

1 General Principles

Let $S(t)$ be a 1-dimensional time signal (sound) to be compressed. It is discretized in time by sampling at regular frequency (example 8000 Hz): $t_0, t_1, t_2 \dots t_N$
 $t_n = t_0 + n * \tau$, $\tau = 1/F$

Each signal measure $S_n = s(t_n)$ is converted from analogic to digital, so discretized and encoded to a fixed integer precision (example: 16 bits).

1.1 different measures of residu

While decomposing a signal S into main components H_k : $S = \sum_k H_k + R$, the residual signal $R = S - \sum_k H_k$ must be as small as possible: $\lim_{k \rightarrow \infty} \|R\| = 0$
 Signal Entropy:

$$E(S) = \sum_n p(S_n) \log(S_n) \quad (1)$$

where $p(S_n)$ is the probability of obtaining value S_n , and $\log(S_n)$ is the number of bits (log in base 2) for encoding value S_n .

Signal Quadratic Variance:

$$Var(S) = \|S\|^2 = \sum_n S_n^2 \quad (2)$$

$Var(S)$ represents the norm of the signal = distance to 0. It is linked to scalar product $\langle f, g \rangle = \sum_n f_n g_n$.

Signal Absolute Area:

$$AbsArea(S) = \sum_n abs(S_n) \quad (3)$$

$AbsArea(S)$ represents the total area of the signal with the x-axis.

Signal Maximum Range:

$$MaxRange(S) = \max_n S_n - \min_n S_n \quad (4)$$

$MaxRange(S)$ represents the height of the horizontal band containing the signal (for a symmetric signal around 0, it is the max distance to 0).

2 Linear Autocorrelation

$\tilde{(X_t)}$ linear prediction for X_t knowing X_{t-1} :

$$\tilde{(X_t)} = AX_{t-1} + B_t, X_t = \begin{pmatrix} x_t \\ x_{t-1} \\ \vdots \\ x_{t-P} \end{pmatrix}, A = \begin{pmatrix} a_1 & a_2 & a_3 & \dots & a_n \\ 1 & 0 & \dots & & \\ 0 & 1 & 0 & & \\ \vdots & & \ddots & & \end{pmatrix} \quad (5)$$

The error between real and predicted value is $\tilde{X}_t - X_t = \begin{pmatrix} x_t - \sum_n a_n x_{t-n} - B_t \\ 0 \\ 0 \\ 0 \end{pmatrix}$

The quadratic error is

$$\begin{aligned}
Q &= \text{Var}(\tilde{X} - X) = \sum_n (\tilde{X}_n - X_n)^2 \\
&= \sum_t (x_t - \sum_n a_n x_{t-n} - B_t) \cdot (x_t - \sum_n a_n x_{t-n} - B_t) \\
&= \sum_t \left((x_t - B_t)^2 - 2(x_t - B_t) \cdot \left(\sum_n a_n x_{t-n} \right) + \left(\sum_{n_1} a_{n_1} x_{t-n_1} \right) \cdot \left(\sum_{n_2} a_{n_2} x_{t-n_2} \right) \right) \\
&= \sum_{n_1, n_2} a_{n_1} a_{n_2} \left(\sum_t x_{t-n_1} x_{t-n_2} \right) \\
&\quad - 2 \sum_n a_n \left(\sum_t (x_t - B_t) x_{t-n} \right) \\
&\quad + \left(\sum_t (x_t - B_t)^2 \right)
\end{aligned} \tag{6}$$

This is a quadratic form of term a_{ij} . It can be written as

$$Q(a) = Q^0 - 2N^t a + a^t M a \tag{7}$$

$$\text{with } M = \begin{pmatrix} & \vdots \\ \dots & \sum_t x_{t-i} x_{t-j} \end{pmatrix}, \quad N = \begin{pmatrix} \vdots \\ \sum_t (x_t - B_t) x_{t-i} \\ \vdots \end{pmatrix}, \quad Q^0 =$$

$$\sum_t (x_t - B_t)^2$$

A simple linear calculation, analogue to quadratic scalar form $ax^2 + bx + c$... gives min for $x = -b/2a$, with value $= c - b^2/2a$.

If M is invertible, the minimum of $Q(a)$ is obtained for $a = M^{-1}N$, and has value $Q^0 - N^t M^{-1}N$. If M is not invertible, a similar result can be obtained with the pseudo inverse.

Note that the result can be interpreted by saying that the variance has reduced by $N^t M^{-1}N$ when applying the autocorrelation procedure.

3 Least Square Method for Harmonics Amplitude

The signal is decomposed in K main harmonics, with a residu: $S(t_n) = \sum_k S_k(t_n) + R(t)$

We supposed we know frequencies w_k and phases ϕ_k (example: by using zero-crossing algorithms, see next), we want to adjust optimal amplitude coefficient c_k to minimize quadratic errors.

For calculation with complex $H_k = c_k \cos(w_k t + \phi_k)$ will be replaced by $H_k = c_k e^{i(w_k t + \phi_k)}$, and $\|x\|^2 = x\bar{x}$

$$\begin{aligned} \|R\|^2 &= \sum_n \|S_n - \sum_k H_k(t_n)\|^2 \\ &= \sum_n \left(S_n - \sum_{k_1} H_{k_1}(t_n) \right) \left(S_n - \sum_{k_2} \bar{H}_{k_2}(t_n) \right) \\ &= \underbrace{\left(\sum_n S_n^2 \right)}_c - \underbrace{\left(\sum_k \left(\sum_n S_n (H_k + \bar{H}_k)(t_n) \right) \right)}_{b_k} + \underbrace{\left(\sum_{k_1, k_2} \sum_n H_{k_1} \bar{H}_{k_2}(t_n) \right)}_{a_{k_1 k_2}} \end{aligned}$$

Using $H_k(t_n) = c_k e^{i(w_k t + \phi_k)}$

it follows $H_k + \bar{H}_k(t) = 2c_k \cos(w_k t + \phi_k)$ (is real, not complex), and

$$b_k = 2c_k \sum_n S_n \cos(w_k t + \phi_k)$$

Then for $a_{k_1 k_2}$, $H_{k_1} \bar{H}_{k_2}(t) = c_{k_1} c_{k_2} e^{i((w_{k_1} - w_{k_2})t + (\phi_{k_1} - \phi_{k_2}))}$

summing twice, complex part of $\dots(k_1 - k_2) + \dots(k_2 - k_1)$ give zero

$$\frac{1}{2}(H_{k_1} \bar{H}_{k_2} + H_{k_2} \bar{H}_{k_1})(t) = \frac{1}{2} c_{k_1} c_{k_2} 2 \cos((w_{k_1} - w_{k_2})t + (\phi_{k_1} - \phi_{k_2}))$$

so

$$a_{k_1 k_2} = c_{k_1} c_{k_2} \sum_n \cos((w_{k_1} - w_{k_2})t_n + \phi_{k_1} - \phi_{k_2})$$

($a_{k_1 k_2}$ is a real symmetric positive matrix)

Finally,

$$\|R\|^2 = \underbrace{\left(\sum_n S_n^2 \right)}_{R^0} - 2 \sum_k c_k \underbrace{\left(\sum_n S_n \cos(w_k t + \phi_k) \right)}_{B_k} + \sum_{k_1, k_2} c_{k_1} c_{k_2} \underbrace{\left(\sum_n \cos((w_{k_1} - w_{k_2})t_n + \phi_{k_1} - \phi_{k_2}) \right)}_{A_{k_1 k_2}} \quad (8)$$

This is a quadratic form on vector variable c_k :

$$\|R\|^2 = R^0 - 2B^t c + c^t A c \quad (9)$$

The minimum is reached for $c = A^{-1}B$, and the minimum is $R^0 - B^t A^{-1}B$

Note that the result can be interpreted by saying that the variance has reduced by $B^t A^{-1}B$ when applying the least-square amplitude fitting.

4 Least Square Method for Harmonics Amplitude Linear Perturbation

The signal is decomposed in K main harmonics, with a residu: $S(t_n) = \sum_k S_k(t_n) + R(t)$

We have known approximations for frequencies w_k , phases ϕ_k and amplitudes c_k .

The K^{th} harmonic $S_k(t) = c_k \cos(w_k t + \phi_k)$ is replaced by modifying the constant amplitude c_k , to obtain a “non-periodic harmonic”: $c_k(t) = c_k^0 + c_k^1(t - t_0)$

This model is usable only for short-time interval!

The calculation done in previous section is slightly modified.

$$\begin{aligned} ||R||^2 = & \left(\sum_n S_n^2 \right) \\ & - 2 \sum_k \left(\sum_n c_k(t_n) S_n \cos(w_k t_n + \phi_k) \right) \\ & + \sum_{k_1, k_2} \left(\sum_n c_{k_1}(t_n) c_{k_2}(t_n) \cos((w_{k_1} - w_{k_2})t_n + \phi_{k_1} - \phi_{k_2}) \right) \end{aligned} \quad (10)$$

We want to expand $c_k(t) = c_k^0 + c_k^1(t - t_0)$, then factorize the variance as a quadratic form on vector term c_k^1

For ease of calculation, lets note $COS_{kn} = \cos(w_k t_n + \phi_k)$ and $COS\Delta_{k_1 k_2 n} = \cos((w_{k_1} - w_{k_2})t_n + \phi_{k_1} - \phi_{k_2})$

$$\begin{aligned} ||R||^2 = & \left(\sum_n S_n^2 \right) \\ & - 2 \sum_k \sum_n (c_k^0 + c_k^1(t_n - t_0)) S_n COS_{kn} \\ & + \sum_{k_1, k_2} \left(\sum_n (c_{k_1}^0 + c_{k_1}^1(t_n - t_0))(c_{k_2}^0 + c_{k_2}^1(t_n - t_0)) COS\Delta_{k_1 k_2 n} \right) \\ = & \sum_n S_n^2 \\ & - 2 \sum_k c_k^0 \sum_n S_n COS_{kn} - 2 \sum_k c_k^1 \left(\sum_n (t_n - t_0) S_n COS_{kn} \right) \\ & + \sum_{k_1, k_2} \left(\sum_n (c_{k_1}^0 c_{k_2}^0 + (c_{k_1}^0 c_{k_2}^1 + c_{k_1}^1 c_{k_2}^0)(t_n - t_0) + c_{k_1}^1 c_{k_2}^1 (t_n - t_0)^2) COS\Delta_{k_1 k_2 n} \right) \\ = & \sum_n S_n^2 \end{aligned}$$

$$\begin{aligned}
& -2 \sum_k c_k^0 \sum_n S_n \cos S_{kn} - 2 \sum_k c_k^1 \left(\sum_n (t_n - t_0) S_n \cos S_{kn} \right) \\
& + \sum_{k_1, k_2} \left(\sum_n (c_{k_1}^0 c_{k_2}^0) \cos \Delta_{k_1 k_2 n} \right) \\
& + \sum_{k_1, k_2} \left(\sum_n \underbrace{(c_{k_1}^0 c_{k_2}^1 + c_{k_1}^1 c_{k_2}^0)}_{=2c_{k_1}^1 c_{k_2}^0} (t_n - t_0) \cos \Delta_{k_1 k_2 n} \right) \\
& + \sum_{k_1, k_2} \left(\sum_n (c_{k_1}^1 c_{k_2}^1 (t_n - t_0)^2) \cos \Delta_{k_1 k_2 n} \right)
\end{aligned} \tag{11}$$

Then. . .

$$\begin{aligned}
\|R\|^2 = & \underbrace{\left(\sum_n S_n^2 - 2 \sum_k c_k^0 \sum_n S_n \cos S_{kn} + \sum_{k_1, k_2} c_{k_1}^0 c_{k_2}^0 \sum_n \cos \Delta_{k_1 k_2 n} \right)}_{R^0} \\
& - 2 \sum_k c_k^1 \underbrace{\left(\sum_n (t_n - t_0) S_n \cos S_{kn} - \sum_{k_2} c_{k_2}^0 \sum_n (t_n - t_0) \cos \Delta_{k k_2 n} \right)}_{B_k} \\
& + \sum_{k_1, k_2} c_{k_1}^1 c_{k_2}^1 \underbrace{\left(\sum_n ((t_n - t_0)^2) \cos \Delta_{k_1 k_2 n} \right)}_{A_{k_1 k_2}}
\end{aligned} \tag{12}$$

Finally,

$$\|R\|^2 = R^0 - 2B^t c^1 + (c^1)^t A (c^1) \tag{13}$$

The minimum is reached for $c^1 = A^{-1}B$, and the minimum is $R^0 - B^t A^{-1}B$. Note that the result can be interpreted by saying that the variance has reduced by $B^t A^{-1}B$ when applying the least-square amplitude linear perturbation procedure.

5 Least Square Method for Harmonics Frequency Linear Perturbation

The signal is decomposed in K main harmonics, with a residu: $S(t_n) = \sum_k S_k(t_n) + R(t)$

The K^{th} harmonic $S_k(t) = c_k \cos(w_k t + \phi_k)$ with $c_k \geq 0$, $\phi_k \in [-\pi, \pi]$ is replaced by modifying the constant frequency w_k , to obtain a “non-periodic harmonic”: $w_k(t) = w_k^0 + w_k^1(t - t_0)$
This model is usable only for short-time interval!