

# Memory BIST Training Workbook

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Contractor/manufacture is:  
Mentor Graphics Corporation  
8005 S.W. Boeckman Road, Wilsonville, Oregon 97070-7777.

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# About This Training Workbook

## Introduction

This course is designed to be a one-day, self-paced training class. The student will use this workbook and run exercises to become familiar with memory Built-In-Self-Test (BIST) concepts. The following are the top level course goals:

- The student will understand the Memory BIST design processes
- The student will gain experience with Mentor Graphics MBISTArchitect™ and Memory BIST-In-Place™ tools
- The student will understand how to find information and problem-solve typical design issues

If taken in its entirety, this training course is intended to introduce design engineers to the V8.2002\_1 version of the Mentor Graphics MBISTArchitect and Memory BIST-In-Place tools.

# Audience

## Primary Audience

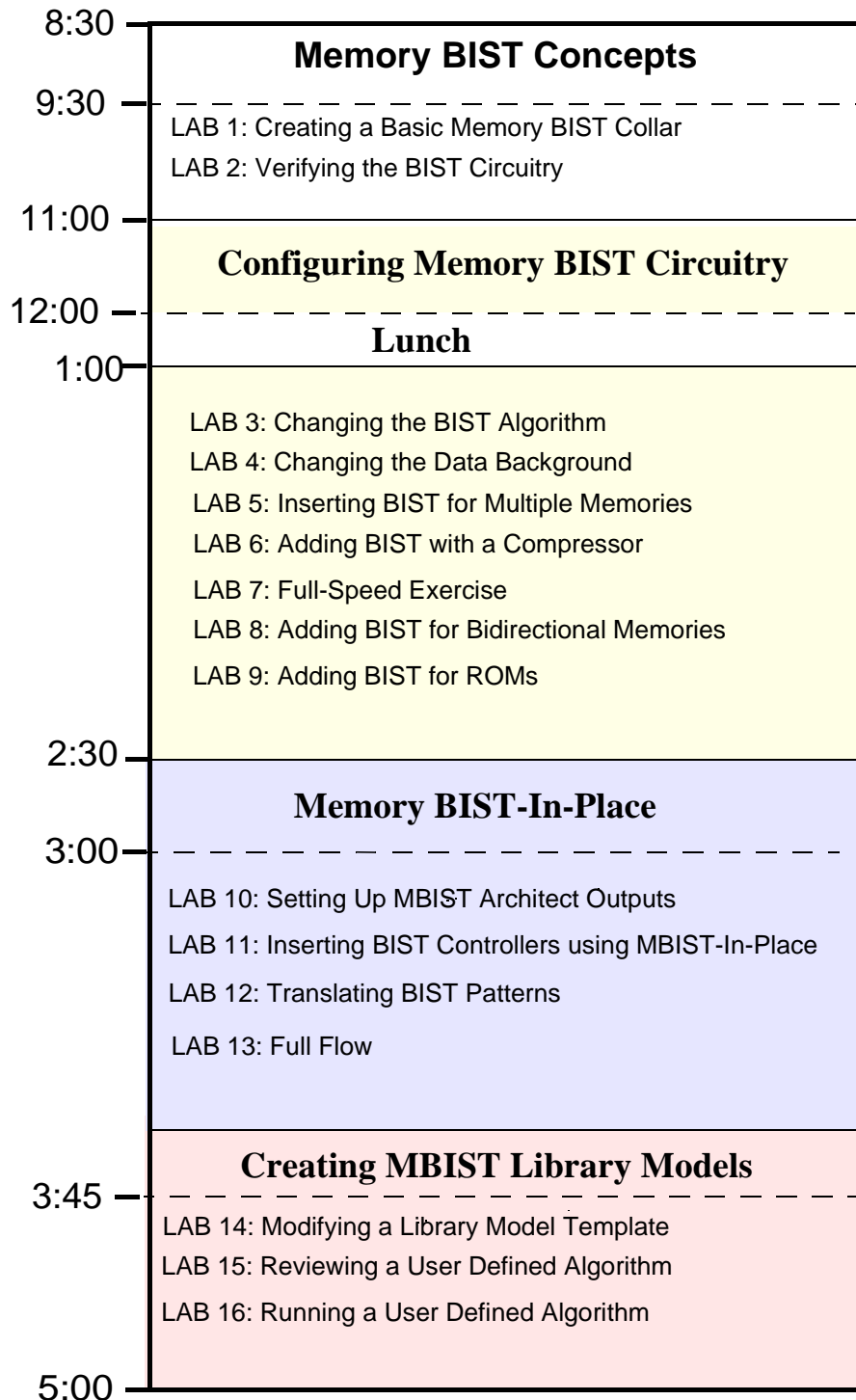
The target student profile is the Electronic Design Engineer using synthesis tools to develop synchronous digital designs. It is assumed that students will be using MBISTArchitect, and optionally, Memory BIST-In-Place tools. This type of student will comprise about 80% of the course attendees and will have the following characteristics:

- They have some limited familiarity with DFT terminology and concepts.
- They are interested in learning how to add memory BIST to their designs.
- As they work with these DFT tools, these engineers want to know “what is this tool doing to my design” (or my design flow) and “how do I control what the tool is doing to my design?”
- These engineers want to know how to analyze the tool-generated reports and modify the tool setup constraints to achieve the test goals that may be imposed on them by their organization.
- These engineers want to be well grounded in the basic tool process flow and be able to respond appropriately when the tools report “problems.”

## Secondary Audience

About 20% of the students will be “test engineers.” These engineers are typically members of a manufacturing test group or an internal CAD group that provides support for design engineers. Test engineers are typically well grounded in their understanding of DFT terms and concepts, but may not have had much experience with DFT tools.

## Course Timeline



# Course Overview

The course is divided into the following five parts:

## Module 1 Memory BIST Concepts

The first module introduces various types of memories, memory BIST concepts, memory testing and fault types.

## Module 2 Generating a Memory BIST

This module introduces you to the typical memory BIST flow, inputs and outputs to MBISTArchitect, and the role of the test bench. It also introduces you to the MBISTArchitect graphical user interface (GUI) and user documentation for memory BIST tools. The lab exercises will give you practice generating a BIST collar and verifying the circuit.

## Module 3 Common BIST Variations

This module highlights a variety of options you can use to customize the memory BIST circuitry to your design. The lab exercises cover tasks you may use when adding memory BIST to your design such as inserting BIST for multiple memories or adding BIST when you have a compressor, ROM, or bidirectional memories. A number of lab exercises are included here to give you a variety of choices. Generally, you will not be expected to complete them all.

## Module 4 Memory BIST-In-Place

This module gives you a basic understanding of how to create, connect, and integrate BIST structures using the Mentor Graphics Memory BIST-In-Place tool. The lab exercises at the end of this module will give you experience in running through the process flow of Memory BIST-In-Place.

## Module 5 Memory Modeling for MBISTArchitect

This module explains how memory devices are modeled inside MBISTArchitect. The lab exercises are designed to give you practice creating a memory model in case your company does not already have the model you are looking for. It also

introduces you to the Mentor Graphics User Defined Algorithm™ function that can be used to generate your own March-type algorithms.

## Prerequisite Knowledge

Prerequisite knowledge in DFT fundamentals is required. The purpose of requiring prerequisites is to (1) reduce “learning overload” which can happen early in the course and (2) help the students move quickly toward learning tool concepts and best practices for getting results.

Generic DFT concepts and terminology can be learned from sources outside Mentor Graphics.

## Acronyms Used in This Workbook

The following is an alphabetical list of the acronyms used in this workbook:

ASIC - Application Specific Integrated Circuit  
ATE - Automatic Test Equipment  
ATPG - Automatic Test Pattern Generation  
AVI - ASIC Vector Interfaces  
BIST - Built-In Self-Test  
BSDL - Boundary Scan Design Language  
CTAF - Core Test Access File  
CTDL - Core Test Description Language  
CUT - Circuit Under Test  
DFT - Design-for-Test  
DRC - Design Rules Checking  
DUT - Device Under Test  
GUI - Graphical User Interface  
HDL - Hardware Description Language  
JTAG - Joint Test Action Group  
LFSR - Linear Feedback Shift Register  
LSB - Least Significant Bit  
MCM - Multi-Chip Module  
MISR - Multiple Input Signature Register

MSB - Most Significant Bit  
PRPG - Pseudo-Random Pattern Generator  
RTL - Register Transfer Level  
SCOAP - Sandia Controllability Observability Analysis Program  
SFP - Single Fault Propagation  
TAP - Test Access Port  
TCK - Test Clock  
TDI - Test Data Input  
TDO - Test Data Output  
TMS - Test Mode Select  
TRST - Test Reset  
WDB - Waveform Data Base

## Customer Support Information

Additional help is available from Mentor Graphics Customer Support using the following phone numbers, email address, and internet site:

<b>DirectConnect</b> (M-F: 6am-5:30pm, PST)	1-800-547-4303
<b>SupportCenter Fax</b>	1-800-684-1795
<b>SupportNet-Email</b>	support_net@mentor.com
<b>SupportNet-Web site</b>	<a href="http://www.mentor.com/supportnet">http://www.mentor.com/supportnet</a>
<b>Mentor DFT Web site</b>	<a href="http://www.mentor.com/dft">http://www.mentor.com/dft</a>

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# Module 1

## Memory BIST Concepts

When you complete this module, you should have a basic understanding of memory testing and memory BIST concepts.

### Objectives

Upon completion of this module, you will be able to:

- List when and why to use Memory BIST (MBIST)
- List the basic advantages and disadvantages of Memory BIST
- Describe some of the common fault types associated with memory testing
- List the common algorithms used by MBIST Architect to test memories

# Embedded Memories

## Embedded Memories

- ♦ **Memory Built-In-Self-Test (MBIST) has been used successfully for years to solve the test issues for embedded memories.**
- ♦ **Most of today's designs contain embedded memories—features associated with memories:**
  - **Memory can consume a large design portion**
  - **Memories are dense, resulting in high defect rates**
  - **Embedded memory quality is critical to whole chip quality**
  - **Memories can have high operating speeds**
  - **Embedded memories can be difficult to exercise efficiently with functional testing**

Most of today's designs contain embedded memories. Here are some common side effects of using embedded memories in chips today:

- **Memory can consume a large design portion and result in high defect rates**

In many designs today, memories may take up a large portion of the design. See [“Typical Architecture with Embedded Memories” on page 1-4](#) for more information on architecture and test options.
- **Embedded memories can be difficult to exercise efficiently with functional or other types of testing.**

Large, complex circuits often contain difficult-to-test portions of logic. Even the most testable designs, if large, can require extensive test generation time, tester pattern memory, and tester application times—all of which are expensive, yet necessary, to adequately test devices in a classic test scenario. Memory BIST solves the problems associated with functional

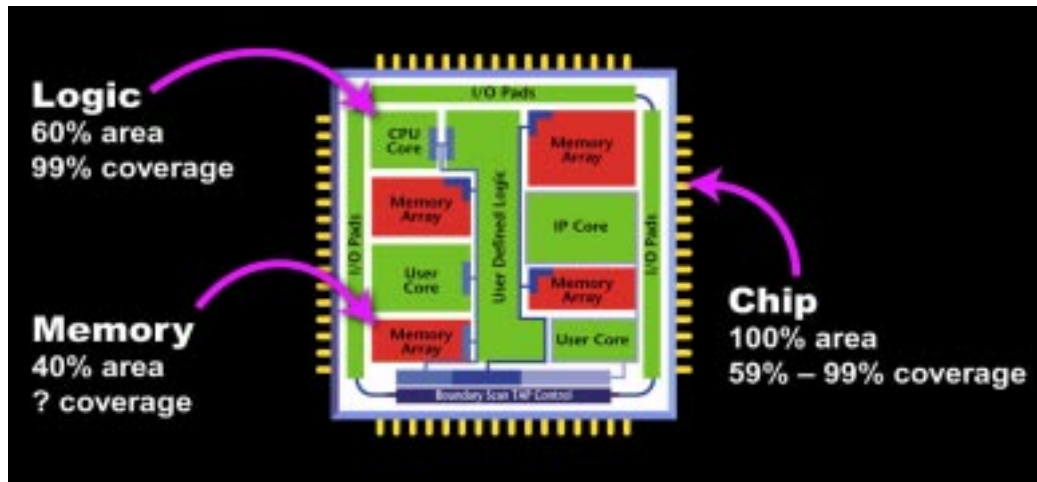


testing, see [“Advantages of Adding BIST” on page 1-13](#) for an overview of memory BIST advantages.

- **Typical ATE testing may not adequately test memories.**  
The bullets in these slides describe features of memories. In Lesson 3, we'll talk about how a memory can fail and what kinds of patterns need to be used to test them well.
- **Memories can have high operating speeds**  
Memories of all sorts, but especially high speed memories, are susceptible to speed-related defects. To ensure high quality memory tests, you need to test for these sorts of defects by running at-speed memory tests. See the [“Full-Speed Overview” on page 3-14](#) for information on using MBISTArchitect to test memory at full access speed.

# Typical Architecture with Embedded Memories

## Typical Architecture with Embedded Memories



This slide shows a typical architecture of a design with embedded memories. Logic takes up 60% of the silicon while memories consume 40%. In order to ensure high quality, you need to thoroughly test these memories. For example, if your test coverage is 99%, but you don't test your memories the whole chip test coverage is much lower and your finished products will be susceptible to test escapes due to the untested memories.

# Types of Memories

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## Types of Memories

- ♦ **General Memory types**
  - Different depth and width
  - Synchronous and asynchronous
  - Multi-port
- ♦ **SRAM**
- ♦ **DRAM**
- ♦ **EPROM & EEPROM  
(Flash)**
- ♦ **ROM**

The memories listed in the slide above are common, your memory BIST circuit will be different depending upon the model selected. In general, SRAM and ROMs commonly use memory BIST to solve the test problems.

- **General Memory Types**

Memories can have different depths and widths, be synchronous and asynchronous, or be multi-port.
- **SRAM**

The most commonly used in our industry is an ASIC type of flow. We do most of our work in this tutorial on variations of an SRAM.
- **DRAM**

A DRAM is not as common as an SRAM, a special process is sometimes required to accommodate DRAMs.

- **EPROM and EEPROM (Flash)**

EPROM and EEPROM are also known as “Flash” memory. Flash is fairly common but is usually not fully functionally tested because of the extremely long access times. Specialized memory BIST could be generated but is not commonly done.

- **ROM**

ROM memories are very common and generally use a slightly different implementation for memory BIST. We’ll talk about BIST architectures in Lesson 3.

# Types of Testing

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## Types of Testing

- ♦ **Functional testing**
- ♦ **Direct access testing**
- ♦ **Memory BIST**

There are several common testing techniques used to test memories.

# Functional Testing

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## Functional Testing

- ♦ Pattern generation can be very difficult
- ♦ Verification can be time consuming
- ♦ Determining quality is difficult and time consuming
- ♦ Reduces amount of external test data to store
- ♦ No functional impacts

Large, complex circuits often contain difficult-to-test portions of logic. Large designs require extensive test generation time, tester pattern memory, and tester application times—all of which are expensive, yet necessary, to adequately test devices in a classic test scenario.

Previously, ATPG functional testing was the only way to test embedded memories. Because memory faults differ from random logic faults and memories reside within larger designs, ATPG does not provide an adequate memory testing solution. Functional testing is inadequate because pattern generation is difficult, verification is time-consuming and it is difficult to determine quality.

# Direct Access Testing

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## Direct Access Testing

- ♦ **Must Mux inputs and outputs to chip pins**
- ♦ **Pattern generation may be difficult (pattern conversion is easy)**
- ♦ **Extensive ATE memory required or memory test hardware**
- ♦ **Routing and timing issues can arise**

Another method of testing memories is direct access testing. This method is usually feasible only if you only have one or two memories and can be accessed off of a bus that is already routed at the top level of the chip. Direct access testing may require special hardware or ATE memory testing equipment. It also requires mux input and output access directly to pins. In addition, pattern generation and verification is still a problem with this method of testing.

# Memory BIST Testing

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## Memory BIST Testing

- ♦ Simplifies pattern generation
- ♦ High quality guaranteed by algorithmic patterns
- ♦ Minimal impact to timing and area

Memory BIST testing addresses memory testing problems. Memory BIST adds a layer of test circuitry around the memory. This circuitry becomes the interface between the high-level system and the memory. This interface minimizes the controllability and observability challenges of testing embedded memories. And the built-in, finite-state machine that provides the test stimulus for the memory greatly reduces the need for an external test set for memory testing.

BIST provides a memory test solution without sacrificing test quality. In many cases, BIST structures can eliminate or minimize the need for external test pattern generation (and thus, tester pattern memory) and tester application time. In addition, a designer can exercise BIST circuitry within a design, running tests at speed due to the proximity of the BIST circuitry to the memory under test. A designer can also run a memory BIST process from within higher levels of the design. See [“Advantages of Adding BIST” on page 1-13](#) for more detailed information on the advantages of using memory BIST.



# When Should You Use Memory BIST?

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## When Should You Use Memory BIST?

- ◆ **User should use Memory BIST:**
  - On medium to large embedded memories
  - On memories that are contained within Intellectual Property (IP) that will be reused
  - On memories that should be tested at speed
  - On devices with multiple embedded memories
  - On devices that are time-to-market critical
  - On devices that run on ATEs with limited capability
  - On SOCs where testing and verification will be difficult

You should use Memory BIST (MBIST):

- **Medium to Large embedded memories**

You should definitely use memory BIST testing on medium-to-large memories. Very small memories must be considered on a case by case basis. On very small arrays, the controller may be larger than the array. Small memories can also be added to an existing MBIST controller so very minimal impact is observed. Alternative solutions such as MacroTest might be a better solution.
- **Memories which are contained within Intellectual Property (IP) that will be reused**

MBIST is a very important part of the reusability and portability of IP. Once the test circuitry is built in, it can be reused and rerun wherever the IP is placed, with no additional work. You only need to ensure that the memory BIST operation on the IP is properly controlled at the chip-level.

- **Memories that should be tested at speed**

Ideally, you should test memories at the same rate or a rate greater than they will be used in the application. Additional types of memory faults will be found if the memory is exercised at full speed. MBISTArchitect has a full speed option that lets you test the memories at full speed, see the [“Full-Speed Overview” on page 3-14](#) for more information.

- **On devices with multiple embedded memories**

Memory BIST controllers can be shared across multiple arrays of different sizes with little incremental area increases. This is practical when the arrays are in relatively close proximity to each other. If arrays are far apart in the chip layout, care must be taken not to have excessive routing overhead.

- **On devices that are time-to-market critical**

Pattern generation and conversion is significantly easier with the use of memory BIST. Verification of the manufacturing patterns is streamlined by the use of an automated tool.

- **On devices that run on ATEs with limited capability**

BIST can reduce the memory, timing, and control signals an ATE would need to test a memory. This may allow the device to be tested on a simpler and cheaper tester.

- **On SOCs where testing and verification will be difficult**

Verification and test generation are the two largest challenges of SOCs. Therefore, Memory BIST improves time-to-market and first pass silicon success.

# Advantages of Adding BIST

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## Advantages of Adding BIST

- ♦ **Enables Intellectual Property (IP) reuse**
- ♦ **Reduces the routing of signals needed at the chip level**
- ♦ **Reduces test application time and simplifies pattern generation**
- ♦ **Reduces amount of test data to store**
- ♦ **Facilitates hierarchical test capabilities -- lets you easily test at model, block, design, and system levels**
- ♦ **Merges test and design, reducing development time**
- ♦ **BIST controller can be shared across memories**

Self-testing provides a number of benefits. First, placing the test circuitry on the chip itself reduces external tester time and expense. Second, it minimizes the difficulty of testing embedded circuitry by providing system-level control signals that run and report status of the test operation. Third, because the circuitry itself generates test stimulus, this eliminates or reduces expensive test pattern generation time. Likewise, it eliminates or reduces the amount of required external test data storage.

Additionally, designs with BIST facilitate hierarchical test capabilities. Hierarchical BIST lends itself to test at the model, block, design, and system levels. For example, a memory BIST controller embedded in an IC can be used to test off-the-shelf memories that are external to the chip.

BIST blends both the design and test disciplines. Merging test into the design process far earlier in the flow reduces the product development cycle.

# Disadvantages of Adding BIST

---

## Disadvantages of Adding BIST

- ♦ **Small area increase**
- ♦ **Adds Mux delay to memory data path**
- ♦ **Not as flexible as direct access testing**
- ♦ **Small routing and timing impact**

Disadvantages of adding BIST include:

- **Small area increase**  
The area increase caused by adding BIST is small and depends on what features you select for your BIST controller and the word and address size of the array. Typically, a controller can range from 400 gates for a simple implementation, to 1500 gates for an implementation that uses many options and several algorithms.
- **Adds Mux delay to memory data path**  
This multiplexor delay depends on the technology you are using. Typically, this is in the range of 200ps. This may be a problem if your designer doesn't have that much margin built into his or her design.
- **Not as flexible as direct access testing**  
There are many types of tests and algorithms to use for memory BIST.

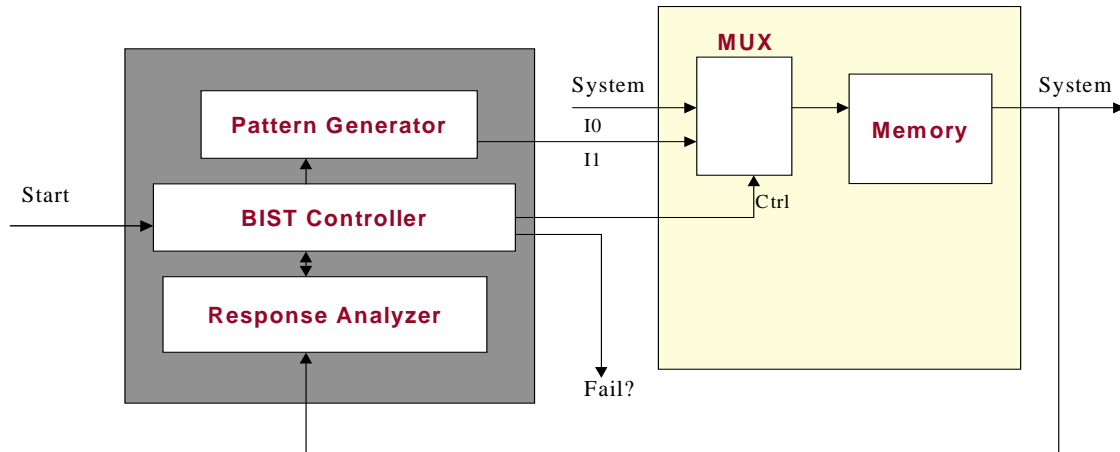
However, these tests are being “hard-wired” into the controller. After they are designed in, they cannot be changed. If you have direct access, you can change your test pattern supplied by the tester and rerun. It is also possible to design a re-configurable controller but this takes additional work and overhead.

- **Small routing and timing impact**

This is usually the reason most designers or managers initially might question the use of memory BIST. However, the routing and timing changes required by MBIST are almost always so small they are insignificant. With 5 or 6 layer metal processes and mux delays of 100 to 200ps, you can probably justify the use of memory BIST.

# Inserting BIST Circuitry

## Inserting BIST Circuitry



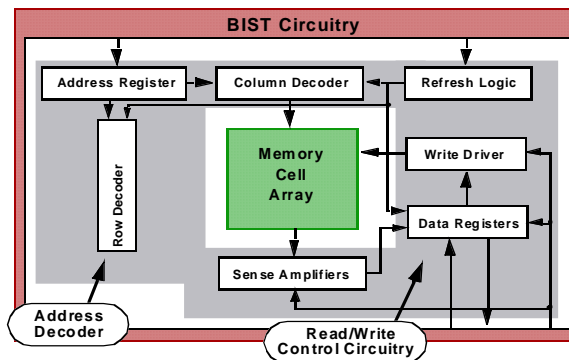
A built-in self-test (BIST) solution can alleviate many of these classic problems by embedding the pattern generator within the silicon. This approach can be automated using MBISTArchitect, which creates the RTL description in either Verilog or VHDL, that tests the memory without external stimulus or access.

# Memory Testing and Fault Types

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## Memory Testing and Fault Types

- ♦ Faults can be found in:
  - Address decoder logic
  - Read/write control logic
  - Memory cell array



# Memory Testing and Fault Types

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## Memory Testing and Fault Types (Continued)

- ♦ **Faults include:**
  - **stuck-at**
  - **transition**
  - **coupling**
  - **neighborhood pattern sensitive**

Memories fail in a number of different ways. The three main parts—address decoder logic, memory cell array, and read/write logic—can each have flaws that cause the device to fail. Memory testing, while similar to random logic testing, focuses on testing for these memory-specific failures.

The basic types of memory faults include stuck-at, transition, coupling, and neighborhood pattern sensitive. The next several slides discuss each of these fault types in more detail.

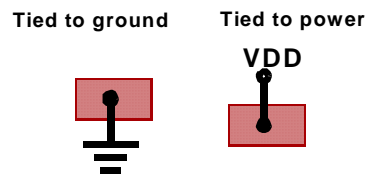


# Stuck-at Faults

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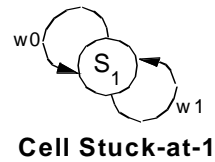
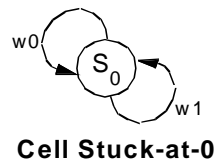
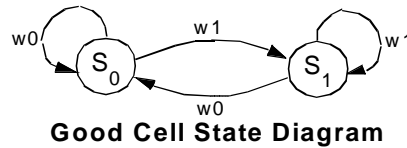
## Stuck-at Faults

- ◆ Applies to Control signals and memory cells
  - Behavior: Value stuck at either 0 or 1 indefinitely (signal/cells acts as though tied to power or ground)



# Stuck-at Faults

## Stuck-at Faults (Continued)



A memory fails if one of its control signals or memory cells remains stuck at a particular value. Stuck-at faults model this behavior, where a signal or cell appears to be indefinitely tied to power (stuck-at-1) or ground (stuck-at-0).

# Transition Faults

---

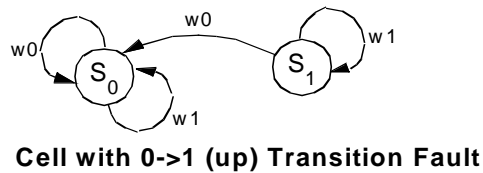
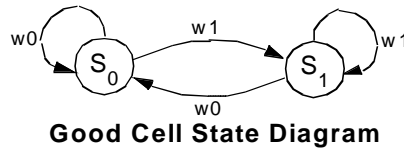
## Transition Faults

- ◆ Applies to: Signal or cell
- ◆ Behavior: Signal or cell cannot transition from 0 to 1 or 1 to 0



# Transition Faults (Continued)

## Transition Faults (Continued)



A memory fails if one of its control signals or memory cells cannot make the transition from either 0 to 1 or 1 to 0. The inability to change from 0 to 1 is called an up transition fault. The inability to change from a 1 to a 0 is called a down transition fault.

As the example shows, a cell may behave normally when a test writes and then reads a 1 value. And it may even transition properly from 1 to 0. However, when undergoing a 0->1 transition, the cell could remain at the 0 state—exhibiting stuck-at-0 behavior from this point on. However, a stuck-at-0 test might not detect this fault if the cell was at the 1 state originally.

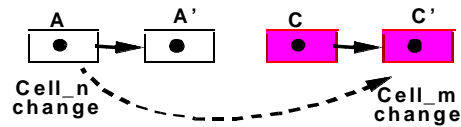
Thus, to ensure the cell can transition normally, a test must write a 1, write a 0, and then read the cell contents, as well as write a 0, write a 1, and then read the cell contents.

# Coupling Faults

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## Coupling Faults

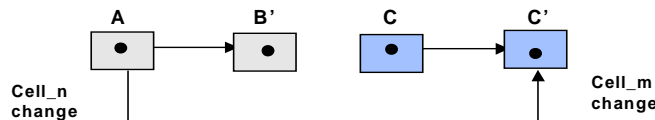
- ♦ Applies to: Memory cells
- ♦ Behavior: A write operation changing one cell's value influences another cell's value.
  - Several types:
    - Inversion (CFin) – Transition in one cell causes inversion of another cell's value



# Coupling Faults (Continued)

## Coupling Faults (Continued)

- ♦ **Idempotent (CFid) - Transition in one cell forces a particular value on another cell**



- ♦ **Bridging (BF) - Short, or bridge between two cells**
- ♦ **State (SCF) - A certain state on one cell forces a value onto another cell**

Memories fail when memory cells do not attain the proper state. This can happen in a number of different ways. In one case, a write operation in one cell can influence the value in another cell. Coupling faults model this behavior.

Coupling faults fall into several categories: inversion, idempotent, bridging, and state.

Inversion coupling faults, commonly referred to as CFins, occur when one cell's transition causes inversion of another cell's value. For example, a 0->1 transition in cell i causes the value in cell j to go from 0 to 1.

Idempotent coupling faults, commonly referred to as CFids, occur when one cell's transition forces a particular value onto another cell. For example, a 0->1 transition in cell i causes the value of cell j to be 0.

Bridge coupling faults, abbreviated as BFs, occur when a short, or bridge, exists between two or more cells or signals. Instead of transition operation, a logic value

triggers the faulty behavior. Bridging faults fall into either the AND bridging fault (ABF) or OR bridging fault (OBF) subcategories. ABFs exhibit AND gate behavior: that is, the bridge has a 1 value only when all the connected cells or signals have a 1 value. OBFs exhibit OR gate behavior: that is, the bridge has a 1 value when any of the connected cells or signals have a 1 value.

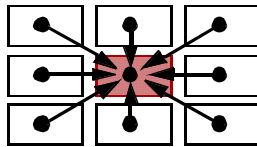
State coupling faults, abbreviated as SCFs, occur when a certain state in one cell causes another specific state in another cell. For example, a 0 value in cell i causes a 1 value in cell j.

# Neighborhood Pattern Sensitive Faults

---

## Neighborhood Pattern Sensitive Faults

- ◆ Applies to: Memory cells
- ◆ Behavior: A set of values or a transition of values in multiple cells influences the value of another cell.





# Neighborhood Pattern Sensitive Faults

---

## Neighborhood Pattern Sensitive Faults (Continued)

♦ **Three types:**

- **Active** – During a certain pattern in neighboring cells, one cell change causes another cell to change value.
- **Passive** – Certain pattern in neighboring cells causes a cell to remain fixed (appear stuck-at).
- **Static** – Certain pattern in neighboring cells forces a cell to a certain state.

Another way in which memory cells can fail involves a write operation on a group of surrounding cells affecting the values of one or more neighboring cells. Neighborhood pattern sensitive faults model this behavior. Neighborhood pattern sensitive faults break down into three categories: active, passive, and static.

An active fault occurs when, given a certain pattern of neighboring cells, one cell value change causes another cell value to change. A passive fault occurs when a certain pattern of neighboring cells cause one cell value to remain fixed.

A static fault occurs when a certain pattern of neighboring cells forces another cell to a certain state.

Because of the complexity and vast number of ways in which these faults can occur, testing for neighborhood pattern sensitive faults remains a very difficult task.

# Testing for Cell Array Faults

## Testing for Cell Array Faults

- **Stuck-at faults:**
  - Require writing 0's in all cells, reading all cells, writing 1's in all cells, and reading again.
- **Transition faults:**
  - Require writing (1->0) and immediately reading 0's at each address, and repeating the process for writing (0->1) and reading 1's.
- **Coupling faults:**
  - Require scanning (writing/reading) all memory cells in ascending order followed by scanning all memory cells in descending order.
- **Neighborhood pattern sensitive faults:**
  - Difficult to detect and require different procedures for different types of these faults.

To detect stuck-at faults, you must place the value opposite the stuck-at fault at the fault location. To detect all stuck-at-1 faults, you must place 0s at all fault locations. To detect all stuck-at-0 faults, you must place 1s at all fault locations.

In order to detect all transition faults in the memory array, a test must transition each cell from 0->1 and then immediately read it. The test must then repeat this process for the 1->0 transition.

Coupling faults involve cells affecting adjacent cells. Thus, to sensitize and detect coupling faults, you must perform a write operation on one cell (j) and later read cell (i). The write/read operation performed in ascending order assumes coupling of a memory cell to any number of cells with lower addresses. Likewise, the write/read operation performed in descending order assumes coupling of a memory cell to any of the cells with higher addresses.

Neighborhood pattern sensitive faults are complex and require a variety of different methods for detection. While currently available, test algorithms for

neighborhood pattern sensitive fault detection require much area overhead and produce very long test sets. Some test algorithms, in conjunction with manual circuit manipulation, can produce test sets for this fault type. However, currently no commercially-available tool alone does an adequate testing job for this memory fault type.

# Memory BIST Algorithms

## Memory BIST Algorithms

- ♦ Numerous memory BIST algorithms exist
  - The more popular memory BIST algorithms include:
    - March A and March B
    - March C, March C-, March C+, March3, and Column March
    - Unique Address
    - Checkerboard
    - ROM Tests
    - Port Interactive Test
    - User Defined Algorithm™

The test industry has generated many different algorithms for memory testing. The following list gives a brief description of some of the more popular ones:

- **March A and March B**

The March A and March B algorithms cover some linked faults, such as idempotent linked faults, transition faults linked with idempotent coupling faults, and inverting faults coupled with idempotent coupling faults.
- **March C+ (March2) Default Algorithm)**

The next few slides discuss the March C+ default algorithm.
- **Other Algorithms**

Other common algorithms include: March C, March C-, March3, Column March, Unique Address, Checkerboard, ROM Test, the Port Interaction Test, and the User Defined Algorithm. For more detailed information on these algorithms, see Chapter 3 in the *Built-In Self-Test Process Guide*.

# Comparing the Algorithms

## Comparing the Algorithms

### Testing at 10Mhz

\*not supported in MBISTArchitect

Algorithm	Fault Coverage	Test Time on 1M RAM
March C-	Detects address, stuck-at, transition, coupling, and unlinked coupling	1.0 seconds
March C+ (default)	Detects all March C- faults and some dynamic faults such as address decoder delay faults	1.3 seconds
MATS*	Detects address and stuck-at faults	0.42 seconds
Unique Address	Detects stuck-at and address faults	0.50 seconds
Checkerboard	Locates stuck-at and memory leakage (refresh) faults	0.52 seconds
Walking 0/1*	Locates stuck-at, address, transition, and coupling faults	2.5 days
GALPAT*	Locates address, stuck-at, transition, coupling, and write recovery faults	5.1 days

This slide provides a comparison of several algorithms that ran on a 1 Megabit RAM. Notice that some of the algorithms required a large amount of time for test completion. This is due to the nature of the algorithm—the number of operations (complexity) required for testing.

For example, the March C- algorithm has a complexity of  $10n$ , where  $n$  is the number of locations in the memory. That is, if you count the number of operations (see the slide depicting the algorithm operations for March C-), you can see that it requires 10 operations at each location to complete its test. In this comparison, the March C- algorithm took 1.0 seconds to complete testing of a 1 M RAM.

While more robust in its fault detection, the GALPAT algorithm, on the other hand, has an order  $n^2$  complexity. A walking target cell and revisiting of this target cell after each read greatly increases this algorithm's complexity. This results in the following equation that describes the complexity of the GALPAT algorithm:  $2(2^N + 2n^2)$ , where  $N$  is the number of address lines and  $n$  is the number of cells in the memory. Because of its complexity, in this comparison, the GALPAT

algorithm would take 5.1 days to complete testing of a 1 M RAM. The MATS algorithm is a modification of the Algorithm Test Sequence (ATS). MATS provides the shortest march test for unlinked stuck-at faults, it detected address and stuck-at faults in .42 seconds.

### Evaluating the Tradeoffs

Selecting one or more algorithms for your BIST design depends on the type of memory you are testing, your test goals, your overall test strategy, and the advice you may receive from in-house memory-test experts or ASIC vendors.

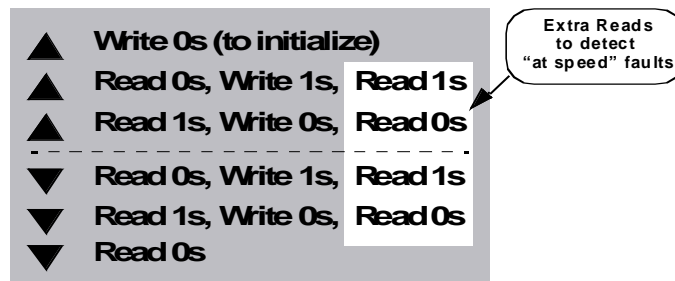
As you can see, as the size of the target memory grows, the complexity of the algorithm plays a very big role in the required test execution time. So, you need to consider the trade-off between robust fault coverage and test execution time when determining which algorithms to use.

# March C+ (March 2)

---

## March C+ (March 2)

- ◆ Default Algorithm for MBISTArchitect
- ◆ Adds extra read to each stage of march
- ◆ Extra read operation immediately after write operation lets you test at-speed
- ◆ Algorithm comprised of 14 operations (14n):



# March C+ (March 2)

---

## March C+ (March 2) Continued

- ♦ Detects the same faults as March C, PLUS some stuck-open faults, and some timing faults if you test at-speed (due to the read immediately after the write)
- ♦ MBISTArchitect refers to March C+ as “March 2”

MBISTArchitect uses this algorithm by default if you don't specify an algorithm and refers to this algorithm as “March 2”.

The March C+ algorithm modifies the original March C algorithm by adding an extra read operation after each stage of the march plus another at the end of the final stage. While increasing the algorithm from  $10n$  (of March C-) to  $14n$ , these extra reads allows additional fault detection, most notably stuck-open faults for all types of RAM.



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# Module 2

## Generating a Memory BIST

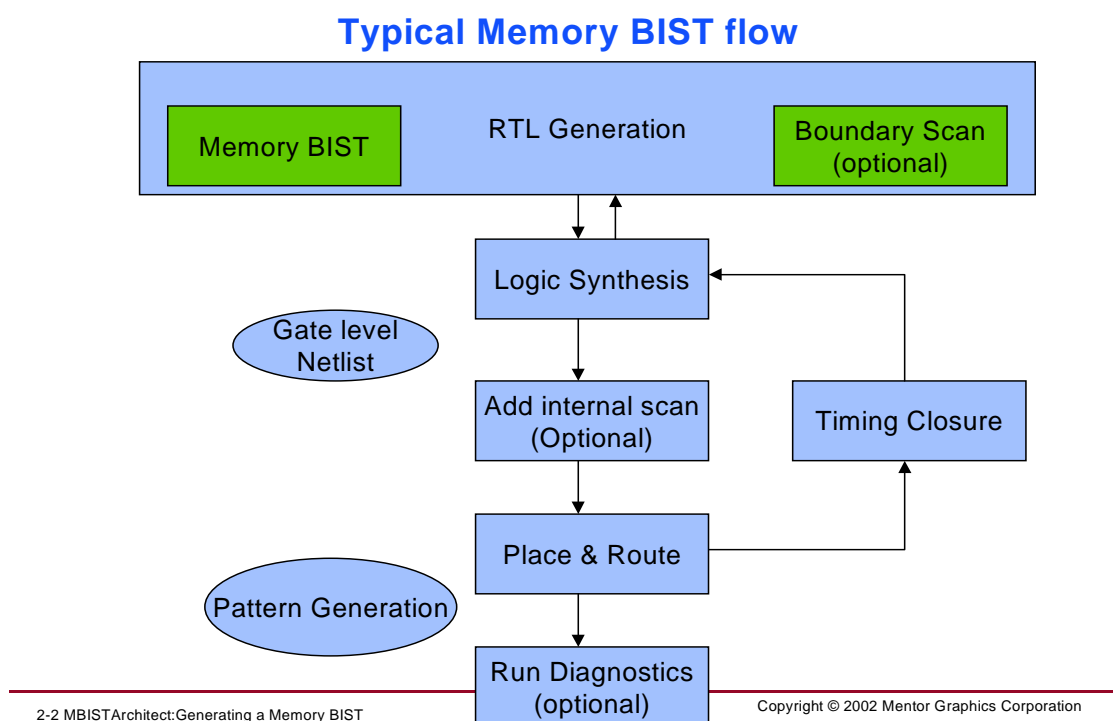
When you complete this module, you should have a basic understanding of memory testing inputs and outputs, how to launch MBISTArchitect, and how to access documentation for MBISTArchitect. The accompanying lab exercises give you hands-on experience generating and verifying a BIST collar for a simple memory device.

### Objectives

Upon completion of this module, you will be able to:

- Describe what a typical Memory BIST flow might look like
- List inputs and outputs to MBISTArchitect
- Show how to start MBISTArchitect in GUI mode
- Describe how to locate and use the documentation resources available through Mentor Graphics for MBISTArchitect

# Typical Memory BIST Flow



Here a typical Memory BIST flow. It contains all the basic steps that are used in an ASIC type flow.

- **Step 1 – Generate RTL or VHDL type code**

Memory BIST RTL code is inserted at this stage. It can be generated by hand. But why would you generate the code manually when you can use tools like MBISTArchitect to generate it for you? You can also add a boundary scan (IEEE 1149.1), sometimes called JTAG, at this time. Boundary scan is a very common method to control test modes such as memory BIST and scan testing. We will talk about this more in Lesson 5.

- **Step 2 – Logic Synthesis**

Tools like Synopsys' Design Compiler, Cadence' Ambit, or Mentor Graphics' Leonardo can handle this step, which turns RTL into a technology specific gate level netlist.

- **Step 3 – Adding Internal Scan**

Memory BIST is not dependent on having a scan design. You can have memory BIST in a scan design or in a non-scan design. In Lesson 5, we will talk further on how to interface scan and memory BIST.

- **Step 4 – Place and Route**

Many tools can be used to generate a physical layout database. This generates a GDSII database that is needed to continue the physical flow.

- **Step 5 – Timing Closure**

Static and Dynamic timing analysis tools can be used to verify timing. An SDF file can be generated and used in a Verilog-based simulation to check for additional timing issues.

- **Step 6 – Pattern Generation**

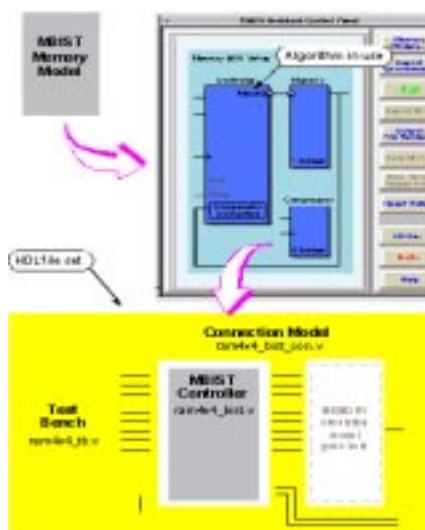
Patterns can be generated at any time but should be verified with back-annotated timing information. Memory BIST patterns can be generated using the testbench generated during the RTL generation.

- **Step 7 – Diagnostics and Debug**

Diagnostics and debug might be required if a part fails on the ATE tester. Memory BIST generated with debugging options included can make this stage possible. We'll also talk about this in Lesson 3.

# MBISTArchitect Inputs and Outputs

## MBISTArchitect Inputs and Outputs



2-3 MBISTArchitect:Generating a Memory BIST

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The only input to MBISTArchitect is one or more “abstract” memory models. These memory models reside in ASCII text files. The model describes the signals on the memory ports and the read/write protocol. These models only serve as a core around which MBISTArchitect builds an RTL BIST collar. The memory model itself does not become part of the final design output.

The MBISTArchitect Control Panel gives you a graphic means to setup and generate the memory BIST circuitry. As shown in the slide, the output is a set of three HDL files:

1. A memory BIST Controller (**ram4x4\_bist.v**). This file includes the finite state machine, the pattern generator, the comparator and memory blocks.
2. A Connection Model (**ram4x4\_bist\_con.v**). This is basically a set of ports and wires. It provides a means for connecting the memory BIST Controller to the memory simulation model and serves as the main interface to your design.

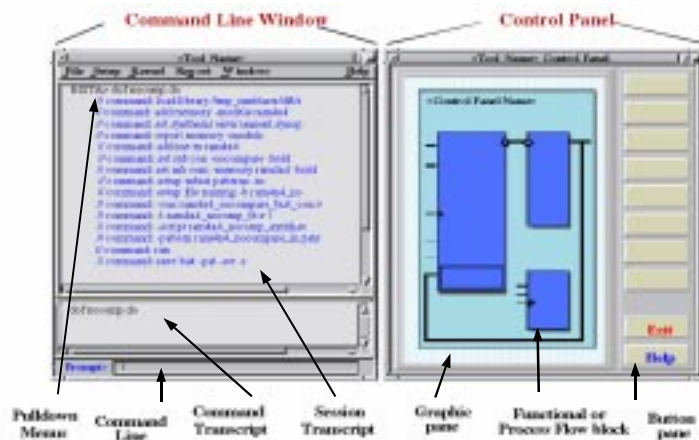
3. Test Bench (**ram4x4\_tb.v**) You can use this test bench to verify the proper working of the memory BIST generated circuit before you include the circuit in your design.

You or your ASIC vendor must supply the memory simulation model. In this case, assume that your simulation model file is named **ram4x4.v**.

# Graphical User Interface

## Graphical User Interface

- ♦ All DFT tools use a similar Graphical User Interface (GUI)
- ♦ When you invoke a tool, it opens:
  - The Command line window
  - The Control Panel windows



2-4 MBISTArchitect:Generating a Memory BIST

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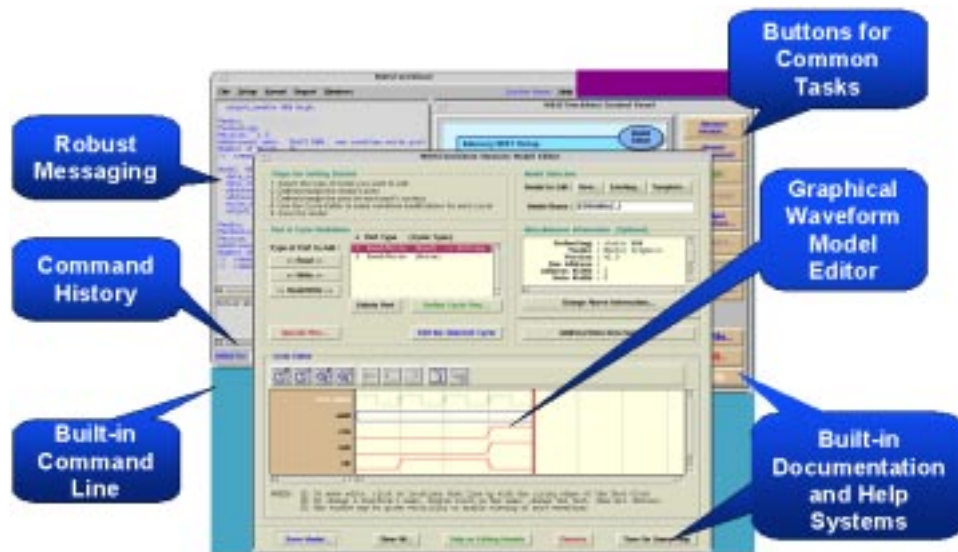
DFT products use two similar graphical user interfaces (GUI): one for BIST products and one for ATPG products. The BIST graphical user interface supports MBISTArchitect, LBISTArchitect, BIST Controller Synthesis, BIST-In-Place, and BSDArchitect.

The slide shows a representation of the GUI elements that are common to both user interfaces. Notice that the graphical user interfaces consist of two windows: the Command Line window and the Control Panel window.

# MBISTArchitect GUI Overview

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## MBISTArchitect GUI Overview



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2-5 MBISTArchitect:Generating a Memory BIST

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The MBISTArchitect GUI provides the following:

- Buttons used for common tasks such as loading memory models, setting the report environment, and obtaining help.
- Command history, command messaging and a command line for entering commands manually.
- A graphical wave form model editor.

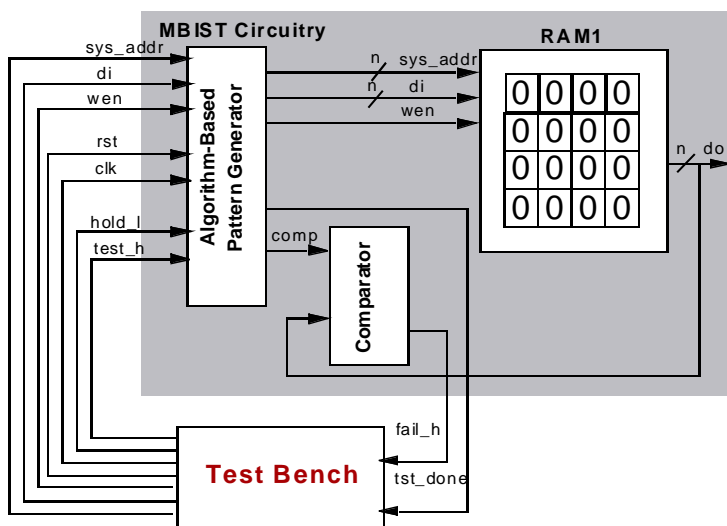
To launch MBISTArchitect in the GUI mode, type:

```
shell > mbistarchitect
```

You will be using the tool in the GUI mode in the first exercise.

# Role of the Test Bench

## Role of the Test Bench



2-6 MBIST Architect: Generating a Memory BIST

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The testbench instantiates and provides stimulus to the connected memory BIST model. A high value on the `tst_done` signal indicates the BIST test has successfully completed. The `fail_h` signal value goes high the first time the BIST controller encounters a miscompare.



# Memory BIST Documentation

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## Memory BIST Documentation

- ♦ You can obtain online help and view Memory BIST documentation using a PDF viewer
- ♦ Refer to these guides:
  - MBISTArchitect Reference Manual
  - Built-In-Self-Test Process Guide

Use the following documentation for information on BIST concepts and how to use the MBISTArchitect and Memory BIST-In-Place tools:

- *MBISTArchitect Reference Manual*—This guide provides reference information for the Mentor Graphics' MBISTArchitect and Memory BIST-In-Place tools. Information contained in this manual includes tool capabilities, a reference for all tool commands, modeling information, and sample tool outputs.
- *Built-In-Self-Test Process Guide*—This guide contains process-oriented information on MBISTArchitect and Memory BIST-In-Place, as well as other Mentor Graphics Design-for-Test (DFT) tools. Use this manual to become familiar with Memory BIST concepts and tool functionality.

Click the **Help** and **Turn on Query Help** buttons to obtain online information and links to this documentation. You will need a PDF viewer to view documentation.

## Module 2 Lab Exercises

- **Setting Up the Training Data**
- **Creating a Basic Memory BIST Collar**  
(20 minutes)
- **Verifying the BIST Circuitry**  
(20 minutes)

# Module 2: Lab Exercises

In this lab, you will use the MBISTArchitect tool to create a memory BIST collar and verify the BIST circuitry. The goal of each exercise is as follows:

**Exercise 1: Creating a Basic Memory BIST Collar**—You will create a basic BIST collar for a simple 4x4 RAM model, then save the BIST circuit as a Verilog file set.

**Exercise 2: Verifying the BIST Circuitry**—You will use ModelSim to verify the memory BIST circuit using an MBISTArchitect-generated test bench.



These exercises should take approximately 40 minutes.

## Getting Started

This section lists the software versions and versions of the training data you will need. It also provides instructions on how to install the training data so that you can run the labs.

## Software Versions

This version of the training data and materials (V8.2002\_1) should be used with the V8.2002\_1 release of all BIST products to ensure that the lab exercises run successfully:

MBISTArchitect	v8.2002_1
Memory BIST-In-Place	v8.2002_1
Acrobat Reader	v4.0 (install from MGC CD)
ModelSim	EE/Plus 5.5f or newer, including both VHDL and Verilog libraries

## Training Files

Training files have been provided for this course. Use the files in this directory to access the training data:

- ***mbist896nwp***—This is the data you need to use to run the exercises.

## Installing the Training Data Files

The data for the lab exercises consists of circuit(s), library parts, and userware (called dofiles). Because you will modify some of the data during the lab exercises, you need to have your own local copy. Use the following procedure to make a local copy of the lab exercise data:

1. Make sure that your `$MCG_HOME` shell variable is set to a MGC\_HOME tree that contains the 8.2002\_1 version of the MBISTArchitect and Memory BIST-In-Place software.
2. The design data for the lab exercises is named ***mbist896nwp***. It is located in the directory named:

`$MGC_HOME/shared/training/mbist896nwp`

Before you can perform the lab exercises, this training data directory must be installed in your Mentor Graphics tree. To check whether the training data is installed, list the contents of the `$MGC_HOME/shared/training` directory by issuing the following operating system command:

```
/bin/ls $MGC_HOME/shared/training
```

If ***mbist896nwp*** does not appear in the displayed list, then either you or your system administrator must install the training package.

3. Before attempting to copy the training directory, ensure you have at least 20 MB of disk space. The uncompressed tar file is approximately 10 MB and the design data is about 10 MB.

4. Copy the training package data from the `MGC_Home` tree to the training directory on your workstation. Specify the pathname where you want your local copy. The pathname that you specify in this step is referred to as *your\_path*.

Copy from: `$MGC_HOME/shared/training/mbist896nwp`

Copy to: `your_path/training/mbistnwp`

5. Include `$MGC_HOME/bin` in your `$PATH` variable.
6. Ensure that `acroread` (Acrobat Reader) is also included in your `$PATH` variable.
7. Define an environment variable named `MBISTNWP` that points to the full pathname where you copied `mbistnwp`. For example, in a C shell enter:

```
$ setenv MBISTNWP <dir_path>/mbistnwp
```

8. Each lab directory, such as *lab1* or *lab2*, contains a *results* subdirectory. You may need to change the permissions of these directories to allow write access.

## Exercise 1: Creating a Basic Memory BIST Collar

This exercise should take approximately 20 minutes to complete.

1. Change to the following working directory:

```
shell> cd $MBISTNWP/mbist1/ram4x4/design
```

2. List the design files you will be using in this exercise.

```
shell> ls -l ram*
```

The *ram4x4.atpg* file is a library file contains a single 4x4 RAM model.  
The *ram4x4.v* file is the corresponding Verilog simulation model.

3. Change to the *ram4x4/results* directory.

```
shell> cd ../results
```

You will work and save your results in this directory.

4. Invoke MBISTArchitect.

```
shell> mbistarchitect
```

This step invokes the MBISTArchitect graphical user interface (GUI). You will be using various aspects of the GUI to create your memory BIST model.

5. Click on the **Memory** block in the Control Panel graphic pane.

This starts the process to load the **ram4x4** model.

- a. Click on the **Browse** button, then navigate to the *\$MBISTNWP/mbist1/ram4x4/design* directory.

- b. Double click on the *ram4x4.atpg* file, then click **Load**.

The **ram4x4** now appears in the Available Models list.

- c. Select the **ram4x4** model, then click **Add**.

This adds the model to the memory models for BIST insertion.

d. Click **OK**.

6. From the command line, report on the models in the library.

MBISTA> **report library models**

The tool should respond:

```
// Error: Command 'report library models' is unknown
```

This is not the proper command name.

7. Use the help command to display the available application commands.

MBISTA> **help**

The tool displays a list of commands similar to those shown on the next page.

ADD Verilog Include	ADD VHdl Library
ADD VHdl Use	ALias
DELeTe ALgorithms	DELeTe DATA Backgrounds
DELeTe DIAGNOSTIC Monitor	DELeTe MBist Algorithms
DELeTe MEMory Models	DELeTe VERilog Include
DELeTe VHdl Library	DELeTe VHdl Use
DOFile	EXIt
HELp	HISTory
LOAD Algorithms	LOAD Library
REPort ALgorithm Steps	REPort ALgorithms
REPort Bist	REPort DATA Backgrounds
REPort DIAGNOSTIC Monitor	REPort ENVironment
REPort MBist Algorithms	REPort MEMory Models
REPort Verilog Include	REPort VERSION Data
REPort VHdl Library	REPort VHdl Use
RESet STate	RUN
SAVE Bist	SAVE HISTory
SET BISTinplace	SET COMmand Editing
SET COMparator Test	SET CONTroller Debug
SET CONTroller Hold	SET DOfile Abort
SET File Compression	SET GZip Options
SET MESSAGE Handling	SET SCan Logic
SET SYNthesis Environment	SET VHdl Configurations
SETup CLock Period	SETup COMparator Failflag
SETup CONTroller Clock	SETup CONTroller Naming
SETup CONTroller Pipeline	SETup CONTroller Reset
SETup DIAGNOSTIC Clock	SETup File Naming
SETup MBist Algorithms	SETup MBist COMPressor
SETup MBist Patterns	SETup MEMory Access
SETup MEMory Clock	SETup MEMory Test
SETup MUX Location	SETup OBServation Scheme
SETup RETention Cycles	SYSTEM

The command you want to use is **REPort MEMory Models**.

1. Get the command usage for Report Memory Models.

**MBISTA> help report memory models**

The tool should display the following usage:

**Usage: REPort MEMory Models [-Library | -Model <model\_name>]**

You want to see the models in the library, so use the -Library switch.



2. Report on the library models.

```
MBISTA> report memory models -library
```

Or, if you want to use minimal typing:

```
MBISTA> rep me m -l
```

The tool displays information on the single model, ram4x4, as follows:

```
Available Memory Models:
Name           Vendor           Technology
-----
ram4x4         sample          sample1
```

This is the memory model around which you want to generate BIST circuitry.

3. From the command line interface, report more information on the current memory model by issuing the following command:

```
MBISTA> report memory models -model ram4x4
```

The tool should display the following:

```
Model  ram4x4
  data_out DO3, DO2, DO1, DO0;
  data_in DI3, DI2, DI1, DI0;
  address A1, A0;
  write_enable WEN low;

Vendor:  sample
Technology:  sample1
Version:  1.0
Additional info:  4x4 RAM, ports = 1rw
Number of Words:  4
```

This RAM has one read/write port and contains four words. It has four data input bits, two address bits, and four data output bits.

4. Click **Run**.

Default BIST circuitry is added to this model.

## 5. Click **Report BIST**

The tool should display the following information:

### Generated BIST structures:

Name	Type
-----	
<code>ram4x4_bist</code>	Memory Bist

MBISTArchitect generates *ram4x4\_bist*, which is the memory BIST controller for the *ram4x4* model.

## 6. Click **Save BIST...**

Verify that the *ram4x4\_bist.v* model will be saved to the *\$MBISTNWP/mbist2/ram4x4/results* directory, then click **OK**. The tool responds by telling you the models it is saving, as such:

```
Saving MBIST Data:
  Saved ram4x4_bist.v
  Saved ram4x4_bist_con.v
  Saved ram4x4_tb.v
```

## 7. Click **View Saved Design Files**

This action brings up the File Viewer window which allows you to view the contents of each of the output files. MBISTArchitect generated the following three Verilog files for the *ram4x4* model:

- *ram4x4\_bist.v* - an HDL model that contains the *ram4x4* BIST controller.

Examine the top-level signals coming out of and going into the BIST controller. Scroll down through the file and notice that the default BIST circuitry includes a comparator.

- *ram4x4\_bist\_con.v* - this HDL model simply instantiates both *ram4x4\_bist* and *ram4x4* and connects them up by default. The ports of this model represent the external interface of the memory BIST collar.

Examine the top-level signals coming out of this connection model. Notice that the system signals (**sys\_**) replace the previous memory input ports and the memory signals (**Con-Test**) are newly created wires that connect the controller to the memory inputs. Notice also that the memory data outputs have a “**\_O**” extension.

There are also three new BIST inputs; **clk** and **rst\_1** drive the new BIST state machine. **test\_h** is a level-sensitive signal that tells the BIST controller to run the test.

**tst\_done** tells your design that the test is finished and has run successfully;

**fail\_h** tells your design that the memory test has failed.

- *ram4x4\_tb.v* - the testbench for the *ram4x4\_bist\_con.v* model.

Examine the top-level signals coming out of this model. Examine the testing that the testbench performs on the *ram4x4\_bist\_con.v* model.

8. Exit the tool.

MBISTA> **exit**

## Exercise 2: Verifying the BIST Circuitry

This exercise should take approximately 20 minutes to complete.

In this exercise, you will use the MBISTArchitect-generated testbench to verify the memory BIST circuitry that you created in the last exercise.

1. Ensure that you are still working in the *\$MBISTNWP/mbist1/ram4x4/results* directory.

2. Set up a work directory.

```
shell> $MGC_HOME/bin/vlib work
```

3. Compile the memory simulation model, all BIST models, and the testbench.

```
shell> $MGC_HOME/bin/vlog ../design/ram4x4.v \
      ram4x4_bist.v ram4x4_bist_con.v ram4x4_tb.v
```

4. Simulate the test driver.

- a. Invoke the ModelSim simulator and load the testbench model.

```
shell> $MGC_HOME/bin/vsim ram4x4_tb
```

- b. Set up the lists by running the following dofile:

```
VSIM 1> do ../design/vsim_setup.do
```

This dofile sets the parameters for the simulation to stop—either **tst\_done** or **fail\_h** going high. It also sets up a List window so you can examine the pertinent signals. If necessary, expand the list window that appears so you can see all the signals. You can also use a wave window, if you choose.

- c. Run the simulation until it is finished.

```
VSIM 2> run -all
```

- d. Run a little more to capture the complete pattern for the **tst\_done** signal.

VSIM 3> **run 50**

- e. Write the displayed list to a file.

VSIM 4> **write list trace.log.m2**

- f. Quit the simulation.

VSIM 5> **quit**

5. Examine the saved list file. Use whatever editor you prefer to view the **trace.log.m2** file you saved.

As you scroll through this file, notice the following things:

- The signals that comprise the columns in this file include (from left to right, fail\_h, tst\_done, the address, the write enable, the data input values, and the data output values).
- The first 650ns of the testbench tests some system signals.
- The March2 algorithm begins at time 1450ns. Remember that the write enable is active low, and the address changes only when the write enable is *not* active (that is, only when the write enable is high). So at time 1450ns, the address is set to 0, the write enable is inactive, and the data on inputs is set to 0. At time 1550ns, the write enable goes low, capturing the input data and writing it to address space 0. At time 1650ns, the write enable again goes inactive, so the address can change to space 1. Thus, the time from 1450ns to 1650ns initializes address space 0 to all 0s, and prepares to initialize the next address space to 0.
- From 1650ns to 2150ns, the March2 algorithm continues to initialize address spaces 1, 2, and 3. This completes the first step in the March2 algorithm: Write 0s to initialize.
- At time 2250ns, with the address set back to space 0, the algorithm reads 0, writes 1, and reads 1. The address space increases to 1, and the

algorithm then reads 0, writes 1, and reads 1. This process repeats for addresses 2 and 3.

- At time 4650ns, with the address set back to space 0, the algorithm reads 1, writes 0, and reads 0. This repeats for addresses 1, 2, and 3.
- At time 7050ns, the algorithm begins the test in reverse address order, reading 0s, writing 1s, and reading 1s.
- At time 9450ns, the algorithm again performs the test in reverse address order, this time reading 1s, writing 0s, and reading 0s.
- At time 12750ns, the **tst\_done** flag goes high indicating the BIST testing is complete.
- **fail\_h** remains low throughout the entire simulation.

---

# Module 3

## Common BIST Variations

When you complete this module, you should have a basic understanding of how to configure memory BIST circuitry, use one BIST controller for multiple memories, add diagnostics, add pipeline registers, use compressors and comparators, use clock constraints, and run MBISTArchitect at full-speed.

### Objectives

Upon completion of this module, you will be able to:

- Insert BIST for multiple memories
- Add BIST with a compressor
- Add BIST for bidirectional memories
- Add BIST for ROMs
- Perform a full-speed BIST test

# Configuring Memory BIST Circuitry

---

## Configuring Memory BIST Circuitry

- ♦ **Add to and/or change the BIST algorithms**
  - Use one BIST controller for multiple memories
  - Use a compressor instead of a comparator
    - Add more system-level BIST control signals
- ♦ **Use one BIST controller for multiple memories**
  - Decreases BIST hardware
    - Memories must be compatible, event sequences must be the same



# Configuring Memory BIST Circuitry

---

## Configuring Memory BIST Circuitry (Continued)

- ♦ **Use a compressor instead of a comparator**
  - **Allows ROM testing**
  - **Reduces diagnostic capability**
  - **Decreases interconnections**

The MBISTArchitect tool provides a common default BIST architecture, however, this default circuitry may not meet all your testing requirements. Thus, MBISTArchitect lets you customize the circuitry it generates in a number of ways.

You can add to or change the default algorithms. For example, if you are adding BIST circuitry to a multiple-port memory model, you may not want to execute the March C+ test on every write port. You may instead want to use the Unique Address algorithm to test just the address and control circuitry for all but the first port.

Another common variation includes using a single BIST controller for multiple memory models. You can add a BIST collar around an individual model or you can create a single BIST controller that controls and tests a number of different compatible memory models.

One common variation includes using a compressor for signature analysis instead of a built-in comparator for direct memory output comparison. You also have less capability to diagnose what failure occurred.

You can add a system-level hold signal that can stop the testing process. You can also define multiple input busses connecting to the memory model to provide further system control.

# Support for Multi-port Memories

---

## Support for Multi-port Memories

- ♦ **MBISTArchitect provides the following features for multi-port memories**
  - **Applies different algorithms to each port**
    - Reduces test application time
- ♦ **Generates a port interaction test**
  - **Produces higher quality tests**
- ♦ **Handles restrictions on simultaneous port access**
  - **Honors read and write constraints for multiple ports**

The MBISTArchitect tool supports testing of multi-port memories. Using this functionality, you can apply different algorithms to each port to reduce test application time. The tool honors the read and write constraints for multiple ports which it uses to handle restrictions on simultaneous read-port access.

# Generate a Comparator Functional Test

## Generate a Comparator Functional Test

- ♦ From controller's finite state machine
- ♦ Add two states
  - comp\_test\_write
  - comp\_test\_read\_fail and comp\_test\_read\_pass
    - Generate Using:
 

```
MBIST> setup observation scheme -compare
              set comparator test -on
```
- ♦ Use with other options
  - Example: Repeat comparator test for each memory prior to any other tests to test the fail flag of each memory independently
 

```
MBIST> setup memory test - sequential
              set comparator test -on
              setup comparator failflag -separate
```

The MBISTArchitect tool provides the ability to test the comparator before running the BIST. This is achieved by adding three states to the controller's finite state machine that inject faulty data into the memory at the beginning of the test. The two states are comp\_test\_write and comp\_test\_read\_fail.

The comparator test first uses the comp\_test\_write state to write known data (background 1) to address zero of all the memories. Then, comp\_test\_read performs a read/compare expecting a mismatch which should raise the fail\_h flag. Next, comp\_test\_read performs a second read/compare expecting a match, thereby resetting the fail\_h flag. When you enable the comparator test, it always precedes all other tests.

To generate the comparator test, use the Setup Observation Scheme command with the -Compare switch. To test the comparator, use the Set Comparator Test command with the -on switch as follows:

**setup observation scheme -compare**  
**set comparator test -on**

The default of the comparator upon invocation of the MBISTArchitect is “-Off”.

Additionally, you can use other MBISTArchitect command options in conjunction with these commands. For example, you can enable the comparator test in combination with the Setup Memory Test command’s sequential memory test (-Sequential) and the comparator fail flag option as shown here:

**setup memory test - sequential**  
**setup comparator failflag -separate**

In this case, the controller repeats the comparator test for each memory prior to the application of any other tests. Thus, testing the fail flag of each memory independently.

# Inserting BIST for Multiple Memories

---

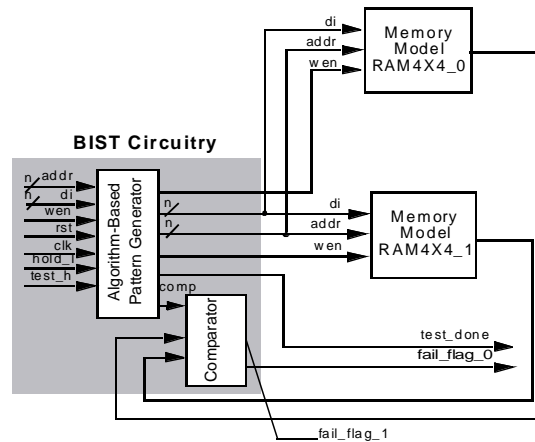
## Inserting BIST for Multiple Memories

- ♦ MBISTArchitect can generate BIST circuitry that tests multiple memory models
- ♦ You can load multiple memory models with the Add Memory Model command
- ♦ MBISTArchitect runs test in parallel

## Inserting BIST for Multiple Memories (Continued)

### Inserting BIST for Multiple Memories (Continued)

#### Example



You can create a single BIST controller that runs BIST on multiple memory models. This can only occur with compatible memory models—those that share the same vendor and technology and have compatible read and write cycle definitions.

You can specify multiple memories using one or more Add Memory Models commands. MBISTArchitect generates BIST circuitry that runs the testing on all memories in parallel.

In this case, the default names become *<first\_model\_added>\_multi.v*, *<first\_model\_added>\_multi\_con.v*, and *<first\_model\_added>\_tb.v*.

# MBISTArchitect Controller Options

---

## BIST Controller Options

- ♦ **The BIST controller performs two primary functions while testing memories under test:**
  - It provides the test stimulus
  - It checks the response
- ♦ **MBISTArchitect contains options that let you determine how to test the memory and how fast the MBIST controller performs memory testing**

MBISTArchitect contains options that let the designer or tester determine how they want to test the memory and how fast the MBIST controller performs memory testing. Using MBISTArchitect options, the designer can match the memory BIST controller speed and hardware required to their own unique needs.

Designers and testers can use MBISTArchitect to test embedded memories at varying speeds. From using the system default with built-in delay cycles, all the way up to using the full-speed option to exercise and test the memory at system cycle speeds, as well as performing timing and stress tests on embedded memories.

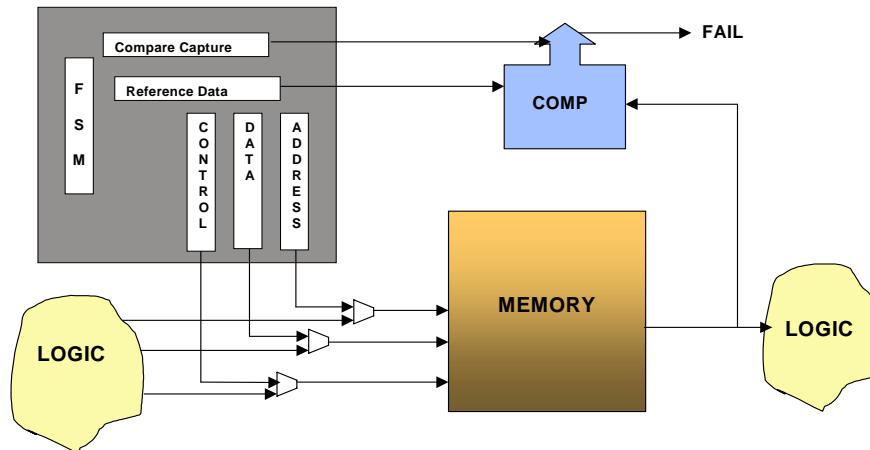
See [“How the BIST Controller Works” on page 3-11](#) for information on how the BIST controller typically works. See [“Full-Speed Overview” on page 3-14](#) for an overview of MBISTArchitect full-speed implementation.



## How the BIST Controller Works

---

### Typical Memory BIST Controller



3-9 MBISTArchitect:Common BIST Variations

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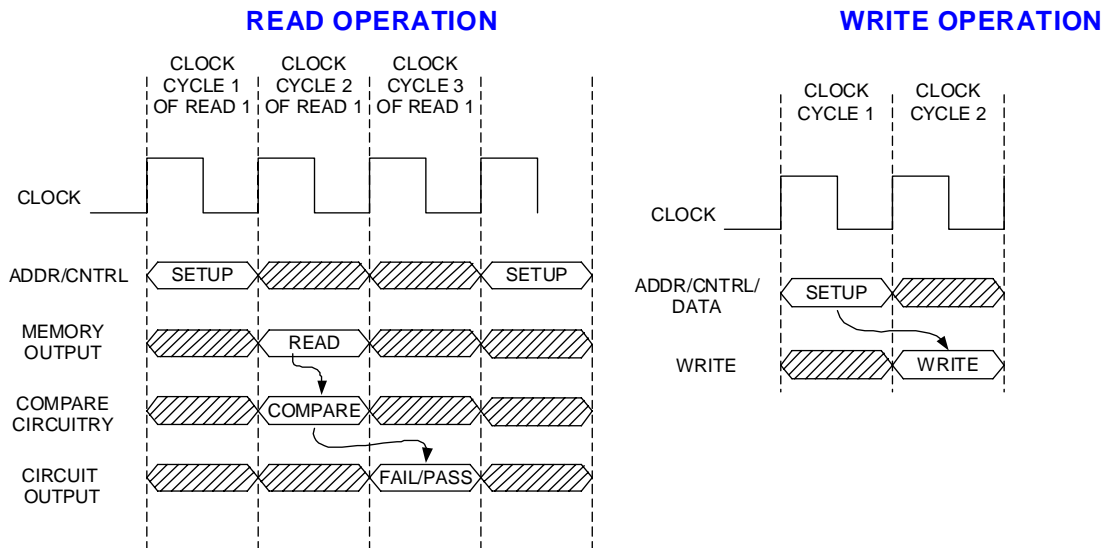
In a typical design with memory BIST, the BIST controller performs two primary functions to the memories under test: 1) it provides the test stimulus, 2) it checks the response. The slide shows that there is one memory tested by one BIST controller. In reality, the BIST controller is much smaller than the memories.

The BIST controller itself is a finite-state machine. The clock controlling its state transitions can be from either an internal clock generator or an external source. To avoid clock synchronization problems during the BIST operation, normally the same clock source controls both the BIST controller and the memories it tests. In this example, we assume all memories are synchronous memories.

See [“Read/Write Operations on Synchronous Memories” on page 3-12](#) for information on how MBISTArchitect performs read/write operations on synchronous memory. See [“Pipelining Read/Write Operations” on page 3-17](#) for information on how MBISTArchitect performs read/write operations when using the full-speed option.

# Read/Write Operations on Synchronous Memories

## Typical Read/Write Operations



3-10 MBISTArchitect:Common BIST Variations

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To properly perform read or write operations for synchronous memories, the BIST controller must first generate read/write setup signals before the memory clock is active. For simplicity, the examples presented in this section assume:

- all read/write setup signals are synchronous signals
- all memories and the BIST controller are activated at rising edge

Since the BIST controller and its memories use the same clock, a typical read/write operation requires two clock cycles. During the first clock cycle, the BIST controller generates all the necessary read/write setup signals for the memories under test. During the second clock cycle, a read/write operation occurs at the edge of memory clock. This is called data latency in single clock memory BIST operation.

## Common BIST Variations

---

In addition, memory BIST controllers typically use comparators to verify the data read out from the memories. Since memory outputs are not ready until the edge of the second clock, the result of the comparator will be captured at the third clock cycle. Therefore, a BIST controller requires:

- three clock cycles to perform a complete read operation
- two clock cycles to finish a write operation

Typically, a memory BIST controller requires six cycles to do two consecutive read operations and four cycles to do two consecutive write operations. Likewise, it requires five cycles to do one read operation followed by one write operation.

# Full-Speed Overview

---

## Full-Speed Approach

- BIST controller running at system speed
- Memory exercised at system speed
- Timing stress testing

Because memories are getting larger and denser, design and test engineers need to ensure higher memory test quality to ensure overall chip quality. Besides static functional tests, timing and stress tests are necessary to detect system operation problems. *At-speed* BIST operation generally means BIST operation is capable of exercising the memories at system clock frequency. However, at-speed operation is not sufficient to detect all timing faults. Even if a BIST controller design is operated in system clock frequency, its data latency prevents testing whether the memory can change the address and read out different data from different addresses at every cycle. Without this limitation, the BIST operation may not ensure adequate memory quality.

MBISTArchitect has a feature called Full-Speed™ BIST operation. Full-speed is used to enhance a single clock memory BIST controller so that it can launch a read or write operation on each active clock edge, thus enabling timing and stress testing as part of the BIST operation. Besides improved test quality, full-speed BIST operation significantly reduces test time. For example, typical consecutive

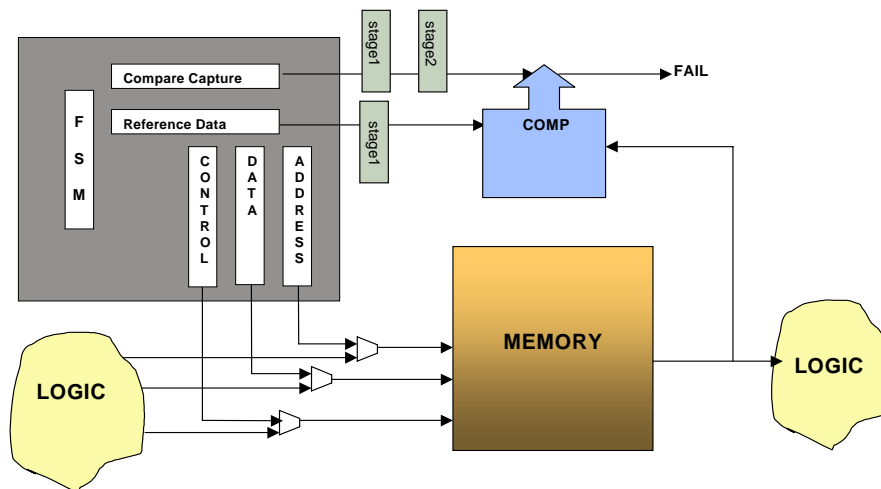
read/write operations require 5 clock cycles which can be done in 2 clock cycles with full-speed BIST operation.

The MBISTArchitect full-speed functionality provides maximum memory BIST controller speed and test performance. Running full-speed at system clock speed tests the memory at the full-speed the system will run.

Full-speed testing reduces testing times. Full-speed can locate defects that will not be detected at slower speeds, thus providing increased fault detection. Full-speed provides the additional benefit of testing whether a memory can change an address and read data from different addresses at every cycle. It enables timing and stress testing as part of the BIST operation because it can launch a read/write operation on each active clock edge.

# Full-speed design with pipeline circuitry

## Pipelined BIST Controller



3-12 MBISTArchitect:Common BIST Variations

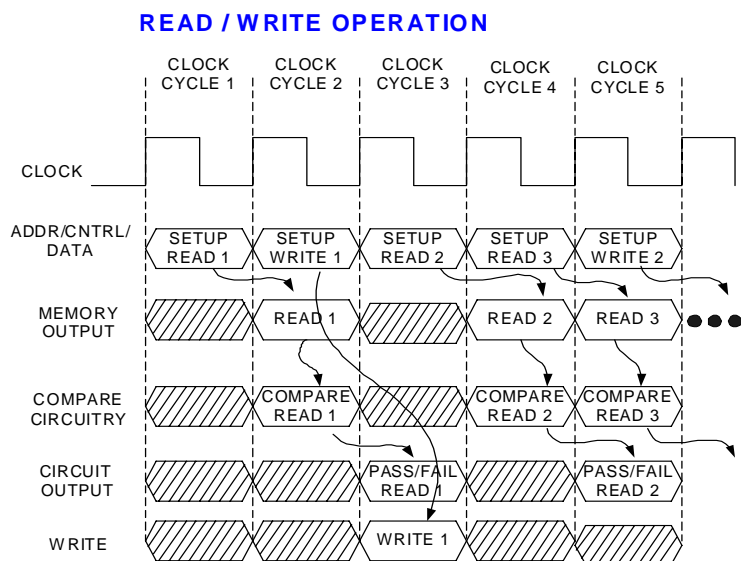
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The slide shows a Full-Speed pipelined BIST Controller.

Since there is data latency in memory BIST controllers, the BIST controller must be pipelined to enable full-speed read/write operation. The pipeline is used to temporally separate the needed action at each cycle of read/write operations. With pipelining, you can model the memory as only taking one clock cycle and then use the pipelining to tear the comparison out of the first cycle and the capture of the comparison result, which happens at the end of the pipeline.

# Pipelining Read/Write Operations

## Full-Speed Pipelined Read/Write Operations



3-13 MBIST Architect: Common BIST Variations

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The slide shows the pipelined design for full-speed memory BIST operation.

A 3-stage pipeline can be used to compress the three cycle read operation into single cycle read. In this case, the first stage does the read setup—which may include read address change, read enable activation, and output enable activation. The second stage activates the read clock and provides the reference data for read data output comparison. The third stage captures the comparison result. Inside the BIST controllers, all signals needed for read operation are generated at the rising edge of the same clock. The following pipelines are also needed:

- A pipeline register to create one-cycle delay at the memory clock signal.
- A pipeline register to create one-cycle delay at the reference data to the comparator.

- A full-speed BIST controller needs a pipeline register to create two-cycle delay at the capture signal that activates the capturing the results of the comparator.

In addition, a 2-state pipeline can compress the two cycle write operation shown earlier, into a single cycle write. The first stage does the write setup which may include write address change, write data change, and write enable activation. The second stage activates the write clock. Similarly, inside the BIST controllers, all signals needed for write operation are generated at the rising edge of the same clock. Here, only the memory clock needs to be delayed one cycle to achieve full-speed operation. As explained earlier, a memory clock is repeated every cycle, the pipeline register to create one-cycle memory clock delay is not needed.

As part of this module, you can perform a Full-Speed exercise to see the results of full-speed testing. See [“Running BIST at Full-Speed” on page 3-61](#).



# Performing Sequential Memory Tests

---

## Performing Sequential Memory Tests

- Apply all test algorithms to all ports of a memory before proceeding to the next memory
  - MBIST> setup memory test -sequential [Interleaved | Contiguous]
- Multiple Memories Sequential Memory Test
- Generate individual fail flags
  - MBIST> setup comparator failflag -separate
- Identifies which memory has failed

MBISTArchitect creates a controller that by default tests multiple memories concurrently. You can specify that the controller test each of these memories sequentially by using the following command option:

### **setup memory test -sequential**

The -Sequential contiguous switch causes the controller to apply all the test algorithms to all the ports of a memory before proceeding to the next memory (this is the system default). Using the -Sequential interleaved switch instructs the MBISTArchitect tool to interleave algorithm steps between memories.

Since the controller tests the memories independently of one another during sequential memory testing, the memory's read/write cycles need no longer be compatible. However, the current MBISTArchitect implementation of sequential memory test does not have this capability.

Additionally, you can generate individual fail flags for multiple memories by using the Setup Memory Test and Setup Comparator Failflag commands as follows:

**setup memory test -sequential**  
**setup comparator failflag -separate**

This specifies separate fail flags for multiple memory tests. This is especially useful in identifying which memory has failed when you specify the sequential memory test option (-Sequential). The default is Common; output a single fail bit regardless of the number of memories.

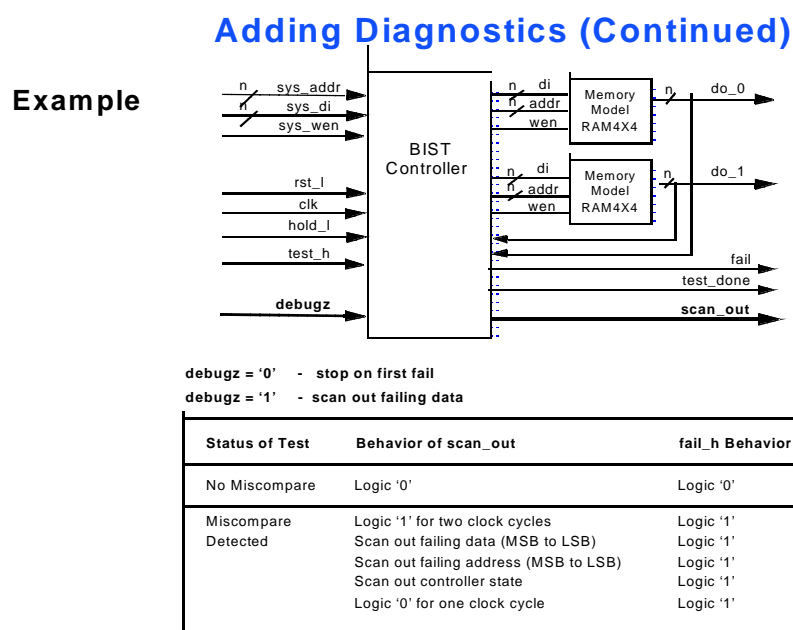
# Adding Diagnostics

---

## Adding Diagnostics

- ♦ Extracts failing data for fault diagnosis process
- ♦ Data scanned out through a serial pin
- ♦ Diagnostics contained within the BIST logic
- ♦ Two modes of operation

# Adding Diagnostics (Continued)



3-16 MBISTArchitect:Common BIST Variations

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MBISTArchitect can give the BIST controller the ability to download the failing data on every occurrence of a mismatch. And, the failing data can be scanned out with a minimal impact on silicon area and routing overhead. You can switch on a diagnostic clock and a diagnostic clock pin named **diag\_clock** is added to the controller pin list. The **diag\_clock** pin is toggled at half BIST clock during the test bench. The BIST controller operates in one of two modes controlled by **debugz**. The modes and operation of the **fail\_h** and **scan\_out** ports is as follows:

**Normal Mode** (**debugz** = '0') When **debugz** is set to '0', the BIST controller performs the default test. In this mode, the **scan\_out** port is set to '0', as no fail data is downloaded. The **fail\_h** port is asserted on the first failure and remains high for the remainder of the test.

**Debug Mode** (debugz = '1') When **debugz** is set to '1', the diagnostic mode is enabled. In this mode, a miscompare will suspend the operation of the BIST controller, and the failing data will be serially scanned out of the controller through **scan\_out** (see the table on the opposite page). Once the failing data has been scanned out, the BIST controller resumes the test and resets fail\_h to 0. At the end of the test, fail\_h is asserted to 1 if there has been any failing data. The scan out operation will repeat on every occurrence of a miscompare.

In order to synthesize the diagnostic functionality into the BIST controller, the following conditions must be met.

1. The BIST controller must use a comparator for verification.
2. Only algorithms supporting the comparator can be used. These include march1, march2, march3, unique address, checkerboard, and topological checkerboard.
3. The hold\_1 signal must be added to the BIST controller.

The diagnostics capability is added by using these commands:

```
setup controller hold -on | -off
setup controller debug -on | -off
set comparator test -on | -off
set comparator failflag {-Common | -SEparate} {-Singlefail | -Multifail}
```

You can also set a slow clock to scan out diagnostic data. The cycle time of the diagnostic clock is two times slower than a BIST clock with a default of 200ns. Use the Setup Diagnostic Clock -Diag\_clock command in conjunction with the Setup Controller Naming -Diag\_clk diag\_clk command to set up a diagnostic clock.

```
setup diagnostic clock diag_clock
setup controller naming -diag_clk diag_clk
```

## Clock Synchronization

There are two clock domains for the diagnostic process in the Memory BIST controller. One clock controls the diagnostic clock domain that scans out diagnostic data to the Automatic Test Equipment (ATE). This clock domain is

usually relatively slow. A second clock domain is run by the bist clock that operates everything except the diagnostic data scan-out and operates at a faster clock speed. In default operation, MBISTArchitect operates with these clocks in a non-synchronized relationship. When you turn synchronization on, these clocks become synchronized by passing information between the domains. For more information, see the “Synchronization between BIST Clock and Diagnostic Clock” section in Chapter 3 of the *Built-In-Self-Test Process Guide*.

# Compressor vs. Comparator

---

## Compressor versus Comparator

- ♦ **Most people use a comparator because:**
  - It stops on the first fail
  - You can add diagnosis capabilities to the BIST controller
- ♦ **Some people use compressors because:**
  - A ROM test requires it

In Memory BIST, the output response analysis is performed either by means of a comparator or a compressor.

A comparator provides some unique benefits such as diagnostic capability at the expense of higher area overhead. Since comparator width is the same as the memory data width, this area overhead increases for wide memories.

However, a comparator has the capability of stopping on the first fail or in the case of debug mode, stopping on every failure and scanning out the data. These features result in good diagnostic capability by providing precise information about failure location.

A compressor, on the other hand, entails relatively less area overhead. Compressor width can be different than that of the memory data width. Output from a memory wider than a compressor has to be fed to the compressor through a properly designed XOR tree.

Since the contents of a ROM are predetermined and cannot be changed by the BIST controller, the expected reference data for a comparator would have to be provided through the duplication of ROM contents. In general, this results in a large area overhead and is unacceptable. Thus, using a compressor for analyzing the ROMs output response is the only viable alternative.

Compressors cannot provide good diagnostic capability since their contents, in general, are checked only at the end of a test. Precise identification of the fault location based on the final content of a compressor is a difficult task.

Compressors can be placed immediately at the output of a memory, or in case of an embedded memory, can be placed downstream. Placing the compressor downstream tests the logic between memory outputs and compressor. However, diagnostic capability will be further worsened since a fault now can be either in the memory or in the intervening logic.

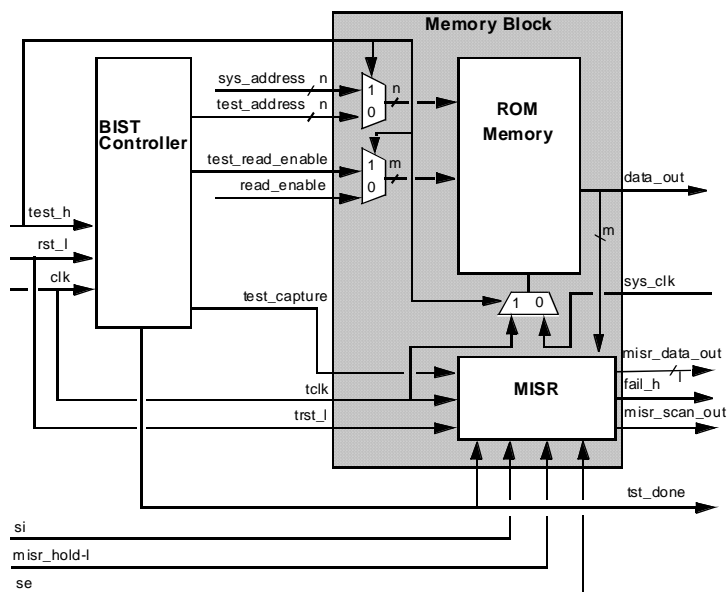
The MBISTArchitect tool has configurations that use a compressor (MISR) to capture the output of the memory under test. You use the Setup MBist Compressor command to define compressor parameters.

Use **Setup MBist Compressor -scan** to scan out the final signature and compare it with the tester. Use **Setup MBist Compressor -localcomparator** to generate a signature comparator in the memory BIST collar (locally).



## BIST using a Compressor

### BIST Using a Compressor



3-18 MBISTArchitect:Common BIST Variations

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Within an MBISTArchitect session, you can generate either a BIST controller with a comparator or a compressor configuration. If you specify a compressor configuration, MBISTArchitect generates a separate HDL model for each compressor(s).

Before you can tell MBISTArchitect to generate a compressor configuration, you must specify that the controller should use a compressor.

For example, the following set of commands generates the compressor shown:

```
shell> $MGC_HOME/bin/mbistarchitect -library dft.lib
```

```
MBIST> add memory model ram4x4 ram8x8 ram8x8
MBIST> set controller hold -on
MBIST> setup observation scheme -compress
MBIST> setup mbist compressor -hold [-localcomparator]
MBIST> run
MBIST> save bist
MBIST> exit
```

# Adding Pipeline Registers

---

## Adding Pipeline Registers

- ◆ Specify number of input and output stages
- ◆ Pipeline registers are separate modules in the BIST controller file
- ◆ Testbench accounts for pipeline stages
  - Example Request:  
MBIST> setup controller pipeline -depth  
input\_depth 2 output\_depth 3



constraints. By specifying the comparator's position, you can control from which pipeline stage you want to take data for comparison.

For example, by entering the following setup command, MBISTArchitect will generate two input and three output pipeline registers:

**setup controller pipeline -depth input\_depth 2 output\_depth 3**

In this case, MBISTArchitect creates the registers as separate instantiations in the connection model and modifies the controller timing to account for the pipeline delay. Notice that pipeline registers can only be added to the address and data paths and not to any other control signals.

Adding pipeline registers is also used in full-speed testing, see [“Full-Speed Overview” on page 3-14](#) for more information.

# Specifying Non-controlled Memory Ports

---

## Specifying Non-controlled Memory Ports

- ♦ Memory ports not to be controlled by BIST Controller
- ♦ Default assertion state is high
  - Test bench holds the signal at the value opposite its assert state
- ♦ Default direction is input
  - Except for “data\_out” and “data\_inout”
- ♦ Define in Library Model’s bist\_definition section

# Specifying Non-controlled Memory Ports

---

## Specifying Non-controlled Memory Ports (Continued)

### ♦ Library Model's bist\_definition section

```
...  
    bist_definition (  
        ...  
        dont_touch port_name assert_state direction;  
        ...  
    ) // end BIST definition
```

- port\_name = pin or bus to be left untouched.
- assert state = “high” (default) or “low”
- direction = “input” or “output”

You can use a clause on the bist\_definition section of the MBISTArchitect memory model to specify which ports on the memory should not be controlled by the BIST Controller. The default assertion state is high and the default direction is input except for “data\_out” and “data\_inout.”

# Specifying Parameters for Memory Clock Signals

---

## Specifying Parameters for Memory Clock Signals

- ♦ **Memory Clock Signal Gate Parameters**
  - **Clock Gating On**
    - Generate multiplexer in path of clock signal
    - MBIST> setup memory clock -control
  - **Clock Gating Off (Default)**
    - Use the system clock for BIST testing
    - MBIST> setup memory clock -system
- ♦ **Synchronize Controller with Memory Clock**
  - **Synchronous with Clock's Rising Edge (Default)**
    - MBIST> setup memory clock -test noinvert
  - **Synchronous with Clock's Falling Edge (inverted)**
    - MBIST> setup memory clock -test invert

---

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The following commands support clock gate control:

SETup MEemory Clock {-**System** | -**Control** | {-**Test** {**Noinvert** | **Invert**}}

The -Control switch specifies that the memory clock should be gated. The default is -System. When the -Control switch is used, the test mode clock is connected to the clock control signals created by the BIST controller.

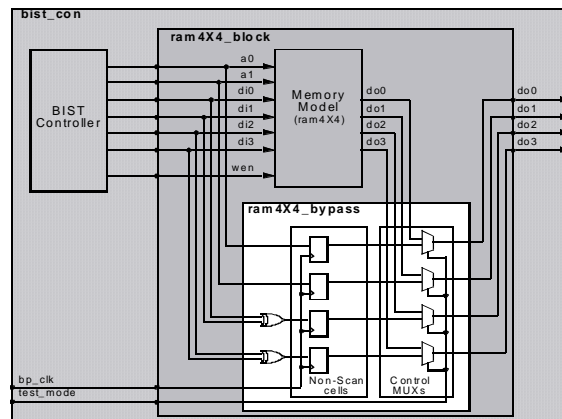
The -Test Noinvert switch lets you specify whether the controller is synchronous with the rising edge or falling edge (inverted) of the clock. The default is to not invert.



# Bypassing Memory in Scan Mode

## Bypassing Memory in Scan Mode

- ◆ Propagates fault efforts around memories
- ◆ Allows high fault coverage for scan and logic BIST designs with embedded memories



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Use these commands:

**SET SCan Logic [-Addr\_observe *integer*] [-Data\_observe *integer*]  
[-NoScan | -Scan] [-Control | -NOControl] [-CNtrl\_observe *integer*]**

Where:

- -Addr\_Observe: Number of cells to observe address
- -Data\_Observe: Number of cells to observe data
- -Scan: Generate scan cells and scan chain (not default)
- -Control: Multiplex bypass cell outputs onto memory cell outputs
- -NOControl: Do not multiplex bypass cell outputs onto memory cell outputs

# Bypassing Memory in Scan Mode (Continued)

---

## Bypassing Memory in Scan Mode (Continued)

### ♦ Commands used:

- **SET SCan Logic** [-Addr\_observe *integer*] [-Data\_observe *integer*] [-NOScan | -Scan] [-Control | -NOControl] [-CNtrl\_observe *integer*]
  - -Addr\_observe : Number of cells to observe address
  - -Data\_observe : Number of cells to observe data
  - -Scan : Generate scan cells and scan chain (not default)
  - -Control : Multiplex bypass cell outputs onto memory cell outputs
  - -NOControl : Do not multiplex bypass cell outputs onto memory cell outputs

You can direct MBISTArchitect to configure scan logic to bypass the memory during scan mode. This is done by XORing all the address lines and all the data input lines to generate a specified number of compressed signals. Each of these compressed signals are captured in scan or non-scan cells. These cells are clocked using a new signal line named **bp\_clk**. If you choose to specify scan cells, MBISTArchitect generates three additional signal lines: scan\_enable, scan\_in, and scan\_out.

The default -Control option is provided to multiplex the scan/non-scan cell output to the memory data output. This is helpful in testing logic on the output side of the memory during scan test. MBISTArchitect inserts one multiplexer for each data output. It connects one input of the multiplexer to memory data output and the other input to the newly inserted scan/non-scan cells. The multiplexer is controlled by the **test\_mode** signal.

The bypass logic created by this command is placed in a hierarchical block called `memory_name_bypass`. Also, a new level of hierarchy called `memory_name_block` is created if it doesn't already exist. This `memory_name_block` is created or modified to contain both the memory and the `memory_name_bypass` blocks. The description of the `memory_name_bypass` and `memory_name_block` are in the same file as the BIST Controller.

To help with testing the logic that surrounds your memory design, MBISTArchitect allows you to add memory bypass circuitry using the Set Scan Logic command. This bypass circuitry compresses the address and data input lines through XOR logic and either scan or non-scan cells into a specified number of output signals. By using the command's default -Control switch, these output signals are multiplexed with the memory data output lines. The multiplexers are controlled by *test\_mode*. When *test\_mode* is asserted high for testing the surrounding logic, the memory is bypassed and the compressed address and data input signals are presented to the data output lines. This allows you control over the downstream logic during testing.

For a detailed description of the MBISTArchitect memory bypass functionality, refer to the Set Scan Logic command description in the *MBISTArchitect Reference Manual*.

# Synthesis Driver File

---

## Synthesis Driver File

- ♦ MBISTArchitect can produce a basic synthesis script:
  - For Synopsys environments
  - Named *<design>\_dcscript* for Synopsys Design Compiler (by default)
  - That you can use as a template or example of a basic synthesis/optimization run

MBISTArchitect can write a basic synthesis script, targeted for Synopsys' Design Compiler tools. You can use this script as a template for synthesizing and optimizing the BIST models MBISTArchitect produces.

While you can change the model's name using Setup File Naming, by default MBISTArchitect names this model *<design>\_dcscript* (for Synopsys Design Compiler).

The Design Compiler can save the BIST controller and BIST block in a single file for Verilog. For VHDL, the BIST controller and BIST block can be saved to separate files.

An example is shown below:

```
/* -----  
// File Type:      Logic Synthesis Script File  
// Date Created:   Wed Feb  6 21:02:00 2002  
// Tool Version:   v8.9_6.02   Wed Feb  6 15:31:55 PST 2002  
// -----  
*/  
  
sh mkdir work  
define_design_lib work -path "./work"  
read -format verilog top_after.v  
  
current_design cti_sab  
uniquify  
compile  
write -format verilog -hierarchy -output "cti_sab_gate.v"  
  
current_design ram4x4_multi_bist  
uniquify  
compile  
write -format verilog -hierarchy -output  
"ram4x4_multi_bist_gate.v"  
  
current_design ram4x4_multi_bist_ram4x4_block_0  
uniquify  
compile  
write -format verilog -hierarchy -output  
"ram4x4_multi_bist_ram4x4_block_0_gate.v"  
  
current_design ram4x4_multi_bist_ram4x4_block_1  
uniquify  
compile  
write -format verilog -hierarchy -output  
"ram4x4_multi_bist_ram4x4_block_1_gate.v"  
  
current_design ram4x4_multi_bist_ram4x4_block_2  
uniquify  
compile  
write -format verilog -hierarchy -output  
"ram4x4_multi_bist_ram4x4_block_2_gate.v"  
exit
```

# Design Compiler Clock Constraints

---

## Design Compiler Clock Constraints

- ◆ **MBISTArchitect lets you define clock constraints within the Synopsis Design Compiler script**

- **Example:**

```
Clock_Period=100
create_clock-period CLOCK_PERIOD RCLK
set dont_touch_network RCLK
create_clock-period CLOCK_PERIOD BIST_CLK
set dont_touch_network BIST_CLK
```

The MBISTArchitect Design Compiler script has been modified to let you define clock constraints. This script uses the currently defined clock width and add clock constraints for the BIST controller clock and any other memory model defined clock signals, where clock gating is disabled.

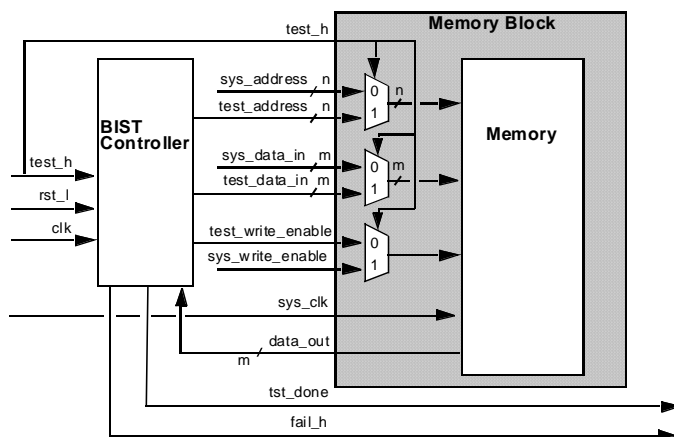
capture and write the output values of the memory itself. You specify this information using the Setup Mbist Patterns command.

While you can change the model's name using Setup File Naming, by default MBISTArchitect names this output *<design>\_bist.pat*.

# Mux-Embedded Memory Support

## Mux-Embedded Memory Support

- ♦ MBISTArchitect supports mux-embedded memory structures
  - The tool uses the inserted muxes within memory blocks to reduce overhead and timing penalties



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MBISTArchitect supports mux-embedded memory structures. It is often to your advantage to design control muxes within memory blocks to reduce overhead and timing penalty. The tool uses the inserted mux to select test or system signals and support by-pass logic signals inside of the memory block. You can design the data output to either separate system and test pins or to a single output for both signals.

## Library Enhancement

To support the mux-embedded memory structure, a library format is used to specify which signals are paired. The following are mux-embedded memory support signals that are used in this library format:

```
port_type sys_name:test_name width active_state
```

A BIST-mode signal is used to select either BIST signals or system signals. This signal is necessary and is used as follows:.

`bist_mode name active_state`

To support embedded bypass logic inside of mux-embedded memories, the following signals should be used:

ATPG\_mode, scan\_clk, scan\_enable, scan\_in, and scan\_out.

`atpg_mode name active_state`

For information on Mux-Embedded Memory Support limitations and examples of the mux-embedded memory support library format, see the “Mux-Embedded Memory Support” section in Appendix A of the *MBISTArchitect Reference Manual*.



## Module 3 Lab Exercises

- **Changing the BIST Algorithm**  
(20 minutes)
- **Changing the Data Background**  
(20 minutes)
- **Inserting BIST for Multiple Memories**  
(15 minutes)
- **Adding BIST with a Compressor**
- **Implementing Full-Speed BIST**  
(20 minutes)
- **Adding BIST for Bidirectional Memories**  
(10 minutes)
- **Adding BIST for ROMs**  
(30 minutes)

## Module 3: Lab Exercises

You may be able to increase test coverage and reduce area by taking the time to configure the BIST circuitry to your design. The following exercises will give you a start at customizing BIST for different design configurations that you may encounter. You may choose to do all exercises or only those that fit your design needs.

**Exercise 3: Changing the BIST Algorithm**—In this exercise, you will generate a new memory BIST collar that uses the March 1 rather than the default March 2 algorithm to reduce test time.

**Exercise 4: Changing the Data Background**—In this exercise, you will improve the test coverage of the March 1 BIST Controller by adding three different data pattern backgrounds.

**Exercise 5: Inserting BIST for Multiple Memories** —In this exercise, you will save area and overhead by sharing a BIST controller for multiple memories.

**Exercise 6: Adding BIST with a Compressor** — In this exercise, you will add a Compressor instead of a Comparator. A Compressor may be used to improve area-overhead or optimize routing.

**Exercise 7: Running BIST at Full-Speed**—In this exercise, you run BIST at “full-speed” meaning it will run at the system clock speed to test the memory at the full speed at which the system will run.

**Exercise 8: Adding BIST for Bidirectional Memories**—In this exercise, you add BIST for bidirectional memories.

**Exercise 9: Adding BIST for ROMs**—In this exercise, you add BIST for ROMs. A ROM requires the use of a compressor.

**Note**

These exercises should take approximately 1 hour and 20 minutes.

### Exercise 3: Changing the BIST Algorithm

This exercise should take approximately 20 minutes to complete.

Selecting one or more algorithms for your BIST design depends on the type of memory you are testing, your test goals, your overall test strategy, and the advice you may receive from in-house memory-test experts and ASIC vendors.

The March 2 (March C+) algorithm is the MBISTArchitect default because it is so commonly used and accepted.

In this exercise, you will direct MBISTArchitect to use the March 1 algorithm when you generate a BIST collar for the **ram4x4** memory. Remember that you can *add* new algorithms as well as change algorithms. You will gain experience adding algorithms in later exercises.

Do the following:

1. Verify that you are in the `$MBISTNWP/mbist1/ram4x4/results` directory.
2. Invoke MBISTArchitect while loading the library at invocation.  

```
shell> mbistarchitect -lib ../design/ram4x4.atpg
```
3. Add the **ram4x4** model to the list of memory models for BIST insertion.
4. Select the **Controller** block to access the Setup Mbist Controller dialog box.
5. Select the Test Algorithms tab, select **March 1**, then click **OK**.
6. Click **Run** to add default BIST circuitry to this model.
7. Save the default outputs by clicking **Save BIST**. After verifying the format and destination directory, click **OK**.

MBISTArchitect prompts you with the following question:

```
One or more of the output files already exist. Do you  
want to overwrite them?
```

Click **No** in the Question dialog box, then **Cancel** the Save Bist Results dialog box.

In Exercise 1, you saved outputs with the same default filenames. Instead of replacing these files, you can give them custom names.

8. Click on the **Output Files Names** button in the Control Panel Window. In the Setup Output File Naming dialog box, change the filenames to the following, then click **OK**:

<u>Old Filename</u>	<u>New Filename</u>
ram4x4_bist.v	ram4x4_m1_bist.v
ram4x4_bist_con.v	ram4x4_m1_bist_con.v
ram4x4_tb.v	ram4x4_m1_tb.v

9. Click **Save BIST**.
10. From the Command Line, list the generated outputs:

```
MBISTA> system ls ram*
```

The System command lets you issue an operating system command. The **ls** command shows the contents of the working directory. You should see:

```
ram4x4_bist.v      ram4x4_m1_bist.v    ram4x4_m1_tb.v
ram4x4_bist_con.v  ram4x4_m1_bist_con.v  ram4x4_tb.v
```

MBISTArchitect generated three new Verilog models for the ram4x4 model:

- *ram4x4\_m1\_bist.v* - a model that contains just the ram4x4 BIST control circuitry,
- *ram4x4\_m1\_bist\_con.v* - the connection model that connects the BIST controller to the ram4x4 simulation model, and
- *ram4x4\_m1\_tb.v* - the testbench that instantiates and tests the *ram4x4\_m1\_bist\_con.v* model.

11. Click **View Saved Design Files**.

MBISTArchitect generated three Verilog models for the ram4x4 model:

Look at *ram4x4\_m1\_bist.v*. This model is very much the same as *ram4x4\_bist.v*, except for the March 1/March 2 algorithm differences. Both algorithms perform the basic March test, with March1 eliminating the RWR operation to reduce the algorithm from 14n to 10n.

12. **Exit** the tool.

13. Compile the outputs.

- a. Set up a new work directory for the March1 test models.

```
shell> $MGC_HOME/bin/vlib work_m1
```

- b. Compile the memory simulation model, all BIST models, and the testbench.

```
shell> $MGC_HOME/bin/vlog -work work_m1 \  
      ../design/ram4x4.v ram4x4_m1_bist.v \  
      ram4x4_m1_bist_con.v ram4x4_m1_tb.v
```

14. Simulate the BIST circuitry

- a. Invoke the ModelSim simulator and load the ram4X4 testbench.

```
shell> $MGC_HOME/bin/vsim -lib work_m1 ram4x4_tb
```

- b. Set up the lists by running the following dofiles:

```
VSIM 2> do ../design/vsim_setup.do
```

- c. Run the simulation until it is finished.

```
VSIM 3> run -all
```

- d. Run a little more to capture the complete pattern for the tst\_done signal.

```
VSIM 4> run 50
```

- e. Write the displayed list to a file.

```
VSIM 5> write list trace.log.m1
```

f. Quit the simulation.

VSIM 6> **quit**

Examine the saved list file.

### Exercise 4: Changing the Data Background

This exercise should take approximately 20 minutes to complete.

This exercise repeats the steps you performed in the last exercise—with one exception. In this exercise, instead of having the March 2 test write words of 0s and 1s, you will tell MBISTArchitect to create a March2 pattern generator that uses 1010, 0010, and 0100 as the data backgrounds.

1. Ensure you are in the `$MBISTNWP/mbist1/ram4x4/results` directory.
2. Invoke MBISTArchitect.
3. Load the *ram4x4.atpg* library from the `../design` directory, (**but do not add the model to the memory list**).
4. Add the ram4x4 model to the list of memory models for BIST insertion.

```
MBISTA> add me m ram4x4
```

5. Change the data background for the March 2 algorithm.

```
MBISTA> add data backgrounds 1010 0010 0100
```

Because you specified three patterns, MBISTArchitect applies the March 2 algorithm three times. In the first March 2 test, the algorithm uses the word value 1010 instead of 0000 and then uses the inverse, 0101, instead of 1111. In the second March 2 test, the algorithm uses 0010 instead of 0000, and then uses the inverse, 1101, instead of 1111. In the third March 2 test, the algorithm uses 0100 instead of 0000, and then uses the inverse, 1011, instead of 1111.

6. Generate the BIST circuitry for this memory model.

```
MBISTA> run
```

7. Set up file naming.

```
MBISTA> setup file naming -bist_model ram4x4_m1db_bist.v -connected \
ram4x4_m1db_bist_con.v -test_bench ram4x4_m1db_tb.v
```

8. Save the output files with the customized names.

MBISTA> **save bist -script**

Because you specified the -Script switch, MBISTArchitect saves a synthesis script file named ***ram4x4\_bist.v\_dcscript*** in addition to the regular outputs.

9. Exit the tool.

MBISTA> **exit**

10. Compile the outputs and simulate the testbench. Create a new work directory called ***work\_m2db*** for the compilation and simulation results. You can use the ***../design/vsim\_setup\_db.do*** file to setup the simulation. Name the trace file ***trace.log.m2db***. If you need assistance with this process, refer back to [Exercise 2: Verifying the BIST Circuitry](#).
11. Observe from the List Window that the background patterns you specified are written to and read from the memory.
12. Exit the simulator.



### Exercise 5: Inserting BIST for Multiple Memories

This exercise should take approximately 15 minutes to complete.

This exercise explores several additional features of MBISTArchitect. First, it creates a single BIST controller for two memories: an 8x4 RAM and a 4x4 RAM. Second, the BIST controller applies the March 2 algorithm to the first write port and the “unique address” algorithm to the second write port. Third, MBISTArchitect produces the BIST controller in VHDL format.

Now that you have become more acquainted with the GUI features, you will be able to utilize the command line to take advantage of minimum typing and other features.

1. Change directories.

```
shell> cd $MBISTNWP/mbist2/multi_ram_dwp/results
```

This is where you will do your work and save your results.

2. Invoke MBISTArchitect.

```
shell> mbistarchitect
```

3. Load the appropriate libraries (you can load only one library at a time):

```
MBISTA>load library ../design/ram4x4.atpg  
MBISTA>load library ../design/ram8x4.atpg
```

4. Add both the ram4x4 and ram8x4 models to the list of memory models for BIST insertion:

```
MBISTA> add mem ram4x4 ram8x4
```

5. Add the “Unique Address” algorithm to port 2.

```
MBISTA> add mbist algorithm 2 unique
```

**Note**

The BIST controller automatically applies the March 2 algorithm to port 1. This command adds the unique address algorithm to port 2 (the second port of the ram8x4 memory), replacing the default March 2 algorithm for this specified port only.

If you issued the Add Mbist Algorithms command Add MBIST Algorithms command again for port 2, the BIST controller would apply both specified algorithms to port 2.

6. Add default BIST circuitry to this model.

```
MBISTA> run
```

7. Save default VHDL-format outputs with the default names.

```
MBISTA> save bist -vhd
```

Note that when you generate a BIST controller for multiple memories, MBISTArchitect names the saved outputs *<first\_memory>\_multi\_bist.vhd*, *<first\_memory>\_multi\_bist\_con.vhd*, and *<first\_memory>\_multi\_tb.vhd*, by default. In this case, you added ram4x4 first with the Add Memory Models command, so “ram4x4” becomes the prefix for each saved file.

8. Examine the generated outputs using the **View Saved Design Files** button in the Control Panel window.
9. Exit the tool.

```
MBISTA> exit
```

10. Compile the outputs and simulate the testbench.

- a. Set up a work directory for the March2/Unique test models.

```
shell> $MGC_HOME/bin/vlib work
```

- b. Compile the core logic, all BIST models, and the testbench.

```
shell> $MGC_HOME/bin/vlog ../design/ram4x4.v \
      ../design/ram8x4.v
shell> $MGC_HOME/bin/vcom -explicit \
      ram4x4_multi_bist.vhd ram4x4_multi_bist_con.vhd \
      ram4x4_multi_tb.vhd
```



### Note

You use **vlog** for compiling Verilog (the original memory models) and **vcom** for compiling VHDL (the MBISTArchitect-generated outputs). In this exercise, you perform mixed Verilog/VHDL simulation using ModelSim after compiling the models.

- a. Invoke the ModelSim simulator and load the testbench model.

```
shell> $MGC_HOME/bin/vsim ram4x4_multi_tb
```

- b. Set up the lists by running the following dofile:

```
VSIM 1> do ../design/vsim_setup.do
```

- c. Run the simulation until it is finished.

```
VSIM 2> run -all
```

- d. Run a little more to capture the complete pattern for the tst\_done signal.

```
VSIM 3> run 50
```

- e. Write the displayed list to a file.

```
VSIM 4> write list trace.log.m2.un
```

- f. Quit the simulation.

```
VSIM 5> quit
```

Examine the saved list file.

The BIST controller runs testing on the RAM4x4 and RAM8x4 memories in parallel, first running the March 2 algorithm on port 1 of each memory, followed by running the Unique Address algorithm on port 2 of RAM8x4.

The Unique Address algorithm places the address value in the address location. For example, the algorithm places address value 0000 in location 0, address value 0001 in location 1, address value 0010 in location 2, and so on. If the address and data widths do not match, the algorithm concatenates the MSB values of the address and places them as the LSB of the data word, to pad the data word to the appropriate size.

In this exercise, the address bus has three bits while the data width has four. So the algorithm pads the data word by duplicating the most significant address bit as the least significant data word bit to increase the word size to four bits. For example, in this case the algorithm places the value 0000 at location 0, 0010 at location 1, 0100 at location 2,..., 1011 at location 5, and so on. Note that you see the data values in reverse bit order (LSB->MSB) during simulation.



**Note**

At this time, the descending March test performs the read/write/read operation in the order 0,3,2,1 (for RAM4x4) or 0,7,6,5,4,3,2,1 (for RAM8x4).

Table 3-1 provides a breakdown of the testbench simulation, and thus, the memory BIST controller operation.

**Table 3-1. March2 and Unique Address Simulation Activity**

Time (ns)	RAM4x4 Activity/Address	RAM8x4 Activity/Address	Algorithm Operation
<b>BEGIN BIST LOGIC TESTING (Initialize for the BIST controller)</b>			
0-2975	Write0 / 0 Read0 / 0	Write0 / 0 Read0 / 0	Test for the system path
<b>BEGIN MARCH 2 TESTING OF RAM4X4 AND RAM8X4 PORT 1</b>			
3050-3750 3750-4550	Write0 / 0->3 Hold WEN off	Write0 / 0->3 Write0 / 4->7 Hold WEN B off	Initialize for March 2 Test

**Table 3-1. March2 and Unique Address Simulation Activity**

<b>Time (ns)</b>	<b>RAM4x4 Activity/Address</b>	<b>RAM8x4 Activity/Address</b>	<b>Algorithm Operation</b>
4650-6850 7050-9250	Read0,Write1,Read1/ 0->3 Hold WEN off	Read0,Write1,Read1/ 0->3 Read0,Write1,Read1/ 4->7 Hold WEN B off	▲ Read0 Write1 Read1
9450-11650 11850-14050	Read1,Write0,Read0/ 0->3 Hold WEN off	Read1,Write0,Read0/ 0->3 Read1,Write0,Read0/ 4->7 Hold WEN B off	▲ Read1 Write0 Read0
14550-16450 16650-18850	Hold WEN off Read0,Write1,Read1/ 3->0	Read0,Write1,Read1/ 7->4 Read0,Write1,Read1/ 3->0 Hold WEN B off	▼ Read0 Write1 Read1
19050-21250 21450-23650	Hold WEN off Read1,Write0,Read0/ 3->0	Read1,Write0,Read0/ 7->4 Read1,Write0,Read0/ 3->0 Hold WEN B off	▼ Read1 Write0 Read0
23685-25250	Read0 / 3->0 Read0 / 3->0	Read0 / 7->4 Read0 / 3->0 Hold WEN B off	▼ Read0
<b>BEGIN UNIQUE ADDRESS TESTING OF RAM8X4 PORT 2</b>			
25550-26950	Hold WEN off	Hold WENA off 0->7	▲ Write address value to address location
27050-28550	Hold WEN off	Hold WENA off 0->7	▲ Read value from address location
28650-30050	Hold WEN off	Hold WENA off 0->7	▲ Write1 to all address locations

**Table 3-1. March2 and Unique Address Simulation Activity**

<b>Time (ns)</b>	<b>RAM4x4 Activity/Address</b>	<b>RAM8x4 Activity/Address</b>	<b>Algorithm Operation</b>
30250-31750	Hold WEN off	Hold WENA off 7->0	▲ Write inverse address value to address location
31850-33250	Hold WEN off	Hold WENA off	▲ Read value from address locations

### Exercise 6: Adding BIST with a Compressor

This exercise should take approximately 40 minutes to complete.

This exercise demonstrates how to generate BIST circuitry that uses a compressor instead of a comparator. Since you have invoked MBISTArchitect and generated BIST circuitry several times in previous exercises, this exercise does not provide as much detail as the previous exercises. If you need assistance, refer to [Exercise 1: Creating a Basic Memory BIST Collar](#).

This exercise again uses the RAM4x4 model—for the sake of both simplicity and comparison to the architectures generated by earlier exercises.

1. Change to the `$MBISTNWP/mbist2/ram4x4/results` directory.
2. Invoke MBISTArchitect, loading the *ram4x4.atpg* library (*../design/ram4x4.atpg*) at invocation.
3. Add the ram4x4 model to the list of memory models for BIST insertion.
4. Specify that the BIST controller should not include a comparator in the architecture.

```
MBISTA> setup observation scheme -compress
```

The -Compress switch tells MBISTArchitect not to include a comparator as part of the controller. In this case, you want to use a compressor for signature analysis, instead of a comparator. You will set up the compressor parameters in the next step.

5. Set up the compressor parameters.

```
MBISTA> setup mbist compressor -low 32
```

This specifies for MBISTArchitect to generate a compressor model associated with RAM4x4 with a MISR length of 32 bits.

Due to character conflicts, the minimum typing for Setup Mbist Compressor is “set mb com” and the minimum typing for Setup Mbist Controller is “set mb con”.

6. Run the BIST circuitry generation process.
7. Set up output file naming.

You already generated default outputs in a previous exercise. Because you do not want MBISTArchitect to overwrite these models, you should give the models generated in this exercise unique names.

```
MBISTA> setup file naming -bist_model ram4x4_nocompare_bist.v \  
-connected ram4x4_nocompare_bist_con.v \  
-test_bench ram4x4_nocompare_tb.v
```

8. Save the default outputs, with the customized names, in Verilog format.

```
MBISTA> save bist -r
```

9. List the generated outputs.

```
MBISTA> system ls *.v
```

Because of the compressor model, this time when you saved MBISTArchitect generated FOUR new Verilog models for the ram4x4 model. These models include:

- *Compressor\_lib.v* - the compressor-only model.
- *ram4x4\_nocompare\_bist.v* - a model that contains just the ram4x4 BIST control circuitry.
- *ram4x4\_nocompare\_bist\_con.v* - the connection model for the controller and the RAM collar.
- *ram4x4\_nocompare\_tb.v* - the testbench that instantiates and tests the *ram4x4\_nocompare\_bist\_con.v* model.

10. Reset the state of MBISTArchitect and make some changes within the session.

Assume you examined the files and decided you want to implement a hold\_1 signal. This signal lets you pause BIST testing with a low value on the hold\_1 signal, retaining the state of the BIST test process. When the hold\_1 signal returns to a high state, the BIST test continues. The hold\_1



signal, among other purposes, enables you to perform data retention testing. If you pause testing between a write and a read, the read performed after testing should display the expected values from the write operation. If not, the memory could have a data retention problem.

Next, assume you decide to generate a synthesis script for the MBISTArchitect outputs as well as a file capturing the BIST-generated inputs to the RAM4x4 memory. You can do this within the current session by resetting the state and running the additional commands as follows:

```
reset state
add me m ram4x4
report memory models
set obs s -compress
set con h -on
set mb com -low 32 -hold
setup file naming -bist ram4x4_nocompare_bist.v -con \
    ram4x4_nocompare_bist_con.v -t ram4x4_nocompare_tb.v \
    -script ram4x4_nocompare_synth.script
run
save bist -scr -r
```

**Hint:** Instead of entering these commands interactively, run the `../design/nocomp.do` dofile.

### 11. Examine the generated outputs.

First look at *ram4x4\_nocompare\_bist.v*. Notice that this model contains two signals that the previous models did not: `test_capture_0` and `hold_1`.

You should also notice that the connection file, *ram4x4\_nocompare\_bist\_con.v*, instantiates the BIST controller and the RAM collar.

### 12. Exit MBISTArchitect.

### 13. Compile the outputs and run the simulation using the following script. Verify that the final signature is 8482e23a.

```
shell> runsim
```

Answer “No” to the question about finishing.

14. Examine the synthesis template script generated.

The synthesis file, *ram4x4\_nocompare\_synth.script*, in your results directory provides a template script for compiling and synthesizing the BIST controller model in the Design Compiler environment.

15. If you have time, and want to explore more of the available algorithms for the compressor architecture, repeat this exercise specifying one of the other algorithms—such as Diagonal or Checkerboard—instead of the default March 2 algorithm.

### Exercise 7: Running BIST at Full-Speed

This exercise will take approximately 20 minutes. We will generate BIST circuitry using the default values for MBISTArchitect with the exception of adding the library and BIST changes required to run at full speed.

**FULL SPEED** is defined as clocking with back to back read/write cycles.

1. Change to the following working directory:

```
shell> cd $MBISTNWP/full_speed/design
```

2. List the design files you will be using in this exercise:

```
shell> ls -ltr *
```

3. Change directories to the *full\_speed/results* directory:

```
cd ../results
```

4. Invoke MBISTArchitect:

```
shell> mbistarchitect
```

5. Click on the **Memory** block in the Control Panel graphic pane. This will start the process to load the “Full speed” memory model.

- a. Click on the **Browse** button, then navigate to *../design* directory.
- b. Double click the *lab13.atpg* file. A list of files displays, click on *Full\_speed.atpg*, then click **Load**. The **Full\_speed** model now appears in the available Models list.
- c. Select the **Full\_speed** model, then click **Add**.
- d. Click **OK**.

You’ve just added the memory model for BIST insertion.

6. Click on the **Controller** block in the Control Panel graphic pane.

Now you can modify the specific settings to enable the generation of the FULL SPEED memory BIST controller. You should now see the “Setup Mbist Controller” panel.

- a. Select the **Controller Options** tab at the top right of the “Setup MBIST Controller” panel.
- b. Make sure that the **System Clock** is selected in the Type of Memory Clock.
- c. Select the **Setup Pipelining...** button at the bottom left of the “Setup MBIST Controller” panel.

You should now see the “Setup Pipeline Staging” panel come up.

- i. Select **Pipeline Stages**
- ii. Select **Add Pipeline Controller Registers of Different Depths**
  - a. Set **#Input Stages** = 0
  - b. Set **#Output Stages** = 2
- iii. Set **Position of the Comparator** = 1
- iv. Select **Placement of Delay Stages**
  - a. Select **No Delay Set**
- v. Click **OK** in “Setup Pipeline Staging.” The following figure displays the settings for this dialog box.

## Common BIST Variations

The screenshot displays the BIST configuration interface with the 'Test Algorithms' tab selected. The main window shows various configuration options for the BIST controller. A dialog box titled 'Setup Pipeline Staging' is open, showing options for pipeline stages and delay registers.

**Controller Options**

- Test Algorithms
- Data Backgrounds
- Controller Naming
- Scan Logic
- Controller Diagnostics
- HDL Language

BIST Clock Period :  ns

Retention Cycles :  x Bist Clock Period

☐ Prevent Simultaneous Access of a Single Address from Multiple Ports During Read/Write Operations

Generate the Following

- ☐ "Hold" Control Signal
- ☐ BIST-in-Place Information
- ☒ Comparator
  - ☐ Comparator Test
  - ☒ Connections to Memory Models
  - ☒ Common Fail Flag
  - ☐ Multiple Fail Flag
- ☐ Multiplexors Located Inside Controller
- ☐ VHDL Configurations

Type of Memory Clock : ☐ Control Clock

Type of Diagnostic Clock : ☒ BIST Clock

- ☐ Synchronize Controller to Falling Edge
- ☐ Synchronous Reset Signal
- ☐ Sequential Test for Memory
  - ☐ Interleave Memory

**Setup Pipeline Staging**

- ☒ Pipeline Stages
  - ☐ Do Not Add Pipeline Controller Registers
  - ☐ Add Pipeline Controller Registers of Equal Depth
    - Depth :
  - ☒ Add Pipeline Controller Registers of Different Depths
    - #Input Stages :  #Output Stages :
- ☒ Position of the Comparator :
- ☒ Placement of Delay Stages
  - ☐ Set Delays for All Controller Pipelines
  - ☐ Number of Input Delay Registers :
  - ☐ Number of Output Delay Registers :
  - ☒ No Delays Set

- d. Click **OK** in the “Setup Mbist Controller” panel.
7. Click **Run** in the Control Panel. This will generate the BIST circuitry.
8. Click **Save BIST** and click OK. This will generate the BIST circuitry and add to the BIST model.

9. Click **View Saved Design Files**.

Next, look at the files you just generated. You should see three new files. (*Full\_speed\_bist.v*, *Full\_speed\_bist\_con.v*, *Full\_speed\_tb.v*). Look at the *\*\_bist.v* file and try to identify the new pipeline registers.

10. What makes the memory model different in AT Speed vs. FULL Speed? Look at the memory models and compare them.

- a. Open up another shell window:

```
shell> cd $MBISTNWP/mbist/full_speed/design
```

- b. Use your text editor to chose and view the library file.

```
shell> vi lab13.atpg
```

You will see two RAM model definitions in this library file. The first is called *At\_speed* and the second is called *Full\_speed*.

Try and identify all the differences between these two models.

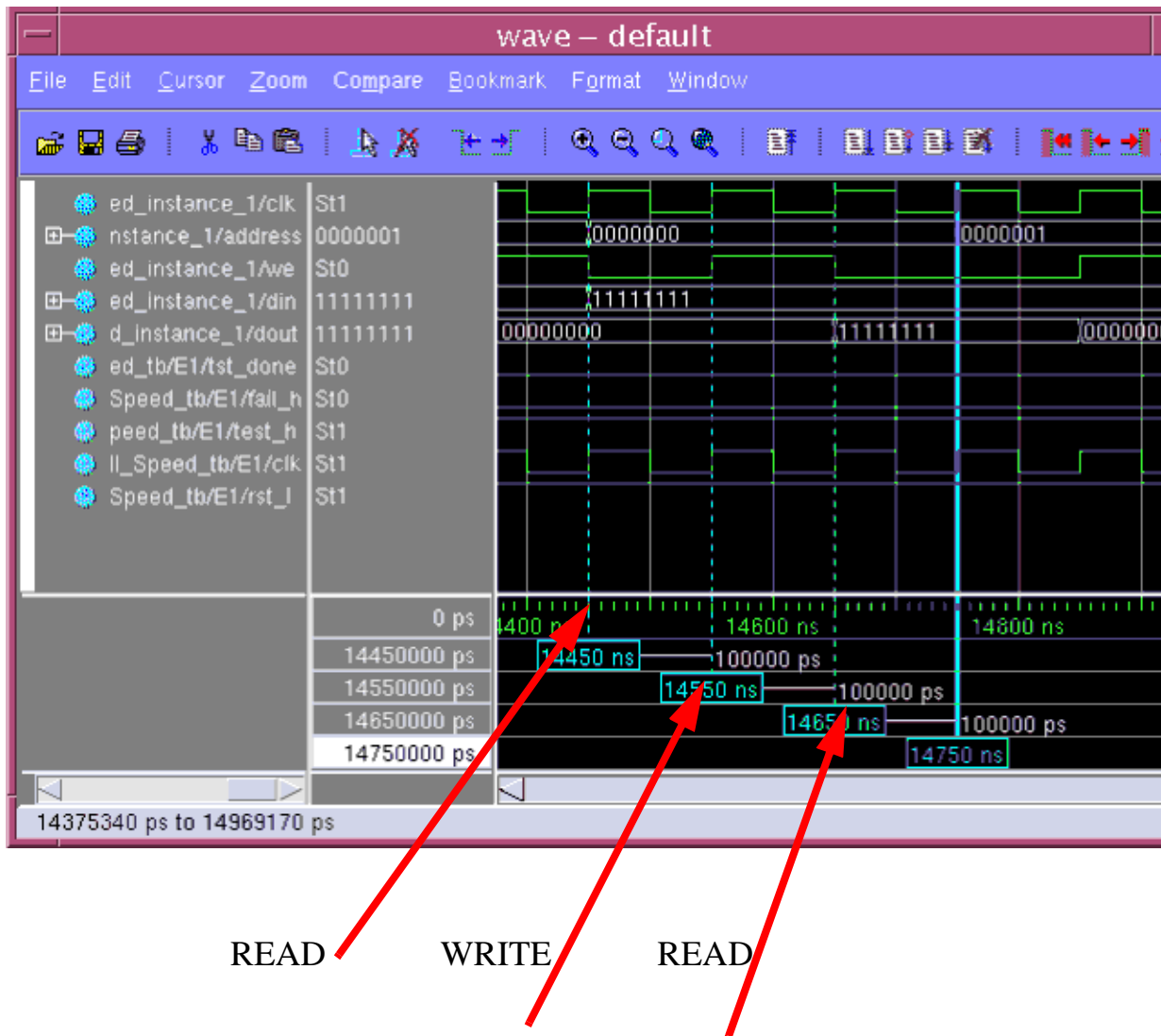
11. Next, we will verify the BIST logic works properly. Use the BIST Controller you just generated with the Verilog model of the memory and resimulate to see if everything works.

- a. Open up another shell window

```
shell> cd $MBISTNWP/mbist/full_speed/results  
shell> runsim
```

- b. A window displays with the message “*Are you sure you want to finish?*” Click **No**.

## Common BIST Variations



c. Review the **wave - default** window pane.

Separated by time cursors are 3 cycles of interest for the first back-to-back RWR operation for address 0.

## Extra Credit

To do additional speed comparisons go through steps 1 - 11 again but this time select the At\_speed model in steps 5b and 5c. Skip step 6.

- Compare the verilog of the \*\_bist.v files. What are the differences and why?
- Compare the Verilog expected data and the number of cycles in the 2 testbench file. What are the differences and why?

## Test Your Knowledge

- Why do you need pipelining stages to test your memory at Full-Speed?
- What two things do you need to change to accomplish Full-Speed Memory BIST?
- What are the advantages and disadvantages of doing Full-Speed memory BIST?

## Lab Summary

You should now be able to take a memory that can perform back-to-back read/write cycle and generate a memory BIST circuit to do Full Speed testing of that memory.



### Exercise 8: Adding BIST for Bidirectional Memories

This exercise should take approximately 10 minutes to complete.

This exercise demonstrates BIST insertion for a RAM with a bidirectional data bus. In this exercise, you will duplicate the default run you performed in another exercise, then examine the generated outputs to understand the circuitry that MBISTArchitect creates.

1. Change to the `$MBISTNWP/mbist2/bram4x4/design` directory.
2. Examine the model defined in `bram4x4.atpg`. Notice the `data_inout` statement declaration for the bidirectional data bus “dio”.
3. Change to the `../results` directory.
4. Invoke MBISTArchitect.
5. Load the `../design/bram4x4.atpg` library.
6. Add the `bram4x4` model to the list of memory models for BIST insertion.
7. Add default BIST circuitry to this model.
8. Save the default outputs, with the default names.
9. Exit the tool.
10. Compile the outputs and the memory model (`../designs/bram4x4.v`) then simulate the testbench. Create a new work directory called *work* for the compilation and simulation results. You can use the `../design/vsim_setup.do` file to setup the simulation and format the transcript. Name the trace file *trace.log*. If you need assistance with this process, refer back to [Exercise 2: Verifying the BIST Circuitry](#).

## Exercise 9: Adding BIST for ROMs

This exercise should take approximately 30 minutes to complete.

In this exercise, you will use a dofile to add BIST circuitry to test a ROM. This exercise uses a ROM64x16 model.

1. Change to the `$MBISTNWP/mbist2/rom64x16/results` directory.
2. Look at the following dofile:

```
shell> more ../design/rom64x16.do
```

The contents should appear as follows:

```
load library ../design/rom64x16.lib
add memory models rom64x16
set obs s -compress
set mb com -low 32
run
report bist
save bist -r
exit
```

This dofile sets up for ROM BIST circuitry generation, runs the insertion, and saves the default outputs with the default names to the current directory.

The ROM BIST insertion process is very automated. MBISTArchitect recognizes memory models without defined write cycles as ROMs. When you add a ROM model during a session, MBISTArchitect automatically sets the algorithm type to ROM. An architecture with a compressor, not a comparator, supports ROM testing. Thus, you must specify that the BIST controller *not* contain a comparator. You then additionally specify for MBISTArchitect to generate a compressor using the Setup Mbist Compressor command.

3. Invoke MBISTArchitect without the GUI using a dofile:

```
shell> mbistarchitect -dofile ../design/rom64x16.do \
        -nogui
```

4. Examine each of the generated files in the current (*rom64x16/results*) directory.

```
rom64x16_bist.v      rom64x16_bist_con.v
Compressor_lib.v     rom64x16_tb.v
```

5. Run the simulation using the following script and examine the results.

```
shell> runsim
```

Answer “No” to the question about finishing.



---

# Module 4

## Memory BIST-In-Place

This module will give you a basic understanding of how to create, connect, and integrate BIST structures using the Memory BIST-In-Place tool. The lab exercises at the end of this module will give you experience in running through the process flow of Memory BIST-In-Place.

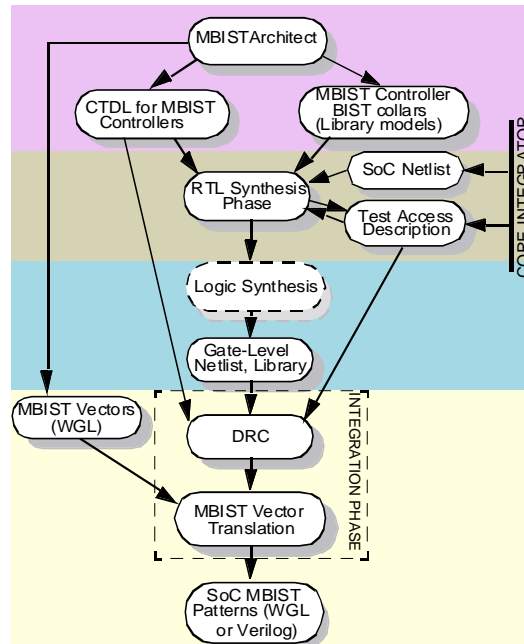
### Objectives

Upon completion of this module, you will be able to:

- Define the Memory BIST-In-Place flow
- Launch the Memory BIST-In-Place tool
- Define the files used in the tool to create BIST structures

# Memory BIST-In-Place Flow

## Memory BIST-In-Place Flow

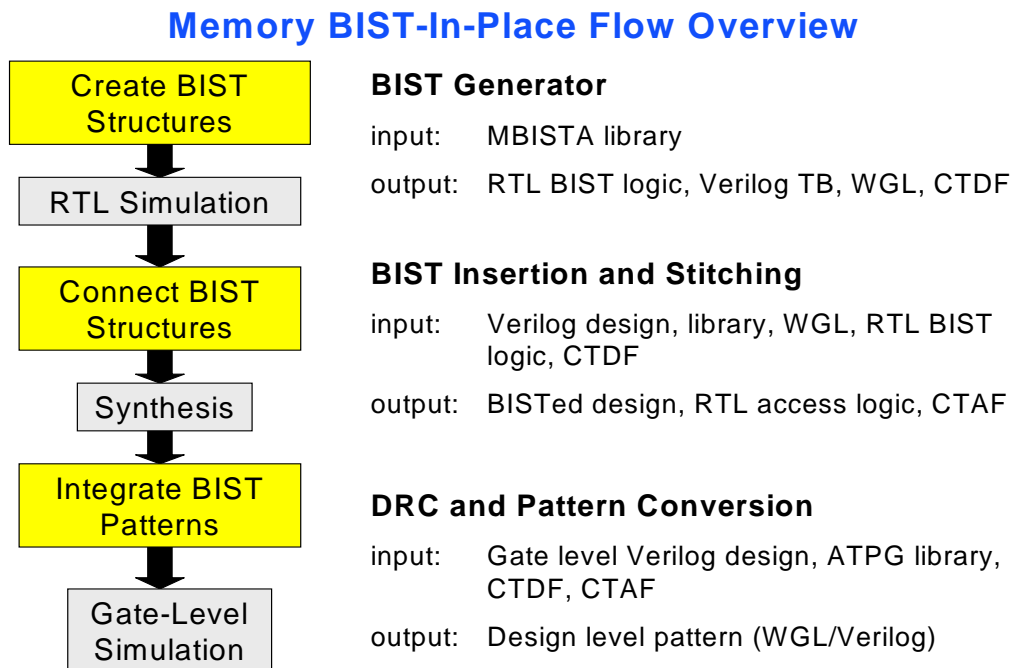


Memory BIST-In-Place automates the insertion of Memory BIST structures for embedded memory test in a System-on-a-Chip (SoC) design. This includes:

- the insertion of BIST collars around the original embedded memories
- the connection of memories to the inserted BIST controller(s)
- the synthesis of access structures in order for BIST controllers to be accessed from the SoC periphery
- pattern translation

The BIST controllers and memory collars are generated from the standalone Mentor Graphics Memory BIST tool, MBISTArchitect.

# Memory BIST-In-Place Flow Overview



4-3 • Memory BIST Training Workbook: MBIST In-Place

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- **Create BIST structures**  
Invoke the MBISTArchitect tool to generate RTL, BIST logic, a Verilog test bench, and WGL and Core Test Description Files.
- **RTL Simulation**  
Run the Mentor Graphics ModelSim tool to simulate the design. You can also run a gate-level simulation later in the process.
- **Connect BIST structures**  
Run Memory BIST-In-Place in the Synthesis mode to connect BIST structures and output a bisted design, RTL access logic, and Core Test Access Files.
- **Synthesis**  
Run the bisted design through a synthesis tool.

- **Integrate BIST patterns**  
Run Memory BIST-In-Place in the Integration mode to perform a design rules check and to generate patterns.
- **Gate-Level Simulation**  
Run the Mentor Graphics ModelSim tool to simulate the design.



# Creating BIST Structures

---

## Creating BIST Structures™

- ◆ **Uses MBISTArchitect**
- ◆ **Requires Memory BIST models for input**
- ◆ **Creates RTL BIST models**
  - BIST controller
  - BIST collar

## NOTES:

# Model Creation

## Creating BIST Structures Model Creation

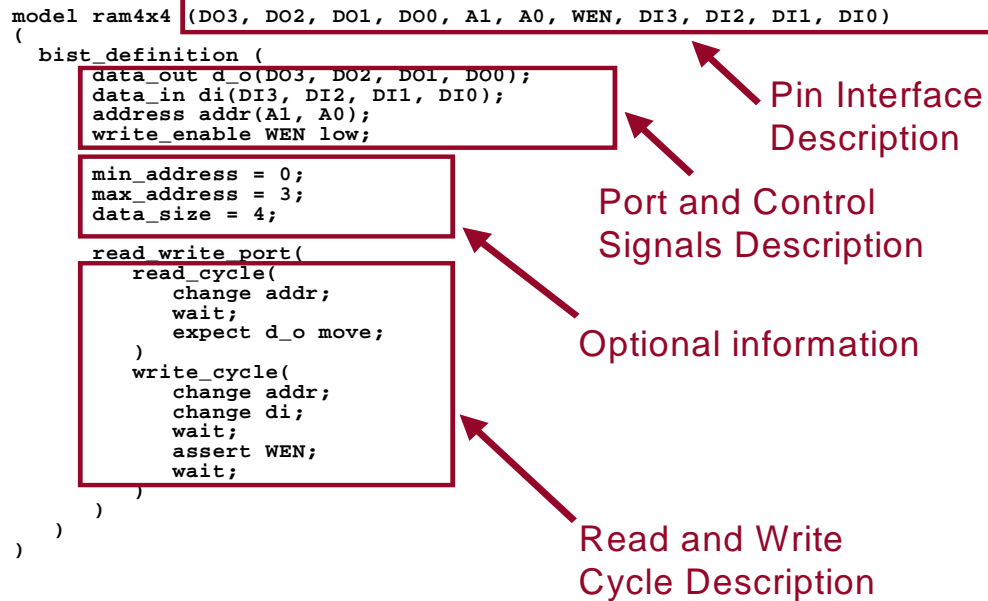
- ♦ **Model description includes:**
  - Pin interface
  - Read/write cycle description
- ♦ **You can create models:**
  - Manually, using basic syntax
  - Graphically, with the MBISTArchitect Model Editor



## NOTES:

# Memory Model Example

## Creating BIST Structures Memory Model Example

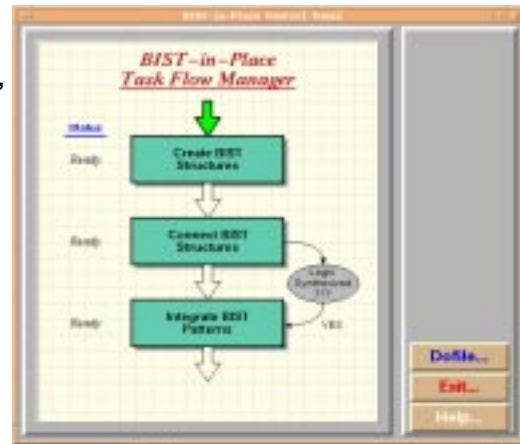


## NOTES:

# Creating BIST Structures Invocation

## Creating BIST Structures Invocation

- ♦ MBISTArchitect point tool invocation
  - `$ mbistarchitect -library lib_name -nogui`
- ♦ BIST-in-Place GUI invocation
  - `$ bistinplace`
  - Click on “Create BIST Structures” step in the Task Flow Manager



## NOTES:

# Basic Command Flow

---

## Creating BIST Structures Basic Command Flow

- ◆ Load Library <library name>
- ◆ Add Memory Model <model name...>
- ◆ Add Mbist Algorithm <port#> <algorithm>
- ◆ Set Bistinplace -on
- ◆ Run
- ◆ Save Bist
- ◆ Exit

## NOTES:

# Creating BIST Structures Results

---

## Creating BIST Structures Results

- ♦ RTL BIST logic (ramname\_bist.v)
  - BIST logic
  - Adds specparams to convey the info of connections between RAM and BIST to MBIP
- ♦ WGL file (ramname\_bist.wgl)
  - Used for pattern conversion in MBIP pattern integration step
- ♦ CTDF (ramname\_bist.v.ctdf)
  - Defines procedures to get in test mode and isolation mode

## NOTES:

## Example of RTL BIST Logic

### Creating BIST Structures

#### Example RTL BIST Logic

```
module ram8x4_multi_bist
specify
  specparam cti_cell_type$ram8x4_multi_bist =
    "mbist_controller";
  specparam bist_cycles$ram8x4_multi_bist =
    "466";
  specparam cti_connect$Test_addra_0 =
    "ram8x4_block_0/Test_addra_0";
  specparam cti_connect$Test_D03_3 =
    "ram8x4_multi_bist_ram4x4_block_3/D03_3";
  specparam cti_pin_type$test_h = "test_h";
  specparam cti_pin_type$clk = "clk";
  specparam cti_pin_type$rst_l = "rst_l";
  specparam cti_cell_type$ram8x4_block_0 =
    "mbist_memory:ram8x4_multi_bist";
endspecify
```

BIST Controller Name

BIST Cycles

BIST Controller & Collar Connection

BIST Control Signal Names

BIST Collar Name

### NOTES:

# Example WGL File

## Creating BIST Structures Example WGL File

```
waveform ram8x4_multi_bist
```

```
  signal
    test_h : input initialp[N];
    clk : input initialp[N];
    rst_l : input initialp[N];
    tst_done : output;
    fail_h : output;
  end;
```

Test Pin Interface

```
timeplate TP0 period 400ns
```

Timeplate definition

```
pattern bist_control ( test_h, clk, rst_l, tst_done, fail_h)
```

Pin Order List

## NOTES:



## Example WGL File (Continued)

---

### Creating BIST Structures Example WGL File

```
vector(+, TP0) := [ 1 1 1 X X ];  
vector(+, TP0) := [ 1 1 0 X X ];
```

BIST Initialization  
Pattern

```
loop 450  
    vector(+, TP0) := [ 1 1 1 0 0 ];  
end  
  
    loop 16  
        vector(+, TP0) := [ 1 1 1 X 0 ];  
        end  
        vector(+, TP0) := [ 1 1 1 1 0 ];  
    end  
end
```

BIST Test Pattern

### NOTES:

# Core Test Description File (CTDF)

Creating BIST Structures

## Core Test Description File (CTDF)

```
core ram8x4_multi_bist =  
  output Test_addr_3[1:0];  
  output Test_WEN_3;  
  output tst_done;  
  output fail_h;  
  input  
    Test_da_o3_0,Test_da_o2_0,Test_da_o1_0,Test_da_o0_0;  
  input test_h;  
  input clk;  
  input rst_l;  
  clock clk;  
  clock_lo rst_l;  
end;
```

Input & Output



## NOTES:

## Core Test Description File (Continued)

---

Creating BIST Structures  
**Core Test Description File (CTDF)**

```
core ram8x4_multi_bist =  
  procedure core_isolate =  
    core ram8x4_multi_bist;  
    timeplate tp1;  
    cycle =  
    hold clk 0;  
    hold test_h 1;  
    expect tst_done 0;  
    expect fail_h 0;  
  end;  
end;
```

Procedure to  
place BIST  
Controller in  
Isolation  
Mode



### NOTES:

# Core Test Description File (Continued)

## Creating BIST Structures

### Core Test Description File (CTDF)

```
core ram8x4_multi_bist =
```

```
  procedure core_test run_bist =
```

```
    core ram8x4_multi_bist;
```

```
    timeplate tp1;
```

```
    probe tst_done, fail_h, clk, rst_l;
```

```
    pattern_file ram8x4_multi_bist.wgl;
```

```
    cycle =
```

```
    hold test_h 1;
```

```
  end;
```

```
end;
```

Procedure to place  
BIST Controller into  
Test Mode

Pins to be monitored  
during BIST test

Test pattern file name

## NOTES:

# Connecting BIST Structures

---

## Connecting BIST Structures

- ◆ Uses Memory BIST-In-Place Synthesis mode
- ◆ Requires
  - RTL or gate-level design (VHDL or Verilog)
  - VHDL or Verilog library
  - BIST design objects created earlier in flow
  - CTDF created earlier in flow
- ◆ Inserts/connects BIST structures within hierarchy and to chip-level I/O

## NOTES:

# Connecting BIST Structures Invocation

## Connecting BIST Structures Invocation

- ♦ Can invoke with GUI or as command line only tool
  - `$ bistinplace design_name -verilog -lverilog verilog_library_name -synthesis -gui|nogui`



## NOTES:

# Example Command Flow (Setup)

---

## Connecting BIST Structures Example Command Flow (Setup)

- ◆ Load RTL bist logic
  - load design object ramname\_bist.v
- ◆ Load CTDF file
  - load core description file ramname\_bist.v.ctdf
- ◆ Define clocks
  - add clock 0 clock\_name

## NOTES:

# Example Continued (Setup)

---

## Connecting BIST Structures Example Continued (Setup)

- ♦ Specify BIST controller location and RAM/RAM collar correspondence
  - add mbist controller <bist\_controller\_pathname>  
<bist\_controller\_module\_name> <memory\_path\_name>  
-Collar <memory\_collar\_module\_name>
- ♦ Switch to synthesis mode
  - set system mode synthesis

## NOTES:



# Example Command Flow (Synthesis)

---

## Connecting BIST Structures Example Command Flow (Synthesis)

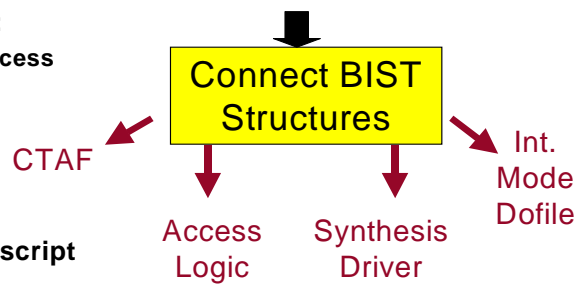
- ♦ **Run**
  - insert access logic
- ♦ **Write out RTL access logic and phase decoder**
  - save design file\_name -replace
- ♦ **Write out CTAF file**
  - save core access file\_name -replace
- ♦ **Write driver files for Design Compiler and Memory BIST-In-Place integration mode**
  - save driver files -logic\_synthesis file\_name -integration file\_name
- ♦ **Exit**
  - exit

## NOTES:

# Connecting BIST Structures Results

## Connecting BIST Structures Results

- ♦ **Core Test Access File (CTAF)**
  - Contains mapping information between BIST controller and design pins
  - Recommended file naming:
    - <design\_name>.ctaf or .access
- ♦ **RTL access logic and phase decoder**
- ♦ **Driver files for downstream tools**
  - Design compiler synthesis script
  - MBIP integration mode



## NOTES:

# Connecting BIST Structures Dofile

---

## Connecting BIST Structures Example Dofile

```
load design objects ram8x4_bist.v
add mbist controller mbistc ram8x4_bist
    /U1/mem_a -collar ram8x4_block_0
load core description ram8x4_bist.v.ctdf
set system mode synthesis
insert access logic
save design core1_rtl.v -replace
save access file core1_rtl.access -replace
save driver files -logic_synthesis dc.do
-bsda bsda.do -replace
exit
```

## NOTES:

# Example CTAF File

## Connecting BIST Structures Example CTAF File

```

core_instance /core_b/mbistc =
  core ram8x4_multi_bist;
  map clk = clkp , rst_1 = rstp ,
      tst_done = cl_ap;
  map fail_h = cl_bp;
end;
procedure core_access =
  timeplate gen_tp2 ;
  core_instance /core_b/mbistc ;
  cycle =
    force clkp 0 ;
    force core_addr_0 1 ;
    force core_addr_1 0 ;
    force cti_core_test_mode 1 ;
    force rstp 1 ;
  end;

```

Annotations:

- BIST Controller Name (points to `core ram8x4_multi_bist;`)
- BIST Controller to Chip Pin Mapping (points to the `map` block)
- Timeplate name (points to `timeplate gen_tp2 ;`)
- BIST Controller Instance Name (points to `core_instance /core_b/mbistc ;`)
- Procedure to Activate a Test Path (points to the `cycle =` block)

## NOTES:

## Example RTL Phase Decoder

### Connecting BIST Structures Example RTL Phase Decoder

```
module mbip_decoder
  assign core_addr_en[0]= core_addr_0;
  assign core_addr_en[1]= core_addr_1;
  always @ (core_addr_en)
  begin : cnt_shf
    case (core_addr_en)
      2'b00:
        begin
          core_select_0 = 1'b1;
          core_select_1 = 1'b0;
          core_select_2 = 1'b0;
        end
      2'b10:
        begin
          core_select_0 = 1'b0;
          core_select_1 = 1'b1;
          core_select_2 = 1'b0;
        end
    endcase
  end
endmodule
```

Core\_address signal  
chooses which BIST  
controller is tested

Core\_address "00"  
assigned to isolate  
all BIST controllers

Core\_address "10"  
assigned to activate  
BIST controller #1

### NOTES:

# Integrating BIST Patterns

---

## Integrating BIST Patterns

- ♦ **Uses Memory BIST-In-Place Integration mode**
- ♦ **Requires**
  - Gate-level netlist
  - ATPG library
- ♦ **Creates design level test vector running the BIST process**
  - Verilog and WGL

## NOTES:

# Integrating BIST Patterns Invocation

---

## Integrating BIST Patterns Invocation

- ♦ Can invoke with GUI or as command line only tool
  - `$ bistinplace design_name -verilog -library atpg_library_name -integration`



## NOTES:

# Integrating BIST Patterns Commands

---

## Integrating BIST Patterns Example Command Flow (Setup)

- ♦ **Load CTDF file**
  - load core description ramname\_bist.v.ctdf
- ♦ **Load CTAF file**
  - load core access design\_name.ctaf
- ♦ **Define clocks**
  - add clocks 0 clock\_name
- ♦ **Switch to integration mode**
  - set system mode integration

## NOTES:



# Continued Example (Integration)

---

## Integrating BIST Patterns Continued Example (Integration)

- ♦ Specify BIST controller name(s) for pattern conversion
  - add pattern translation -all
- ♦ Run
  - run
- ♦ Write out chip-level test patterns
  - save pattern file\_name [-verilog|-wgl] -replace

## NOTES:

# Integrating BIST Patterns Dofile

---

## Integrating BIST Patterns Example Dofile

```
load core description ram8x4_bist.v.ctdf
load core access my_design.ctaf
add clocks 0 clock_clk1
add clocks 1 reset_rst0
set system mode integration
add pattern translation -all
run
save patterns my_pats -verilog -replace
save patterns my_pats.wgl -wgl -replace
exit
```

### NOTES:

# Integrating BIST Patterns Results

## Integrating BIST Patterns Results

- ♦ Verilog pattern for simulation
- ♦ WGL pattern for tester

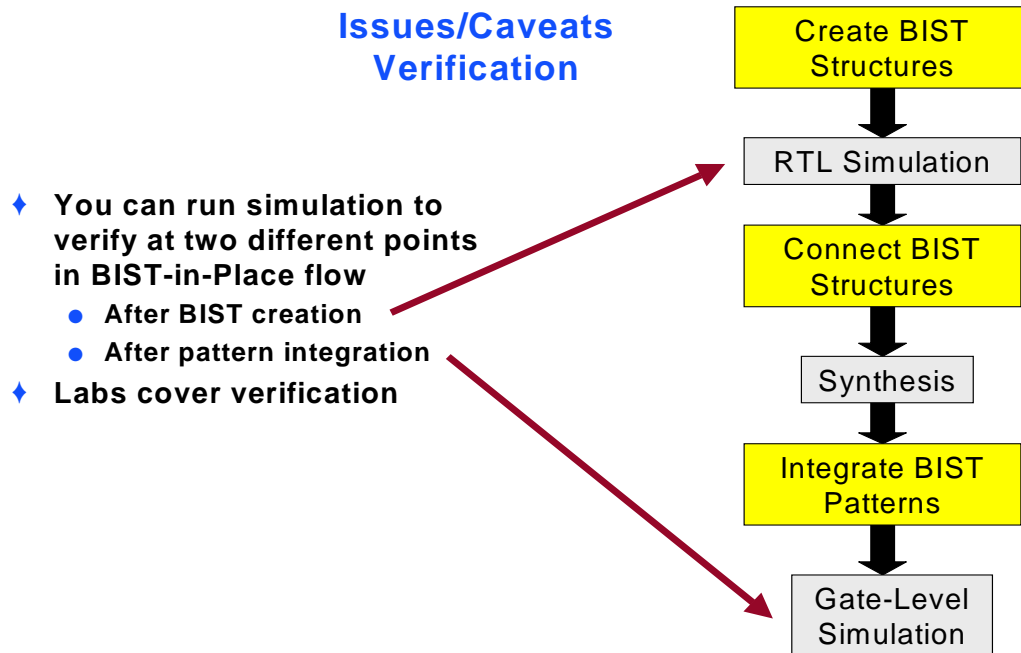


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## NOTES:

# Verification



## NOTES:

# I/O Pads

---

### Issues/Caveats I/O Pads

- ◆ Designs with I/O pads attributes need to be added in Verilog library

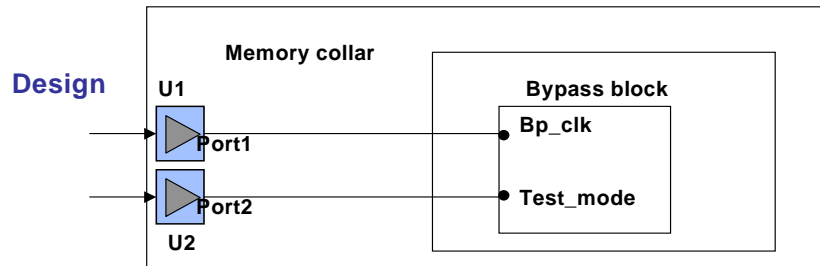
```
module iopad1 (pad, cin, i, oen) ;
  inout pad;
  output cin;
  input i, oen;
  bufif0      U1 (pad,i,oen);
  buf         U2 (cin,pad);
  specify
    specparam cti_cell_type$iopad1 = "io_pad_bidi" ;
    specparam cti_pin_type$pad = "io_pin" ;
    specparam cti_pin_type$cin = "data_in" ;
    specparam cti_pin_type$i = "data_out" ;
    specparam cti_pin_type$oen = "output_enable_n" ;
    //specparam cti_pin_type$oen = "output_enable";
    //if enable is active high.
  endspecify
endmodule
```

## NOTES:

# Global Signal Connections

## Global Signal Connections

- ◆ MBISTArchitect lets you make a connection for bypass logic
- ◆ Use the Set Global Pin command to specify the global pins for pin types `clock_bypass` and `control_bypass`
- ◆ For example, use this command in MBIP synthesis mode:
  - SET Global Pin -clock\_bypass U1/port1
  - SET Global Pin -control\_bypass U2/ port2



## NOTES:

# BSDArchitect/ Memory BIST-In-Place Integration

---

### BSDArchitect / Memory BIST-In-Place Integration

- ◆ **Memory BIST-In-Place creates a BSDA dofile**
- ◆ **BSDA reserves an instruction register and creates a data register for memory BIST**
- ◆ **BSDA uses the information in the dofile to run memory BIST**
- ◆ **Example dofile:**

```
save driver file -bsda bsda.do -inst mbist -reg mbist_reg -op 0011
add external register mbist_reg 2
add bscan instr mbist -reg mbist_reg -code 0011
add port connection clk buf TCK
add port connection cti_core_test_mode buf mbist
add port connection rst_1 mbist nand update_dr
add nontop port core_addr_0 core_addr_1
set testbench para -tck 200
set external_register interface mbist_reg \
-capture out1 out2 -update core_addr_0 acore_addr_1
set mbist interface -instr mbist -shift_in 10 10 \
-shift_out xx 10 -cycle 0 156
run
save bscan -r
```

## NOTES:





## Module 4: Lab Exercises

The following exercises take you through the Memory BIST-In-Place process flow illustrated in this lesson.

**Exercise 10: Setting Up MBISTArchitect Outputs** —You will create a memory BIST structure using MBISTArchitect, then generate the files needed for use with Memory BIST-In-Place.

**Exercise 11: Inserting BIST Controllers using Memory BIST-In-Place** —You will insert the BIST controller and synthesize the design.

**Exercise 12: Translating BIST Patterns to the SoC Level** —As the final step in the Memory BIST-In-Place process flow, you will translate the BIST patterns to the chip-level.

**Exercise 13: Full Flow Exercise** —You will run through the entire process again using a different design. To make things faster, you will run through various scripts which take you through the process.

## Exercise 10: Setting Up MBISTArchitect Outputs

The purpose of this exercise is to use MBISTArchitect to create the output files needed by Memory BIST-In-Place. You will generate a BIST structure for a design, *mbip.v*, that has three 4x4 RAMs and one 8x4 RAM.

Now that you are familiar with the MBISTArchitect GUI and its command line interface, we will invoke MBISTArchitect through Memory BIST-In-Place. The Memory BIST-In-Place GUI provides a task flow manager that makes creating BIST structures easier. For every command that the flow guide executes, do the following:

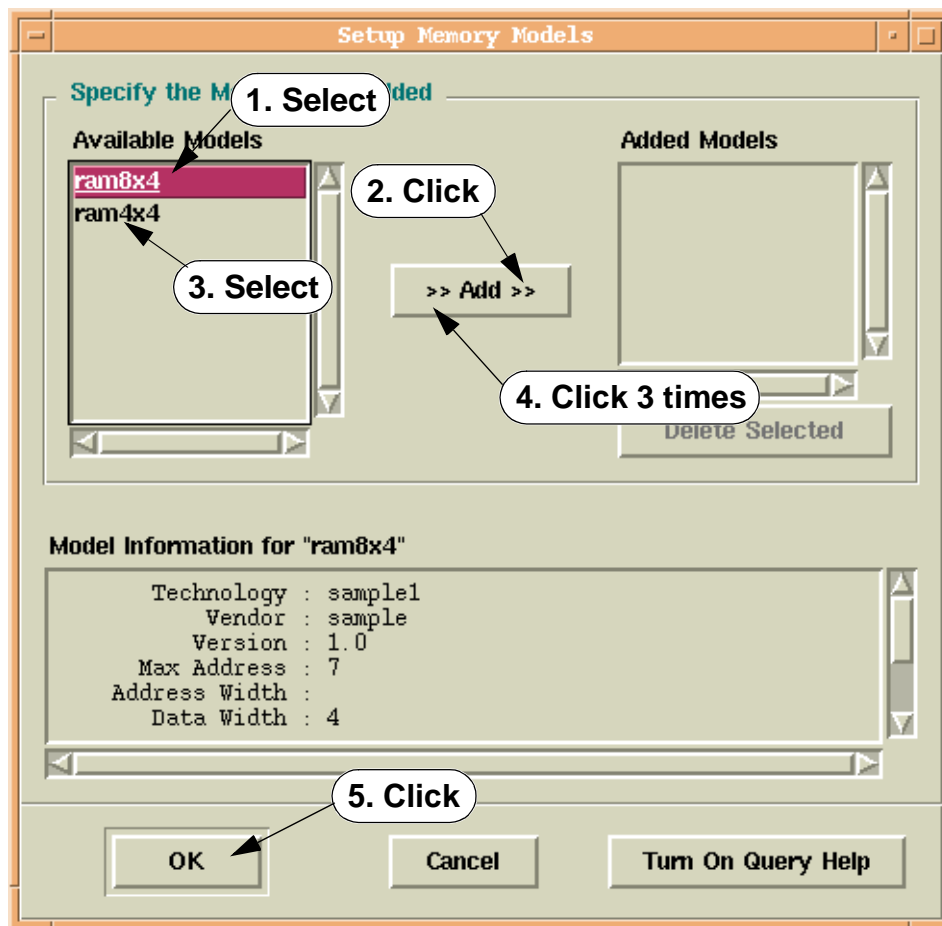
1. Change to the `$MBISTNWP/mbist4/ram8x4` directory.
2. Invoke Memory BIST-In-Place  

```
shell> $MGC_HOME/bin/bistinplace
```
3. From the GUI, click on **Create BIST Structures**.

The BIST Structures Creating Flow Guide opens to the first step.

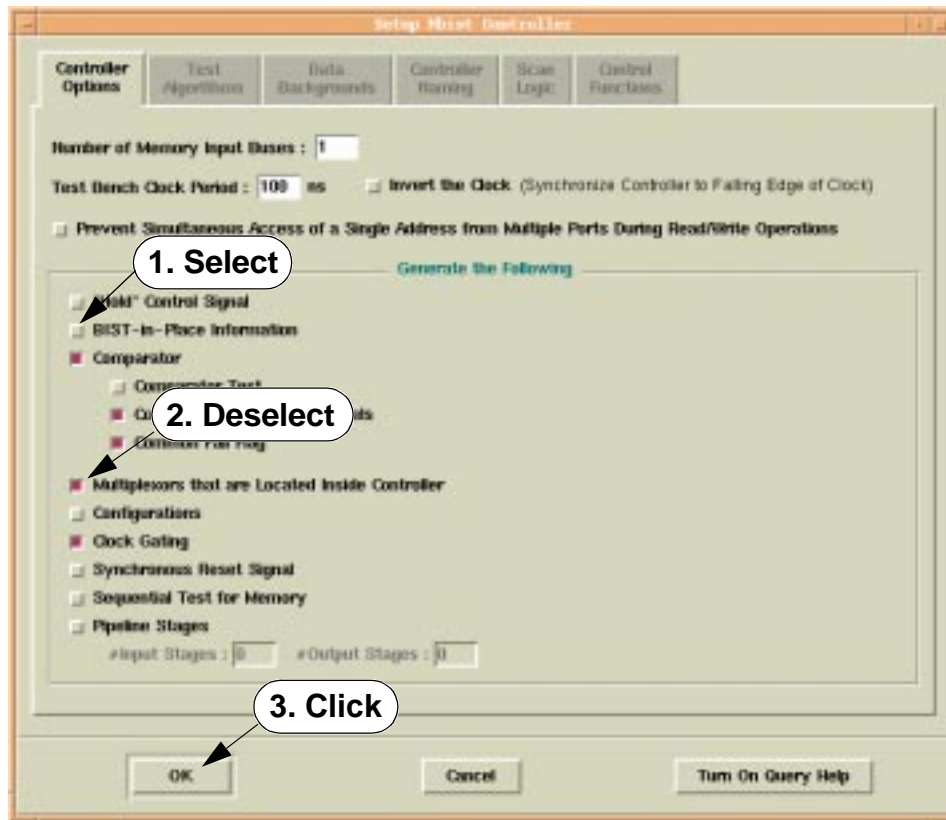


4. Load the *mbist.lib* design library. Click **Click Here to Set Up...** to set up the Load Libraries information. In the new window, select the *mbist.lib* library and click **Load**. Close this window.
5. Click **Next >>>** to move to the Add Memories step, then set up the Add Memories information as shown. You will be adding one 8x4 RAM and three 4x4 RAMs for BIST insertion. This shares multiple RAMs with one BIST controller.



6. Continue to the next step in the flow guide. Since you will be using the default algorithm setting, there is no need to set up any information for this step.
7. Continue to the next step in the flow guide. You are now at the Specify Controller Options step. For this exercise, you will create multiplexers

outside of the controller, in the BIST collar block, and create the necessary output files for use with Memory BIST-In-Place.



8. From this point on, you will want to use the default settings. Click **Next >>>** until you get to the Generate BIST Logic step.
9. Click **Next >>>** to run the BIST circuitry generation process.
10. Click **Next >>>** until you get to the Save Results step. Set this up to generate BIST-In-Place files, then click **OK**.
11. Click **Next >>>** twice and Close Flow Guide.
12. From the MBISTArchitect command line window, list the generated outputs.

**MBISTA> system ls ram8x4\***

Because you specified for the tool to save Memory BIST-In-Place information, MBISTArchitect generated a total of FIVE files. These files include:

- *ram8x4\_multi\_bist.v* — The RTL-level BIST logic.
- *ram8x4\_multi\_bist\_con.v* — The connection model for the controller and the RAM collar.
- *ram8x4\_multi\_tb.v* — The testbench that instantiates and tests the BIST model.
- *ram8x4\_multi\_bist.v.ctdf* — The CTDF file.
- *ram8x4\_multi\_bist.wgl* — The WGL pattern file.

13. Examine the generated outputs.
14. Exit MBISTArchitect.
15. Exit Memory BIST-In-Place.
16. Compile the model outputs and simulate the testbench to verify the BIST structure using the given script.

```
shell> runmsim
```

Answer “No” to the question about finishing.

## Exercise 11: Inserting BIST Controllers using Memory BIST-In-Place

In this exercise, you will be continuing through the design flow of Memory BIST-In-Place, building on the data created in the previous exercise. This exercise steps you through the process of inserting BIST controllers on the RTL level.

1. Ensure that you are still in the *\$MBISTNWP/mbist4/ram8x4* directory.
2. Invoke Memory BIST-In-Place in synthesis mode.

```
shell> $MGC_HOME/bin/bistinplace MBIP.v \
        -verilog -lverilog vlib -synthesis -nogui
```

Memory BIST-In-Place has two modes upon which you can invoke the tool: synthesis and integration. Here we invoked in synthesis mode in order to replace a RAM with the BISTed RAM generated in the previous exercise. This mode also creates access logic to a BIST controller and a connection to an SoC. All outputs in this mode are at the RTL level.

The `-lverilog` switch specifies the Verilog RAM library used in the design file.

3. Load the BISTed RAM information.

```
SETUP> load design objects ram8x4_multi_bist.v
```

4. Schedule the insertion of the BIST controller into the SoC design. Actual insertion does not take place until you transition the tool into the Synthesis mode.

```
SETUP> add mbist controller core_b/mbistc ram8x4_multi_bist \
        mem_a -c ram8x4_multi_bist_ram8x4_block_0 \
        core_b/mem_b -c ram8x4_multi_bist_ram4x4_block_1 \
        core_c/mem_c -c ram8x4_multi_bist_ram4x4_block_2 \
        core_c/core_e/mem_d -c ram8x4_multi_bist_ram4x4_block_3
```

This command places the BIST controller in `/core_b/mbistc`.

You can also run the *add\_mbist.do* dofile to keep from having to type the whole thing.

5. Load the core test description file.

```
SETUP> load core description ram8x4_multi_bist.v.ctdf
```

This file contains information on how to test and isolate a BIST controller.

6. Switch to Synthesis mode.

```
SETUP> set system mode synthesis
```

7. Insert access logic.

```
BISTINPLACE> insert access logic
```

The Insert Access Logic command initiates all the actions specified during the setup mode. These include:

- Replacement of memories by the BIST collar equivalents.
- Connection of the BIST collars to the BIST controllers.
- Insertion of the MUXes to provide access to the BIST controller from SOC pins.
- Insertion of logic to provide isolation conditions for the BIST controller.

8. Save the results.

- a. Save the RTL level access logic to core1\_rtl.v and modified SoC netlist to BIP\_cti.v

```
BISTINPLACE> save design core1_rtl.v -all -replace
```

- b. Save the CTAF file which includes information on how to access the BIST controller from the SoC level.

```
BISTINPLACE> save access file core1_rtl.access -replace
```

- c. Save the script files necessary for downstream tools.

```
BISTINPLACE> save driver files -logic_synthesis dc.do -include \
MBIP_cti.v -integration int.do -replace
```

The scripts saved are as follows:

- *dc.do* — Design Compiler script for the RTL access logic.
- *-include BIP\_cti.v* — Inserts an include statement in the SoC level netlist with the names of the synthesis-generated files.
- *int.do* — Script file for Memory BIST-In-Place integration mode.

9. Exit the tool.
10. At this point, you will synthesize the RTL design using a logic synthesis tool such as Design Compiler. Since we cannot run Design Compiler here, examine the *runDC* and *dc.do* scripts provided in this directory.



### Exercise 12: Translating BIST Patterns to the SoC Level

In this exercise, you will be continuing through the design flow of Memory BIST-In-Place, building on the data created in the previous exercise. This exercise steps you through the process of translating the BIST patterns to the SoC level.

1. Ensure that you are still in the *\$MBISTNWP/mbist4/ram8x4* directory.
2. Examine the dofile *int.do*. It should look something like this:

```
load core description ram8x4_multi_bist.v.ctdf
load core access core1_rtl.access
add clock 0 clkp
add clock 1 rstp
set gate report error
set gate level design
set drc hand c2 ignore
set system mode int
report cores
add pattern translation -all
run
save patterns mapped.v -verilog -replace
save patterns mapped.wgl -wgl -replace
exit -d
```

The first step is to load in the core test description file (*ram8x4\_multi\_bist.v.ctdf*), which describes how to get in to test mode and isolation mode of the BIST controller, and core access file (*core1\_rtl.access*), which describes the procedure for accessing the BIST controller for test purposes.

After defining the design's clocks (clock=clkp and reset=rstp), you set the system mode to cti or integration mode. This initiates a set of design rules checks. Then you tell the tool to translate all patterns (Note: If you have multiple memories and/or multiple controllers, you could translate patterns for only a subset of these), and then run, which creates chip-level vectors to control the BIST operation. You then save patterns in both Verilog (for simulation/verification) and WGL (for test program) formats.

3. Invoke Memory BIST-In-Place in integration mode. You will be invoking on the dofile you created in the previous exercises.

```
shell> $MGC_HOME/bin/bistinplace MBIP_cti.v -verilog \  
      -lib atpglib -integration -dof int.do -nogui
```

4. Scroll back through the transcript to see the results of the steps described previously.
5. Verify the chip-level BIST test patterns.

This step performs a final simulation of the chip-level BIST operation, simulating them to ensure there are no mismatches.

```
shell> runfinalsim
```

Answer “No” to the question about finishing. You should see the comment “No error between simulated and expected patterns.”

### Exercise 13: Full Flow Exercise

This exercise follows the same process flow as the previous exercises, but gives you the opportunity to work on a different design. This exercise demonstrates the use of MBISTArchitect for generating BISTed memory models. In addition, this exercise takes you through the process of inserting BISTed memories and connecting the BIST circuitry at the chip-level with Memory BIST-in-Place (an option to MBISTArchitect). This exercise goes through the entire chip-level memory BIST process.

1. Change to the `$MBISTNWP/mbist4/picdram/data` directory.
2. Invoke MBISTArchitect.

First you will invoke MBISTArchitect to generate a BIST structure for a RAM called *picdram* in the design *design\_noscan.v*.

```
shell> $MGC_HOME/bin/mbistarchitect
```

3. Load a design library and add memories.

Click on the Memory models block in the MBISTArchitect Control panel.

The MBISTArchitect library is located in `../libs/ram.atpg`. Click the **Browse...** button to find and select the appropriate library. Navigate up one level and into the *libs* directory. Select the *ram.atpg* library, click **OK** in the File Browser dialog and then click **Load**. You should see two models appear in the Available Models field.

The next step is to Add Memories. This means you are choosing the memory models you want to BIST from the library that you just loaded.

Select *picdram* from “Available Models” and click **>> Add >>**. You should see this model description listed under “Added Models.” If you click on *picdram*, you can view model information in the “Model Information” area.

Click **OK** to close the Setup Memory Models dialog box.

### 4. Specify algorithms.

Click on the line between the Controller and RAM blocks in the Control Panel. Here you can see the list of all available algorithms the tool supports. The March2 algorithm, which is the default (shown on the Controller block), is the algorithm we'll be using in this lab. It is already selected by default, so you can just **Cancel** out of this dialog.

### 5. Specify controller options.

You have a lot of flexibility in setting up the Memory controller. Click on the Controller block in the Control panel to see these options.

In this case, we want to put multiplexors under the Memory collar block (as opposed to putting them in the controller block), since there is only one memory being BISTed. Therefore, unselect the option to create a configuration that has Multiplexors Located Inside Controller.

We also want to turn clock gating off so make sure you unselect the option Clock Gating (the system clock is used for the memory).

Also, we want to insert BIST-in-place in the design, so check the option for BIST-in-Place information. Click **OK** when you are finished.

### 6. Generate BIST logic.

We are now ready to perform a Memory BIST generation. You need only click the **Run** button to generate the BIST logic. Notice how the compressor block disappears from the Control Panel. That is because we did not choose to use a compressor, but instead are using a comparator to determine whether the memory passes the BIST process.

### 7. Save the results.

The last step is to save all results. Click on the **Save BIST...** button. When the Save BIST Results dialog appears, check all the boxes to select all the options. Then click **OK**.

The tool writes out a total of six files, which you can see in the transcript area:

- *piccdram\_bist.v* — BIST Model
- *picdram\_bist\_con.v* — Connected Model (RAM collar and BIST controller)
- *picdram\_tb.v* — Test bench
- *picdram\_bist.v\_dcscript* — DC synthesis script
- *picdram\_bist.v.ctdf* — CTFD file
- *picdram\_bist.wgl* — WGL format pattern file

### 8. Exit MBISTArchitect.

You have just created BIST structures for your memory model - so you now have a “BISTed” memory. In other words, you have a memory model with a BIST collar, a BIST controller to control the BIST operation for this memory, as well as other files (testbench, core test definition, DC synthesis script, and WGL pattern file) that will be used downstream.

You are now ready to insert these BIST structures into the chip-level design. For this process we will run a series of scripts.

### 9. Verify the operation of the BISTed model.

You are now going to run a simulation of the RTL BIST model you created. To do this, execute the following command:

```
shell> runsim1
```

Answer “No” to the question about finishing.

This script compiles the BIST design objects and runs the generated testbench on the model.

Notice the march2 algorithm as its shown in the Wave window. Expand the Wave window. You can **Zoom > Zoom Full** to see the whole BIST process, or zoom into various parts by clicking your middle mouse button and drawing a box around a particular area. You may also need to expand the leftmost area where the signals are displayed to see their full names.

Basically what you are seeing is the test clock (clk), the reset signal (rst\_l), the test signal (test\_h), the test\_done signal (tst\_done), the fail flag (fail\_h), followed by the clock, address, we, din, and dout of the memory model. Notice the read/write operations and the address incrementing up and down the address space, as occurs during the march test. The tst\_done signal goes high when the BIST operation completes.

Use **File > Quit** from the ModelSim EE window to close Modelsim, and this time enter Yes, that you want to quit.

#### 10. Run BIST-in-Place synthesis.

The next step is to insert the BISTed memory and controller into the design. We will do this via a script that runs the Memory BIST-in-Place tool in synthesis mode. The end result is that we will have an RTL design that includes the inserted memory model with BIST collar, BIST controller, access logic, phase decoder, and all the appropriate connections. To perform this operation, execute the following command:

```
shell> runsyn
```

Scroll up through the transcript of BISTINPLACE. The main steps that were performed include copying the original design, inserting the controller, connecting the controller to the memory, and replacing the memory with the BISTed memory. The tool then creates access logic to the chip-level, mapping the controller I/O to chip-level pins. The tool then saves the design (mbip\_rtl.v) and access file (mbip.access).

The design file (mbip\_rtl.v) now needs to undergo synthesis, as the next phase of BIST-in-place, integration mode, requires a gate-level design. Integration mode also uses the access file (mbip.access) as described in the next step.

### 11. Run BIST-in-Place integration

In a normal design flow, you would synthesize the RTL design created during BIST-in-Place synthesis (mbip\_rtl.v) to gates. However, due to time constraints, we will use a design that has already been synthesized. Therefore, the final step is to perform rules checking on the gate-level design to ensure safe testing when the access path is sensitized and then create chip-level patterns to initiate the memory BIST operation. This is all done in the integration phase of BIST-in-Place.

To view the steps the tool will perform, view the integration script, runint. It should look as follows:

```
$MGC_HOME/bin/bistinplace design_noscan_cti.v -verilog \  
-lib ../libs/atpglib -int -nogui <<!!  
load core description picdram_bist.v.ctdf  
load core access mbip.access1  
add cl 0 ramclk1  
add cl 1 clk2  
set sys m cti  
add pattern translation -all  
run  
save pattern mapped.v -verilog -r  
save pattern mapped.wgl -wgl -r  
!!
```

The first step is to invoke BIST-in-Place on the synthesized Verilog design (design\_noscan\_cti.v). You then load in the core test description file (picdram\_bist.v.ctdf), which describes how to get in to test mode and isolation mode of the BIST controller, and core access file (mbip.access1), which describes the procedure for accessing the BIST controller for test purposes.

After defining the design's clocks (clock=ramclk1 and reset=clk2), you set the system mode to cti or integration mode. This initiates a set of design rules checks. Then you tell the tool to translate all patterns (Note: If you have multiple memories and/or multiple controllers, you could translate patterns for only a subset of these), and then run, which creates chip-level vectors to control the BIST operation. You then save patterns in both Verilog (for simulation/verification) and WGL (for test program) formats.

To run the integration process, execute the script:

```
shell> runint
```

### 12. Verify the chip-level BIST test patterns.

This step performs a final simulation of the chip-level BIST operation, simulating them to ensure there are no mismatches.

To perform this checking, execute the following command:

```
shell> runsim2
```

You should see the comment “No error between simulated and expected patterns.”



---

# Module 5

## Memory Modeling for MBISTArchitect

This module gives you a basic understanding of how to create, load and verify MBISTArchitect memory models. The lab exercises at the end of this module also give you experience creating, verifying and troubleshooting a variety of memory model types.

### Objectives

Upon completion of this module, you will be able to:

- Define inputs and outputs.
- Understand clocking schemes.
- Understand memory models.
- Understand troubleshooting procedures.

# A Memory Model:

---

## A Memory Model:

- ♦ Is an abstract data model that defines the memory ports and the read/write protocol of each port
- ♦ Is the only “design” input to MBISTArchitect
- ♦ Is not a simulation model
  
- ♦ Uses a basic DFT library model description
- ♦ Adds its own constructs to support BIST insertion
- ♦ Ignores the constructs it does not need

The MBISTArchitect tool uses an abstract data model that defines the memory ports and read/write protocol of each port. This model adds its own constructs to support BIST insertion. The memory model is the only input to MBISTArchitect. See the next slide for an example of a memory model and a description of memory model syntax.

You can add or change memory models using the Memory Model Editor in the MBISTArchitect Control Panel. See [“Memory Model Editor” on page 5-4](#) for a sample of the Memory Model Editor.

# Memory Model Syntax

## Memory Model Syntax

```

model model_name(list_of_pins) {
    bist_definition (
        address <name> (list_of_pins);
        data_in <name> (list_of_pins);
        data_out <name> (list_of_pins);
        data_inout <name> (list_of_pins);
        clock <pin> <active_state>;
        write_enable <pin> <active_state>;
        read_enable <pin> <active_state>;
        output_enable <pin> <active_state>;
        chip_enable <pin> <active_state>;
        control <pin> <active_state>;
        dont_touch <name> <active_state> <dir>;

        tech = <tech_name>;
        vendor = <vendor_name>;
        version = "number";
        message = "message_text";
        address_size = <number>;
        min_address = <lowest address>;
        max_address = <highest address>;
        data_size = <data_bus_bits>;
        addr_inc = <number>;

        write_port (
            write_cycle ( ... )
        )
        read_port (
            read_cycle ( ... )
        )
    ) // end bist_definition
} // end model description

```

**Input/Output Definitions**

**Memory Identification**

**Memory Size Information**

**Read and Write Cycles**

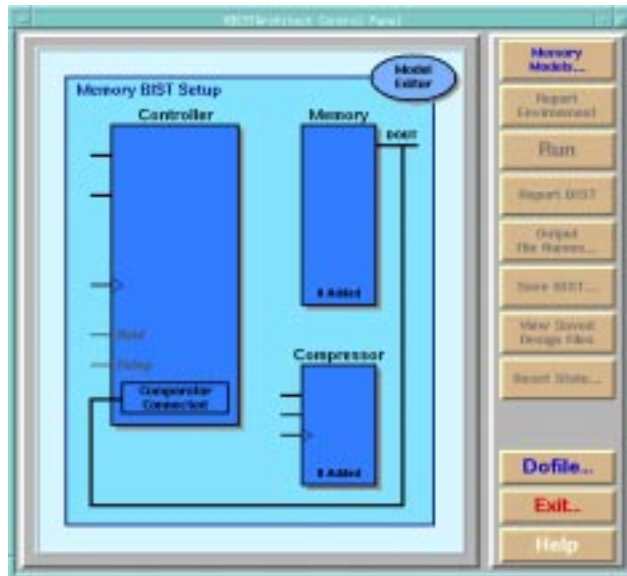
The MBISTArchitect tool shares the library format used by the DFT/ATPG tools FastScan, FlexTest, and DFTAdvisor. You need only to add the special “bist\_definition” section if you have an existing memory model in the DFT library format. MBISTArchitect does not use the gate-level simulation primitive information found in the primitive construct. The other DFT/ATPG tools use this information, but MBISTArchitect simply ignores it.

The term “pin” in this context refers to the individual inputs and outputs of the memory at the cell boundary. A pin can be defined as a scalar bit or an array. An array represents a bus and is sometimes referred to as a “wide pin”. The pin name must exactly match the port names specified in the associated Verilog or VHDL simulation model (both in name and case).

See [“Loading Library Files and Models” on page 5-5](#) for instructions on how to load the library file, add a memory model, and run MBISTArchitect to generate memory BIST logic.

# Memory Model Editor

## Memory Model Editor



You can add or change memory models using the Memory Model Editor in the MBISTArchitect Control Panel. For more information on how to use the Memory Model Editor, refer to “Using the Memory Model Editor” in Chapter 3 of the *MBISTArchitect Reference Manual*.

# Loading Library Files and Models

---

## Loading Library Files and Models

- ♦ Follow these steps to run MBISTArchitect, load libraries add memories, and generate :
- ♦ Launch MBISTArchitect
- ♦ Load a library
- ♦ Add a model
- ♦ Run MBISTArchitect
- ♦ Save the output and exit

Follow these steps to invoke, set up, and run the MBISTArchitect tool using the minimum set of commands needed to generate memory BIST logic.

### 1. Invoke MBISTArchitect.

To invoke MBISTArchitect, enter the following command at the shell:

```
shell> $MGC_HOME/bin/mbistarchitect
```

### 2. Load a Library.

After tool invocation, you must load a DFT library that contains the memory model(s) for which to add BIST logic. To load a DFT library interactively during the session, enter:

```
MBISTA> load library dft.lib
```

Where *dft.lib* is the name of the library. You can also load a library at invocation by using the -Lib switch.

### 3. Add a Memory Model.

The next step is to add a memory model from the loaded library to the BIST configuration. For example:

```
MBISTA> add memory models ram4x4
```

Where *ram4x4* is the name of the memory model for which you want to add BIST logic.

### 4. Run MBISTArchitect.

After you have loaded a library and added a memory model, you can run MBISTArchitect to generate default BIST logic:

```
MBISTA> run
```

### 5. Save the Output.

MBISTArchitect saves files in Verilog (default) or VHDL format. After memory BIST generation, you need to save the output:

```
MBISTA> save bist
```

### 6. To end an MBISTArchitect session, enter:

```
MBISTA> exit
```

# Defining Inputs/Outputs

## Defining Inputs/Outputs

<b>Address</b>	←	<b>Address bus</b>
<b>data_in</b>	←	<b>Data input bus</b>
<b>data_out</b>	←	<b>Data output bus</b>
<b>data_inout</b>	←	<b>Data bus (bidirectional)</b>
<b>Clock</b>	←	<b>Memory clock(s)</b>
<b>write_enable</b>	←	<b>Control signals</b>
<b>read_enable</b>	←	
<b>output_enable</b>	←	
<b>chip_enable</b>	←	
<b>control</b>	←	<b>Additional control signals if reserved keywords insufficient</b>
<b>dont_touch</b>	←	<b>Pins that are not controlled or observed by the BIST controller</b>

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**Defining Buses** You should define the address and data buses in the same manner as the simulation model for the memory. If a bus in your simulation model is declared as an array, then declare the same bus in your memory model header as an array. Consider the following memory model header segment:

```
model ram4x4 (A, DI, DO, WEN)
( bist_definition (
    address A (array = 1:0);
    data_in DI (array = 3:0);
    data_out DO (array = 3:0);
    ...

```

When MBISTArchitect reads this model, it assumes the address and data ports on the HDL model are declared as arrays and will use the STD\_LOGIC\_VECTOR as the data type when generating the matching bus in the BIST controller. You can change the signal type to STD\_CLOGIC\_VECTOR to specify it at the end of each statement.

Now consider the following memory model header segment:

```
model ram4x4 (DO3, DO2, DO1, DO0, A1, A0, WEN, DI3, DI2, DI1,
DI0)
( bist_definition (
    data_out d_o(DO3, DO2, DO1, DO0);
    data_in di(DI3, DI2, DI1, DI0);
    address addr(A1, A0);
    ...
```

Each bus element in the model header is declared as an individual scalar bit, the same as the simulation model. Notice that the `bist_definition` segment allows you to collect the individual bit under a single bus name and the ordering is significant. MBISTArchitect assumes that the bit order is from most significant (MSB) to least significant (LSB). MBISTArchitect uses this pin ordering when it connects the BIST controller to the RAM model. Thus, mismatches between the specified library pin ordering and the HDL model pin ordering can result in an improperly-connected BIST controller.

**Memory Clocks** You can define one or more memory clocks for synchronous memories.

**Control Signals** The active state can be either **high** (default) or **low**. During the read and write cycles, control signals always remain at the value *opposite* this state except when explicitly asserted. The following example declares an active low write enable named “wrt”:

```
write_enable wrt low;
```

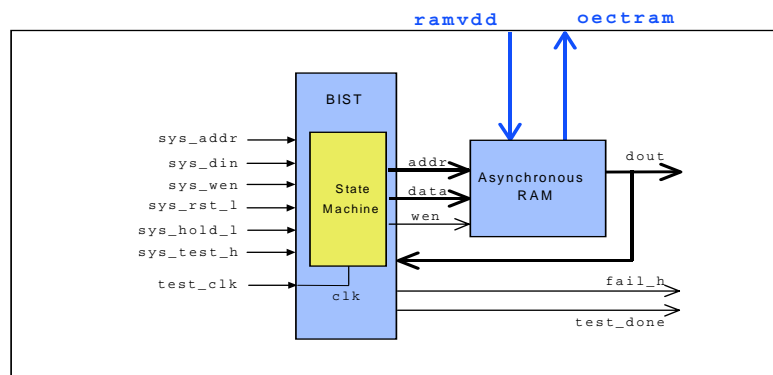
If the control signal operates a bidirectional data bus, the active state required to control tri-state buffers for the data bus follows the signal’s active state. You must specify either **tri\_l** or **tri\_h** to define this tri-state output buffer control state. When you define a model with a bidirectional data bus, you must specify a tri-state output control state for at least one of the defined control signals.



# The Dont\_touch Keyword

## The Dont\_touch Keyword

- ♦ **Memory pins not connected to the BIST controller**
  - **Examples: supply pins, status output pins**



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The `dont_touch` keyword allows you to specify pins that have no need to be controlled by the BIST controller. The syntax for specifying `dont_touch` ports is as follows:

```
dont_touch pin_name assert_state direction;
```

The *assert state* is either **high** (default) or **low** and defines the pin's active state. `Dont_touch` pins always remain at the value *opposite* their assert state. The *direction* is either **input** or **output**.

The default is input for all vector types except “data\_out” and “data\_inout.” The following example declares two `dont_touch` ports — an active low input port named “clr” and an active high output port named “refcntso”:

```
dont_touch clr low;
dont_touch refcntso out;
```

# Understanding Clocking Schemes

---

## Understanding Clocking Schemes

- ♦ **Asynchronous Memory**
  - No memory clock input
  - A change in inputs starts a read or write cycle
- ♦ **Gated Memory Clock**
  - In system mode, the memory clock connects through a mux to the system clock
  - In test mode, the memory clock connects to a controller-related clock signal
- ♦ **Non-Gated Memory Clock**
  - Memory clock (mem\_clk) connects to the system memory clock

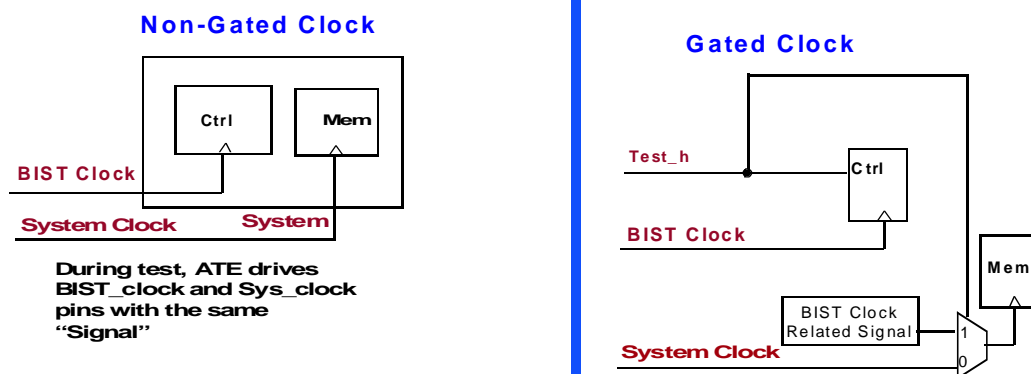
## The Primary Goal

Your job in creating a memory model is to define a read and write cycle that meets the minimum timing constraints as specified by the manufacturer, but at the same time runs at the fastest test speed. Introducing just one extra test clock cycle in a read operation, for example, can increase the total test time for a March2 algorithm by 50%.

There are different clocks that you need to reference when defining the read and write cycles for a memory model. This discussion refers to the “BIST clock” as the primary input clock to the BIST controller. This clock is named “clk” by default and is used to advance the BIST state machine to the next state. As the state machine enters each state, memory control signals are asserted or de-asserted and memory bus values can be changed. The term “memory clock” (mem\_clk) refers to the clock input to a synchronous memory. A memory can have one or more clock inputs. Asynchronous memories don’t have a clock input. During testing, a “test clock” is also generated as described next.

# Understanding Clocking Schemes

## Understanding Clocking Schemes (Continued)



**MBISTArchitect has a variety of clock connection options. Use these commands to control the clock connection:**

```
Setup Memory Clock [-System|-Test[Noinvert|Invert]]-Control]
Set controller clock [-positive|-negative]
```

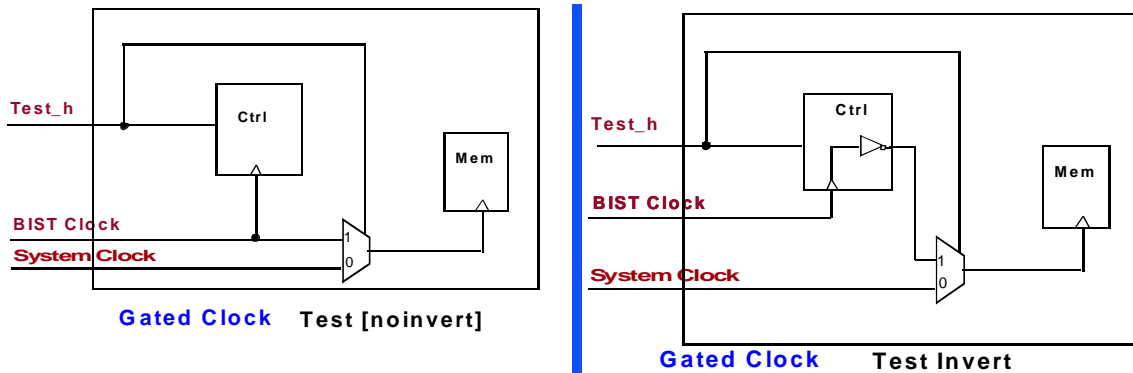
- **Non-Gated** (also referred to as System)—During test (and during actual system use), the memory clock is driven by a system clock.
- **Gated** —There is a MUX (gate) attached to the clock input of the memory. During system use, the MUX is set so the memory clock is driven by a system clock. During test mode, the MUX is set so that the memory is driven by a BIST related clock. There are three important variations of this described in the following clocking diagrams.

The key advantage of the Non-Gated approach is that it greatly simplifies getting clock timing correct for normal system use. Depending on the clock-tree generation process and the severity of the minimal skew requirements, using the Non-Gated approach can be almost mandatory. It is the default mode.

The major disadvantage to this approach is that the tester must be set up to drive the system clock input and the bist clock input with the same signal. There are potential skew issues with this due to tester limitations. However, these are often mitigated by the tester clock being much slower than the expected system clock.

# Clock Connections

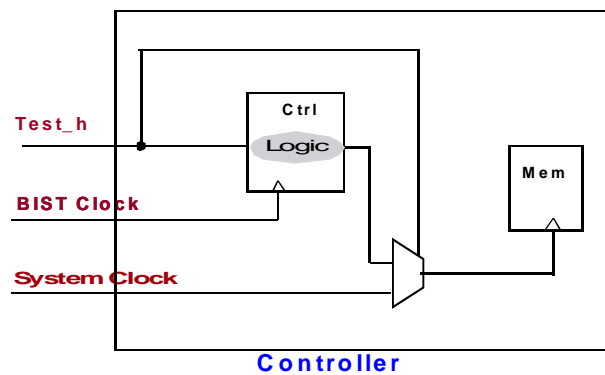
## Clock Connection



**MBISTArchitect has a variety of clock connection options. Use these commands to control the clock connection:**

```
Setup Memory Clock [-System|-Test[Noinvert|Invert]]|-Control]
Set controller clock [-positive|-negative]
```

## Clock Connection (Continued)



**Note:** Fastest possible clock out of ctrl is 1/2 rate of the BIST clock or even slower

**MBISTArchitect has a variety of clock connection options. Use these commands to control the clock connection:**

```
Setup Memory Clock [-System|-Test[Noinvert|Invert]]|-Control]
Set controller clock [-positive|-negative]
```

When Memory clocking is set to anything other than -System, there will be a MUX instantiated which will select between a system clock and a BIST-related signal. This MUX will be very obvious when you use commands that place MUX related RTL in the collar. Otherwise, it may be buried in the BIST controller RTL.

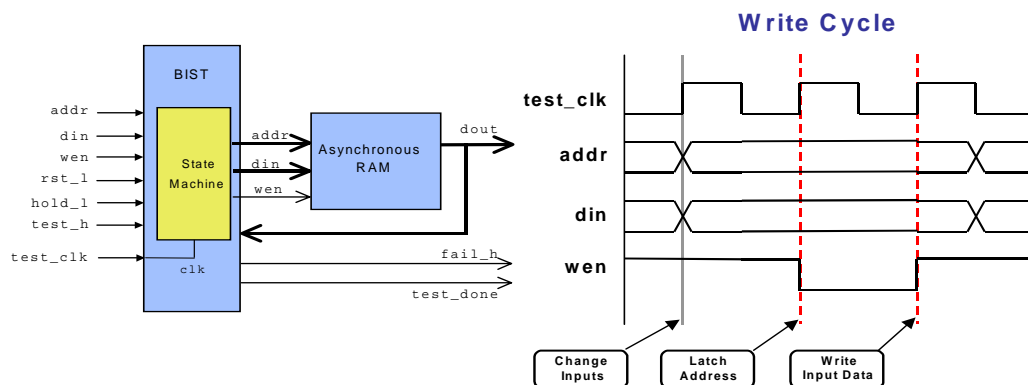
- When the SETup MEMory Clocking -Test command is specified, the BIST clock will be sent to the BIST controller and directly to the MUX.
- When the SETup MEMory Clocking -Test INVERT command is specified, the BIST clock will be routed to the BIST controller. It will be passed through an inverter and then routed to the MUX.
- When the SETup MEMory Clocking -Controller command is specified, the BIST controller will use internal logic to drive a signal that is to act as the clock. That signal will be routed to the MUX. The BIST controller itself is a synchronous, single-phase clock, design. So, it cannot change the state of this clock signal any faster than once per BIST clock. So, the clock it generates cannot be any faster than one-half the speed of the BIST clock (one clock for going high, one-clock for going low). Depending on the operation needed, the BIST controller may keep the output clock constant for several BIST clock cycles. Often, this is to calculate and set up conditions for a rising clock edge on the memory.

There is a separate command: SET COntroller Clock [-positive | negative] that can be used in conjunction with the clock connection command to deal with memories that lock their inputs on falling edges rather than rising edges. Also, the two commands can be used to effect half-cycle phase shift which can overcome timing violation issues. This is discussed later in this workbook.

# No Memory Clock

## No Memory Clock

- ◆ Asynchronous memory
- ◆ test\_clk drives the BIST state machine



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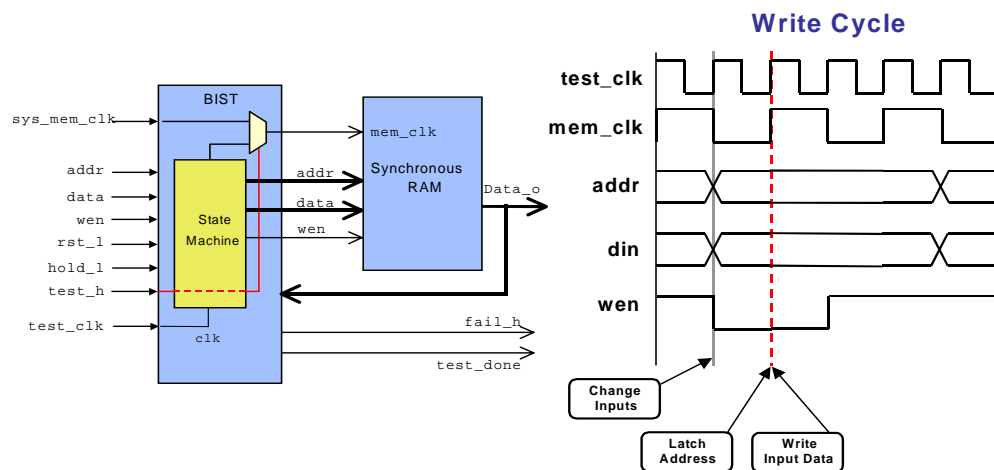
Asynchronous memories don't have a clock input, so a change in one or more of the inputs starts a read or write cycle. In this write cycle example, a change on the **addr** address bus starts the write cycle. After a minimum settling time, the **wen** write enable signal goes active low which causes the memory to latch the new address. Sometime before the **wen** signal goes inactive high, new data is placed on the **din** bus and allowed to settle. When **wen** goes inactive high, the data is written to memory and the write cycle ends.

# A Gated Memory Clock

## A Gated Memory Clock

While under test, **mem\_clk** is driven by the BIST controller.

Set Memory Clock -control



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If the memory model is synchronous, the default is -system. Notice in the illustration that a reference clock called **test\_clk** drives the BIST state machine. When **test\_h** goes active, the multiplexer prevents the **sys\_clk** from reaching the memory and the BIST state machine drives the **mem\_clk** input. The BIST controller drives this clock input as it would any other control signal. Notice that the **mem\_clk** frequency is half the **test\_clk** frequency (at best), and that the controller has total control over the memory clock.



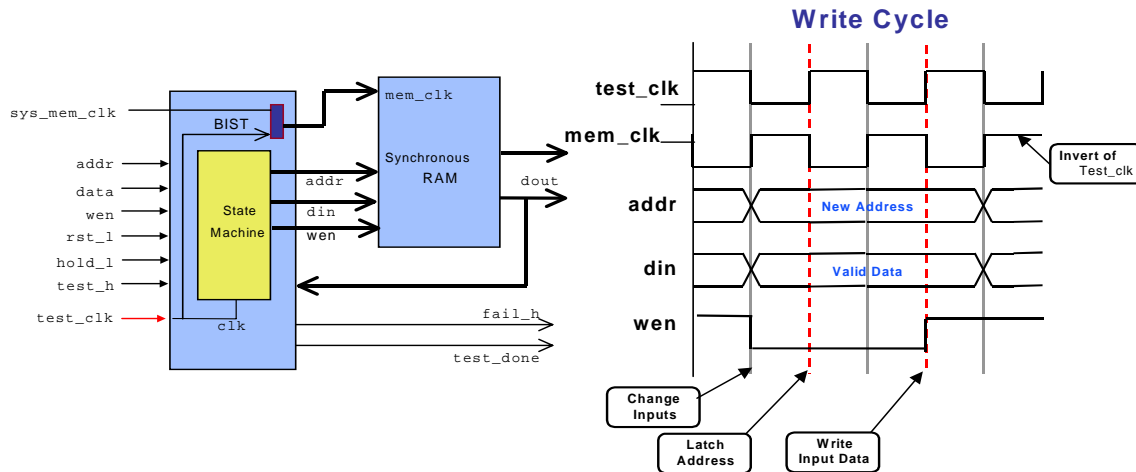


# An Inverted BIST Clock

## An Inverted BIST Clock

**Non-gated mem\_clk is 180 out of phase with test\_clk**

Setup Memory Clock -Test invert



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You can use a setup command in MBISTArchitect to tell the BIST state machine to respond to the falling edge of the clock. In this case, the falling edge of the clock input to the state machine causes the memory input buses and control lines to change. One half cycle later, the rising edge of the clock input to the synchronous memory captures the input data. This scheme reduces the write cycle from four cycles to two, and thus cuts the write cycle test time in half.

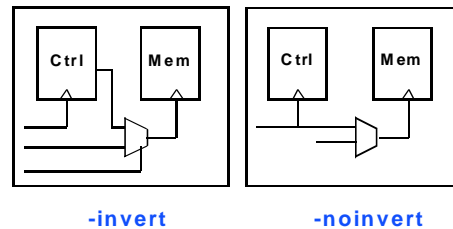
# Test Clock

## A Test Clock

**Memory clock is connected to a mux. In test mode, the clock is driven from a signal generated by the BIST controller.**

Setup Memory Clock -Test

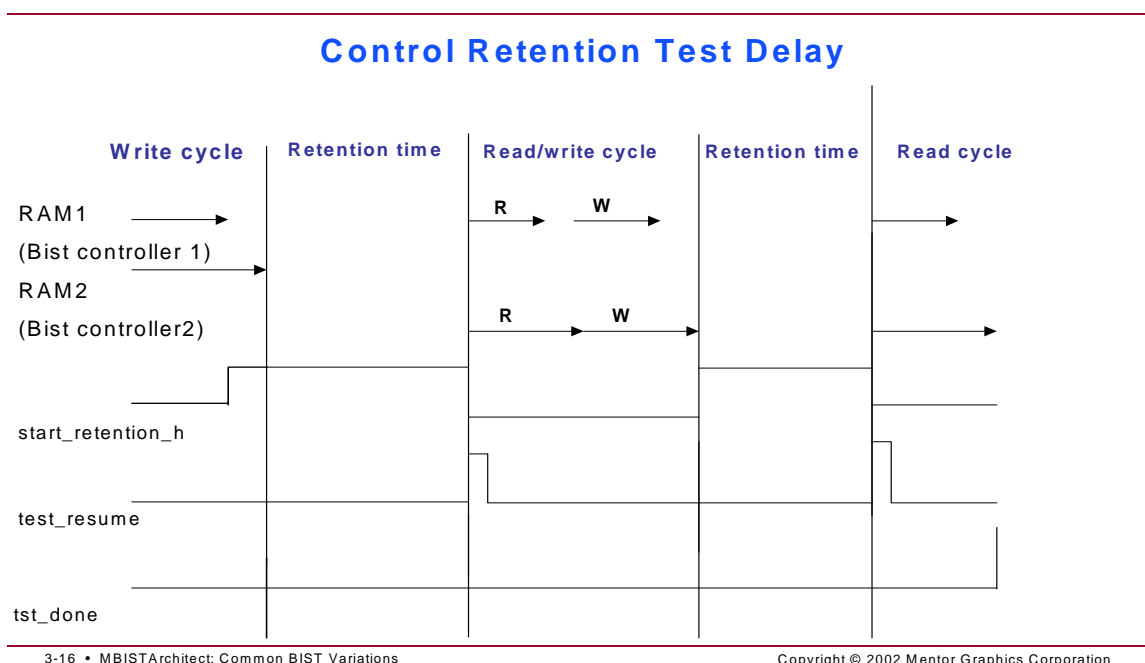
Test Clock



When using the test clock scheme, the memory clock connects to a mux. In system mode, the clock is driven from the system memory clock. In test mode, the clock is driven from a signal generated by the BIST controller. This signal is a reassignment of the BIST controller clock. The generated RTL will be modified for the controller assigned test clock scheme to include the controller assignment of the clock and the clock mux.

Two types of test clock connection are supported, -noinvert and -invert.

## Control Retention Test Delay



MBISTArchitect lets you control the delay value used in the WGL and simulation test bench when waiting to assert the resume signal. This is used to continue the BIST session following a retention test synchronization delay.

You can specify the delay value, as a multiple of the number of controller clock cycles. The default value is 100 cycles. The report environment reports the delay value.

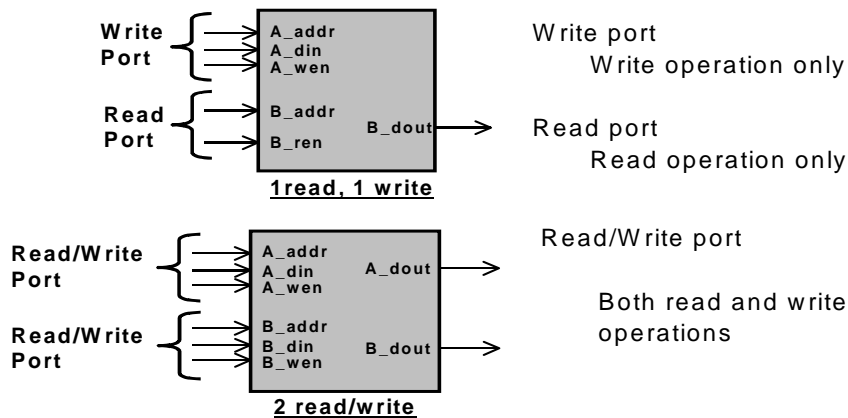
The diagnostics capability is added by using the Setup Retention Cycles command followed by a value defining the delay in cycles. For example, to set a delay value of 50 cycles, enter:

**setup retention cycles 50**

# Memory Ports

## Memory Ports

- ♦ **Memory ports define their read and write capability**
  - Can have any number of read ports, write ports or read/write ports



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A memory component can have any number of read ports, write ports, or read/write ports. The memory model syntax can match the port scheme of any memory component.

# Defining Memory Ports

---

### Defining Memory Ports

- ♦ Each unique port requires its own port definition
- ♦ Port definitions are not explicitly labeled
- ♦ MBISTArchitect identifies a port by the signals controlled within the read/write cycles

```
read_port (
  read_cycle (
    ....
  )
)
```

```
write_port (
  write_cycle (
    ....
  )
)
```

```
read_write_port (
  read_cycle (
    ....
  )
  write_cycle (
    ....
  )
)
```

- ♦ **Write Port**
  - Contains write cycle only
- ♦ **Read Port**
  - Contains read cycle only
- ♦ **Read/Write Port**
  - Contains both a read cycle and write cycle

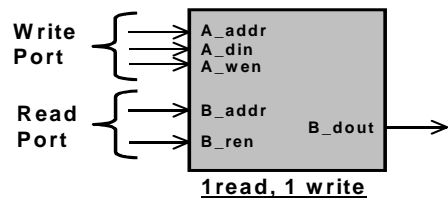
Each unique port requires its own port definition and the definitions are not explicitly labeled. MBISTArchitect identifies a port by the signals controlled within the read/write cycles. Only the write port is identified as a port for the add mbist algorithm command.

The first port that is defined within the bist\_definition is referred to as port #1 and the MBISTArchitect Model Editor will enter a comment identifying it as such. The second port defined in the model will be referred to as port #2, and so on.

# Port Definition Example 1

## Port Definition Example 1

♦ Example : 1 read, 1 write memory



```

write_port (
  write_cycle (
    change A_addr;
    change A_din;
    wait;
    assert A_wen;
    wait;
  )
)

read_port (
  read_cycle (
    change B_addr;
    wait;
    assert B_ren;
    wait;
    expect B_dout;
    wait;
  )
)

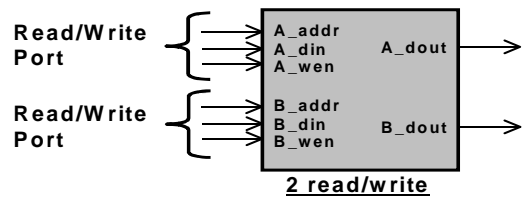
```

## NOTES:

## Port Definition Example 2

### Port Definition Example 2

♦ **Example : 2 read/write memory**

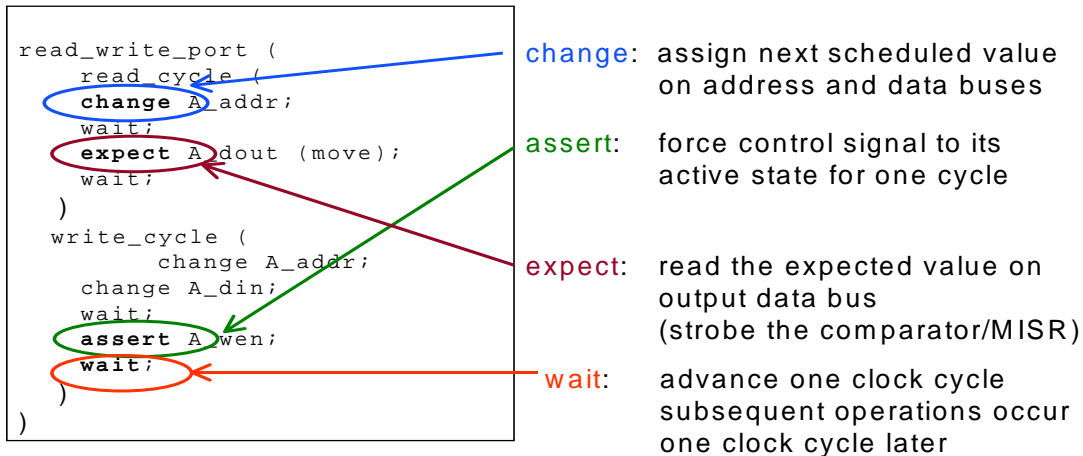


```
read_write_port (
  read_cycle (
    change A_addr;
    wait;
    expect A_dout;
    wait;
  )
  write_cycle (
    change A_addr;
    change A_din;
    wait;
    assert A_wen;
    wait;
  )
)
read_write_port (
  read_cycle (
    change B_addr;
    wait;
    expect B_dout;
    wait;
  )
  write_cycle (
    change B_addr;
    change B_din;
    wait;
    assert B_wen;
    wait;
  )
)
```

### NOTES:

# Read/Write Cycle Syntax

## Read/Write Cycle Syntax



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You use *event* statements to describe the action of the inputs and outputs during a read and write cycle. You use the **change** statement to assign the next scheduled value on the address bus and data buses.

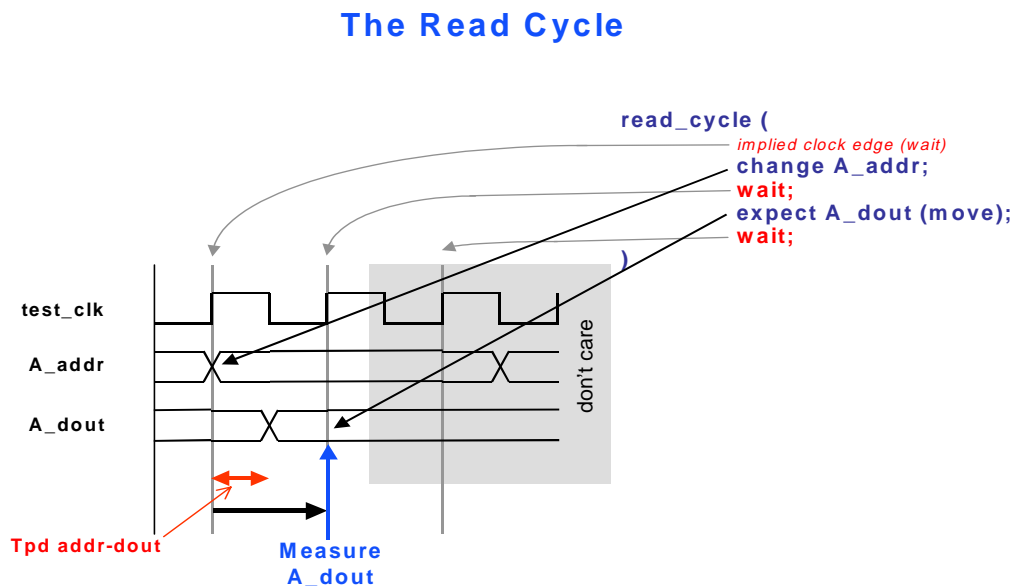
You use the **assert** statement to force a control signal to its active state during that test clock cycle. The control signal returns to its inactive state on the leading edge of the next test clock cycle, unless asserted again with another assert statement.

The **expect** statement tells the BIST controller that the data on the specified bus is valid, starting with the leading edge of that test clock cycle. This tells the BIST controller that it can read the data for use with the comparator or MISR.

Inserting a **wait** statement is like inserting the leading edge of the next test clock cycle (the clock that drives the BIST state machine). Any signals described in the change, assert, or expect statements that follow the wait statement become active or are valid on the leading edge of that clock cycle.



# The Read Cycle



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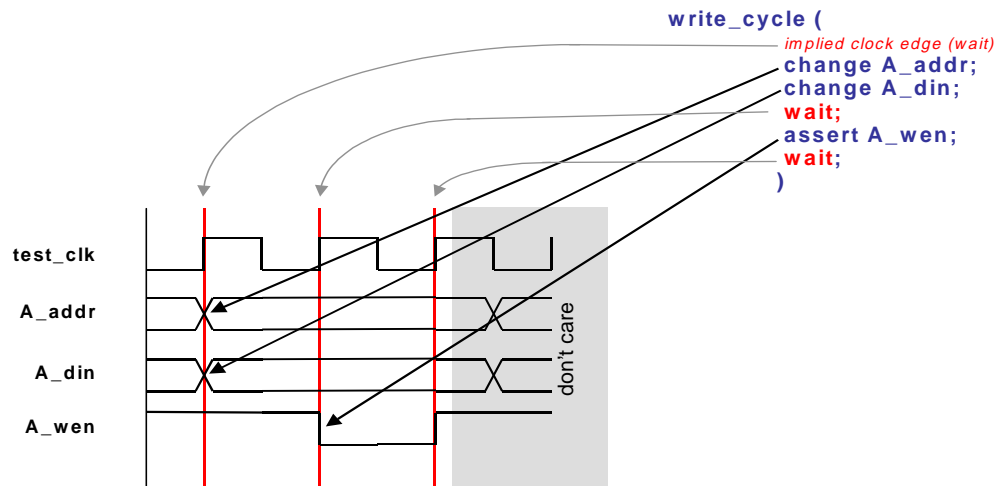
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A read or write cycle often starts with a memory input becoming active, such as an address bus or a chip enable line. This change occurs on the rising edge of the clock that is advancing the state of the BIST controller. You can also think of a wait statement as a rising edge of the reference clock and a read or write cycle as starting with an implied wait statement, even though the wait statement is not explicitly written into the model. You should place an explicit wait statement at the end of a read or write cycle to mark the end of the cycle.

In this example, the change in address marks the beginning of an asynchronous read cycle. After a specified period, the data appears on the output data bus. You can assume that this period is less than one test clock cycle. The next wait statement marks the next clock edge and the expect statement following that tells MBISTArchitect that the data is valid and it is okay to measure the output at that clock edge.

# The Write Cycle

## The Write Cycle



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In this asynchronous write cycle, the change in address occurs on the rising edge of the `test_clk` and marks the beginning of an asynchronous write cycle. The new input data is also changed at this time, although typically, the change could occur later in the cycle without violating the timing constraints. The next wait statement marks the next rising edge of the `test_clk`. `A_wen` goes active low, which latches the address into the memory. On the next clock edge (wait statement), `A_wen` is released because it is not explicitly activated in the memory model. This action writes the data to memory and ends the write cycle.

An **expect** statement can include an optional **move** modifier that specifies when an event executes. The **move** modifier means that the MBISTArchitect tool can move this event to a later clock cycle when optimizing the BIST structure. The **move** option usually applies to data outputs. The MBISTArchitect tool uses the **move** option only when it is trying to optimize circuitry while combining read and write cycles together to form read/write/read cycles or other large cycles.

# Interpreting Data Sheets

---

## Interpreting Datasheets

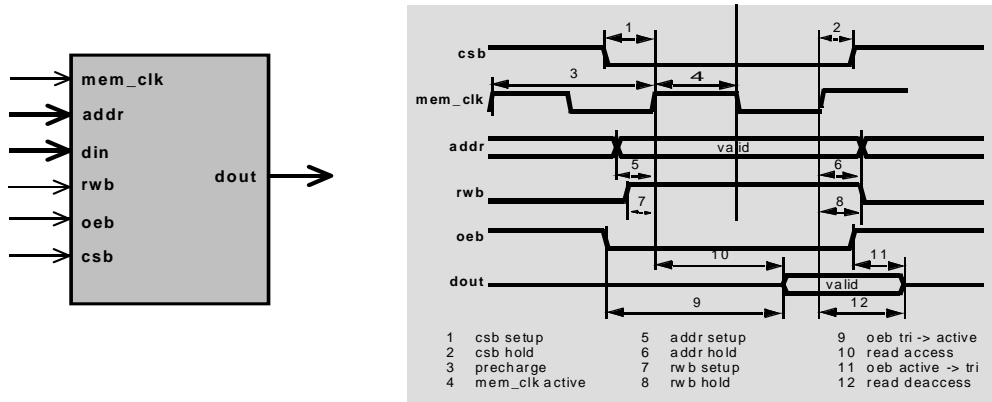
- ♦ Read and Write cycles can be determined from datasheets
- ♦ Dependent timing constraints are handled with “wait” statements
  - Setup and hold constraints
  - Sequential behavior

## NOTES:

# A Synchronous RAM Example

## A Synchronous RAM Example

### 1 read/write - synchronous RAM

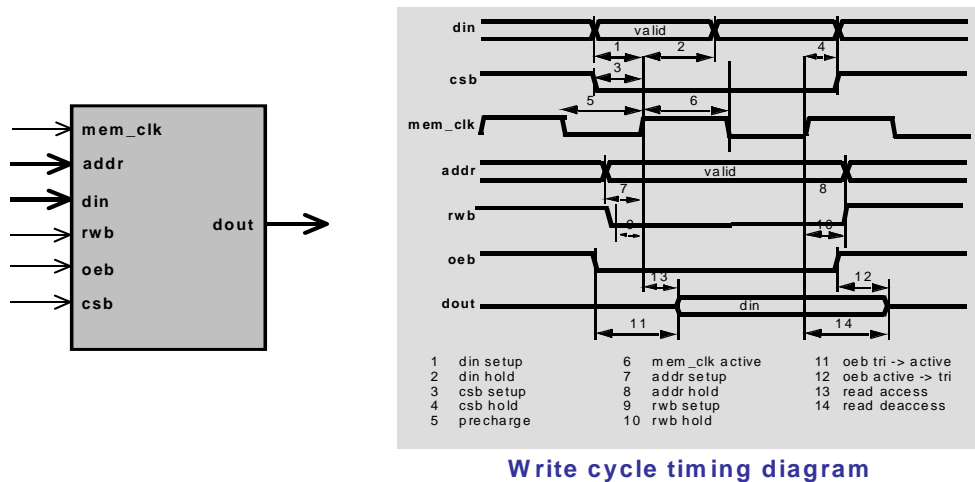


Read cycle timing diagram

### NOTES:

# A Synchronous RAM Example

## A Synchronous RAM Example (Continued)



Write cycle timing diagram

## NOTES:

# A Synchronous RAM Example

## A Synchronous RAM Example (Continued)

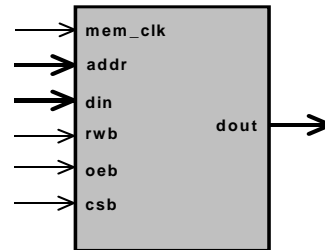
### ◆ Input/Output Definitions

```

model ram_1rw (addr, din, rwb, oeb, csb, mem_clk,
dout)(
  bist_definition (
    address addr (array = 4:0;);
    data_in din (array = 3:0;);
    data_out dout (array = 3:0;);
    output_enable oeb low;
    write_enable rwb low;
    chip_enable csb low;
    clock mem_clk high;

    tech = "technology_1";
    vendor = "acme_silicon";
    version = "1.0";
    message = "Synchronous SRAM, 1rw";
    address_size = 5;
    min_address = 0;
    max_address = 31;
    data_size = 4;
  )
);

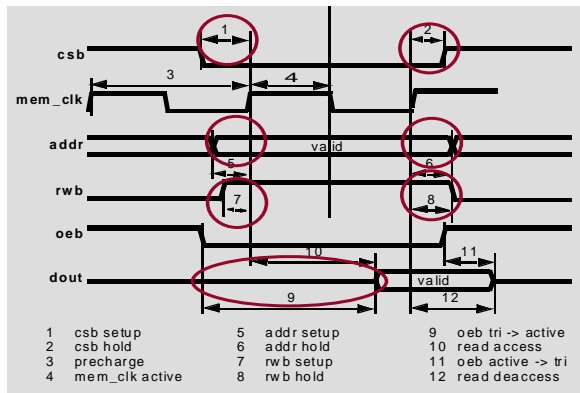
```



## NOTES:

# Interpreting the Read Cycle Timing

## Interpreting the Read Cycle Timing



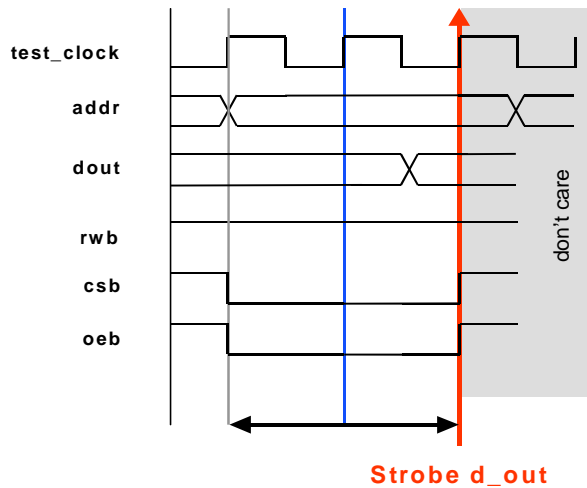
Read cycle timing diagram

- ♦ Read cycle is synchronized by mem\_clock rising edge
- ♦ Look for dependencies
  - csb setup before mem\_clk
  - addr setup before mem\_clk
  - Tpd oeb to dout
  - Tpd mem\_clk to dout
  - csb hold after mem\_clk
  - addr hold after mem\_clk
  - rwb setup and hold???
  - No - rwb is inactive
- ♦ No other dependencies

## NOTES:

# Defining the Read Cycle

## Defining the Read Cycle



### Read dependencies

- csb setup before mem\_clk
- addr setup before mem\_clk
- Tpd oeb to dout
- rwb remains inactive
- Tpd mem\_clk to dout
- csb hold after mem\_clk

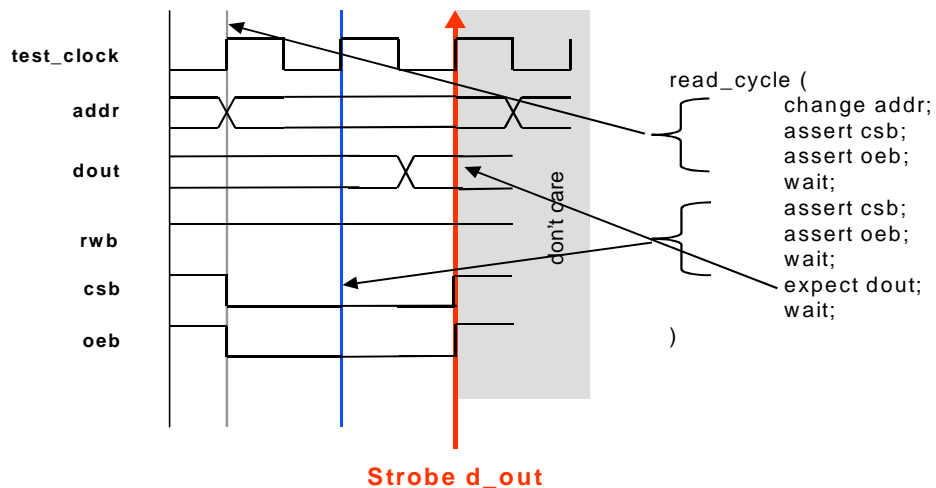
Set memory clock  
Set controller clock

## NOTES:



## Defining the Read Cycle

### Defining the Read Cycle (Continued)



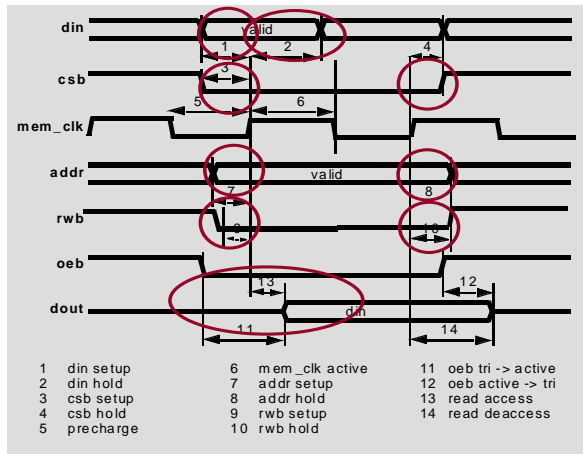
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### NOTES:

# Interpreting the Write Cycle Timing

## Interpreting the Write Cycle Timing



Write cycle timing diagram

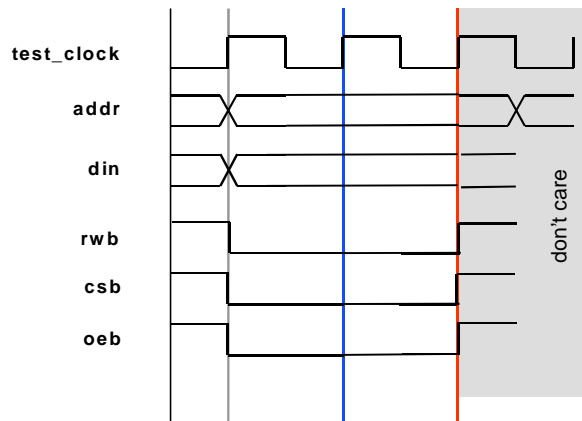
- ♦ Write cycle is synchronized by mem\_clk rising edge
- ♦ Look for dependencies
  - csb setup before mem\_clk
  - addr setup before mem\_clk
  - rwb setup before mem\_clk
  - Tpd mem\_clk to data valid
  - csb hold after mem\_clk
  - addr hold after mem\_clk
  - rwb hold after mem\_clk
  - oeb setup and hold?
    - No - oeb is for observe only
    - oeb can be asserted
- ♦ No other dependencies

## NOTES:

# Defining the Write Cycle

---

## Defining the Write Cycle



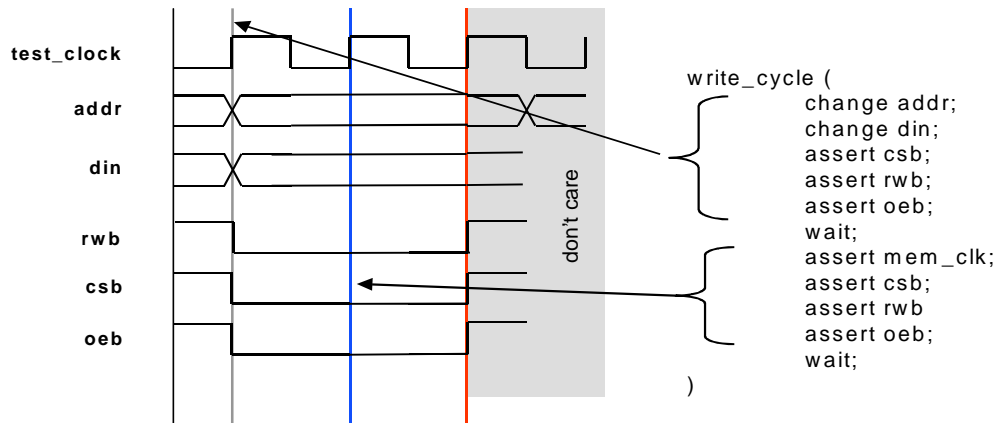
### ♦ Write dependencies

- csb setup before mem\_clk
- addr setup before mem\_clk
- din setup before mem\_clk
- rwb setup before mem\_clk
- Tpd oeb to dout
- Tpd mem\_clk to data valid
- csb hold after mem\_clk
- addr hold after mem\_clk
- rwb hold after mem\_clk

## NOTES:

# Defining the Write Cycle

## Defining the Write Cycle (Continued)



## NOTES:

# Defining Constant Values

---

### Defining Constant Values

- ♦ Some signals can be held constant during both read and write cycles
  - For example, output enable, chip enable
- ♦ These signals can be redefined in the memory model
  - Redefine the active state to be the *inactive state*
  - Remove the assert statements from the read and write cycles

#### Original Description

```
bist_definition (  
    address addr (array = 4:0;);  
    data_in din (array = 3:0;);  
    data_out dout (array = 3:0;);  
    output_enable oeb low;  
    write_enable rwb low;  
    chip_enable csb low;  
    clock mem_clk high;
```

#### Modified Description

```
bist_definition (  
    address addr (array = 4:0;);  
    data_in din (array = 3:0;);  
    data_out dout (array = 3:0;);  
    output_enable oeb HIGH;  
    write_enable rwb low;  
    chip_enable csb HIGH;  
    clock mem_clk high;
```

## NOTES:

# Defining Constant Values

---

## Defining Constant Values (Continued)

### ♦ Modified read and write cycles

```
read_cycle (                write_cycle (
    change addr;            change addr;
    wait;                   change din;
    assert mem_clk;         assert rwb;
    wait;                   wait;
    expect dout;            assert mem_clk;
    wait;                   assert rwb;
)                           wait;
                           )
```

## NOTES:

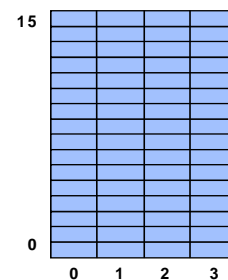
# Logical to Physical Mapping

## Logical to Physical Mapping

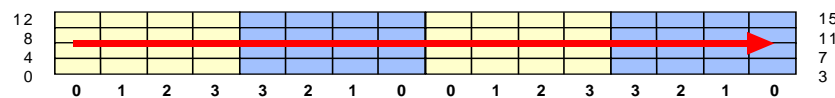
### ♦ Example : 64-bit RAM

- 16 words, 4 bits per word
- 16 physical columns by 4 physical rows
- 4 words per row

#### Logical Addressing



#### Physical Addressing



Externally, the memory illustrated may appear to consist of sixteen 4-bit words. The internal physical layout of a memory is organized in a two-dimensional matrix, in this case a common word per row configuration. Memory designers use different physical configurations in an effort to reduce cell space, reduce power consumption, increase yield (by including spare rows and columns), and accommodate mapping to standard chip pin assignments. In this example, there are four words per row.

# The Effect of Physical Topology

## The Effect of Physical Topology

### ♦ Apply checkerboard algorithm

- Should ensure inversion between every bit

Physical topology compromises algorithm effectiveness

12	0	1	0	1	1	0	1	0	0	1	0	1	1	0	1	0	15
8	0	1	0	1	1	0	1	0	0	1	0	1	1	0	1	0	11
4	0	1	0	1	1	0	1	0	0	1	0	1	1	0	1	0	7
0	0	1	0	1	1	0	1	0	0	1	0	1	1	0	1	0	3
Data bits	0	1	2	3	3	2	1	0	0	1	2	3	3	2	1	0	

**No inversions**

A checkerboard algorithm detects stuck-at-faults and shorts between adjacent cells by writing alternating 1's and 0's to cells as viewed from a logical layout. When the physical layout differs, the inversion of the bits between adjacent cells doesn't always happen, as shown in the illustration.



# Allowing for Physical Topology

## Allowing for Physical Topology

### ♦ Solution

- Adjust the data pattern to fit the physical topology
- Data inverted at addresses 1, 3, 4, 6, 9, 11, 12, 14

	12	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1	0	15
Address	8	0	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1	11
	4	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1	0	7
	0	0	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1	3
Data bits		0	1	2	3	0	1	2	3	0	1	2	3	0	1	2	3	

## NOTES:

# The Checkerboard Algorithm

---

## The Checkerboard Algorithm

- ♦ Supports basic “columns per row” architectures
- ♦ Library keywords define topology
  - `top_column` = number columns (words) per row
  - `top_word` defines muxing on address decoder
- ♦ Select algorithm with the following command:
  - Add Mbist Algorithms Checkerboard

```
top_column = 4;  
top_word = 0;
```

The Checkerboard algorithm reads the physical topology information from the memory model and adjusts the output patterns to create the proper checkerboard pattern among physically adjacent cells. When you are creating the memory model, you must include the physical topology information by placing the following lines within the memory model `bist_definition`. Often this information is not found in a standard memory data book and you must request it from the manufacturer. **`top_column=<value>`** tells the algorithm the number of words per row. The `<value>` can be any integer greater than 0. The algorithm uses this value to ensure that the first word of each row is different than the first word of the previous row, thus creating a checkerboard pattern. **`top_word=<value>`** tells if multiplexers in the column address decoder. A multiplexer is used to select between the bits of two words that are interleaved. If this is the case, then writing all 1's to one word and all 0's to the other creates a checkerboard pattern. 1 indicates there are multiplexers, 0 indicates there are not.

You must use the **Setup Observation Scheme -Compare** command when you use the Checkerboard algorithm to compare algorithms. In addition, multiple memories of different topologies can share the same controller. It is only necessary that each memory model contain its own `top_column` and `top_word` statements.

# Descrambling Functions

## Descrambling Functions

- ♦ **Descrambling provides most flexible topological mapping**
  - Used for more complex topological mapping
  - Defines where data is inverted
  - Addresses can also be descrambled
  - Must use “Setup Mux location -controller”

```
descrambling_definition (
  data_in (
    data0_desc = data0 XOR ((addr3 AND (NOT addr0) OR (addr0 AND (NOT addr3)));
    data1_desc = data0 XOR ((addr3 AND (NOT addr0) OR (addr0 AND (NOT addr3)));
    data2_desc = data0 XOR ((addr3 AND (NOT addr0) OR (addr0 AND (NOT addr3)));
    data3_desc = data0 XOR ((addr3 AND (NOT addr0) OR (addr0 AND (NOT addr3)));
  )
)
```

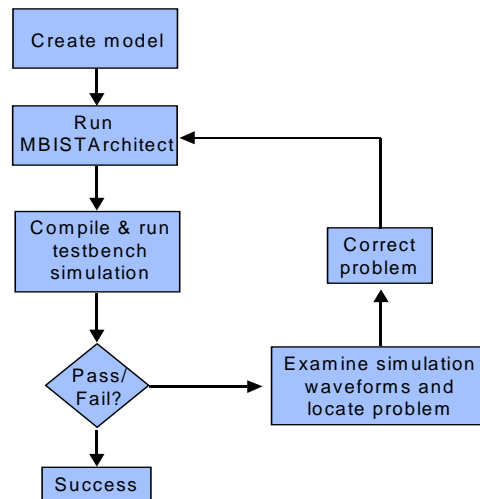
If your memory uses a scheme that “translates” an external address to an internal address or translates the input data in some way for internal storage, you must describe this translation to the Topchecker algorithm. Otherwise, an accurate checkerboard pattern cannot be generated. Usually this kind of information is not found in a standard data book and you must request it from the manufacturer.

The **address** subsection defines the descrambling for the address bus and the **data\_in** subsection defines the descrambling for the data input bus. For each address/data line of the memory there must be a line in the corresponding subsection. For example, if the width of the address bus is 4, there must be four lines in the **address** subsection of the descrambling definition section of the memory model. Similarly, if the width of the data bus is 8, there must be eight lines in the **data** section of the descrambling definition section of the memory model. The names of the descrambled address/data lines are arbitrary but the order of the statements in each section is important.

The first statement corresponds to the LSB and the last to the MSB. The supported Boolean operators are BUF, INV, AND, NAND, OR, NOR, XOR, XNOR. Finally, you must define BOTH address and data\_in subsections, regardless of whether or not scrambling information exists for both.

# Validating a Memory Model

## Validating a Memory Model



♦ Validation is performed

♦ Memory model errors will result in incorrect BIST controller

## NOTES:

# User Defined Algorithm

## User Defined Algorithm™

- ♦ You can define your March-type algorithm
- ♦ Not supported in User Defined Algorithm function:
  - Access to multiple ports at the same time
  - Non-March type algorithms (such as Galpat)
  - Example of UDA:

0	0	0	0	0
0	0	0	0	0
0	0	0	1	0
0	0	0	0	0

1. Write all 0 except base cell
2. Read first cell
3. Read base cell
4. Repeat 2-3 for all cells

Prior to inclusion of the Mentor Graphics User Defined Algorithm™ function, all of the test algorithms available in the MBISTArchitect tool were precoded into the tool. Adding a new algorithm required engineering work at the factory to support the new algorithm. The User Defined Algorithm functionality removes the pre-coded test algorithms and replaces them with algorithm definitions, loaded from files, which you can modify prior to BIST generation. All of the algorithms pre-configured as part of the MBISTArchitect tool, except the comparator test and port interaction tests, are defined within the software using this facility.

You can use User Defined Algorithms to define a class of simple March-type algorithms. This capability lets you define algorithms that perform a single memory access operation or more complex activity formed from read and write operations, at each address of a range of memory addresses.

When the memory BIST kernel is active, you can use User Defined Algorithm commands to load algorithms into the tool and change the set of available algorithms. The UDA algorithms use the UDA language that follows a Verilog-like style. Algorithms are composed of these parts:

- Tests
- Repetitions
- Steps

Use these commands when working with User Defined Algorithms: Load Algorithms, Delete Algorithms, and Report Algorithms.

User Defined Algorithm Exercises are available at the end of this module so that you can get experience using this MBISTArchitect feature.



# Troubleshooting a Memory Model

---

## Troubleshooting a Memory Model

- ◆ **Three major causes of mismatches**
- ◆ **Incorrect memory model description**
  - Re-examine datasheet and memory model
- ◆ **Additional MBISTArchitect commands required**
  - Some memories will require some setup in MBISTArchitect
- ◆ **Incorrect simulation model or inaccurate datasheet**
  - Intended behavior correct, but still getting problems

## NOTES:

# Troubleshooting Example: March2

---

## Troubleshooting Example: March2

▲(Wr0), ▲(R0, Wr1, R1), ▲(R1, Wr0, R0), ▼(R0, Wr1, R1), ▼(R1, Wr0, R0)

- ♦ Is data being written correctly during ▲(Wr0)?
  - Yes : write operation is correct
  - No : problem with write operation
- ♦ Is data being read correctly on R0 of ▲(R0,W1,R1)?
  - Yes : read operation is correct
  - No : problem with read operation
- ♦ Is data being read correctly on R1 of ▲(R0,W1,R1)?
  - Yes : read, write, and rwr operation is correct
  - No : problem optimizing read and write cycles to rwr
- ♦ Does incorrect behavior apply to all addresses?
  - Yes : problem is not address dependent
  - No : there may be a problem interfacing last ▲(W0) and first ▲(R0,W1,R1) operations

## NOTES:

## Module 5 Lab Exercises

- **Using the Model Editor**  
(20 minutes)
- **Reviewing a User Defined Algorithm**  
(20 minutes)
- **Running a User Defined Algorithm File**  
(20 minutes)

## Module 5: Lab Exercises

The following exercises introduce you to the Model Editor and User Defined Algorithms.

**Exercise 14: Modifying a Template to Match Your Memory Specifications** — You will use the model editor to make a working copy of a template, modify the template, and save the model.

**Exercise 15: Reviewing a User Defined Algorithm** — You will review a User Defined Algorithm (UDA) that has been modified to change the March1 algorithm.

**Exercise 16: Running a User Defined Algorithm File** — You will run the User Defined Algorithm reviewed in the previous exercise and simulate a memory model which uses the algorithm.

### Exercise 14: Modifying a Template to Match Your Memory Specifications

The purpose of this exercise is to give you step-by-step instruction on how to use the Model Editor. You will invoke the Model Editor, make a working copy of a template that comes close to the RAM model you need, modify the working copy to conform to the specifications of your particular RAM, then save the model. Before you begin, you should be aware of the following characteristics of the Model Editor:

- The Model Editor works on the principle of “Correct-by-Construction (CBC)”. It will only read and write a complete and syntactically-correct model file.
- The Model Editor works on selected objects. Therefore, in most cases, you must first select an object before you modify or replace that object.
- The Model Editor allows you to save multiple models to one file. Therefore, during a save operation, only a model with exactly the same name is overwritten in that file. If you save a model by a different name, that model will be appended to the existing models in the file.

## Becoming Familiar with your RAM Input/Output Specifications

1. Examine the following model information:

<p style="text-align: center;"><b>RAM4x16</b></p> <p style="text-align: center;"><b>1-Port Asynchronous RAM with 4 words by 16 bits</b></p> <p>Technology: Newest</p> <p>Version: 1.00</p> <p>Date: 4/29/96</p> <p>Inputs/Outputs</p> <p>a1:0 - Address lines.</p> <p>d15:0 - Data inputs.</p> <p>q15:0 - data outputs (tri-state).</p> <p>oe - Output Enable - active low.</p> <p>wrt - Write control line. A high state enables writing, a low state enables reading.</p> <p>Miscellaneous Info: Input and Output buses are defined as wide-pins (arrays) on the simulation model.</p>
--

## Make a Working Copy of a Similar Template

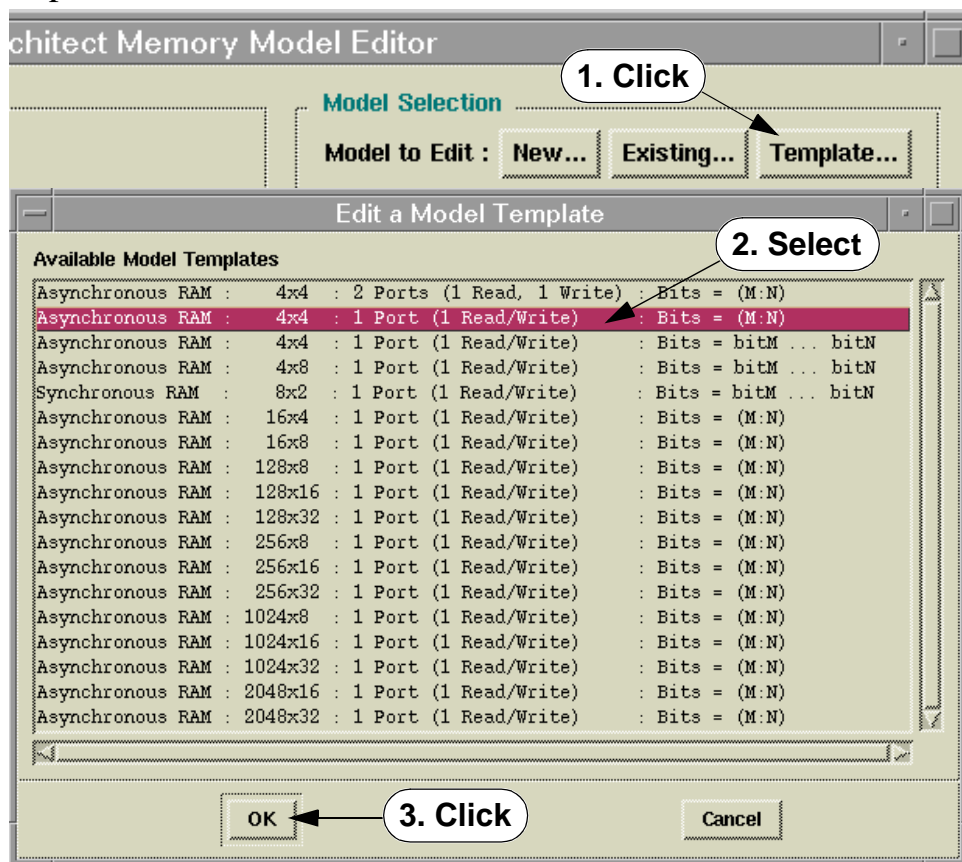
1. Move to the **mbist3** directory.

```
shell> cd $MBISTNWP/mbist3/ram4X16/results
```

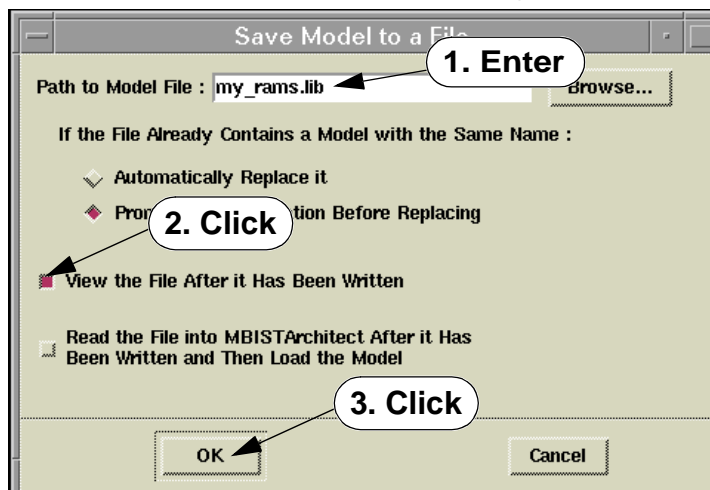
2. Invoke MBIST Architect:

```
shell> mbistarchitect
```

- Click the Model Editor button, then follow the directions below to select a template:



- Change the model name to `MY_RAM4X16_bussed`.
- Click **Save Model...**, then do the following.



The File Viewer window should appear with the template displayed. At this point, only the model name has changed.

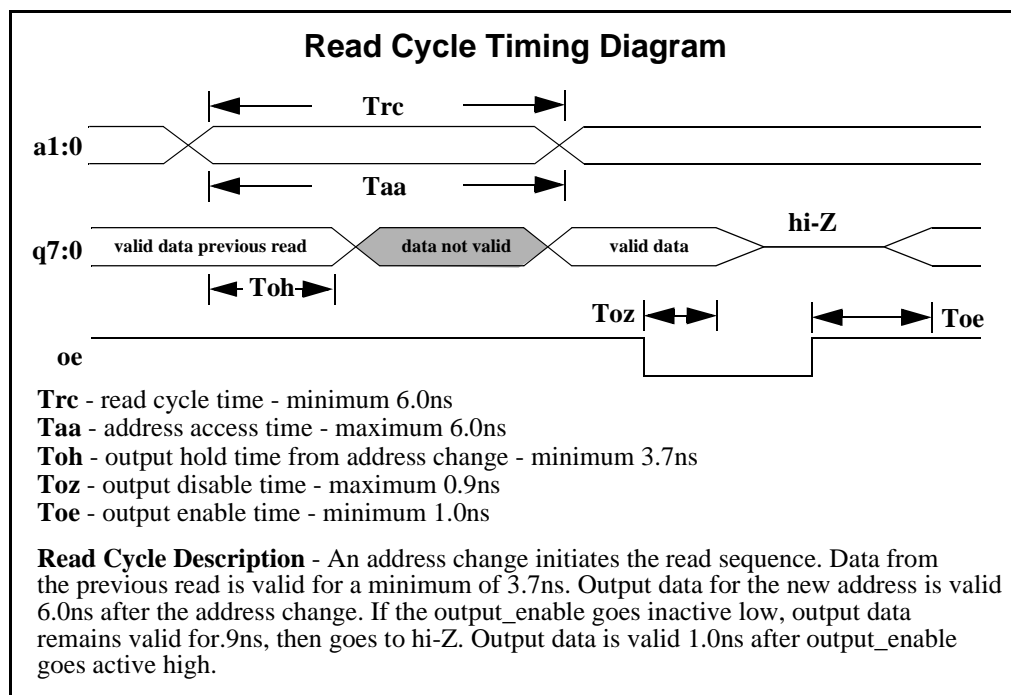
**BEST PRACTICE:** Keep the File Viewer window open and off to the side, so you can examine the updated status of the file after you execute each **Save Model...** command.

### Edit the Miscellaneous Information

1. Bring the Model Editor window to the front, then click on **Change Above Information**.
2. Change the **Data Width** to 16.
3. Change the **Message Text** to read “4X16 RAM”.
4. Click **OK**.

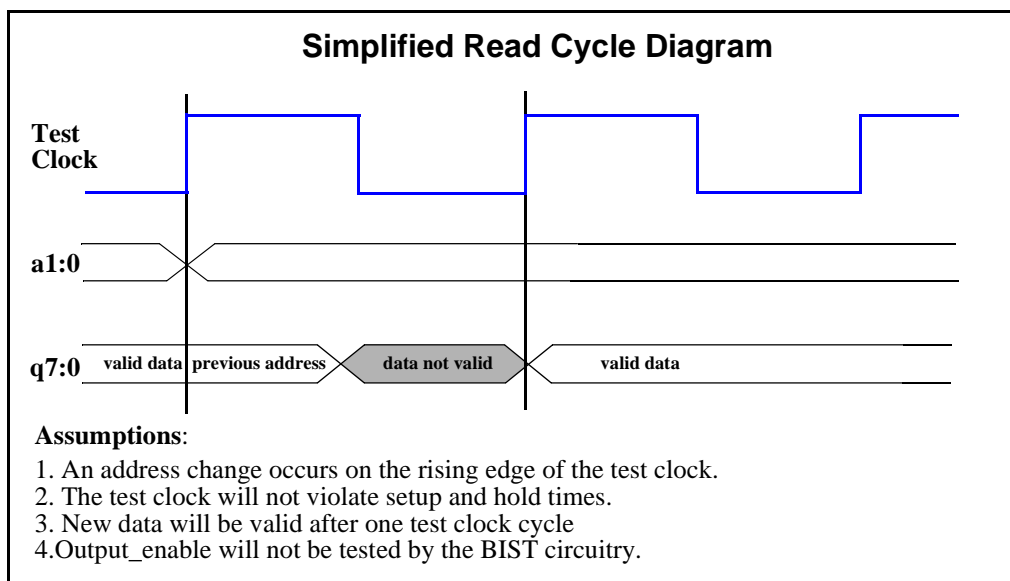
### Edit the Read Cycle Definition

1. Examine the following vendor timing diagram:





It is helpful to draw a simplified event-driven diagram that uses the test clock as a reference. For this first exercise, you can use the following diagram:



2. Verify that the Read Cycle for Port #1 is selected for editing. If not, select **"1 Read/Write (Read)"**, then click **Edit the Selected Cycle**.
3. Click **Define Cycle Pins**, then select the Address definition on the right side of the form.
4. Change the **Name** from "addr" to "a", click >> **Add** >>, then click **Change Selected**.
5. Select the **Data\_OUT** definition, then change the name from "do" to "q" and change the bus width to "15:0", click >> **Add** >>, then click **Change Selected**.
6. Add an Output\_Enable signal called oe. Since you will not be testing this signal, define it as active low even though it is active high. This causes the

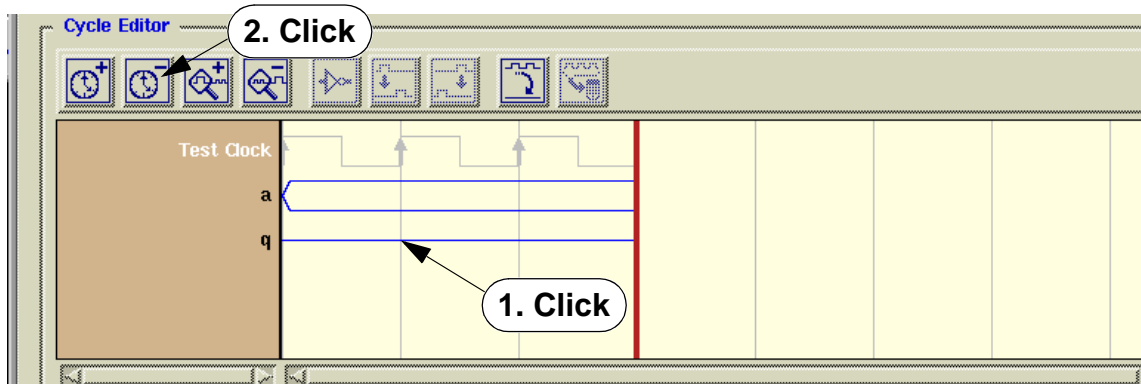
BIST controller to hold it in what it thinks is the inactive state (high), when in fact it is the active state (for example, output always enabled).



**Note**

In the Write Cycle editing session that follows, you will define the write\_enable “wrt” signal as active “high”. You can assume this signal is low during the read cycle, therefore you don’t have to define it as part of the read- cycle protocol.

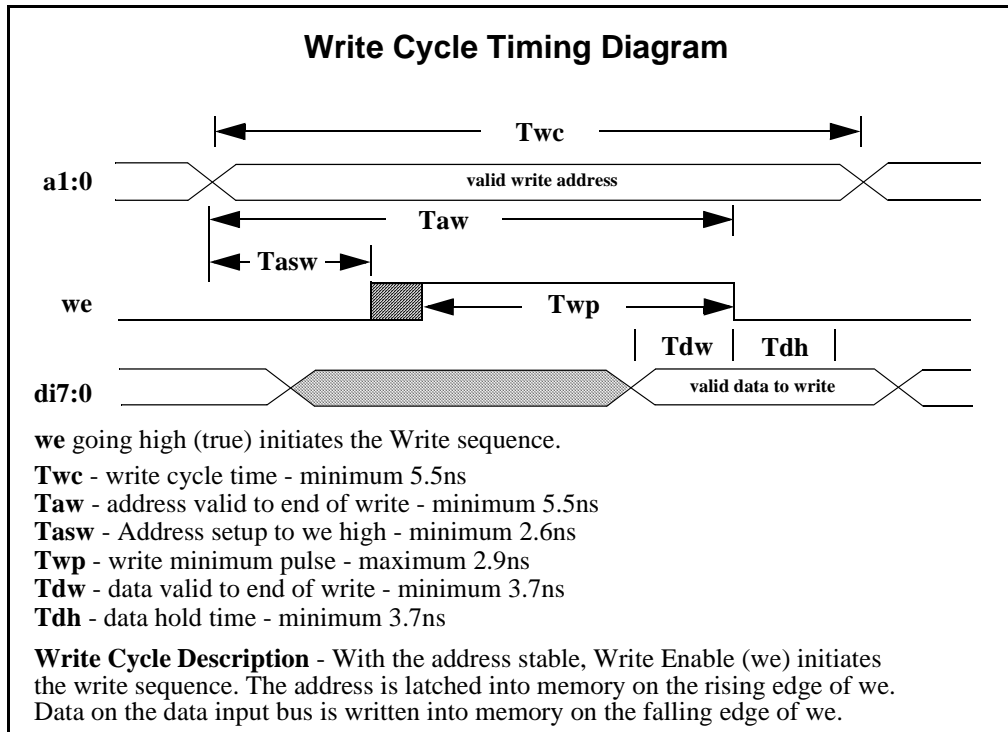
7. End the Cycle Pin editing by clicking **OK**.
8. Look at the Cycle Editor timing diagram. The output data should be valid one test clock cycle after a valid address change, so click on the “q” signal line where shown in Step 1 below. Also, the read cycle is complete within two test clock cycles, so you should shorten the Read Cycle by one test clock cycle, as shown in Step 2 below:



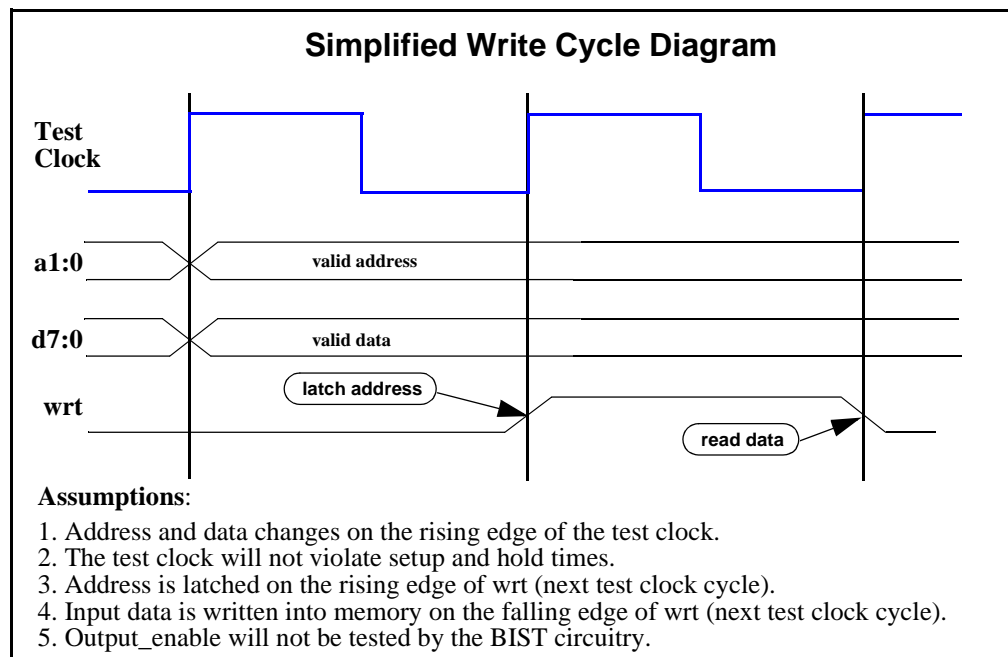
### Edit the Write Cycle Definition

In the following sequence, you will learn how to Import a signal definition from the Read Cycle to the Write Cycle.

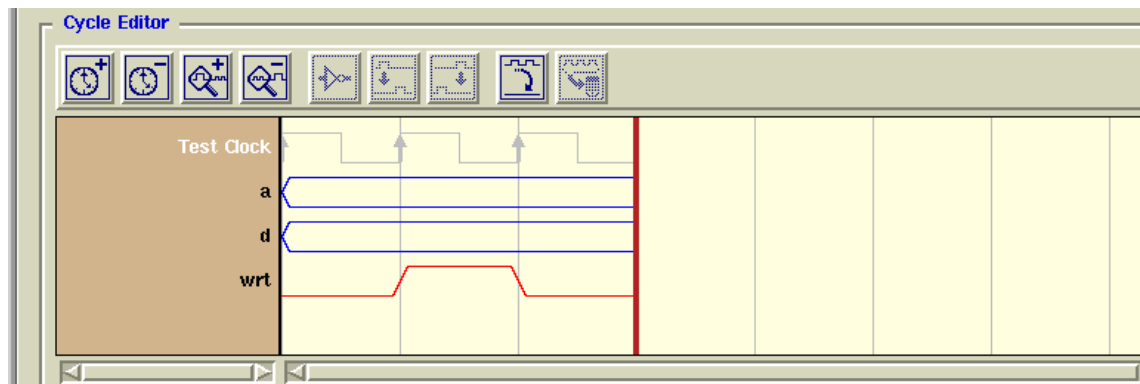
### 1. Examine the following Write Cycle timing diagram



The simplified event-driven diagram below uses the test clock as a reference:



2. Select “**1 Read/Write (Write)**” in the Model Editor window, then click **Edit the Selected Cycle**.
3. Click **Define Cycle Pins**, select the Address definition, then click **Import Pin...**
4. Select the Address definition, click **OK**, click >> **Add** >>, then click **Change Selected**.
5. Select the **Data\_IN** definition. Change the name from “di” to “d”, change the bus width to “15:0”, then >> **Add** >> the input to the Write Cycle definition.
6. Change the Write\_Enable input to an active high “wrt” signal, then click **OK**.
7. Look at the Cycle Editor timing diagram and make it conform to the illustration below:



8. End the Modeling Editing session by clicking on **Save Model...**, then click **OK**.
9. Examine the updated status of the Read and Write Cycle in the File Viewer window. Verify that it corresponds with your understanding of what the syntax should be.

### Summary

In this exercise, you modified a template to match the specifications of your particular RAM model. You changed the model name, changed the bus width specification, then modified the read cycle protocol. You then imported the read cycle specification to modify the write cycle protocol. You are now ready to invoke MBISTArchitect on this model and create a bist collar for it.

## Exercise 15: Reviewing a User Defined Algorithm

MBISTArchitect contains a User Defined Algorithm (UDA) feature that lets you create your own algorithms. The UDA functionality removes the pre-coded test algorithms and replaces them with algorithm definitions contained in files, which you can modify prior to BIST generation. You would typically create a user defined algorithm if you wanted to modify one of the memory test algorithms.

In this exercise, we will show you an example of a user defined algorithm. The next exercise shows you how to load a dofile that references this algorithm and to run the dofile in MBISTArchitect.

### Reviewing an Algorithm File

1. Move to the *mbist3/uda/design* directory.

```
shell> cd $MBISTNWP/mbist3/uda/design
```

2. The file named **marchA.dsc** is an algorithm file that has been created to modify the existing March 1 algorithm. Use your favorite text editor or **vi** to open this file.
3. This file contains the following sections:
  - Definition
  - Steps
  - Algorithm repetition

The Definition section contains the test name, a summary of the test, and size. This is followed by an algorithm definition that defines the actions to be taken in the algorithm. In this example, it defines the read and write operations performed during the up and down memory test.

### Definition Section

```
marchA

Summary:
    test example for a marching algorithm named marchA

Size: Copyright (C) Mentor Graphics Corporation 1999 All Rights
Reserved

10n

Algorithm:
    up    - write 0
    up    - read 0, write 1
    up    - read 1, write 0
    down  - read 0, write 1
    down  - read 1, write 0
    down  - read 0
```

The Steps section declares the basic activity across the address space of the memory ports. The step includes the following:

- **addr**  
The address clause defines what happens to the address register during the step of the algorithm.
- **data**  
A string that defines what data values will be used by the operation applied at each address visited by the algorithm step.
- **operation**  
A string that defines the activity, such as a read or write, that is performed at each address visited by the algorithm step.

## Steps Section

```
step wSeedUp;
  addr min, max, up, 1;
  data seed;
  operation w;

step rwInvSeedUp;
  addr min, max, up, 1;
  data invSeed;
  operation rw;

step rwSeedUp;
  addr min, max, up, 1;
  data seed;
  operation rw;

step rwInvSeedDown;
  addr min, max, down, 1;
  data invSeed;
  operation rw;

step rwSeedDown;
  addr min, max, down, 1;
  data seed;
  operation rw;

step rSeedDown;
  addr min, max, down, 1;
  data seed;
  operation r;
```

The Repetition section defines the action that will be taken in the algorithm. It includes the following:

- **seed**  
A string that specifies a common default value to be used by all the steps in the repetition.
- **keywords and steps**  
The begin and end keywords surround the body of the repetition declaration, which is a sequence of step references.



### Repetition Section

```
repetition marchA;  
  seed 0;  
begin  
  step wSeedUp;  
  step rwInvSeedUp;  
  step rwSeedUp;  
  step rwInvSeedDown;  
  step rwSeedDown;  
  step rSeedDown;  
end  
  
test marchA;  
  repetition marchA;
```

Once you have finished reviewing the sample algorithm, close the text editor. In the next exercise, you will load a dofile that references the MarchA algorithm and run the dofile in MBISTArchitect.

## Exercise 16: Running a User Defined Algorithm File

In this exercise, you will review a dofile that loads the user defined algorithm reviewed in [Exercise 15: Reviewing a User Defined Algorithm](#). You will also run the dofile in MBISTArchitect and synthesize the design.

### Running an Algorithm Dofile

1. Move to the *mbist3/uda/design* directory.

```
shell> cd $MBISTNWP/mbist3/uda/design
```

2. Use your favorite text editor or **vi** to open the *ram4x4.do* file. This file contains commands required to load the design, memory model and the MarchA algorithm.

#### **ram4x4.do sample**

```
loa li ../design/ram4x4.atpg
add me m ram4x4
load algorithm marchA.dsc
add mbis alg 1 marchA
run
save bist -replace
exit
```

3. Once you have finished reviewing the sample algorithm, close the text editor.
4. You are now ready to run MBISTArchitect and load this dofile. Change to the **results** directory.

```
shell> cd $MBISTNWP/mbist3/uda/results
```

Type the following command to launch MBISTArchitect and run the dofile:

```
shell > mbistarchitect -nogui -dofile ../design/ram4x4.do
```

MBISTArchitect will load the design, memory models, and MarchA algorithm. It will create BIST circuitry and create the following files, it will also save these files and exit the tool:

```
ram4x4_bist.v
ram4x4_bist_con.v
ram4x4_tb.v
```

### Verifying the BIST Circuitry

Next, you will use the MBISTArchitect-generated testbench to verify the memory BIST circuitry created by running the dofile.

1. Ensure that you are still working in the *\$MBISTNWP/mbist1/uda/results* directory.
2. Set up a work directory.

```
shell> $MGC_HOME/bin/vlib work
```

3. Compile the memory simulation model, all BIST models, and the testbench.

```
shell> $MGC_HOME/bin/vlog ../design/ram4x4.v ram4x4_bist.v\
ram4x4_bist_con.v ram4x4_tb.v
```

4. Simulate the test driver.
  - a. Invoke the QuickHDL simulator and load the testbench model.

```
shell> $MGC_HOME/bin/vsim ram4x4_tb
```

5. Set up the lists by running the following dofile:
  - a. VSIM 1> **do ../design/vsim\_setup.do**
  - b. This file sets the parameters for the simulation to stop due to **tst\_done** or **fail\_h** going high. It also sets up a List window so you can examine pertinent signals.

6. Run the simulation until it is finished.

VSIM 2> **run -all**

a. Write the displayed list to a file.

VSIM 2> **write list trace.log.uda**

b. Quit the simulation.

VSIM 4> **quit**

7. Examine the saved list file. Use whatever editor you prefer to view the *trace.log.uda* file you saved.

- The signals that comprise the columns in this file include (from left to right): *tst\_done*, *fail\_h*, the address, the write enable, the data input values, and the data output values.
- The first portion of the testbench tests some system signals.
- The MarchA algorithm is performed as follows:

-W(up)	1450-2150ns
-RW(up)	2250-3750ns
-RW(up)	3850ns-5350ns
-RW(down)	5450-6950ns
-RW(down)	7050-8550ns
-R(down)	8650-9550ns

# Trademark Information

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