

Capacitance Measurement



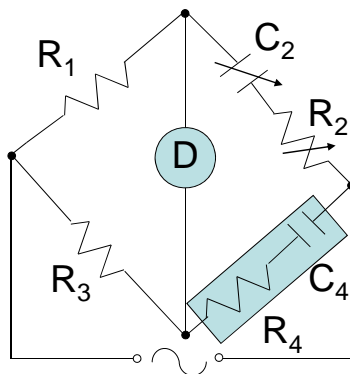
EIE 240 Electrical and Electronic Measurement
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Series Capacitance Bridge

Capacitor can be considered to be a pure capacitance in series with, or sometimes in parallel with, a pure resistance.



Impedances,

$$Z_1 = R_1$$

$$Z_2 = R_2 - j/\omega C_2$$

$$Z_3 = R_3$$

$$Z_4 = R_4 - j/\omega C_4$$

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Series Capacitance Bridge (Cont'd)

Balanced bridge,

$$Z_4 = Z_2 Z_3 / Z_1$$

$$R_4 - j/\omega C_4 = (R_2 - j/\omega C_2) R_3 / R_1$$

$$= R_2 R_3 / R_1 - j(R_3 / \omega C_2 R_1)$$

Real part: $R_4 = R_2 R_3 / R_1$

Imagination part: $C_4 = C_2 R_1 / R_3$

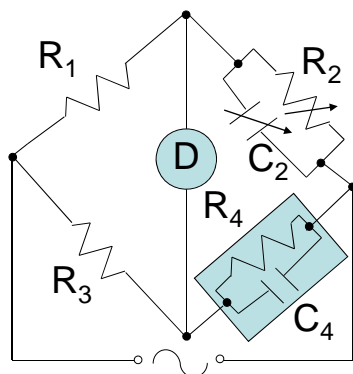
Used for low D = 0.001-0.1

$$D = 1/Q = \omega R_4 C_4 = \omega R_2 C_2$$

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Parallel Capacitance Bridge

Used for D = 0.05-50



Impedances,

$$Z_1 = R_1$$

$$Z_2 = 1 / (1/R_2 + j\omega C_2)$$

$$= R_2 / (1 + j\omega C_2 R_2)$$

$$Z_3 = R_3$$

$$Z_4 = R_4 / (1 + j\omega C_4 R_4)$$

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Parallel Capacitance Bridge (Cont'd)

Balanced bridge,

$$Z_4 = Z_2 Z_3 / Z_1$$

$$R_4 / (1 + j\omega C_4 R_4) = R_2 R_3 / (1 + j\omega C_2 R_2) R_1$$

$$R_1 R_4 + j\omega C_2 R_1 R_2 R_4 = R_2 R_3 + j\omega C_4 R_2 R_3 R_4$$

Real part: $R_1 R_4 = R_2 R_3$

$$R_4 = R_2 R_3 / R_1$$

Imagination part: $C_2 R_1 R_2 R_4 = C_4 R_2 R_3 R_4$

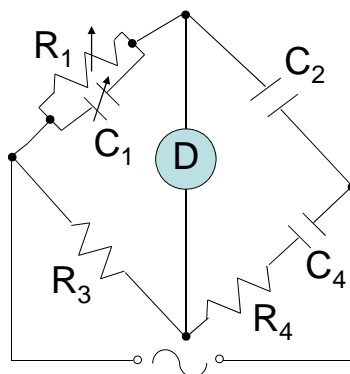
$$C_4 = C_2 R_1 / R_3$$

$$D = \omega R_4 C_4 = \omega R_2 C_2$$

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Schering Bridge

Used for very low D



Impedances,

$$Z_1 = R_1 / (1 + j\omega C_1 R_1)$$

$$Z_2 = 1 / j\omega C_2$$

$$Z_3 = R_3$$

$$Z_4 = R_4 - j / \omega C_4$$

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Schering Bridge (Cont'd)

Balanced bridge,

$$\begin{aligned} Z_4 &= Z_2 Z_3 / Z_1 \\ R_4 - j/\omega C_4 &= R_3(1+j\omega C_1 R_1) / j\omega C_2 R_1 \\ &= (\omega C_1 R_1 R_3 - jR_3) / \omega C_2 R_1 \\ &= R_3 C_1 / C_2 - j(R_3 / \omega R_1 C_2) \end{aligned}$$

Real part: $R_4 = R_3 C_1 / C_2$

Imagination part: $1/C_4 = R_3 / R_1 C_2$

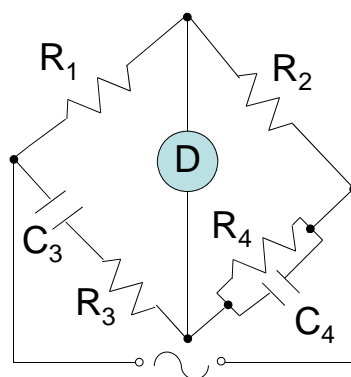
$$C_4 = C_2 R_1 / R_3$$

$$D = \omega R_4 C_4 = \omega R_1 C_1$$

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Wien Bridge

Used as frequency-dependent circuit



Impedances,

$$Z_1 = R_1$$

$$Z_2 = R_2$$

$$Z_3 = R_3 - j/\omega C_3$$

$$Z_4 = R_4 / (1 + j\omega C_4 R_4)$$

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Wien Bridge (Cont'd)

Balanced bridge,

$$Z_4 = Z_2 Z_3 / Z_1$$

$$R_4 / (1 + j\omega C_4 R_4) = R_2 (R_3 - j/\omega C_3) / R_1$$

$$R_1 R_4 / R_2 = R_3 + R_4 C_4 / C_3 + j(\omega C_4 R_3 R_4 - 1/\omega C_3)$$

Imagination part: $\omega C_4 R_3 R_4 = 1/\omega C_3$

$$C_4 R_4 = 1/\omega^2 C_3 R_3$$

Real part: $R_1 R_4 / R_2 = R_3 + R_4 C_4 / C_3$

$$R_4 = (R_3 R_2 C_3 + R_2 R_4 C_4) / R_1 C_3$$

$$= (R_3 R_2 C_3 + R_2 / \omega^2 C_3 R_3) / R_1 C_3$$

$$= R_2 (\omega^2 C_3^2 R_3^2 + 1) / (\omega^2 C_3^2 R_1 R_3)$$

and $C_4 = 1/\omega^2 C_3 R_3 R_4$

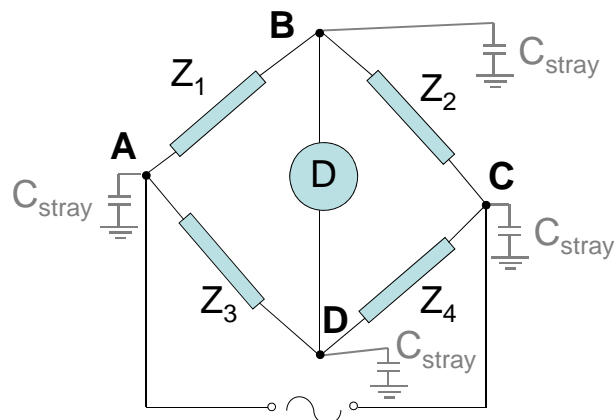
$$= C_3 R_1 / R_2 (\omega^2 C_3^2 R_3^2 + 1)$$

$$D = \omega R_4 C_4 = 1/\omega R_3 C_3$$

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Stray Impedance

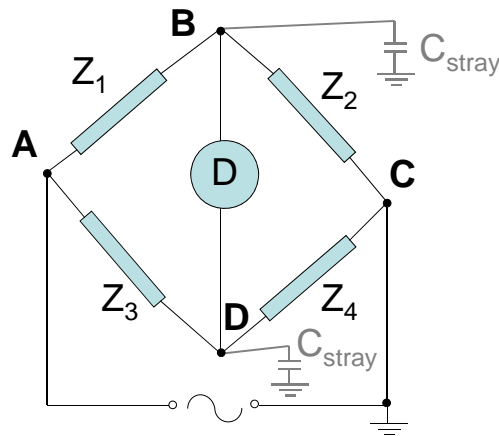
There are stray capacitances between the various element and the ground and it may affect bridge balance.



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Stray Impedance (Cont'd)

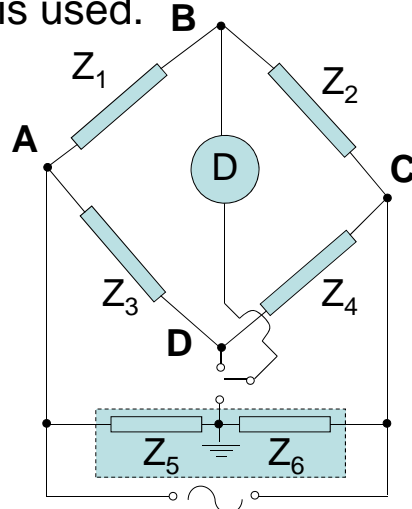
The stray capacitances can be reduced by earthing one side of AC supply.



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Stray Impedance (Cont'd)

To minimize stray capacitances between the detector terminals and earth, Wagner earth is used.

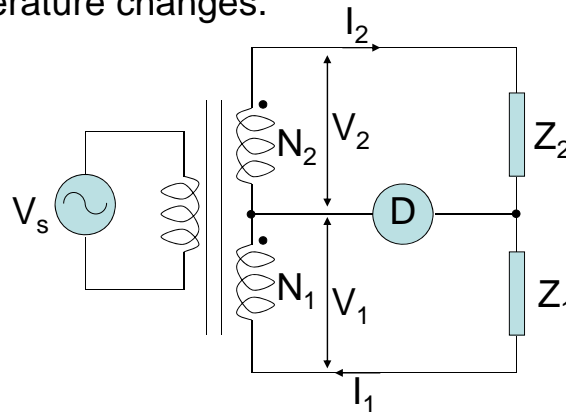


To ensuring that the points B and D of a balanced bridge are at ground potential

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Transformer Ratio Bridges

Not only varying the impedances, but bridge can be also balanced by varying the turns ratio of a transformer. There is a small number of standard resistors and capacitors and no effect of temperature changes.



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Single Ratio Transformer Bridge

Tap a transformer \Rightarrow voltage divider of V_s

$$V_1 = kN_1 = I_1 Z_1 \Rightarrow I_1 = kN_1 / Z_1$$

$$V_2 = kN_2 = I_2 Z_2 \Rightarrow I_2 = kN_2 / Z_2$$

To balance the bridge or no current through the detector, $\textcircled{D} = \text{Null}$

$$I_1 = I_2 \Rightarrow Z_1 / Z_2 = N_1 / N_2$$

Impedance Ratio	Turn Ratio
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Single Ratio Transformer Bridge (Cont'd)

- Resistance Measurement

Z_1 = Unknown resistor R_x

Z_2 = Standard resistor R_s

$$R_x = R_s \frac{N_1}{N_2}$$

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Single Ratio Transformer Bridge (Cont'd)

- Capacitance Measurement

Z_1 = Unknown $C_x \parallel R_x$ (leakage resistance)

$$= 1/(1/R_x + j\omega C_x)$$

$$= R_x / (1 + j\omega R_x C_x)$$

Z_2 = Standard $C_s \parallel R_s$

$$= R_s / (1 + j\omega R_s C_s)$$

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Single Ratio Transformer Bridge (Cont'd)

- Capacitance Measurement (Cont'd)

Balanced, $Z_1 / Z_2 = N_1 / N_2$

$$1/Z_1 = (N_2/N_1) 1/Z_2$$

$$(1+j\omega R_x C_x)/R_x = (N_2/N_1) (1+j\omega R_s C_s)/R_s$$

$$1/R_x + j\omega C_x = (N_2 / N_1 R_s) + j\omega C_s N_2/N_1$$

Real part: $R_x = R_s N_1/N_2$

Imagination part: $C_x = C_s N_2/N_1$

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Single Ratio Transformer Bridge (Cont'd)

- Inductance Measurement

$$Z_1 = \text{Unknown } L_x \parallel R_x$$

$$= 1/(1/R_x + 1/j\omega L_x)$$

$$= 1/(1/R_x - j/\omega L_x)$$

$$Z_2 = \text{Standard } C_s \parallel R_s$$

$$= 1/(1/R_s + j\omega C_s)$$

$$= 1/(1/R_s - j\omega C_s) \quad , \text{ Reversed Current}$$

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Single Ratio Transformer Bridge (Cont'd)

- Inductance Measurement (Cont'd)

Balanced, $1/Z_1 = (N_2/N_1) 1/Z_2$

$$1/R_x - j/\omega L_x = (N_2/N_1) (1/R_s - j\omega C_s)$$

Real part: $R_x = R_s N_1/N_2$

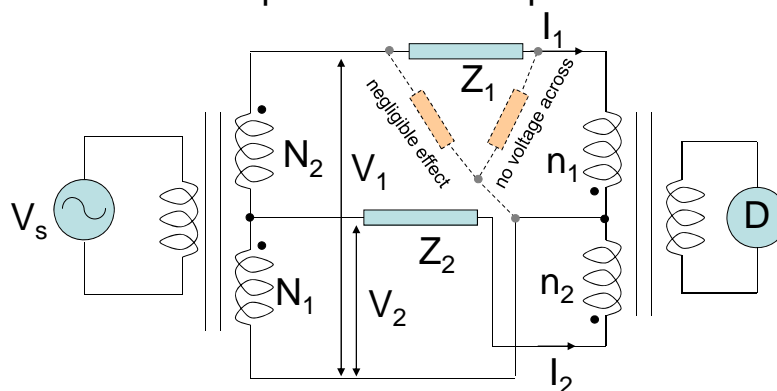
Imagination part: $1/\omega L_x = (N_2/N_1)\omega C_s$

$$L_x = N_1 / N_2 \omega^2 C_s$$

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Double Ratio Transformer Bridge

To measure the impedance of components *in Situ*.



$$I_1 = V_1 / Z_1 = k(N_1 + N_2)/Z_1$$

$$I_2 = V_2 / Z_2 = kN_2/Z_2$$

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Double Ratio Transformer Bridge (Cont'd)

Balanced, null current or zero magnetic flux,

$$n_1 I_1 = n_2 I_2$$

$$n_1(N_1+N_2)/Z_1 = n_2 N_2/Z_2$$

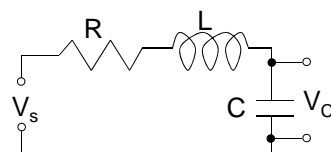
$$Z_1 = Z_2 \frac{n_1(N_1+N_2)}{n_2 N_2}$$

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Q-Meter

RLC Series Resonance

$$\begin{aligned} Z &= R + j\omega L + 1/j\omega C \\ &= R + j(\omega L - 1/\omega C) \end{aligned}$$

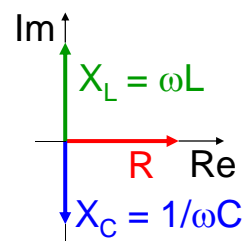


Resonant frequency

$$\omega_0 L = 1/\omega_0 C$$

$$\omega_0^2 = 1/LC$$

$$\omega_0 = 1/\sqrt{LC}$$



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Q-Meter (Cont'd)

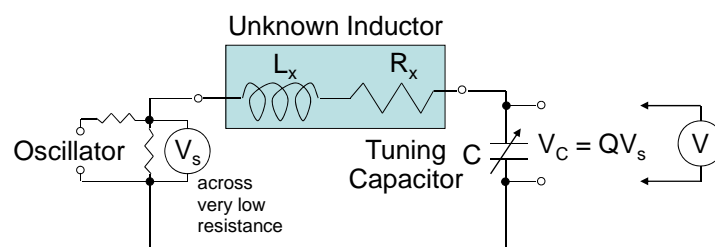
$$\begin{aligned}
 I_0 &= V_s / R \\
 V_C &= I_0 X_C \\
 &= (V_s / R)(1 / \omega_0 C) \\
 &= (1 / \omega_0 RC) V_s \\
 &= Q V_s \\
 &\propto Q
 \end{aligned}$$

where $Q = \text{Reactance/Resistance}$

$$\begin{aligned}
 &= \omega_0 L / R \\
 &= \omega_0 (1 / \omega_0^2 C) / R \\
 &= 1 / \omega_0 RC \\
 &\text{(Unloaded)}
 \end{aligned}$$

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Q-Meter (Cont'd)



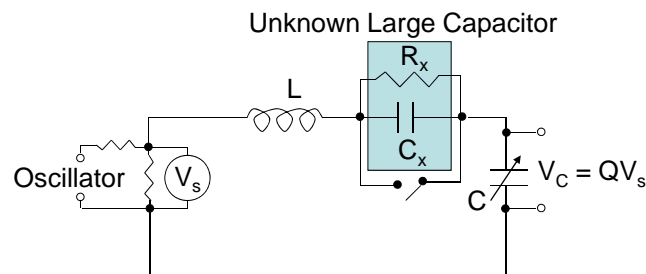
Tuning to resonance, $\omega_0 = 1 / \sqrt{LC}$

$$L_x = 1 / \omega_0^2 C$$

$$R_x = 1 / Q \omega_0 C$$

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Q-Meter: Low Impedance Measurement



Short circuit \Rightarrow tuning $C = C_1$, $L_1 = L$, R_1

$$Q_1 = 1/\omega_0 R_1 C_1$$

$$R_1 = 1/\omega_0 C_1 Q_1$$

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Q-Meter: Low Impedance Measurement (Cont'd)

Remove short circuit \Rightarrow then tuning $C = C_2$

$$C_2 \text{ \& } C_x = C_1, L_2 = L, R_2 = R_1 \text{ \& } R_x$$

$$\text{Series capacitors, } C_1 = 1 / (1/C_x + 1/C_2)$$

$$= C_2 C_x / (C_x + C_2)$$

$$C_1 C_x + C_1 C_2 = C_2 C_x$$

$$C_x = C_1 C_2 / (C_2 - C_1)$$

$$Q_2 = 1/\omega_0 R_2 C_2$$

$$R_2 = 1/\omega_0 C_2 Q_2$$

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Q-Meter: Low Impedance Measurement (Cont'd)

$$R_1 = R_2 - R_x$$

$$R_x = R_2 - R_1 \quad , \text{ Leakage resistance}$$

$$= 1/\omega_0 C_2 Q_2 - 1/\omega_0 C_1 Q_1$$

$$= (C_1 Q_1 - C_2 Q_2) / (\omega_0 C_1 C_2 Q_1 Q_2)$$

$$Q_x = 1/\omega_0 R_x C_x$$

$$= (\omega_0 C_1 C_2 Q_1 Q_2)(C_2 - C_1) /$$

$$\omega_0 (C_1 Q_1 - C_2 Q_2) C_1 C_2$$

$$= Q_1 Q_2 (C_2 - C_1) / (C_1 Q_1 - C_2 Q_2)$$

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Q-Meter: Low Impedance Measurement (Cont'd)

If the unknown component is an inductor,

$$L_x = 1/\omega_0^2 C_x$$

$$= (C_2 - C_1) / \omega_0^2 C_1 C_2$$

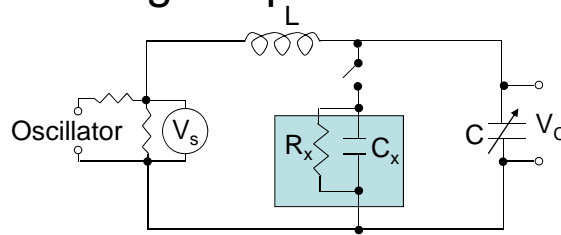
If the unknown component is a pure resistor
(no reactance),

$$R_x = (C_1 Q_1 - C_2 Q_2) / (\omega_0 C_1 C_2 Q_1 Q_2)$$

$$= (Q_1 - Q_2) / \omega_0 C_1 Q_1 Q_2 \quad , C_1 = C_2$$

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Q-Meter: High Impedance Measurement



For high resistance, inductance > 100mH, or capacitance < 400 pF

Open circuit and tune $C = C_1$, $L_1 = L$, R_1

Then short circuit and tune $C = C_2$

$$C_2 // C_x = C_1, L_2 = L, R_2 = R_x // R_1$$

$$C_x + C_2 = C_1 \Rightarrow C_x = C_1 - C_2$$

$$R_2 = R_x // R_1 = R_x R_1 / (R_x + R_1) \Rightarrow R_x = R_1 R_2 / (R_1 - R_2)$$

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Q-Meter: High Impedance Measurement (Cont'd)

Parallel RLC $\Rightarrow Q = \omega_0 RC$ (loaded)

$$R_1 = Q_1 / \omega_0 C_1 \text{ and } R_2 = Q_2 / \omega_0 C_2$$

Therefore

$$\begin{aligned} R_x &= Q_1 Q_2 \omega_0 C_1 C_2 / \omega_0^2 C_1 C_2 (Q_1 C_2 - Q_2 C_1) \\ &= Q_1 Q_2 / \omega_0 (Q_1 C_2 - Q_2 C_1) \end{aligned}$$

$$\begin{aligned} Q_x &= \omega_0 R_x C_x \\ &= \omega_0 Q_1 Q_2 (C_1 - C_2) / \omega_0 (Q_1 C_2 - Q_2 C_1) \\ &= Q_1 Q_2 (C_1 - C_2) / (Q_1 C_2 - Q_2 C_1) \end{aligned}$$

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Q-Meter: High Impedance Measurement (Cont'd)

For unknown inductance,

$$\begin{aligned} L_x &= 1/\omega_0^2 C_x \\ &= 1/\omega_0^2 (C_1 - C_2) \end{aligned}$$

For pure resistance,

$$\begin{aligned} R_x &= Q_1 Q_2 / \omega_0 (Q_1 C_2 - Q_2 C_1) \\ &= Q_1 Q_2 / \omega_0 C_1 (Q_1 - Q_2) \quad , C_1 = C_2 \end{aligned}$$

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