

Department of Electrical Engineering
University of Arkansas



ELEG5693 Wireless Communications Propagation and Noise Part II

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OUTLINE

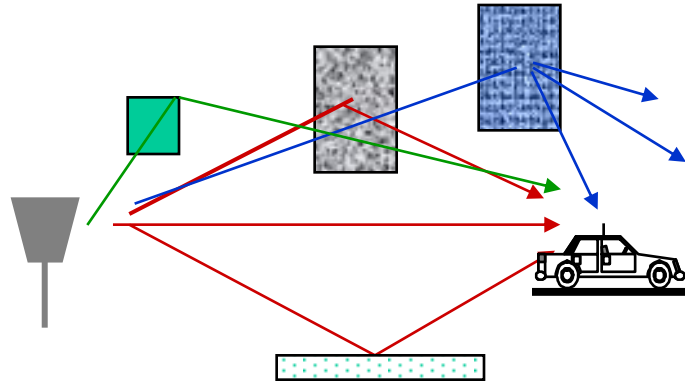
- Wireless channel
- Path loss
- Shadowing
- **Small scale fading**
- Simulation model
- Channel classifications
- Noise and interference

FADING: WHAT IS FADING?

- **Path loss and shadowing is caused by large objects that are distant from MS.**
 - Even the MS is moving, the change in the relative position between MS and those distant large objects is small.
 - Therefore, the impairments caused by those large distant objects change very slow with respect to (w.r.t.) time and position.
 - Shadowing is also referred to as **large scale fading**.
- **Small scale fading is caused by the effects of objects that are close to MS.**
 - The movement of MS w.r.t. nearby small objects will dramatically change the reflection or diffractions of propagated signals.
 - The signal at receiver (sum of the signals from all multiple paths) will change **rapidly** with the movement of MS.

Small scale fading: rapid fluctuation of the received signals over short distance.

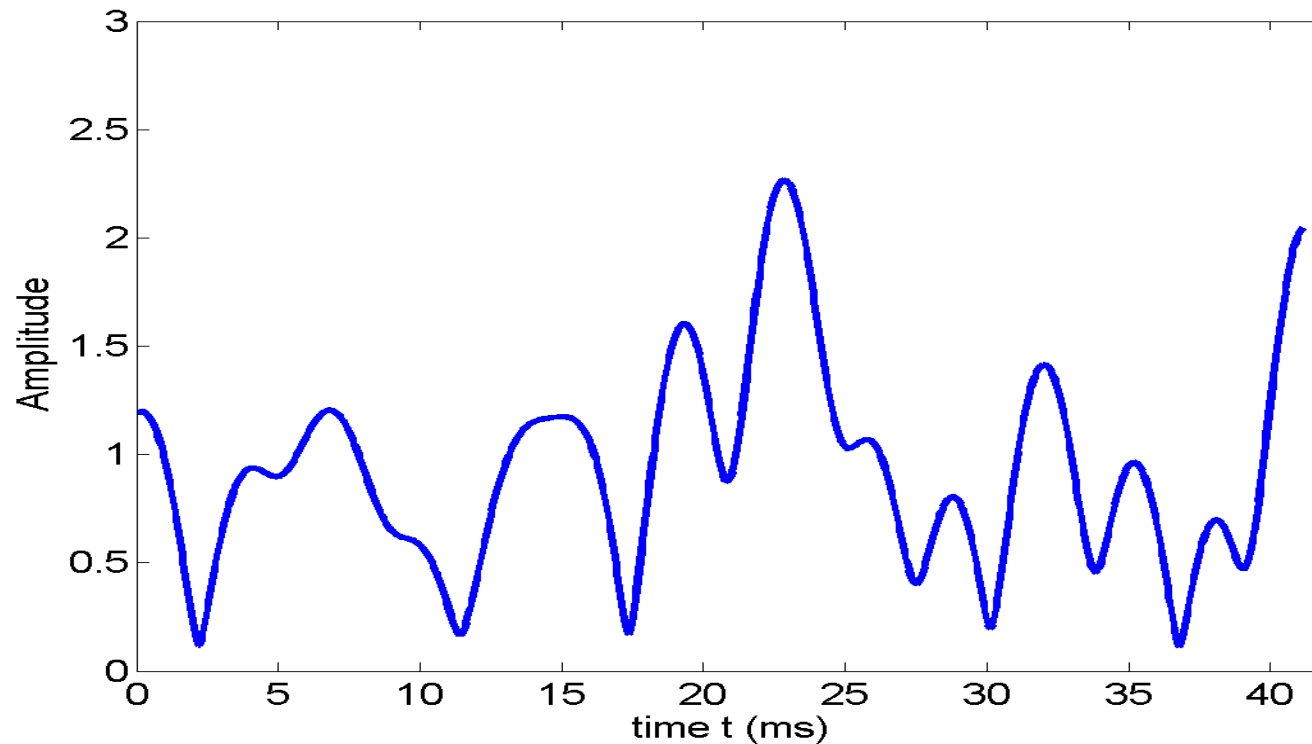
FADING: WHAT IS FADING?



- **Random # of multipath components**

- The amplitude, phase, and frequency of each component change w.r.t. the movement of MS.
- The signal at the receiver is the summation of all the multipath components → the amplitude, phase, and frequency of the received signal at receiver change w.r.t. the movement of MS.
- The movement of surrounding objects (e.g. vehicles) will also cause the time variation of the signals.

FADING: AN EXAMPLE

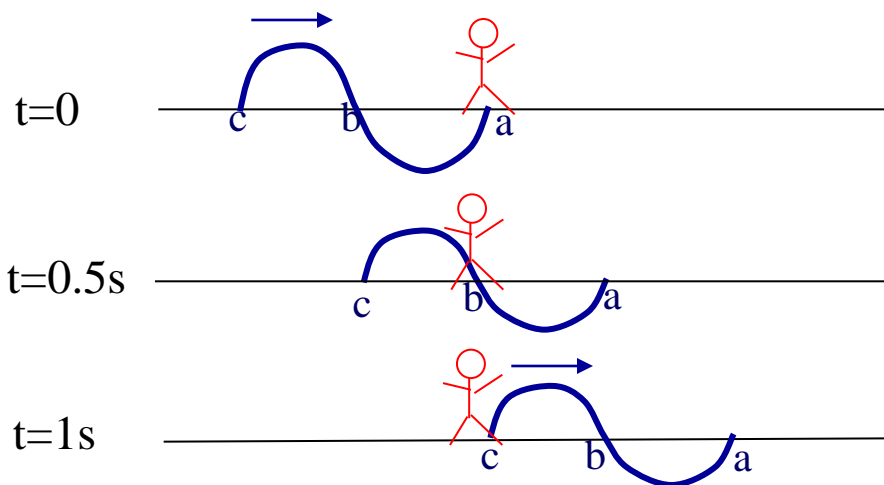


- **The rate of variation depends on two factors:**
 - Relative movement speed between Tx and Rx
 - Speed of surrounding objects

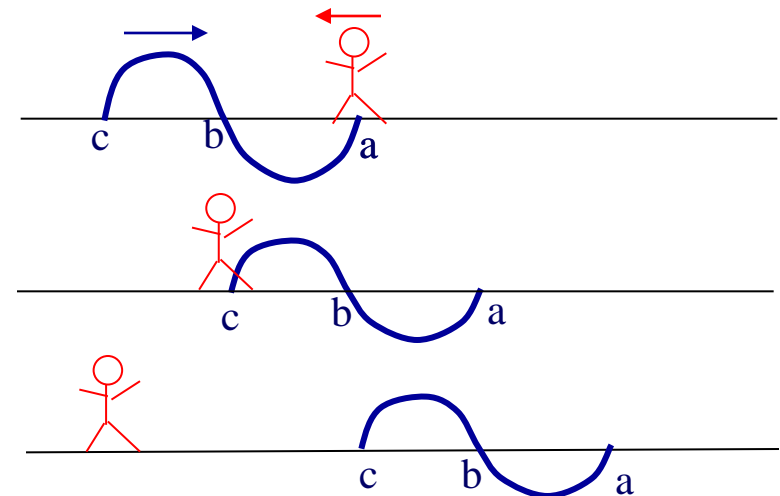
FADING: DOPPLER

- **What is Doppler?**

- The whistle of the train coming from opposite direction sounds different with the train passing by.
 - The pitch of the sound (determined by sound frequency) is changing.
- Rx signal frequency will change if the Rx is moving w.r.t. Tx.
- Signal frequency change due to the relative movement between Tx and Rx is called **Doppler effects**.



$T = 1s$



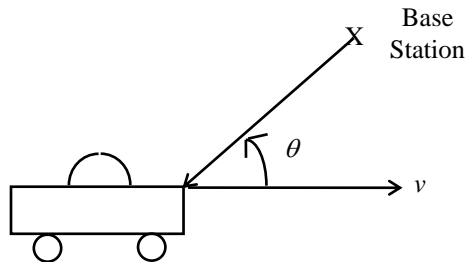
$T = 0.5s$

FADING: DOPPLER

- **Consider Tx sends out a sinusoid with frequency 1Hz**
 - If Rx moves toward Tx, the signal observed by Rx will have a shorter period → frequency increased
 - If Rx moves away from Tx, the signal observed by Rx will have a longer period → frequency decreased
- **The amount of frequency change is called Doppler shift**
 - Doppler shift depends on
 - Relative speed between Tx and Rx
 - The frequency of the original signal

FADING: DOPPLER

- Relationship between speed and Doppler shift



$$\Delta f = \frac{v}{\lambda} \cos \theta$$

v : relative speed λ : wavelength

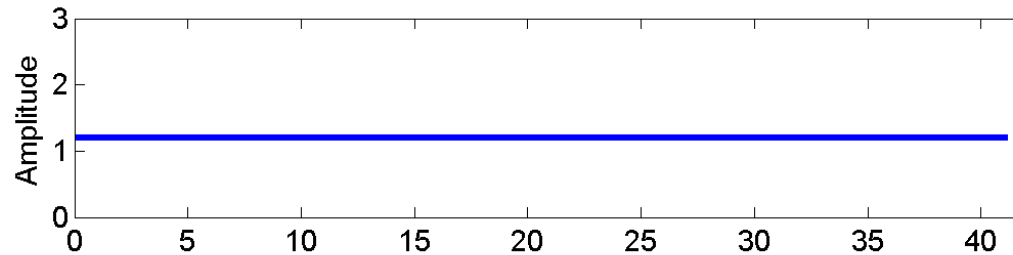
- Maximum Doppler shift:

$$f_D = \frac{v}{\lambda}$$

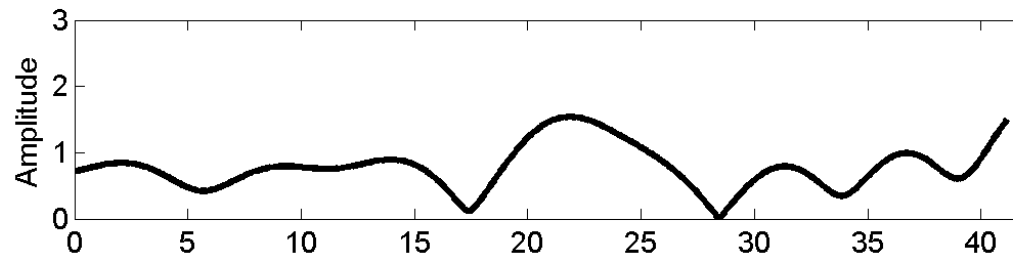
- Example:** find the maximum Doppler shift of 900MHz system with mobile speed 120km/Hr

FADING: DOPPLER

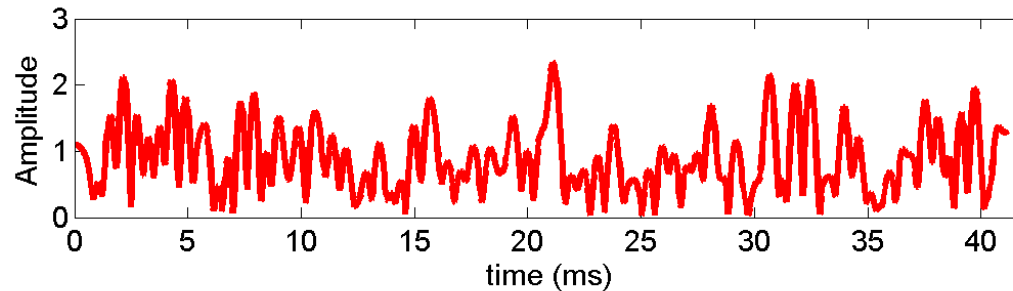
- At given frequency
 - $v \uparrow \Rightarrow f_D \uparrow \Rightarrow$ channel changes more rapidly



$f_D = 0\text{ Hz}$



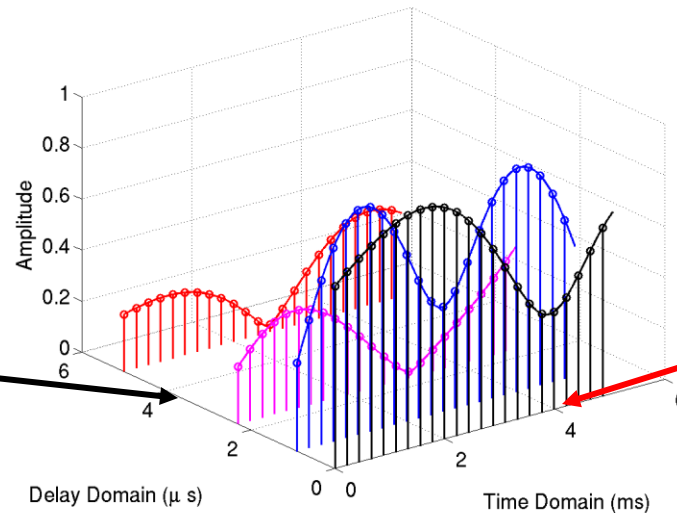
$f_D = 100\text{ Hz}$



$f_D = 1000\text{ Hz}$

FADING: IMPULSE RESPONSE

- The impulse response of fading is time-varying!



Relative multipath
delay τ

Time variation t

N multipath components

$$c(t, \tau) = \sum_{n=1}^N \alpha_n(t) \cos[2\pi f_c t + \varphi_n(t)] \times \delta(\tau - \tau_n)$$

- f_c : system operating frequency (e.g. 900MHz, 1.8GHz)
- t : the time variation (both amplitude and phase changes with respect to time)
- τ : relative delay between multipath components
- $\varphi_n(t)$: depends on path distance and Doppler shift $(2\pi f_D t)$

FADING: IMPULSE RESPONSE

- Complex baseband representation**

$$\begin{aligned}
 c(t, \tau) &= \sum_{n=1}^N \alpha_n(t) \operatorname{Re} \left[e^{j2\pi f_c t + j\varphi_n(t)} \right] \times \delta(\tau - \tau_n) \\
 &= \operatorname{Re} \left\{ e^{j2\pi f_c t} \left[\sum_{n=1}^N \alpha_n(t) e^{j\varphi_n(t)} \times \delta(\tau - \tau_n) \right] \right\}
 \end{aligned}$$

$$h(t, \tau) = \sum_{n=1}^N \alpha_n(t) e^{j\varphi_n(t)} \times \delta(\tau - \tau_n)$$

- Maximum delay spread**

- The time interval between the first multipath and the last multipath

$$\tau_{\max} = \tau_N - \tau_1$$

FADING: FLAT FADING

- **Flat fading**

- Maximum delay spread $\tau_{\max} \ll$ system symbol period T_s
- Relative to the symbol period, all the multipath components arrive at almost the same time → **Doesn't need to consider the delay variable τ**

$$h(t) = \sum_{n=1}^N \alpha_n(t) e^{j\varphi_n(t)} = \sum_{n=1}^N \alpha_n(t) \cos[\varphi_n(t)] + j \cdot \sum_{n=1}^N \alpha_n(t) \sin[\varphi_n(t)]$$

- Inphase component

$$h_I(t) = \sum_{n=1}^N \alpha_n(t) \cos[\varphi_n(t)]$$

- Quadrature component

$$h_Q(t) = \sum_{n=1}^N \alpha_n(t) \sin[\varphi_n(t)]$$

FADING: FLAT FADING

$$h(t) = h_I(t) + j \cdot h_Q(t)$$

- **Both $h_I(t)$ and $h_Q(t)$ are the sum of many multipath components**
 - Each multipath component is a random process
 - $h_I(t)$ and $h_Q(t)$ are random processes
- **Central limit theorem**
 - The sum of N independent and identically distributed (i.i.d.) random variables tends to Gaussian distribution when N is large enough.
- **Based on central limit theorem, at time t , both $h_I(t)$ and $h_Q(t)$ are Gaussian distributed!**

FLAT FADING: RAYLEIGH FADING

- If there is **no LOS** between Tx and Rx
 - $h_I(t)$ and $h_Q(t)$ are zero-mean Gaussian distributed $\sim N(0, \sigma^2)$
 - The amplitude (or envelope) of $h(t)$

$$|h(t)| = \sqrt{h_I^2(t) + h_Q^2(t)}$$

- The fading envelope $|h(t)|$ follows **Rayleigh** distribution

$$f_{|h(t)|}(z) = \frac{z}{\sigma^2} \exp\left(-\frac{z^2}{2\sigma^2}\right)$$

- Average power of fading

$$E[|h(t)|^2] = E[h_I^2(t)] + E[h_Q^2(t)] = 2\sigma^2$$

FLAT FADING: RICIAN FADING

- If there is LOS component

- $h_I(t)$ and $h_Q(t)$ are non-zero-mean Gaussian distributed $\sim N(s, \sigma^2)$
- The fading envelope

$$|h(t)| = \sqrt{h_I^2(t) + h_Q^2(t)}$$

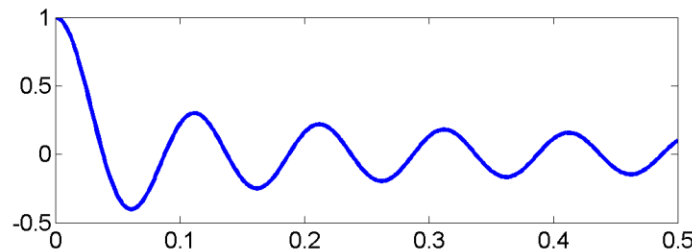
follows Rician distribution

FLAT FADING: TIME DOMAIN CORRELATION

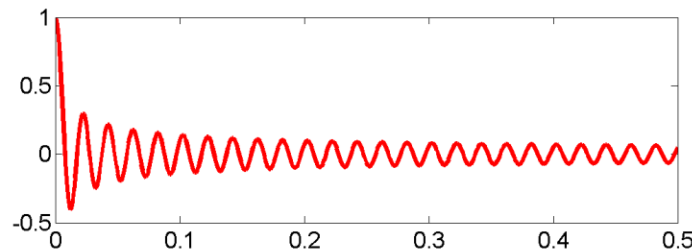
- The time domain correlation of $h(t)$ is

$$R_h(\alpha) = E[h(t + \alpha)h^*(t)] = P_h \cdot J_0(2\pi f_D \alpha)$$

- $J_0(x)$: zero-order Bessel function of the first kind



$f_D = 10\text{Hz}$

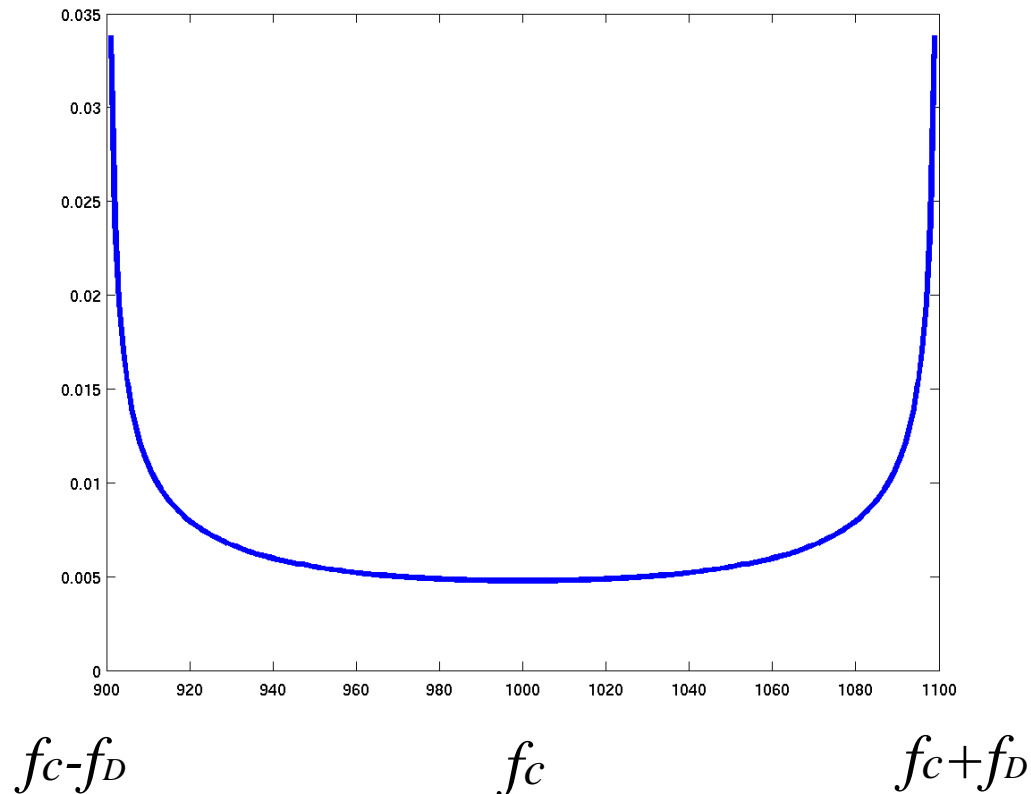


$f_D = 50\text{Hz}$

- Generally speaking, for given time interval α
 - Larger speed $v \rightarrow$ larger $f_D \rightarrow$ smaller $|R_h(\alpha)|$.

FLAT FADING: POWER SPECTRAL DENSITY

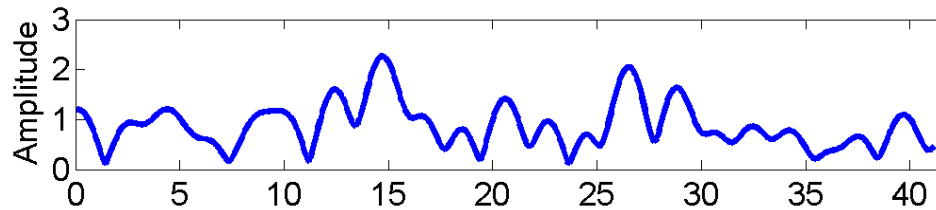
- Power spectral density is the Fourier transform of auto-correlation function.



OUTLINE

- Wireless channel
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- Small scale fading
- **Simulation model**
- Channel classifications
- Noise and interference

SIMULATOR



$$h(t) = h_I(t) + j \cdot h_Q(t)$$

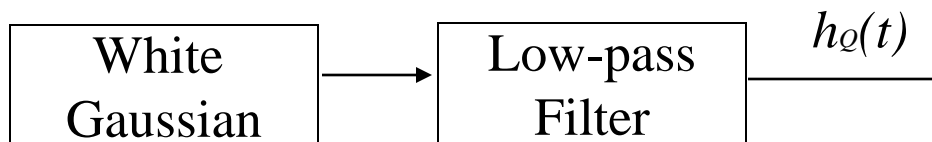
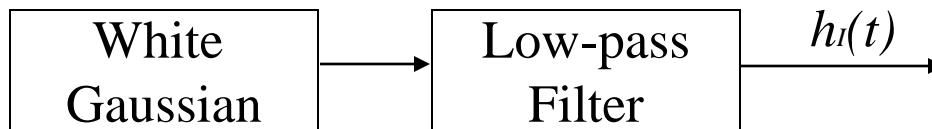
- **Flat Rayleigh fading is a random process**

- At any time instant t , $|h(t)| = \sqrt{h_I^2(t) + h_Q^2(t)}$ is Rayleigh distributed.
 - Both the real part $h_I(t)$ and the imaginary part $h_Q(t)$ are zero mean Gaussian distributed.
- The auto-correlation function must satisfy

$$R(\tau) = E[h(t + \tau)h^*(t)] = J_0(2\pi f_D \tau)$$

SIMULATOR

- **How to generate flat Rayleigh fading with computer program?**
 - Method 1: Filtered Gaussian noise
 - Rely on low-pass filter to introduce the time-domain correlation among symbols



- The low-pass filter is hard to design.

SIMULATOR

- Method 2: Sum-of-sinusoid**

$$h_I(nT_s) = \frac{1}{\sqrt{M}} \sum_{m=1}^M \cos \left\{ 2\pi f_D \cos \left[\frac{(2m-1)\pi + \theta}{4M} \right] \cdot nT_s + \alpha_m \right\}$$

$$h_Q(nT_s) = \frac{1}{\sqrt{M}} \sum_{m=1}^M \sin \left\{ 2\pi f_D \cos \left[\frac{(2m-1)\pi + \theta}{4M} \right] \cdot nT_s + \beta_m \right\}$$

$$h(nT_s) = h_I(nT_s) + j \cdot h_Q(nT_s)$$

$\theta, \alpha_m, \beta_m$: uniformly distributed in $[0, 2\pi]$

M : a constant. The larger, the more accurate. Usually 8 or 16.

T_s : time duration between samples.

$\mathbf{h} = \text{Rayleigh}(\mathbf{N}, \mathbf{f}_D, \mathbf{T}_s)$

$\mathbf{h} = [h(0T_s), h(1T_s), h(2T_s), \dots, h((N-1)T_s)]$

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- Wireless channel
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- Noise and interference

CLASSIFICATION

- **Fading**
 - Amplitude and phase distortions of transmitted signal
- **Classification criteria**
 - Scale
 - Large scale fading, small scale fading
 - Small scale fading
 - Flat fading v.s. frequency selective fading
 - Fast fading v.s. slow fading
 - Rayleigh fading v.s. Rician fading

CLASSIFICATION: SCALE

- **Large scale fading**

- Path loss (signal power loss as a function of distance)
 - Due to distance between Tx and Rx, reflection of large objects
- Shadowing
 - Obstruction from large objects

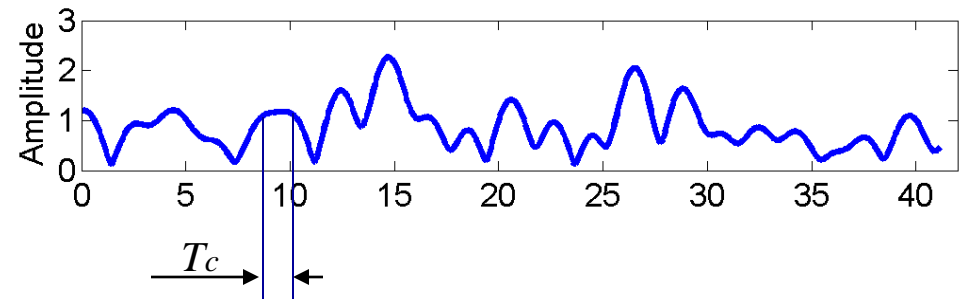
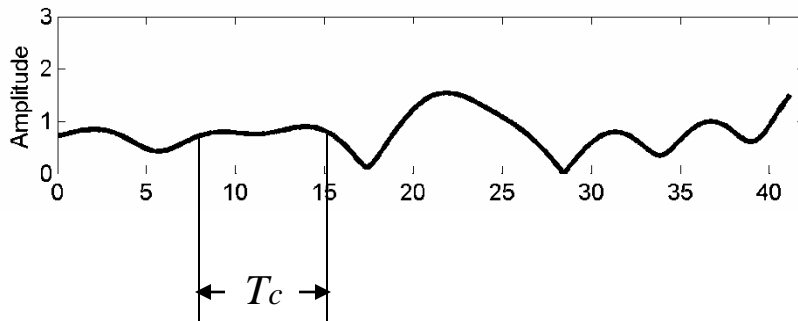
- **Small scale fading**

- Amplitude and phase distortions from local objects → highly sensitive to locations of MS
- Due to the superposition of multiple electromagnetic waveforms
- Caused by two independent propagation mechanisms
 - (1) time dispersion (delay spread)
 - Determines frequency selective or flat
 - (2) frequency dispersion (Doppler spread)
 - Determines fast or slow

CLASSIFICATION: FAST FADING V.S. SLOW FADING

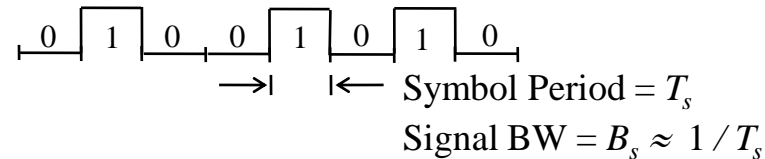
- **The time domain variation of fading is determined by maximum Doppler spread f_D**
 - Doppler shift: signal frequency change due to relative movement between Tx and Rx.
 - Larger speed $v \rightarrow$ larger $f_D \rightarrow$ channel varies faster.
- **Coherence time T_c**
 - The time period over which the channel is strongly correlated (didn't change too much)
 - Inverse proportional to f_D

$$T_c \approx \frac{1}{2f_D}$$



CLASSIFICATION: FAST FADING V.S. SLOW FADING

- System symbol period v.s. signal bandwidth



- **Fast fading**

– If $T_s > T_c$, or $B_s < f_D$

- $T_s > T_c$: channel changes within one symbol period → fast fluctuation

- **Slow fading**

– If $T_s \ll T_c$, or $B_s \gg f_D$

- $T_s \ll T_c$: channel keeps constant during several symbol periods → slow amplitude fluctuation.

(Coherence time T_c , Doppler spread f_D) is related to fast fading or slow fading

CLASSIFICATION: FAST FADING V.S. SLOW FADING

- **Example: A cell phone user is in a vehicle moves at a speed of 120km/hr. The carrier frequency is 1800MHz.**
 - (a) What is the maximum Doppler spread?
 - (b) What is the coherence time of the channel?
 - (c) The symbol period of a system is 3ms. Is the system experiencing fast fading or slow fading?
 - (d) The symbol rate of IS-136 system is 24.3ksym/s. Is the system experiencing fast fading or slow fading?

CLASSIFICATION: FLAT V.S. FREQUENCY SELECTIVE

- Maximum delay spread**

- The time interval between the first multipath and the last multipath

$$\tau_{\max} = \tau_N - \tau_1$$

- Mean delay spread**

$$\bar{\tau} = \sum_{n=1}^N \frac{P_n}{P_{total}} \cdot \tau_n$$

$$P_{total} = \sum_{n=1}^N P_n : \text{ the total power of the all multipath}$$

P_n : the average power of the n th multipath

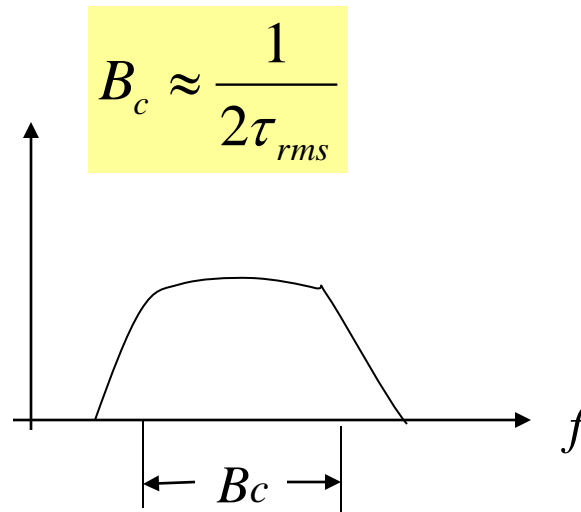
- Root mean square (rms) delay spread**

$$\tau_{rms} = \sqrt{\sum_{n=1}^N \frac{P_n}{P_{total}} \cdot (\tau_n - \bar{\tau})^2} = \sqrt{\sum_{n=1}^N \frac{P_n}{P_{total}} \cdot \tau_n^2 - \bar{\tau}^2}$$

CLASSIFICATION: FLAT V.S. FREQUENCY SELECTIVE

- **Coherence bandwidth B_c**

- The bandwidth over which the channel is strongly correlated (didn't change too much)
 - The spectrum over coherence bandwidth is almost “flat”
- Inverse proportional to rms delay spread

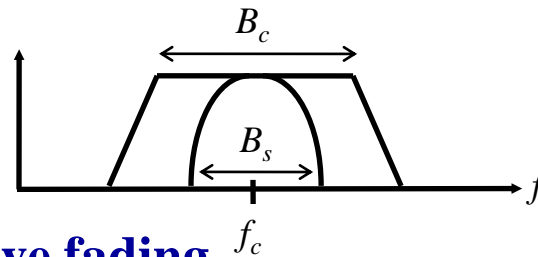


(Coherence bandwidth B_c , rms delay spread τ_{rms}) is related to fast fading or slow fading

CLASSIFICATION: FLAT V.S. FREQUENCY SELECTIVE

- **Flat fading**

- If $B_s \ll B_c$, or $T_s \gg \tau_{rms}$
- $B_s \ll B_c$: signal bandwidth \ll channel bandwidth
- $T_s \gg \tau_{rms}$: relative arrival time between multipath components is negligible
 - Doesn't need to consider delay variable $h(t, \tau) \rightarrow h(t)$



- **Frequency selective fading**

- If $B_s \gg B_c$, or $T_s \ll \tau_{rms}$
- $B_s \gg B_c$: signal bandwidth \gg channel bandwidth
 - Signal spectrum will be seriously distorted by channel!
- $T_s \ll \tau_{rms}$: symbol period smaller than rms delay spread
 - The relative arrival time between the multipath components is no longer negligible!

CLASSIFICATION: FREQUENCY SELECTIVE FADING

- **Frequency selective fading**

- rms delay spread $\tau_{rms} \gg$ system symbol period T_s
 - The relative arrival time between the multipath components is no longer negligible!
- The N multipath components are divided into L clusters
 - Within each cluster, there are still many multipath components
 - Multipath components belonging to the l th cluster arrives at approximately the same time τ_l .

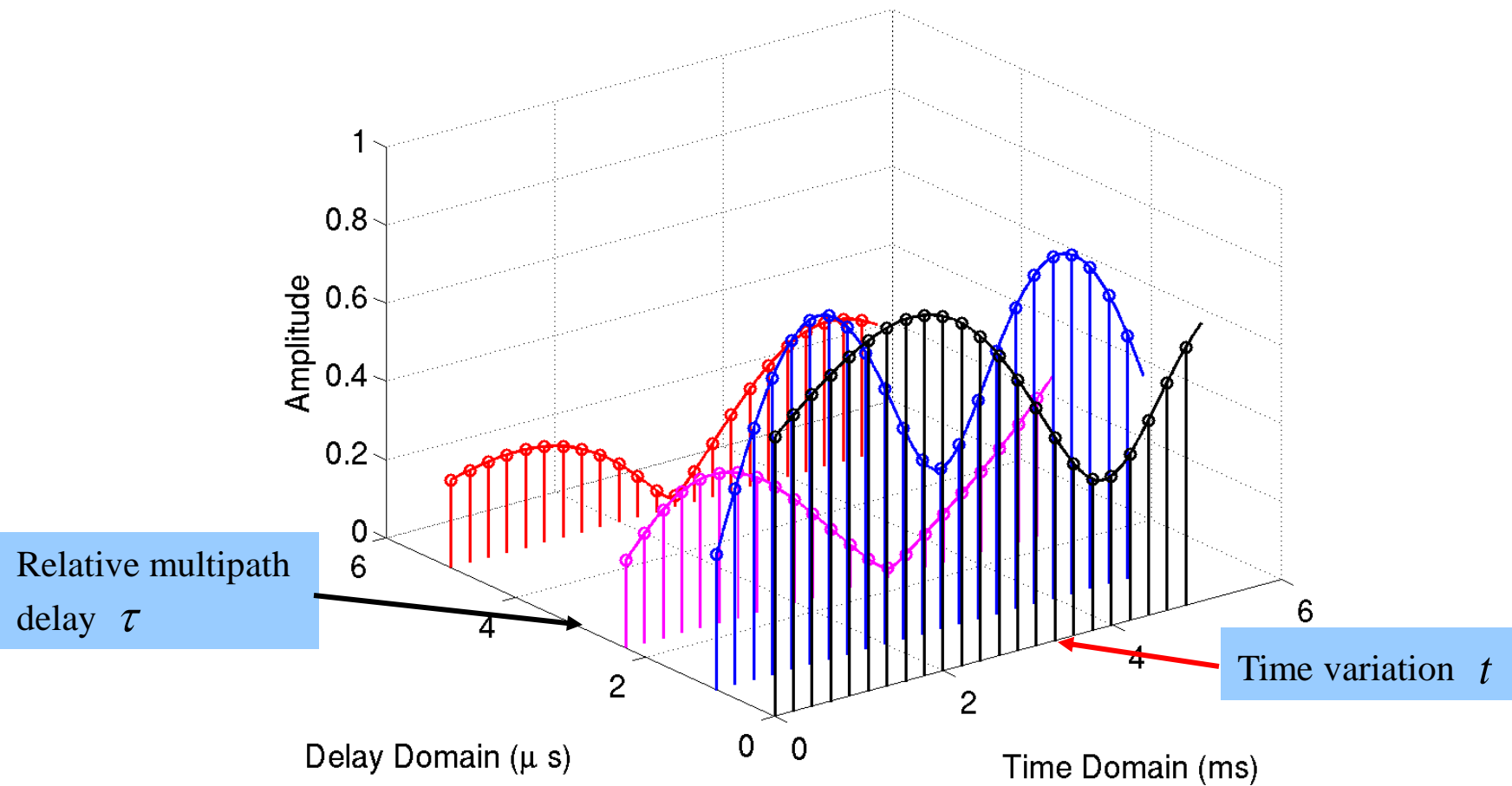
$$h(t, \tau) = \sum_{l=1}^L h_l(t) \times \delta(\tau - \tau_l) = \sum_{l=1}^L [h_{ll}(t) + jh_{ql}(t)] \times \delta(\tau - \tau_l)$$

- $h_l(t)$ is the sum of all the multipath components within the same cluster
 - Resolvable multipath component
- The inphase and quadrature components of $h_l(t)$ are Gaussian distributed.
- The frequency selective fading can be viewed as the combination of multiple flat fading
 - Each branch (cluster) $h_l(t)$ can be viewed as flat fading

CLASSIFICATION: FREQUENCY SELECTIVE FADING

- Each branch of frequency selective fading can be viewed as a flat fading
 - All the properties discussed for flat fading can be directly applied to each branch of frequency selective fading
 - E.g. Inphase and quadrature components are Gaussian distributed.
 - fading envelope: Rayleigh v.s. Rician

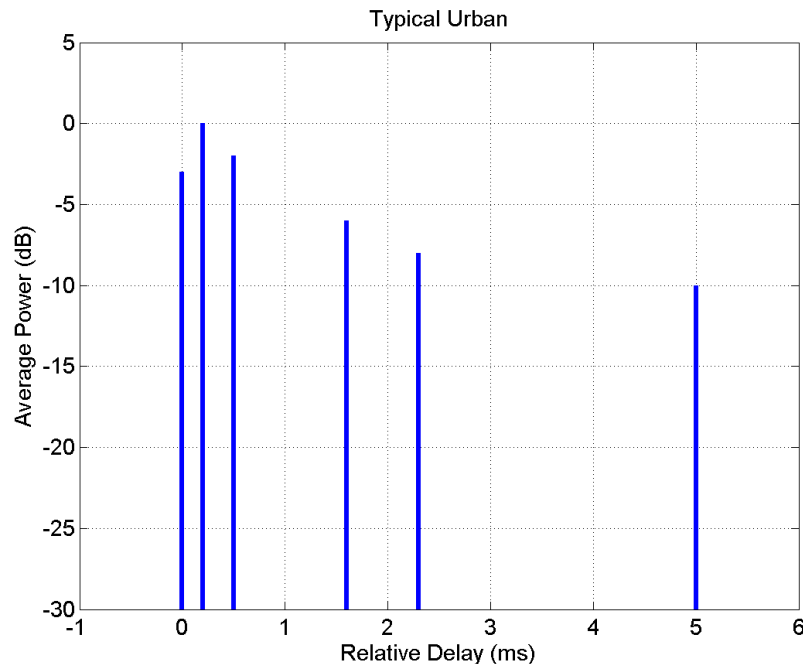
CLASSIFICATION: FREQUENCY SELECTIVE FADING



CLASSIFICATION: FREQUENCY SELECTIVE FADING

- **Power delay profile**

- The average power of each resolvable multipath component, w.r.t. the relative delay



Relative delay (ms)	0	0.2	0.5	1.6	2.3	5.0
Average power	0.1897	0.3785	0.2388	0.0951	0.0600	0.0379

CLASSIFICATION: FREQUENCY SELECTIVE FADING

- **Example:**

- (a) Find the maximum delay spread, mean delay spread, and rms delay spread of the following power delay profile.
- (b) What is the coherence bandwidth of the channel?
- (c) For a system with symbol rate 0.5KHz, is this a flat fading or frequency selective fading?
- (d) For a system with symbol rate 1000KHz, is this a flat fading or frequency selective fading?

Relative delay (ms)	0	1
Average power	0.4	0.6

CLASSIFICATION: RAYLEIGH V.S. RICIAN

- **Fading envelope**

$$|h(t)| = \sqrt{h_I^2(t) + h_Q^2(t)}$$

- **Rayleigh fading**

- Fading envelope $|h(t)|$ follows Rayleigh distribution
- Non LOS (no dominant multipath components)

- **Rician fading**

- Fading envelope $|h(t)|$ follows Rician distribution
- One dominant component (LOS) along with weaker multipath signals

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NOISE

- **Noise and interference**

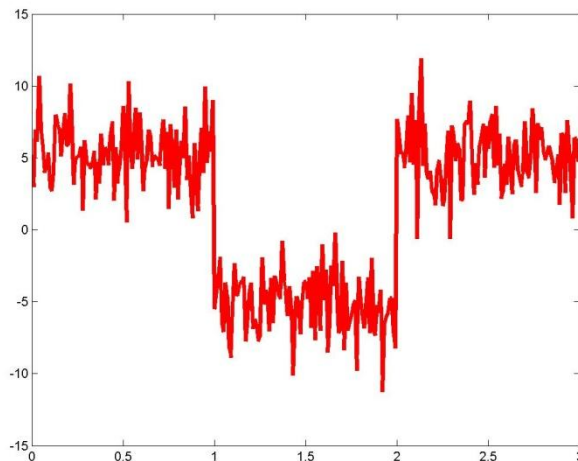
- Unwanted electrical signals interfering with the desired signal
- Arises from outside natural or artificial sources
 - Artificial source: noise from automobile ignition, signal from other communication system, etc.
 - Natural source: thermal noise, atmospheric disturbances.

- **Noise v.s. fading**

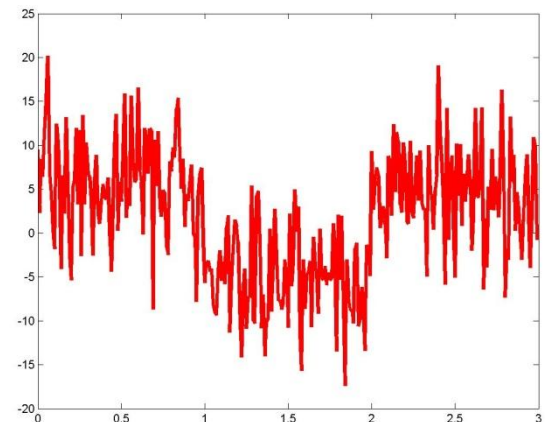
- Noise arises from outside sources
 - fading arises from the signal propagation itself
- Noise is added to the desired signal → the desired signal is buried by noise (noise only has negative effects on signal).
 - Fading results in signal power fluctuation → signal power may become larger or become smaller (fading might benefit system performance).
- Noise is present in all communication systems.
 - Fading is unique to wireless communications.

NOISE: SIGNAL TO NOISE RATIO

- **Signal to noise ratio (SNR):**
 - the ratio of the signal power to the noise power **at the receiver**.
 - $SNR = S/N$, with S being the signal power, and N being the noise power observed by the receiver.
 - High SNR \rightarrow Signal is strong, and noise is weak \rightarrow Better communication quality.
 - Improve SNR \rightarrow Improve Tx power \rightarrow More power consumption



High SNR



Low SNR

NOISE: THERMAL NOISE

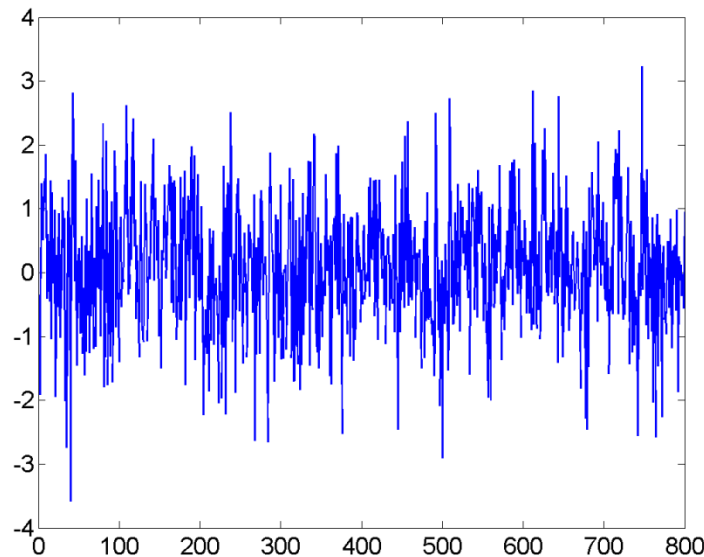
- **Thermal noise**

- At the temperature **above** 0K (absolute temperature, =-273 centigrade), the electrons inside the conductor will move randomly.
- This random movement of electrons will cause random voltage fluctuations of the transmitted signals.
- Temperature increase → electrons movement becomes stronger → noise power increase

NOISE: THERMAL NOISE

- **Thermal noise is a random process**
 - At any time instant, the random voltage due to thermal noise follows **Gaussian** distribution with zero mean.
 - The random voltage is caused by the sum of the motions of a large number of electrons (central limit theorem).
 - The noise samples at any two different time instants are **uncorrelated**. Auto-correlation function

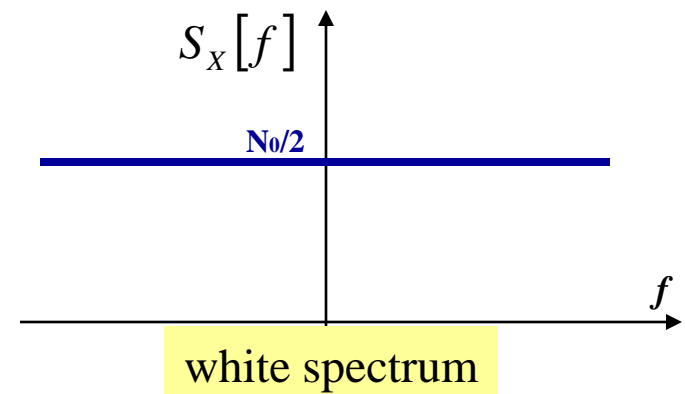
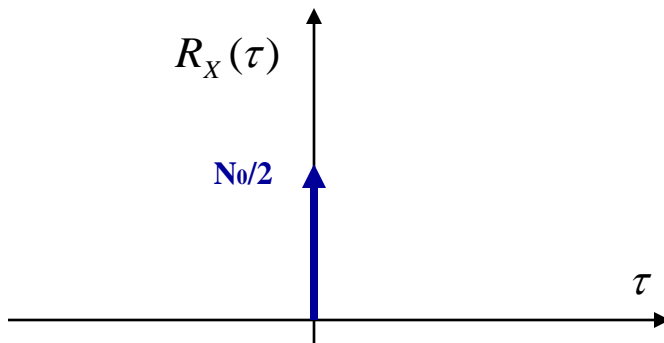
$$R_x(\tau) = E[n(t)n^*(t-\tau)] = \frac{N_0}{2} \delta(\tau)$$



NOISE: THERMAL NOISE

- **Power spectral density**
 - Fourier transform of autocorrelation function

$$S_x(f) = \int_{-\infty}^{+\infty} \frac{N_0}{2} \delta(\tau) e^{-j2\pi f\tau} d\tau = \frac{N_0}{2}$$



Additive White Gaussian Noise (AWGN)

NOISE: COCHANNEL INTERFERENCE

- **Cochannel interference (CCI)**
 - Due to frequency reuse, cells using the same frequency range will generate interference with each other.
 - Unique to cellular system.

