

LAB-6

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6.1 b:

Given, $x(t) = e^{jt}$

On calculating its FT, we get

$$X(j\omega) = \frac{4 \sin(2\omega)}{2\omega}$$

- As the input signal is rectangular pulse of unit amplitude, therefore its continuous time FT results in **sinc function**.
- The sinc is completely **real**.
- Hence, its **imaginary** part is **0**.
- The **phase** plot continuously shifts from $\pi/2$ to $-\pi/2$ due to shift in $X(j\omega)$ values from positive to negative values in **oscillatory** fashion.

6.1 c:**Observations:**

- On changing T from 2 to 1, the **signal is compressed** in time by factor of 2 while its **Fourier transform is expanded** by the same factor.
- Also there is **change in amplitude**.
- On changing T from 2 to 4, the **signal is expanded** in time by factor of 2 while its **Fourier transform is compressed** by the same factor.
- Also there is **change in amplitude**.

FT Property:

Frequency Scaling Property: If a function is expanded in time by a quantity a , the Fourier Transform is compressed in frequency by same amount and vice-versa.

Hence, if $x(t)$ -----FT-----> $X(jw)$

Then $x(at)$ -----FT-----> $\frac{1}{|a|}X(\frac{jw}{|a|})$

6.1 d:

(a). Given, $x(t) = e^{jt}$

On calculating its FT, we get

$$X(j\omega) = \frac{2 \sin(\pi(1-\omega))}{1-\omega}$$

Observations:

- The shape is **sinc function**
- $\text{Im}(X(j\omega))$ is **0** because $X(j\omega)$ is purely real as calculated.

(b). Given, $x(t) = \cos t$

$$= (e^{jt} + e^{-jt})/2$$

On calculating its FT, we get

$$X(j\omega) = \frac{2\omega \sin(\pi\omega)}{1-\omega^2}$$

Observations:

- The shape is symmetric **sinc function**.
- From the symmetry, we observe that **$X(j\omega)$ is even** which is evident from the expression.
- **$\text{Im}(X(j\omega))$ is 0** because $X(j\omega)$ is purely real as calculated.

6.1 e:

(a). Given, $x(t) = 1+t$ for $t \in [-1, 0)$
 $= 1-t$ for $t \in [0, 1]$

On calculating its FT, we get

$$X(j\omega) = \left(\frac{\sin(\frac{\omega}{2})}{\frac{\omega}{2}} \right)^2$$

Observations:

- The shape is **sinc function**
- **Im $\{X(j\omega)\}$ is 0** because $X(j\omega)$ is purely real as calculated.

6.2 : A signal and its samples

The signal has been plotted.

6.3 : Fast Fourier Transform(Radix-2)

When $N=2$, $X(k) = \sum_{n=0}^{1} x(n) e^{\frac{-j\pi kn}{2}}$

where $x(n)$ is the signal obtained on decomposing the input signal into odd and even components until its length is 2.

6.4 : Filtering of periodic signals with LTI systems

6.4 a:

- The relationship between input and output signal can be given as:

$$\mathbf{b_k=a_k(H(jw))}$$

- From the relation, we observe that there is **no change in periodicity** of output signal.

6.4 b:

(b).Ideal Low Pass Filter(LPF)

Given, $x(t)=\cos t$

$$=(e^{jt}+e^{-jt})/2$$

$$F\{\cos(t)\}=\pi(\delta(w + 1) + \delta(w - 1))$$

On comparing with the standard equation:

$$x(t)=\sum_{k=-N}^N a_k e^{jkt}$$

$$a_{-1}=a_1=\frac{1}{2}$$

$$a_0=0$$

0 otherwise

$$\rightarrow A=[\frac{1}{2}, 0, \frac{1}{2}]$$

Given, $w_0=1$

- The filter will only allow frequencies range such that $k\omega_0 < |\omega_c|$
- Also k ranges between -1 to 1
- Hence, allowed frequencies = $\omega_c < -1$ or $\omega_c > 1$

Case I: $\omega_c = 2$

- Hence $b_k = a_k$ for $k = -1, 0, 1$

Case II: $\omega_c = 0.5$

- Hence $b_k = 0$ for $k = -1, 0, 1$

(c). Ideal High Pass Filter (HPF)

Given, $x(t) = \cos t$

$$= (e^{jt} + e^{-jt})/2$$

As calculated above,

$$\rightarrow A = \left[\frac{1}{2}, 0, \frac{1}{2}\right]$$

Given, $\omega_0 = 1$

- The filter will only allow frequencies range such that $k\omega_0 > |\omega_c|$
- Also k ranges between -1 to 1
- Hence, allowed frequencies = $-1 < \omega_c < 1$

Case I: $\omega_c = 2$

- Hence $b_k = 0$ for $k = -1, 0, 1$

Case II: $w_c=0.5$

- Hence $b_k=a_k$ for $k=-1,0,1$

(d).Non-Ideal Filter(LPF)

Given, frequency response, $H(w)=\frac{G}{a+w}$ where $a>0, G>0$

Observations:

- The output signal has a **phase shift**
- The **amplitude** of output signal is **decreased**.

Complex valued nature of LTI system

- Complex valued nature of LTI system corresponds to **phase shift in frequency domain**.
- On taking inverse FT, this corresponds to **time scale shift in time domain**.

6.5 : Reconstruction methods**Comparison between the quality of three interpolations**

- The quality of interpolated signal is **not uniform** throughout.
- The signal is **better** interpolated in **middle** than at ends.
- If we ignore ends,
Quality of interpolation: **sinc>Linear>Zero Order**
- End error occurs due to **unavailability of sufficient samples at the end**.

Maximum Absolute Error(MAE) between original and reconstructed signal in interval [0.25,1.75]

- **Zero order hold:** 3
- **Linear hold:** 0.3827
- **Sinc:** 0.6350

This is evident from above explanation.

6.6 : Sampling of non-band limited signal

t_samples vector: $-1 : T_s : 1$ where T_s is sampling interval.

Observations:

- **On decreasing sampling interval, sampling frequency increases.**
- **Hence, the number of samples increases and signal is better reconstructed.**

6.7 : Audio signals

filename	/Users/ashish...
CompressionM...	'Uncompressed'
NumChannels	2
SampleRate	8000
TotalSamples	268237
Duration	33.5296
Title	'Impact Moderato'
Comment	[]
Artist	'Kevin MacLeod'
BitsPerSample	16

Rest explanation is given in code.

6.8 : Aliasing

Given, $x(t)=\cos(5\pi t)$

Nyquist rate: $2*5\pi=10\pi$ rad/sec

$$\omega_s=2\pi/T_s$$

Observations:

- On **increasing sampling interval**, the **error** or deviation between original and reconstructed signal **increases**,

Reason:

- When $T_s \leq 0.2$ s or $\omega_s \geq 10\pi$ rad/sec, the **sampling frequency is above the Nyquist rate**. Hence the signal is **properly reconstructed**.
- As the **sampling frequency decreases**, the signal is **imperfectly reconstructed** which is evident from the plot.

