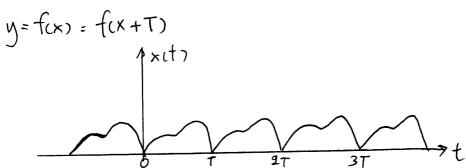


- Fourier Series is named after a French mathematical Physicist by the name "Jean Baptiste Joseph, Baron de Fourier (March 21, 1768-Many 16, 1830".

- He stated that "no matter how complicated it is, a wave that is periodic, with a pattern that repeats itself, consists of the sum of many simple waves."

- The concept of The Fourier series is based on representing periodic signal as a sum of harmonically related sinusoidal functions. With the aid of Fourier series, complex periodic wantorms can be represented in terms of sinusoidal functions, whose properties are familiar to us.

- Periodic function



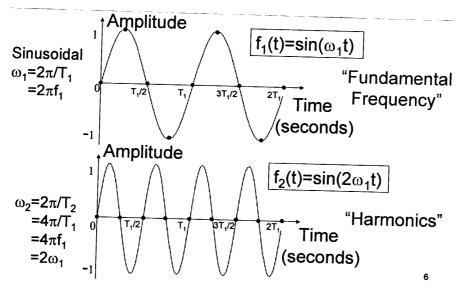
This signal maybe represented by the sum of a series of sine and/or cosine functions plus a dc term. The resulting series is called a Fourier series.

The lowest frequ of the sinusoidal components is a frequent, given by

This frequestion is the day of will be interest multiple of

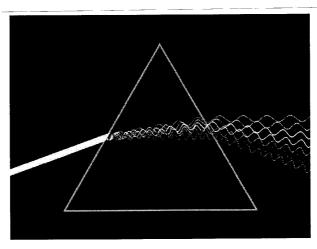
All other frequent the signal will be integer multiples of the fundamental. These various components are referred to as harmonics, with the order of a given harmonic indicated by the ratio of its frequent to the fundamental frequency to the fundamental frequency.

Example



hote: angular velocity:  $\omega = \frac{\theta}{t}$  [rad/s]  $rad = \frac{\pi}{180} \times degree$ 

Example Prism = Fourier transform

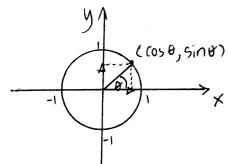


Any transmission system through which a given signal passes must have a bandwidth sufficiently large to pass all significant freeze of the signal. In a purely mathematical sense, many common waveforms theoretically contains an infinite number of harmonics. So, must we need an infinite bandwidth to process such signals?

So, the goal is to write this known signal fet, (or xit)) in time domain fct) = ao + \(\frac{2}{n=1}\) [an(os(nwt) + bnsin(nwt)] = ao + a, cos(wt) + b, sincut)

+ 
$$a_2(os(awt) + b_2sin(awt) + \dots$$
  
+  $a_2(os(awt) + b_2sin(awt) + \dots$ 

so why sine and cosine 9 Because they are "sinusoidal basis functions" in Euclidean space (R).



their inner product costs. Sint = 0 they are orthogonal to each other

so, if we know ao, a., ...an, b., .... bn, we should be able to write our fct) as an infinite sum of sine & cosine functions. So, how do we find them?

use these formulas:

here Tormolus:
$$a_0 = \frac{1}{T} \int_0^T x(t) dt = \frac{\text{area under curve in one cycle}}{\text{period } T}$$

$$a_n = \frac{2}{T} \int_0^T x(t) \cos(n\omega t) dt$$

$$b_n = \frac{2}{T} \int_0^T x(t) \sin(n\omega t) dt$$

To prove such relations, we first review some important trigonometric

#### Trigonometric Formulas

- $sin(-\theta) = -sin\theta$
- $cos(-\theta) = cos\theta$
- $sin(\theta_1 \pm \theta_2) = sin\theta_1 cos\theta_2 \pm sin\theta_2 cos\theta_1$
- $\cos(\theta_1 \pm \theta_2) = \cos\theta_1 \cos\theta_2 \mp \sin\theta_1 \sin\theta_2$
- $\sin^2\theta + \cos^2\theta = 1$
- $tan^2\theta + 1 = sec^2\theta$
- $1 + \cot^2\theta = \csc^2\theta$
- $\sin 2\theta = 2\sin\theta\cos\theta$
- $\cos 2\theta = \cos^2 \theta \sin^2 \theta = 2\cos^2 \theta 1 = 1 2\sin^2 \theta$
- $e^{j\theta} = \cos\theta + j \sin\theta$  where  $j = \sqrt{-1}$
- $\sin\theta = 1/2j (e^{j\theta} e^{-j\theta})$
- $\cos\theta = 1/2 (e^{j\theta} + e^{-j\theta})$
- For sinusoidal signal,  $\omega = 2\pi/T = 2\pi f$

Integrate both sides
$$\int_{0}^{T} f(t) dt = \int_{0}^{T} a_{0} dt + \sum_{n=1}^{\infty} \left[ \int_{0}^{T} a_{n}(os(n\omega t)dt + \int_{0}^{T} b_{n} sin(n\omega t)dt) \right]$$

$$= a_{0} \int_{0}^{T} dt + \sum_{n=1}^{\infty} \left[ a_{n} \int_{0}^{T} cos(n\omega t)dt + b_{n} \int_{0}^{T} sin(n\omega t)dt \right]$$

$$= a_{0} \int_{0}^{T} dt + \sum_{n=1}^{\infty} \left[ a_{n} \times 0 + b_{n} \times 0 \right]$$

= 
$$a_0 = \frac{a_0 T}{T} \int_0^T f(t) dt$$

[Finding an] fet) = ao + 2 [an (os (nwt) + bnsin (nwt)] -multiply both sides by coscut) fct) (05 (wt) = a, (052 (wt) + b, sin(wt) (05 (wt) + a= (05 (2wt) (05 (wt) + b= sin (2 wt) (05 (wt)  $\int_{0}^{\infty} \int_{0}^{\infty} \int_{0$ - integrate both sides + a2 ( T cos (2 wt) cos ( wt) dt + b2 ( T sin(2 wt) (os ( wt) dt Lit's look at each of these terms on the right (1) ao stos cut, at = 0 (2)  $a_1 \int_{0}^{T} \cos^2(\omega t) dt = a_1 \int_{0}^{T} \frac{(1+(os(2\omega t)))}{2} dt = a_1 \left[ \int_{0}^{T} dt + \int_{0}^{T} \cos(2\omega t) dt \right]$  $= \frac{\alpha_1}{9} \left[ T + 0 \right] = \frac{\alpha_1 T}{9}$ 3 b, JT sin(wt)(os(wt)dt = bix L JT sin(wt) d(sinwt) = bi sin (wt) | T = 0 ( a2) Teoscowtocoscovtodt = a2 ( [1-2sin2covto]desincot) = a= [ (Td(sin(wt)) - 2 ) sin2(wt)d(sin(wt))] = ar [ sin(wt)] T - 2 sin3(wt)] T  $= \frac{g_2(0)}{\omega}(0) = 0$  = 0  $= \int_0^T \sin(2\omega t) \cos(\omega t) dt = b2 \times 2 \int_0^T \sin(\omega t) \cos(\omega t) \cos(\omega t) dt$ = 2b2 /Toin(wt) cos 2cout) dt = -2b2 /Tcos 2cout) d(coscort)  $= -\frac{ab_2}{3(1)} (o3(wt))T = 0$ 

G 0

(F)

Therefore,

$$\int_{0}^{T} \sin(3\omega t) \cos(\omega t) dt = 0$$
Therefore,

$$\int_{0}^{T} \text{fct}_{2}(\cos(\omega t)) dt = a_{0}(0) + a_{1}(\frac{T}{2}) + b_{1}(0) + a_{2}(0) + b_{2}(0) + \dots$$

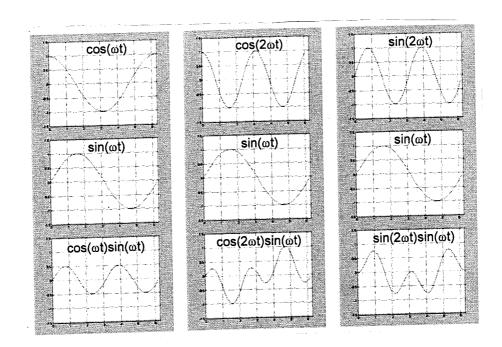
$$a_{1} = \frac{a_{1}}{T} \int_{0}^{T} \text{fct}_{2}(\cos(\omega t)) dt$$
From their result, we conclude:

$$a_{1} = \frac{a_{1}}{T} \int_{0}^{T} \text{fct}_{3}(\cos(\omega t)) dt$$
Finding by

$$f(t) = a_{0} + \sum_{n=1}^{\infty} [a_{n}(\cos(n\omega t) + b_{n}\sin(n\omega t))]$$
multiply both sides with sincusty and integrate:

Finding by  $f(t) = a_0 + \sum_{n=1}^{\infty} \left[ a_n (os(nwt) + b_n sin(nwt)) \right]$ multiply both side with sin(wt) and integrate:  $f(t) sin(wt) = a_0 sin(wt) + a_1 (os(wt) + in(wt) + b_1 sin^2(wt) + a_2 (os(2wt) + sin(wt) + b_2 sin(2wt) + sin(wt) + a_2 (os(2wt) + sin(wt) + a_2 + a_$ 

(b)



some good points to note: For mand nanintegers:

( sin (mwt) cos (nwt) dt = 0

```
\int_{-\infty}^{\infty} \sin(m\omega t) \sin(n\omega t) dt = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \cos((m-n)\omega t) - \cos((m+n)\omega t) dt
           ( 00 LOS ( b, - 02) = (050, LOS 02 + SIN 4, SIN b)
                                         (0) (4,+ 62) = (0)6, (0) 6, - 9/n 4, Sin 62 )
      \frac{1}{2} \int_{0}^{T} \cos((m-n)\omega t) - (\cos((m+n)\omega t)) dt = \frac{1}{2} \frac{x}{2\omega(m-n)} \sin((m-n)\omega t) \int_{0}^{T} \frac{1}{2\omega((m+n)\omega t)} dt
                                                                                                                                                                                                                                                                                                    = \frac{1}{2\omega} \left[ \frac{\sin(m-h)\omega T}{(m-h)} - \frac{\sin(m+h)\omega T}{(m+h)} \right]
= \frac{1}{2\omega} \left[ \frac{\sin(2\pi (m-n))}{(m-n)} - \frac{\sin(2\pi (m+n))}{(m+n)} \right]
\text{(D) For m = n; the whole thing is 0 is sina <math>\pi k = 0 (k = integer)

\text{(D) For m = n; the terms } \frac{\sin(m-n)}{(m-n)} = \frac{0}{0} \quad \text{(m) where } \frac{1}{1+0} = \frac{1}{1+0} =
   Ut m = n = x; \frac{\sin 2\pi x}{x} \Rightarrow \frac{d(\sin 2\pi x)/dx}{dx} = \frac{(\cos 2\pi x)^{x} 2\pi}{1} = 2\pi
                                                                      \frac{1}{2\omega} \times \left[2\pi - 0\right] = \frac{\pi}{\omega} = \frac{\pi}{2\pi f} = \frac{1}{2}
```

Decomposition: we know fet) and need to find Fourier coefficients 
$$a_n$$
,  $b_n$ 

$$a_n = \frac{2}{T} \int_0^T f(t) \cos(2\pi nt/T) dt = \frac{2}{T} \int_{t=0}^T f(t) \cos(2\pi nt/T) \Delta t$$

$$b_n = \frac{2}{T} \int_0^T f(t) \sin(2\pi nt/T) dt = \frac{2}{T} \int_{t=0}^T f(t) \sin(2\pi nt/T) \Delta t$$

let's do some examples:

There are 3 forms of Fourier Series:

- sine cosine form
- amplitude phase form
- complex exponential form
- 1) sine-cosine form: s we just studied a faw minutes ago

① amplitude phase form: 
$$x(t) = C_0 + \sum_{n=1}^{\infty} c_n (os(nwt + \phi_n))$$

or  $x(t) = C_0 + \sum_{n=1}^{\infty} c_n sin(nwt + \phi_n)$ 

where  $c_n = \sqrt{a_n + b_n}$ ,  $c_0 = a_0$ 

$$\phi_n = tan^{-1} \left( \frac{b_n}{a_n} \right)$$
 $\theta_n = \psi_n + 90^{\circ}$ 

(3) Complex Exponential form:

$$x(t) = \sum_{n=-\infty}^{\infty} \overline{X}_n e^{jn\omega t}$$

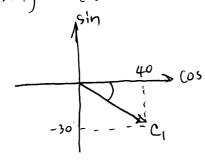
$$\overline{X}_0 = a_0 = C_0 ; \overline{X}_n = \underline{a_n - j\beta_n} \text{ for } n \neq 0$$

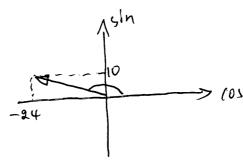
$$\overline{X}_n = X_n e^{j\phi_n}$$

xit) = 18 + 40 cos 2000 Tit - 30 sin 2000 Tit -24 (05 4000TT + 105 in 4000TT + Express the signal in (a) amplitude-phase form (b) complex exponential form

a) ... The amplitude phase form should be 
$$x(t) = 18 + C_1 \cos(2000\pi t + \phi_1) + C_2(0) (4000\pi t + \phi_2)$$
 or 
$$x(t) = 18 + C_1 \sin(2000\pi t + \phi_1) + C_2 \sin(4000\pi t + \phi_2)$$

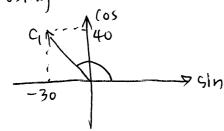
using the cosine as basis function



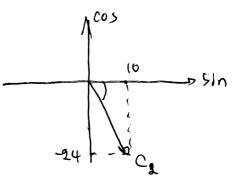


×(t) = 18 + 50 (05 (2000 nt - 36.87) + 26 (05 (4000 nt + 157.38))

using the sme as basis function



$$\overline{C}_1 = -30 + 40j = 50 \angle 126.87^{\circ}$$
 $\overline{C}_2 = 10 - j24 = 26 \angle -67.38^{\circ}$ 

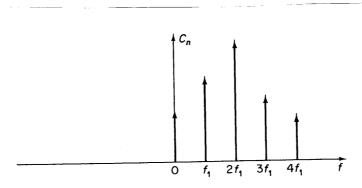


-- xct) = (8+50 sin (2000 Tt + 126.87°) + 26 sin (4000 Tt -67.38°)

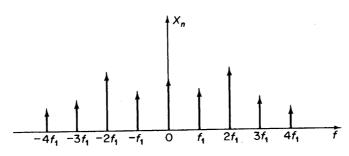
b) (umplex Exponential torm  $\overline{X}_{1} = \frac{40 - j(-30)}{2} = 20 + j15 = 25 \angle 36.87^{\circ}$   $\overline{X}_{-1} = \text{complex conjugate of } \overline{X}_{1} = 25 \angle -36.87^{\circ} = 25 - j15$   $\overline{X}_{2} = \frac{-94 - j10}{2} = -12 - j5 = 13 \angle -157.38^{\circ}$   $\overline{X}_{-2} = \text{cumplex conjugate of } \overline{X}_{2} = -12 + j5 = 13 \angle 157.38^{\circ}$   $\overline{X}_{-2} = \text{cumplex conjugate of } \overline{X}_{2} = -12 + j5 = 13 \angle 157.38^{\circ}$   $\overline{X}_{-1} = 18 + 25 e^{j} (2000\pi t + 36.87^{\circ}) + 25 e^{j} (2000\pi t + 36.87^{\circ})$   $+ 13 e^{j} (4000\pi t - 157.38^{\circ}) + 13 e^{j} (4000\pi t - 157.38^{\circ})$ 

One of the most useful form for displaying the Fourier Series of a signal is by means of a graphical plot showing the relative strengths of the components as a function of freqs. Such a plot is called "the freq spectrum" of a given signal.

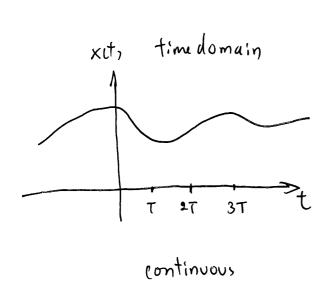
There are two different kinds of spectrum pluts: one-sided and two-sided

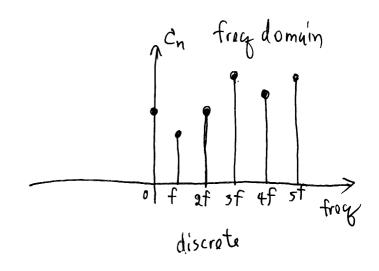


**FIGURE 9–4** Typical one-sided amplitude-frequency spectrum.



**FIGURE 9–5**Two-sided amplitude spectrum corresponding to Figure 9–4.



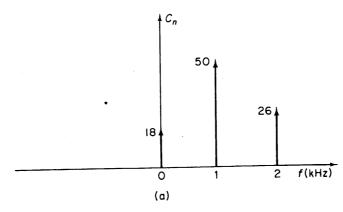


Example From tu previous example, plot tu spectrum plots
a) single-sided plot:

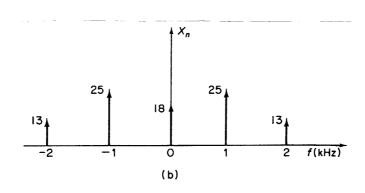
xit)= 16+50(05(2000 Tt-36.87)+26(05(4000 Tt+157.38°)

Frequency (Hz)	0	1000	2000
Amplitude	18	50	26

The one-sided plot is shown in Figure 9-6(a).



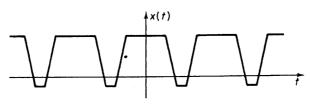
Frequency (Hz)	0	±1000	±2000
Amplitude	18	25	13



### Fourier Series Symmetry Conditions

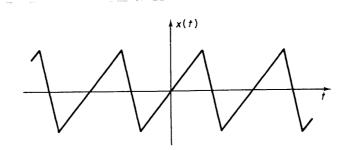
In many cases, there are certain properties that may be used to simplify the computation of spectrum. The types of criteria to be studied in this section are the "symmetry" conditions.

$$x(-t) = x(t)$$



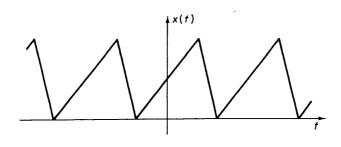
all bn = 0 (sine functions disappear!)

### @ odd function:



all  $a_n = 0$ 

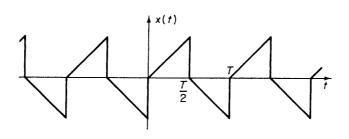
or it can be disgussed by the presence of a dc component.



ao = 0

 $a_n = 0$ 

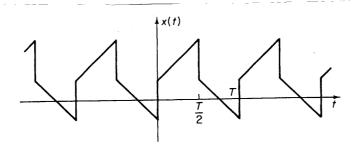
 $b_n \neq 0$ 



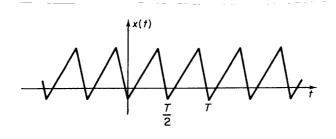
mure will only be odd-numbered harmonics present in the series.

(n=1,3,5,7,...)

or with a presence of a dc component:



(Full wave symmetry: xct + I) = xct)



Thur will only be even-numbered harmonics present in the series (n=0,2,4,6,---)

# Table of summary on the symmetry

**TABLE 9–1**Fourier series symmetry conditions

Sine—cosine form:  $x(t) = A_0 + \sum_{n=1}^{\infty} (A_n \cos n\omega_1 t + B_n \sin n\omega_1 t), \qquad \omega_1 = 2\pi f_1 = \frac{2\pi}{T}$ 

Amplitude-phase form:  $x(t) = C_0 + \sum_{n=1}^{\infty} C_n \cos(n\omega_1 t + \phi_n) = C_0 + \sum_{n=1}^{\infty} C_n \sin(n\omega_1 t + \theta_n), \qquad C_n = \sqrt{A_n^2 + B_n^2}$ 

Complex exponential form:  $x(t) = \sum_{n=-\infty}^{\infty} \overline{X}_n e^{jn\omega_1 t}$ ,  $\overline{X}_n = \frac{A_n - jB_n}{2}$ , for  $n \neq 0$   $\overline{X}_0 = A_0$ 

Condition	$A_n(\text{except } n = 0)$	$B_n$	$\overline{X}_n$	Comments
General	$\frac{2}{T} \int_{0}^{T} x(t) \cos n\omega_{1} t dt$	$\frac{2}{T} \int_0^T x(t) \sin n\omega_1 t  dt$	$\frac{1}{T}\int_0^T x(t) e^{-jn\omega_1 t} dt$	
Even function $x(-t) = x(t)$	$\frac{4}{T} \int_0^{T/2} x(t) \cos n\omega_1 t  dt$	0	$\frac{2}{T}\int_0^{T/2} x(t) \cos n\omega_1 t  dt$	One-sided forms have only cosine terms $\overline{X}_n$ terms are real
Odd function $x(-t) = -x(t)$	0	$\frac{4}{T}\int_0^{T/2} x(t) \sin n\omega_1 t  dt$	$\frac{-2j}{T}\int_0^{T/2}x(t)\sin n\omega_1tdt$	One-sided forms have only sine terms $\overline{X}_n$ terms are imaginary
Half-wave symmetry	$\frac{4}{T}\int_0^{T/2} x(t) \cos n\omega_1 t  dt$	$\frac{4}{T}\int_0^{T/2} x(t) \sin n\omega_1 t  dt$	$\frac{2}{T}\int_0^{T/2}x(t)\varepsilon^{-jn\omega_1t}dt$	Odd-numbered harmonics only
$x\left(t + \frac{T}{2}\right) = -x(t)$ Full-wave symmetry $x\left(t + \frac{T}{2}\right) = x(t)$	$\frac{4}{T}\int_0^{T/2}x(t)\cos n\omega_1tdt$	$\frac{4}{T}\int_0^{T/2} x(t) \sin n\omega_1 t  dt$	$\frac{2}{T}\int_0^{T/2} x(t) e^{-jn\omega_1 t} dt$	Even-numbered harmonics only

# some common periodic want forms and thun Fourier Series

**TABLE 9–2**Some common periodic signals and their Fourier series

Signal x(t)	Fourier series
Square wave	$\frac{4A}{\pi} \left(\cos \omega_1 t - \frac{1}{3} \cos 3\omega_1 t + \frac{1}{5} \cos 5\omega_1 t - \frac{1}{7} \cos 7\omega_1 t + \cdots\right)$
Triangular wave	$\frac{8A}{\pi^2} \left(\cos \omega_1 t + \frac{1}{9} \cos 3\omega_1 t + \frac{1}{25} \cos 5\omega_1 t + \cdots\right)$
Sawtooth wave	$\frac{2A}{\pi} \left( \sin \omega_1 t - \frac{1}{2} \sin 2\omega_1 t + \frac{1}{3} \sin 3\omega_1 t - \frac{1}{4} \sin 4\omega_1 t + \cdots \right)$
Half-wave rectified cosine	$\frac{A}{\pi} \left( 1 + \frac{\pi}{2} \cos \omega_1 t + \frac{2}{3} \cos 2\omega_1 t - \frac{2}{15} \cos 4\omega_1 t + \frac{2}{35} \cos 6\omega_1 t - \cdots (-1)^{n/2+1} \frac{2}{n^2 - 1} \cos n\omega t + \cdots \right)$ n even
Full-wave rectified cosine	$\frac{2A}{\pi} (1 + \frac{2}{3}\cos 2\omega_1 t - \frac{2}{15}\cos 4\omega_1 t + \frac{2}{35}\cos 6\omega_1 t - \cdots (-1)^{n/2+1} \frac{2}{n^2-1}\cos n\omega_1 t + \cdots) $ n even
	$Ad\left[1 + 2\left(\frac{\sin \pi d}{\pi d} \cos \omega_1 t + \frac{\sin 2\pi d}{2\pi d} \cos 2\omega_1 t + \frac{\sin 3\pi d}{3\pi d} \cos 3\omega_1 t + \cdots\right)\right]$ $d = T/T$