# FIR Filter Design for BandPass Filter

Sandeepa H.K.C.A. dept. of Electronic and Telecommunication Engineering, University of Moratuwa.

Abstract—In this document we try to implement FIR filter design using windowing method in conjunction with the Kaiser window. And also, there are several specifications to show. In addition to that accuracy of this design will be checked using given function and resultant plots in both time and frequency domain. The software is used for project is MATLABR2018a.

#### I. INTRODUCTION

This project is used to implement FIR (Finite Duration Impulse Response) filter design using given specification in the project description. When we try to implement filter design using windowing method this is also call as Fourier series method, there are more techniques but in here we use most famous windowing method call as Kaiser windowing method. Using MATLABR2018a software and building functions these calculations for designing were implemented.

#### II. BASIC THEORY

When we design the filters there are two main techniques. Closed form design technique and Optimized design technique. In here we use closed form technique.

Since the frequency response is periodic, we can represent it as a Fourier Series as below.

$$H(e^{j\omega T}) = \sum_{n=-\infty}^{n=\infty} h(nT)e^{-j\omega nT}$$

where,

$$h(nT) = \frac{1}{\omega_s} \int_{-\omega_s}^{\omega_s} H(e^{j\omega T}) e^{j\omega nT} d\omega$$

Define the impulse response of the Filter as h(nT) and then by substituting  $z=e^{j\omega T}$  we obtain the Transfer function of the filter as follows.

$$H(z) = \sum_{n = -\infty}^{n = \infty} h(nT)z^{-n}$$

In the windowing method we truncate the impulse response using multiplication with window function in time domain as follows,

$$h_w(nT) = h(nT)w(nT)$$

Hence, the frequency response of the modified filter is given by the convolution integral.

$$H_{w}(e^{j\omega T}) = \frac{T}{2\pi} \int_{0}^{\frac{2\pi}{T}} H(e^{j\omega'T}) W(e^{j(\omega-\omega')T}) d\omega'$$

With these theory aspects filtering can be done by using several windows such as Rectangular, Hann, Hamming, Blackman, Dolph-Chebyshev and Kaiser. If Rectangular window is easy it creates more ripple ratio. In order to reduce that effect and good approximation Kaiser window is selected Kaiser window function is,

$$w_{K}(nT) = \begin{cases} \frac{I_{0}(\beta)}{I_{0}(\beta)}, & for |n| \leq (N-1)/2\\ 0, & otherwise \end{cases}$$

where,

$$\beta = \alpha \sqrt{1 - \left(\frac{2n}{N-1}\right)^2}$$

and

$$I_0(x) = 1 + \sum_{k=1}^{\infty} \left[ \frac{1}{k!} \left( \frac{x}{2} \right)^k \right]^2$$

Using that knowledge try to implement Band Pass filter using Kaiser Window.

#### III. IMPLEMENTATION

# A. Finding Parameters for given Index(Your Index)

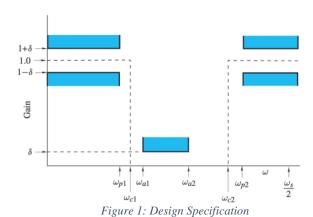
In the project description it mentions how can we select the basic parameters according to your index.

For given index number 180564F, the parameters are,

| Maximum passband ripple, Ãp      | $0.08~\mathrm{dB}$ |
|----------------------------------|--------------------|
| Minimum stopband attenuation, Ãa | 51dB               |
| Lower passband edge, $\Omega p1$ | 700 rad/s          |
| Upper passband edge, $\Omega$ p2 | $1100 \; rad/s$    |
| Lower stopband edge, $\Omega$ a1 | 550 rad/s          |
| Upper stopband edge, $\Omega$ a2 | 1200 rad/s         |
| Sampling frequency, $\Omega$ s   | 3200 rad/s         |

# B. Calculations

Parameters are defined as follows,



Critical transition width can be calculated as follows,

$$B_t = \min \left[ \left( \omega_{p1} - \omega_{a1} \right), \left( \omega_{a2} - \omega_{p2} \right) \right]$$

Idealized frequency response for a BandPass filter is given as.

$$H(e^{j\omega t}) = \begin{cases} 1 & for - \omega_{c2} \le \omega \le -\omega_{c1} \\ 1 & for & \omega_{c2} \le \omega \le \omega_{c1} \\ 0 & Otherwise \end{cases}$$

where cutoff frequencies,

$$\omega_{c1}=\omega_{p1}-\left(\frac{B_t}{2}\right)$$
 ,  $\omega_{c2}=\omega_{p2}+\left(\frac{B_t}{2}\right)$ 

By using inverse Fourier Transform impulse response can be obtained,

$$h[nT] = \begin{cases} \frac{2}{\omega_s} (\omega_{c1} - \omega_{c2}), & for \ n = 0\\ \frac{1}{n\pi} (\sin(\omega_{c2}nT) - \sin(\omega_{c1}nT)), & Otherwise \end{cases}$$

#### C. Determine $\delta$ value

The parameter  $\delta$  is defined in a way that, actual passband ripple  $A_p \leq \tilde{A}_p$ 

The actual minimum stopband attenuation,  $Aa \ge \tilde{A}a$ 

Let 
$$\delta = \min(\delta_p, \delta_a)$$

where 
$$\delta_p = \frac{10^{0.05 \tilde{A}_p} - 1}{10^{0.05 \tilde{A}_p} + 1}$$
 and  $\delta_a = 10^{-0.05 \tilde{A}a}$ 

# D. Actual Stopband Attenueation

After finding correct value of delta then move to find out Actual Attenuation at the stop band as follows,

$$A_a = -20\log(\delta)$$

# E. Detaremine parameter D and α

Using A<sub>a</sub> which is determined earlier.

$$D = \begin{cases} 0.9222 & for A_a \le 21\\ \frac{A_a - 7.95}{14.36} & for A_a > 21 \end{cases}$$

and

$$\alpha = \begin{cases} 0 & for 21 \le A_a \\ 0.5842(A_a - 21)^{0.4} + 0.07886(A_a - 21) & for 21 < A_a \le 50 \\ 0.07886(A_a - 21) & for A_a > 50 \end{cases}$$

The smallest, odd value which satisfies the following condition is chosen as N.

$$N \ge \frac{\omega_s D}{B_t} + 1$$

## IV. RESULTS

By going through the above defined equations, the parameters can be calculated as follows,

| Parameter | Value  |
|-----------|--------|
| delta p   | 0.0046 |
| delta a   | 0.0028 |
| $A_a$     | 51.00  |
| D         | 2.9979 |
| α         | 4.6614 |
| N         | 97     |

By doing calculation using MATLAB the results can be generated as figures. It helps to figure out our implementation.

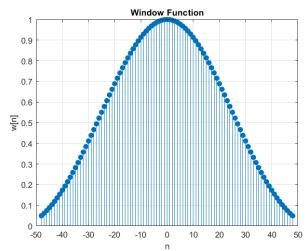


Figure 2: Impulse Response of Kaiser Window

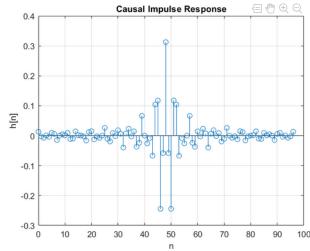


Figure 3: Causal Impulse Response

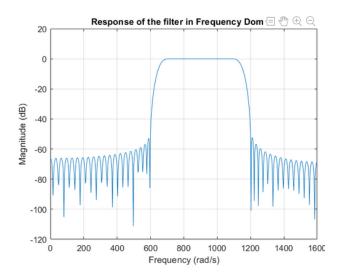


Figure 4: Magnitude Response of the Filter

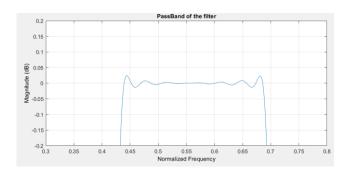


Figure 5: Magnitude Response of the passband

# V. VALIDATION OF THE DESIGNED FILTER

For validations following function is used and the omega values are mentioned below,

$$x(nT) = \sum_{i=1}^{3} \sin\left(\Omega_{i} nT\right)$$

where,

| $\omega_1$ | 275        |
|------------|------------|
| $\omega_2$ | 900 rad/s  |
| <b>ω</b> 3 | 1400 rad/s |

for the validations.

when we give the above input to our designed filter it should filter out the  $\omega_2$  because it defined using passband frequency.

The observations are in both time domain and frequency domain and also have the expectation vs ideal output results. Using these results, we can clearly say that this bandpass filter works properly.

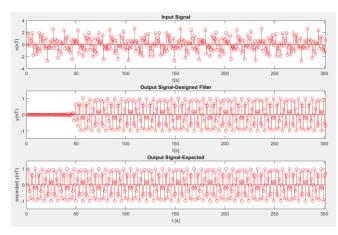


Figure 6: Time Domain Representation for input signal, designed filter's output and expected output

Expected output and filter output are almost same.

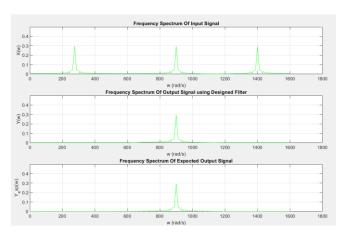


Figure 7: Frequency Domain Representation for input signal, designed filter's output and expected output

# VI. CONCLUSION

The implementation of FIR Filter design is efficient way because it gives the almost same output as the ideal filter. Its ripple ratio is very small compared to the other windows. And also, the methods that we used to implement is very helpful for students and it can be easily implemented.

## ACKNOWLEDGMENT

I'd like to express my heartfelt gratitude to Dr. Chamira Edussooriya for her invaluable assistance in this project.

# REFERENCES

A. Antoniou, "Digital Signal Processing," 2005. [Online]. Available: www.ece.uvic.ca/~dsp. [Accessed 29 12 2016]

En.wikipedia.org. 2021. *Kaiser window*. [online] Available at: <a href="https://en.wikipedia.org/wiki/Kaiser\_window">https://en.wikipedia.org/wiki/Kaiser\_window</a>> [Accessed 5 March 2021].

# **Appendix**

```
1 clc;
2 clear all;
3 output = calcMyFilterSpecs(5, 6, 4); % call the function to get values
  for the given index
4 Ws = output(7); % set sample frequency value
5 T=2*pi/Ws; % set the sampling period
6 delta_p = (10^{(0.05*output(1))} - 1)/(10^{(0.05*output(1))} + 1); %
  determine delta p value using Ap
7 delta_a = (10^{(-0.05)} * output(2))); % determine delta a value using Aa
8 delta = min(delta a, delta p); % find delta value
9 Actual Aa = (-20)*log10(delta); % finding actual stopband attenuation
10 Bt = min((output(3) - output(5)), (output(6) - output(4))); % create
  transition width
11 Wc1 = output(3) - Bt/2; % define the cutoff frequencies
12 Wc2 = output(4) + Bt/2;
13 valAlfa = getAlfa(Actual Aa); % call the function to get value alpha
14 valD = getD(Actual Aa); % call the function to get value D
15 calcN = ceil((output(7)*valD)/Bt + 1); % calculate current N value and
  round it to the nearest integer
16 finalN = calcN + 1 - rem(calcN, 2); % get the correct N value
17 M = finalN - 1;
18 tau = M/2;
19 range1_n = -tau: 1: tau; % define range of n value
20 %----- Create Kaiser Window ------
21 valBeta = valAlfa*(1 - ((2*range1 n)/M).^2).^0.5; % Determine beta value
22 iO alpha = bessel(valAlfa); % call the bessel function to fing window
23 i0 beta = bessel(valBeta);
24 win = i0 beta/i0 alpha; % find the window function for specified n values
25 stem(range1 n, win, "fill")
26 %ylim([0 1])
27 grid on;
28 xlabel("n");
29 ylabel("w[n]");
30 title("Window Function")
31 %----- create impulse response -----
32 impulse response = hnT(range1 n, Wc1, Wc2, Ws, T); function for ideal filter
33 stem(range1 n, impulse response)
34 xlabel("n"); ylabel("h[n]");
35 grid on
36 title ("Ideal Filter Response");
37 \text{ range2 n} = 0:1:M;
38 stem(range2 n, impulse response); %causal response
39 title ("Causal Impulse Response");
40 grid on;
```

```
41 % Determine the impulse response of the filter - with causal condition
42 imp = impulse response.*win;
43 stem(range2 n, imp);
44 title ("Impulse Response of the filter - Causal");
45 xlabel("n"); ylabel("h[n]")
46 grid on;
47 %----- compute the magnitude response -----
48 [h, W] = freqz(imp);
49 W = W/T;
50 h = 20*log10(abs(h));
51 plot(W, h);
52 title ("Response of the filter in Frequency Domain");
53 xlabel("Frequency (rad/s)"); ylabel("Magnitude (dB)")
54 grid on;
55 % -----
56 fvtool(imp) % determine magnitude response using function
57 lower limit = round(length(W)/(Ws/2)*Wc1);
58 upper limit = round(length(W)/(Ws/2)*Wc2);
59 w pass = W(lower limit:upper limit);
60 h pass = h(lower limit:upper limit);
61 axis([-inf, inf, -0.2, 0.2]); %zooming version of pass band
62 grid on;
63 title("PassBand of the filter");
64 xlabel("Normalized Frequency"); ylabel("Magnitude (dB)");
65 plot(w pass,h pass);
66 sample = 300; %select an appropriate samples
67 nfft = 2^nextpow2(sample); % create length
68 W 1 = output (5)/2;
                                    % define omega values
69 W 2 = (output(3) + output(4))/2;
70 \text{ W}^{-}3 = (\text{output}(6) + \text{output}(7)/2)/2;
71 nT = 0: T: sample*T;
72 %----- Create an input Signal -----
73 XnT = sin(W 1*nT) + sin(W 2*nT) + sin(W 3*nT);
74 X W = fft(XnT, nfft);
75 X 1 = T*abs(X W(1:nfft/2 + 1));
76 %----- Output Signal of the filter -----
77 HnT = fft(imp, nfft);
78 Y W = HnT.*X W;
79 y\bar{t} = ifft(Y \bar{W}, nfft);
80 Y = T*abs(Y W(1: nfft/2 + 1));
```

```
81 \%----- Expected output of the filter (as the ideal one)-----
82 expected ynT = sin(W 2*nT);
83 expected YW = fft(expected ynT, nfft);
84 Y 1 = T*abs(expected YW(1: nfft/2 + 1));
85 %----- time domain plots for comparison -----
86 figure,
87 subplot(3, 1, 1);
88 stem(1:(sample+1), XnT(1:(sample+1)), 'r');
89 title('Input Signal');
90 xlabel('t(s)'); ylabel('x(nT)');
91 axis([0, (sample+1), -4, 4]);
92 subplot(3, 1, 2);
93 stem(1:(sample+1),yt(1:(sample+1)), 'r');
94 title('Output Signal-Designed Filter'); xlabel('t(s)'); ylabel('y(nT)');
95 axis([0, (sample+1), -1.5, 1.5]);
96 subplot(3, 1, 3);
97 stem(1:(sample+1), expected ynT(1:(sample+1)), 'r');
98 title('Output Signal-Expected');
99 xlabel('t (s)'); ylabel('expected y(nT)');
     axis([0, (sample+1), -1.5, 1.5]);
100
101
     %----- frequency domain plots for comparison ------
102
     w = Ws*(0:1/nfft:1/2);
103
     figure, subplot(3, 1, 1);
104
     plot(w, X 1, "g");
105
     title('Frequency Spectrum Of Input Signal');
     xlabel('w (rad/s)'); ylabel('X(w)');
106
     grid on;
107
108
     axis([0, 1800, 0, 0.5]);
109
     subplot(3, 1, 2);
     plot(w, abs(Y'), "g");
110
     title('Frequency Spectrum Of Output Signal using Designed Filter');
111
112
     xlabel('w (rad/s)'); ylabel('Y(w)');
113
     grid on;
114
     axis([0, 1800, 0, 0.5]);
     subplot(3, 1, 3);
115
     plot(w, abs(Y_1'), "g");
116
117
     title('Frequency Spectrum Of Expected Output Signal');
     xlabel('w (rad/s)'); ylabel('Y exp(w)');
118
119
     grid on;
     axis([0, 1800, 0, 0.5]);
120
```

```
121
    %----- choose parameter alfa ------
122
    function [valAlfa] = getAlfa(Attenuation)
123
    if Attenuation <= 21</pre>
         valAlfa = 0;
124
    else
    if Attenuation > 50
         valAlfa = 0.1102* (Attenuation - 8.7);
    else
         valAlfa = 0.5842* (Attenuation - 21)^0.4 + 0.07886* (Attenuation-
    21);
    end
125
    end
126
    end
127
    128
    function [valD] = getD(Attenuation)
129
    if Attenuation <= 21</pre>
         valD = 0.9222;
130
    else
         valD = (Attenuation - 7.95)/(14.36);
131
    end
132
    end
133
    %----- window function ------
134
    function [i0 x] = bessel(x)
135
    i0 x = 1;
    for k = 1: 10
       i0 x = i0 x + (((x/2).^k)*(1/factorial(k))).^2;
    end
136
    end
137
    %----- ideal filter function ------
138
    function [impulse_response] = hnT(n, Wc1, Wc2, Ws, T)
139
    impulse response=zeros(size(n));
140
    for c1 = 1:length(n)
    if n(c1) == 0
         impulse response(c1)=2*(Wc2-Wc1)/Ws;
         impulse response(c1) = (\sin(n(c1)*Wc2*T) -
    \sin(Wc1*n(c1)*T))/(n(c1)*pi);
    end
141
    end
142
    end
```

```
143 %Function for take values the last integers of Index Number
144
     function output= calcMyFilterSpecs(A, B, C)
145
     Ap = 0.03 + (0.01 * A);
146
     Aa = 45 + B;
147
     Wp1 = (C * 100) + 300;
     Wp2 = (C * 100) + 700;

Wa1 = (C * 100) + 150;
148
149
150
     Wa2 = (C * 100) + 800;
     Ws = 2*((C * 100) + 1200);
151
152
     output = [Ap, Aa, Wp1, Wp2, Wa1, Wa2, Ws]
153
     end
```