



Table 7.1

Parameter Symbol	Parameter Description	Typical Parameter Value		Units
		n-Channel	p-Channel	
$V_{T0}$	Threshold voltage ( $V_{BS} = 0$ )	$0.7 \pm 0.15$	$-0.7 \pm 0.15$	V
$K'$	Transconductance parameter (in saturation)	$110.0 \pm 10\%$	$50.0 \pm 10\%$	$\mu\text{A}/\text{V}^2$
$\gamma$	Bulk threshold parameter	0.4	0.57	$\text{V}^{1/2}$
$\lambda$	Channel length modulation parameter	$0.04 (L = 1 \mu\text{m})$ $0.01 (L = 2 \mu\text{m})$	$0.05 (L = 1 \mu\text{m})$ $0.01 (L = 2 \mu\text{m})$	$\text{V}^{-1}$
$2 \phi_F $	Surface potential at strong inversion	0.7	0.8	V

7.1 In the op amp of Fig 7.1,  $(W/L)_{1-8}=100/0.5$ ,  $I_{SS}=1\text{mA}$ , and  $V_{b1}=1.7\text{V}$ . Assume that  $\gamma=\lambda=0$ .  $V_{DD}=3\text{V}$ .

- What is the maximum allowable input CM level?
- What is  $V_X$ ?
- What is the maximum allowable output swing if the gate of  $M_2$  is connected to the output? (make sure that  $M_2$  and  $M_4$  work in saturation region)
- What is the acceptable range of  $V_{b2}$ ?

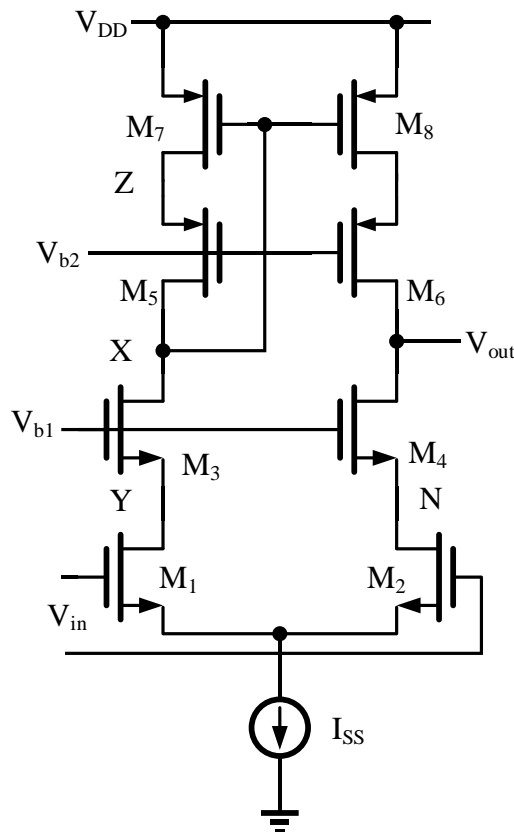


Fig 7.1

解:

(a)

$$\begin{aligned} V_{in,CM,max} &= V_Y + V_{TH1} = V_{b1} - V_{GS3} + V_{TH1} \\ &= V_{b1} - V_{OD3} - V_{TH3} + V_{TH1} = V_{b1} - V_{OD3} \end{aligned}$$

$$V_{OD3} = V_{GS3} - V_{TH} = \left( \frac{2I_{D3}}{K_n \left( \frac{W}{L} \right)_3} \right)^{\frac{1}{2}} = 0.213V$$

$$V_{in,CM,max} = 1.7 - 0.213 = 1.487V$$

(b)

$$\frac{1}{2}I_{SS} = \frac{1}{2}K_p \left( \frac{W}{L} \right)_7 (V_{DD} - V_X - |V_{TH7}|)^2$$

解得  $V_X = 1.984V$

(c)

M2 和 M4 均工作在饱和区，所以

$$V_{out} - V_{TH} \leq V_N$$

$$V_{out} \geq V_{b1} - V_{TH}$$

且有  $V_N = V_{b1} - V_{GS4}$

整理得  $V_{b1} - V_{TH} \leq V_{out} \leq V_{b1} - (V_{GS4} - V_{TH})$

$$\text{所以 } \Delta V_{swing} = 2V_{TH} - V_{GS4} = 2V_{TH} - V_{GS3} = 1.4 - 0.913 = 0.487V$$

(d)

$V_Z < V_X + V_{TH}$ ，且  $V_{b2} = V_Z - |V_{GS5}|$ ，所以

$$V_{b2} < V_X + V_{TH} - |V_{GS5}| = 1.984 + 0.7 - 1.016 = 1.668V$$

$$\text{且 } V_{b2} > V_X - V_{TH} = 1.984 - 0.7 = 1.284V$$

7.2 Suppose the circuit of Fig 7.2 is designed with  $I_{SS}$  equal to 1mA,  $I_{D9}$ - $I_{D12}$  equal to 0.5mA, and  $(W/L)_{9-12} = 100/0.5$ .  $V_{DD} = 3V$ ,  $\lambda = 0$ .

(a) What CM level is required at X and Y?

(b) If  $I_{SS}$  requires a minimum voltage of 400mV, choose the minimum dimensions of  $M_1$ - $M_8$  to allow a peak-to-peak swing of 200mV at X and at Y, assume that  $M_1$ - $M_4$  are of the same size and  $M_5$ - $M_8$  are of the same size.

(c) Give the expression of the overall voltage gain.

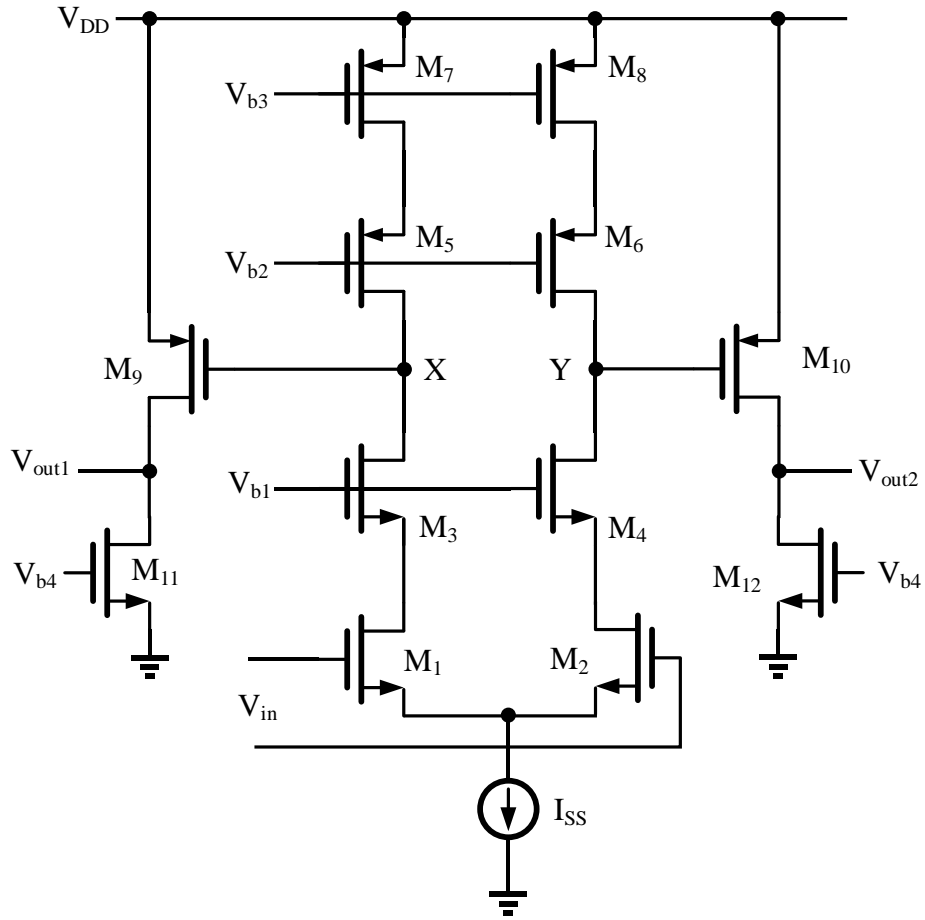


Fig 7.2

解:

(a)

$$|V_{GS9}| = |V_{TH}| + \left[ \frac{2I_{D9}}{K_p \left( \frac{W}{L} \right)_9} \right]^{\frac{1}{2}} = 1.016V$$

$$V_{XY,CM} = V_{DD} - |V_{GS9}| = 1.984V$$

(b)

$V_X$  摆幅为 200mV, 所以  $V_{Xmax} = 2.084V$ ,  $V_{Xmin} = 1.884V$

$$V_{OD7} = V_{OD5} = \frac{V_{DD} - V_{Xmax}}{2} = 0.458V$$

$$V_{OD1} = V_{OD3} = \frac{V_{Xmin} - V_{ISS}}{2} = 0.742V$$

$$\left(\frac{W}{L}\right)_{5-8} = \frac{I_{SS}}{K_p V_{OD5,7}^2} = 95.35$$

$$\left(\frac{W}{L}\right)_{1-4} = \frac{I_{SS}}{K_n V_{OD1,3}^2} = 16.51$$

(c)

$$A_V = g_{m1} (g_{m3} r_{o3} r_{o1} // g_{m5} r_{o5} r_{o7}) g_{m9} (r_{o9} // r_{o11})$$

7.3 Consider the amplifier of Fig 7.3, where  $(W/L)_{1-4}=50/0.5$  and  $I_{SS}=I_1=0.5\text{mA}$ .  $\lambda_p=0.2$  and  $\lambda_n=0.1$ . (此题图有所修改, 在答案最后)

(a) Estimate the poles at nodes X and Y by multiplying the small-signal resistance and capacitance to ground. Assume that  $C_X=C_Y=0.5\text{pF}$ . What is the phase margin for unity-gain feedback?

(b) If  $C_X=0.5\text{pF}$ , what is the maximum tolerable value of  $C_Y$  that yields a phase margin for  $60^\circ$  for unity-gain feedback?

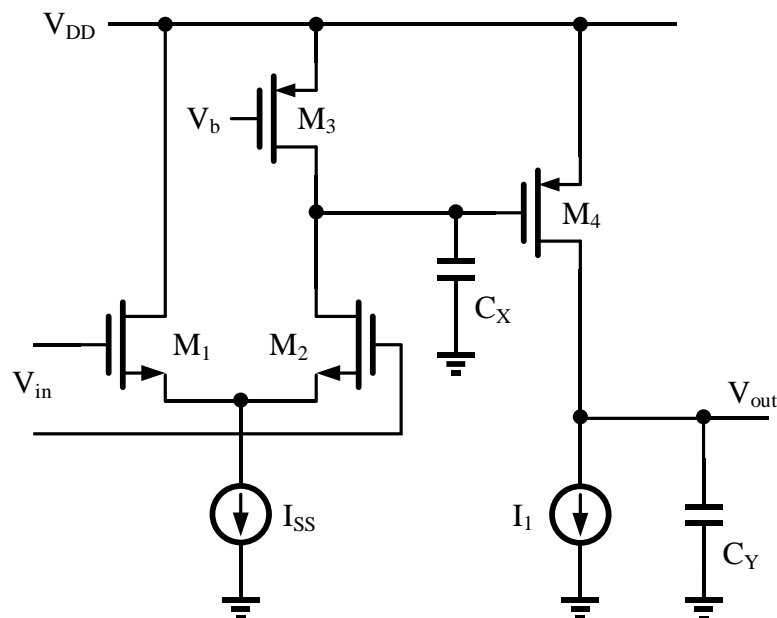


Fig 7.3

解:

(a)

$$r_{o3} = \frac{1}{\lambda_p I_{D3}} = 20k\Omega, \quad r_{o2} = \frac{1}{\lambda_n I_{D2}} = 40k\Omega$$

$$\omega_{px} = \frac{1}{C_X (r_{o2} // r_{o3})} = 150 \times 10^6 \text{ rad} / s$$

$$g_{m4} = \sqrt{2I_1 K_n \left(\frac{W}{L}\right)_4} = 3.3 \times 10^{-3}$$

$$\omega_{py} = \frac{1}{C_Y (g_{m4})^{-1}} = 6.6 \times 10^9 \text{ rad} / s$$

$$g_{m2} = \sqrt{I_{SS} K_n \left( \frac{W}{L} \right)_4} = 2.35 \times 10^{-3}$$

低频增益为:  $g_{m2} (r_{o2} // r_{o3}) = 31.33 \rightarrow 29.9 \text{ dB}$  (增益转换为 dB 模式)

所以增益交点对应频率为:  $20 \log \frac{\omega_1}{\omega_{px}} = 29.9$ ,  $\omega_1 = 4.69 \times 10^9 \text{ rad} / s$

$$PM = 180^\circ - \tan^{-1} \frac{\omega_1}{\omega_{px}} - \tan^{-1} \frac{\omega_1}{\omega_{py}} = 56.43^\circ$$

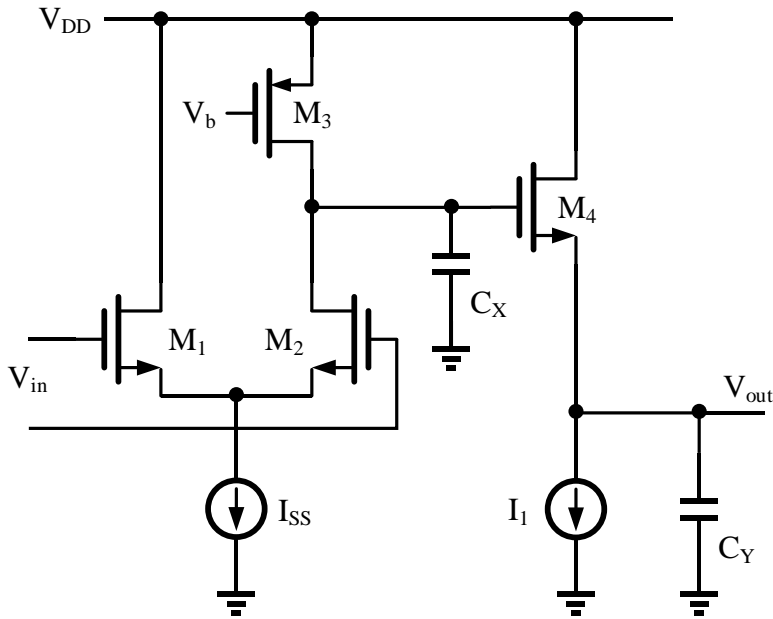
(b)

$$PM = 180^\circ - \tan^{-1} \frac{\omega_1}{\omega_{px}} - \tan^{-1} \frac{\omega_1}{\omega_{py}} = 60^\circ$$

设  $\omega_{px}$  不变, 解得  $\omega_{py} = 8.43 \times 10^9 \text{ rad} / s = \frac{1}{C_Y (g_{m4})^{-1}}$

解得  $C_Y = 391 \text{ fF}$

此题图应改为:



7.4 In the circuit of Fig.7.4, assume that  $(W/L)_{1-4} = 100/1$ ,  $C_1 = C_2 = 0.5$  pF, and  $I_{SS} = 1$  mA. Calculate the small-signal time constant( $\tau$ ) of the circuit (tips: the transfer function of the amp can be expressed by  $\frac{A}{1+\tau s}$  )

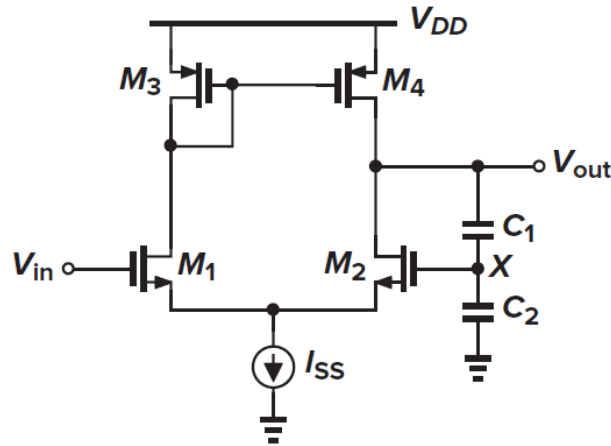
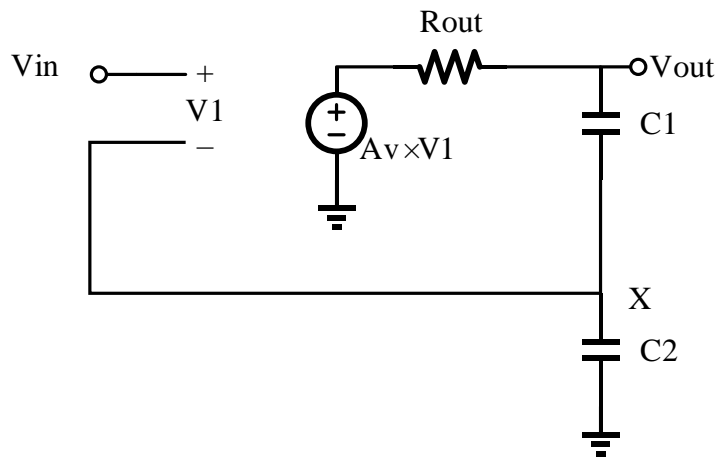


Fig 7.4

解:



$$A_v = g_{m1}(r_{o2}/r_{o4}) = 73.7$$

$$R_{out} = r_{o2}/r_{o4} = 22.2k$$

$$V_{in} = V_1 + V_x$$

$$V_x = V_{out} \frac{C_1}{C_1 + C_2}$$

$$V_{out} = A_v V_1 \left[ \frac{\frac{1}{C_1 // C_2 \cdot S}}{R_{out} + \frac{1}{C_1 // C_2 \cdot S}} \right] = A_v V_1 \left[ \frac{1}{C_1 // C_2 \cdot R_{out} \cdot S + 1} \right]$$

$$= \frac{A_v}{C_1 // C_2 \cdot R_{out} \cdot S + 1} (V_{in} - V_{out} \frac{C_1}{C_1 + C_2})$$

$$\frac{V_{out}}{V_{in}} = \frac{\frac{A_v}{1 + A_v \frac{C_1}{C_1 + C_2}}}{1 + \frac{(C_1 // C_2) R_{out} S}{1 + A_v \frac{C_1}{C_1 + C_2}}}$$

$$\tau = \frac{\frac{C_1 C_2}{C_1 + C_2} R_{out}}{1 + A_v \frac{C_1}{C_1 + C_2}} = 1.466 \times 10^{-10}$$