1

Digital Signal Processing EE3900

M Avinash AI21BTECH11018

November 2, 2022

CONTENTS

1. Definitions

1. The unit step function is

$$u(t) = \begin{cases} 1 & t > 0 \\ \frac{1}{2} & t = 0 \\ 0 & t < 0 \end{cases}$$
 (1.1)

2. The Laplace transform of g(t) is defined as

$$G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt$$
 (1.2)

2. Laplace Transform

1. In the circuit, the switch S is connected to position P for a long time so that the charge on the capacitor becomes $q_1 \mu C$. Then S is switched to position Q. After a long time, the charge on the capacitor is $q_2 \mu C$.

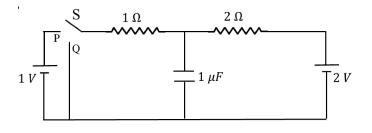


Fig. 2.1.

2. Draw the circuit using latex-tikz. **Solution:** The following code yields Fig.??

wget https://github.com/AvinashNayak27/ Spice/blob/master/Tikz%20Circuits/2.2.tex

3. Find q_1 .

Solution: Before switching S to Q: At steady state, which achieved when switch S is at P

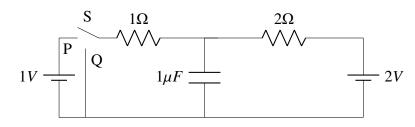


Fig. 2.2. Given Circuit

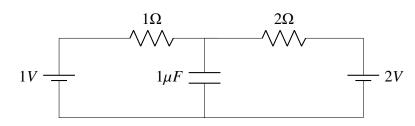


Fig. 2.3. Before switching S to Q

for long time capacoitor behaves as an open switch, hence current through capacitor is 0, Let i be the current flowing in the circuit at steady state. Applying KVL,

$$1 - i - 2i - 2 = 0 \tag{2.1}$$

$$3i = -1 \Rightarrow i = \frac{-1}{3}A\tag{2.2}$$

Potential Difference across the capacitor at steady state is

$$1 - \left(\frac{-1}{3}\right) = \frac{4}{3}V\tag{2.3}$$

$$q_1 = \frac{4}{3} \cdot 1$$
 (2.4)
= $\frac{4}{3}\mu C$ (2.5)

$$=\frac{4}{3}\mu C\tag{2.5}$$

4. Show that the Laplace transform of u(t) is $\frac{1}{s}$ and find the ROC.

Solution: We know that Laplace Transform fo function f(t) is given as F(s),

$$F(s) = \int_0^\infty f(t)e^{-st} dt \qquad (2.6)$$

(2.7)

For u(t), we have,

$$F(s) = \int_0^\infty u(t)e^{-st} dt \qquad (2.8)$$

Using (??),

$$F(s) = \int_0^\infty u(t)e^{-st} dt$$
 (2.9)

$$= \int_0^\infty e^{-st} dt \tag{2.10}$$

$$= -\left(0 - \frac{1}{s}\right) \tag{2.11}$$

$$=\frac{1}{s} \tag{2.12}$$

ROC is Re(s) > 0 since for s > 0, $e^{-st} < \infty$ for $t \to \infty$

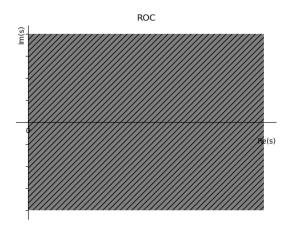


Fig. 2.4.

5. Show that

$$e^{-at}u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} \frac{1}{s+a}, \quad a > 0$$
 (2.13) Fig. 2.6.

and find the ROC.

Solution: From ??,

$$F(s) = \int_0^\infty u(t)e^{-at}e^{-st} dt$$
 (2.14)

$$= \int_0^\infty u(t)e^{-(s+a)t} dt$$
 (2.15)

$$= \int_{0}^{\infty} e^{-(s+a)t} dt$$
 (2.16)

$$= -\left(0 - \frac{1}{s+a}\right) \tag{2.17}$$

$$=\frac{1}{s+a}\tag{2.18}$$

ROC is

$$Re(s) + a > 0 \Rightarrow Re(s) > -a$$
 (2.19)

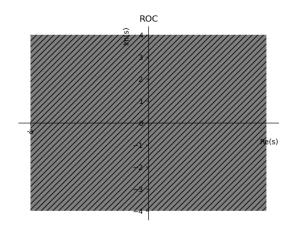
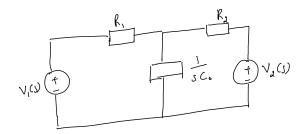


Fig. 2.5.

6. Now consider the following resistive circuit transformed from Fig. ?? where



$$u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} V_1(s)$$
 (2.20)

$$2u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} V_2(s)$$
 (2.21)

Find the voltage across the capacitor $V_{C_0}(s)$. **Solution:**

$$R_{eff} = \frac{1}{1 + \frac{1}{2}} = \frac{2}{3}\Omega \tag{2.22}$$

$$V_{eff} = \frac{1}{1 + \frac{1}{2}} = \frac{2}{3}V \tag{2.23}$$

$$V_{C_0}(s) = V_S(s) \frac{C_0}{C_0 + R_{eff}}$$
 (2.24)

$$= \left(\frac{4}{3s}\right) \left(\frac{\frac{1}{s}}{\frac{1}{s} + \frac{2}{3}}\right) \tag{2.25}$$

$$= \frac{3+4s}{3s\left(s+\frac{3}{2}\right)} \tag{2.26}$$



Solution: Running the following code gives the plot.

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/2.7.py

Using (??),

$$\frac{3+4s}{3s\left(s+\frac{3}{2}\right)} = \frac{2}{3s} + \frac{2}{3(\frac{3}{2}+s)}$$
 (2.27)

Apply inverse Laplacian Transform,

$$V_{C_0}(s) \stackrel{\mathcal{L}^{-\infty}}{\longleftrightarrow} V_{C_0}(t)$$
 (2.28)

$$\mathcal{L}^{-1} \left[V_{C_0}(s) \right] = \mathcal{L}^{-1} \left[\frac{2}{3s} + \frac{2}{3(\frac{3}{2} + s)} \right]$$
 (2.29)

$$= \mathcal{L}^{-1} \left[\frac{2}{3s} \right] - \frac{2}{3} \mathcal{L}^{-1} \left[\frac{1}{\frac{3}{2} + s} \right]$$

Since,

$$\mathcal{L}^{-1} \left[\frac{1}{s} \right] = u(t) \tag{2.31}$$

$$\mathcal{L}^{-1}\left[\frac{1}{s-a}\right] = e^{at}u(t) \tag{2.32}$$

Using the above equations,

$$V_{C_0}(t) = \frac{2}{3} \left(1 + e^{\frac{-3}{2}t} \right) u(t)$$
 (2.33)

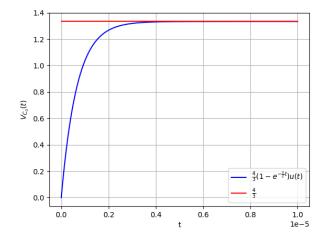


Fig. 2.7. Plot of $V_{C_0}(t)$

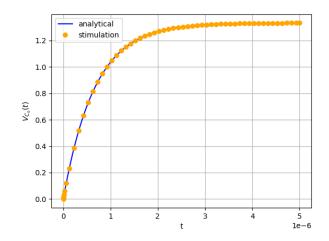


Fig. 2.8.

8. Verify your result using ngspice.

Solution:

9. Obtain Fig. ?? using the equivalent differential equation

Solution: Results obtained can be verified by running the following code.

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/2.8.cir

And is plotted using the below code.

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/2.8.py

Using Kirchoff's junction law

$$\frac{v_c(t) - v_1(t)}{R_1} + \frac{v_c(t) - v_2(t)}{R_2} + \frac{dq}{dt} = 0 \quad (2.34)$$

where q(t) is the charge on the capacitor On taking the Laplace transform on both sides of this equation

$$\frac{V_c(s) - V_1(s)}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + \left(sQ(s) - q(0^-)\right) = 0 \tag{2.35}$$

But $q(0^{-}) = 0$ and

$$q(t) = C_0 v_c(t) \tag{2.36}$$

$$\implies Q(s) = C_0 V_c(s)$$
 (2.37)

Thus

$$\frac{V_c(s) - V_1(s)}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + sC_0V_c(s) = 0$$

(2.38)

$$\Longrightarrow \frac{V_c(s) - V_1(s)}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + \frac{V_c(s) - 0}{\frac{1}{sC_0}} = 0$$
(2.39)

which is the same equation as the one we obtained from Fig. ??

3. Initial Conditions

1. Find q_2 in Fig. ??.

Solution: At steady state capacitor behaves as an open switch. Hence $V_{C_0} = V_{1\Omega}$.

Let i be the current in the circuit. Using KVL,

$$2 - 2i - i = 0 \implies i = \frac{2}{3}$$
 (3.1)

$$V_{1\Omega} = i \times 1 = \frac{2}{3}V \tag{3.2}$$

$$V_{C_0} = \frac{q_2}{C_0} = V_{1\Omega} = \frac{2}{3}$$
 (3.3)

$$\implies q_2 = \frac{2}{3}\mu C \tag{3.4}$$

2. Draw the equivalent s-domain resistive circuit when S is switched to position Q. Use variables R_1, R_2, C_0 for the passive elements. Use latextikz.

Solution:

3. $V_{C_0}(s) = ?$

Solution: Let voltage across capacitor be V. Using KCL at node in Fig. $\ref{eq:condition}$?

$$\frac{V-0}{R_1} + \frac{V-\frac{2}{s}}{R_2} + sC_0\left(V - \frac{4}{3s}\right) = 0$$
 (3.5)

$$\implies V_{C_0}(s) = \frac{\frac{2}{sR_2} + \frac{4C_0}{3}}{\frac{1}{R_1} + \frac{2}{R_2} + sC_0}$$
 (3.6)

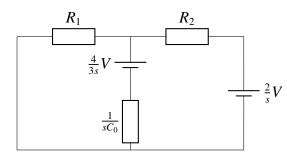


Fig. 3.1. After switching S to Q

4. $v_{C_0}(t) = ?$ Plot using python.

Solution: Running the following code gives the plot.

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/3.4.py

From (??),

$$V_{C_0}(s) = \frac{4}{3} \left(\frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right) + \frac{2}{R_2 \left(\frac{1}{R_1} + \frac{1}{R_2} \right)} \left(\frac{1}{s} - \frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right)$$
(3.7)

Taking an inverse Laplace Transform,

$$v_{C_0}(t) = \frac{4}{3}e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}u(t) + \frac{2}{R_2\left(\frac{1}{R_1} + \frac{1}{R_2}\right)}\left(1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}\right)u(t)$$
(3.8)

Substituting values gives

$$v_{C_0}(t) = \frac{2}{3} \left(1 + e^{-(1.5 \times 10^6)t} \right) u(t)$$
 (3.9)

5. Verify your result using ngspice.

Solution: Results obtained can be verified by running the following code.

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/3.5.cir

Runningn the below code plots the figure ??, and verifies our result.

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/3.5.py

6. Find $v_{C_0}(0-)$, $v_{C_0}(0+)$ and $v_{C_0}(\infty)$.

Solution: From the initial conditions,

$$v_{C_0}(0-) = \frac{q_1}{C} = \frac{4}{3}V \tag{3.10}$$

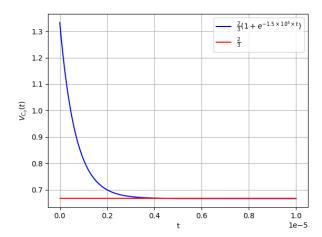


Fig. 3.2. Plot of $V_{C_0}(t)$

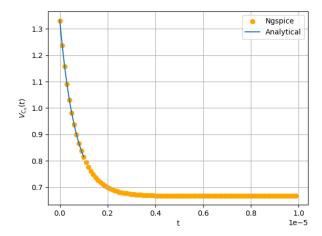


Fig. 3.3. ngspice plot of $V_{C_0}(t)$

Using (??),

$$v_{C_0}(0+) = \lim_{t \to 0+} v_{C_0}(t) = \frac{4}{3}V \tag{3.11}$$

$$v_{C_0}(\infty) = \lim_{t \to \infty} v_{C_0}(t) = \frac{2}{3}V$$
 (3.12)

7. Obtain Fig. ?? using the equivalent differential equation

Solution: Using Kirchoff's junction law

$$\frac{v_c(t) - 0}{R_1} + \frac{v_c(t) - v_2(t)}{R_2} + \frac{\mathrm{d}q}{\mathrm{d}t} = 0$$
 (3.13)

where q(t) is the charge on the capacitor. On taking the Laplace transform on both sides of

the equation (??), we get,

$$\frac{V_c(s) - 0}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + sQ(s) - q(0^-) = 0$$
(3.14)

But $q(0^-) = \frac{4}{3}C_0$ and

$$q(t) = C_0 v_c(t)$$
 (3.15)

$$\implies Q(s) = C_0 V_c(s)$$
 (3.16)

Thus

$$\frac{V_c(s) - 0}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + \left(sC_0V_c(s) - \frac{4}{3}C_0\right) = 0$$
(3.17)

$$\implies \frac{V_c(s) - 0}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + \frac{V_c(s) - \frac{4}{3s}}{\frac{1}{sC_0}} = 0$$
(3.18)

which is the same equation as the one we obtained from Fig. ??

4. BILINEAR TRANSFORM

4.1. In Fig. ??, consider the case when *S* is switched to *Q* right in the beginning. Formulate the differential equation

Solution: The differential equation is the same as before

$$\frac{v_c(t) - 0}{R_1} + \frac{v_c(t) - v_2(t)}{R_2} + \frac{dq}{dt} = 0 \quad (4.1)$$

i.e.,
$$\frac{v_c(t)}{R_1} + \frac{v_c(t) - v_2(t)}{R_2} + C_0 \frac{dv_c}{dt} = 0$$
 (4.2)

but with a different initial condition

$$q(0^{-}) = q(0) = 0$$
 (4.3)

4.2. Find H(s) considering the outure voltage at the capacitor

Solution: On taking the Laplace transform on both sides of this equation

$$\frac{V_c(s)}{R_1} + \frac{V_c(s) - V_2(s)}{R_2} + sQ(s) - 0 = 0$$
(4.4)

 $\implies V_c(s) \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + sC_0 V_c(s) = \frac{V_2(s)}{R_2}$ (4.5)

$$\Longrightarrow \frac{V_c(s)}{V_2(s)} = \frac{\frac{1}{R_2}}{\frac{1}{R_1} + \frac{1}{R_2} + sC_0}$$
 (4.6)

The transfer function is thus

$$H(s) = \frac{\frac{1}{R_2 C_0}}{s + \frac{1}{R_1 C_0} + \frac{1}{R_2 C_0}}$$
(4.7)

On substituting the values, we get

$$H(s) = \frac{5 \times 10^5}{s + 1.5 \times 10^6} \tag{4.8}$$

4.3. Plot H(s). What kind of filter is it?

Solution: Download the following Python code that plots Fig. ??

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/4.3.py

Run the codes by executing

python 4.3.py

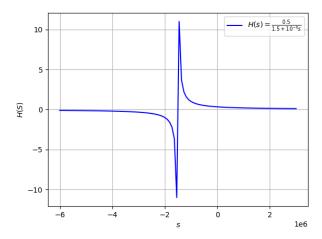


Fig. 4.1. Plot of H(s)

Consider the frequency-domain transfer function by putting $s = \omega$

$$H(j\omega) = \frac{5 \times 10^5}{j\omega + 1.5 \times 10^6}$$
 (4.9)

$$\implies |H(j\omega)| = \frac{5 \times 10^5}{\sqrt{\omega^2 + 2.25 \times 10^{12}}} \quad (4.10)$$

As ω increases, $|H(j\omega)|$ decreases

In other words, the amplitude of high-frequency signals gets diminished and they get filtered out

Therefore, this is a low-pass filter

4.4. Using trapezoidal rule for integration, formulate the difference equation by considering

$$y(n) = y(t)|_{t=n}$$
 (4.11)

Solution:

$$\frac{v_c(t)}{R_1} + \frac{v_c(t) - v_2(t)}{R_2} + C_0 \frac{dv_c}{dt} = 0 \quad (4.12)$$

$$\implies C_0 \frac{dv_c}{dt} = \frac{2u(t) - v_c(t)}{R_2} - \frac{v_c(t)}{R_1} \qquad (4.13)$$

$$\implies v_c(t)|_{t=n}^{n+1} = \int_n^{n+1} \left(\frac{2u(t) - v_c(t)}{R_2 C_0} - \frac{v_c(t)}{R_1 C_0}\right) dt$$

$$(4.14)$$

By the trapezoidal rule of integration

$$\int_{a}^{b} f(t)dt \approx \frac{b-a}{2}(f(a) + f(b))$$
 (4.15)

Consider $y(t) = v_c(t)$

$$y(n+1) - y(n) = \frac{1}{R_2 C_0} (u(n) + u(n+1))$$
$$-\frac{1}{2} (y(n+1) + y(n)) \left(\frac{1}{R_1 C_0} + \frac{1}{R_2 C_0} \right)$$
(4.16)

Thus, the difference equation is

$$y(n+1)\left(1 + \frac{1}{2R_1C_0} + \frac{1}{2R_2C_0}\right)$$

$$= y(n)\left(1 - \frac{1}{2R_1C_0} - \frac{1}{2R_2C_0}\right)$$

$$+ \frac{1}{R_2C_0}\left(u(n) + u(n+1)\right) \quad (4.17)$$

4.5. Find H(z)

Solution: Let $\mathcal{Z}{y(n)} = Y(z)$

On taking the Z-transform on both sides of the difference equation

$$zY(z)\left(1 + \frac{1}{2R_1C_0} + \frac{1}{2R_2C_0}\right)$$

$$= Y(z)\left(1 - \frac{1}{2R_1C_0} - \frac{1}{2R_2C_0}\right)$$

$$+ \frac{1}{R_2C_0}\left(\frac{1}{1 - z^{-1}} + \frac{z}{1 - z^{-1}}\right) \quad (4.18)$$

$$Y(z)\left(z + \frac{z}{2R_1C_0} + \frac{z}{2R_2C_0} - 1 + \frac{1}{2R_1C_0} + \frac{1}{2R_2C_0}\right)$$

$$= \frac{1}{R_2C_0} \frac{1+z}{1-z^{-1}} \quad (4.19)$$

Also

$$v_2(t) = 2 \qquad \forall t \ge 0 \qquad (4.20)$$

$$\implies x(n) = 2u(n) \tag{4.21}$$

$$\implies X(z) = \frac{2}{1 - z^{-1}} \qquad |z| > 1 \qquad (4.22)$$

Thus, the transfer function in z-domain is

$$H(z) = \frac{Y(z)}{X(z)}$$

$$= \frac{\frac{1+z}{2R_2C_0}}{z + \frac{z}{2R_1C_0} + \frac{z}{2R_2C_0} - 1 + \frac{1}{2R_1C_0} + \frac{1}{2R_2C_0}}$$

$$= \frac{\frac{1+z^{-1}}{2R_2C_0}}{1 + \frac{1}{2R_1C_0} + \frac{1}{2R_2C_0} - z^{-1} + \frac{z^{-1}}{2R_1C_0} + \frac{z^{-1}}{2R_2C_0}}$$

$$(4.24)$$

On substituting the values

$$H(z) = \frac{2.5 \times 10^5 (1 + z^{-1})}{7.5 \times 10^5 + 1 + (7.5 \times 10^5 - 1)z^{-1}}$$
(4.26)

with the ROC being

$$|z| > \max\left(1, \left| \frac{7.5 \times 10^5 - 1}{7.5 \times 10^5 + 1} \right| \right)$$
 (4.27)

$$\implies |z| > 1 \tag{4.28}$$

4.6. How can you obtain H(z) from H(s)?

Solution: The Z-transform can be obtained from the Laplace transform by the substitution

$$s = \frac{2}{T} \frac{1 - z^{-1}}{1 + z^{-1}} \tag{4.29}$$

where T is the step size of the trapezoidal rule (1 in our case)

This is known as the bilinear transform Thus

$$H(z) = \frac{\frac{1}{R_2C_0}}{2\frac{1-z^{-1}}{1+z^{-1}} + \frac{1}{R_1C_0} + \frac{1}{R_2C_0}}$$

$$= \frac{\frac{\frac{1+z^{-1}}{2R_2C_0}}{1-z^{-1} + \left(\frac{1}{2R_1C_0} + \frac{1}{2R_2C_0}\right)(1+z^{-1})}$$

$$= \frac{\frac{\frac{1+z^{-1}}{2R_2C_0}}{1+\frac{1}{2R_1C_0} + \frac{1}{2R_2C_0} - z^{-1} + \frac{z^{-1}}{2R_1C_0} + \frac{z^{-1}}{2R_2C_0}}$$

$$= \frac{2.5 \times 10^5(1+z^{-1})}{7.5 \times 10^5 + 1 + (7.5 \times 10^5 - 1)z^{-1}}$$

$$(4.30)$$

which is the same as what we obtained earlier 4.7. Find y(n) from H(z) and verify whether $y(n) = y(t)|_{t=n}$

Solution:

$$Y(z) = H(z)X(z)$$

$$= \frac{2.5 \times 10^{5}(1+z^{-1})}{7.5 \times 10^{5} + 1 + (7.5 \times 10^{5} - 1)z^{-1}} \frac{2}{1-z^{-1}}$$

$$= \frac{\frac{2}{3}}{1-z^{-1}} - \frac{\frac{2}{3}}{7.5 \times 10^{5} + 1 + (7.5 \times 10^{5} - 1)z^{-1}}$$
(4.34)

On taking the inverse Z-transform by considering the ROC to be |z| > 1, we get

$$y(n) = \frac{2}{3}u(n) - \frac{2}{3}\frac{1}{7.5 \times 10^5 + 1} \left(-\frac{7.5 \times 10^5 - 1}{7.5 \times 10^5 + 1} \right)^n u(n)$$

$$= \frac{2}{3} \left(1 - \frac{(1 - 7.5 \times 10^5)^n}{(1 + 7.5 \times 10^5)^{n+1}} \right) u(n) \quad (4.38)$$

If we are sampling the signal at intervals of $T \ll 10^{-5}$, say [parse-numbers = false] 10^{-7} , i.e., $n = 10^{-7}$, 2×10^{-7} , ...

$$y(n) \approx \frac{2}{3} \left(1 - \frac{1 - 7.5 \times 10^5 n}{1 + 7.5 \times 10^5 n} \right) u(n)$$
 (4.39)

Now

$$Y(s) = H(s)X(s) (4.40)$$

$$=\frac{5\times10^5}{s+1.5\times10^6}\frac{2}{s}\tag{4.41}$$

$$= \frac{10^6}{1.5 \times 10^6} \left(\frac{1}{s} - \frac{1}{s + 1.5 \times 10^6} \right) \quad (4.42)$$

On taking the inverse Laplace transform by considering the ROC to be $\Re(s) > 0$, we get

$$y(t) = \frac{2}{3} \left(1 - e^{-1.5 \times 10^6 t} \right) u(t)$$
 (4.43)

But

$$e^{-1.5 \times 10^{6} t} = \frac{e^{-0.75 \times 10^{6} t}}{e^{0.75 \times 10^{6} t}}$$

$$\approx \frac{1 - 7.5 \times 10^{5} t}{1 + 7.5 \times 10^{5} t}$$
 when $t \ll 10^{-6}$

$$(4.45)$$

Therefore

$$y(t) \approx \frac{2}{3} \left(1 - \frac{1 - 7.5 \times 10^5 t}{1 + 7.5 \times 10^5 t} \right) u(t)$$
 (4.46)

$$\therefore y(n) = y(t)|_{t=n} \tag{4.47}$$

Download the following codes for simulation and plotting Fig. ?? respectively

wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/4.7.cir wget https://github.com/AvinashNayak27/ Spice/blob/master/codes/4.7.py

Run the codes by executing

```
ngspice 4.7.cir
python 4.7.py
```

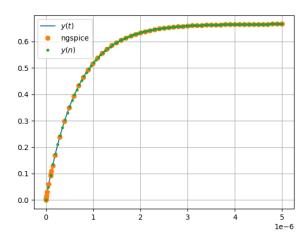


Fig. 4.2. Plots of y(t) and y(n)