

## CALCOLO DELLE PRESTAZIONI NEL CASO DI MODULAZIONE PSK

$$x[k] \equiv e^{i\theta_m}$$

$$\theta_m = \frac{2\pi}{M} (m-1) \quad (\text{PSK})$$

$$x[k] = \underbrace{\cos(\theta_m)}_{x_c[k]} + j \underbrace{\sin(\theta_m)}_{x_s[k]}$$

Poiché il simbolo  $x[k]$  si può scrivere nello stesso modo di un simbolo QAM, il calcolo delle prestazioni segue quello fatto per la QAM.

Ricogliamo adesso la densità spettrale di potenza delle componenti in base e quadratura del rumore in banda passante.

Supponiamo di avere una modulazione QAM.

Sì è visto che le segnali di uscita del ricevitore in assenza di cross-talk, e:

$$y_c(t) = \sum_{k=-\infty}^{+\infty} x_c[k] h(t - kT_s) + n_c(t) \otimes h_n(t) \quad (\text{RAMO IN FASE})$$

$$y_s(t) = \sum_{k=-\infty}^{+\infty} x_s[k] h(t - kT_s) + n_s(t) \otimes h_n(t) \quad (\text{RAMO IN QUADRATURA})$$

$$h(t) = p(t) \otimes c(t) \otimes h_n(t) = p(t) \otimes h_n(t)$$

Per calcolare le prestazioni del sistema di comunicazione è necessario conoscere la DSP dei processi di rumore  $n_c(t)$  e  $n_s(t)$ .

Per processi stazionari

$$\begin{cases} R_{n_c n_c}(t) = R_{n_s n_s}(t) \\ R_{n_s n_c}(t) = -R_{n_s n_c}(-t) = -R_{n_c n_c}(t) \end{cases}$$

$\Downarrow$  TF

$$\begin{cases} S_{ncnc}(f) = S_{nsns}(f) \\ S_{nsnc}(f) = -j S_{nsnc}(-f) \end{cases}$$

reale e pari

immaginario puro e dispari

N.B.  $R_{nsnc}(t) = -R_{nsnc}(-t)$

per  $t=0 \Rightarrow R_{nsnc}(0) = 0$

$$R_{nsnc}(0) = E\{n_s(t_1) \cdot n_c(t_1)\} = 0$$



$n_s(t_1)$  e  $n_c(t_1)$  sono  
V.A. incorrelate e  
se congiuntamente  
Gaussiane anche indipendenti

Affinché le porte in fase e quadratura siano indipendenti, non ci deve essere cross-talk e le V.A. di rumore devono essere incorrelate.

$$R_n(t) = R_{ncnc}(t) \cos(2\pi f_c t) - R_{nsnc}(t) \sin(2\pi f_c t) =$$

$$= \text{Re}\{R_{\tilde{n}}(t) e^{j2\pi f_c t}\}$$

$$R_{\tilde{n}}(t) = R_{ncnc}(t) + j R_{nsnc}(t)$$

TRASFORMATA DI FOURIER

$$S_{\tilde{n}}(f) = S_{ncnc}(f) + j S_{nsnc}(f)$$



$$\begin{cases} S_{ncnc}(f) = \frac{S_{\tilde{n}}(f) + S_{\tilde{n}}(-f)}{2} \\ j S_{nsnc}(f) = \frac{S_{\tilde{n}}(f) - S_{\tilde{n}}(-f)}{2} \end{cases}$$

$$S_n(f) = \frac{1}{2} S_{\tilde{n}}(f-f_0) + \frac{1}{2} S_{\tilde{n}}(f+f_0)$$

Osservazioni

(1) per un processo porta-banda

$$S_n(f) = 0 \quad |f| \geq 2f_0$$

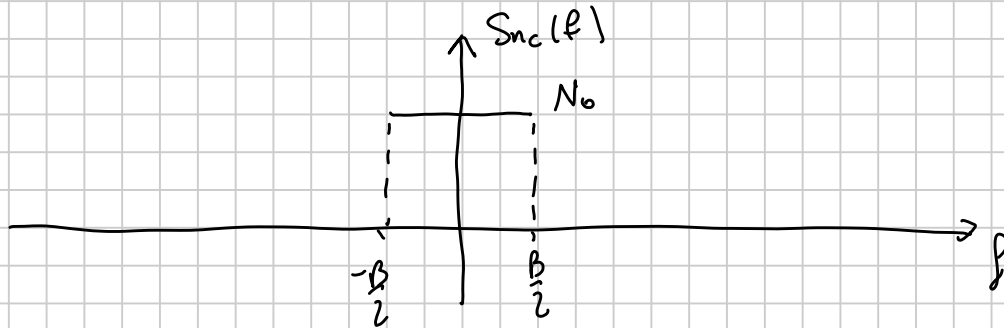
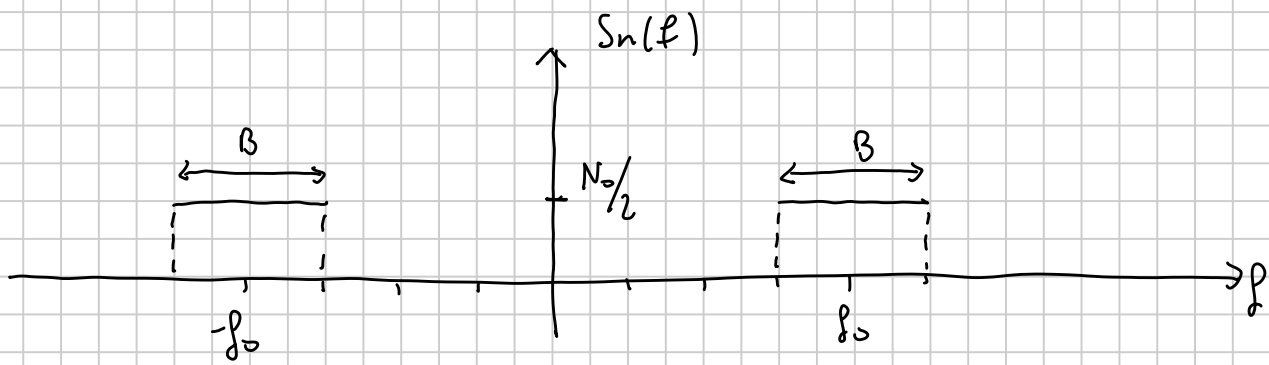
(2) Per definizione la DSP dell'involuppo complesso  $\tilde{n}(t)$

$$S_{\tilde{n}}(f) = 0 \quad \text{per } f < -f_0$$

$\Downarrow$   $S_n(f)$  è reale e pari.

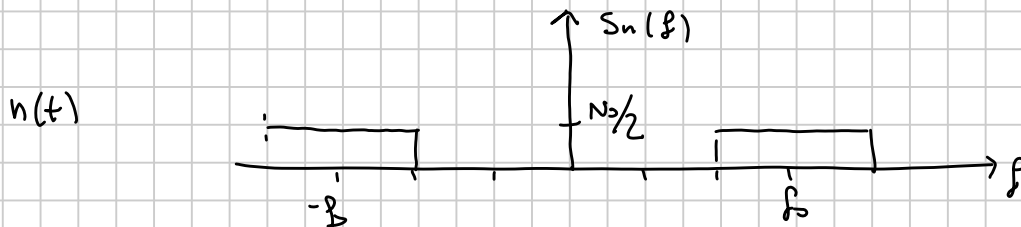
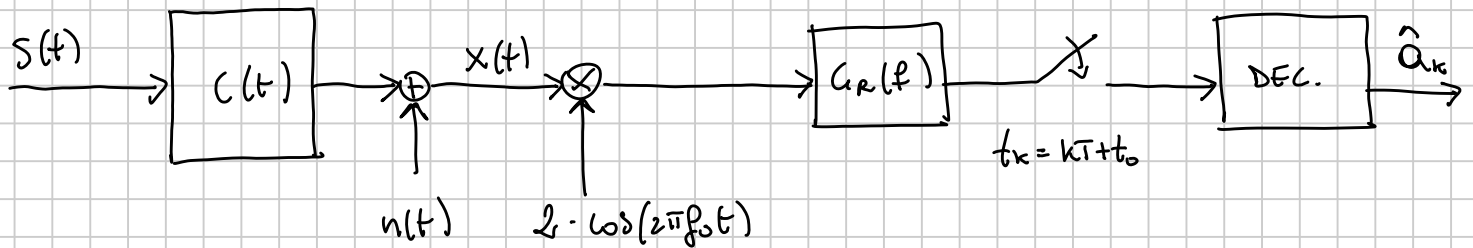
$$S_{n_c}(f) = S_{n_s}(f) = \begin{cases} S_n(f+f_0) + S_n(f-f_0) & |f| \leq f_0 \\ 0 & \text{altrove} \end{cases}$$

$$j S_{n_s n_c}(f) = -j S_{n_c n_s}(f) = \begin{cases} S_n(f+f_0) - S_n(f-f_0) & |f| \leq f_0 \\ 0 & \text{altrove} \end{cases}$$



$$S_{nsnc}(f) = -S_{ncns}(f) = 0$$

Esercizio PAM banda passante



$$n(t) = \sum_{k=-\infty}^{+\infty} a_k p(t - kT) \cos(2\pi f_0 t)$$

$a_k \in A_s = \{\pm 1\}$  indip. ed equiprobabili

$$p(f) = \begin{cases} \sqrt{T} \cos(\pi f T / 2) & |f T| \leq 1 \\ 0 & \text{altrove} \end{cases}$$

$$G_R(f) = P(f)$$

$$C(f) = \begin{cases} e^{-j[2\pi(f-f_0)T + \pi/6]} & |f - f_0| \leq \frac{1}{T} \\ e^{-j[2\pi(f+f_0)T - \pi/6]} & |f + f_0| \leq \frac{1}{T} \\ 0 & \text{altrove} \end{cases}$$

Calcoliamo:

$$1) P_E(M) = P_E(b)$$

2)  $E_s$  energia media per simbolo trasmesso

Soluzione :

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k \left[ \underbrace{p(t-kT) \cos(2\pi f_0 t)}_{g(t)} \right] \otimes c(t) + n(t)$$

$$q(t) = [p(t - kT) \cos(2\pi\beta t)] \otimes c(t)$$

$$Q(f) = \left[ P(f) e^{-j2\pi kT f} \otimes \left( \frac{\delta(f-f_0) + \delta(f+f_0)}{2} \right) \right] \cdot C(f)$$

$$Q(f) = \left[ \frac{1}{2} P(f-f_0) e^{-j2\pi kT(f-f_0)} + \frac{1}{2} P(f+f_0) e^{-j2\pi kT(f+f_0)} \right] \cdot C(f)$$

$$= \frac{1}{2} P(f-f_0) e^{-j2\pi kT(f-f_0)} e^{-j2\pi(f-f_0)\tau} e^{-j\pi/6} +$$

$$+ \frac{1}{2} P(f+f_0) e^{-j2\pi kT(f+f_0)} e^{-j2\pi(f+f_0)\tau} e^{+j\pi/6}$$

$$Q(f) = \frac{1}{2} P(f-f_0) e^{-j2\pi(f-f_0)(kT+\tau)} e^{-j\pi/6} +$$

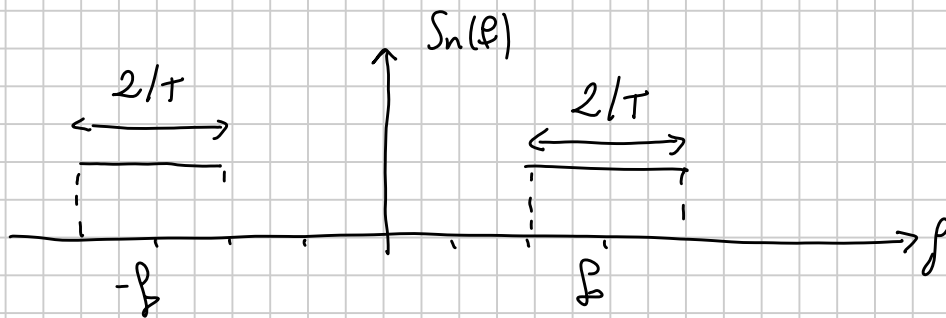
$$\frac{1}{2} P(f+f_0) e^{-j2\pi(f+f_0)(kT+\tau)} e^{+j\pi/6}$$

$$q(t) = p(t - kT - \tau) \cos(2\pi\beta t + \pi/6)$$

$$g(t) = p(t - kT - \tau) \left[ \cos(2\pi\beta t) \cos(\pi/6) - \sin(2\pi\beta t) \sin(\pi/6) \right]$$

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k p(t - kT - \tau) \left[ \cos(2\pi\beta t) \cos(\pi/6) - \sin(2\pi\beta t) \sin(\pi/6) \right] + n(t)$$

$n(t)$  processo Gaussiano stocastico aditivo passa-banda



$$n(t) = n_c(t) \cos(2\pi\beta t) - n_s(t) \sin(2\pi\beta t)$$

Il segnale demodulato  $z(t)$  sarà:

$$z(t) = x(t) \cdot 2 \cos(2\pi\beta t) = z_s(t) + z_n(t)$$

$$z_s(t) = 2 \sum_{k=-\infty}^{+\infty} a_k p(t - kT - \tau) \left[ \cos^2(2\pi\beta t) \cos(\pi/6) - \sin(2\pi\beta t) \cdot \cos(2\pi\beta t) \sin(\pi/6) \right]$$



$$2 \cos^2(\alpha) = 1 + \cos(2\alpha)$$

$$2 \sin(\alpha) \cos(\alpha) = \sin(2\alpha)$$

$$z_s(t) = \sum_k Q_k p(t - kT - \tau) \left[ \left( 1 + \cos(4\pi\beta t) \right) \cos(\pi/6) - \sin(4\pi\beta t) \sin(\pi/6) \right]$$

$$z_n(t) = n(t) \cdot 2 \cos(2\pi\beta t)$$

$$n(t) = n_c(t) \cos(2\pi\beta t) - n_s(t) \sin(2\pi\beta t)$$

$$z_n(t) = n_c(t) \left( 1 + \cos(4\pi\beta t) \right) - n_s(t) \sin(4\pi\beta t)$$

$g_n(p)$  è un filtro adattato e passa-basso

$$y(t) = z(t) \otimes g_n(t) = z_s(t) \otimes g_n(t) + z_n(t) \otimes g_n(t)$$

$$y_s(t) = z_s(t) \otimes g_n(t) = \sum_{k=-\infty}^{+\infty} Q_k \cos(\pi/6) p(t - kT - \tau) \otimes g_n(t)$$

$$y_n(t) = z_n(t) \otimes g_n(t) = n_c(t) \otimes g_n(t)$$

$$q(t) = p(t) \otimes g_p(t) = p(t) \otimes p(t)$$

$$Q(f) = P^2(f) = \begin{cases} T \cos^2(\pi f T/2) & |fT| \leq 1 \\ 0 & \text{altrove} \end{cases}$$

$$= \begin{cases} \frac{T}{2} (1 + \cos(\pi f T)) & |fT| \leq 1 \\ 0 & \text{altrove} \end{cases}$$

FUNZIONE COSENO

RIALZATO GN

$$\alpha = 1$$



Soddisfa il criterio di Nyquist

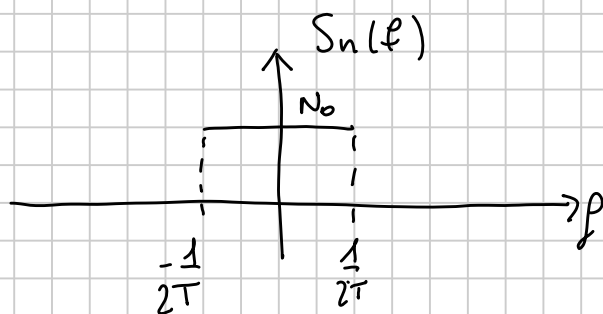
$$\sum_k Q(f - \frac{k}{T}) = T \Rightarrow q(0) = 1$$

$$y_s(t) = \sum_k a_k \cos(\pi/6) q(t - kT - \tau)$$

$$t_i = iT + t_0 = iT + \tau \Rightarrow t_0 = \tau$$

$$y_s[k] = a_k q(0) \cos(\pi/6) = a_k \cos(\pi/6)$$

$$g_n(t) = n_c(t) \otimes g_n(t)$$

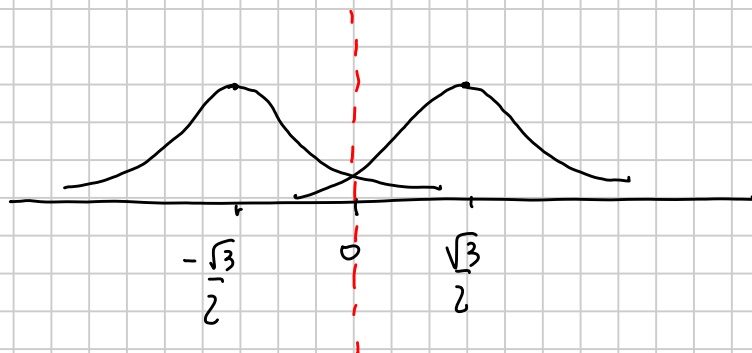


$$\zeta_p^2 = N_0 \int g_n(t)^2 dt = N_0 \int G_n^2(f) df = N_0 \int P^2(f) df = N_0$$

$$y[k] = a_k \cos(\pi/6) + n_{cp}[k]$$

$$a_k \in \{\pm 1\}$$

$$\hat{a} = 0$$



$$P_E(u) = P_E(b) = \frac{1}{2} \Pr\{\hat{a}_k = 1 | a_k = -1\} + \frac{1}{2} \Pr\{\hat{a}_k = -1 | a_k = 1\}$$

$$\Pr\{\hat{a}_k = 1 | a_k = -1\} = \Pr\{\hat{a}_k = -1 | a_k = 1\}$$

$$P_E(u) = \Pr\{\hat{a}_k = 1 | a_k = -1\} = Q\left(\frac{\sqrt{3}/2}{\sqrt{N_s}}\right) = Q\left(\sqrt{\frac{3}{4N_0}}\right)$$

$E_s$  per simbolo  $T_x$

$$s(t) = \sum_i a_i p(t - iT) \cos(2\pi f_b t)$$

$$E_{s1} \Rightarrow s_1(t) = 1 \cdot p(t) \cos(2\pi f_b t) \quad (a_k = 1)$$

$$E_{s1} = \int_{-\infty}^{+\infty} s_1^2(t) dt = \int_{-\infty}^{+\infty} p^2(t) \cos^2(2\pi f_b t) dt \quad (a_k = -1)$$

$$= \int_{-\infty}^{+\infty} p^2(t) \left( \frac{1}{2} + \frac{1}{2} \cos(4\pi f_b t) \right) dt \approx \frac{1}{2} \int_{-\infty}^{+\infty} p^2(t) dt = \frac{1}{2} E_p$$

$$E_{s2} \Rightarrow s_2(t) = -1 \cdot p(t) \cos(2\pi f_b t)$$

$$E_{s2} = \int_{-\infty}^{+\infty} p^2(t) \cos^2(2\pi f_b t) dt = E_{s1}$$

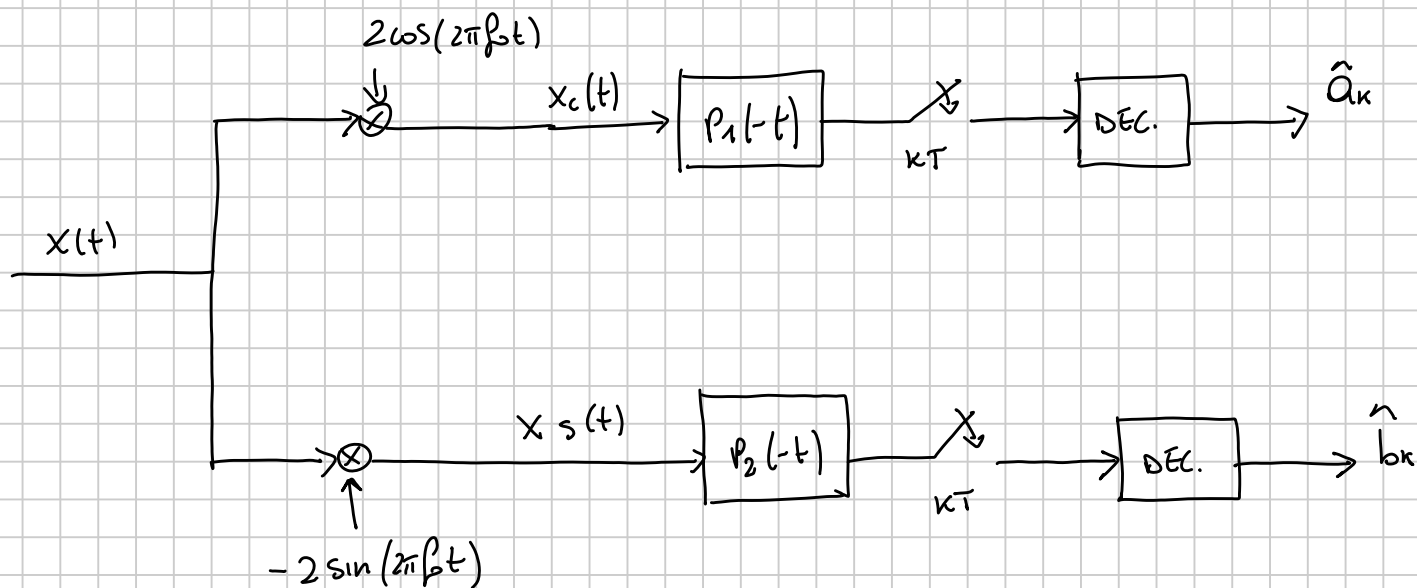
$$E_s = \frac{1}{2} E_{s1} + \frac{1}{2} E_{s2} = E_{s1} = \frac{1}{2} E_p = \frac{1}{2}$$

$E_s \Rightarrow$  energia medio per intervallo di segnalazione del segnale  $T_x$

$$1) \quad E_s = P_s T_s \quad P_s = \int_{-\infty}^{+\infty} S_s(f) df$$

$$2) \quad E_s = \int_0^T s^2(t) dt$$

# QAM in banda passante

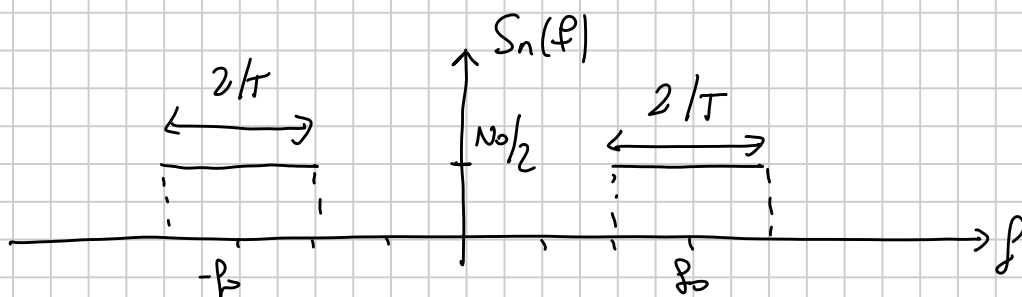


$$x(t) = \sum_{i=-\infty}^{+\infty} a_i p_1(t-iT) \cos(2\pi f_0 t) - \sum_{i=-\infty}^{+\infty} b_i p_2(t-iT) \sin(2\pi f_0 t) + n(t)$$

$$a_i \in \{\pm 2\} \quad b_i \in \{\pm 1\} \quad \text{indip. ed equiprobabili}$$

$$p_1(f) = \begin{cases} T \sqrt{1-f^2} & |f| \leq 1 \\ 0 & \text{altrove} \end{cases}$$

$$p_2(f) = \begin{cases} \sqrt{T} \sin(\pi |f| T/2) & |f| \leq 1 \\ 0 & \text{altrove} \end{cases}$$



## RAMO IN FASE

$$x_c(t) = x(t) \cdot 2 \cos(2\pi\beta t) =$$

$$= \sum_i a_i p_1(t-iT) 2 \cos^2(2\pi\beta t) - \sum_i b_i p_2(t-iT) 2 \sin(2\pi\beta t) 2 \cos(2\pi\beta t) +$$
$$n(t) \cdot 2 \cos(2\pi\beta t)$$

$$n(t) = n_c(t) \cos(2\pi\beta t) - n_s(t) \sin(2\pi\beta t)$$

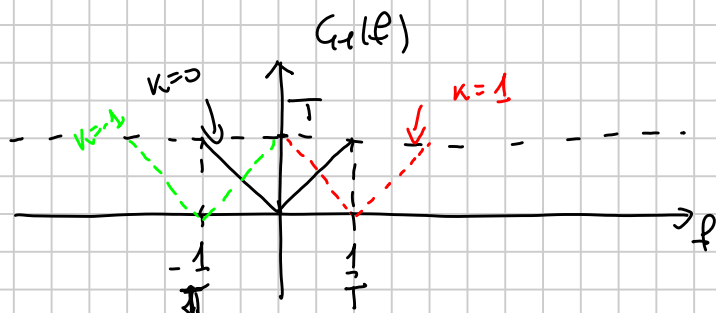
$$= \sum_i a_i p_1(t-iT) (1 + \cos(4\pi\beta t)) - \sum_i b_i p_2(t-iT) \sin(4\pi\beta t) +$$
$$n_c(t) (1 + \cos(4\pi\beta t)) - n_s(t) \sin(4\pi\beta t)$$

7) J'etablis in ricezione  $p_1(t)$  e F.A. e posso - posso

$$y_c(t) = \sum_i a_i p_1(t-iT) \otimes p_1(-t) + n_c(t) \otimes p_1(-t)$$

$$g_1(t) = p_1(t) \otimes p_1(-t)$$

$$G_1(f) = P_1(f) \cdot P_1^*(f) = |P_1(f)|^2 = P_1^2(f) = \begin{cases} T^2 |f| & |f| \leq 1 \\ 0 & \text{altrove} \end{cases}$$



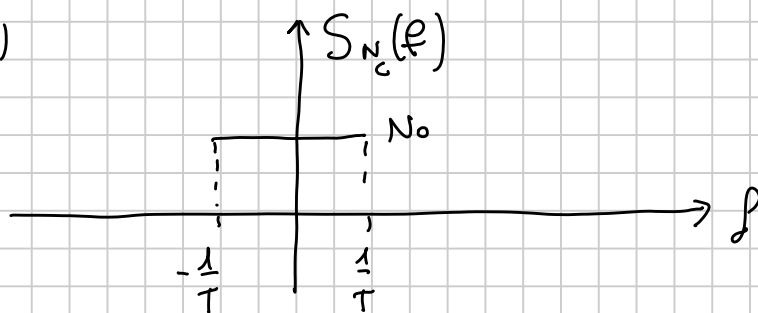
$$\sum_k G_1(f - k/T) = T \quad \Rightarrow \quad g_1(0) = 1$$

$$g_1(k) = 0 \quad k \neq 0$$

Altre uscita del compressore

$$y_c[k] = a_k + n_{cp_1}[k]$$

$$n_{cp_1}(t) = n_c(t) \otimes p_1(-t)$$



RAHO IN QUADRATURA

$$x_s(t) = x(t) \cdot (-2 \sin(2\pi\beta t)) =$$

$$= \sum_i a_i p_1(t-iT) (-2 \sin(2\pi\beta t) \cdot \cos(2\pi\beta t)) + \sum_i b_i p_2(t-iT) 2 \sin^2(2\pi\beta t)$$

$$+ n(t) (-2 \sin(2\pi\beta t))$$

$$n(t) = n_c(t) \cos(2\pi\beta t) - n_s(t) \sin(2\pi\beta t)$$

$$= - \sum_i a_i p_1(t-iT) \sin(4\pi\beta t) + \sum_i b_i p_2(t-iT) (1 - \cos(4\pi\beta t))$$

$$-n_c(t) \sin(4\pi f t) + n_s(t) (1 - \cos(4\pi f t))$$

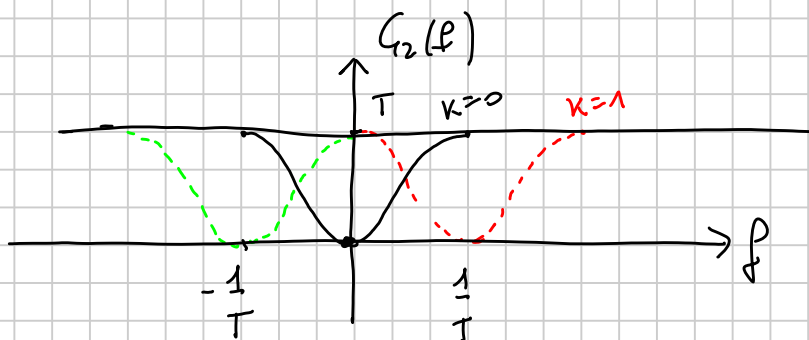
2) Fieho in ruazione e<sup>-</sup>  $p_2(-t) \Rightarrow$  F.A. e postu busto

$$y_s(t) = \sum_i b_i p_2(t-iT) \otimes p_2(-t) + n_s(t) \otimes p_2(-t)$$

$$g_2(t) = p_2(t) \otimes p_2(-t)$$

$$G_2(f) = P_2(f) \cdot P_2^*(f) = |P_2(f)|^2 = P_2^2(f) = \begin{cases} T \sin^2(\pi |f| T/2) & |f| \leq 1 \\ 0 & \text{altrove} \end{cases}$$

$$G_2(f) = \begin{cases} \frac{T}{2} (1 - \cos(\pi f T)) & |f| \leq 1 \\ 0 & \text{altrove} \end{cases}$$



$$\sum G_2(f - k/T) = T \Rightarrow g_2(0) = 1$$

$$\frac{T}{2} (1 - \cos(\pi f T)) + \frac{T}{2} (1 - \cos(\pi (f - \frac{1}{T}) T)) =$$



$$= \frac{I}{2} - \frac{I}{2} \cos(\pi f T) + \frac{I}{2} - \frac{I}{2} \cos(\pi f T - \pi) = I$$

$= -\cos(\pi f T)$

All'uscita del compressore abbiamo:

$$y_s[k] = b_k + n_{sp}[k]$$

$$\begin{cases} y_c[k] = a_k + n_{cp}[k] = a_k + v_c \\ y_s[k] = b_k + n_{sp}[k] = b_k + v_s \end{cases}$$

$$S_{v_c}(f) = S_{n_c}(f) |P_1(f)|^2$$

$$S_{n_c}(f) = S_{n_s}(f)$$

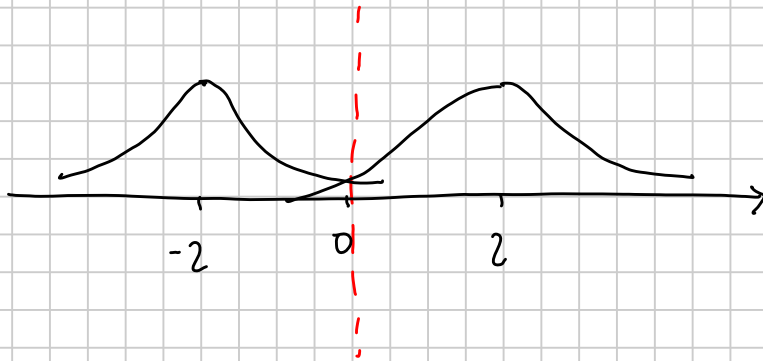
$$S_{v_s}(f) = S_{n_s}(f) |P_2(f)|^2$$

$$\sigma_{v_c}^2 = \int_{-\infty}^{+\infty} S_{n_c}(f) |P_1(f)|^2 df = N_0 \int_{-\infty}^{+\infty} P_1^2(f) df = N_0 \int_{-\infty}^{+\infty} G_1(f) df = N_0 g_1(0) = N_0$$

$$\sigma_{v_s}^2 = \int_{-\infty}^{+\infty} S_{n_s}(f) |P_2(f)|^2 df = N_0 \int_{-\infty}^{+\infty} P_2^2(f) df = N_0 \int_{-\infty}^{+\infty} G_2(f) df = N_0 g_2(0) = N_0$$

## RAMO IN FASE

$$y_c[k] = a_k + v_c$$

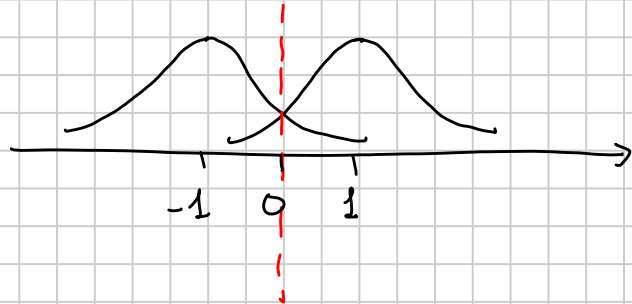


$$P_c = \frac{1}{2} \Pr\{\hat{a}_k = -2 \mid a_k = 2\} + \frac{1}{2} \Pr\{\hat{a}_k = 2 \mid \hat{a}_k = -2\} = \Pr\{\hat{a}_k = -2 \mid a_k = 2\}$$

$$P_c = Q\left(\frac{2}{\sqrt{N_0}}\right)$$

## RAMO IN QUADRATURA

$$y_s[k] = b_k + v_s$$



$$P_s = \frac{1}{2} \Pr\{\hat{b}_k = -1 \mid b_k = 1\} + \frac{1}{2} \Pr\{\hat{b}_k = 1 \mid b_k = -1\} = \Pr\{\hat{b}_k = -1 \mid b_k = 1\}$$

$$P_s = Q\left(\frac{1}{\sqrt{N_0}}\right)$$

$$P_E^{\text{QAM}}(4) = P_c (1 - P_s) + P_s (1 - P_c) + P_s P_c$$

Calcoliamo adesso l'energia media per simbolo trasmesso.

$$S_1(t) = 2p_1(t) \cos(2\pi f_s t) - p_2(t) \sin(2\pi f_s t) \quad (2, 1)$$

$$S_2(t) = 2p_1(t) \cos(2\pi f_s t) + p_2(t) \sin(2\pi f_s t) \quad (2, -1)$$

$$S_3(t) = -2p_1(t) \cos(2\pi f_s t) - p_2(t) \sin(2\pi f_s t) \quad (-2, 1)$$

$$S_4(t) = -2p_1(t) \cos(2\pi f_s t) + p_2(t) \sin(2\pi f_s t) \quad (-2, -1)$$

$$E_{n1} = \int_{-\infty}^{+\infty} S_1^2(t) dt = \int_{-\infty}^{+\infty} 4p_1^2(t) \cos^2(2\pi f_s t) dt + \int_{-\infty}^{+\infty} p_2^2(t) \sin^2(2\pi f_s t) dt =$$

$$= 4 \int_{-\infty}^{+\infty} p_1^2(t) \left( \frac{1}{2} + \frac{1}{2} \cos(4\pi f_s t) \right) dt + \int_{-\infty}^{+\infty} p_2^2(t) \left( \frac{1}{2} - \frac{1}{2} \cos(4\pi f_s t) \right) dt =$$

$$= 4 \cdot \frac{1}{2} \int_{-\infty}^{+\infty} p_1^2(t) dt + \frac{1}{2} \int_{-\infty}^{+\infty} p_2^2(t) dt = 2 E_{p1} + \frac{1}{2} E_{p2}$$

$$E_{n2} = E_{n3} = E_{n4} = E_{n1} = 2 E_{p1} + \frac{1}{2} E_{p2}$$

$$E_s = \frac{1}{4} E_{n1} + \frac{1}{4} E_{n2} + \frac{1}{4} E_{n3} + \frac{1}{4} E_{n4} = 2 E_{p1} + \frac{1}{2} E_{p2}$$

$E_s$  energia media per intervallo di segnalazione del segnale TDMSS.

$$E_s = E \left\{ \int_0^T r^2(t) dt \right\}$$

$$E \{ a_i \cdot a_{i+k} \} = \begin{cases} E \{ a_i^2 \} & k=0 \\ (E \{ a_i \})^2 & k \neq 0 \end{cases} = \begin{cases} 4 & k=0 \\ 0 & k \neq 0 \end{cases}$$

$$E \{ a_i^2 \} = \frac{1}{2} (4) + \frac{1}{2} (4) = 4$$

$$R_a[k] = 4 \delta[k]$$

$$E \{ b_i \cdot b_{i+k} \} = \begin{cases} E \{ b_i^2 \} & k=0 \\ (E \{ b_i \})^2 & k \neq 0 \end{cases} = \begin{cases} 1 & k=0 \\ 0 & k \neq 0 \end{cases}$$

$$R_b[k] = \delta[k]$$

$$E \{ a_i b_{i+k} \} = E \{ a_i \} \cdot E \{ b_{i+k} \} = 0 \quad \forall k$$

$$f_s = E \left\{ \int_0^T n^2(t) dt \right\} =$$

$$= E \left\{ \int_0^T \left( \sum_i a_i p_1(t-iT) \cos(2\pi\beta t) - \sum_i b_i p_2(t-iT) \sin(2\pi\beta t) \right) \right.$$

$$\left. \left( \sum_k a_k p_1(t-kT) \cos(2\pi\beta t) - \sum_k b_k p_2(t-kT) \sin(2\pi\beta t) \right) dt \right\}$$

$$= \int_0^T \sum_i \sum_k E\{a_i a_k\} p_1(t-iT) p_1(t-kT) \cos^2(2\pi\beta t) dt +$$

$$- \int_0^T \sum_i \sum_k E\{a_i b_k\} p_1(t-iT) p_2(t-kT) \cos(2\pi\beta t) \sin(2\pi\beta t) dt +$$

$$- \int_0^T \sum_i \sum_k E\{a_k b_i\} p_1(t-kT) p_2(t-iT) \cos(2\pi\beta t) \sin(2\pi\beta t) dt$$

$$+ \int_0^T \sum_i \sum_k E\{b_i b_k\} p_2(t-iT) p_2(t-kT) \sin^2(2\pi\beta t) dt$$

$$E_S = \int_0^T \sum_i 4 P_1^2(t-iT) \cos^2(2\pi f t) dt + \int_0^T \sum_i P_2^2(t-iT) \sin^2(2\pi f t) dt =$$

$$= \frac{1}{2} \int_0^T \sum_i 4 P_1^2(t-iT) dt + \frac{1}{2} \int_0^T \sum_i P_2^2(t-iT) dt$$

$$\int_0^T \sum_i P(t-iT) dt = \sum_i \int_0^T P(t-iT) dt = \quad \alpha = t-iT$$

$$= \sum_i \int_{iT}^{iT+T} P(\alpha) d\alpha = \int_{-\infty}^{+\infty} P(\alpha) d\alpha$$

$$E_S = 4 \frac{1}{2} \int_{-\infty}^{+\infty} P_1^2(t) dt + \frac{1}{2} \int_{-\infty}^{+\infty} P_2^2(t) dt = 2 E_{P_1} + \frac{1}{2} E_{P_2}$$