

THANGAL KUNJU MUSALIAR COLLEGE OF ENGINEERING

KOLLAM – 691 005



ELECTRONICS AND COMMUNICATION ENGINEERING

22ECL509

DIGITAL SIGNAL PROCESSING LAB

LABORATORY RECORD

YEAR 2024-25

THANGAL KUNJU MUSALIAR COLLEGE OF ENGINEERING

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ELECTRONICS AND COMMUNICATION ENGINEERING

LABORATORY RECORD

YEAR 2024-25

*Certified that this is a Bonafide Record of the work done by
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Electronics and Communication Branch) in the **Digital Signal Processing**
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SIMULATION OF BASIC TEST SIGNALS

Aim:

To generate continuous and discrete waveforms for the following:

1. Unit Impulse Signal
2. Bipolar Pulse Signal
3. Unipolar Pulse Signal
4. Ramp Signal
5. Triangular Signal
6. Sine Signal
7. Cosine Signal
8. Exponential Signal
9. Unit Step Signal

Theory:

1. Unit Impulse Signal:

- A signal that is zero everywhere except at one point, typically at $t=0$ where its value is 1.
- Mathematically $\delta(t) = \begin{cases} \infty & t = 0 \\ 0 & t \neq 0 \end{cases}$

2. Unit Step Signal:

- A signal that is zero for all negative time values and one for positive time values.
- Mathematically $u(t) = \begin{cases} 1 & t \geq 0 \\ 0 & t < 0 \end{cases}$

3. Ramp Signal:

- A signal that increases linearly with time.
- Mathematically $r(t) = \begin{cases} t & t \geq 0 \\ 0 & t < 0 \end{cases}$

4. Sine Signal:

- A continuous periodic signal. It oscillates smoothly between -1 and 1.
- Mathematically: $y(t) = A \sin(2\pi ft)$

5. Cosine Signal:

- A continuous periodic signal like the sine wave but phase-shifted by $\pi/2$.
- Mathematically: $y(t) = A \cos(2\pi ft)$

6. Square wave-bipolar:

- A pulse signal that alternates between positive and negative values, usually rectangular in shape. It switches between two constant levels (e.g., -1 and 1) for a defined duration.
- Mathematically $p(t) = A$ for $|t| \leq \tau/2$, $p(t) = 0$ otherwise

7. Square wave-unipolar:

- A pulse signal that alternates between zero and a positive value. It remains at zero for a specified duration and then jumps to a positive constant level (e.g., 0 and 1).
- Mathematically $p(t) = A$ for $|t| \leq \tau/2$, $p(t) = 0$ otherwise (assuming A is positive)

8. Exponential Signal:

- A signal that increases or decreases exponentially with time. The rate of growth or decay is determined by the constant a .
- Mathematically: $e^{(at)}$

Program:

```
clc;
clear all;
close all;
t=-5:1:5;
y1=[zeros(1,5),ones(1,1),zeros(1,5)];
subplot(3,3,1);
stem(t,y1);
xlabel('time index');
ylabel('amplitude');
title('Unit impulse');

y2=[zeros(1,5),ones(1,6)];
subplot(3,3,2);
stem(t,y2);
xlabel('time index');
ylabel('amplitude');
title('Unit step signal');

t3=0:1:5;
y3=t3;
subplot(3,3,3);
plot(t3,y3);
hold on;
stem(t3,y3);
xlabel('time index');
ylabel('amplitude');
title('Ramp Signal');
legend("Continuous","Discrete");
```



```

t4=0:0.01:1;
f4=4;
subplot(3,3,4);
plot(t4,sin(2*pi*f4*t4));
hold on;
stem(t4,sin(2*pi*f4*t4));
xlabel('time');
ylabel('amplitude');
title('Sine wave');
legend("Continuous","Discrete");

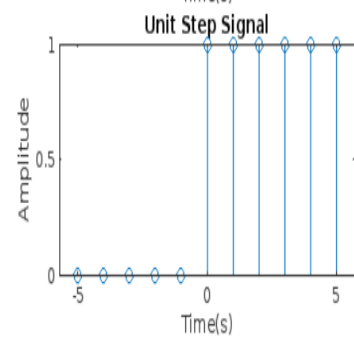
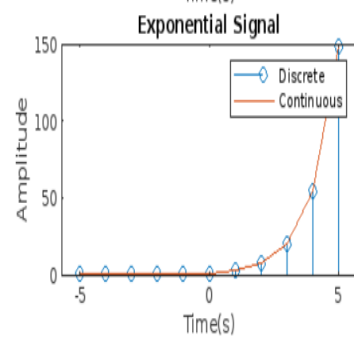
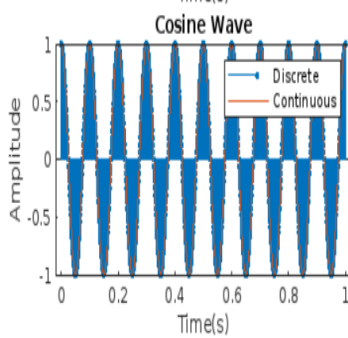
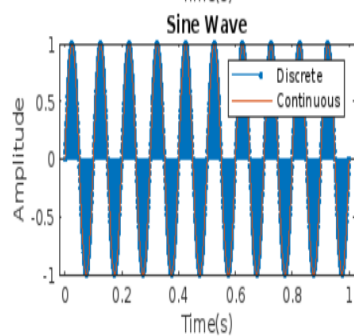
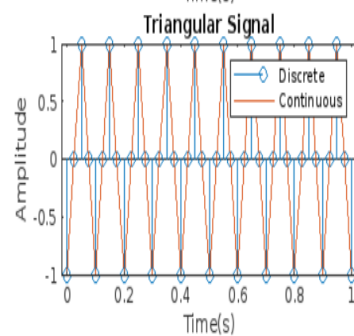
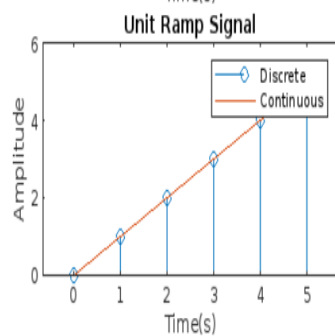
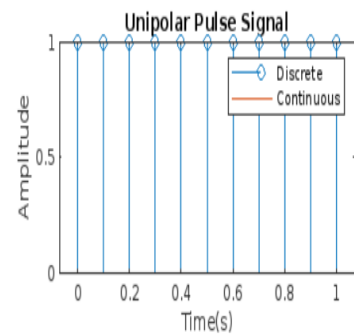
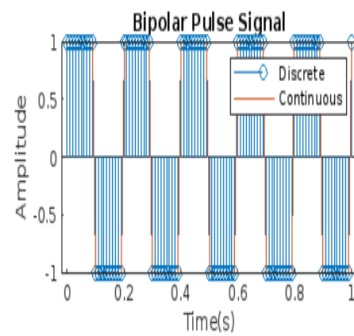
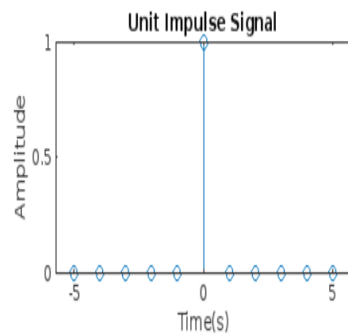
subplot(3,3,5);
plot(t4,cos(2*pi*f4*t4));
hold on;
stem(t4,cos(2*pi*f4*t4));
xlabel('time');
ylabel('amplitude');
title('Cosine wave');
legend("Continuous","Discrete");

subplot(3,3,6);
t6=0:0.0001:1;
f6=10;
plot(t6,square(2*pi*f6*t6));
xlabel('time');
ylabel('amplitude');
title('Bipolar Pulse');

subplot(3,3,7);
plot(t6,abs(square(2*pi*f6*t6)));

```

Observation:



```

xlabel('time');
ylabel('amplitude');
title('Unipolar Pulse');

t8=0:0.25:50;
f8=5;
subplot(3,3,8);
plot(t8,sawtooth(2*pi*f8*t8));
hold on;
stem(t8,sin(2*pi*f8*t8));
xlabel('time');
ylabel('amplitude');
title('Triangular wave');
legend("Continuous","Discrete");

t9=0:0.1:5;
y9=exp(-1*t9);
subplot(3,3,9);
plot(t9,y9);
hold on;
stem(t9,y9);
xlabel('time');
ylabel('amplitude');
title('Decaying Exponential signal');
legend("Continuous","Discrete");

```

Result:

Generated and Verified various Continuous and Discrete waveforms for basic test signals.

VERIFICATION OF SAMPLING THEOREM

Aim:

To verify sampling theorem using sinusoidal signal in matlab.

Theory:

The Sampling Theorem, also known as the Nyquist-Shannon Sampling Theorem, states that a continuous signal can be completely reconstructed from its samples if the sampling frequency is greater than twice the highest frequency present in the signal. This critical frequency is known as the Nyquist rate.

$$f_s \geq 2 \cdot f_{\max}$$

where:

f_s is the sampling frequency (rate at which the signal is sampled),

f_{\max} is the highest frequency present in the signal.

Applications:

- Digital audio and video processing
- Communication systems
- Image processing
- Medical imaging

Program:

```
%verification of sampling theorem
clc;
clear all;
close all;
%original signal
t=0:0.01:1;
fm=10;
```



```

y=sin(2*pi*fm*t);
figure;
subplot(2,2,1);
plot(t,y);
hold on;
stem(t,y,".");
xlabel("Time");
ylabel("Amplitude");
title("Original signal");
legend("Continuous","Discrete");

```

```

%undersampling
fs1=fm;
t1=0:1/fs1:1;
y1=sin(2*pi*fm*t1);
subplot(2,2,2);
plot(t1,y1);
hold on;
stem(t1,y1,'. ');
xlabel("Time");
ylabel("Amplitude");
title("Undersampling");
legend("Continuous","Discrete");

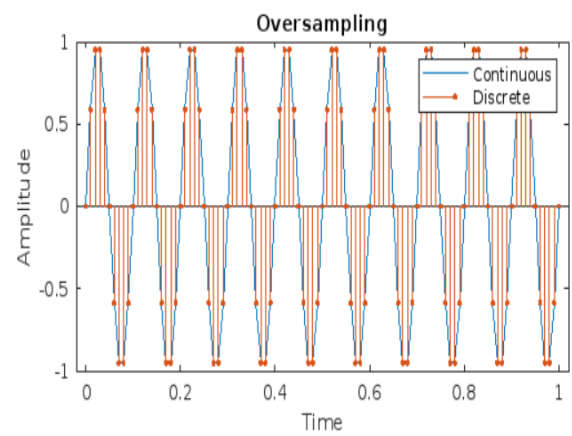
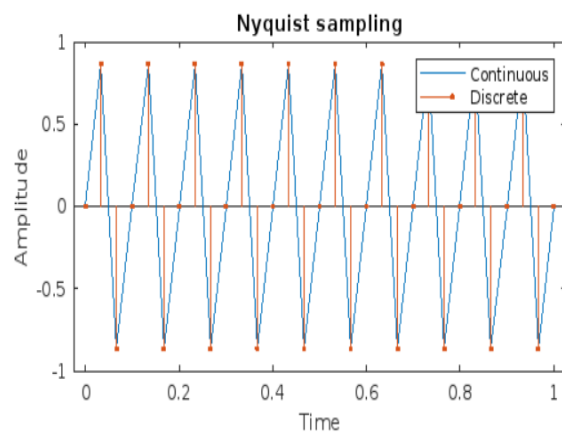
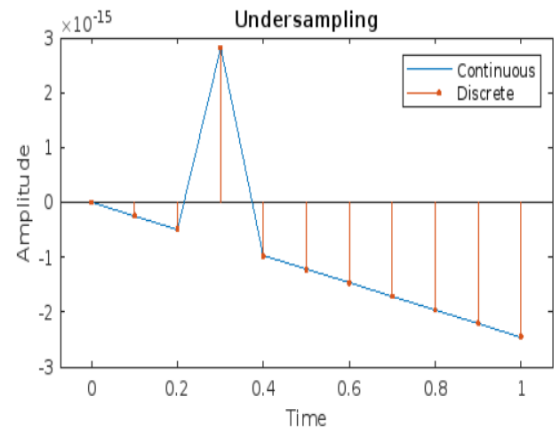
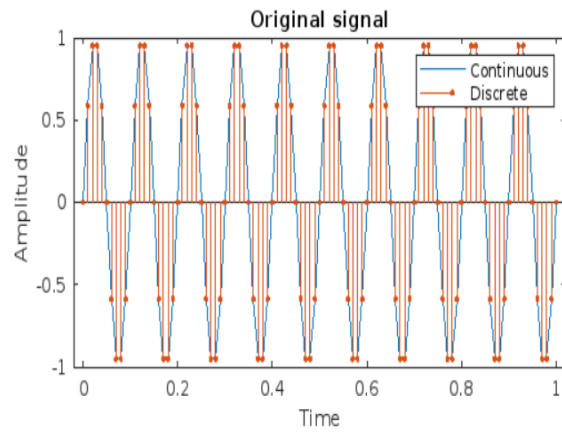
```

```

% Nyquist samling
fs2=3*fm;
t2=0:1/fs2:1;
y2=sin(2*pi*fm*t2);
subplot(2,2,3);
plot(t2,y2);

```

Observation:



```

hold on;
stem(t2,y2,'.');
xlabel("Time");
ylabel("Amplitude");
title("Nyquist sampling");
legend("Continuous","Discrete");

```

```

% Oversampling
fs3=10*fm;
t3=0:1/fs3:1;
y3=sin(2*pi*fm*t3);
subplot(2,2,4);
plot(t3,y3);
hold on;
stem(t3,y3,'.');
xlabel("Time");
ylabel("Amplitude");
title("Oversampling")
legend("Continuous","Discrete");

```

Result:

Verified Sampling Theorem using MATLAB.

LINEAR CONVOLUTION

Aim:

- a) To find linear convolution using built in function.
- b) To find linear convolution without using built in function.

Theory:

Linear convolution is a mathematical operation used to combine two signals to produce a third signal.

It's a fundamental operation in signal processing and systems theory.

Mathematical Definition:

Given two signals, $x(t)$ and $h(t)$, their linear convolution is defined as:

$$y(t) = x(t) * h(t) = \int x(\tau)h(t - \tau) d\tau$$

Applications:

- **Filtering:** Convolution is used to filter signals, removing unwanted frequencies or noise.
- **System Analysis:** The impulse response of a system completely characterizes its behaviour, and convolution can be used to determine the output of the system given a known input.
- **Image Processing:** Convolution is used for tasks like edge detection, blurring, and sharpening images.

System Analysis: The impulse response of a system completely characterizes its behaviour,

and convolution can be used to determine the output of the system given a known input.

Image Processing: Convolution is used for tasks like edge detection, blurring, and sharpening

images.

Program:

a) With built-in function:

```
clc;
clear all;
close all;
x=input('Enter input x(n):');
x_ind=input('Enter index of x(n):');
h=input('Enter input h(n):');
h_ind=input('Enter index of h(n):');
y=conv(x,h);
y_ind=min(x_ind)+min(h_ind):max(x_ind)+max(h_ind);
disp(y);
figure();
subplot(3,1,1);
stem(x_ind,x);
xlabel('Time index');
ylabel('Amplitude');
title('Input x(n)');
subplot(3,1,2);
stem(h_ind,h);
xlabel('Time index');
ylabel('Amplitude');
title('Input h(n)');
subplot(3,1,3);
stem(y_ind,y);
xlabel('Time index');
ylabel('Amplitude');
title('Linear convolution');
```


b) Without built-in function

```
clc;
clear all;
close all;
x=input('Enter input x(n):');
x_ind=input('Enter index of x(n):');
h=input('Enter input h(n):');
h_ind=input('Enter index of h(n):');
x_len=length(x);
h_len=length(h);
y_ind=min(x_ind)+min(h_ind):max(x_ind)+max(h_ind);
y=zeros(y_ind);
for i=1:x_len
    for j=1:h_len
        y(i+j-1)=y(i+j-1)+x(i)*h(j);
    end
end
disp(y);
figure();
subplot(3,1,1);
stem(x_ind,x);
xlabel('Time index');
ylabel('Amplitude');
title('Input x(n)');
subplot(3,1,2);
stem(h_ind,h);
xlabel('Time index');
```

Observation:

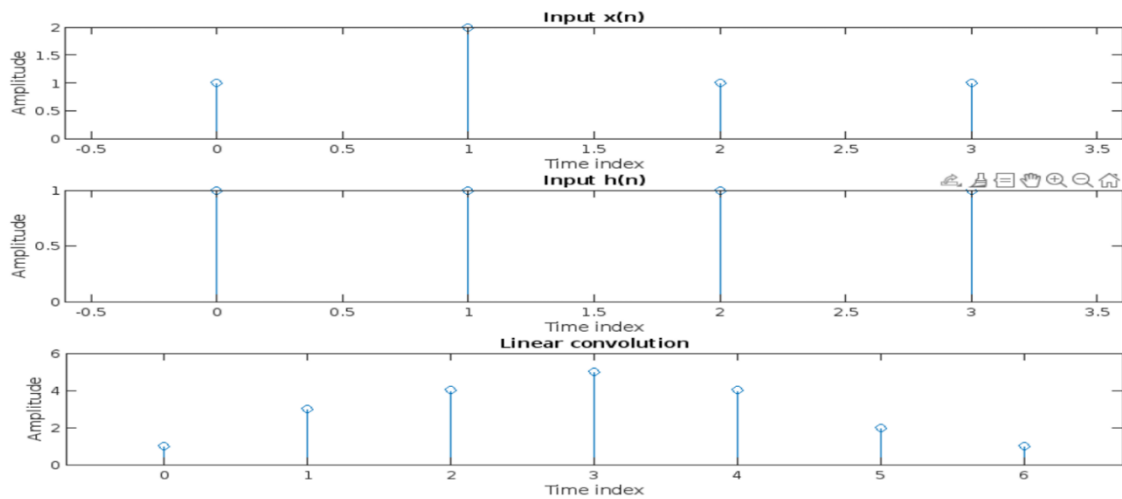
a) Enter input $x(n)$: [1 2 1 1]

Enter index of $x(n)$: 0:3

Enter input $h(n)$: [1 1 1 1]

Enter index of $h(n)$: 0:3

1 3 4 5 4 2 1



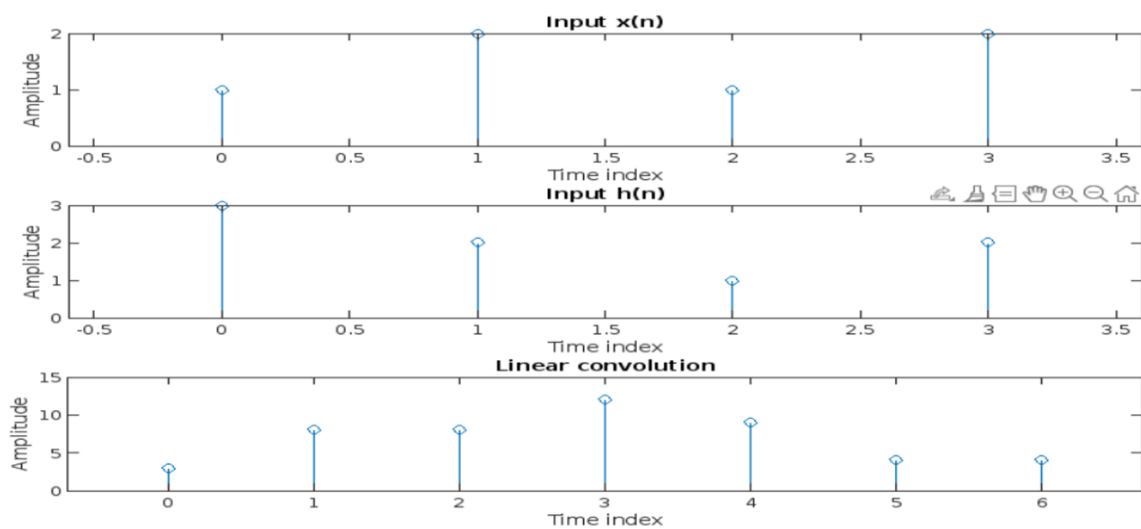
b) Enter input $x(n)$: [1 2 1 2]

Enter index of $x(n)$: 0:3

Enter input $h(n)$: [3 2 1 2]

Enter index of $h(n)$: 0:3

3 8 8 12 9 4 4



```
label('Amplitude');  
title('Input h(n)');  
subplot(3,1,3);  
stem(y_ind,y);  
xlabel('Time index');  
ylabel('Amplitude');  
title('Linear convolution');
```

Result:

Performed Linear Convolution using with and without built-in function.

Circular Convolution

Aim:

To find circular convolution

- a) Using FFT and IFFT.
- b) Using Concentric Circle Method.
- c) Using Matrix Method.

Theory:

Circular convolution is a mathematical operation that is like linear convolution but is performed in a periodic or circular manner. This is particularly useful in discrete-time signal processing where signals are often represented as periodic sequences.

Mathematical Definition:

Given two periodic sequences $x[n]$ and $h[n]$, their circular convolution is defined as:

$$y[n] = (x[n] \circledast h[n]) = \sum_{k=0}^{N-1} x[k]h[(n-k) \bmod N]$$

Applications:

- Discrete-Time Filtering: Circular convolution is used for filtering discrete-time signals.
- Digital Signal Processing: It's a fundamental operation in many digital signal processing algorithms.
- Cyclic Convolution: In certain applications, such as cyclic prefix OFDM, circular convolution is used to simplify the implementation of linear convolution.

Program:

- a) Using FFT and IFFT

```
clc;  
clear all;  
close all;
```

Observation:

Using FFT and IFFT:

16 14 16 14

```

x=[1,2,3,4];
xn=0:3;
h=[1,2,1,0];
hn=0:3;
figure;
subplot(3,1,1);
stem(xn,x);
xlabel('n');
ylabel('x[n]');

subplot(3,1,2);
stem(hn,h);
xlabel('n');
ylabel('h[n]');

%circular convolution

nx=length(x);
nh=length(h);
N=max(nx,nh);
newx=[x zeros(1,N-nx)];
newh=[h zeros(1,N-nh)];
x1=fft(newx);
H1=fft(newh);
Y=x1.*H1;
y=ifft(Y);

```


Observation:

$$h = 1 \times 4$$

1 2 3 4

$$x = 1 \times 4$$

2 1 2 1

$$x = 1 \times 4$$

1 2 1 2

$$x = 1 \times 4$$

2 1 2 1

$$x = 1 \times 4$$

1 2 1 2

14 16 14 16

b) Using Concentric Circle Method

```
clc;
close all;
clear all;
x=[2 1 2 1];
x=x(:,end:-1:1);
%% h[ 1 2 1 2]
h=[1 2 3 4]
for i=1:length(x)
    x=[x(end) x(1:end-1)]
    h1=h;
    y(i)=sum(x.*h1);
end
disp(y);
```

c) Using Matrix Method

```
clc;
clear all;
close all;
xn=[2,1,2,1];
hn=[1,2,3,4];
h=[];
hn=hn(:,end:-1:1);
disp(hn);
for i =1:length(hn)
    hn=[hn(end) hn(1:end-1)];
    h=[h;hn];
end
```

Observation:

4 3 2 1

1 4 3 2

2 1 4 3

3 2 1 4

4 3 2 1

14

16

14

16

```
disp(h);  
%%x=flipdim(x,1);  
y=h*xn';  
disp(y);
```

Result:

Performed Circular Convolution using:

- a) FFT and IFFT
- b) Concentric Circle method
- c) Matrix method and verified result

LINEAR CONVOLUTION USING CIRCULAR **CONVOLUTION AND VICE VERSA**

Aim:

- a) To perform Linear Convolution using Circular Convolution.
- b) To perform Circular Convolution using Linear Convolution.

Theory:

Performing Linear Convolution Using Circular Convolution.

1. **Zero-Padding:** Pad both sequences $x[n]$ and $h[n]$ with zeros to a length of at least $2N-1$, where N is the maximum length of the two sequences. This ensures that the circular convolution will not wrap around and introduce artificial periodicity.
2. **Circular Convolution:** Perform circular convolution on the zero-padded sequences.
3. **Truncation:** Truncate the result of the circular convolution to the length $N_1 + N_2 - 1$, where N_1 and N_2 are the lengths of the original sequences $x[n]$ and $h[n]$, respectively.

Performing Circular Convolution Using Linear Convolution.

1. **Zero-Padding:** Pad both sequences $x[n]$ and $h[n]$ to a length of at least $2N-1$, where N is the maximum length of the two sequences.
2. **Linear Convolution:** Perform linear convolution on the zero-padded sequences.
3. **Modulus Operation:** Apply the modulus operation to the indices of the linear convolution result, using the period N . This effectively wraps around the ends of the sequence, making it circular

Program:**a) Linear Convolution using Circular Convolution**

```
clc;
close all;
clear all;
x = [1 2 3 4];
h = [1 1 1 ];
l=length(x);
m=length(h);
n=l+m-1;
x=[x zeros(1,n-l)];
h=[h zeros(1,n-m)];
x1=fft(x);
h1=fft(h);
y1=x1.*h1;
y=ifft(y1);
disp(y);
```

b) Circular convolution using Linear Convolution

```
clc;
close all;
clear all;
x = [1 2 3 4];
h = [1 1 1 ];
y=conv(x,h);
l=length(x);
m=length(h);
n=max(l,m);
```


Observation:

a) Linear Convolution using Circular Convolution

1 3 6 9 7 4

b) Circular convolution using Linear Convolution

8 7 6 9

```
r1=y(1:n);  
r2=[y(n+1:end) zeros(1:length(y)-n)];  
r=r1+r2;  
disp(r);
```

Result:

Performed a) Linear Convolution using Circular Convolution;

b) Circular Convolution using Linear Convolution and verified result.

DISCRETE FOURIER TRANSFORM & INVERSE DISCRETE FOURIER TRANSFORM

Aim:

- 1) DFT using inbuilt function and without using inbuilt function. Also plot magnitude and phase plot of DFT
- 2) IDFT using inbuilt function and without using inbuilt function.

Theory:

Discrete Fourier Transform (DFT):

The Discrete Fourier Transform (DFT) is a mathematical transformation used to analyze the frequency content of discrete signals. For a sequence $x[n]$ of length N , the DFT is defined as:

$$X[k] = \sum_{n=0}^N x[n] \cdot e^{-j\frac{2\pi}{N}nk}, k = 0, 1, 2, \dots, N-1$$

- $X[k]$ is the DFT of the sequence $x[n]$.
- The exponential factor represents $e^{-j\frac{2\pi}{N}nk}$ the complex sinusoidal basis functions.
- The DFT maps the time-domain signal into the frequency domain.

Inverse Discrete Fourier Transform (IDFT):

The Inverse Discrete Fourier Transform (IDFT) is used to convert a frequency-domain sequence $X[k]$ back into its time-domain sequence $x[n]$. The IDFT is defined as:

$$X[n] = \frac{1}{N} \sum_{k=0}^N X[k] \cdot e^{j\frac{2\pi}{N}nk}, n = 0, 1, 2, \dots, N-1$$

- The IDFT takes the frequency components $X[k]$ and reconstructs the original sequence $x[n]$.
- The exponential factor $e^{j\frac{2\pi}{N}nk}$ is the inverse of the DFT's complex sinusoidal basis functions.

Application:

- Spectrum (Analysis)
- Filtering
- Compression
- Modulation
- Convolution
- Demodulation
- Equalization

Observation:

1.DFT

enter sequence:[1 1 1 0]

enter the N point:8

X

Columns 1 through 7

3.0000 + 0.0000i 1.7071 - 1.7071i 0.0000 - 1.0000i 0.2929 + 0.2929i 1.0000 + 0.0000i

0.2929 - 0.2929i -0.0000 + 1.0000i

Column 8

1.7071 + 1.7071i

round(X)

Columns 1 through 7

3.0000 + 0.0000i 2.0000 - 2.0000i 0.0000 - 1.0000i 0.0000 + 0.0000i 1.0000 + 0.0000i

0.0000 + 0.0000i 0.0000 + 1.0000i

Column 8

2.0000 + 2.0000i

fft

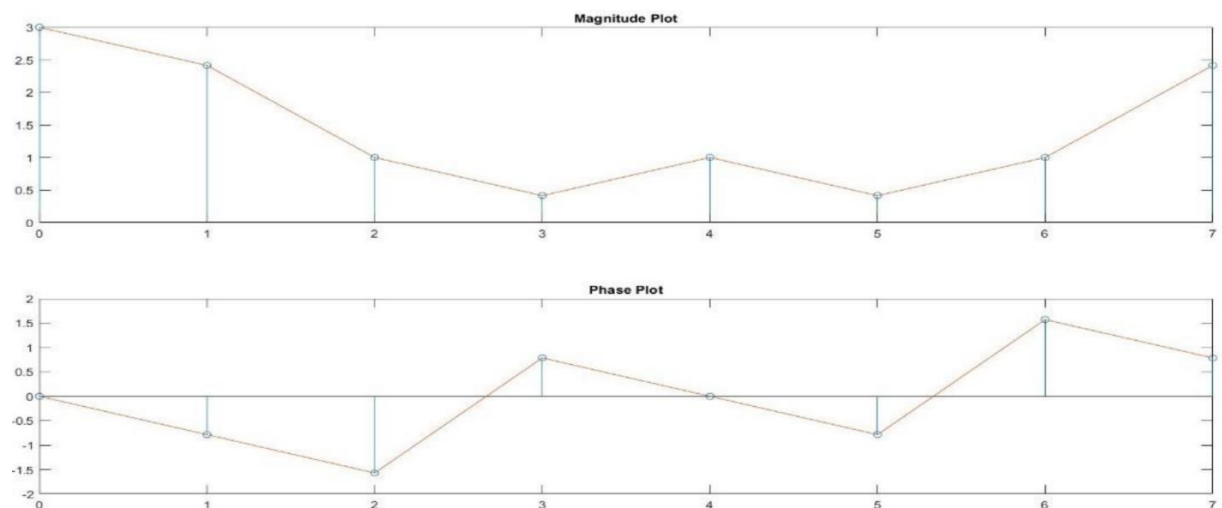
Columns 1 through 7

3.0000 + 0.0000i 1.7071 - 1.7071i 0.0000 - 1.0000i 0.2929 + 0.2929i 1.0000 + 0.0000i

0.2929 - 0.2929i 0.0000 + 1.0000i

Column 8

1.7071 + 1.7071i



- Restoration
- Detection
- Estimation

Program:

1)DFT

```

clc;
clear all;
close all;
x=input("enter sequence:");
N=input("enter the N point:");
l=length(x);
x=[x zeros(1,N-l)];
X=zeros(1,N);
for k=0:N-1
    for n=0:N-1
        X(k+1)=X(k+1)+x(n+1)*exp(-1j*2*pi*n*k/N);
    end
end
disp('X');
disp(X);
disp('round(X)');
disp(round(X));
%verification
disp('fft');
disp(fft(x));
%plotting
k=0:N-1;
magX=abs(X);
phaseX=angle(X);
subplot(2,1,1);

```

Observation:

2. IDFT

enter sequence: $[3.0000 + 0.0000i \ 1.7071 - 1.7071i \ 0.0000 - 1.0000i \ 0.2929 + 0.2929i$
 $1.0000 + 0.0000i \ 0.2929 - 0.2929i \ 0.0000 + 1.0000i]$

enter the n point:8

x

$0.7866 - 0.2134i$

$0.6982 - 0.0000i$

$0.7866 + 0.2134i$

$-0.0000 + 0.3018i$

$0.2134 + 0.2134i$

$0.3018 - 0.0000i$

$0.2134 - 0.2134i$

$0.0000 - 0.3018i$

round(x)

1

1

1

0

0

0

0

0

ifft

Columns 1 through 7

$0.7866 - 0.2134i \ 0.6982 + 0.0000i \ 0.7866 + 0.2134i \ 0.0000 + 0.3018i \ 0.2134 +$
 $0.2134i \ 0.3018 + 0.0000i \ 0.2134 - 0.2134i$

Column 8

$0.0000 - 0.3018i$

```

stem(k,magX);
title("Magnitude Plot");
hold on;
plot(k,magX);
subplot(2,1,2);
stem(k,phaseX);
hold on;
title("Phase Plot");
plot(k,phaseX);

```

2)IDFT

```

clc;
clear all;
close all;
X=input("enter sequence:");
N=input("enter the n point:");
l=length(X);
X=[X zeros(1,N-1)];
x=zeros(N,1);
for k=0:N-1
    for n=0:N-1
        x(n+1)=x(n+1)+X(k+1)*exp(1j*2*pi*n*k/N);
    end
end
x=1/N.*x;
disp('x');
disp(x);
disp('round(x)');
disp(round(x));
%verification
disp('ifft');
disp(ifft(X));

```


Observation:

c)DFT using twiddle factor

Using twiddle factor

$$10.0000 + 0.0000i \quad -2.0000 + 2.0000i \quad -2.0000 - 0.0000i \quad -2.0000 - 2.0000i$$

Using built-in function

$$10.0 \quad .0000i \quad -2.0000 + 2.0000i \quad -2.0000 + 0.0000i \quad -2.0000 - 2.0000i$$

d)IDFT using twiddle factor

Using twiddle factor

$$1.0000 + 0.0000i \quad 2.0000 + 0.0000i \quad 3.0000 - 0.0000i \quad 4.0000 - 0.0000i$$

Using built-in function

$$1 \quad 2 \quad 3 \quad 4$$

3)DFT using twiddle factor

```
clc;
clear;
close all;
x=[1 2 3 4];
N=length(x);
n=0:N-1;
k=n';
W=exp(-1i*2*pi*(k*n)/N);
X=x*W;
disp('Using twiddle factor');
disp(X);
disp('Using built-in function');
disp(fft(x));
```

4)IDFT using twiddle factor

```
clc;
clear;
close all;
X=[10 -2 + 2i -2 -2 - 2i];
N=length(X);
n=0:N-1;
k=n';
W=exp(1i*2*pi*(k*n)/N);
x=X*W/N;
disp(x);
disp(ifft(X));
```


Result:

Performed

1)DFT using inbuilt function and without using inbuilt function. Also plotted magnitude and phase plot of DFT.

2)IDFT using inbuilt function and without using inbuilt function.
and verified the result.

PROPERTIES OF DFT

Aim:

Verify following properties of DFT using Matlab/Scilab.

1. Linearity Property
2. Parseval's Theorem
3. Convolution Property
4. Multiplication Property

Theory:

1. Linearity Property

The linearity property of the DFT states that if you have two sequences $x_1[n]$ and $x_2[n]$, and their corresponding DFTs are $X_1[k]$ and $X_2[k]$, then for any scalar a and b :

$$\text{DFT}\{a \cdot x_1[n] + b \cdot x_2[n]\} = a \cdot \text{DFT}\{x_1[n]\} + b \cdot \text{DFT}\{x_2[n]\}$$

2. Parseval's Theorem

Parseval's theorem states that the total energy of a signal in the time domain is equal to the total energy in the frequency domain. For a sequence $x[n]$ and its DFT $X[k]$:

$$\sum_{n=0}^{N-1} |x[n]|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X[k]|^2$$

3. Convolution Property

The convolution property of the DFT states that the circular convolution of two sequences in the time domain is equivalent to the element-wise multiplication of their DFTs in the frequency domain:

$$\text{DFT}\{x_1[n] \otimes x_2[n]\} = \text{DFT}\{x_1[n]\} \cdot \text{DFT}\{x_2[n]\}$$

4. Multiplication Property

The multiplication property of DFT states that pointwise multiplication in the time domain corresponds to circular convolution in the frequency domain:

$$\text{DFT}\{x_1[n] \cdot x_2[n]\} = \frac{1}{N} \text{DFT}\{x_1[n]\} \otimes \text{DFT}\{x_2[n]\}$$

Observation:

1. Linearity Property:

enter first sequence[1 2 3 4]

enter sequence sequence:[1 1 1 1]

x =

1 2 3 4

enter value of 'a':2

enter value of 'b':3

LHS

$32.0000 + 0.0000i - 4.0000 + 4.0000i - 4.0000 + 0.0000i - 4.0000 - 4.0000i$

RHS

$32.0000 + 0.0000i - 4.0000 + 4.0000i - 4.0000 + 0.0000i - 4.0000 - 4.0000i$

Linearity property verified

2. Parseval's Theorem:

enter first sequence:[1 2 3 4]

enter second sequence:[1 1 1 1]

LHS

10

RHS

10

Parseval's Theorem verified

Program:

1. Linearity Property

```
clc;
clear all;
close all;
x=input("enter first sequence");
h=input("enter second sequence:");
lx=length(x);
lh=length(h);
if lx>lh
    h=[h zeros(1,lx-lh)]
else
    x=[x zeros(1,lh-lx)]
end
a=input("enter value of 'a':");
b=input("enter value of 'b':");
lhs=fft((a.*x)+(b.*h));
rhs=a.*fft(x)+b.*fft(h);
disp('LHS');
disp(lhs);
disp('RHS');
disp(rhs);
if lhs==rhs
    disp('Linearity property verified');
else
    disp('Linearity property not verified');
end
```

2. Parseval's Theorem

```
clc;
clear all;
```

Observation:

3.Convolution Property:

enter first sequence[1 2 3 4]

enter sequence sequence:[1 1 1 1]

LHS

10 10 10 10

RHS

10 10 10 10

Circular Convolution verified

```

close all;
x=input("enter first sequence:");
h=input("enter second sequence:");
N=max(length(x),length(h));
xn=[x zeros(1,N-length(x))];
hn=[h zeros(1,N-length(h))];
lhs=sum(xn.*conj(hn));
rhs=sum(fft(xn).*conj(fft(hn)))/N;
disp('LHS');
disp(lhs);
disp('RHS');
disp(rhs);
if lhs==rhs
    disp("Parseval's Theorem verified");
else
    disp("Parseval's Theorem not verified");
end

```

3.Convolution Property

```

clc;
clear all;
close all;
x=input("enter first sequence");
h=input("enter sequence sequence:");
N=max(length(x), length(h));
xn=[x zeros(N-length(x))];
hn=[h zeros(N-length(h))];
Xn=fft(xn);
Hn=fft(hn);
lhs=cconv(xn,hn,N);
rhs=ifft(Xn.*Hn);
disp('LHS');

```

Observation:

4.Multiplication Property

enter first sequence[1 2 3 4]

enter sequence sequence:[1 1 1 1]

LHS

Columns 1 through 3

$10.0000 + 0.0000i - 2.0000 + 2.0000i - 2.0000 + 0.0000i$

Column 4

$-2.0000 - 2.0000i$

RHS

Columns 1 through 3

$10.0000 + 0.0000i - 2.0000 + 2.0000i - 2.0000 + 0.0000i$

Column 4

$-2.0000 - 2.0000i$

Multiplication property verified

```

disp(lhs);
disp('RHS');
disp(rhs);
if lhs==rhs
    disp('Circular Convolution verified')
else
    disp('Circular Convolution not verified');
end

```

4. Multiplication Property

```

clc;
clear all;
close all;
x=input("enter first sequence");
h=input("enter sequence sequence:");
N=max(length(x), length(h));
xn=[x zeros(N-length(x))];
hn=[h zeros(N-length(h))];
lhs=fft(xn.*hn);
Xn=fft(xn);
Hn=fft(hn);
rhs=(cconv(Xn,Hn,N))/N;
disp('LHS');
disp(lhs);
disp('RHS');
disp(rhs);
if lhs==rhs
    disp('Multiplication property verified');
else
    disp('Multiplication property not verified');
end

```


Result:

Performed and verified the following properties of DFT:

- 1.Linear Property
- 2.Parseval's Theorem
- 3.Convolution Property
- 4.Multiplication Property

OVERLAP ADD AND OVERLAP SAVE METHOD

Aim:

Implement overlap add and overlap save method using Matlab/Scilab.

Theory:

Both the Overlap-Save and Overlap-Add methods are techniques used to compute the convolution of long signals using the Fast Fourier Transform (FFT). The direct convolution of two signals, especially when they are long, can be computationally expensive. These methods allow us to break the signals into smaller blocks and use the FFT to perform the convolution more efficiently.

Overlap-Save Method:

The Overlap-Save method deals with circular convolution by discarding the parts of the signal that are corrupted by wrap-around effects. Here's how it works:

1. Block Decomposition: The input signal is divided into overlapping blocks. If the filter has length M and we use blocks of length N , the overlap is M samples, so each block has $N - M + 1$ new samples and $M - 1$ samples from the previous block.
2. FFT and Convolution: Each block is convolved with the filter using FFT. However, because of circular convolution, the result contains artifacts due to the overlap.
3. Discard and Save: We discard the first $M - 1$ samples from each block (the part affected by the wrap-around) and save the remaining samples. This gives us the correct linear convolution.

Overlap-Add Method :

The Overlap-Add method, on the other hand, handles circular convolution by adding overlapping sections of the convolved blocks. Here's how it works:

1. Block Decomposition: The input signal is split into non-overlapping blocks of size N . Each block is then zero-padded to a size of $2N - M$, where M is the length of the filter.
2. FFT and Convolution: Each block is convolved with the filter using FFT. Since the blocks

are zero-padded, the convolution produces valid linear results, but the output blocks overlap.

3. Overlap and Add: After convolution, the results of each block overlap by samples. These overlapping regions are added together to form the final output.

Program:

1. Overlap Add:

```
clc;
clear all;
close all;

% User input for the input sequence
x = input('Enter the input sequence x : ');
% User input for the impulse response
h = input('Enter the impulse response h : ');
% Section length for overlap-save
L = length(h); % Length of impulse response
% Initialization
N = length(x);
M = length(h);
% Pad input x with zeros
x_padded = [x, zeros(1, L - 1)];
% Prepare the output array
y = zeros(1, N + M + 1);
% Calculate the number of sections
num_sections = (N + L - 1) / L; % Calculate number of sections
% Process sections
for n = 0:num_sections-1
    % Determine the current section
    start_idx = n * L + 1;
    end_idx = start_idx + L - 1;
    % Ensure the section does not exceed the bounds
```

Observation:

1.Overlap Add:

Enter the input sequence x : [3 -1 0 1 3 2 0 1 2 1]

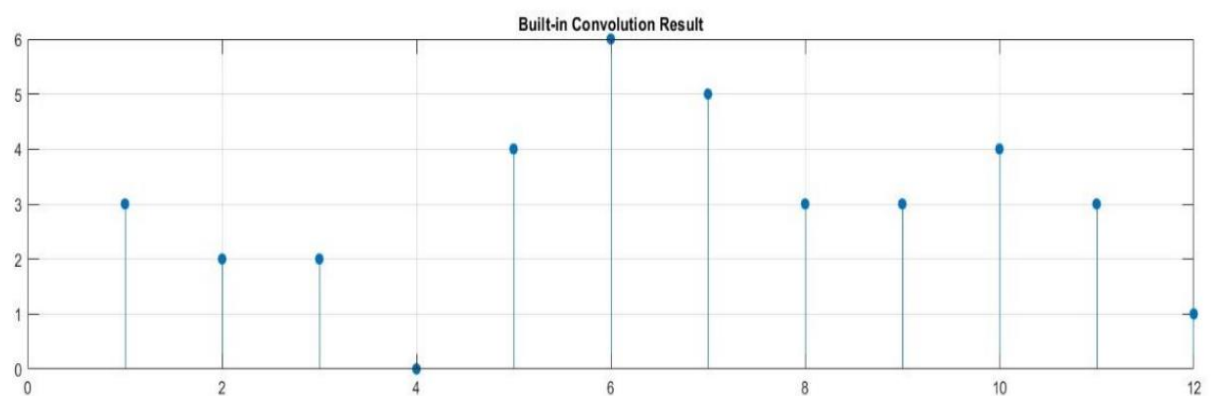
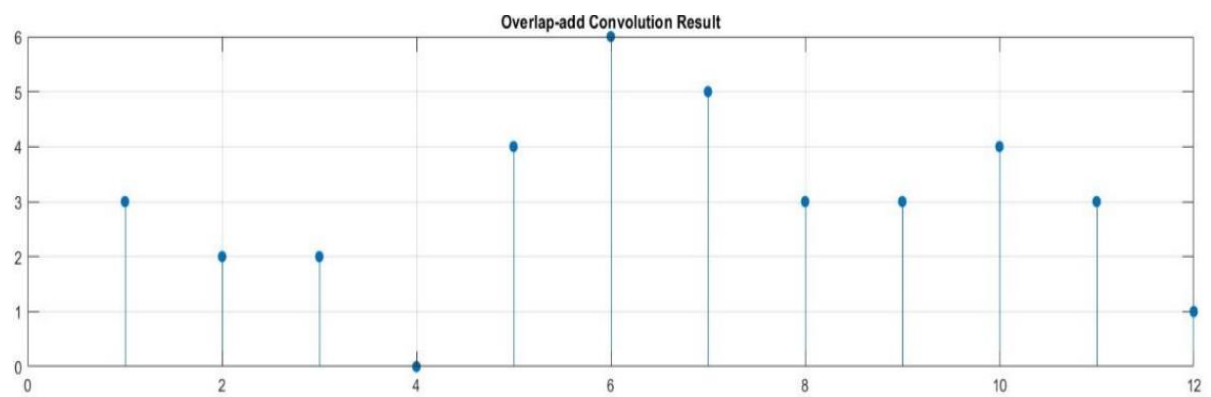
Enter the impulse response h : [1 1 1]

Overlap-add convolution result:

3 2 2 0 4 6 5 3 3 4 3 1

Built-in convolution result:

3 2 2 0 4 6 5 3 3 4 3 1



```

x_section = x_padded(start_idx:min(end_idx, end));
% Convolution
conv_result = conv(x_section, h);
% Save the results to the output
y(start_idx:start_idx + length(conv_result) - 1)
=y(start_idx:start_idx + length(conv_result) - 1) + conv_result;
end
% Trim the output to the valid part
y = y(1:N + M - 1);
% Compare with built-in convolution
y_builtin = conv(x, h);
% Display results
disp('Overlap-add convolution result:');
disp(y);
disp('Built-in convolution result:');
disp(y_builtin);
% Plotting results
figure;
subplot (2, 1, 1);
stem(y, 'filled');
title('Overlap-add Convolution Result');
grid on;
subplot(2, 1, 2);
stem(y_builtin, 'filled');
title('Built-in Convolution Result');
grid on;

```

2.Overlap Save:

```

clc;
clear all;
close all;

```

Observation:

2.Overlap Save:

Enter 1st sequence: [3 -1 0 1 3 2 0 1 2 1]

Enter 2nd sequence: [1 1 1]

Fragmented block size: 3

Using Overlap and Save method

3 2 2 0 4 6 5 3 3 4 3 1

Verification

3.0000 2.0000 2.0000 0 4.0000 6.0000 5.0000 3.0000 3.0000 4.0000

3.0000 1.0000

```

% Input the sequences and block size
x = input("Enter 1st sequence: ");
h = input("Enter 2nd sequence: ");
N = input("Fragmented block size: ");
% Call the overlap-save function
y = ovrlsav(x, h, N);
disp("Using Overlap and Save method");
disp(y);
disp("Verification");
disp(cconv(x,h,length(x)+length(h)-1));
% Define the overlap-save method function
function y = ovrlsav(x, h, N)
    if (N < length(h))
        error("N must be greater than the length of h");
    end
    Nx = length(x); % Length of input sequence x
    M = length(h); % Length of filter sequence h
    M1 = M - 1; % Length of overlap
    L = N - M1; % Length of non-overlapping part
    % Zero-padding for input and filter sequences
    x = [zeros(1, M1), x, zeros(1, N-1)];
    h = [h, zeros(1, N - M)];
    % Number of blocks
    K = floor((Nx + M1 - 1) / L);
    % Initialize the output matrix Y
    Y = zeros(K + 1, N);
    % Perform block convolution using circular convolution
    for k = 0:K
        xk = x(k*L + 1 : k*L + N); % Extract block of input sequence
        Y(k+1, :) = cconv(xk, h, N); % Circular convolution
    end
end

```



```
% Extract valid part from the result and concatenate  
Y = Y(:, M:N)';  
y = (Y(:))';  
end
```

Result:

Implemented overlap add and overlap save method using Matlab/Scilab.

IMPLEMENTATION OF FIR FILTERS

Aim:

Implement various FIR filters using different windows

1. Low Pass Filter
2. High Pass Filter
3. Band pass Filter
4. Band stop Filter

Theory:

Design of FIR Filters Using Window Methods

In FIR (Finite Impulse Response) filter design, the goal is to create a filter with specific frequency response characteristics, such as low-pass, high-pass, band-pass, or band-stop. Using

window methods, we can shape the filter response by applying a window function to an ideal filter impulse response.

Step 1: Define the Ideal Impulse Response

The ideal impulse response, $h_{\text{ideal}}(n)$, of a low-pass filter with a cutoff frequency f_c is given

by:

$$h_{\text{ideal}}(n) = \sin(2 * \pi * f_c * (n - (N - 1) / 2)) / (\pi * (n - (N - 1) / 2))$$

Where:

- f_c : Normalized cut off frequency
- N : Filter length
- n : Sample index

Step 2: Select an Appropriate Window Function

The choice of window affects the trade-off between the main lobe width and the sidelobe levels.

Common windows include the Rectangular, Hamming, Hanning, Blackman, and Kaiser windows.

Window Type	Formula
Rectangular	$w(n) = 1$
Triangular	$w(n) = 1 - 2*abs(n) / (N - 1)$
Hamming	$w(n) = 0.54 + 0.46 * \cos(2 * \pi * n / (N - 1))$
Hanning	$w(n) = 0.5 * (1 + \cos(2 * \pi * n / (N - 1)))$
Blackman	$w(n) = 0.42 + 0.5 * \cos(2 * \pi * n / (N - 1)) + 0.08 * \cos(4 * \pi * n / (N - 1))$
Kaiser	$w(n) = I_0(\beta * \sqrt{1 - (2 * n / (N - 1) - 1)^2}) / I_0(\beta)$

Step 3: Apply the Window to the Ideal Impulse Response

The windowed impulse response is computed as:

$$h(n) = h_{ideal}(n) * w(n)$$

Step 4: Construct the FIR Filter

The final impulse response $h(n)$ defines the FIR filter coefficients that can be used in filtering algorithms.

Filters:

$$\begin{aligned}
 \text{Lowpass:} \quad h(n) &= \begin{cases} \frac{\Omega_c}{\pi} & n = 0 \\ \frac{\sin(\Omega_c n)}{n\pi} & \text{for } n \neq 0 \end{cases} \quad -M \leq n \leq M \\
 \text{Highpass:} \quad h(n) &= \begin{cases} \frac{\pi - \Omega_c}{\pi} & n = 0 \\ -\frac{\sin(\Omega_c n)}{n\pi} & \text{for } n \neq 0 \end{cases} \quad -M \leq n \leq M \\
 \text{Bandpass:} \quad h(n) &= \begin{cases} \frac{\Omega_H - \Omega_L}{\pi} & n = 0 \\ \frac{\sin(\Omega_H n)}{n\pi} - \frac{\sin(\Omega_L n)}{n\pi} & \text{for } n \neq 0 \end{cases} \quad -M \leq n \leq M \\
 \text{Bandstop:} \quad h(n) &= \begin{cases} \frac{\pi - \Omega_H + \Omega_L}{\pi} & n = 0 \\ -\frac{\sin(\Omega_H n)}{n\pi} + \frac{\sin(\Omega_L n)}{n\pi} & \text{for } n \neq 0 \end{cases} \quad -M \leq n \leq M
 \end{aligned}$$

Advantages and Disadvantages of Window-Based FIR Design

Advantages:

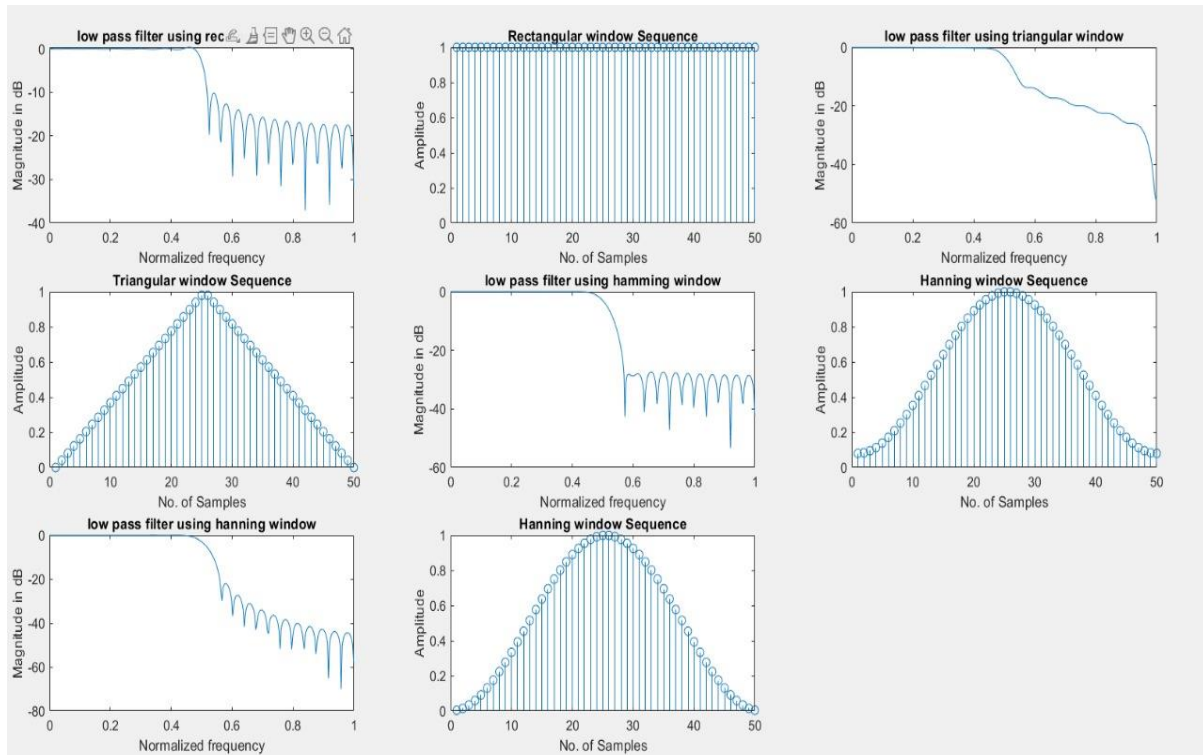
- Simplicity: Windowing is straightforward and does not require iterative optimization.
- Control over Leakage: Different windows provide different control over sidelobes and main lobe width.

Disadvantages:

- Fixed Frequency Response: Once the window is chosen, the frequency response characteristics are determined.
- Trade-Off Limitations: Some applications require specific frequency responses that cannot be perfectly achieved using standard windows.

Observation:

1. LOW PASS FILTER



Program:

1. LOW PASS FILTER

```
clc;
clear all;
close all;
wc=0.5*pi;
N = 50;
alpha = (N-1)/2;
eps = 0.001;
n = 0:1:N-1;
hd = (sin(wc*(n-alpha+eps)))/(pi*(n-alpha+eps));
wr = boxcar(N);
wt=bartlett(N);
wh=hamming(N);
whn=hanning(N);
hn1 = hd.*wr';
hn2 = hd.*wt';
hn3 = hd.*wh';
hn4 = hd.*whn';
w = 0:0.01:pi;
h1 = freqz(hn1,1,w);
h2 = freqz(hn2,1,w);
h3 = freqz(hn3,1,w);
h4 = freqz(hn4,1,w);
subplot(3,3,1);
plot(w/pi,10*log10(abs(h1)));
title('low pass filter using rectangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,2);
```



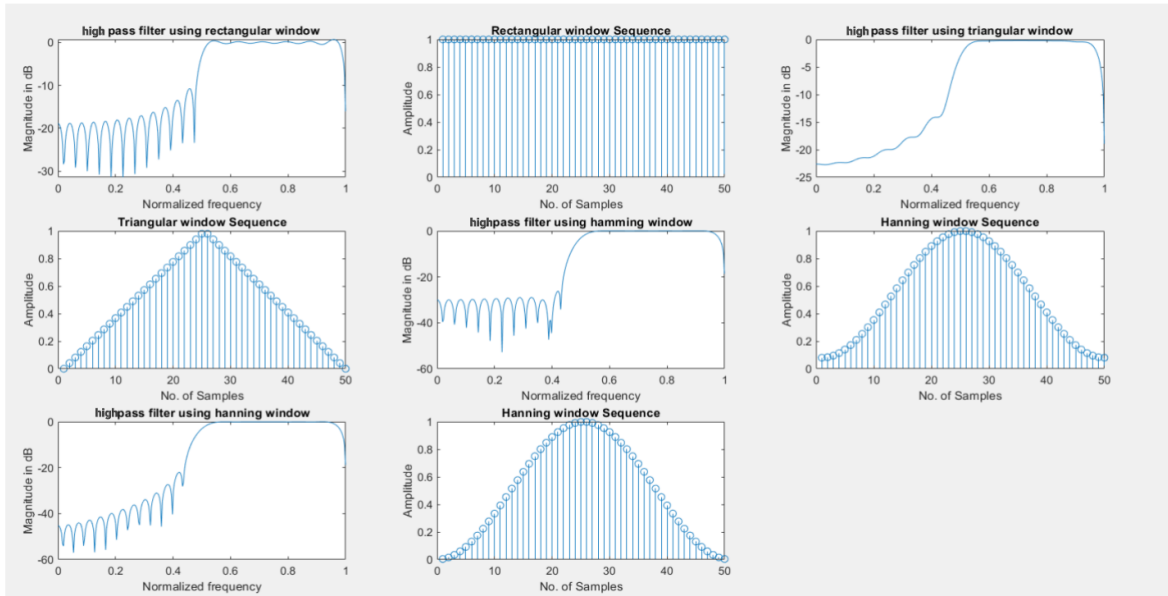
```

stem(wr);
title('Rectangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,3);
plot(w/pi,10*log10(abs(h2)));
title('low pass filter using triangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,4);
stem(wt);
title('Triangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,5);
plot(w/pi,10*log10(abs(h3)));
title('low pass filter using hamming window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,6);
stem(wh);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,7);
plot(w/pi,10*log10(abs(h4)));
title('low pass filter using hanning window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,8);

```


Observation:

2.HIGH PASS FILTER



```

stem(whn);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');

```

2. HIGH PASS FILTER

```

clc;
clear all;
close all;
wc=0.5*pi;
N = 50;
alpha = (N-1)/2;
eps = 0.001;
n = 0:1:N-1;
hd=(sin(pi*(n-alpha+eps))-sin(wc*(n-alpha+eps)))/(pi*(n-alpha+eps));
wr = boxcar(N);
wt=bartlett(N);
wh=hamming(N);
whn=hanning(N);
hn1 = hd.*wr';
hn2 = hd.*wt';
hn3 = hd.*wh';
hn4 = hd.*whn';
w = 0:0.01:pi;
h1 = freqz(hn1,1,w);
h2 = freqz(hn2,1,w);
h3 = freqz(hn3,1,w);
h4 = freqz(hn4,1,w);
subplot(3,3,1);
plot(w/pi,10*log10(abs(h1)));

```



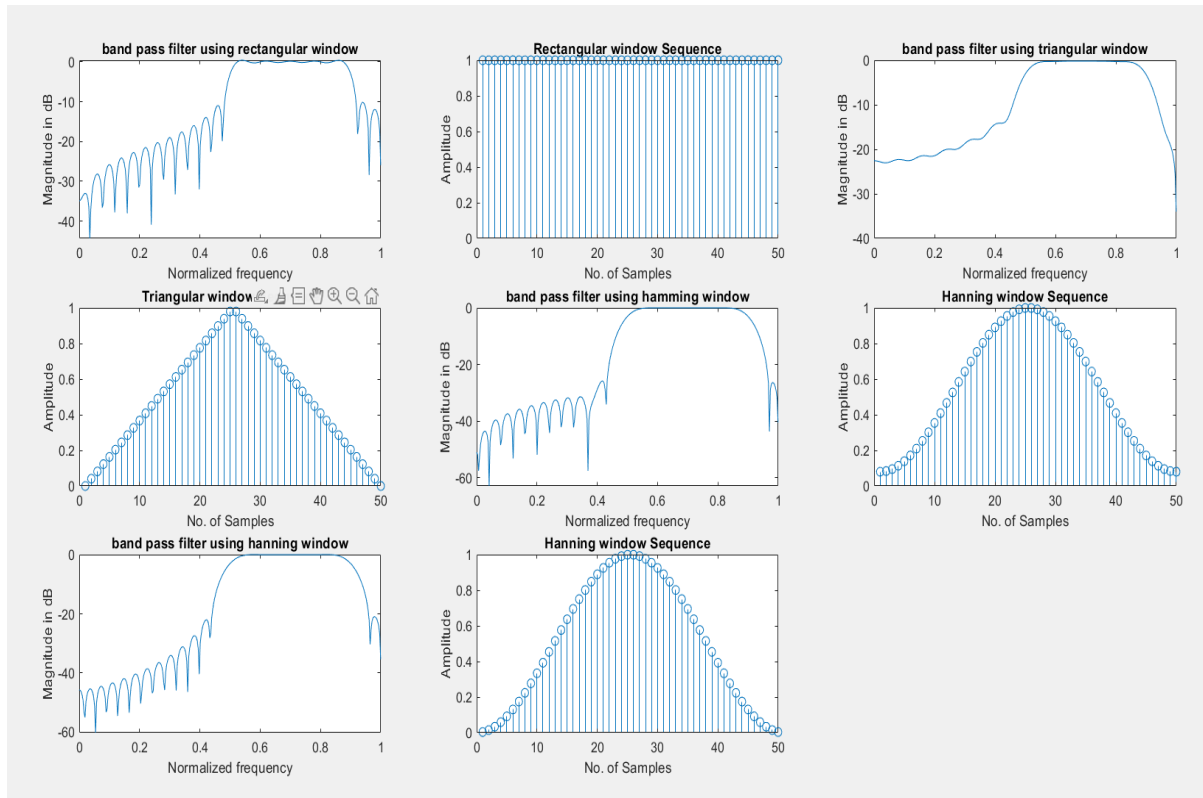
```

title('high pass filter using rectangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,2);
stem(wr);
title('Rectangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,3);
plot(w/pi,10*log10(abs(h2)));
title('high pass filter using triangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,4);
stem(wt);
title('Triangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,5);
plot(w/pi,10*log10(abs(h3)));
title('high pass filter using hamming window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,6);
stem(wh);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,7);
plot(w/pi,10*log10(abs(h4)));

```

Observation:

3.BANDPASS FILTER



```

title('high pass filter using hanning window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,8);
stem(whn);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');

```

3. BANDPASS FILTER

```

clc;
clear all;
close all;
wc1=0.5*pi;
wc2=0.9*pi;
N = 50;
alpha = (N-1)/2;
eps = 0.001;
n = 0:1:N-1;
hd=(sin(wc2*(n-alpha+eps))-sin(wc1*(n-alpha+eps)))./(pi*(n-alpha+eps));
wr = boxcar(N);
wt=bartlett(N);
wh=hamming(N);
whn=hanning(N);
hn1 = hd.*wr';
hn2 = hd.*wt';
hn3 = hd.*wh';
hn4 = hd.*whn';
w = 0:0.01:pi;
h1 = freqz(hn1,1,w);

```



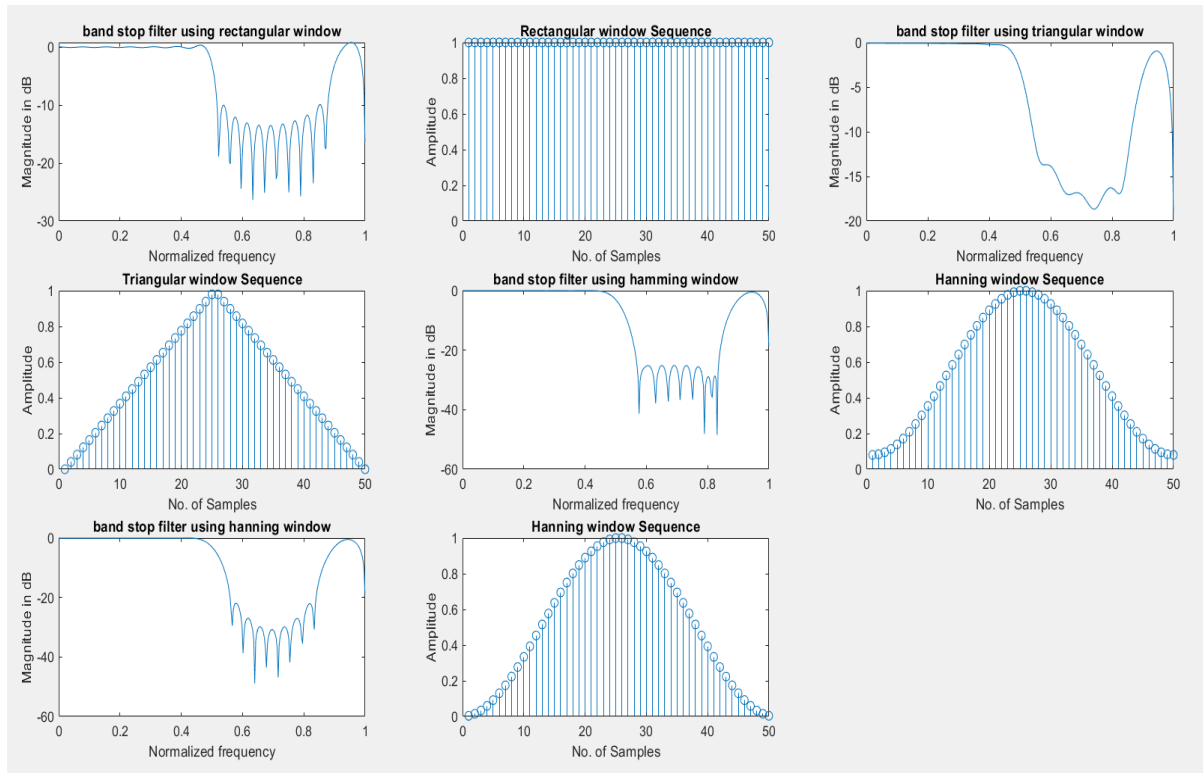
```

h2 = freqz(hn2,1,w);
h3 = freqz(hn3,1,w);
h4 = freqz(hn4,1,w);
subplot(3,3,1);
plot(w/pi,10*log10(abs(h1)));
title('band pass filter using rectangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,2);
stem(wr);
title('Rectangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,3);
plot(w/pi,10*log10(abs(h2)));
title('band pass filter using triangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,4);
stem(wt);
title('Triangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,5);
plot(w/pi,10*log10(abs(h3)));
title('band pass filter using hamming window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,6);
stem(wh);

```


Observation:

4.BANDSTOP FILTER



```

title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,7);
plot(w/pi,10*log10(abs(h4)));
title('band pass filter using hanning window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,8);
stem(whn);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');

```

4. BANDSTOP FILTER

```

clc;
clear all;
close all;
wc1=0.5*pi;
wc2=0.9*pi;
N = 50;
alpha = (N-1)/2;
eps = 0.001;
n = 0:1:N-1;
hd =(sin(wc1*(n-alpha+eps))-sin(wc2*(n-alpha+eps))+sin(pi*(n-
alpha)))/(pi*(n-alpha+eps));
wr = boxcar(N);
wt=bartlett(N);
wh=hamming(N);
whn=hanning(N);
hn1 = hd.*wr';

```



```

hn2 = hd.*wt';
hn3 = hd.*wh';
hn4 = hd.*whn';
w = 0:0.01:pi;
h1 = freqz(hn1,1,w);
h2 = freqz(hn2,1,w);
h3 = freqz(hn3,1,w);
h4 = freqz(hn4,1,w);
subplot(3,3,1);
plot(w/pi,10*log10(abs(h1)));
title('band stop filter using rectangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,2);
stem(wr);
title('Rectangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,3);
plot(w/pi,10*log10(abs(h2)));
title('band stop filter using triangular window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,4);
stem(wt);
title('Triangular window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,5);
plot(w/pi,10*log10(abs(h3)));

```



```

title('band stop filter using hamming window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,6);
stem(wh);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');
subplot(3,3,7);
plot(w/pi,10*log10(abs(h4)));
title('band stop filter using hanning window');
xlabel('Normalized frequency');
ylabel('Magnitude in dB');
subplot(3,3,8);
stem(whn);
title('Hanning window Sequence');
xlabel('No. of Samples');
ylabel('Amplitude');

```

Result:

Implemented various FIR filters using different windows

- 1.Low Pass Filter
- 2.High Pass Filter
- 3.Band pass Filter
- 4.Band stop Filter

FAMILIARIZATION OF THE ANALOG AND DIGITAL INPUT AND OUTPUT PORTS OF DSP BOARD

Aim:

Familiarization of the analog and digital input and output ports of DSP Boards.

Theory:

TMS 320C674x DSP CPU

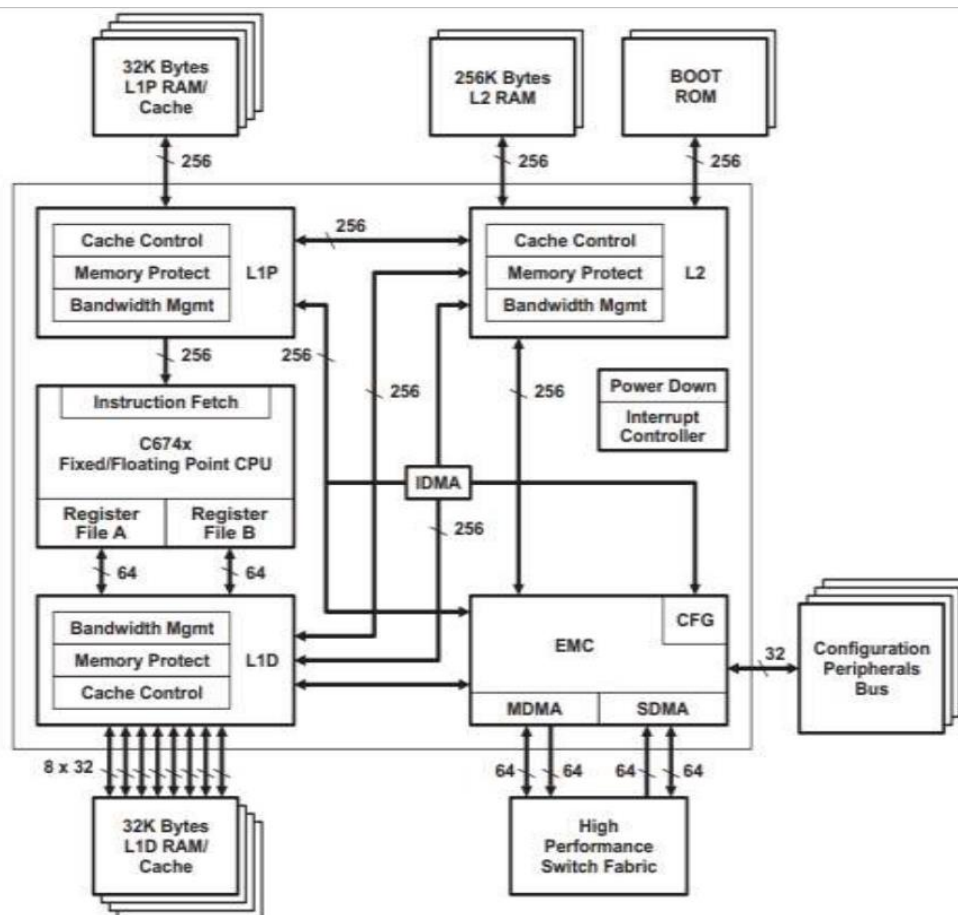


FIGURE: TMS320C 674X DSP CPU BLOCK DIAGRAM

The TMS320C674X DSP CPU consists of eight functional units, two register files, and two data paths as shown in Figure. The two general-purpose register files (A and B) each contain 32 32-bit registers for a total of 64 registers. The general-purpose registers can be used for data or can be data address pointers. The data types supported include packed 8-bit data, packed 16-bit data, 32-bit data, 40-bit data, and 64-bit data. Values larger than 32 bits, such as 40-bit-long or 64-bit-long values are stored in register pairs, with the 32 LSBs of data placed in an even register and the remaining 8 or 32 MSBs in the next upper register (which

is always an odd-numbered register). The eight functional units (.M1, .L1, .D1, .S1, .M2, .L2, .D2, and .S2) are each capable of executing one instruction every clock cycle. The .M functional units perform all multiply operations. The .S and .L units perform a general set of arithmetic, logical, and branch functions. The .D units primarily load data from memory to the register file and store results from the register file into memory.

Multichannel Audio Serial Port (McASP):

The McASP serial port is specifically designed for multichannel audio applications.

Its key features are:

- Flexible clock and frame sync generation logic and on-chip dividers
- Up to sixteen transmit or receive data pins and serializers
- Large number of serial data format options, including: – TDM Frames with 2 to 32 time slots

per frame (periodic) or 1 slot per frame (burst) – Time slots of 8,12,16, 20, 24, 28, and 32 bits

– First bit delay 0, 1, or 2 clocks – MSB or LSB first bit order – Left- or right-aligned datawords

within time slots

- DIT Mode with 384-bit Channel Status and 384-bit User Data registers
- Extensive error checking and mute generation logic
- All unused pins GPIO-capable
- Transmit & Receive FIFO Buffers allow the McASP to operate at a higher sample rate by making it more tolerant to DMA latency.
- Dynamic Adjustment of Clock Dividers – Clock Divider Value may be changed without resetting the McASP. The DSK board includes the TLV320AIC23 (AIC23) codec for input and output.

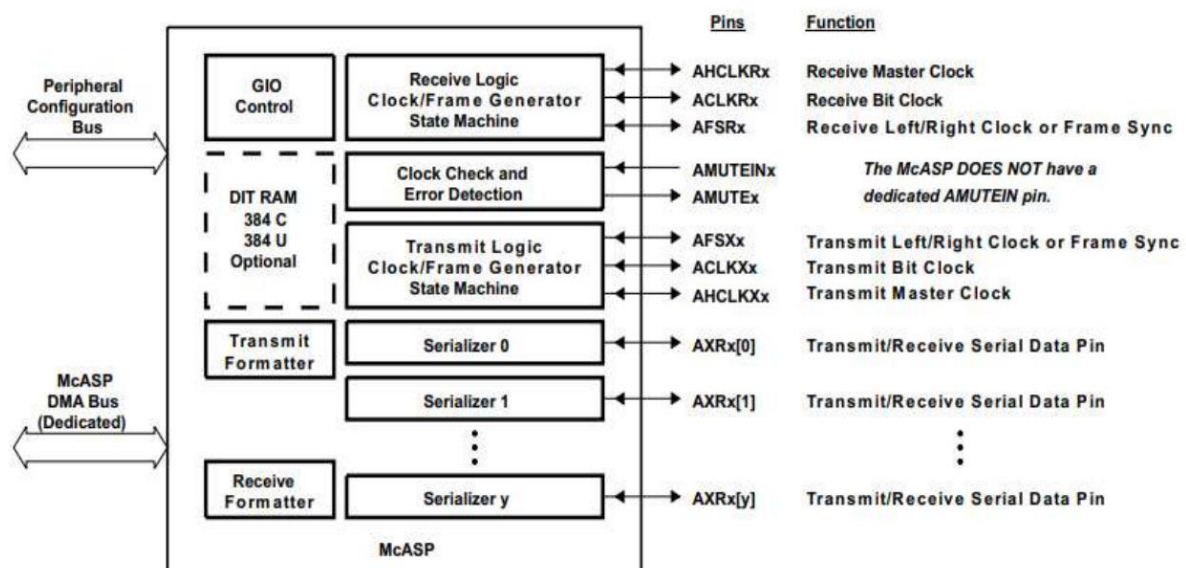
The ADC circuitry on the codec converts the input analog signal to a digital representation to be processed by the DSP. The maximum level of the input signal to be converted is determined

by the specific ADC circuitry on the codec, which is 6 V p-p with the onboard codec. After the

captured signal is processed, the result needs

to be sent to the outside world. DAC, which performs the reverse operation of the ADC. An output filter smooths out or reconstructs the output signal. ADC, DAC, and all required filtering

functions are performed by the single-chip codec AIC23 on board the DSK.



Result:

Familiarized the input and output ports of dsp board.

GENERATION OF SINE WAVE USING DSP KIT

Aim:

To generate a sine wave using DSP Kit.

Theory:

Sinusoidal are the smoothest signals with no abrupt variation in their amplitude, the amplitude witnesses gradual change with time. Sinusoidal signals can be defined as a periodic signal with waveform as that of a sine wave. The amplitude of sine wave increases from a value of 0 at 0° angle to a maximum value of 1 at 90° , it further reaches its minimum value of -1 at 270° and then returns to 0 at 360° . After any angle greater than 360° , the sinusoidal signal repeats the values so we can say that period of sinusoidal signal is 2π i.e. 360° . If we observe the graph, we can see that the amplitude varying gradually with a maximum value of 1 and a minimum value of -1. We can also observe that the wave begins to repeat its value after a period or angle value of 2π hence periodicity of sinusoidal signal is 2π .

$$y(t) = A \sin(\omega t + \phi) + C$$

Procedure:

1. Open Code Composer Studio, Click on File - New – CCS Project

Select the Target – C674X Floating point DSP , TMS320C6748 , and

Connection – Texas Instruments XDS 100v2 USB Debug Probe and Verify.

Give the project name and select Finish.

2. Type the code program for generating the sine wave and choose

File – Save As and then save the program with a name including 'main.c'.

Delete the already existing main.c program.

3. Select Debug and once finished, select the Run option.

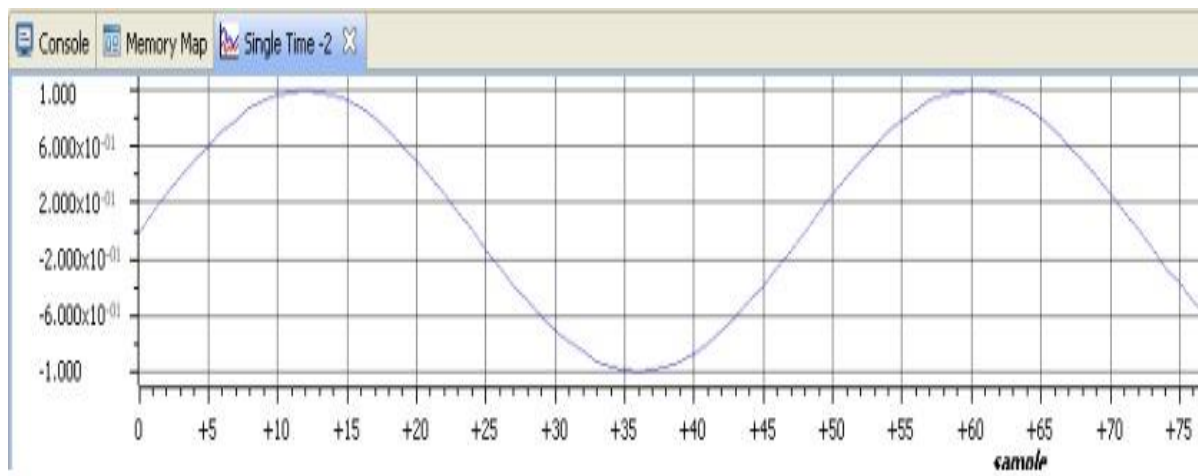
4. From the Tools Bar, select Graphs – Single Time.

Select the DSP Data Type as 32-bit Floating point and time display unit as second(s).

Change the Start address with the array name used in the program(here,s).

5. Click OK to apply the settings and Run the program or click Resume in CCS.

Observation:



Program:

```
#include<stdio.h>
#include<math.h>
#define pi 3.14159
float s[100];
void main()
{
    int i;
    float f=100, Fs=10000;
    for(i=0;i<100;i++)
        s[i]=sin(2*pi*f*i/Fs);
}
```

Result:

Generated sine wave using DSP Kit.

LINEAR CONVOLUTION USING DSP KIT

Aim:

To perform linear convolution of two sequences using DSP Kit.

Theory:

Linear convolution is one of the fundamental operations used extensively in signal and system in electrical

engineering. It has applications in areas like audio processing, signal filtering, imaging, communication systems and

more. In simple terms, linear convolution is the process of combining two signals or functions to produce a third

signal or function.

Formally, the linear convolution of two functions $f(t)$ and $g(t)$ is defined as:

The formula for linear convolution of two discrete signals $x[n]$ and $h[n]$ is given by:

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] \cdot h[n - k]$$

In the context of linear convolution in DSP, this operation is applied to digital signals. DSP systems utilize algorithms to perform convolution efficiently, often leveraging Fast Convolution methods to handle large datasets and real-time processing.

Procedure:**1. Set Up New CCS Project**

Open Code Composer Studio.

Go to File → New → CCS Project.

Target Selection: Choose C674X Floating point DSP, TMS320C6748.

Connection: Select Texas Instruments XDS 100v2 USB Debug Probe.

Name the project and click Finish.

2. Write and Configure the Program

Write the C code for generating and storing a sine wave, configuring it to access data at specified memory locations.

Assign the input Xn and filter Hn values to specified addresses:

Xn: Start at 0x80010000, populate subsequent values at offsets like 0x80010004 for each additional input.

Hn: Start at 0x80011000 with similar offsets for additional values.

Lengths of Xn and Hn should be defined at 0x80012000 and 0x80012004, respectively.

3. Configure Output Location in Code

In the code, configure the output to store convolution results at specific memory addresses starting from 0x80013000, with each result at an offset of 0x04.

4. Save the Program

Go to File → Save As and save the code with a filename like main.c.

Remove any default main.c program that might exist in the project.

5. Build and Debug the Program

Select Debug to build and load the program on the DSP.

Once the build is complete, select Run to execute.

6. Execute and Verify Output

In the Debug perspective, click Resume to run the code.

Use the Memory Browser in Code Composer Studio to verify the output at the memory location

0x80013000:

Check 0x80013000 for the first convolution result, 0x80013004 for the second, and so on.

Cross-check the values with the expected convolution results for accuracy.

Program:

```
#include<fastmath67x.h>
#include<math.h>
void main()
{
    int *Xn,*Hn,*Output;
    int *XnLength,*HnLength;
    int i,k,n,l,m;
    Xn=(int *)0x80010000; //input x(n)
```

Observation:

Xn

0x80010000 – 1

0x80010004 – 2

0x80010008 – 3

Hn

0x80011000 – 1

0x80011004 – 2

XnLength

0x80012000 – 3

HnLength

0x80012004 – 2

Output

0x80013000 – 1

0x80013004 – 4

0x80013008 – 7

0x8001300C – 6

```

Hn=(int *)0x80011000; //input h(n)
XnLength=(int *)0x80012000; //x(n) length
HnLength=(int *)0x80012004; //h(n) length
Output=(int *)0x80013000; // output address
l=*XnLength; // copy x(n) from memory address to variable l
m=*HnLength; // copy h(n) from memory address to variable m
for(i=0;i<(l+m-1);i++) // memory clear
{
Output[i]=0; // o/p array
Xn[l+i]=0; // i/p array
Hn[m+i]=0; // i/p array
}
for(n=0;n<(l+m-1);n++)
{
for(k=0;k<=n;k++)
{
Output[n] =Output[n] + (Xn[k]*Hn[n-k]); // convolution operation.
}
}
}

```

Result:

Performed Linear Convolution using DSP Kit.