



# ANALOG SIGNAL PROCESSING



ECE 210 & 211  
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Zhejiang University / University of Illinois at Urbana-Champaign Institute



# Objectives

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- **Bode plot (continued)**
- **LTI system response to co-sinusoids input**
- **LTI system response to multifrequency co-sinusoids input**
- **Fourier coefficients of periodic signals**
- **Simplification by Symmetrical considerations**
- **Circuit interpretation for Fourier series**

# Objectives

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# Code Plot

Normally, a transfer function looks like (in general form)

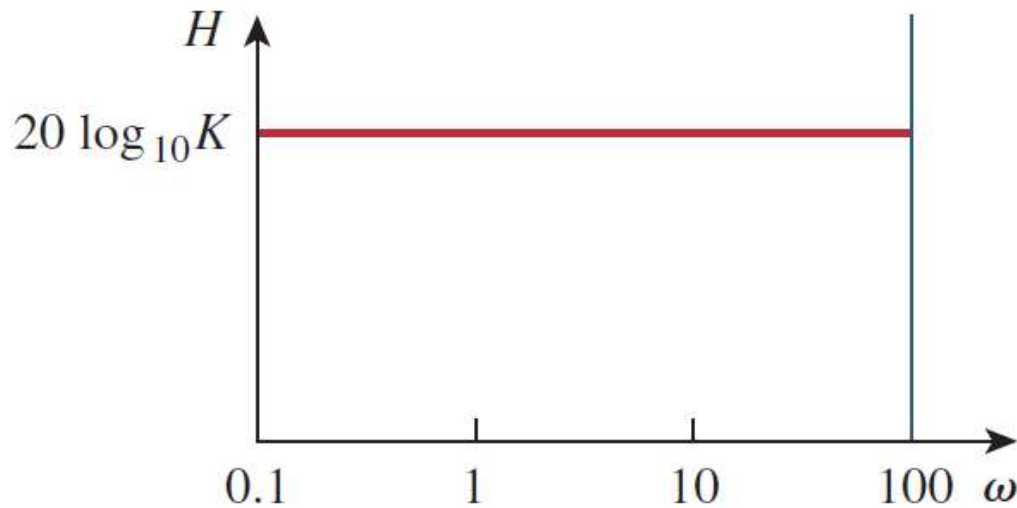
$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$

The initial terms has 7 parts of equation

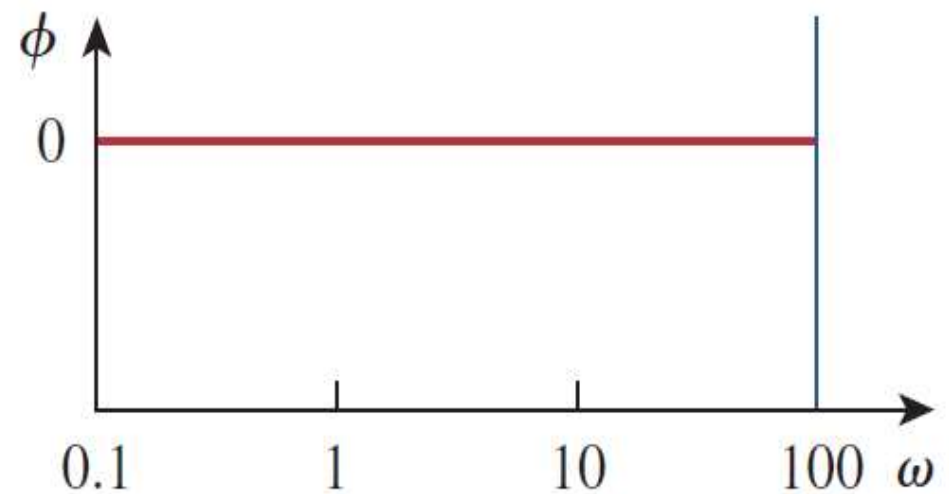
$$\begin{cases} (j\omega)^{+1} = \text{Zero at origin} \\ (j\omega)^{-1} = \text{Pole at origin} \end{cases}$$

# Bode Plot – Gain K

$$H(\omega) = \frac{\boxed{K}(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$



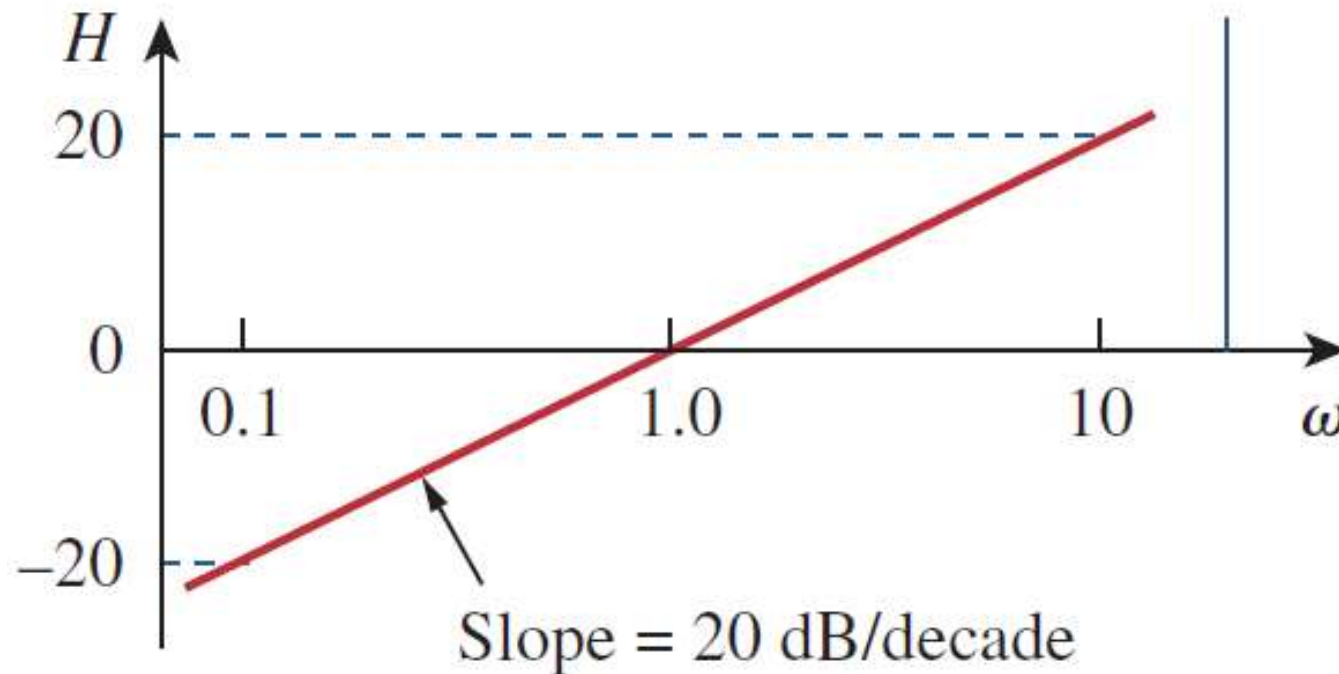
**Magnitude**



**Phase**

# Bode Plot -Pole/zero at the origin

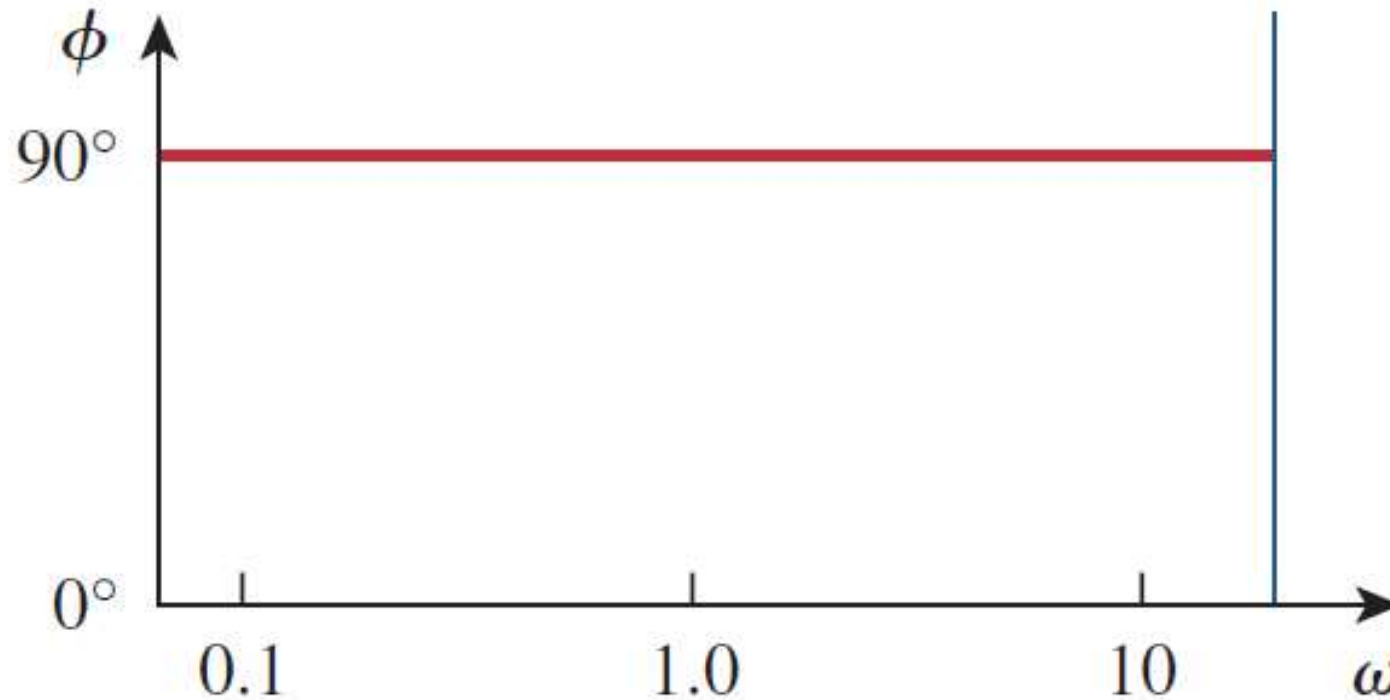
$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$



**Magnitude**

# Bode Plot -Pole/zero at the origin

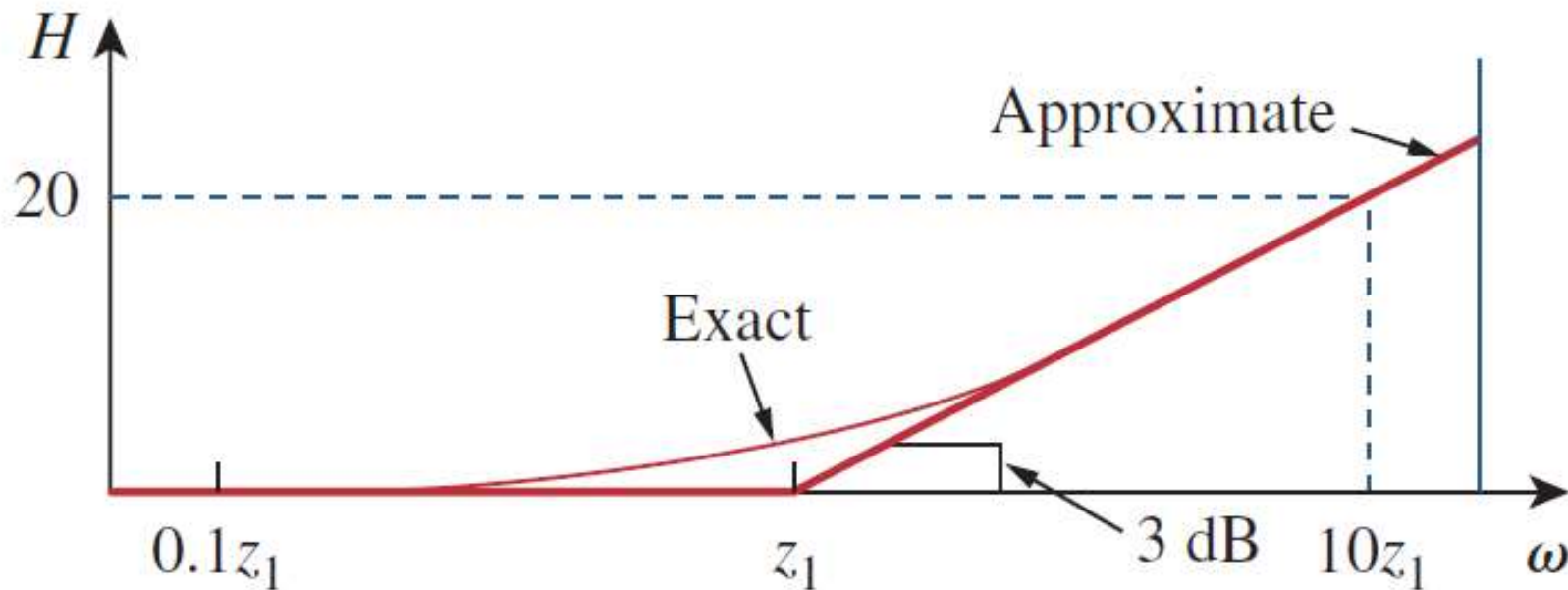
$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$



**Phase**

# Bode Plot – Simple pole/zero

$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$

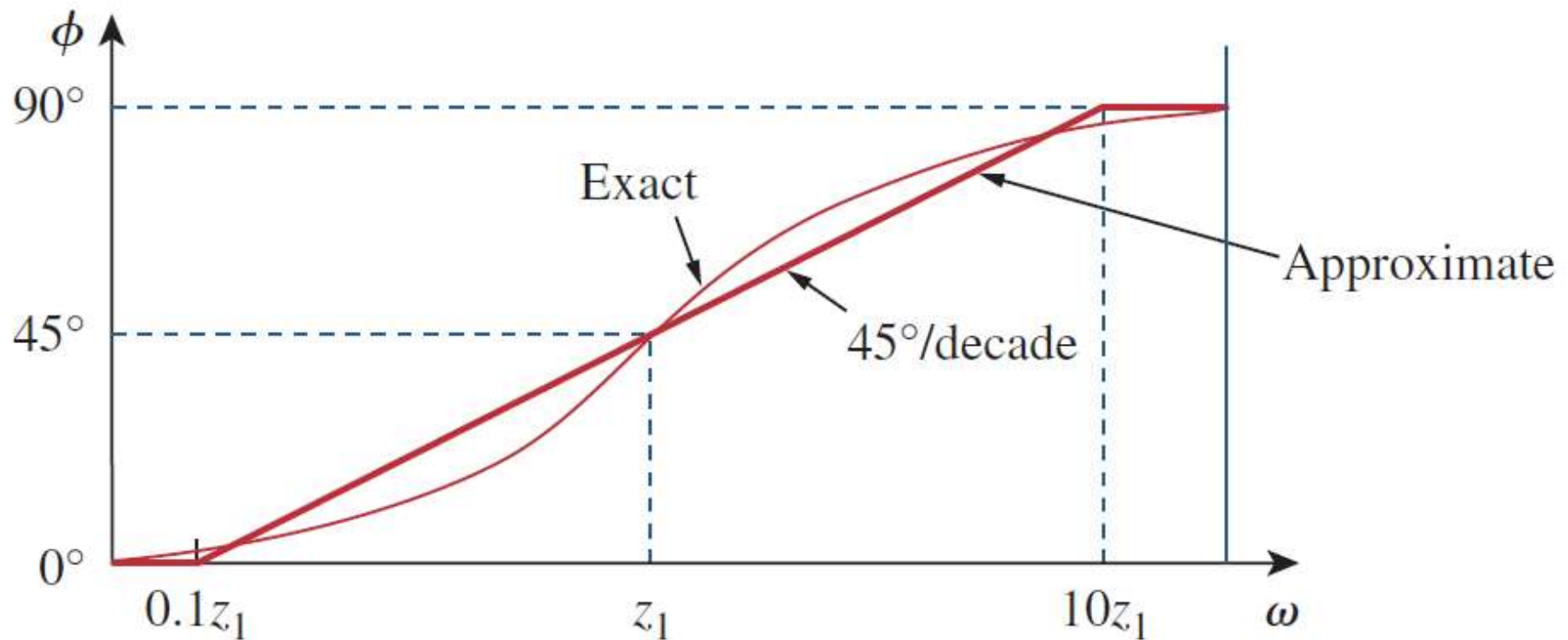


**Magnitude**



# Bode Plot – Simple pole/zero

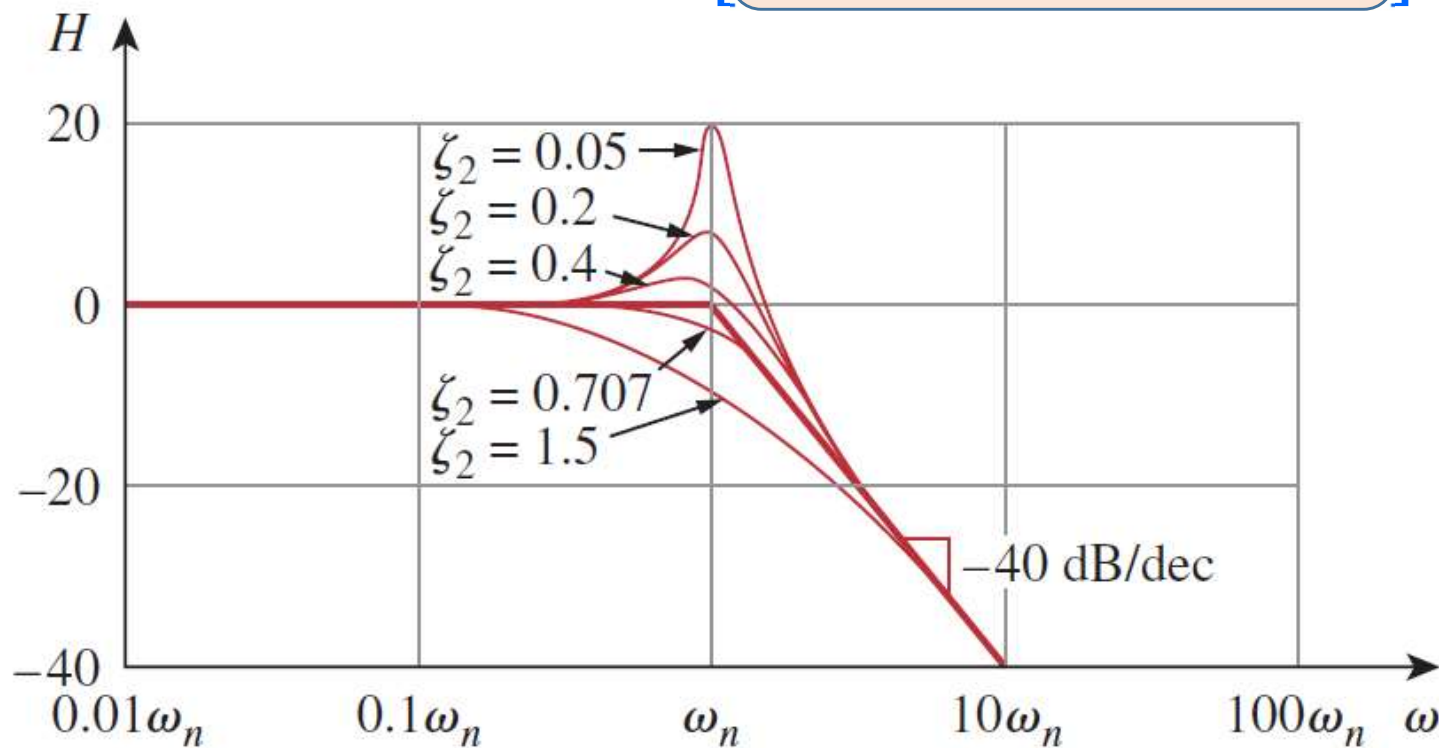
$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$



**Phase**

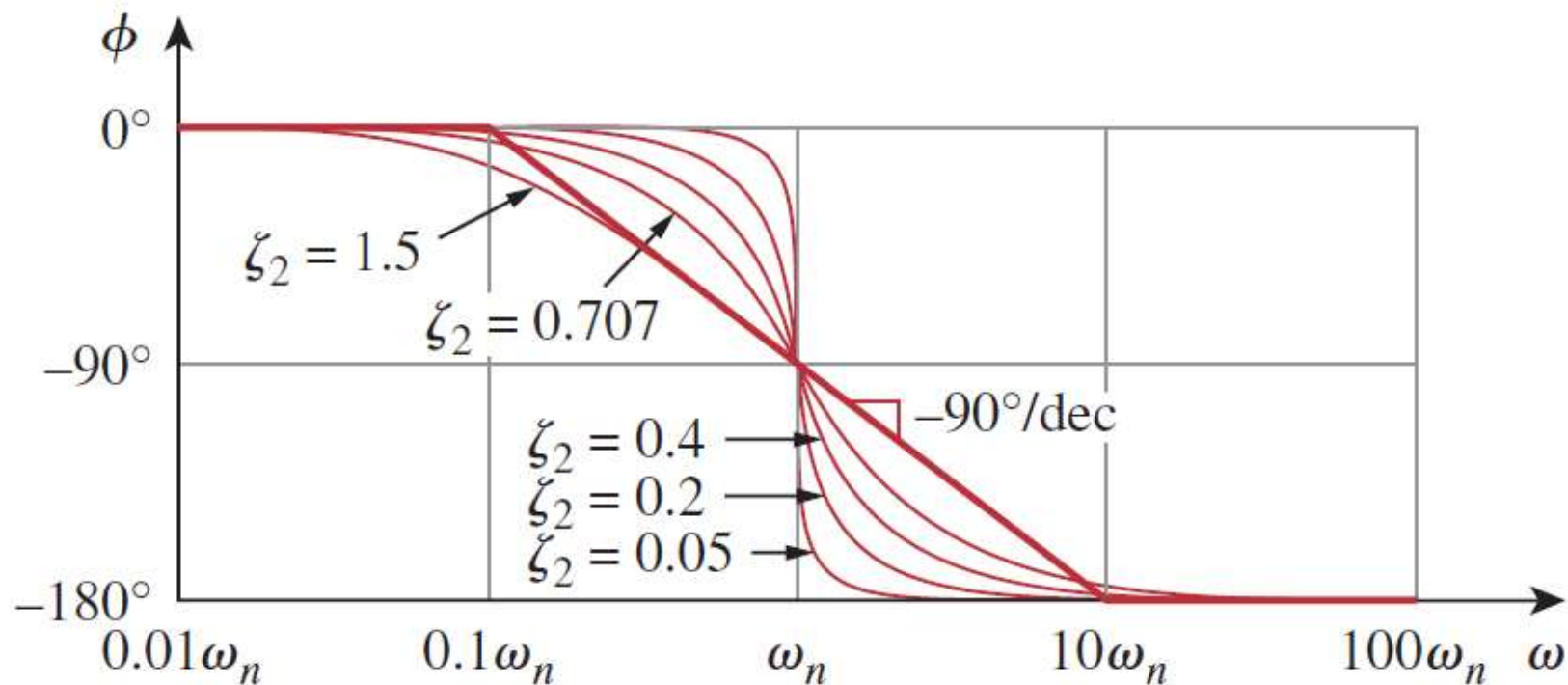
# Bode Plot – Quadratic pole/zero

$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$



# Bode Plot – Quadratic pole/zero

$$H(\omega) = \frac{K(j\omega)^{\pm 1} \left(1 + \frac{j\omega}{z_1}\right) \left[1 + j2\xi_1 \frac{\omega}{\omega_k} + \left(\frac{j\omega}{\omega_k}\right)^2\right] \dots}{\left(1 + \frac{j\omega}{p_1}\right) \left[1 + j2\xi_2 \frac{\omega}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right] \dots}$$



**Phase**

# Bode Plot – Example 1

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**Question:** Construct the Bode plots for the transfer function?

$$H(\omega) = \frac{200j\omega}{(j\omega + 2)(j\omega + 10)}$$

# Bode Plot – Example 1

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## Solution:

We first put  $H(\omega)$  in the **standard form** by dividing out the poles and zeros. Thus,

$$H(\omega) = \frac{10j\omega}{\left(1 + j\frac{\omega}{2}\right)\left(1 + j\frac{\omega}{10}\right)}$$
$$= \frac{10|j\omega|}{\left|1 + j\frac{\omega}{2}\right|\left|1 + j\frac{\omega}{10}\right|} \angle \left(90^\circ - \tan^{-1}\frac{\omega}{2} - \tan^{-1}\frac{\omega}{10}\right)$$

# Bode Plot – Example 1

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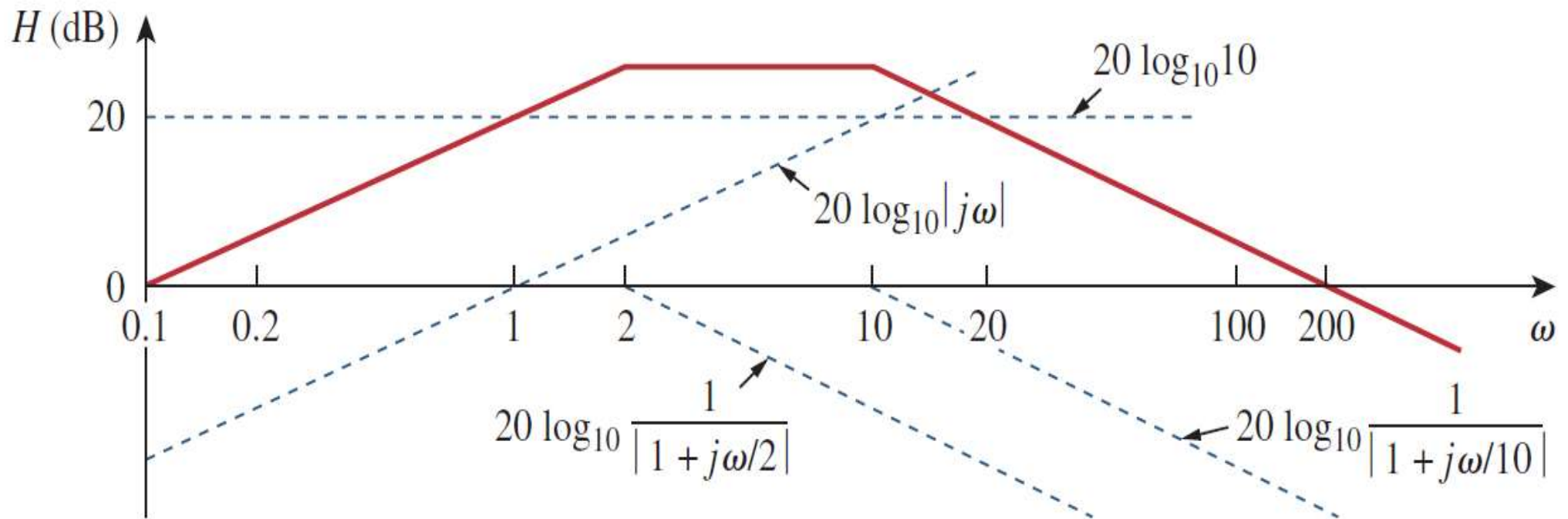
Hence , the magnitude and phase will be,

$$H_{dB} = 20\log_{10}10 + 20\log_{10}|j\omega| - 20\log_{10}\left|1 + \frac{j\omega}{2}\right| - 20\log_{10}\left|1 + \frac{j\omega}{10}\right|$$

$$\phi = 90^\circ - \tan^{-1}\frac{\omega}{2} - \tan^{-1}\frac{\omega}{10}$$

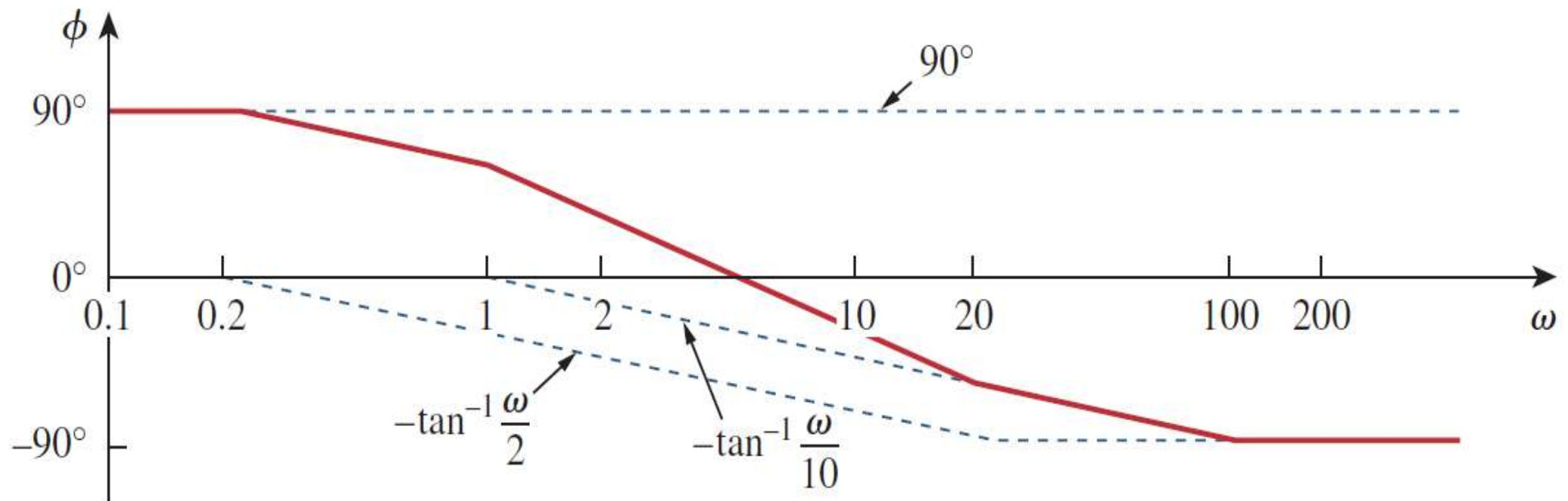
# Bode Plot – Example 1

$$H_{dB} = 20\log_{10}10 + 20\log_{10}|j\omega| - 20\log_{10}\left|1 + \frac{j\omega}{2}\right| - 20\log_{10}\left|1 + \frac{j\omega}{10}\right|$$



# Bode Plot – Example 1

$$\phi = 90^\circ - \tan^{-1} \frac{\omega}{2} - \tan^{-1} \frac{\omega}{10}$$





## Bode Plot – Example 2

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**Question:** Construct the Bode plots for the transfer function?

$$H(\omega) = \frac{j\omega + 10}{j\omega(j\omega + 5)^2}$$

## Bode Plot – Example 2

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**Solution:** Putting  $H(\omega)$  in the **standard form**,

$$H(\omega) = \frac{0.4 \left(1 + j \frac{\omega}{10}\right)}{j\omega \left(1 + j \frac{\omega}{5}\right)^2}$$

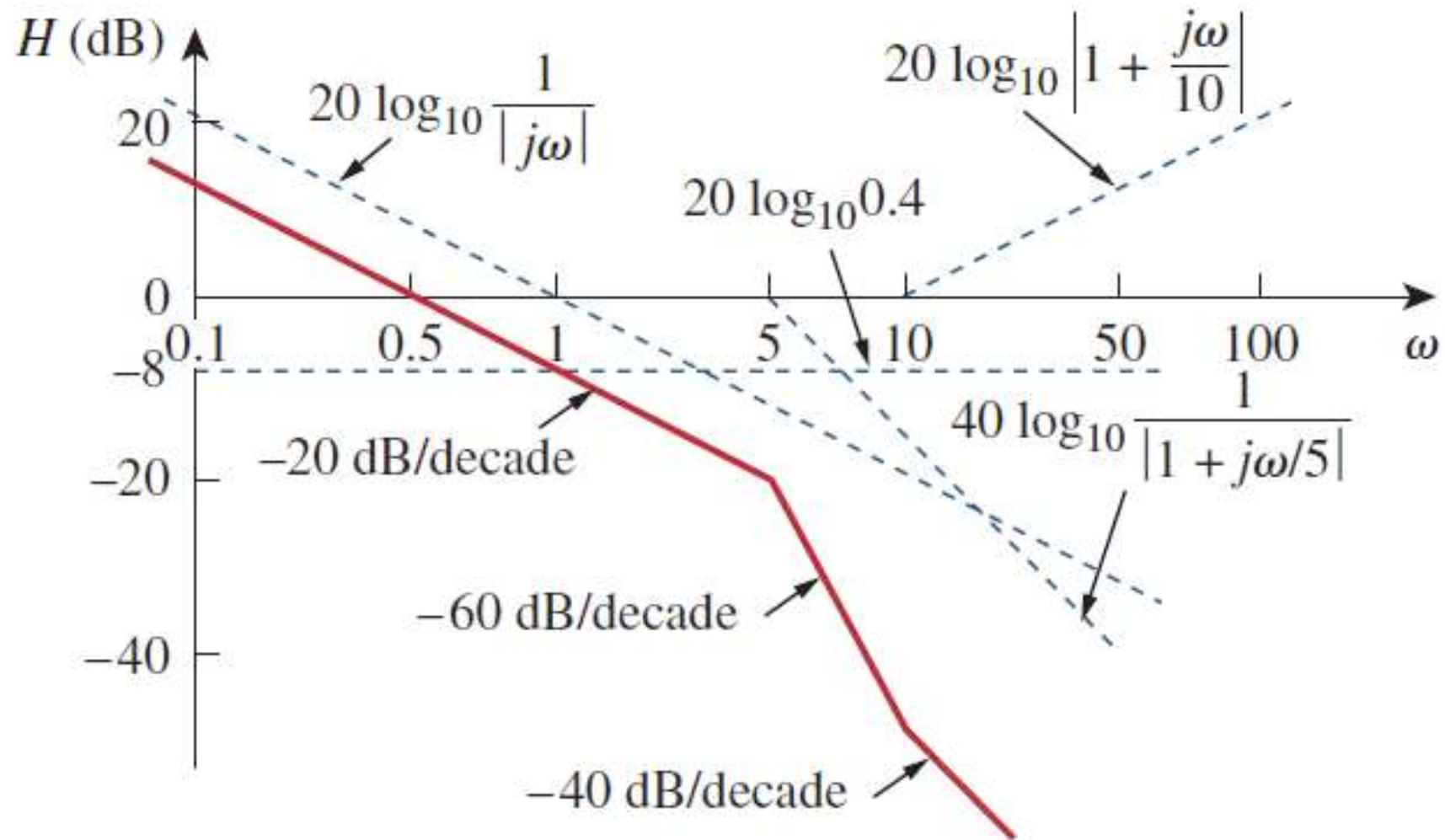
Hence , the magnitude and phase will be,

$$H_{dB} = 20\log_{10}0.4 + 20\log_{10} \left|1 + \frac{j\omega}{10}\right| - 20\log_{10}|j\omega| - 40\log_{10} \left|1 + \frac{j\omega}{5}\right|$$

$$\phi = 0^\circ + \tan^{-1} \frac{\omega}{10} - 90^\circ - 2\tan^{-1} \frac{\omega}{5}$$

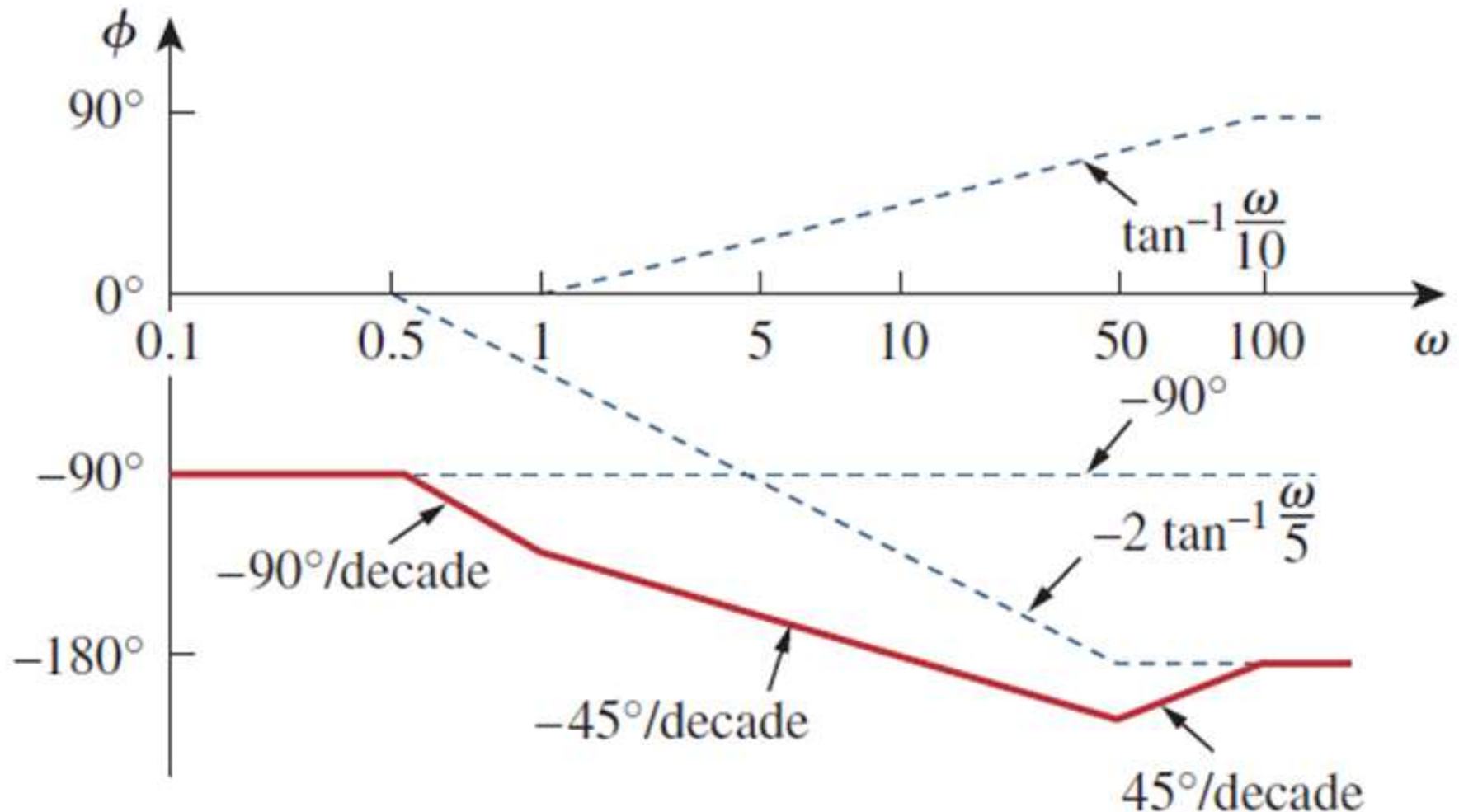
# Bode Plot – Example 2

$$H_{dB} = 20\log_{10}0.4 + 20\log_{10}\left|1 + \frac{j\omega}{10}\right| - 20\log_{10}|j\omega| - 40\log_{10}\left|1 + \frac{j\omega}{5}\right|$$



# Bode Plot – Example 2

$$\phi = 0^\circ + \tan^{-1} \frac{\omega}{10} - 90^\circ - 2 \tan^{-1} \frac{\omega}{5}$$



# Objectives

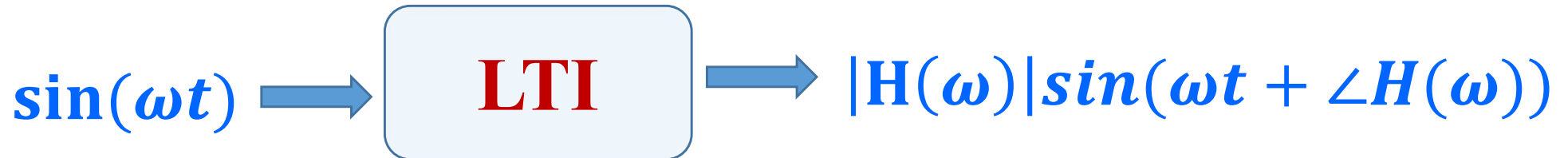
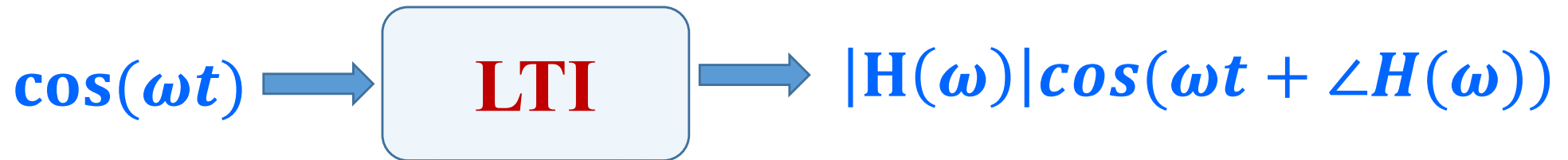
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- Bode plot (continued)
- **LTI system response to co-sinusoids input**
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- Simplification by Symmetrical considerations
- Circuit interpretation for Fourier series

# LTI system response to co – sinusoids

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For the steady – state systems,



and for linearity,



# LTI system response to co – sinusoids

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From this we can conclude that,

$$|F|\cos(\omega t + \theta) \xrightarrow{\quad} \boxed{\text{LTI}} \xrightarrow{\quad} |H(\omega)||F|\cos(\omega t + \theta + \angle H(\omega))$$

and

$$|F|\sin(\omega t + \theta) \xrightarrow{\quad} \boxed{\text{LTI}} \xrightarrow{\quad} |H(\omega)||F|\sin(\omega t + \theta + \angle H(\omega))$$

# LTI system response to co – sinusoids

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It is interesting to note that,

- The LTI systems converts their **co-sinusoidal inputs** of **frequency  $\omega$**  into **co-sinusoidal outputs** having the **same frequency** and following amplitude and phase:
- Output amplitude = Input amplitude *multiplied* by  **$|H(\omega)|$**
- Output phase = Input phase *plus*  **$\angle H(\omega)$**



## Co – sinusoids input: Example 3

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**Question:** Determine the steady-state system responses  $y_1(t)$  and  $y_2(t)$  for given driving functions  $f_1(t)$  and  $f_2(t)$  for transfer function  $H(\omega)$ ?

$$H(\omega) = \frac{1}{1 + j\omega} = \frac{1}{\sqrt{1 + \omega^2}} \angle -\tan^{-1}(\omega)$$

$$f_1(t) = 1 \cos(0.5t) \text{ V}$$

$$f_2(t) = 1 \cos(2t) \text{ V}$$

## Co – sinusoids input: Example 3

**Solution:**

We apply the input-output relation previously shown,

$$y_1(t) = \underbrace{|H(0.5)|}_{\text{Magnitude}} 1 \cos\left(0.5t + \underbrace{\angle H(0.5)}_{\text{Phase}}\right) V$$

**Magnitude**

**Phase**

$$H(0.5) = \frac{1}{\sqrt{1 + 0.5^2}} = 0.894$$

$$\angle H(0.5) = -\tan^{-1}(0.5) = -26.56^\circ$$

$$y_1(t) = 0.894 \cos(0.5t - 26.56^\circ) V$$

## Co – sinusoids input: Example 3

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Similarly, for  $f_1(t)$ ,

$$y_2(t) = \underbrace{|H(2)|}_{\text{Magnitude}} 1 \cos(2t + \underbrace{\angle H(2)}_{\text{Phase}}) V$$

**Magnitude**

**Phase**

$$H(2) = \frac{1}{\sqrt{1 + 2^2}} = 0.447$$

$$\angle H(2) = -\tan^{-1}(2) = -63.46^\circ$$

$$y_2(t) = 0.447 \cos(2t - 63.46^\circ) V$$

## Co – sinusoids input: Example 4

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**Question:** An LTI system has system response as low pass filter denoted as  $H(\omega)$  converts input

$$f(t) = 2\sin(12t)$$

into a steady-state output

$$y(t) = \sqrt{2}\sin(12t + \theta) \text{ for some real valued } \theta$$

Determine  $H(12)$  and also compare the *average power* that the  $f(t)$  and  $y(t)$  would deliver to  $1 \Omega$  resistor?

## Co – sinusoids input: Example 4

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**Solution:** First we write,

$$|H(12)| = \frac{Y}{F} = \frac{\sqrt{2}}{2} = \frac{1}{\sqrt{2}}$$

From the statement of the question ,

$$H(12) = \frac{1}{\sqrt{2}} e^{j\theta}$$

The average power per ohm for input  $f(t)$ ,

$$P_f = \frac{1}{2} |F|^2 = 2$$

While for output,

$$P_y = \frac{1}{2} |Y|^2 = 1$$

## Co – sinusoids input: Example 4

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Comparing both input and output ,

$$\frac{P_y}{P_f} = \frac{1}{2}$$

which makes  $\omega = 12 \frac{rad}{sec}$ , the half power frequency of the filter  $H(\omega)$

## Co – sinusoids input: Example 5

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**Question:** What is the steady-state response of the system

$$H(\omega) = \frac{2 + j\omega}{4 + j\omega} \quad \text{to a DC input } f(t) = 5?$$

**Solution:** Since,

$$H(0) = \frac{2 + j0}{4 + j0} = 0.5$$

The steady-state response will be,

$$y(t) = H(0)5 = 2.5$$

# Objectives

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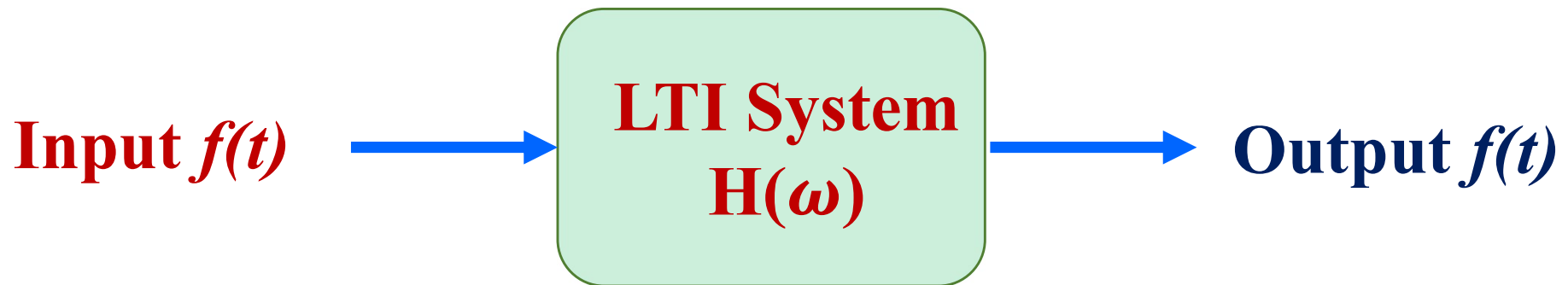
- Bode plot (continued)
- LTI system response to co-sinusoids input
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# LTI Sys. response: Multi-frequency Input

We will apply following methods to evaluate LTI response to multifrequency input

- Principle of Superposition
- Input-output relation for co-sinusoids



$$f(t) = \sum_{n=1}^N |F_n| \cos(\omega_n t + \theta_n)$$

$$y(t) = \sum_{n=1}^N |H(\omega_n)| |F_n| \cos(\omega_n t + \theta_n + \chi(\omega_n))$$

# Multi-frequency Input – Example 6

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**Question:** The input of a low pass filter

$$H(\omega) = \frac{1}{1 + j\omega}$$

is  $f(t) = 1\cos(0.5t) + 1\cos(\pi t)$

**Determine the system output  $y(t)$ ?**

# Multi-frequency Input – Example 6

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**Solution:** According to relation shown for multifrequency,

$$y(t) = y_1(t) + y_2(t)$$

$$y(t) = |H(0.5)|1 \cos(0.5t + \angle H(0.5)) + |H(\pi)|1 \cos(\pi t + \angle H(\pi))$$

For  $y_1(t)$

$$H(0.5) = \frac{1}{\sqrt{1 + 0.5^2}} = 0.894$$

$$\angle H(0.5) = -\tan^{-1}(0.5) = -26.56^\circ$$

$$y_1(t) = 0.894 \cos(0.5t - 26.56^\circ) \text{ V}$$

# Multi-frequency Input – Example 6

---

For  $y_2(t)$

$$H(\pi) = \frac{1}{\sqrt{1 + \pi^2}} = 0.303$$

$$\angle H(\pi) = -\tan^{-1}(\pi) = -72.34^\circ$$

$$y_2(t) = 0.303 \cos(\pi t - 72.34^\circ) \text{ V}$$

$$y(t) = y_1(t) + y_2(t)$$

$$y(t) = 0.894 \cos(0.5t - 26.56^\circ) + 0.303 \cos(\pi t - 72.34^\circ) \text{ V}$$

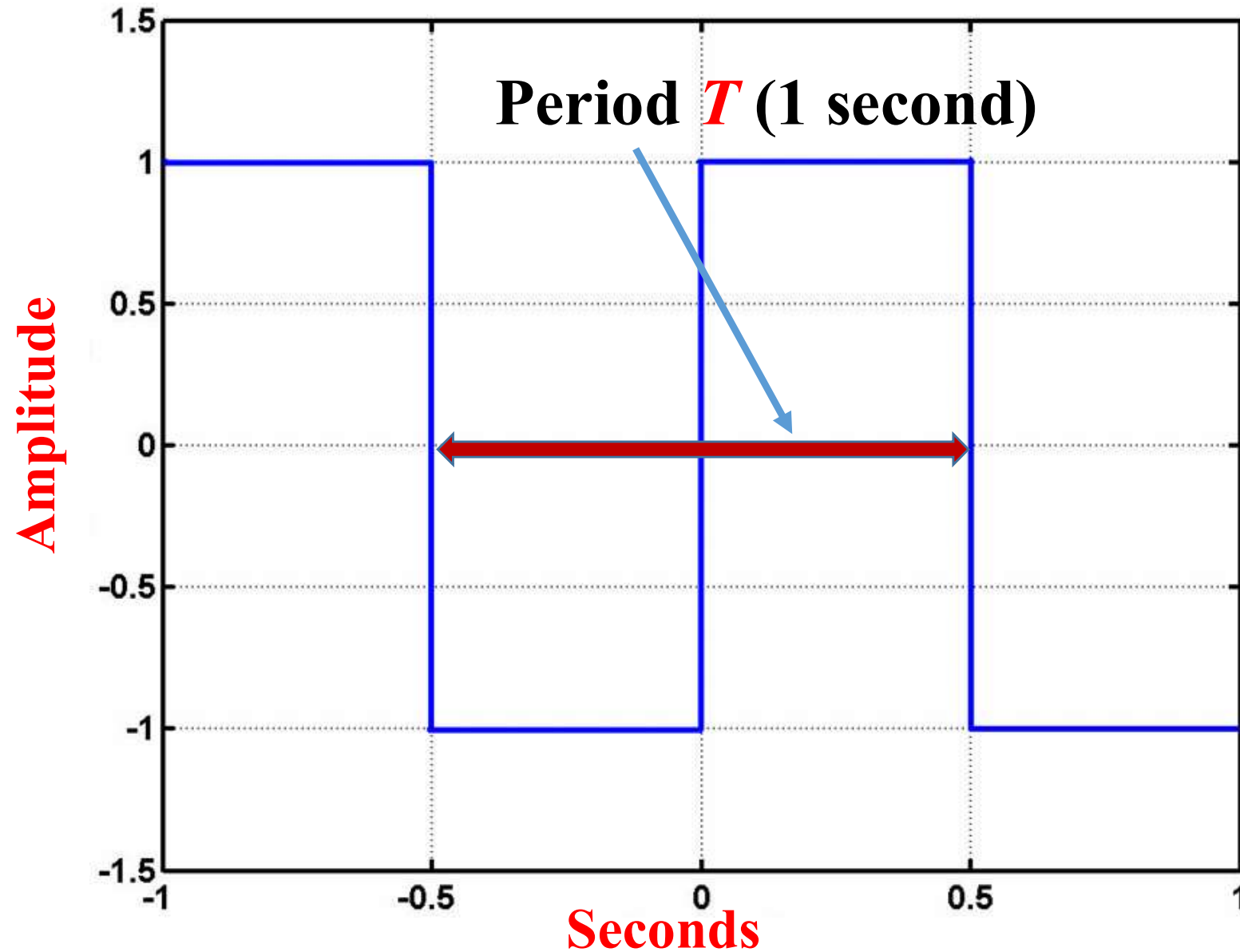
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# Fourier Series and coefficients

Periodic Signal  $f(t) = f(t + nT)$



# Fourier Series and coefficients

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Fourier stated that :

- A non-sinusoidal **periodic function** can be expressed as an ***infinite sum*** of sinusoidal functions

Periodic Signal:  $f(t) = f(t + nT)$

Where ***n*** is any integer and ***T*** is the period of signal

- Any practical periodic function of frequency  **$\omega_o$**  can be expressed as an ***infinite sum of sine or cosine functions*** that are integral multiples of  **$\omega_o$**

# Fourier Series and coefficients

Periodic Signal is expressed as per *Fourier's theorem* as,

$$f(t) = a_0 + a_1 \cos \omega_0 t + b_1 \sin \omega_0 t + a_2 \cos 2\omega_0 t + b_2 \sin 2\omega_0 t + a_3 \cos 3\omega_0 t + b_3 \sin 3\omega_0 t + \dots$$

which can be reduced to,

$$f(t) = \underbrace{a_0}_{\text{dc}} + \underbrace{\sum_{n=1}^{\infty} (a_n \cos n\omega_0 t + b_n \sin n\omega_0 t)}_{\text{ac}}$$

$\omega_o = \frac{2\pi}{T}$  is the fundamental frequency (rad/s)



# Fourier Series and coefficients

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## Decomposition

- $\cos n\omega_o t$  and  $\sin n\omega_o t$  are  $n$ -th harmonics of  $f(t)$
- Even harmonic for even  $n$  and odd for odd  $n$
- $a_n$  and  $b_n$  are called *Fourier coefficients*
- $a_o$  is the dc component of average value of  $f(t)$
- $a_n$  and  $b_n$  are the amplitudes of the sinusoids in the ac component

# Fourier Series and coefficients

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- It resolves  $f(t)$  into a **dc component** and **an ac component** comprising **an infinite series of harmonic sinusoids**

**Dirichlet conditions** to satisfy Fourier series

1.  $f(t)$  should be single valued everywhere
2.  $f(t)$  has a finite number of finite discontinuities in any one period
3.  $f(t)$  has a finite number of maxima and minima in any one period
4. The integral  $\int_{t_o}^{t_o+T} |f(t)| dt < \infty$  for any  $t_o$

# Fourier Series and coefficients

---

**DC component  
(constant)**

$$a_0 = \frac{1}{T} \int_0^T f(t) dt$$

**AC component  
Harmonics**

$$a_n = \frac{2}{T} \int_0^T f(t) \cos n\omega_0 t dt$$

$$b_n = \frac{2}{T} \int_0^T f(t) \sin n\omega_0 t dt$$

# Fourier Series and coefficients

Another form of expressing Fourier series is in terms of amplitude and phase components,

$$f(t) = a_0 + \sum_{n=1}^{\infty} A_n \cos(n\omega_0 t + \phi_n)$$

$$A_n = \sqrt{a_n^2 + b_n^2}$$

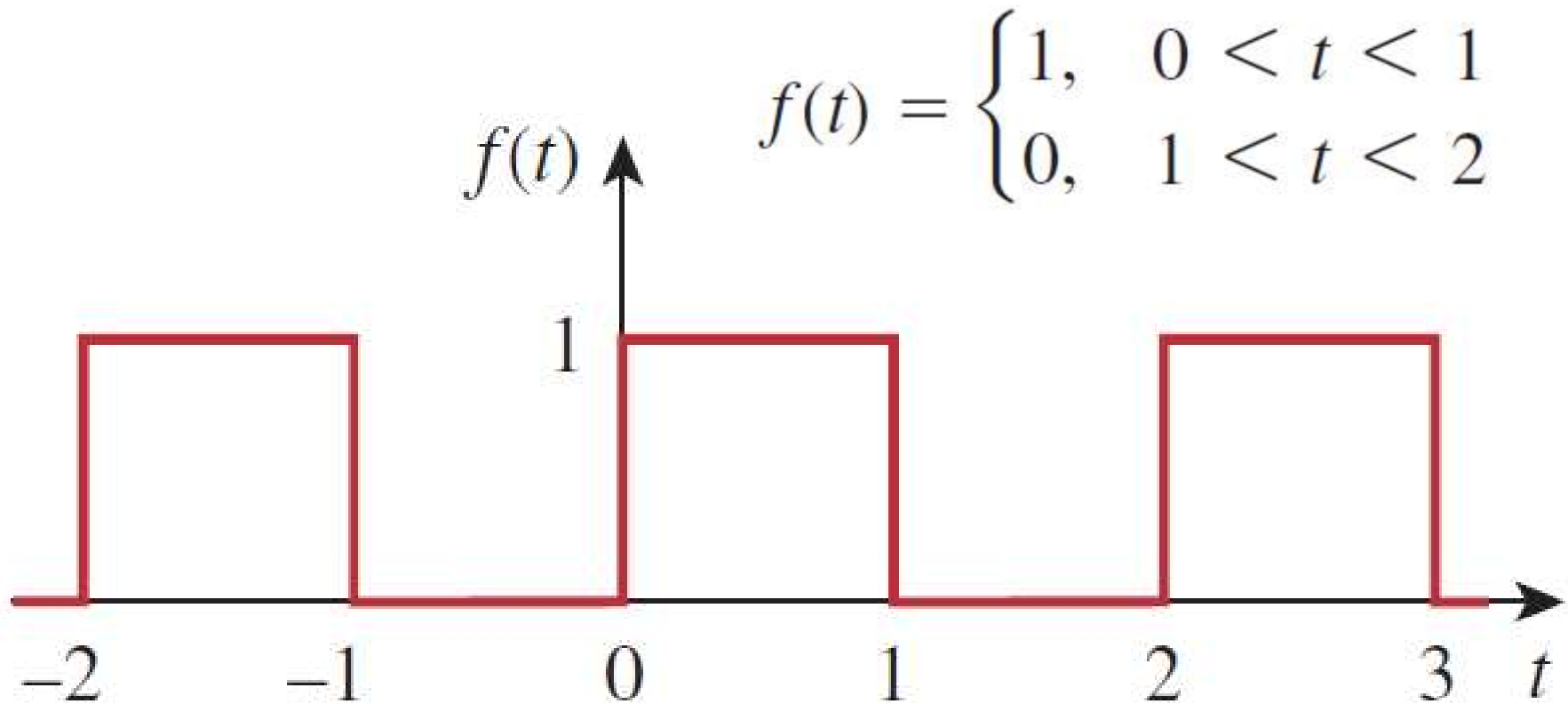
$$\phi_n = -\tan^{-1} \frac{b_n}{a_n}$$

$$\text{Or, } A_n \angle \phi_n = a_n - jb_n$$

- The **frequency spectrum** of a signal consists of the plots of the amplitudes and phases of the harmonics versus frequency

# Fourier Series – Example 7

**Question:** Determine the Fourier series of the waveform shown . Obtain the amplitude and phase spectra?



# Fourier Series – Example 7

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**Solution:** From the waveform,

$$\omega_o = \frac{2\pi}{T} = \frac{2\pi}{2} = \pi \text{ as } T = 2 \text{ s}$$

$$a_0 = \frac{1}{T} \int_0^T f(t) dt = \frac{1}{2} \left[ \int_0^1 1 dt + \int_1^2 0 dt \right] = \frac{1}{2} t \Big|_0^1 = \frac{1}{2}$$

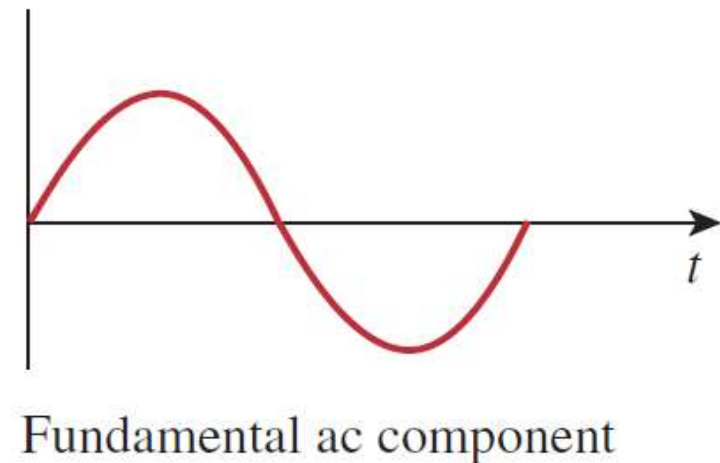
$$a_n = \frac{2}{T} \int_0^T f(t) \cos n\omega_o t dt = 0$$

$$b_n = \frac{2}{T} \int_0^T f(t) \sin n\omega_o t dt = \begin{cases} \frac{2}{n\pi}, & n = \text{odd} \\ 0, & n = \text{even} \end{cases}$$

# Fourier Series – Example 7

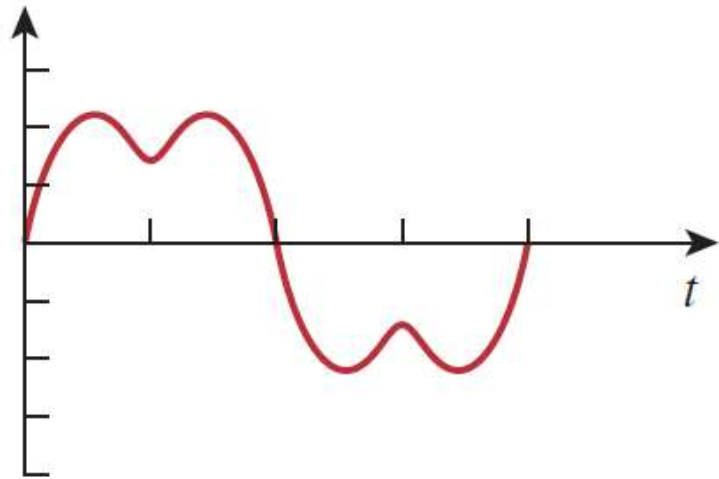
$$f(t) = \frac{1}{2} + \frac{2}{\pi} \sin \pi t + \frac{2}{3\pi} \sin 3\pi t + \frac{2}{5\pi} \sin 5\pi t + \dots$$

$$f(t) = \frac{1}{2} + \frac{2}{\pi} \sum_{k=1}^{\infty} \frac{1}{n} \sin n\pi t, \quad n = 2k - 1$$

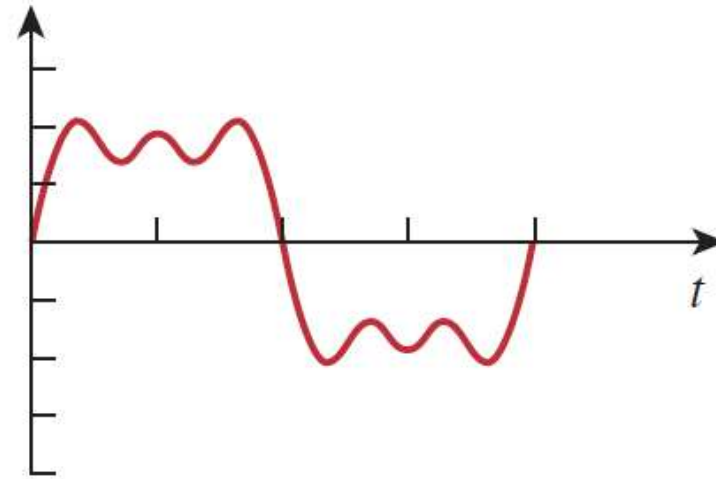


# Fourier Series – Example 7

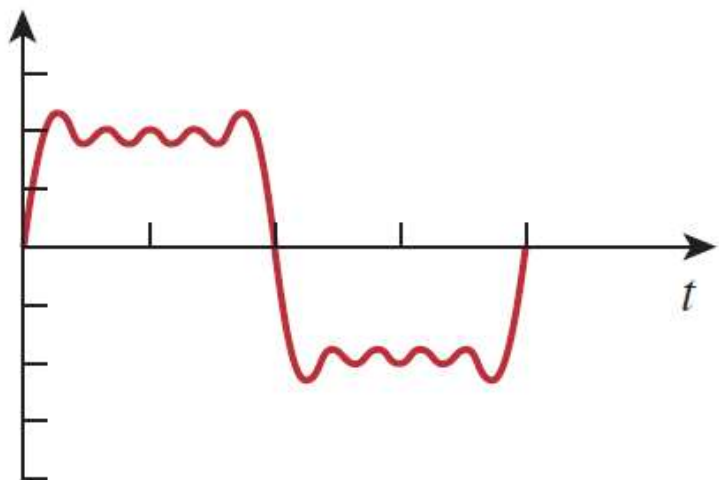
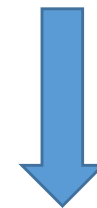
ac components harmonics



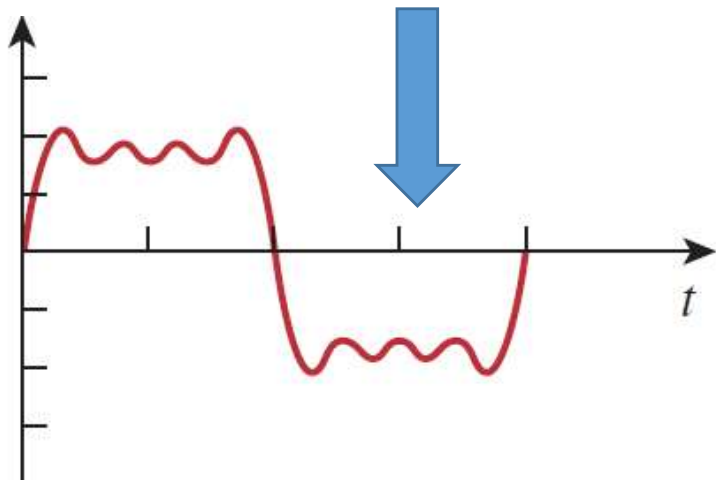
Sum of first two ac components



Sum of first three ac components



Sum of first five ac components



Sum of first four ac components



# Fourier Series – Example 7

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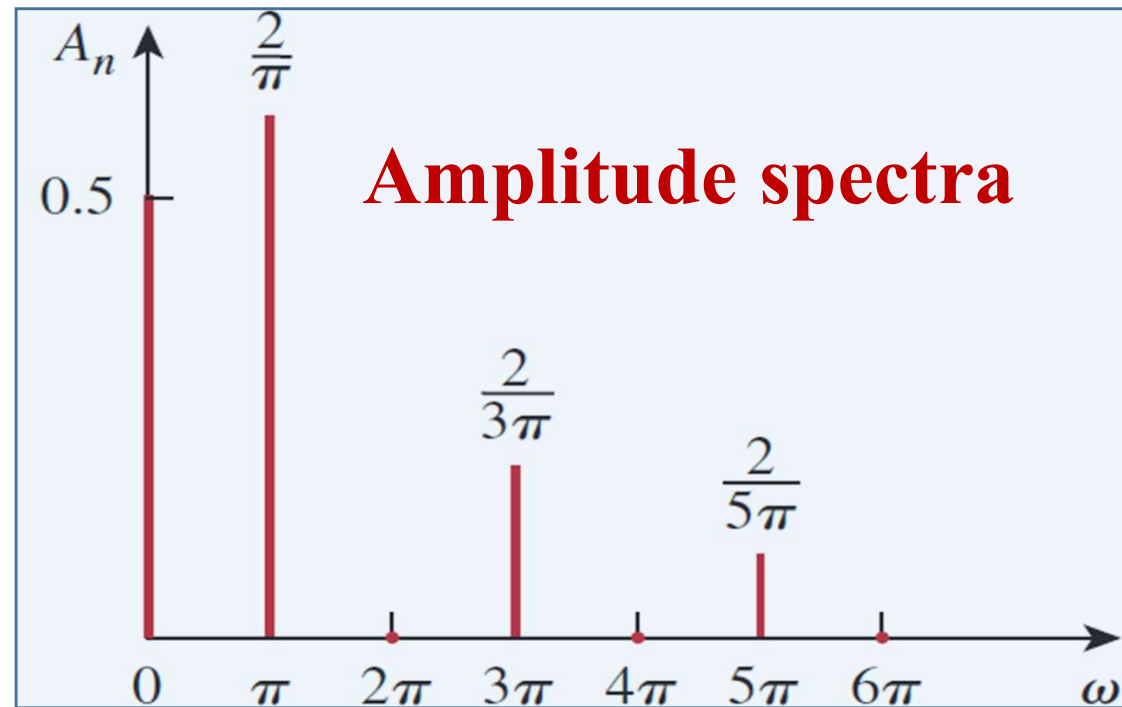
## Another approach

### Amplitude and phase spectra for the signal

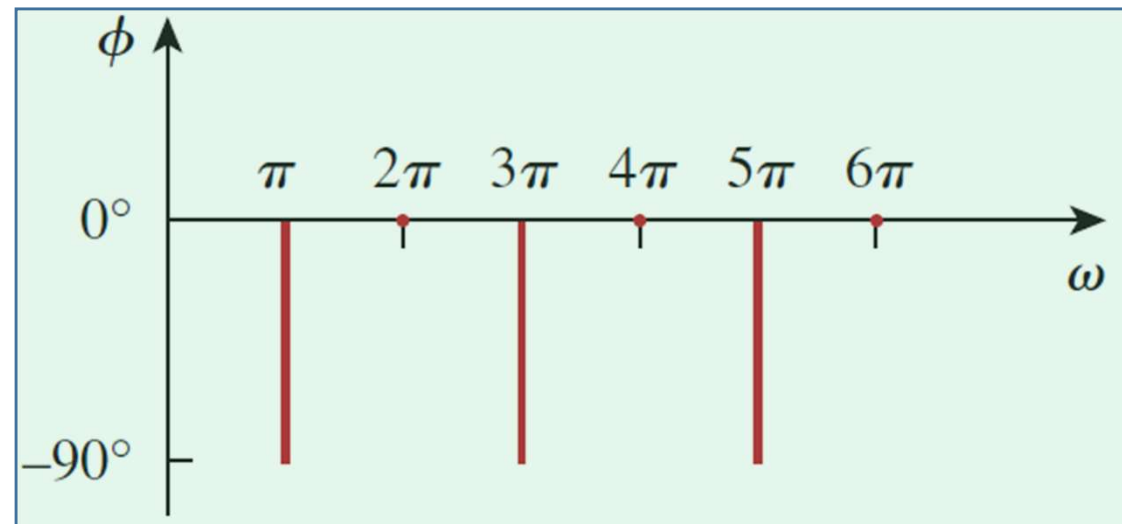
$$A_n = \sqrt{a_n^2 + b_n^2} = |b_n| = \begin{cases} \frac{2}{n\pi}, & n = \text{odd} \\ 0, & n = \text{even} \end{cases}$$

$$\phi_n = -\tan^{-1} \frac{b_n}{a_n} = \begin{cases} -90^\circ, & n = \text{odd} \\ 0, & n = \text{even} \end{cases}$$

# Fourier Series – Example 7



**Phase spectra**



# Objectives

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- Bode plot (continued)
- LTI system response to co-sinusoids input
- LTI system response to multifrequency co-sinusoids input
- Fourier coefficients of periodic signals
- **Simplification by Symmetrical considerations**
- Circuit interpretation for Fourier series

# Symmetry considerations

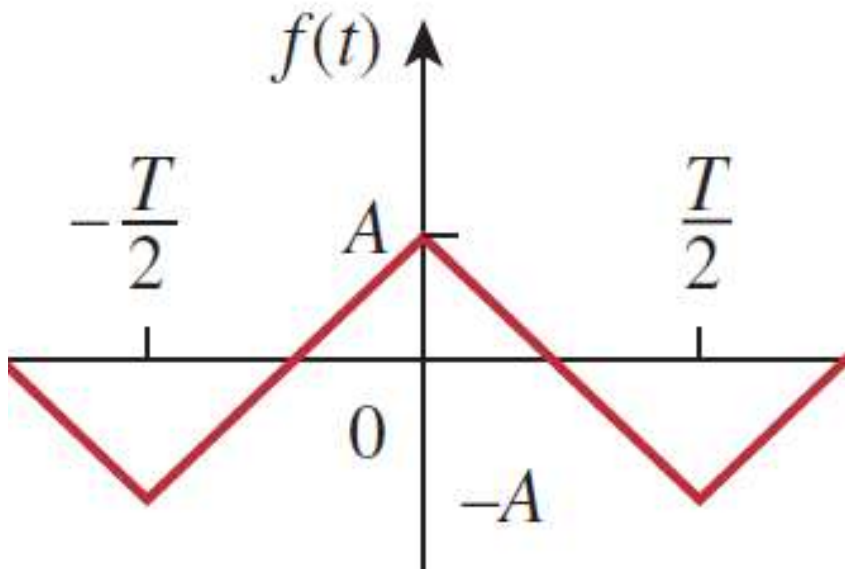
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- To avoid tedious calculations of integrals, symmetry is considered as it exists for sinusoidal and co – sinusoidal
  - Even Symmetry
  - Odd Symmetry
  - Half-wave symmetry

# Even Symmetry

- A function  $f(t)$  is *even* if its plot is symmetrical about the vertical axis; that is,

$$f(t) = f(-t)$$



$$a_0 = \frac{2}{T} \int_0^{T/2} f(t) dt$$

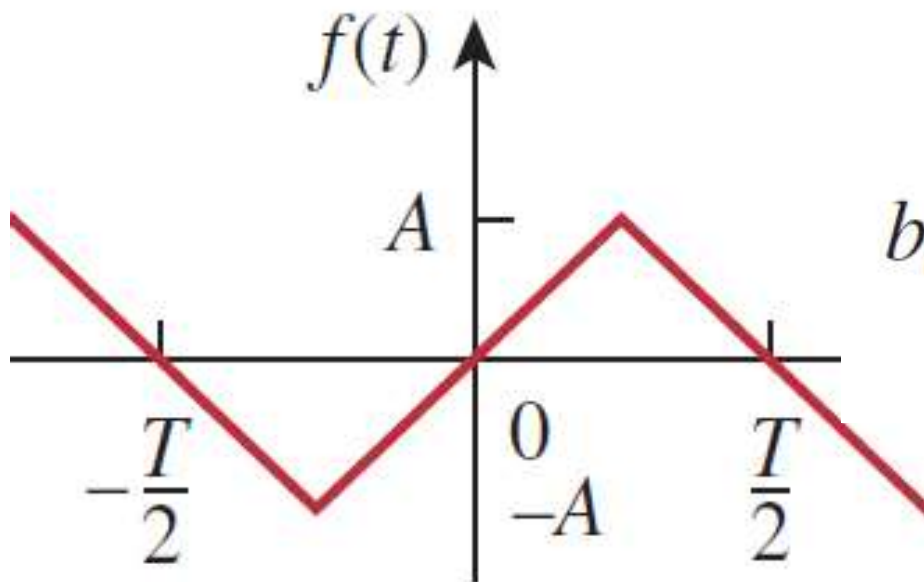
$$a_n = \frac{4}{T} \int_0^{T/2} f(t) \cos n\omega_0 t dt$$

$$b_n = 0$$

# Odd Symmetry

- A function  $f(t)$  is *odd* if its plot is asymmetrical about the vertical axis; that is,

$$f(-t) = -f(t)$$



$$a_0 = 0, \quad a_n = 0$$
$$b_n = \frac{4}{T} \int_0^{T/2} f(t) \sin n\omega_0 t \, dt$$

# Half – wave Symmetry

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**A function is half-wave (odd) symmetric if**

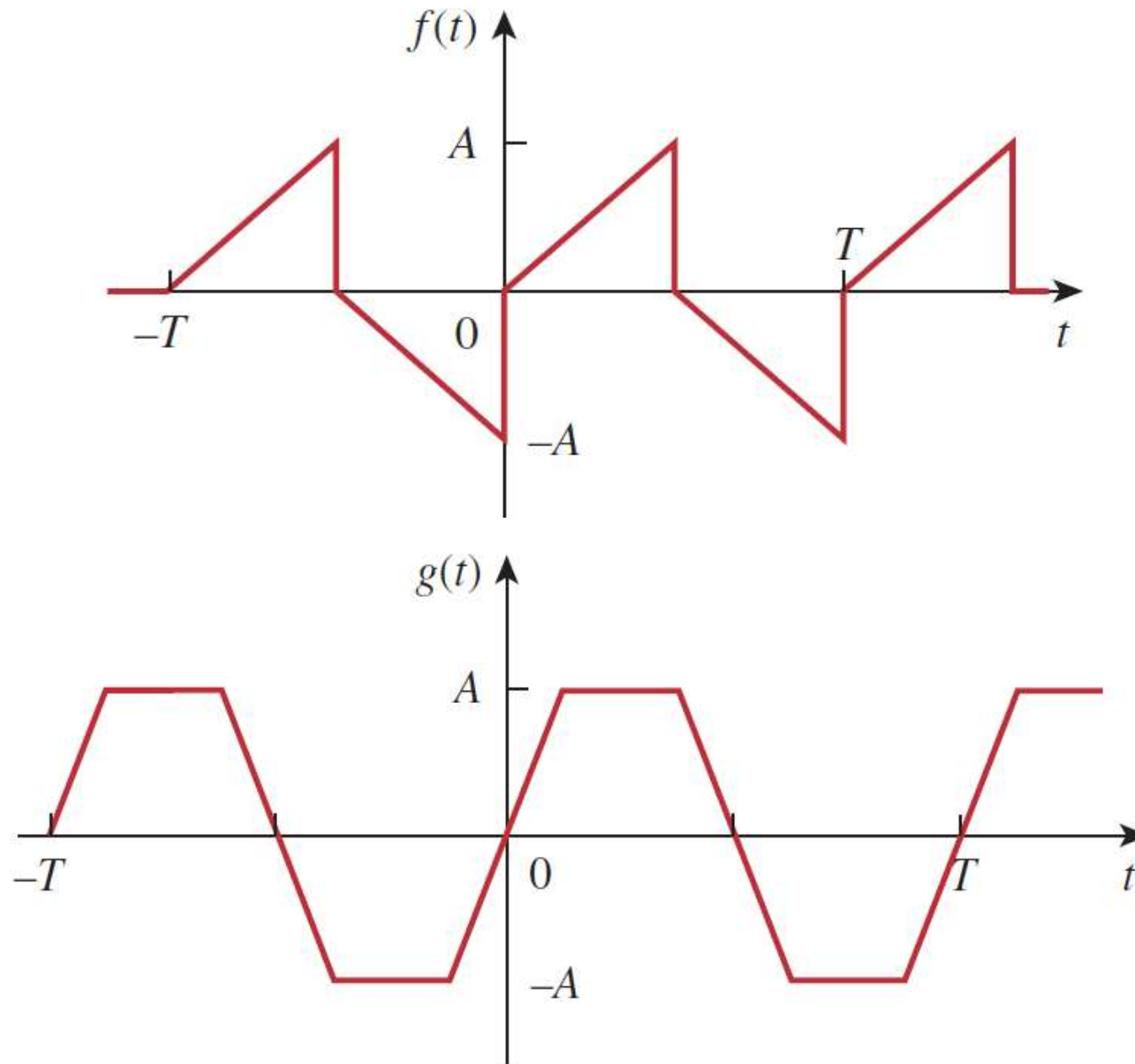
$$f\left(t - \frac{T}{2}\right) = -f(t)$$

$$a_0 = 0$$

$$a_n = \begin{cases} \frac{4}{T} \int_0^{T/2} f(t) \cos n\omega_0 t \, dt, & \text{for } n \text{ odd} \\ 0, & \text{for } n \text{ even} \end{cases}$$

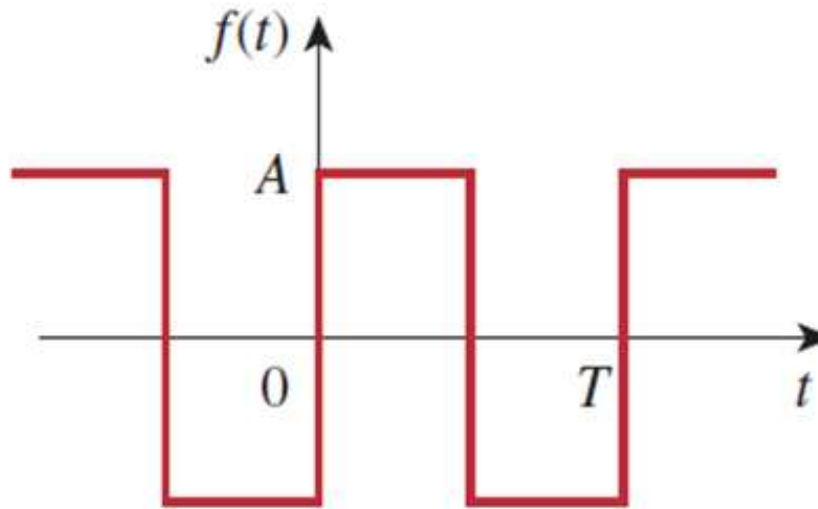
$$b_n = \begin{cases} \frac{4}{T} \int_0^{T/2} f(t) \sin n\omega_0 t \, dt, & \text{for } n \text{ odd} \\ 0, & \text{for } n \text{ even} \end{cases}$$

# Half – wave Symmetry

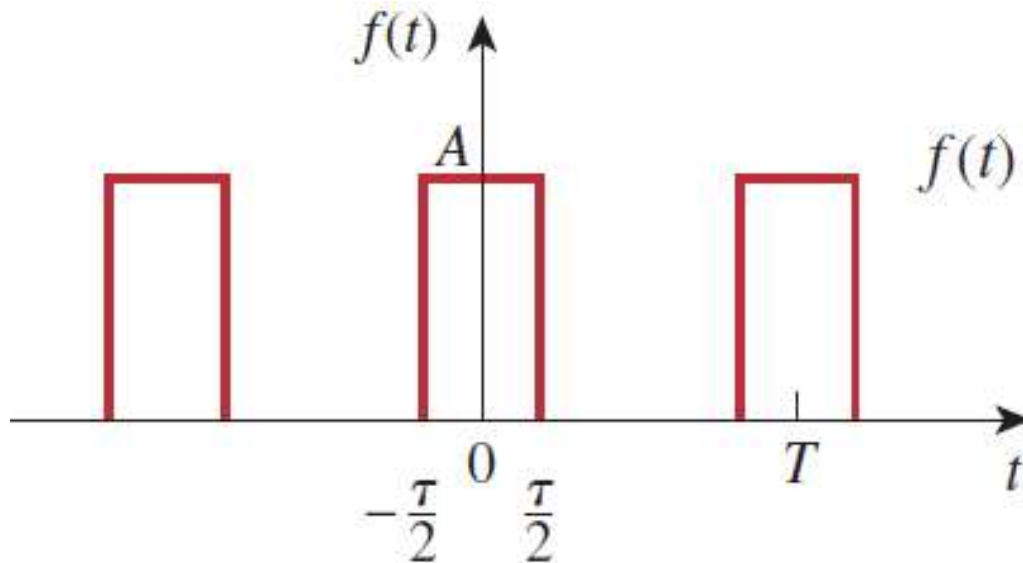




# Fourier series of typical signals

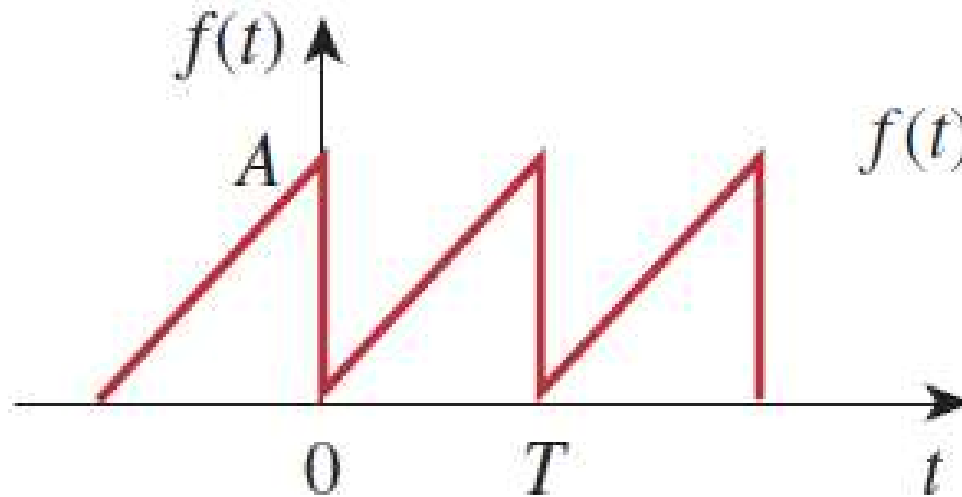


$$f(t) = \frac{4A}{\pi} \sum_{n=1}^{\infty} \frac{1}{2n-1} \sin(2n-1)\omega_0 t$$

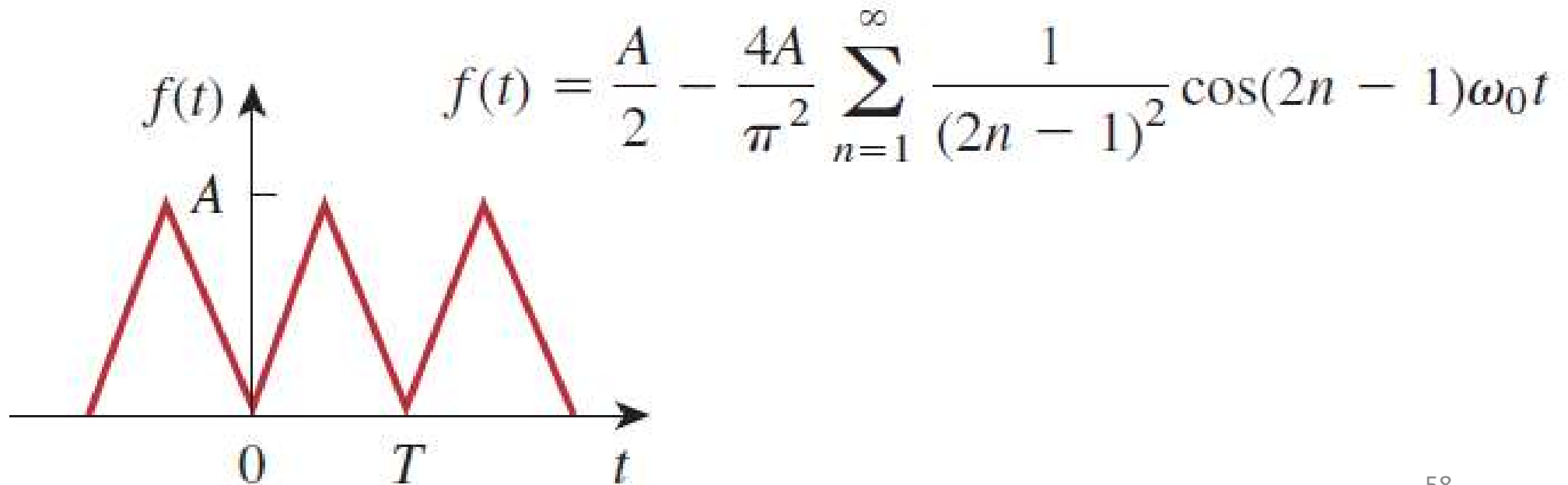


$$f(t) = \frac{A\tau}{T} + \frac{2A}{T} \sum_{n=1}^{\infty} \frac{1}{n} \sin \frac{n\pi\tau}{T} \cos n\omega_0 t$$

# Fourier series of typical signals

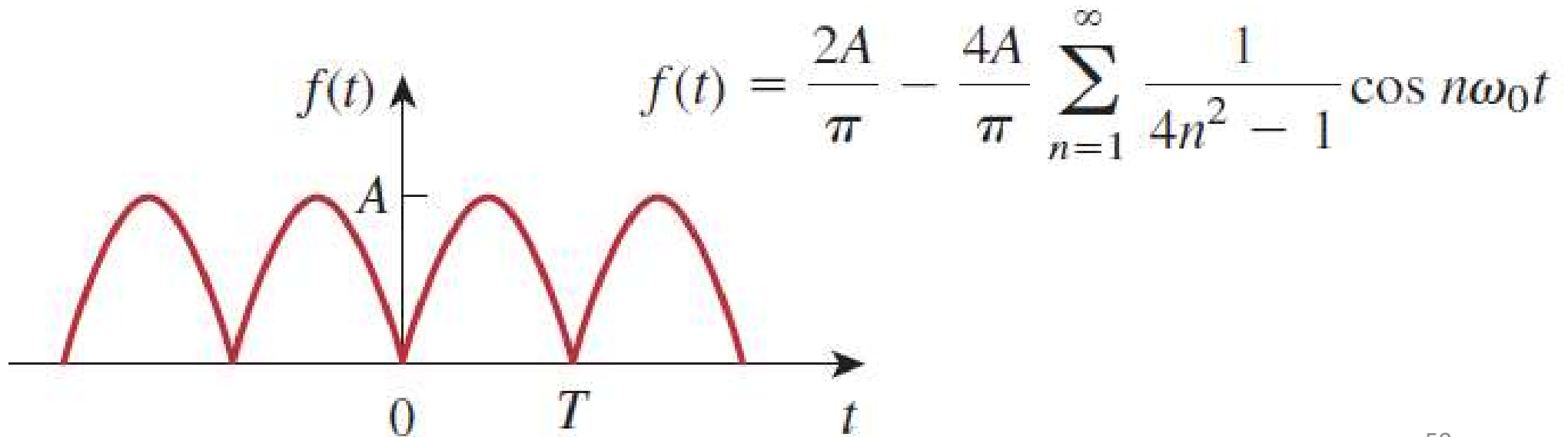
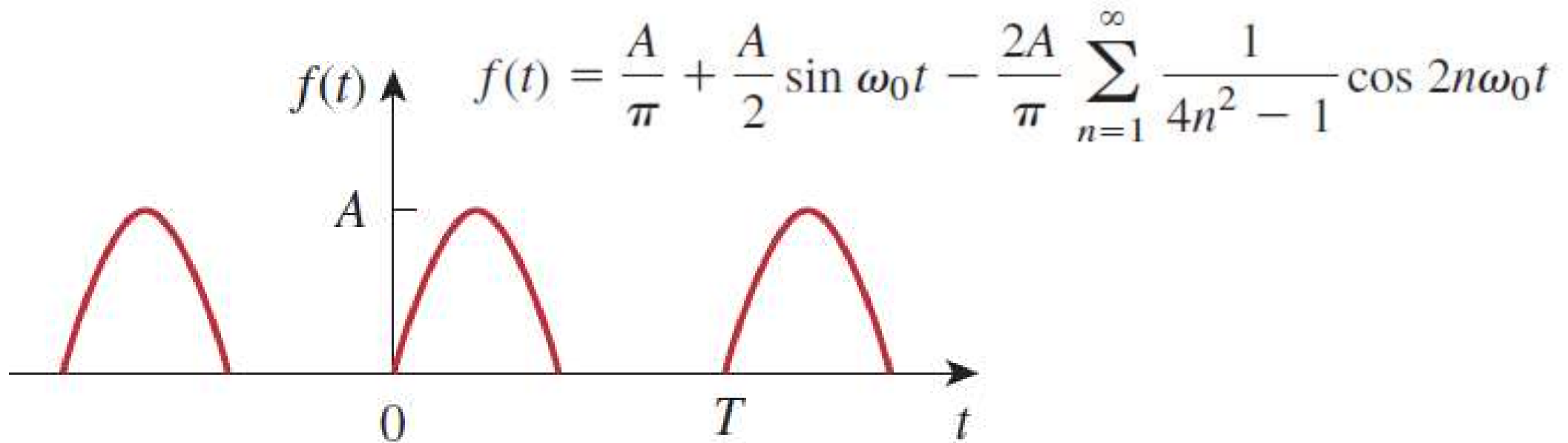


$$f(t) = \frac{A}{2} - \frac{A}{\pi} \sum_{n=1}^{\infty} \frac{\sin n\omega_0 t}{n}$$



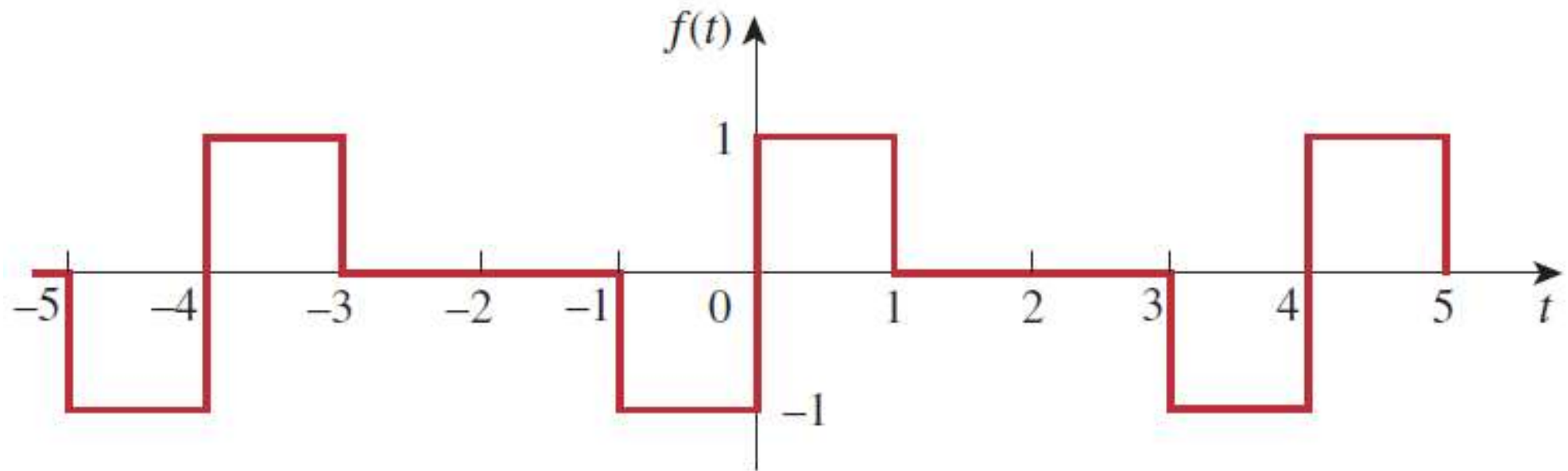
$$f(t) = \frac{A}{2} - \frac{4A}{\pi^2} \sum_{n=1}^{\infty} \frac{1}{(2n-1)^2} \cos(2n-1)\omega_0 t$$

# Fourier series of typical signals



# Symmetrical Signals – Example 8

**Question:** Find Fourier series expansion for the given  $f(t)$ ?



# Symmetrical Signals – Example 8

**Solution:** The function  $f(t)$  is an odd function, So,

$$a_o = a_n = 0$$

The period is  $T = 4$ , hence,  $\omega_o = \pi/2$  so that,

$$\begin{aligned} b_n &= \frac{4}{T} \int_0^{T/2} f(t) \sin n\omega_o t \, dt \\ &= -\frac{2}{n\pi} \cos \frac{n\pi t}{2} \Big|_0^1 = \frac{2}{n\pi} \left( 1 - \cos \frac{n\pi}{2} \right) \end{aligned}$$

$$f(t) = \frac{2}{\pi} \sum_{n=1}^{\infty} \frac{1}{n} \left( 1 - \cos \frac{n\pi}{2} \right) \sin \frac{n\pi}{2} t$$

# Objectives

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- Bode plot (continued)
- LTI system response to co-sinusoids input
- LTI system response to multifrequency co-sinusoids input
- Fourier coefficients of periodic signals
- Simplification by Symmetrical considerations
- **Circuit interpretation for Fourier series**

# Circuit application of Fourier series

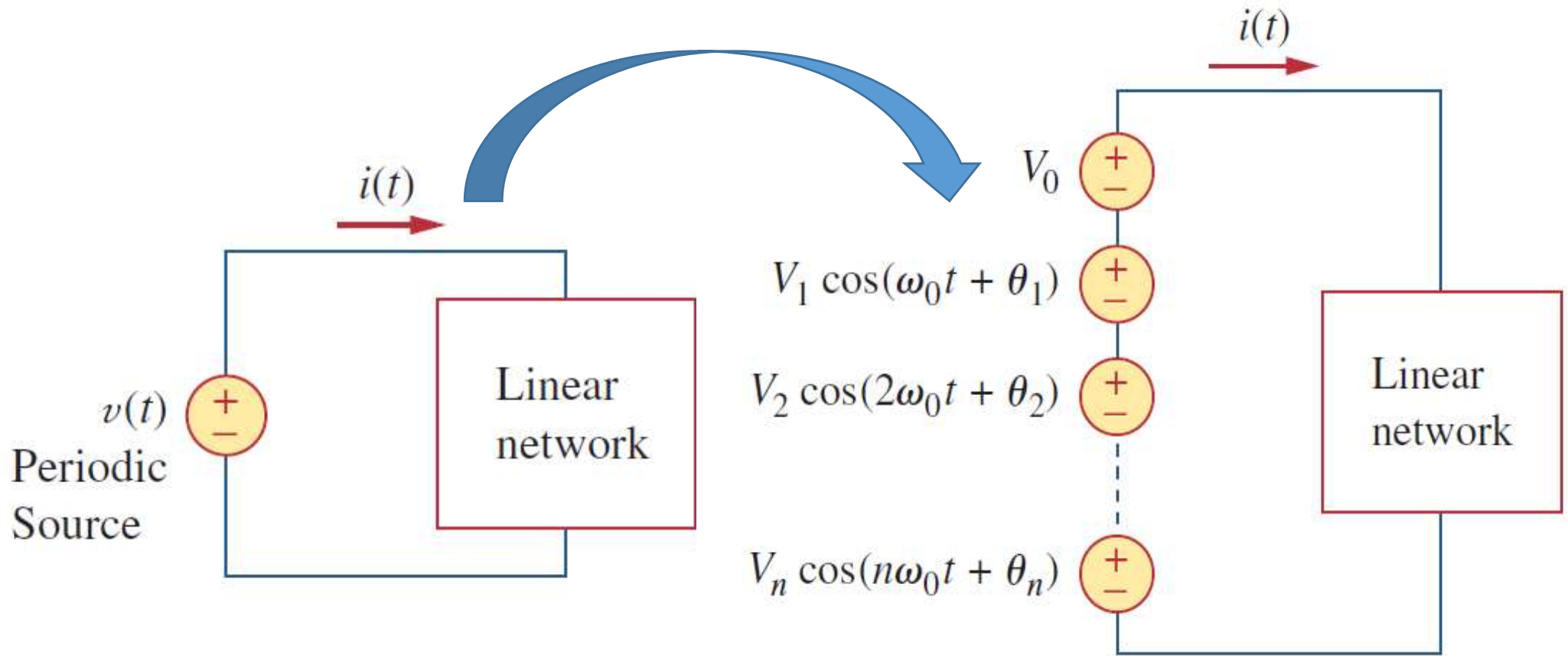
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## Steps for Applying Fourier Series

- 1. Express the excitation as a Fourier series**
- 2. Transform the circuit from the time domain to the frequency domain**
- 3. Find the response of the dc and ac components in the Fourier series**
- 4. Add the individual dc and ac responses using the superposition principle**

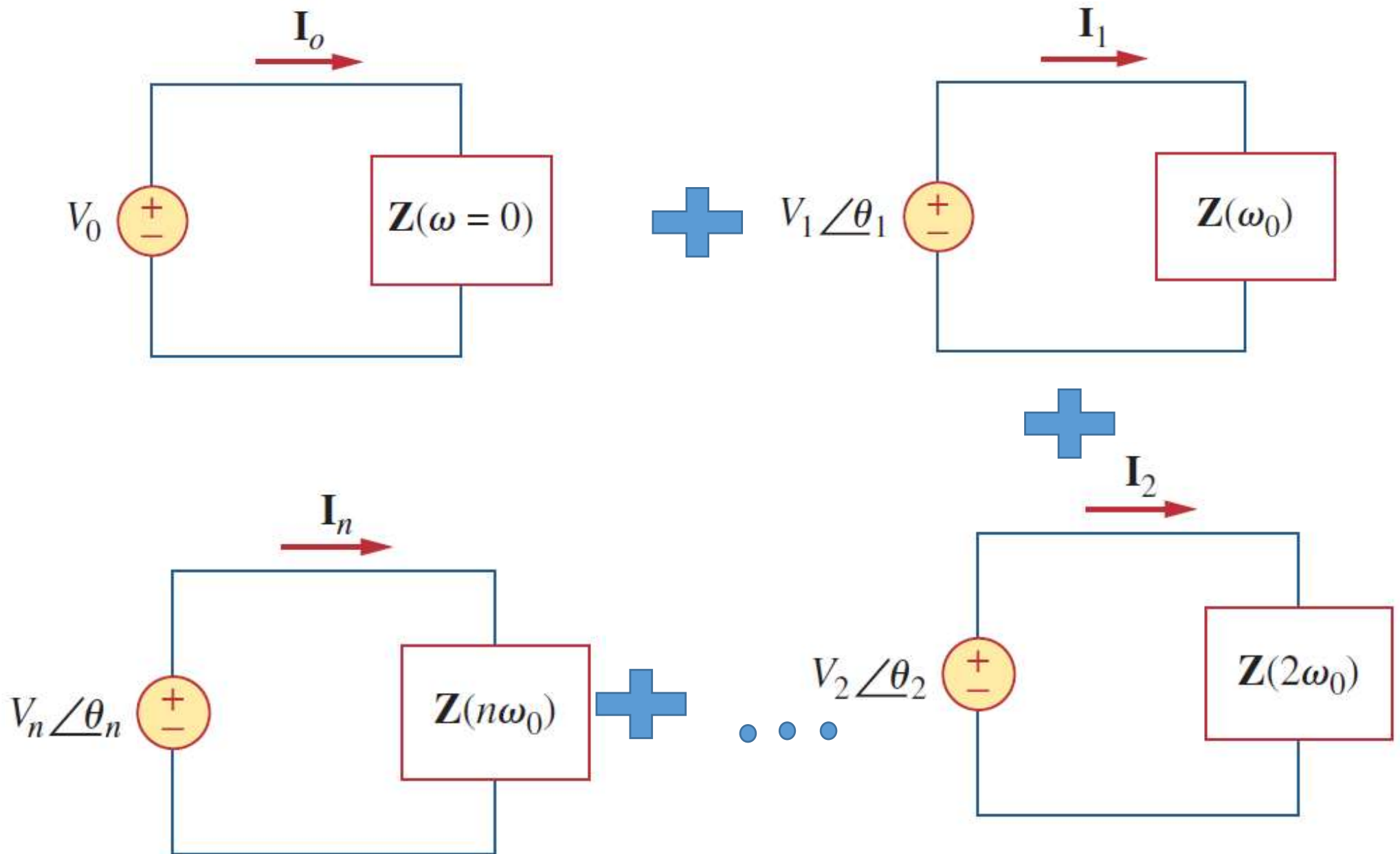
# Circuit application of Fourier series

## Superposition of all phasors of periodic sources





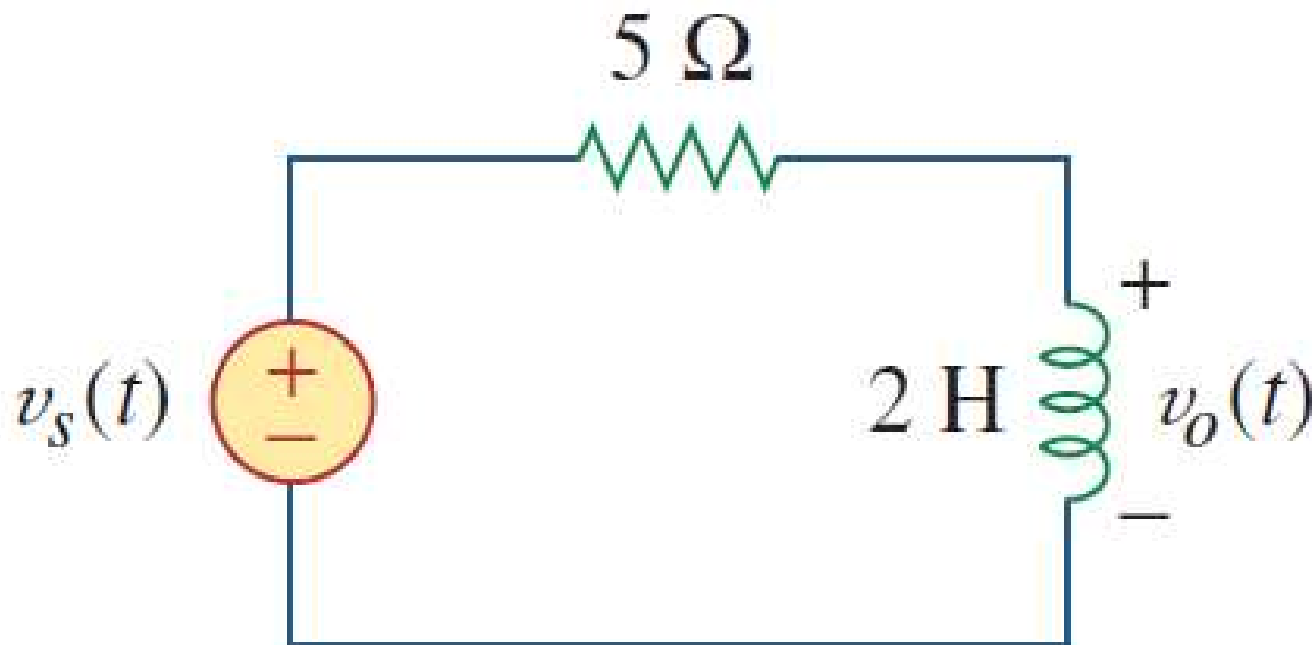
# Circuit application of Fourier series



# Circuit application – Example 9

**Question:** Find  $v_o(t)$  in the circuit by using value of  $v_s(t)$  given below?

$$v_s(t) = \frac{1}{2} + \frac{2}{\pi} \sum_{k=1}^{\infty} \frac{1}{n} \sin n\pi t, \quad n = 2k - 1$$



# Circuit application – Example 9

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**Solution:** By using voltage division rule,

$$\mathbf{V}_o = \frac{j\omega_n L}{R + j\omega_n L} \mathbf{V}_s = \frac{j2n\pi}{5 + j2n\pi} \mathbf{V}_s$$

For DC component,  $\omega_o = 0$  or  $n = 0$ ,

$$V_s = \frac{1}{2}, \quad \text{so,} \quad V_o = 0$$

This is expected, since the inductor is a short circuit to dc. For the  $n$ -th harmonic,

$$\mathbf{V}_s = \frac{2}{n\pi} \angle -90^\circ$$

## Circuit application – Example 9

and the corresponding response is,

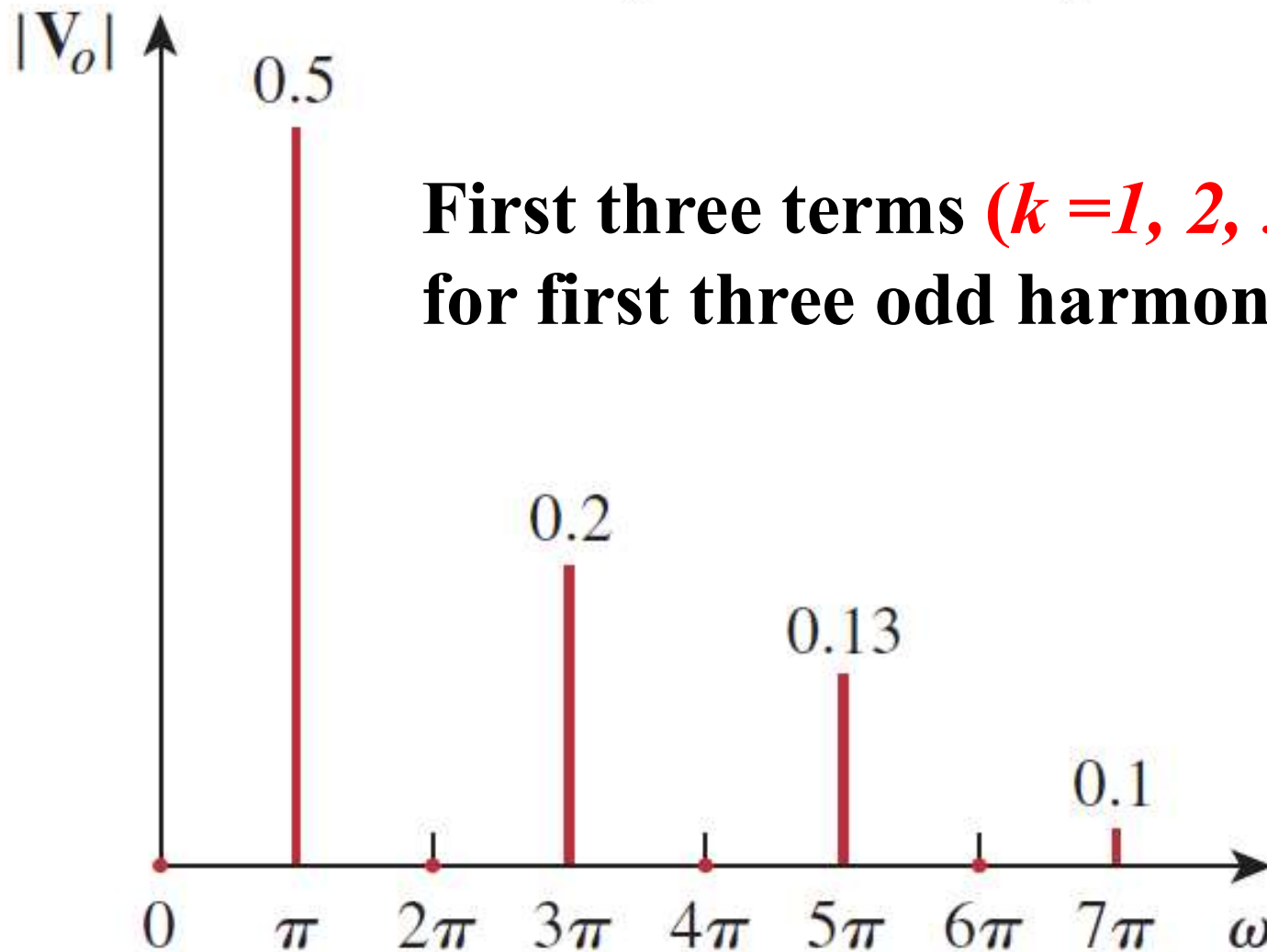
$$\begin{aligned} \mathbf{V}_o &= \frac{2n\pi \angle 90^\circ}{\sqrt{25 + 4n^2\pi^2} \angle \tan^{-1} 2n\pi/5} \left( \frac{2}{n\pi} \angle -90^\circ \right) \\ &= \frac{4 \angle -\tan^{-1} 2n\pi/5}{\sqrt{25 + 4n^2\pi^2}} \end{aligned}$$

converting into time domain,

$$v_o(t) = \sum_{k=1}^{\infty} \frac{4}{\sqrt{25 + 4n^2\pi^2}} \cos\left(n\pi t - \tan^{-1} \frac{2n\pi}{5}\right),$$
$$n = 2k - 1$$

# Circuit application – Example 9

$$v_o(t) = 0.4981 \cos(\pi t - 51.49^\circ) + 0.2051 \cos(3\pi t - 75.14^\circ) + 0.1257 \cos(5\pi t - 80.96^\circ) + \dots \text{ V}$$



**First three terms ( $k=1, 2, 3$  or  $n=1, 3, 5$ )  
for first three odd harmonics**

# System response – Example 10

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**Question:** The input of a linear system,

$$H(\omega) = \frac{2 + j\omega}{3 + j\omega}$$

is the periodic function,

$$f(t) = \sum_{n=-\infty}^{\infty} \frac{n}{1 + n^2} e^{-jn4t}$$

Find the Fourier coefficients of  $Y_n$  of the periodic system output  $y(t)$ ?

# Circuit application –Example 10

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**Solution:** The input Fourier coefficients are,

$$F_n = \frac{n}{1 + n^2}$$

and the fundamental frequency is  $\omega_o = 4 \text{ rad/s}$

$$H(n\omega_o) = H(n4) = \frac{2 + jn4}{3 + jn4}$$

and the Fourier coefficients of the system output  $Y_n$

$$Y_n = H(n\omega_o)F_n = \frac{2 + jn4}{3 + jn4} \cdot \frac{n}{1 + n^2}$$

# Summary

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- The Bode plot describes the magnitude and phase of transfer function in terms of function of frequency
- The LTI systems converts their co-sinusoidal inputs of frequency  $\omega$  into co-sinusoidal outputs having the same frequency: amplitude multiplied and phase added for input output relation
- The linearity and superposition are key principles for applying multifrequency input co-sinusoids



# Summary

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- Any practical periodic function of frequency  $\omega_0$  can be expressed as an infinite sum of sine or cosine functions that are integral multiples of  $\omega_0$
- The Dirichlet conditions are sufficient to satisfy Fourier series
- Using symmetrical simplification reduces tedious job of calculations in Fourier transform
- Superposition and input-output relation jointly solve Fourier series for periodic excitations

# Further reading

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1. Ch. 4 (page 166-181) and Ch. 6 (page 185-208 ), E. Kudeki and D. C. Munson, *Analog Signals and Systems*, Prentice Hall, 2008.
2. Ch. 4 (page 613-619) and Ch. 17 (page 760-780), C. K. Alexander and M. Sadiku, *Fundamentals of Electric Circuits*, 5<sup>th</sup> ed., McGraw-Hill, 2013.
3. Ch. 12 (page 597-607) and Ch. 15 (page 751-773), J. D. Irwin, and R. M. Nelms. *Basic Engineering Circuit Analysis*, 10th ed. Wiley, 2011.

## Preview:

1. Ch. 7 (page 226-247), E. Kudeki and D. C. Munson, *Analog Signals and Systems*, Prentice Hall, 2008.

# Homework 8

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**Deadline: 10:00 PM, 20<sup>th</sup> April, 2022**

**Thank you!**