

## Experiment-7

### Common Source Amplifier

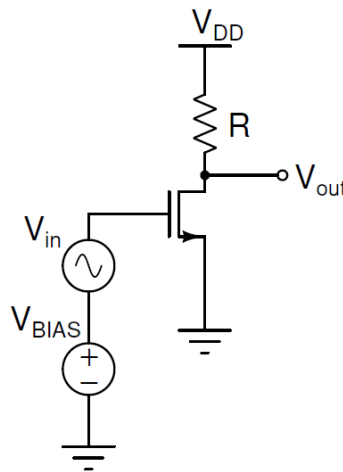
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#### Part1:

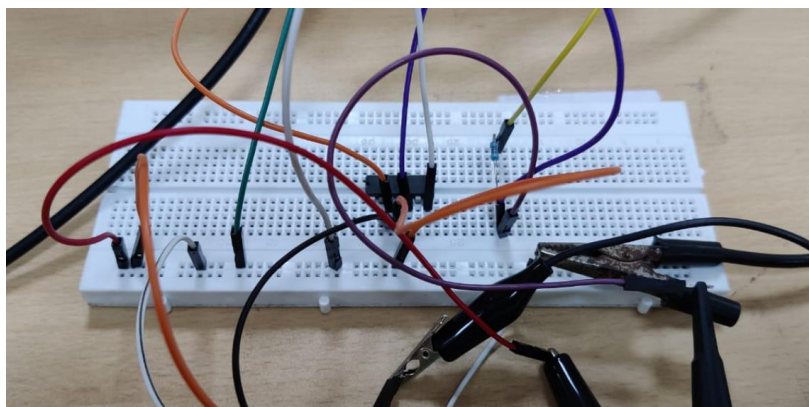
Study of body-effect on gain of CS amplifier.

Schematic view of the circuit:



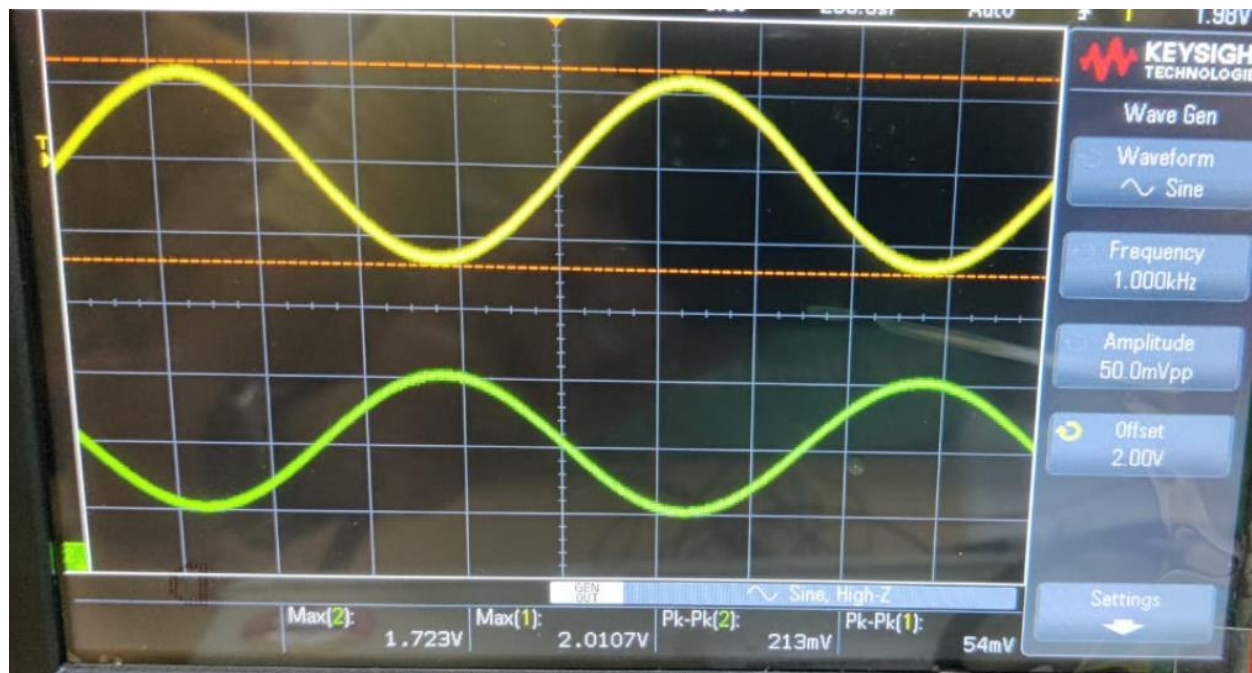
Given:  $V_{DD} = 5V$ ,  $V_{BIAS} = 2V$ ,  $V_T = 1.8V$ ,  $R_L = 4.7k\Omega$ .

Circuit:



(a) Initially the Body terminal (pin 7) is grounded.

$$V_{DD} = 5V, V_{BIAS} = 2V, V_T = 1.8V.$$



Channel 1 is connected to  $V_{GS}$  and output is connected over channel 2, therefore for  $V_{GS} = 2V$  the  $V_{DS} = 1.723V$ .

$$I_{DS} = V_{DD} - V_{DS}/R_L$$

$$\text{Therefore } I_{DS} = 95.04 \mu A$$

$$\text{And using } I_{DS} = \mu_n C_{ox} W/L (V_{GS} - V_T)^2.$$

$$\text{Let: } \mu_n C_{ox} W/L = K.$$

$$I_{DS} = K (V_{GS} - V_T)^2.$$

$$I_{DS} = K(0.44)^2(0.44).$$

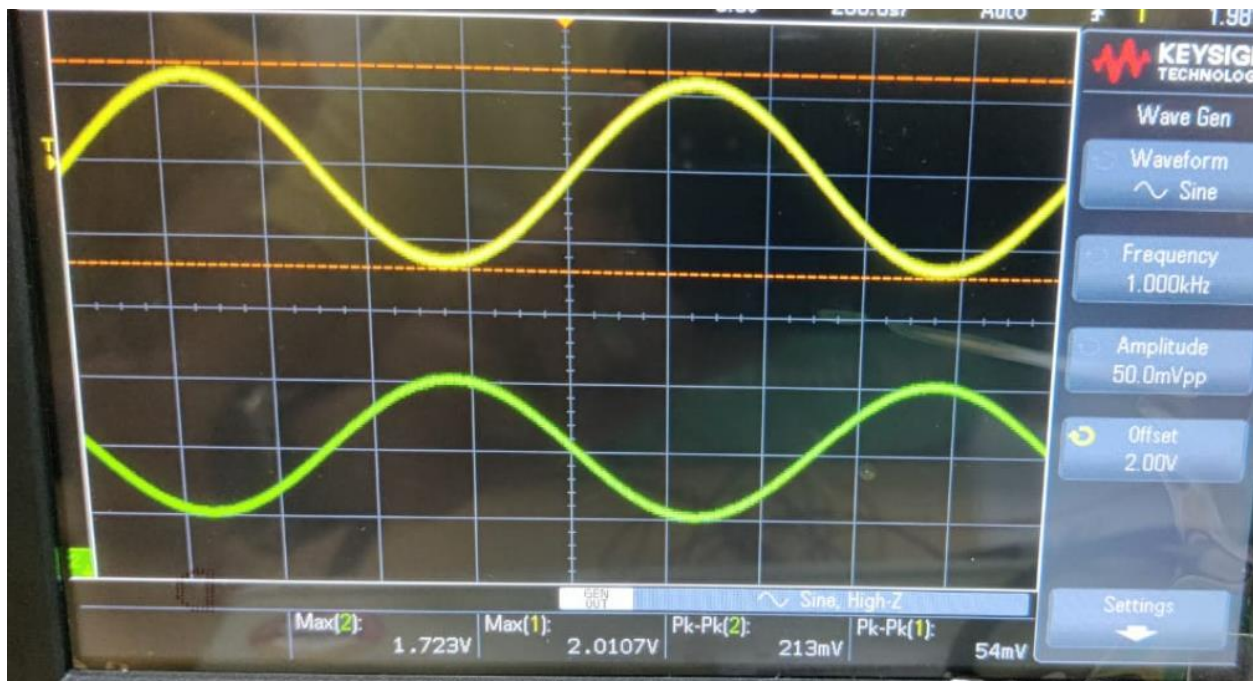
$$95.04 \mu A / 0.1936 = K$$

$$K = 4.909 \times 10^{-4} A/V^2.$$

$$g_m = K \cdot 0.45$$

$$g_m = 2.209 \times 10^{-4} A/V$$

(b) AC signal: 50mVpp

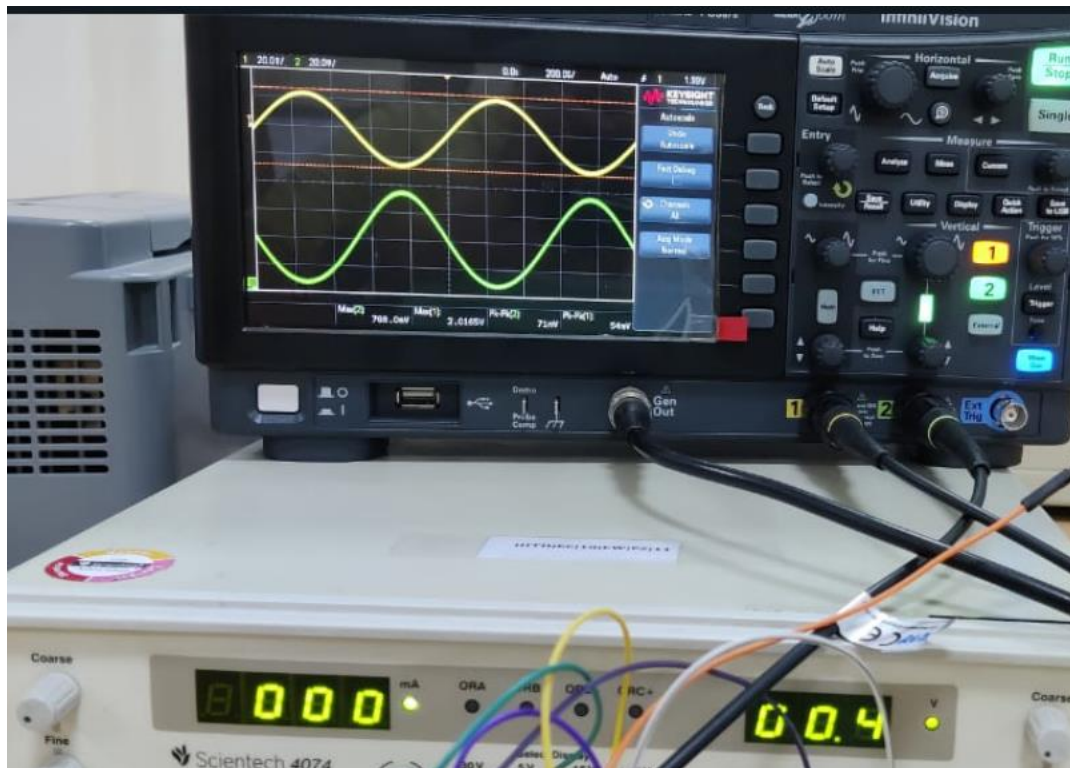


Gain (Practical):  $213/54 = 3.944 = g_m \times R_L$

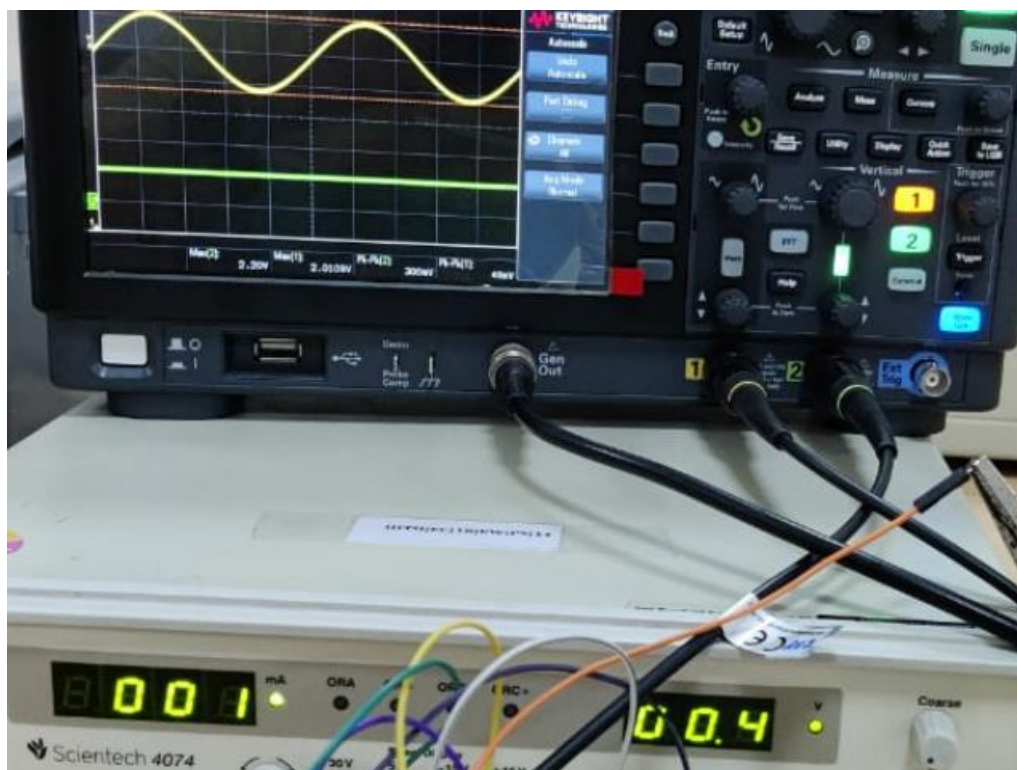
$0.83 \times 10^{-3} = 8.3 \times 10^{-4}$ .

(C) In order to see body effect, we connect  $V_{SS}$  to DC supply, on increasing  $V_{SS}$  value,  $V_{TH}$  decrease hence the gain should increase, this is because, as  $V_{SS}$  increases, more electrons flow into p-substrate, hence MOS requires low voltage to cause the inversion.

$$V_{SS} = 0.4$$



$$V_{SS} = -0.4$$





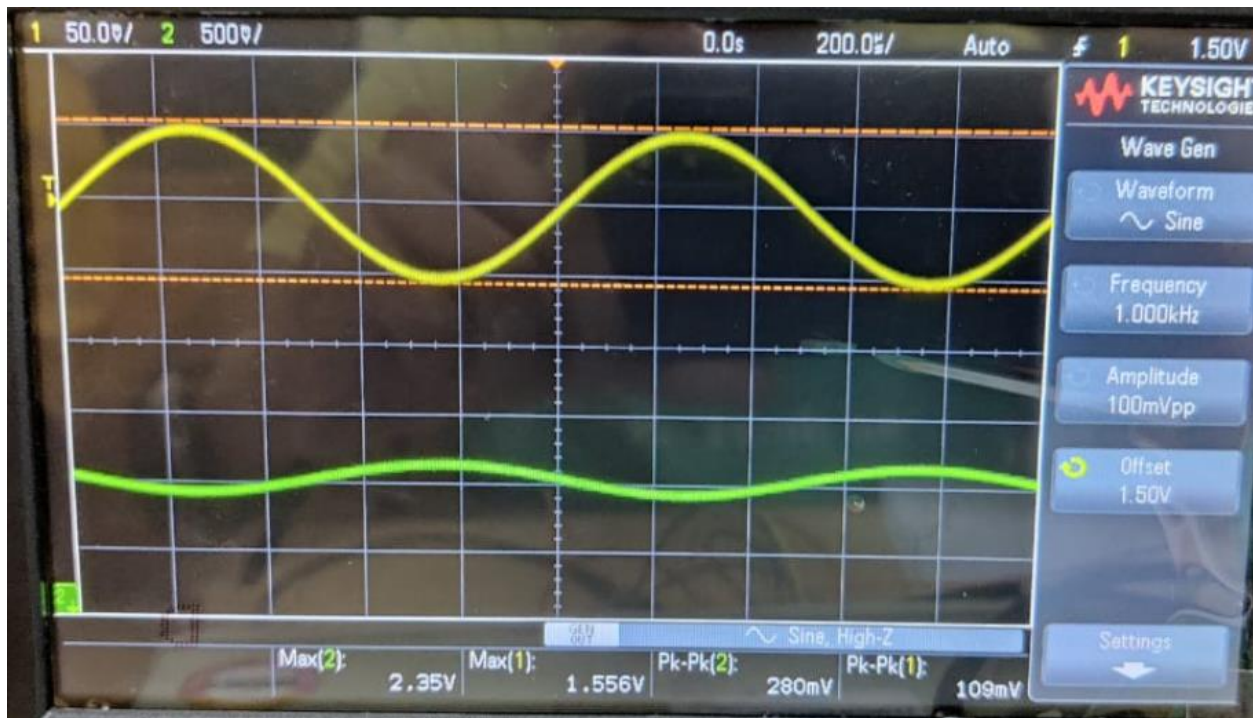
$V_{SS}$	$V_{OUT}$	Gain	Gm
0	213mV	4.26	$0.906 \cdot 10^{-3}$
0.4	400mV	8	$1.7021 \cdot 10^{-3}$
-0.4	72mV	1.44	$0.0306 \cdot 10^{-3}$

## 2. Effect of BIAS points on Gain of common source amplifier

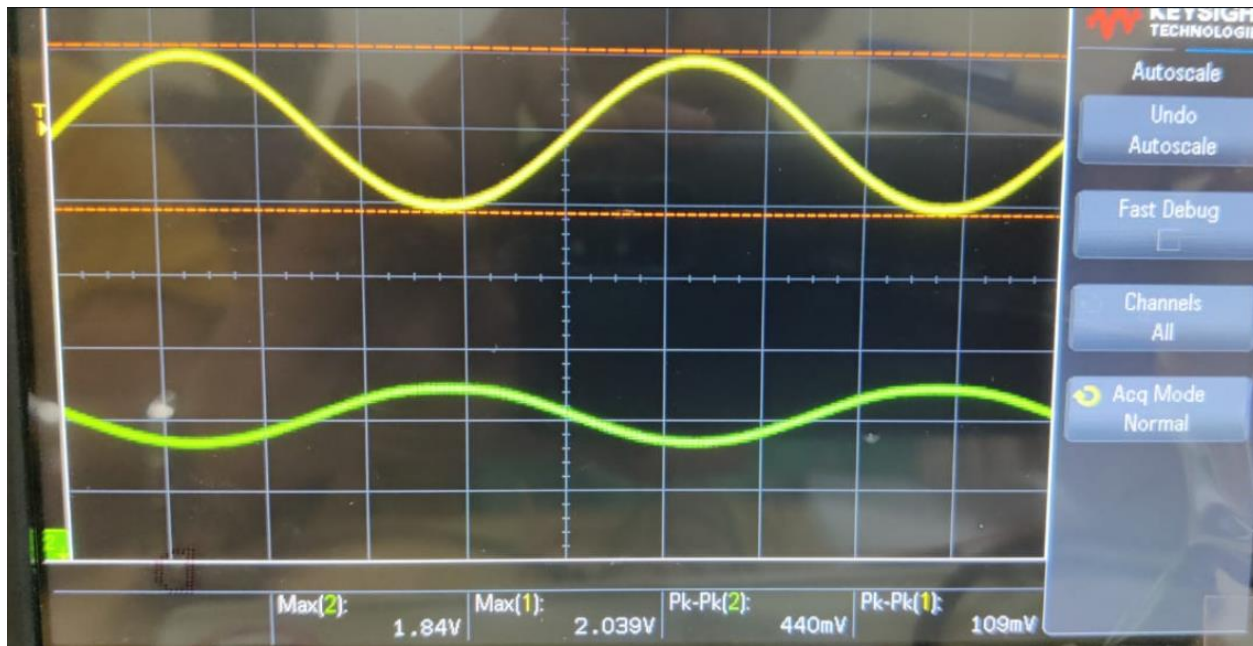
Now we will increase the offset value of V-Bias, with AC amplitude=100m<sub>VPP</sub>

$V_{BIAS}$	$V_{OUT}(AC)$	Gain	$Gm \cdot 10^3$
1.5	280mV	2.8	0.5957
2	440mV	4.4	0.9361

$V_{BIAS} = 1.5V$

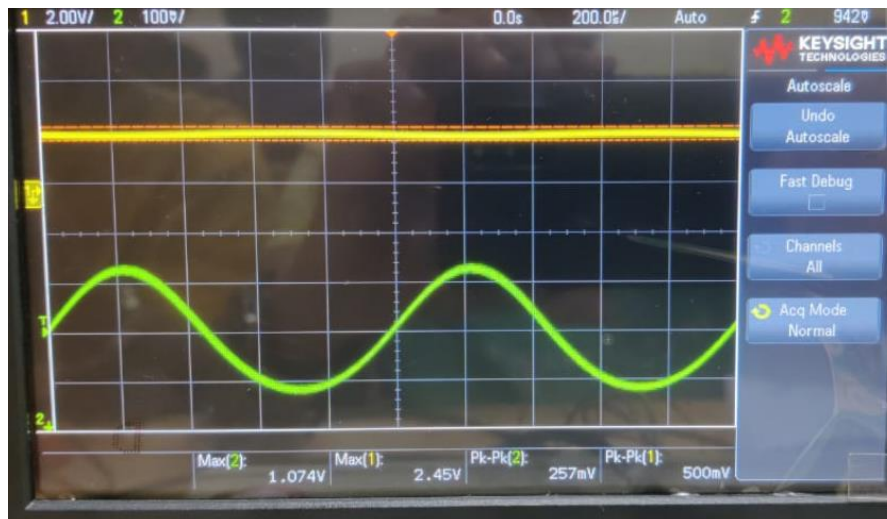


$$V_{\text{BIAS}} = 2\text{V}$$



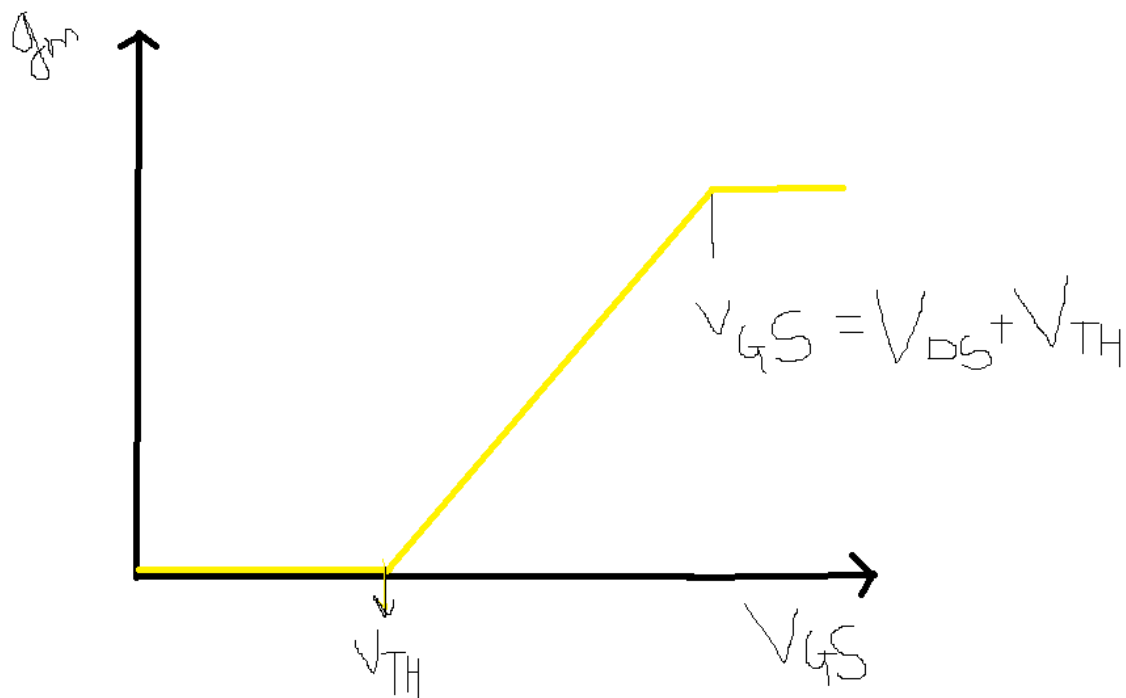
We are not able to increase  $V_{\text{BIAS}}$  any more due to probe restriction, as we increase it more, the  $V_{\text{BIAS}}$  shoot's up to 500mVpp.

The abnormality is shown below:



Explanation for above  $g_m$  trend is, when  $V_{GS}$  is less than  $V_{TH}$  there is nearly 0 gain as current is negligible, as the circuit enters saturation mode, the  $g_m$  increases linearly with  $V_{GS}$ , this is because  $g_m = k(V_{GS} - V_{TH})$  in saturation region. Then in linear region it is constant with  $V_{GS}$  and linearly increases with  $V_{DS}$ .

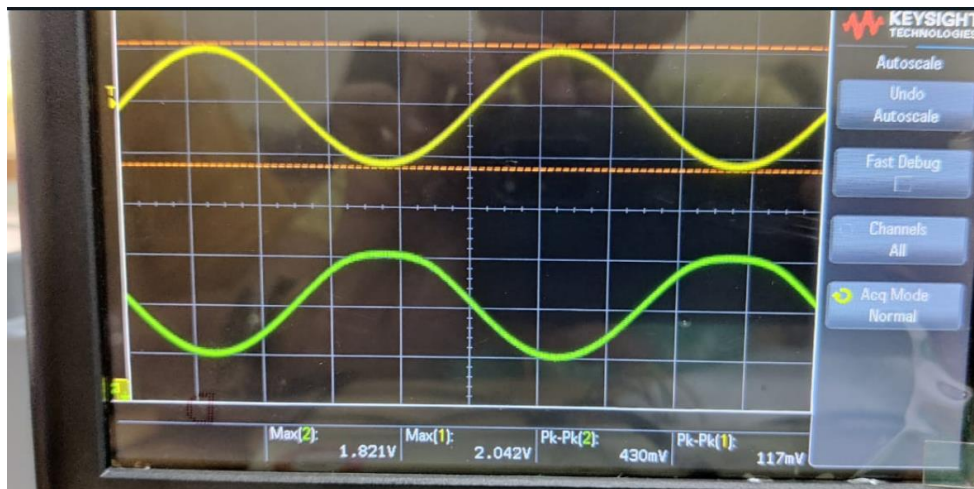
Graph:



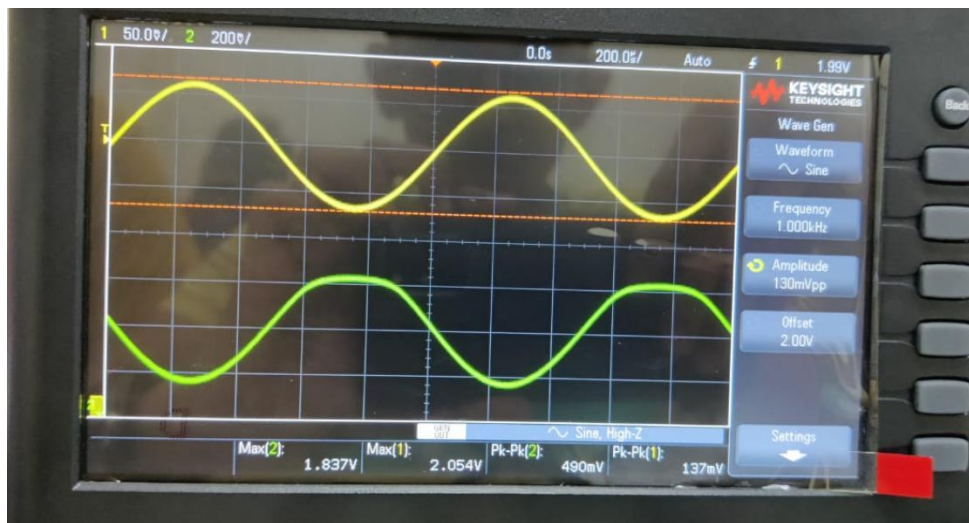
### 3. Effect of small signal input swing on gain of common source amplifier

For low values of ( $V_{BIAS}$ ) AC the output comes out to be a perfect sine wave, but as we increase the input, the wave starts getting clipped, this is because due to higher AC voltage the saturation condition  $V_{GS} - V_{TH} \leq V_{DS}$  is violated at amplitude =  $V_{DS} + V_{TH}$ , and so the circuit, enters the linear mode.

$(V_{\text{BIAS}}) \text{ AC} = 110\text{mVpp}$

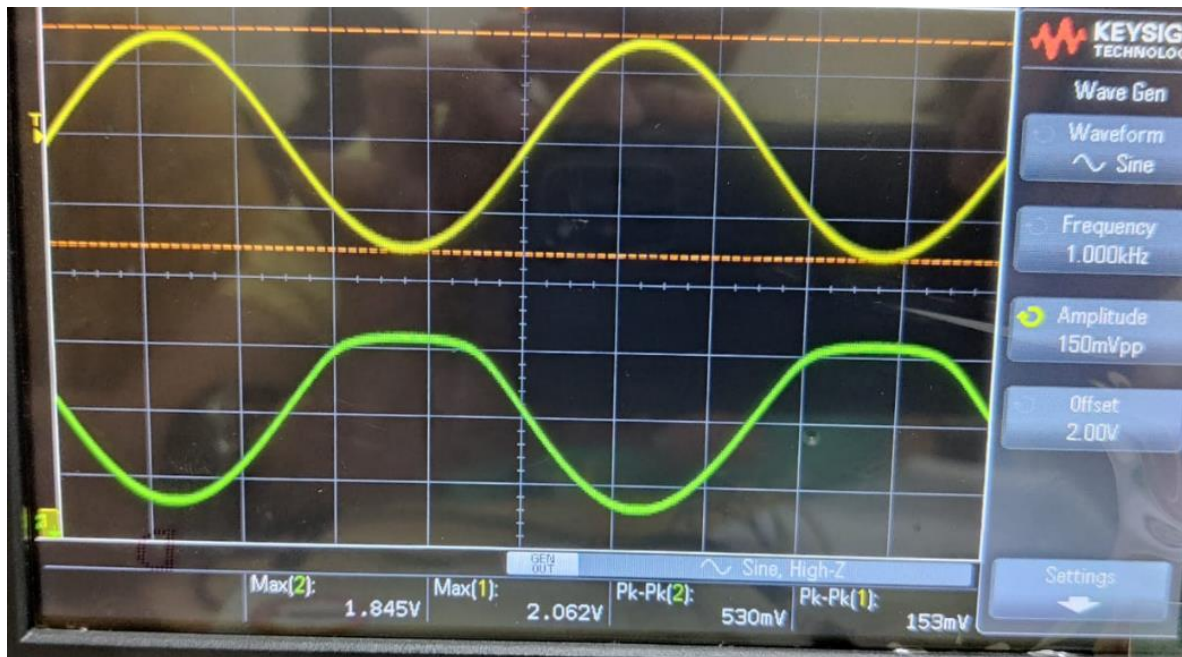


$(V_{\text{BIAS}}) \text{ AC} = 130\text{mVpp}$

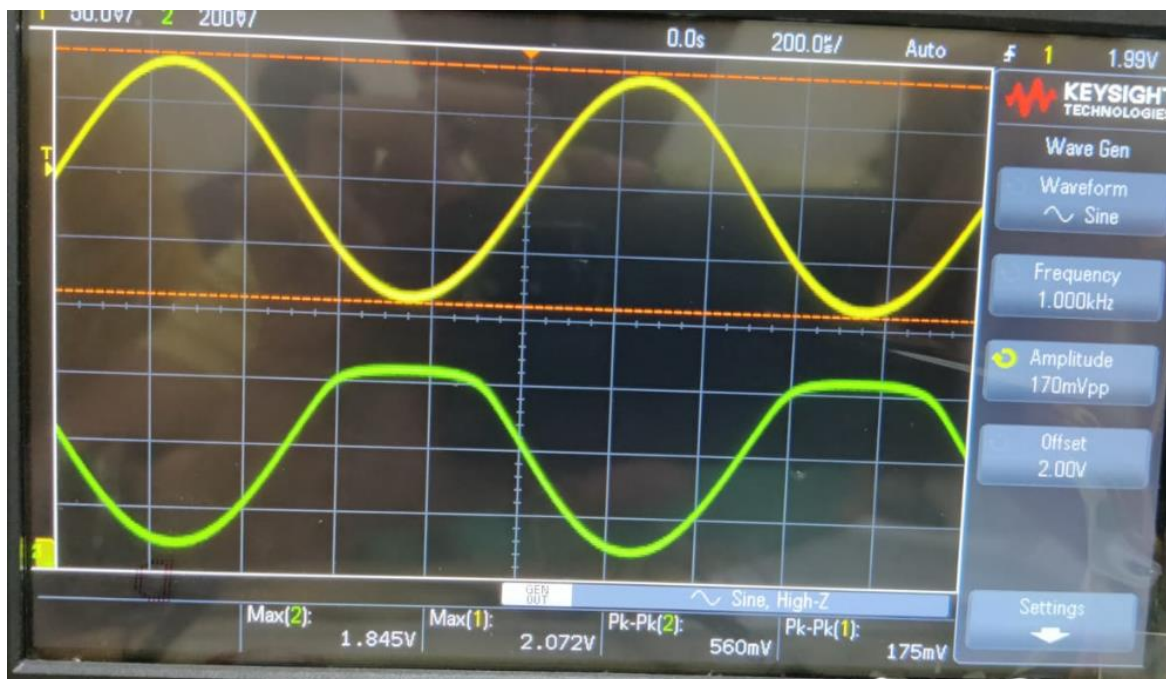


$(V_{\text{BIAS}}) \text{ AC} = 150\text{mVpp}$

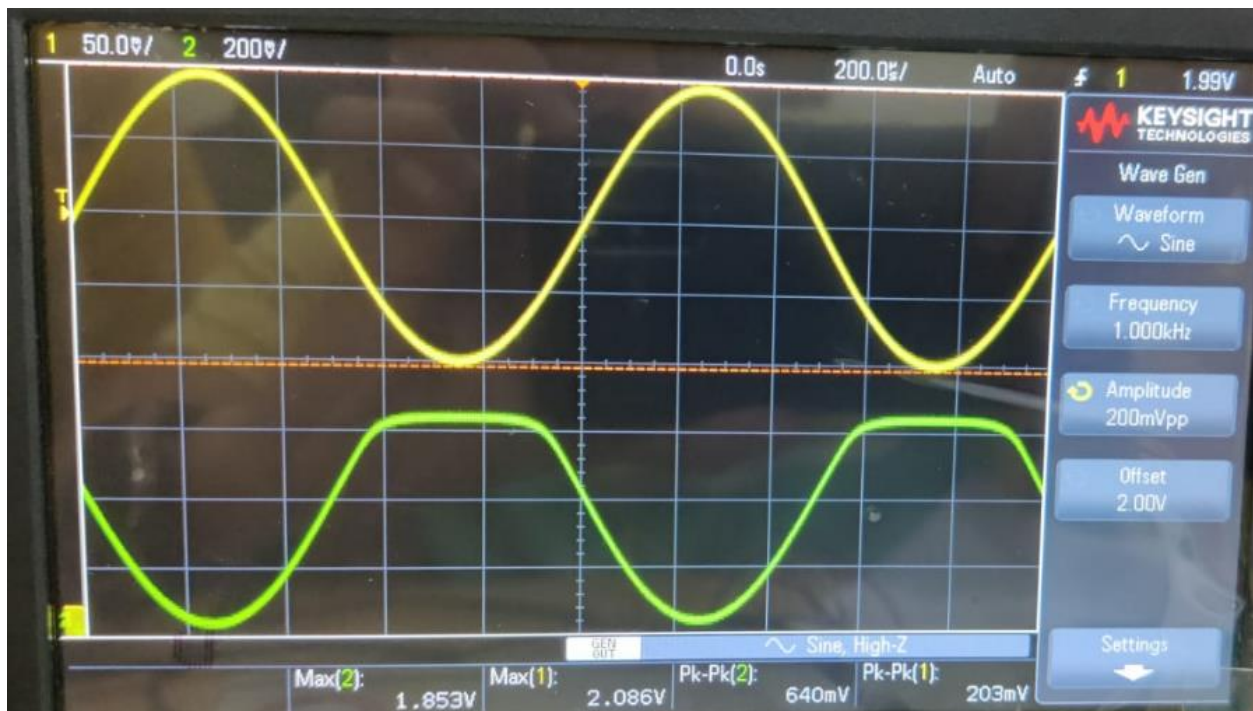




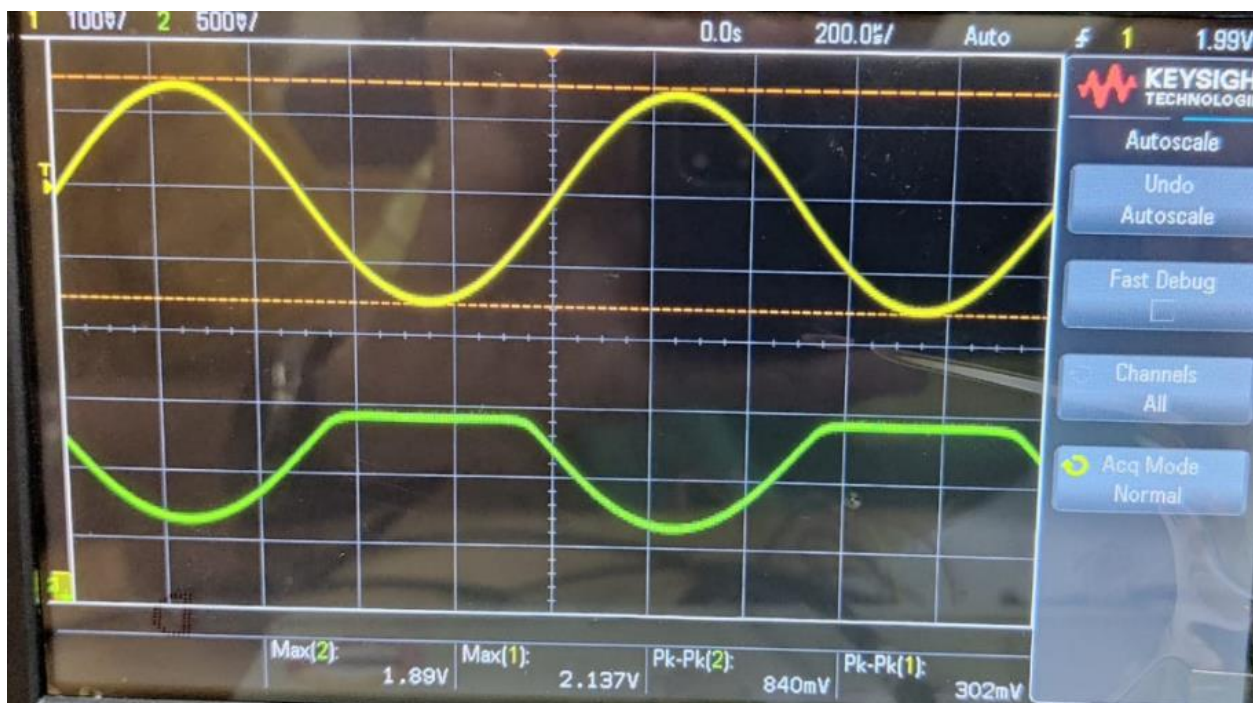
$(V_{\text{BIAS}}) \text{ AC} = 175\text{mVpp}$



$(V_{\text{BIAS}}) \text{ AC} = 200\text{mVpp}$



$(V_{\text{BIAS}})_{\text{AC}} = 300\text{mVpp}$

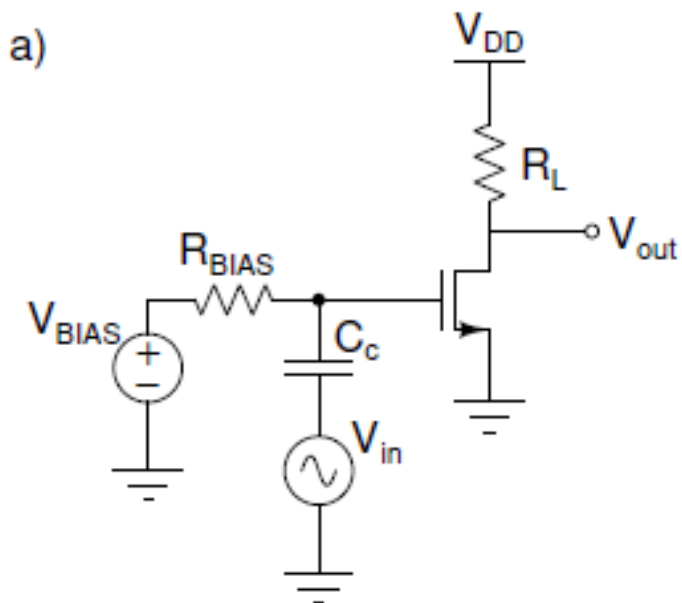


$V_{\text{INPUT AC}}$	$V_{\text{OUTPUT AC}}$	Gain	$g_m * 10^3$
110mV	440mV	4	0.85
135mV	430mV	3.185	0.677
150mV	530mV	3.53	0.75710
175mV	560mV	3.2	0.6808
200mV	640mV	3.2	0.6808
300mV	840mV	2.8	0.5957

As the  $V_{\text{INPUT AC}}$  the gain increases till the point there is no clipping, as the clipping starts the gain decreases due to decrease in output voltage.

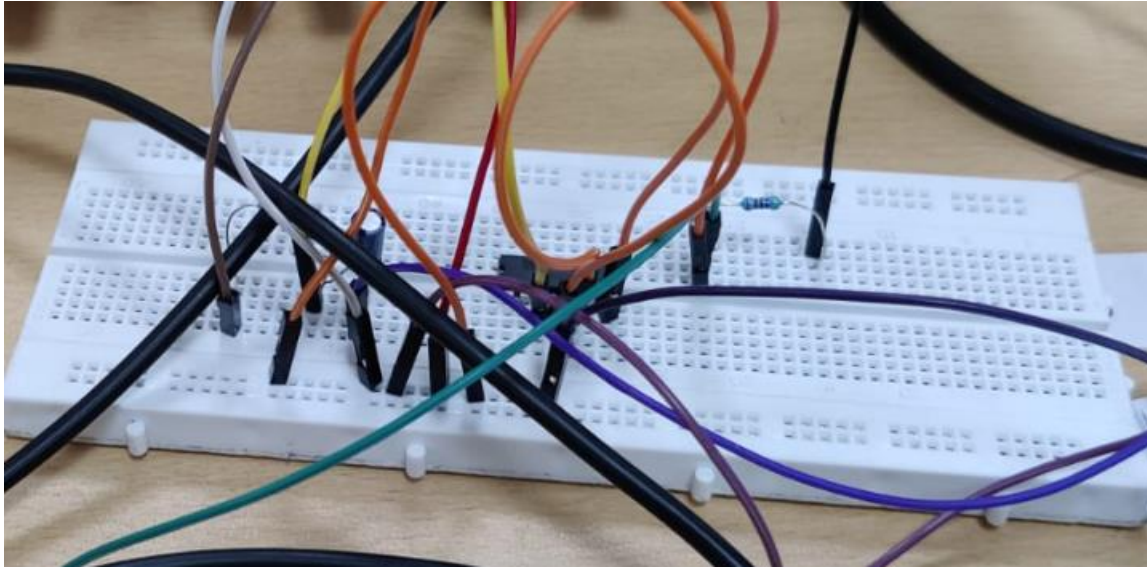
#### 4. CS amplifier with external coupling

Schematic view of the circuit:





Circuit:



The difference in gain is observable at high frequencies when due to capacitances there comes a frequency factor into play, below is the graph showing gain vs frequency.

It is evident that as the frequency increases the gain of the circuit decreases.

$$V_{in} = 107\text{mV}$$

$$V_{out} = 302\text{mV}$$

$$\text{Gain: } 2.8224$$

$$G_m = 0.6 * 10^{-3}.$$

The value of gain and  $g_m$  decreased this is because of frequency term in the denominator of the gain.

$$-g_m * R_L / (1 + C_C * R_{BIAS} * \omega).$$

$$V_{DS} = 1.5\text{V}.$$

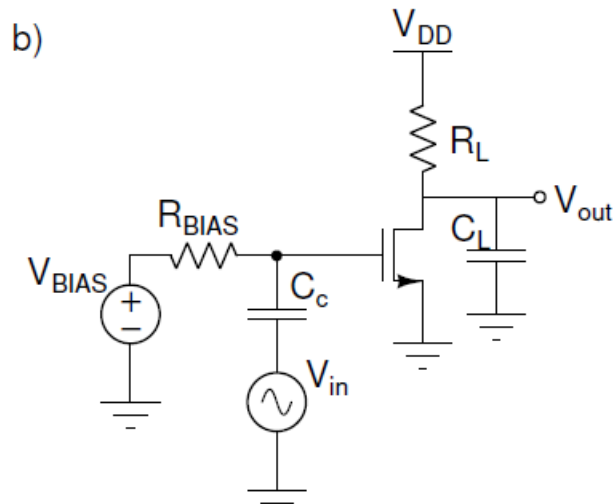
$$I_{DS} = (V_{DD} - V_{DS}) / R_L$$

$$I_{DS} = 1.06 * 10^{-2}.$$



$$\mu C_{ox} \frac{W}{L} = I_{DS} / (V_{DS} - V_{TH})^2 = 5.47 \cdot 10^{-2}.$$

(d) Schematic view of circuit:



-3dB bandwidth: 513KHz.

Transfer function:  $-g_m R_L / (1 + \omega R_{BIAS} C_C)(1 + \omega R_L C_L)$

Poles:  $1/R_{BIAS} C_C 2\pi$  and  $1/R_L C_L 2\pi$

Pole1: 0.15KHz

Pole2: 0.07KHz