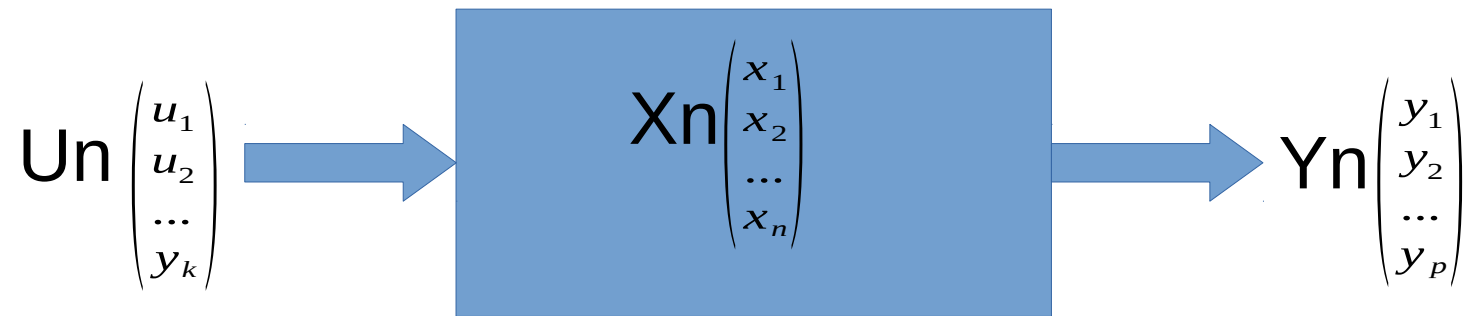
if ($AC, A^2C, A^3C, \dots A^nC$) has dimension N
 \Rightarrow then system is observable

Notice Transposition:

if ($AB, A^2B, A^3B, \dots A^nB$) has dimension N
 \Rightarrow then system is controllable

Observation Model Equation

Continuous / Discrete case



Discrete case :

$$Y_n = f(X_n, U_n, W_n)$$

with $W_n \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_n \end{pmatrix}$: white noise (normal / gaussian)

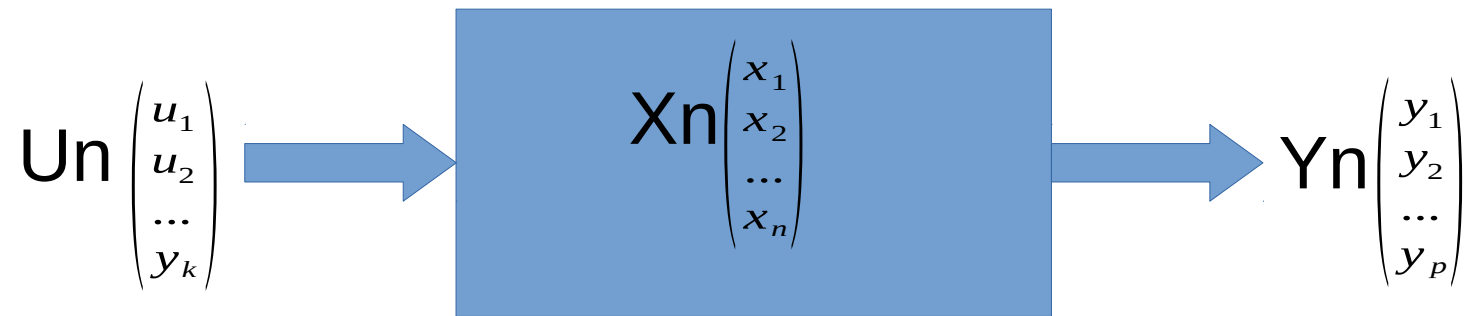
Continuous case :

$$Y_t = f(X_t, U_t, W_t)$$

with $W_t \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_n \end{pmatrix}$: white noise (brownian)

Linear Observation Model Equation

Continuous / Discrete case



$C = \text{matrix}(p, n)$, $F = \text{matrix}(p, k)$ in general $= 0$, $E = \text{matrix}(p, p)$

Discrete case :

$$Y_n = C X_n (+ F U_n) + E W_n$$

with $W_n \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_k \end{pmatrix}$: white noise (normal / gaussian)

Continuous case :

$$Y_t = C X_t (+ F U_t) + E W_t$$

with $W_t \begin{pmatrix} w_1 \\ w_2 \\ \dots \\ w_k \end{pmatrix}$: white noise (brownian)

Sensor (so Measures) have many Weaknesses !

White Noise : random additive noise each time you measure

Other weaknesses :

Model Error ... what you measure is not really what you think, your equation are wrong

Biased Noise : the average measure is shifted compared to real
you need to re-calibrate your sensor

Discrete Sampling Time : your electronic sensor is not very fast ...example every 10ms

Sampling Not Strictly Regular : every 10ms? but sometime 9ms, sometime 11ms

Integer Rounding (=Quantification) : your Analog To Digital converter is only precise
to 10 bits for instance (= 1024 values)

Time Delay : you get results 5ms late after they are really measured

Missing/Irregular Measures Arrival : sometime you don't have measure,
or measures are not regular at all (example: satellites)

Handling Irregular Measures

Example of handling “Missing/Irregular Measures Arrival”

When receiving measures :

measure set1 => then use equation1

$$Y_n = C_1 X_n + E_1 W_n$$

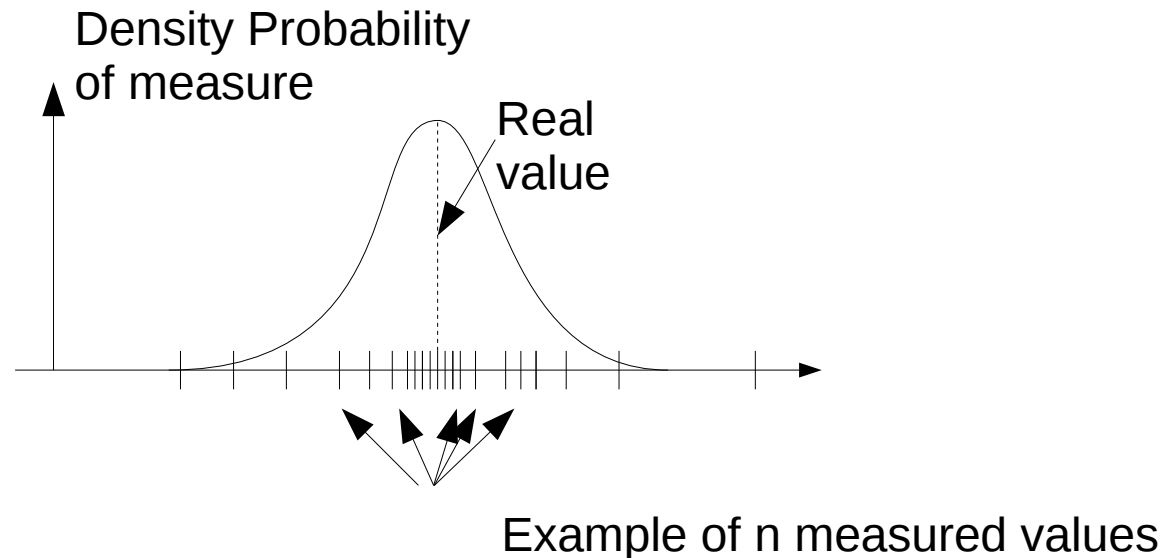
measure set2 => then use equation2

$$Y_n = C_2 X_n + E_2 W_n$$

measure set3 => then use equation3

$$\dots$$

Measures Noise term: $E W_n$

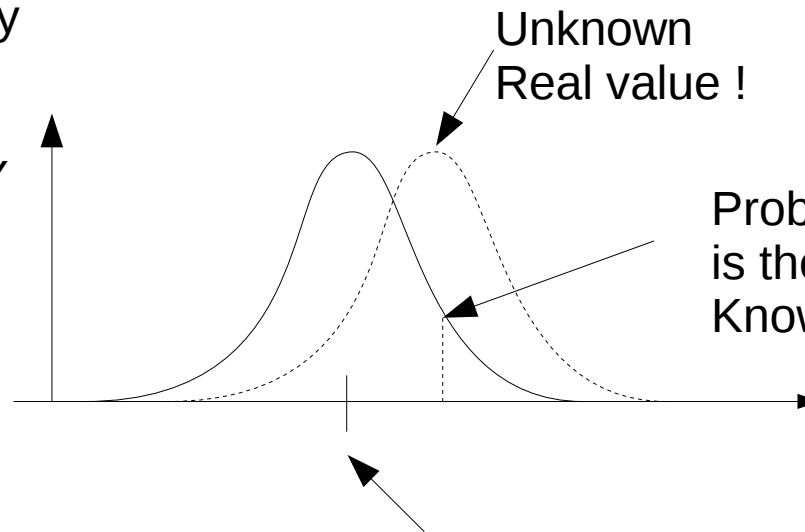


(White) Noise =
repeating N measures give N different values
following a Gaussian distribution

Biased = when the average is shifted

Noise Measure \Rightarrow Real Value?

Density Probability
of real value
Knowing
noised measure Y



Probability that this shifted value
is the Real value
Knowing the (noised) measure Y

Example of 1 measured value : Y

Merge N noised measures : Average

$$\textit{Average}(Y_{1..n}) = \frac{1}{N} (Y_1 + Y_2 + \dots Y_n)$$

If all measures Y_i have precisions $\pm\sigma$

Then Average(Y) has precision $\pm \frac{\sigma}{\sqrt{N}}$

(Theorem called “Law of Big Numbers”, for gaussians distributions)

Probability Expectation (synonym: Average, Mean)

$E(Y)$ = Expected of Y = average value of Y
... for all random events
weighed by their probabilities

Discrete case: given $(Y_1, P_1), (Y_2, P_2) \dots (Y_n, P_n)$
where Y_i : value of event i with probability P_i ,

$$E(Y) = \sum_i P_i Y_i$$

Continuous case: given density probability $(Y(x), P(x)dx)$
where Y_x value for $x < x+dx$ with probability $P(x)$

$$E(Y) = \int_{x=-\infty}^{+\infty} Y(x) P(x) dx$$

Probability Variance (=centered moment of order 2)

$V(Y)$ = Variance of Y = weighted average value of
square of centered ($Y-E(Y)$)
... for all random events

$$\text{Var}(Y) = \sum_i P_i \bar{Y}_i^2 \dots \text{where } \bar{Y}_i = Y_i - E(Y)$$

$$\text{Var}(Y) = \int_{x=-\infty}^{+\infty} Y(\bar{x})^2 P(x) dx$$

Var is moment of order 2

$E(Y)$ was moment of order 1

... we can also define moment 3, 4 ...

Standard Deviation (= square root of variance)

By definition:

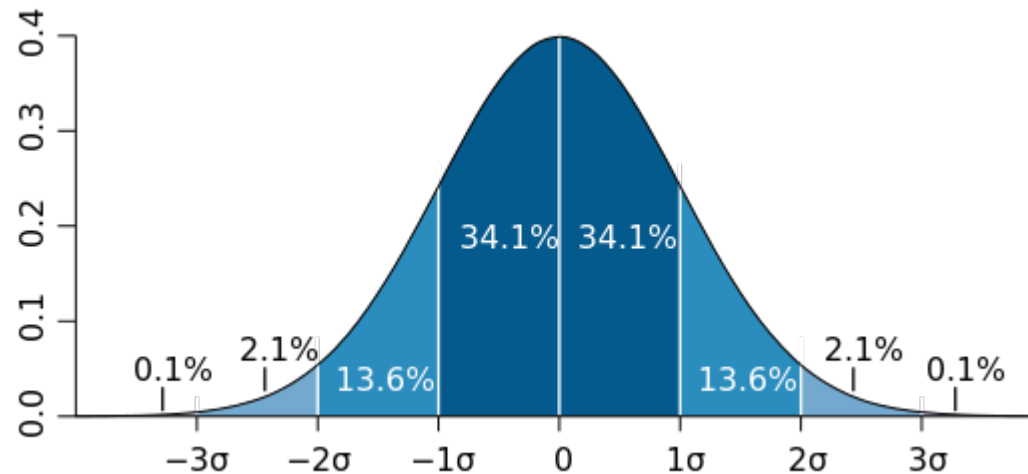
$$\mathit{Var} = \sigma^2$$

Equivalent:

$$\sigma = \sqrt{\mathit{Var}}$$

Standard Deviation of X has the same unit as X , $E(X)$
(if X is a distance, stddev is also distance)

Standard Deviation for Normal Distribution (Gaussian)



Probability ~68% that value in $[y - \sigma, y + \sigma]$

Probability ~95% that value in $[y - 2\sigma, y + 2\sigma]$

Probability ~99.7% that value in $[y - 3\sigma, y + 3\sigma]$

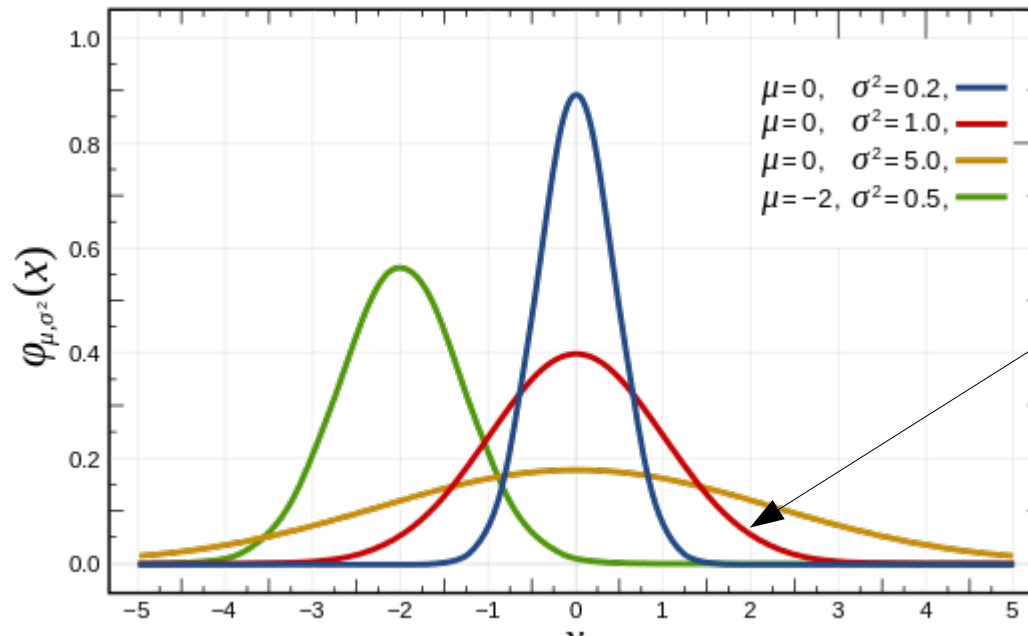
Gaussian Distribution

... defined by (Expect, StdDev)
Normale when StdDev=1



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$$f(x; \mu, \sigma^2) = \frac{1}{\sigma\sqrt{2\pi}} e^{-\frac{1}{2}\left(\frac{x-\mu}{\sigma}\right)^2}$$



Normale when
 $\sigma = 1$

Central Limit Theorem (~ Law of Big Numbers ...)

Theorem:

Given N random independent measures $Y_1, Y_2 \dots Y_n$

Then centered average multiplied by \sqrt{N}
Converges to gaussian distribution $N(0, \sigma^2)$

$$\sqrt{n} \bar{Y}_{1..n} \rightarrow N(0, \sigma^2)$$

... For large enough N ,
the N -average follow a gaussian $N(\mu, \frac{\sigma}{\sqrt{N}})$

Back To Measures Merge...

Sensor Merge point-of-view:

Suppose you have 2 measures Y_1 , Y_2
with different stddeviations

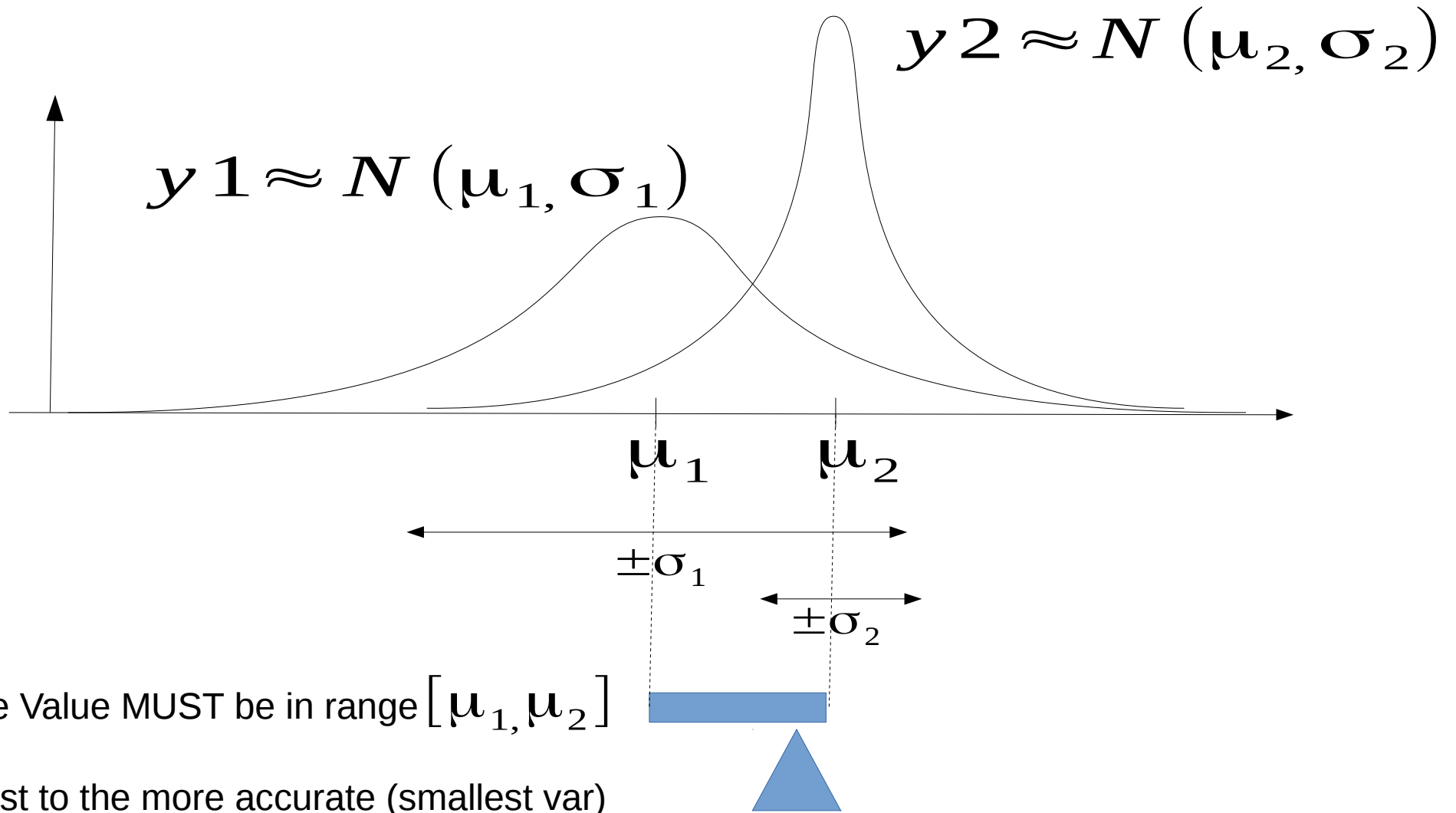
=> how to “weight-average” the 2 values ?

Kalman Filter point-of-view:

Suppose you have a-priori estimated prediction Y_1
and a measure Y_2

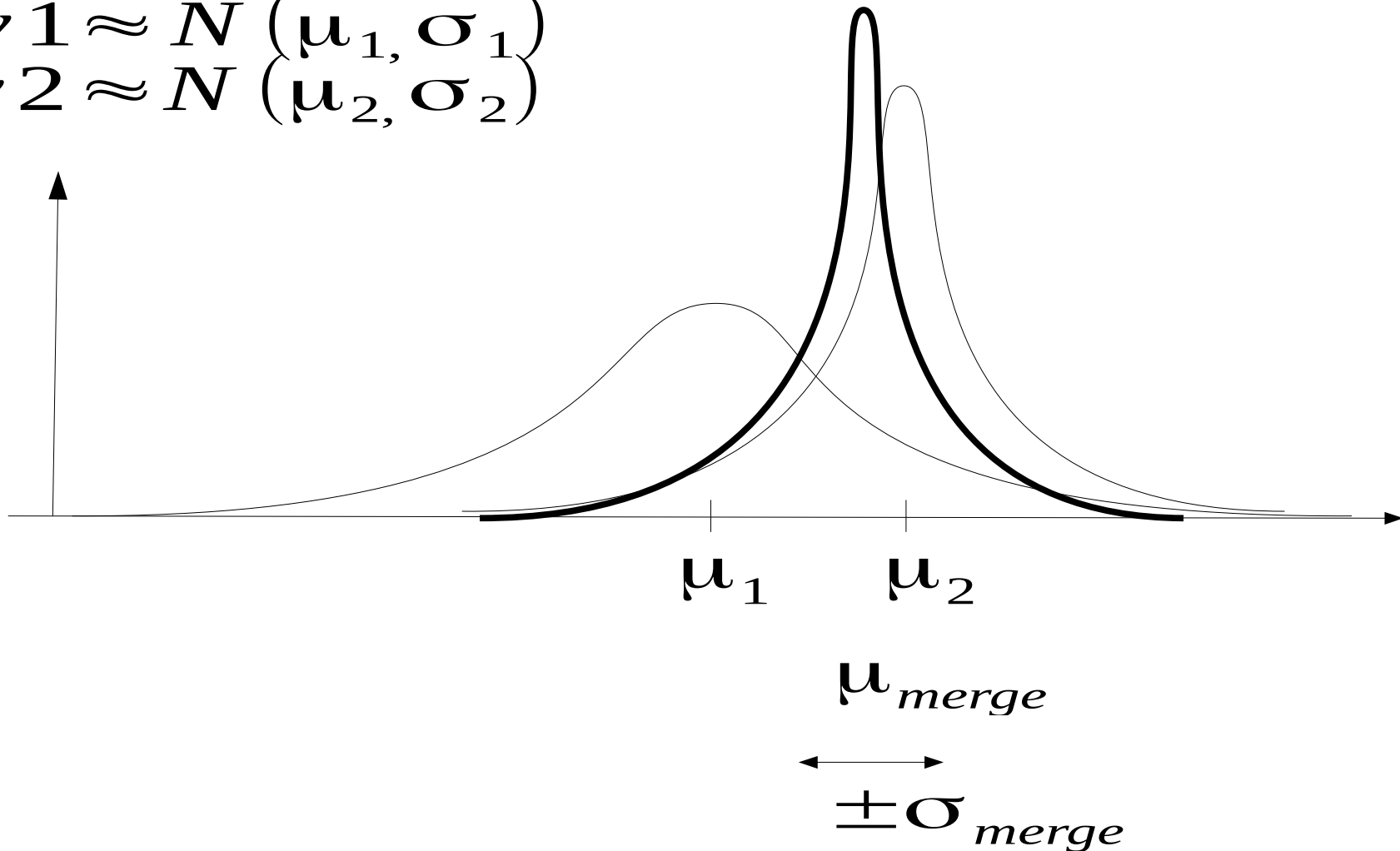
=> how to “weight-average” the 2 values ?

Merge 2 Gaussians ...



Merge (multiply) 2 Gaussians => proportional to a Gaussian

$$\begin{aligned} y_1 &\approx N(\mu_1, \sigma_1) \\ y_2 &\approx N(\mu_2, \sigma_2) \end{aligned}$$



Probability of 2 independent Events => Product of Probability

If A and B are two INDEPENDENT random events,
Then the probability of simultaneous event “A and B” is

$$P(A \wedge B) = P(A) \cdot P(B)$$

Conditional probability (not independent): $P(A|B) = \frac{P(A \wedge B)}{P(B)}$

Y1 and Y2 are both dependent of Yreal

$$P(Y_1 \wedge Y_2 | Y_{real}) = P(Y_1 | Y_{real}) \cdot P(Y_2 | Y_{real})$$

Computing Weigth for product of 2 Gaussians

Remember... $e^n = \underbrace{e \cdot e \dots e}_n$ $e^a \cdot e^b = e^{a+b}$

So

$$Y_1(x) \cdot Y_2(x) = \left(\frac{1}{\sqrt{2\pi}\sigma_1} e^{-\frac{(x-\mu_1)^2}{2\sigma_1^2}} \right) \left(\frac{1}{\sqrt{2\pi}\sigma_2} e^{-\frac{(x-\mu_2)^2}{2\sigma_2^2}} \right)$$

$$Y = k e^{-\frac{1}{2\sigma_1^2\sigma_2^2}(\sigma_2^2(x-\mu_1)^2 + \sigma_1^2(x-\mu_2)^2)}$$

Quadratic form ax^2+bx+c

$$= (\sigma_1^2 + \sigma_2^2) x^2 - 2(\sigma_2^2 \mu_1 + \sigma_1^2 \mu_2) x + \dots$$

$$= (\sigma_1^2 + \sigma_2^2) \left(x - \frac{\sigma_2^2 \mu_1 + \sigma_1^2 \mu_2}{\sigma_1^2 + \sigma_2^2} \right)^2 + \dots$$

Rewrite $a(x+b/2a)^2 + \dots$

Result Gaussian Product

$$Y_1(\mu_1, \sigma_1) \cdot Y_2(\mu_2, \sigma_2) = cst \cdot e^{-\frac{1}{2} \frac{\sigma_1^2 + \sigma_2^2}{\sigma_1^2 \sigma_2^2} \left(x - \frac{\sigma_2^2 \mu_1 + \sigma_1^2 \mu_2}{\sigma_1^2 + \sigma_2^2} \right)^2}$$

$$= cst \cdot Y(\mu, \sigma)$$

With

$$\mu = \frac{\sigma_2^2 \mu_1 + \sigma_1^2 \mu_2}{\sigma_1^2 + \sigma_2^2}$$

$$\frac{1}{\sigma^2} = \frac{\sigma_1^2 + \sigma_2^2}{\sigma_1^2 \sigma_2^2}$$

Re-interpret Expectation Weighted Sum

$$\mu = \frac{\sigma_2^2 \mu_1 + \sigma_1^2 \mu_2}{\sigma_1^2 + \sigma_2^2} \Rightarrow \mu = \underbrace{\frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2}}_{\text{weight for } \mu_1} \mu_1 + \underbrace{\frac{\sigma_1^2}{\sigma_1^2 + \sigma_2^2}}_{\text{Complementary weight for } \mu_2} \mu_2$$

weight for μ_1

Complementary weight for μ_2

Weights are in $[0, 1]$
(0% to 100%)

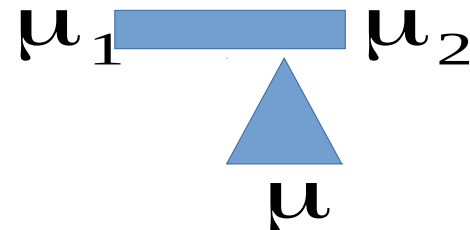
$$0 \leq \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} \leq 1$$

$$0 \leq \frac{\sigma_1^2}{\sigma_1^2 + \sigma_2^2} \leq 1$$

Sum of both Weights = 1
(sum is 100%)

$$\frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} + \frac{\sigma_1^2}{\sigma_1^2 + \sigma_2^2} = 1$$

Result value is in range $[\mu_1, \mu_2]$



Closest to the more accurate (smallest var)

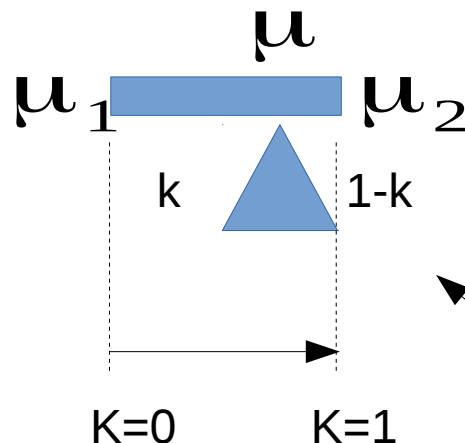
Re-interpret weight as Kalman Gain

posing

$$K = \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2}$$

$$\mu = K \mu_1 + (1 - K) \mu_2$$

Choose Letter “K” in loving memory of Rudolph E Kalman



When $\sigma_1 \ll \sigma_2$

Y2 not Accurate... Y1 better

=> more weight on Y2

$K \sim 1$ $1-K \sim 0$

=> closest to μ_2

When $\sigma_1 \gg \sigma_2$

Y1 not Accurate... Y2 better

=> more weight on Y2

$K \sim 0$ $1-K \sim 1$

=> closest to μ_2

$K = \text{remember } Y1 / \text{correct } Y2 \text{ ratio}$

Re-interpret Standard Deviation (variance)

$$\frac{1}{\sigma^2} = \frac{\sigma_1^2 + \sigma_2^2}{\sigma_1^2 \sigma_2^2} \quad \Rightarrow \quad \sigma^2 = \frac{\sigma_1^2 \sigma_2^2}{\sigma_1^2 + \sigma_2^2}$$

Also see as Sum of inverse of square

$$\frac{1}{\sigma^2} = \frac{1}{\sigma_1^2} + \frac{1}{\sigma_2^2}$$

Remarks:

If $\sigma_1 = \sigma_2 = \sigma_i$ then $\frac{1}{\sigma^2} = \frac{2}{\sigma_i^2}$ so $\sigma = \frac{\sigma_i}{\sqrt{2}}$

Remember... Big Numbers Law (with $N=2!$) : $\text{Average}(Y_1 \dots Y_N) \rightarrow N \left(\mu, \frac{\sigma}{\sqrt{N}} \right)$

Re-interpret StdDev with K

$$\left. \begin{aligned} \sigma^2 &= \frac{\sigma_1^2 \sigma_2^2}{\sigma_1^2 + \sigma_2^2} \\ K &= \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} \end{aligned} \right\} \Rightarrow \sigma^2 = K \sigma_1^2 = (1 - K) \sigma_2^2$$

If Y1 is the a-priori estimation
and Y2 is a measurement value

Then the new variance of a-posteriori estimation
is reduced by a factor (1-K)

