(1)

反馈网络通过密勒补偿电容 C_c 增大二级运放电路的相位裕度。

(2)

$$C_c \geq 0.22 C_L$$

$$\frac{g_{m6}}{g_{m2}}>$$
 10

$$GBW = rac{g_{m1}}{C_c}$$

$$SRpprox rac{I_5}{C_c}$$

$$A_d=rac{g_{m1}}{g_{ds2}+g_{ds4}}$$

$$A_2 = -g_{m6} rac{1}{g_{ds6} + g_{ds7}}$$

$$A_t = A_d A_2 = -rac{g_{m1}}{g_{ds2} + g_{ds4}} rac{g_{m6}}{g_{ds6} + g_{ds7}}$$

$$V_{cm,max} = V_{DD} - \sqrt{rac{I_5}{eta_3}} - V_{T03}(max) + |V_{T01}|(min)$$

$$V_{cm,min} = V_{SS} + V_{DS5,sat} + V_{GS1}$$

$$\therefore V_{cm,min} = V_{SS} + \sqrt{rac{I_5}{eta_1}} + V_{T01}(max) + V_{DS5,sat}$$

$$CMRR = rac{2g_{m1}g_{m3}}{(g_{ds2} + g_{ds4})g_{ds5}}$$

$$P = (V_{DD} - V_{SS})(I_5 + I_6)$$

(1) 选定补偿电容 C_c

$$C_c \geq 0.22 C_L \pm C_L = 10 pF$$

 $\therefore C_c = 3 pF$

(2) 计算 $(W/L)_3$ 、 $(W/L)_4$

$$SRpprox rac{I_5}{C_c} \ dots I_5 = C_c \cdot (SR) = 30 \mu A \ (W/L)_3 = rac{I_5}{\mu_n C_{ox} [V_{DD} - V_{in}(max) - V_{T03}(max) + |V_{T01}|(min)]^2} \ dots \mu_n C_{ox} = 110 \mu A/V^2, V_{DD} = 2.5 V, \ V_{T03}(max) = 0.85 V, |V_{T01}|(min) = 0.55 V \ dots (W/L)_3 = (W/L)_4 pprox 6.82 \ rac{lpha(W/L)_3}{2} = (W/L)_4 = 7$$

(3) 计算 $(W/L)_1$ 、 $(W/L)_2$

$$g_{m1}pprox 94.3 \mu s \ (W/L)_1=(W/L)_2=rac{g_{m1}^2}{2\mu_p C_{ox} I_1}pprox 5.92 \ rac{\mathbb{R}(W/L)_1=(W/L)_2=6}$$

(4) 计算 $(W/L)_5$

$$V_{DS5,sat} = V_{cm,min} - V_{SS} - V_{GS1} = -0.33V$$
 $(W/L)_5 pprox 11$

(5) 计算 $(W/L)_6$

$$g_{m6} \geq 10 g_{m2}$$
 取 $g_{m6} = 943 \mu s$ $g_{m4} = 2 \sqrt{rac{110 imes 10^{-6}}{2} imes 7 imes (15 imes 10^{-6})} pprox 152 \mu s$ 取 $(W/L)6 = 44$

(6) 计算 $(W/L)_7$

$$I_6 = rac{g_{m6}^2}{2\mu_n C_{ox} \cdot (W/L)_6} \ dots I_6 = 93.04 \mu A \ (W/L)_7 = (W/L)_5 imes rac{I_6}{I_5} pprox 34$$

(7) 计算 $(W/L)_8$

设电流源电流为 $45\mu A$

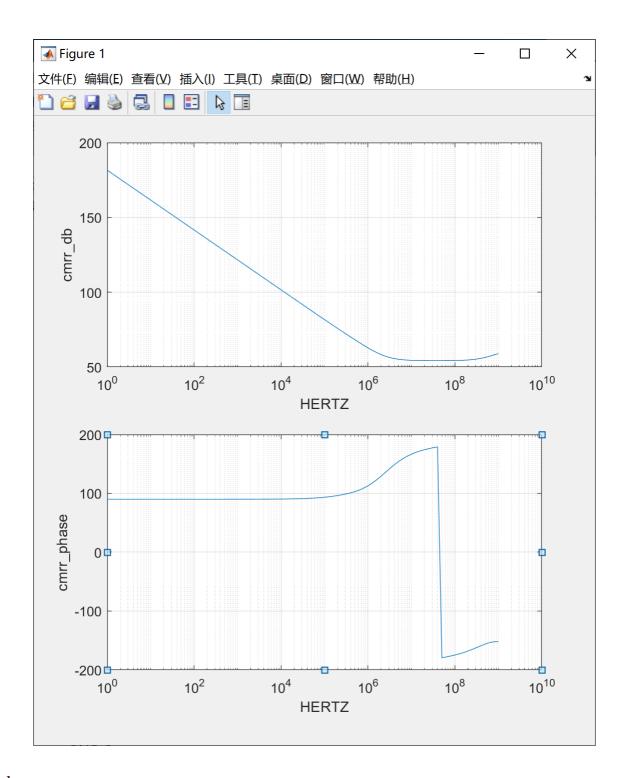
$$(W/L)_8 = (W/L)_5 = 11$$

$$\therefore A_t = A_d A_2 = rac{g_{m1}}{g_{ds2} + g_{ds4}} rac{g_{m6}}{g_{ds6} + g_{ds7}} pprox 78 dB$$

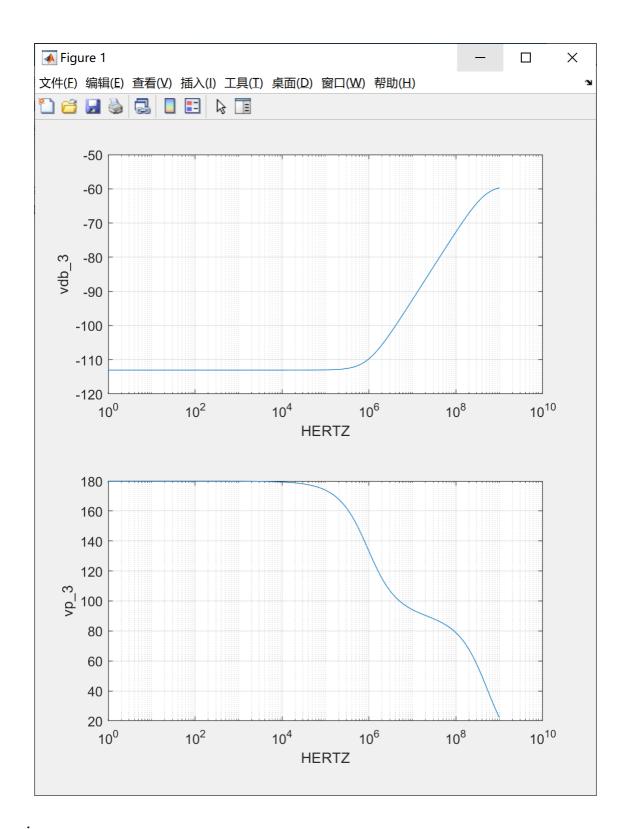
 $CMRR\approx 83dB$

仿真结果

CMRR



open_loop



unity_gain

