

Audio Signal Processing For Guitar Pedal Application

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1 Equalizer

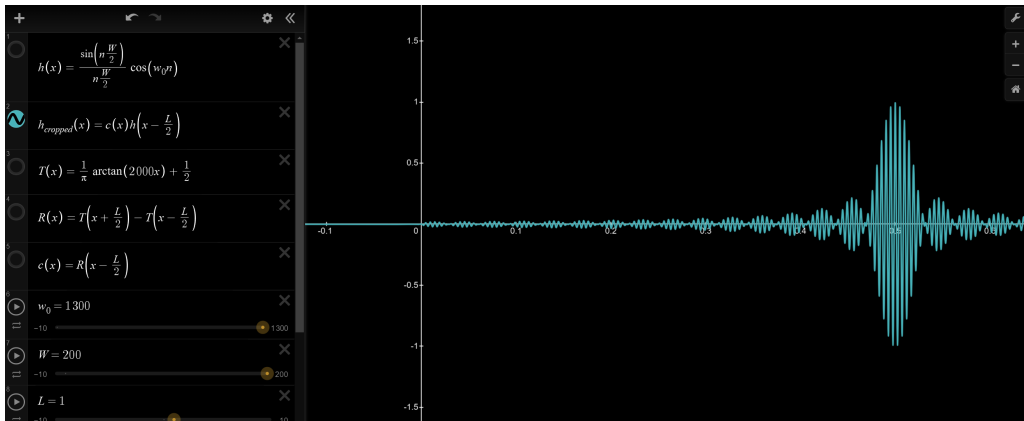
Equalizers are comprised of various band pass filters. A band pass filter may be represented in the Fourier domain as

$$\begin{aligned}
 H(j\omega) &= \text{rect}\left(\frac{1}{W}(\omega - \omega_0)\right) + \text{rect}\left(\frac{1}{W}(\omega + \omega_0)\right) \\
 h[n] &= \mathfrak{F}_{\text{Discrete}}^{-1}\{H(j\omega)\}(t) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(j\omega) e^{-jn\omega} d\omega \\
 h[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \left(\Theta\left(\omega + \omega_0 + \frac{W}{2}\right) - \Theta\left(\omega - \omega_0 - \frac{W}{2}\right) + \Theta\left(\omega - \omega_0 + \frac{W}{2}\right) - \Theta\left(\omega + \omega_0 - \frac{W}{2}\right) \right) e^{-jn\omega} d\omega \\
 h[n] &= \frac{1}{2\pi} \left(\int_{\omega_0 - \frac{W}{2}}^{\omega_0 + \frac{W}{2}} e^{-jn\omega} d\omega + \int_{-\omega_0 - \frac{W}{2}}^{-\omega_0 + \frac{W}{2}} e^{-jn\omega} d\omega \right) \\
 h[n] &= \frac{1}{2\pi nj} \left(e^{-jn(\omega_0 + \frac{W}{2})} - e^{-jn(\omega_0 - \frac{W}{2})} + e^{-jn(-\omega_0 + \frac{W}{2})} - e^{-jn(-\omega_0 - \frac{W}{2})} \right) \\
 h[n] &= \frac{1}{2\pi nj} (e^{jn\frac{W}{2}} - e^{-jn\frac{W}{2}})(e^{jn\omega_0} + e^{-jn\omega_0}) \\
 h[n] &= \frac{2}{\pi n} \sin\left(\frac{W}{2}n\right) \cos(\omega_0 n) = \frac{W}{\pi} \frac{\sin(\frac{W}{2}n)}{\frac{W}{2}n} \cos(\omega_0 n) \\
 h[n] &= \frac{W}{\pi} \text{sinc}\left(\frac{nW}{2}\right) \cos(\omega_0 n)
 \end{aligned}$$

There are some problems with this solution, though. Most importantly, the system is non-causal. To make it causal, we need to shift our signal and multiply it by some *cropping* function, $c[n]$.

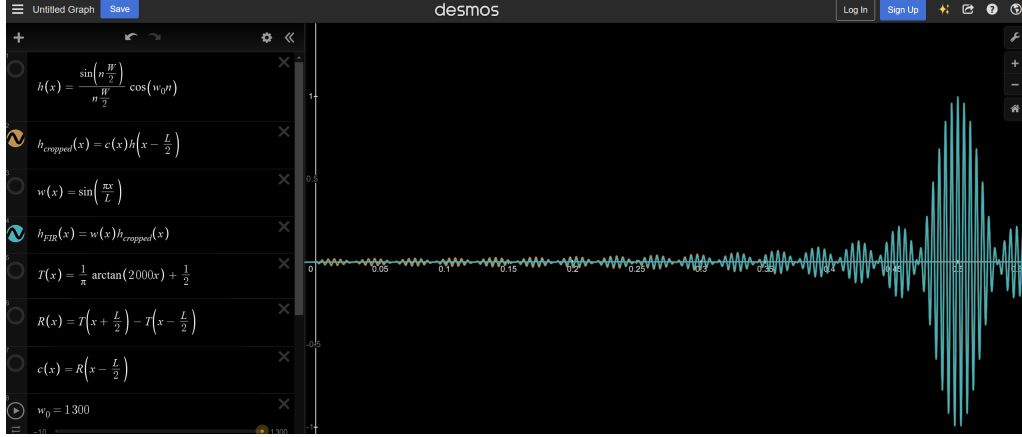
$$\begin{aligned}
 c[n] &= \Theta\left(n + \frac{L}{2}\right) - \Theta\left(n - \frac{L}{2}\right) \\
 h_{\text{cropped}}[n] &= c[n] \times (h[n] * \delta(L/2)) \\
 h_{\text{cropped}}[n] &= \left(\Theta\left(n + \frac{L}{2}\right) - \Theta\left(n - \frac{L}{2}\right) \right) \frac{W}{\pi} \text{sinc}\left(\frac{(n - \frac{L}{2})W}{2}\right) \cos\left(\omega_0\left(n - \frac{L}{2}\right)\right)
 \end{aligned}$$

For my specifications ($\omega_0 = 1300$, $W = 200$) and using $L = 1$, we get the following cropped sinc:



But this introduces yet ANOTHER problem: discontinuities. These greatly affect the frequency response. As a solution, we multiply by a *window* function which effectively smooths out the function. I will use the sine window, $w[n] = \sin\left(\frac{n\pi}{L}\right)$. We thus get the following function and accompanying graph (with the orange being the cropped function, and the blue being the windowed):

$$h_{\text{FIR}}[n] = \sin\left(\frac{n\pi}{L}\right) \left(\Theta\left(n + \frac{L}{2}\right) - \Theta\left(n - \frac{L}{2}\right) \right) \frac{W}{\pi} \text{sinc}\left(\frac{(n - \frac{L}{2})W}{2}\right) \cos\left(\omega_0\left(n - \frac{L}{2}\right)\right)$$



We're now at the finish line: we just need to calculate the output,

$$y[n] = (x * h)[n] := \sum_{m=-M}^M x[n-m]h[m]$$

The final equation representing our system, which is characterized by ω_0 and W , is:

$$y[n] = \sum_{m=-M}^M x[m] \sin\left(\frac{n\pi}{L}\right) \left(\Theta\left(n + \frac{L}{2}\right) - \Theta\left(n - \frac{L}{2}\right) \right) \frac{W}{\pi} \text{sinc}\left(\frac{(n - \frac{L}{2})W}{2}\right) \cos\left(\omega_0\left(n - \frac{L}{2}\right)\right)$$