

# VOLTAGE REGULATOR

Lecture 11

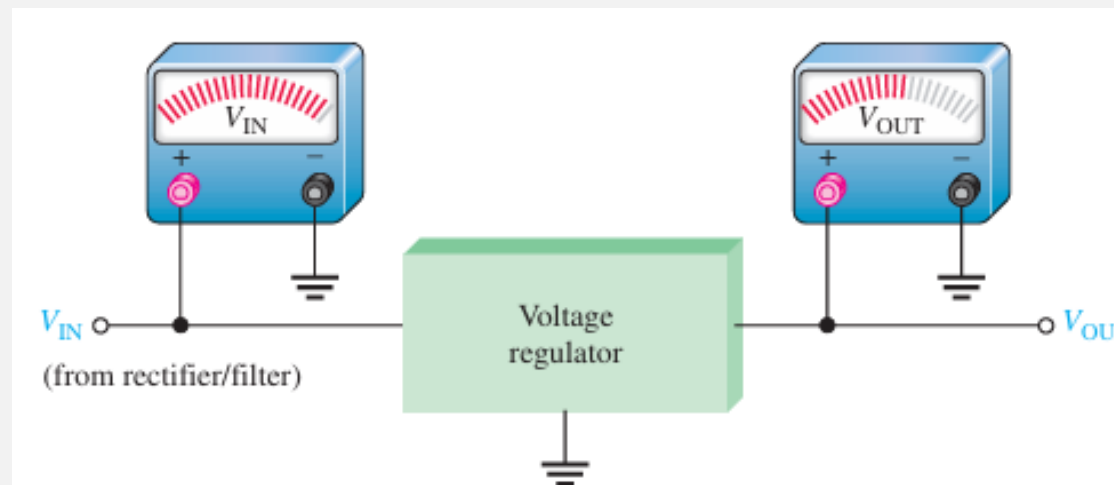
# OVERVIEW

- Voltage Regulation
- Basic Linear Series Regulators
- Basic Linear Shunt Regulators
- Basic Switching Regulators
- Integrated Circuit Voltage Regulators

# VOLTAGE REGULATION

Two basic categories of voltage regulation:

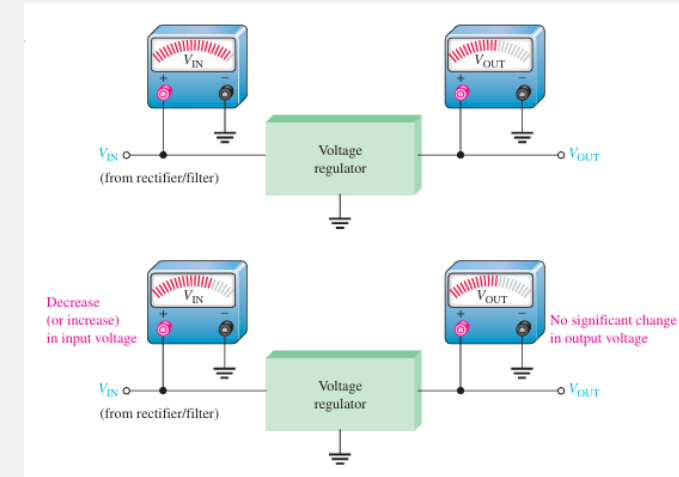
- ❑ **Line Regulation** – maintain a nearly constant output voltage when the input voltage varies.
- ❑ **Load Regulation** – maintain a nearly constant output voltage when the load varies.



# LINE REGULATION

- When the AC input (line) voltage of a power supply changes, an electronic circuit called a regulator maintains a nearly constant output voltage. Line regulation can be defined as the percentage change in the output voltage for a given change in the input voltage. When taken over a range of input voltage values, line regulation is expressed as a percentage by the following formula:

$$\text{Line regulation} = \left( \frac{\Delta V_{OUT}}{\Delta V_{in}} \right) 100\%$$



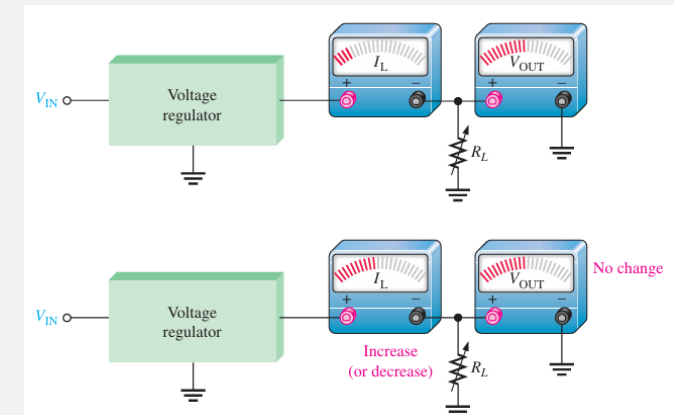
- Line regulation can also be expressed in units of %/V. For example, a line regulation of 0.05%/V means that the output voltage changes 0.05 percent when the input voltage increases or decreases by one Volt. Line regulation can be calculated using the following formula:

$$\text{Line regulation} = \left( \frac{\Delta V_{OUT}/V_{OUT}}{\Delta V_{in}} \right) 100\%$$

# LOAD REGULATION

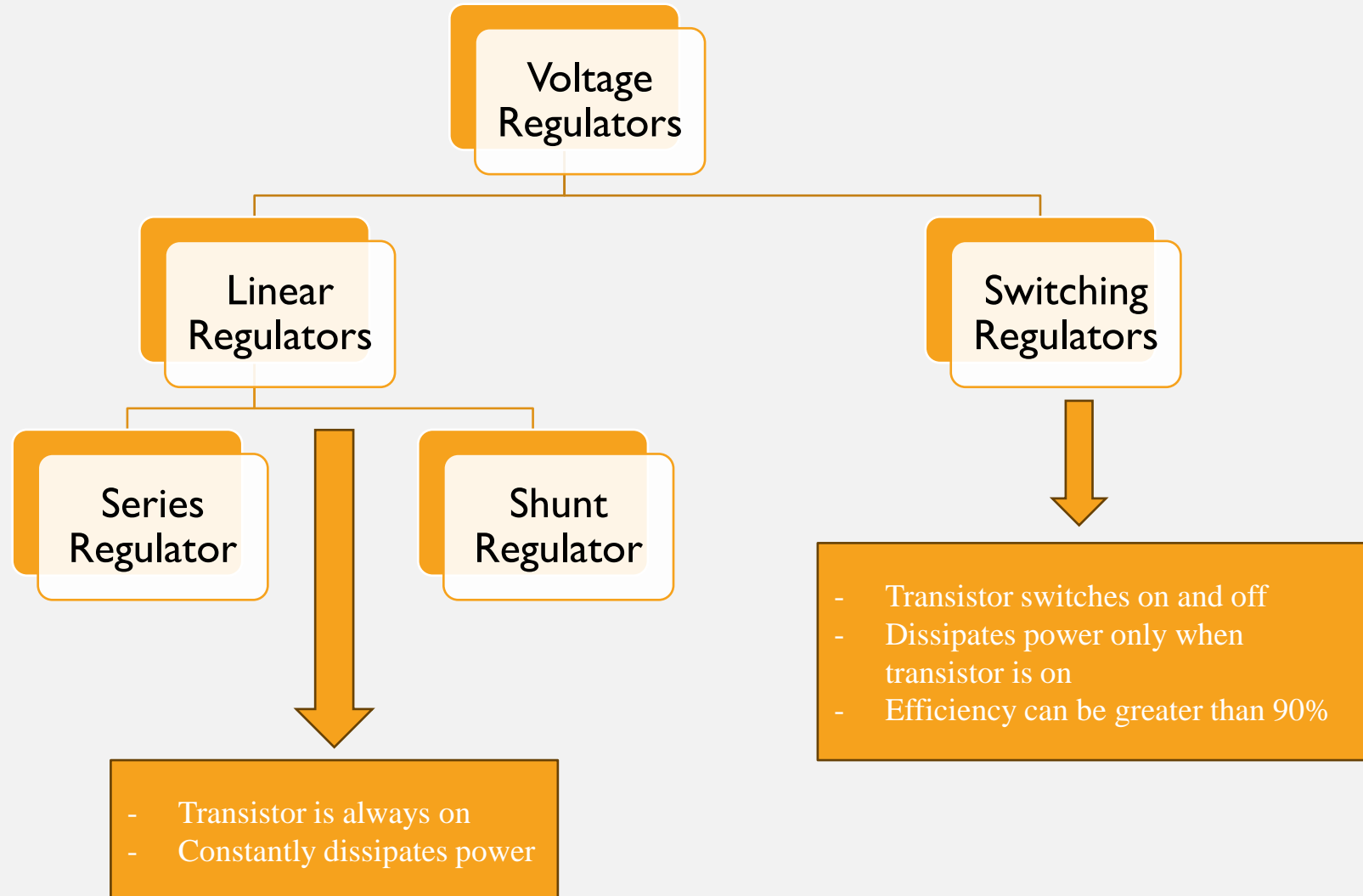
- When the amount of current through a load changes due to a varying load resistance, the voltage regulator must maintain a nearly constant output voltage across the load.
- A change in load current has practically no effect on the output voltage of a regulator (within certain limits).
- **Load regulation** can be defined as the percentage change in output voltage for a given change in load current. One way to express load regulation is as a percentage change in output voltage from no-load (NL) to full-load (FL):

$$\text{Load regulation} = \left( \frac{V_{NL} - V_{FL}}{V_{FL}} \right) 100\%$$



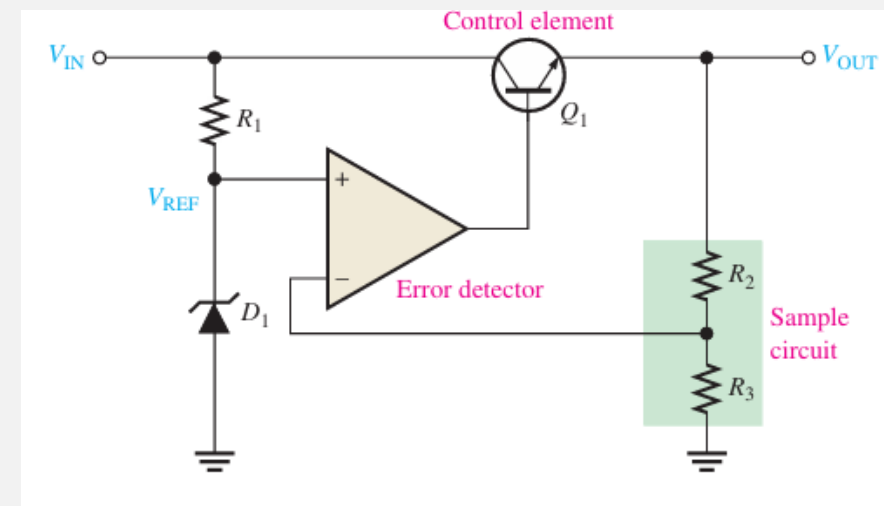
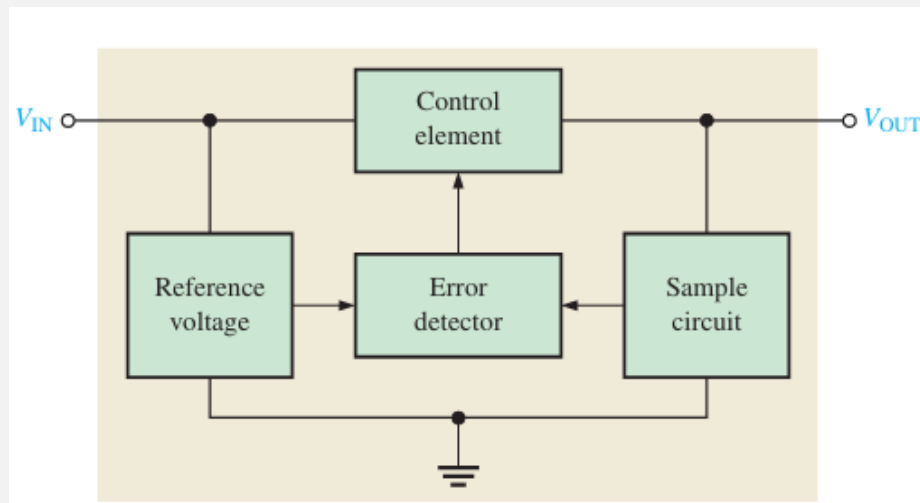
- Alternately, the load regulation can be expressed as a percentage change in output voltage for each mA change in load current. For example, a load regulation of  $0.01\%/mA$  means that the output voltage changes 0.01 percent when the load current increases or decreases 1 mA.

# BASIC LINEAR SERIES REGULATORS



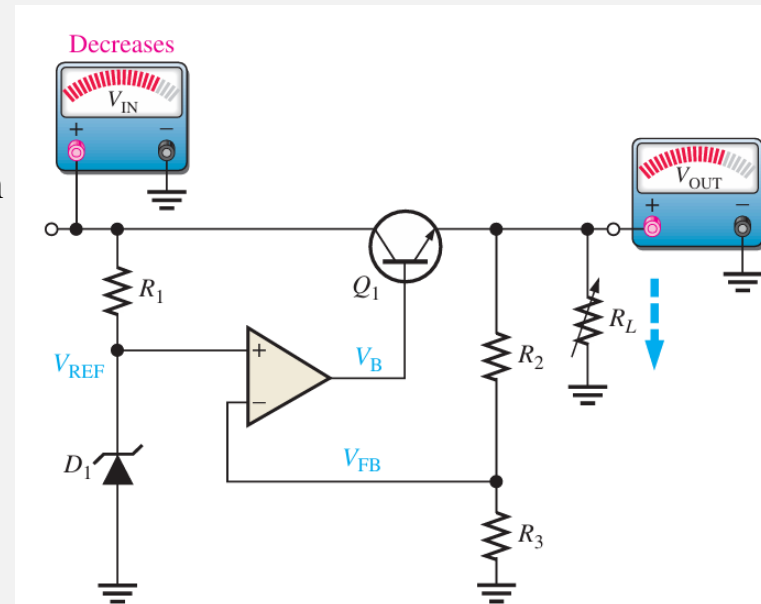
# SERIES REGULATOR

- The control element is a pass transistor in series with the load between the input and output.
- The output sample circuit senses a change in the output voltage.
- The error detector compares the sample voltage with a reference voltage and causes the control element to compensate in order to maintain a constant output voltage.



# SERIES REGULATOR

- The resistive voltage divider formed by  $R_2$  and  $R_3$  senses any change in the output voltage.
- When the output tries to decrease because of a decrease in  $V_{IN}$  or because of an increase in  $I_L$  caused by a decrease in  $R_L$ , a proportional voltage decrease is applied to the op-amp's inverting input by the voltage divider.
- Since the zener diode ( $D_1$ ) holds the other op-amp input at a nearly constant reference voltage,  $V_{REF}$ , a small difference voltage (error voltage) is developed across the op-amp's inputs. This difference voltage is amplified, and the op-amp's output voltage,  $V_B$ , increases.
- This increase is applied to the base of  $Q_1$ , causing emitter voltage  $V_{OUT}$  to increase until the voltage to the inverting input again equals the reference (zener) voltage.





# SERIES REGULATOR

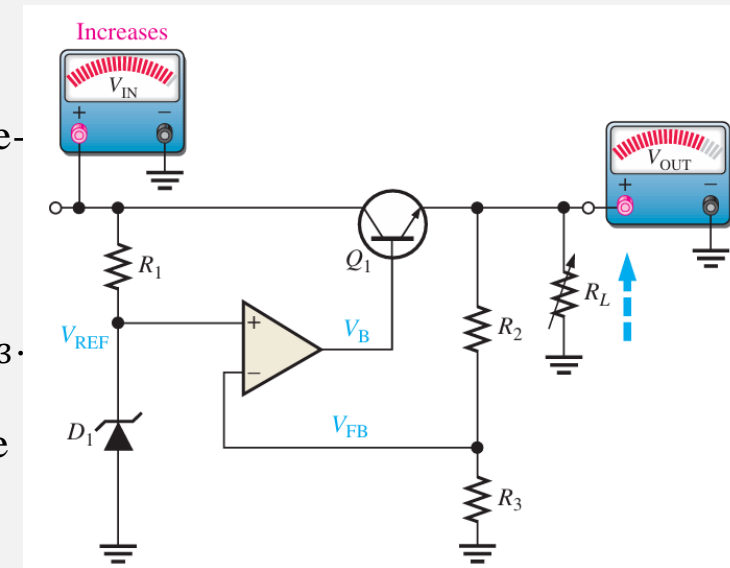
- The op-amp in the series regulator is actually connected as a noninverting amplifier where the reference voltage  $V_{REF}$  is the input at the noninverting terminal, and the  $R_2/R_3$  voltage divider forms the negative feedback circuit. The closed-loop voltage gain is:

$$A_{cl} = 1 + \frac{R_2}{R_3}$$

- Therefore, the regulated output voltage of the series regulator (neglecting the base-emitter voltage of  $Q_1$ ) is

$$V_{OUT} \cong \left(1 + \frac{R_2}{R_3}\right) V_{REF}$$

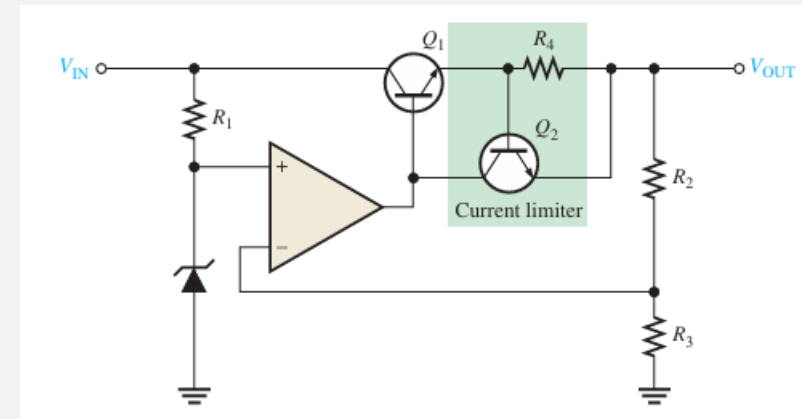
- The output voltage is determined by the zener voltage and the resistors  $R_2$  and  $R_3$ .
- The output voltage is relatively independent of the input voltage, and therefore, regulation is achieved (as long as the input voltage and load current are within the specified limits).
- When  $V_{IN}$  or  $R_L$  increases,  $V_{OUT}$  attempts to increase.
- The feedback voltage,  $V_{FB}$ , also attempts to increase, and as a result,  $V_B$ , applied to the base of the control transistor, attempts to decrease, thus compensating for the attempted increase in  $V_{OUT}$  by decreasing the  $Q_1$  emitter voltage.
- When  $V_{IN}$  (or  $R_L$ ) stabilizes at its new higher value, the voltages return to their original values, thus keeping  $V_{OUT}$  constant as a result of the negative feedback.



# SHORT CIRCUIT OR OVERLOAD PROTECTION

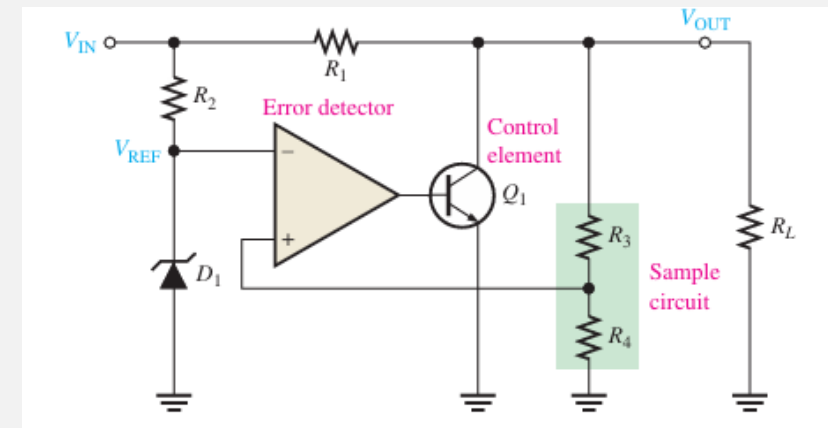
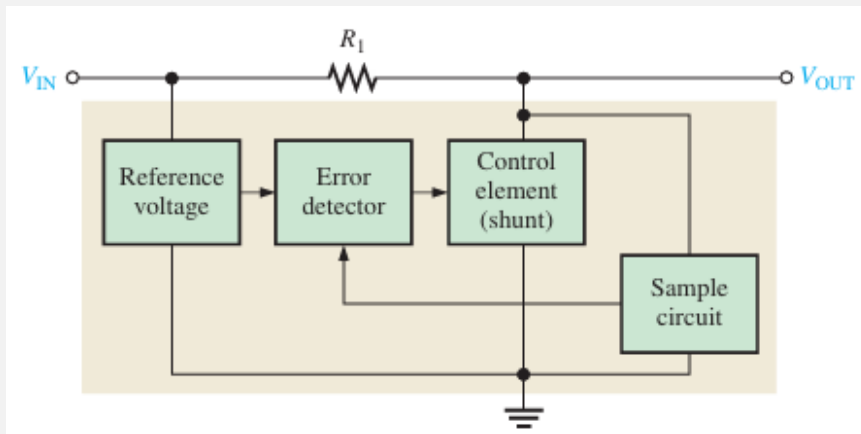
- If an excessive amount of load current is drawn, the series-pass transistor can be quickly damaged or destroyed.
- Most regulators use some type of excess current protection in the form of a current-limiting mechanism.
- The current-limiting circuit consists of transistor  $Q_2$  and resistor  $R_4$ .
- The load current through  $R_4$  produces a voltage from base to emitter of reaches a predetermined maximum value, the voltage drop across  $R_4$  is sufficient to forward-bias the base-emitter junction of  $Q_2$ , thus causing it to conduct.
- Enough op-amp output current is diverted through  $Q_2$ , to reduce the  $Q_1$  base current, so that  $I_L$  is limited to its maximum value,  $I_{L(\max)}$ .
- Since the base-to-emitter voltage of  $Q_2$ , cannot exceed approximately  $0.7\text{ V}$ , the voltage across  $R_4$  is held to this value, and the load current is limited to

$$I_{L(\max)} = \frac{0.7\text{ V}}{R_4}$$



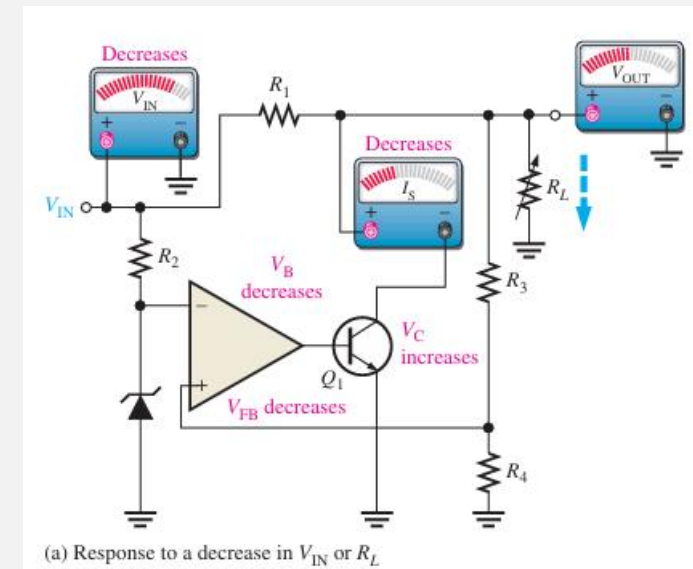
# SHUNT REGULATORS

- The control element is a transistor,  $Q_1$ , in parallel with the load.
- A resistor,  $R_1$ , is in series with the load.



# SHUNT REGULATORS

- The operation of the circuit is similar to that of the series regulator, except that regulation is achieved by controlling the current through the parallel transistor  $Q_1$ .
- When the output voltage tries to decrease due to a change in input voltage or load current caused by a change in load resistance, the attempted decrease is sensed by  $R_3$  and  $R_4$  and applied to the op-amp's noninverting input.
- The resulting difference voltage reduces the op-amp's output ( $V_B$ ), driving  $Q_1$  less, thus reducing its collector current (shunt current) and increasing the collector voltage. Thus, the original decrease in voltage is compensated for by this increase, keeping the output nearly constant.



# SHUNT REGULATORS

- The opposite action occurs when the output tries to increase.
- With  $I_L$  and  $V_{OUT}$  constant, a change in the input voltage produces a change in shunt current ( $I_S$ ) as follows:

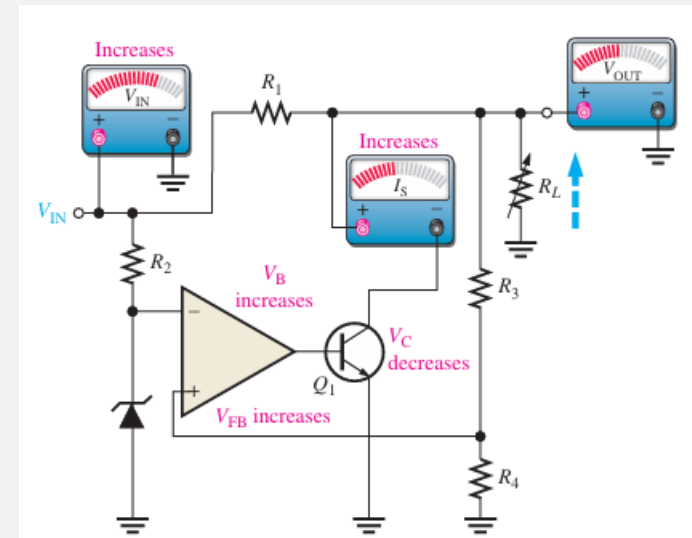
$$\Delta I_S = \frac{\Delta V_{IN}}{R_1}$$

- With a constant  $V_{IN}$  and  $V_{OUT}$ , a change in load current causes an opposite change in shunt current. If  $I_L$  increases,  $I_S$  decreases, and vice versa.

$$\Delta I_S = -\Delta I_L$$

- The shunt regulator is less efficient than the series type but offers inherent short-circuit protection.
- If the output is shorted ( $V_{OUT} = 0$ ), the load current is limited by the series resistor  $R_1$  to a maximum value as follows ( $I_S = 0$ ).

$$I_{L(max)} = \frac{V_{IN}}{R_1}$$



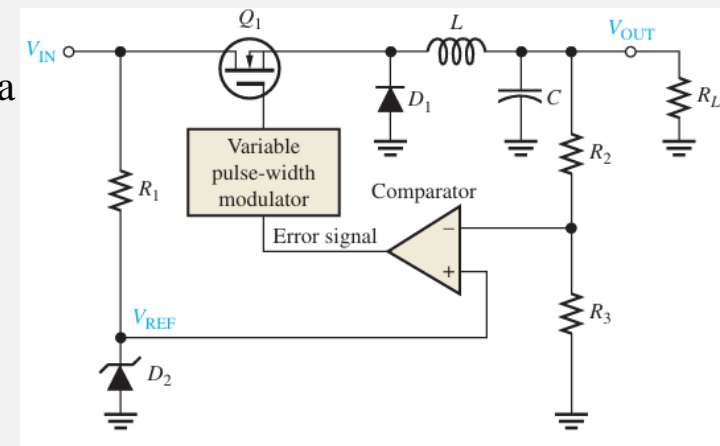
# BASIC SWITCHING REGULATORS

- Switching regulators are designed for various power levels.
- They range in power levels from less than one watt for some battery-operated portable equipment to hundreds and thousands of watts in major applications.
- The requirements for the application determine the particular design, but all switching regulators require feedback to control the on-off time for the switch.
- Three basic configurations of switching regulators are
  - ❖ Step-down,
  - ❖ Step-up
  - ❖ Inverting
- In some cases, such as a laptop computer, all three types may be employed for various parts of the system; for example, the display typically will use an inverting type, the microprocessor would use a step-down type, and the disk drive may use a step-up type.

# STEP-DOWN CONFIGURATION

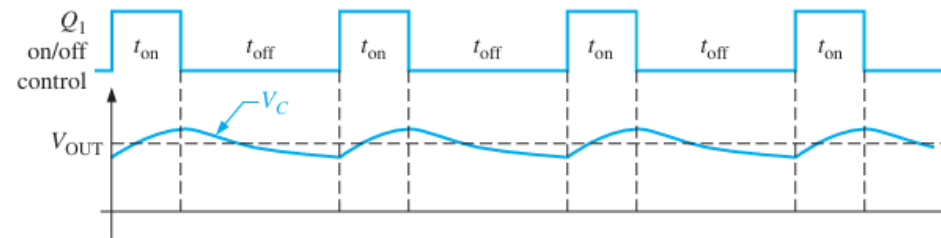
- In the step-down configuration (also called a *buck converter*), the output voltage is always less than the input voltage.
- The basic control element is a high-speed switch, which opens and closes rapidly from a control circuit that senses the output, and it adjusts the on-time and the off-time to keep the desired output.
- When the switch is closed, the diode is off and the magnetic field of the inductor builds, storing energy.
- When the switch opens, the magnetic field collapses, keeping nearly constant current in the load.
- A path for the load current is provided through the forward-biased diode (as long as the load resistance is not too large).
- The capacitor smoothes the DC to a nearly constant level.

- Figure shows a basic step-down switching regulator using a D-MOSFET switching transistor.
- MOSFET transistors can switch faster than BJTs and have been improved in recent years, so they have become the preferred type of switching device, provided that the off-state voltage is not too high.

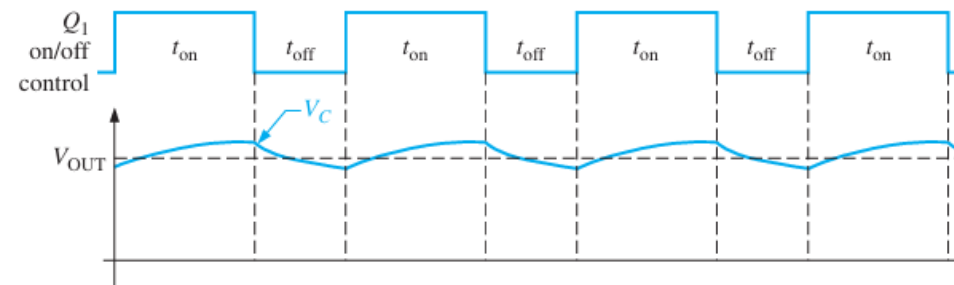


# SWITCHING REGULATOR WAVEFORMS

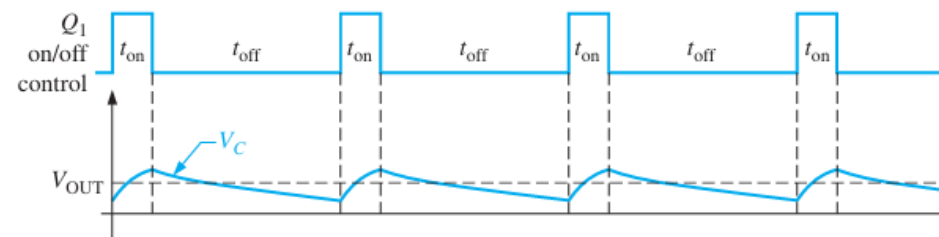
- The  $V_C$  waveform is shown for no inductive filtering to illustrate the charge and discharge action (ripple).
- $L$  and  $C$  smooth  $V_C$  to a nearly constant level, as indicated by the dashed line for  $V_{OUT}$ .



(a)  $V_{OUT}$  depends on the duty cycle.



(b) Increase the duty cycle and  $V_{OUT}$  increases.

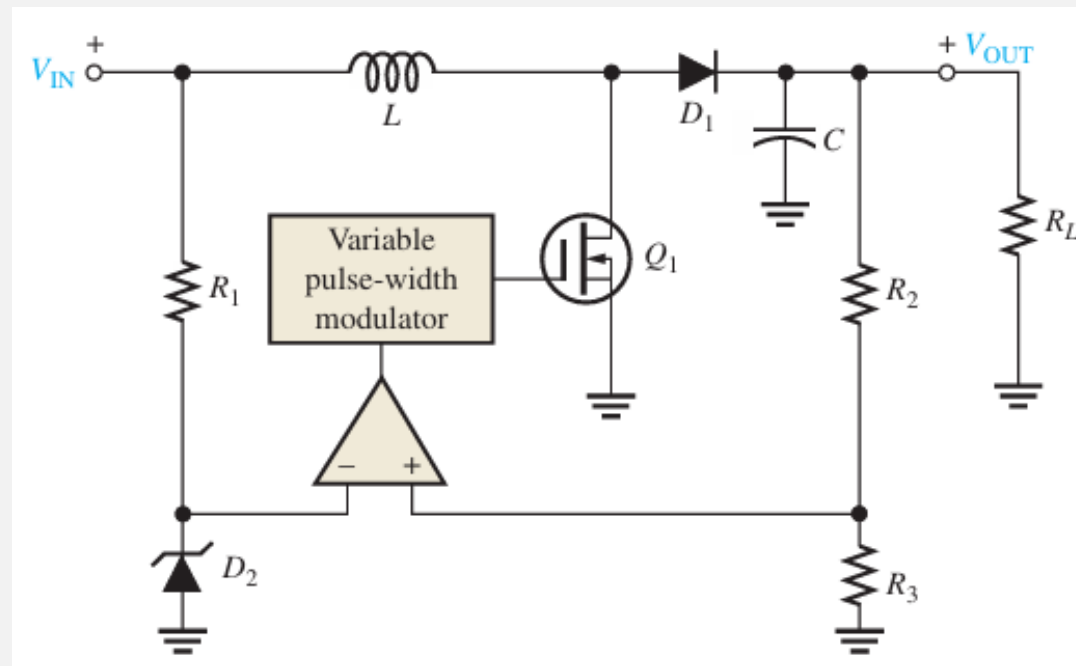


(c) Decrease the duty cycle and  $V_{OUT}$  decreases.



# STEP-UP CONFIGURATION

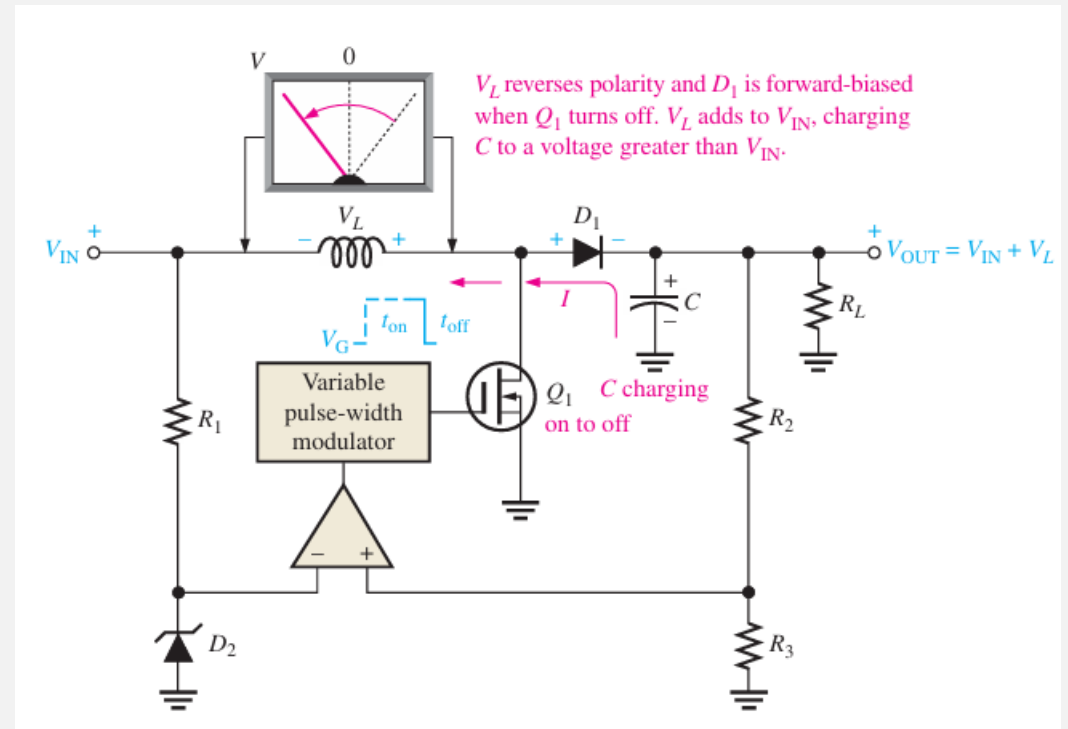
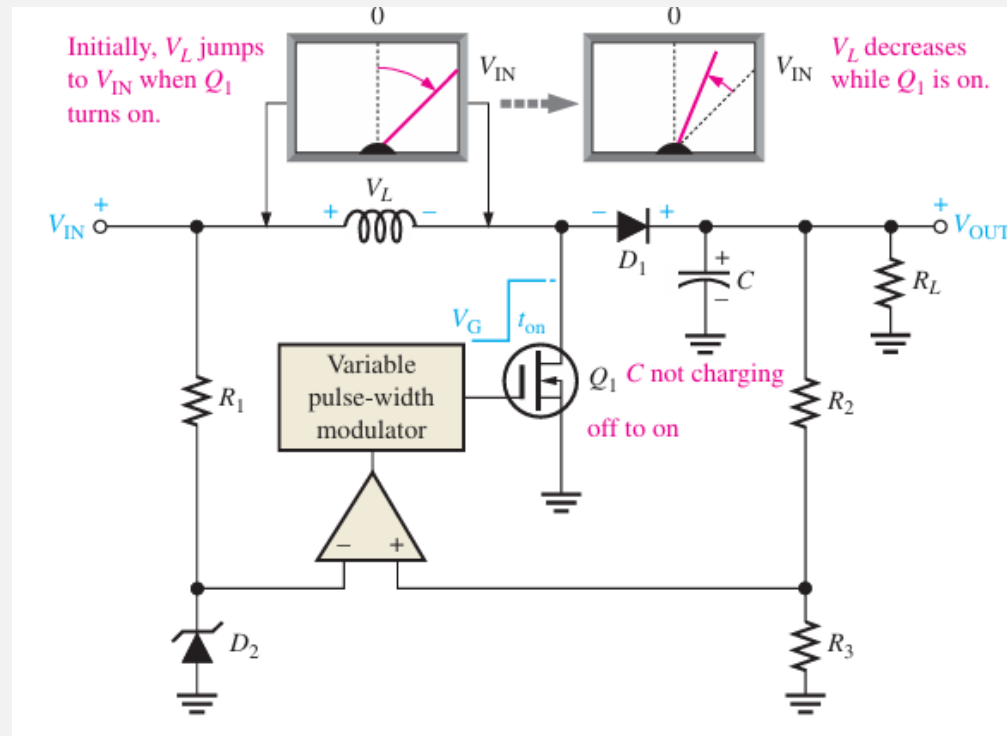
- A basic step-up type of switching regulator (sometimes called a *boost converter*) is shown in Figure below, where transistor  $Q_1$  operates as a switch to ground.



# STEP-UP CONFIGURATION

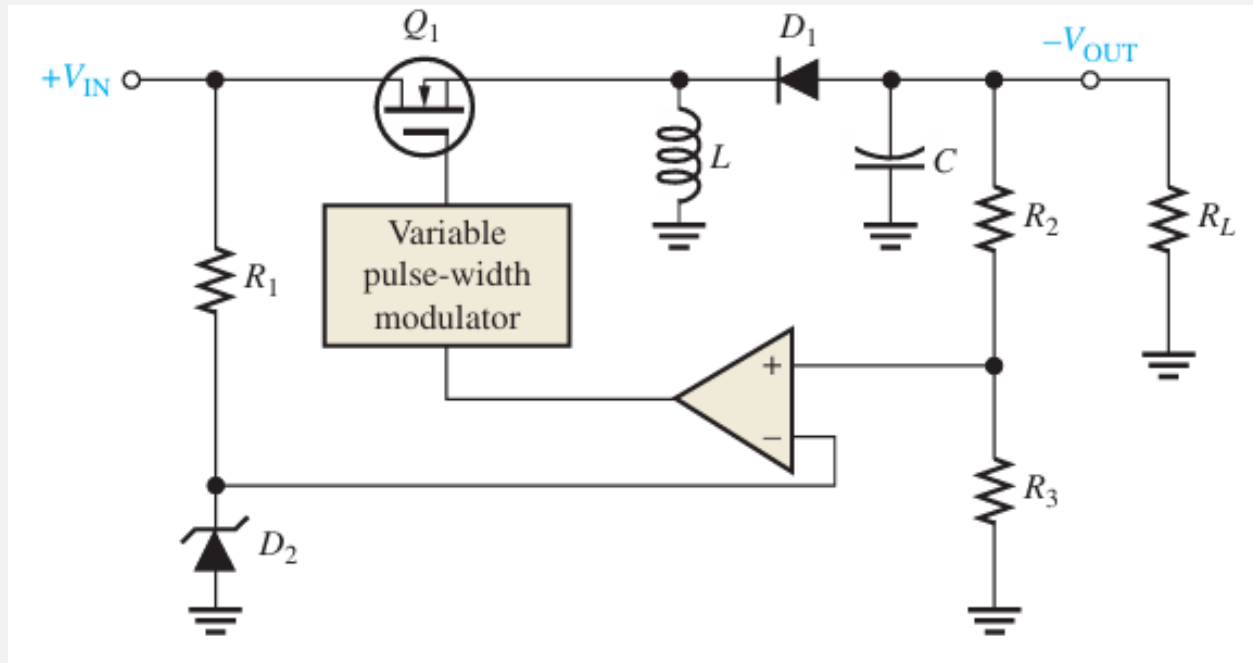
- When  $Q_1$  turns on, a voltage equal to approximately  $V_{IN}$  is induced across the inductor with a polarity.
- During the on-time ( $t_{on}$ ) of  $Q_1$ , the inductor voltage,  $V_L$ , decreases from its initial maximum and diode  $D_1$  is reverse-biased.
- The longer  $Q_1$  is on, the smaller  $V_L$  becomes.
- During the on-time, the capacitor only discharges an extremely small amount through the load.
- When  $Q_1$  turns off, the inductor voltage suddenly reverses polarity and adds to  $V_{IN}$ , forward-biasing diode  $D_1$  and allowing the capacitor to charge.
- The output voltage is equal to the capacitor voltage and can be larger than  $V_{IN}$  because the capacitor is charged to  $V_{IN}$  plus the voltage induced across the inductor during the off-time of  $Q_1$ .
- The output voltage is dependent on both the inductor's magnetic field action (determined by  $t_{on}$ ) and the charging of the capacitor (determined by  $t_{off}$ ).
- Voltage regulation is achieved by the variation of the on-time of  $Q_1$  (within certain limits) as related to changes in  $V_{OUT}$  due to changing load or input voltage.
- If  $V_{OUT}$  tries to increase, the on-time of  $Q_1$  will decrease, resulting in a decrease in the amount that  $C$  will charge.
- If  $V_{OUT}$  tries to decrease, the on-time of  $Q_1$  will increase, resulting in an increase in the amount that  $C$  will charge. This regulating action maintains  $V_{OUT}$  at an essentially constant level.

# STEP-UP CONFIGURATION



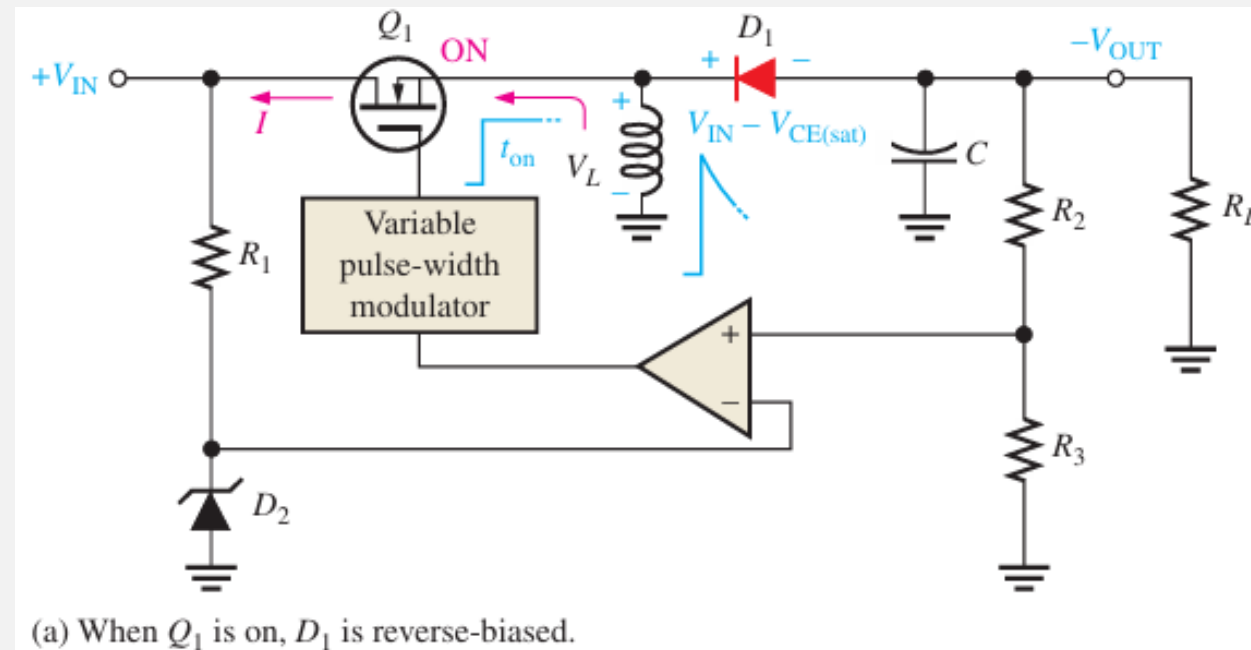
# VOLTAGE-INVERTER CONFIGURATION

- ❑ Produces an output voltage that is opposite in polarity to the input.
- ❑ Sometimes called a *buck-boost* converter.



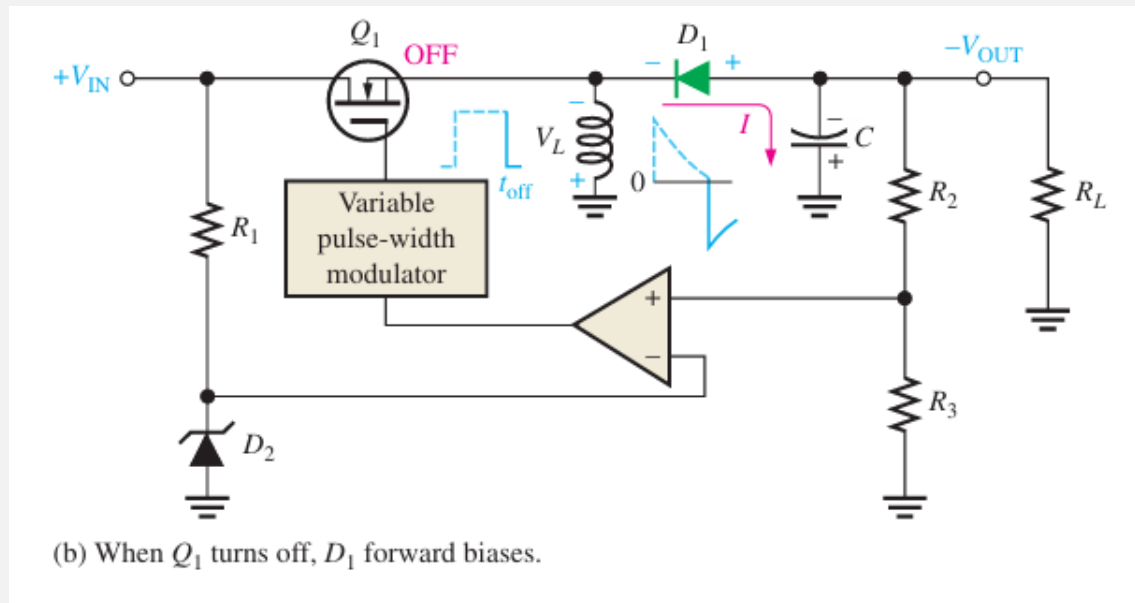
# VOLTAGE-INVERTER CONFIGURATION

- When  $Q_1$  turns on, the inductor voltage jumps to approximately  $V_{IN} - V_{CE(sat)}$  and the magnetic field rapidly expands, as shown.
- While  $Q_1$  is on, the diode is reverse-biased and the inductor voltage decreases from its initial maximum.



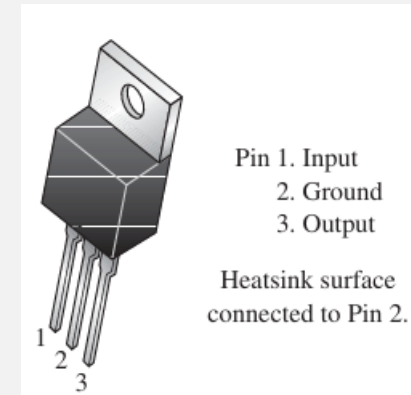
# VOLTAGE-INVERTER CONFIGURATION

- When  $Q_1$  turns off, the magnetic field collapses and the inductor's polarity reverses.
- This forward-biases the diode, charges  $C$ , and produces a negative output voltage, as indicated.
- The repetitive on-off action of  $Q_1$  produces a repetitive charging and discharging that is smoothed by the  $LC$  filter action.



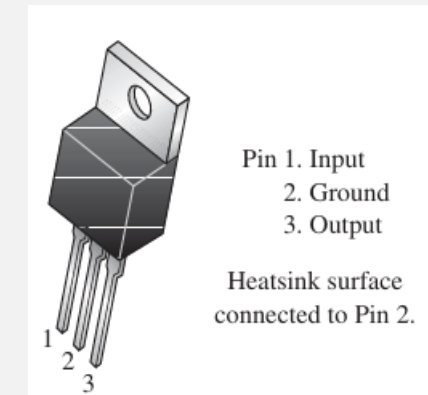
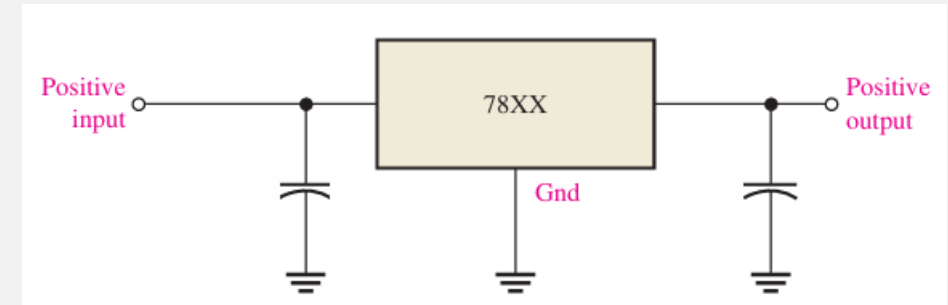
# INTEGRATED CIRCUIT VOLTAGE REGULATORS

- Several types of both linear and switching regulators are available in integrated circuit (IC) form.
- Generally, the linear regulators are three-terminal devices that provide either positive or negative output voltages that can be either fixed or adjustable.



# FIXED POSITIVE LINEAR VOLTAGE REGULATORS

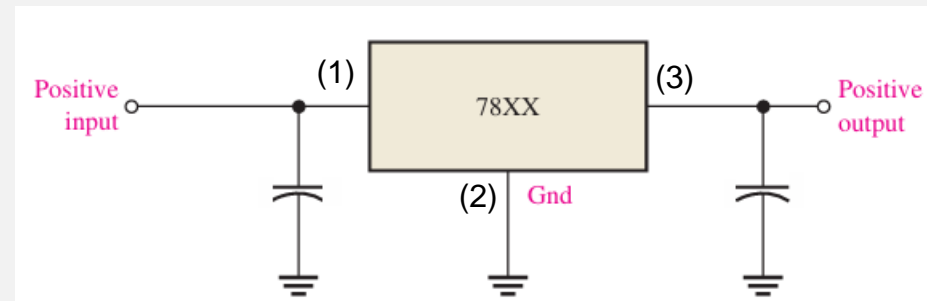
- 78XX series of IC regulators is representative of three-terminal devices that provide a fixed positive output voltage.
- The three terminals are input, output, and ground as indicated in the standard fixed voltage configuration
- The last two digits in the part number designate the output voltage. For example, the 7805 is a +5.0 V
- For any given regulator, the output voltage can be as much as 4% of the nominal output.
- regulator. For any given regulator, the output voltage can
- A 7805 may have an output from 4.8 V to 5.2 V but will remain constant in that range.





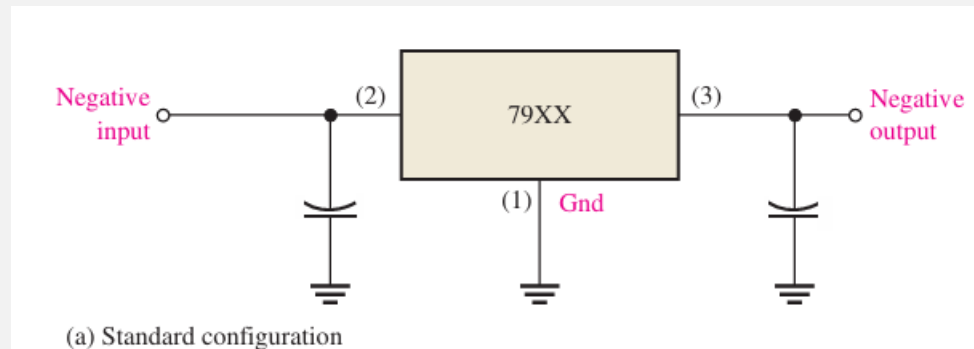
# FIXED POSITIVE LINEAR VOLTAGE REGULATORS

- Capacitors, although not always necessary, are sometimes used on the input and output.
- The output capacitor acts basically as a line filter to improve transient response.
- The input capacitor filters the input and prevents unwanted oscillations when the regulator is some distance from the power supply filter such that the line has a significant inductance.
- The 78XX series can produce output currents up to in excess of 1 A when used with an adequate heat sink.
- The input voltage must be approximately 2.5 V above the output voltage in order to maintain regulation.
- The circuits have internal thermal overload protection and short-circuit current-limiting features.
- Thermal overload occurs when the internal power dissipation becomes excessive and the temperature of the device exceeds a certain value.
- Almost all applications of regulators require that the device be secured to a heat sink to prevent thermal overload.



# FIXED NEGATIVE LINEAR VOLTAGE REGULATORS

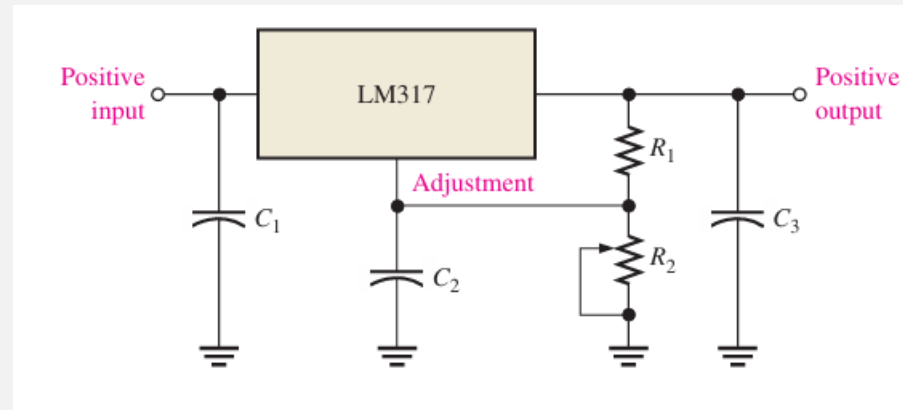
- The 79XX series is typical of three-terminal IC regulators that provide a fixed negative output voltage.
- This series is the negative-voltage counterpart of the 78XX series and shares most of the same features and characteristics except the pin numbers are different than the positive regulators.



Type number	Output voltage
7905	-5.0 V
7905.2	-5.2 V
7906	-6.0 V
7908	-8.0 V
7912	-12.0 V
7915	-15.0 V
7918	-18.0 V
7924	-24.0 V

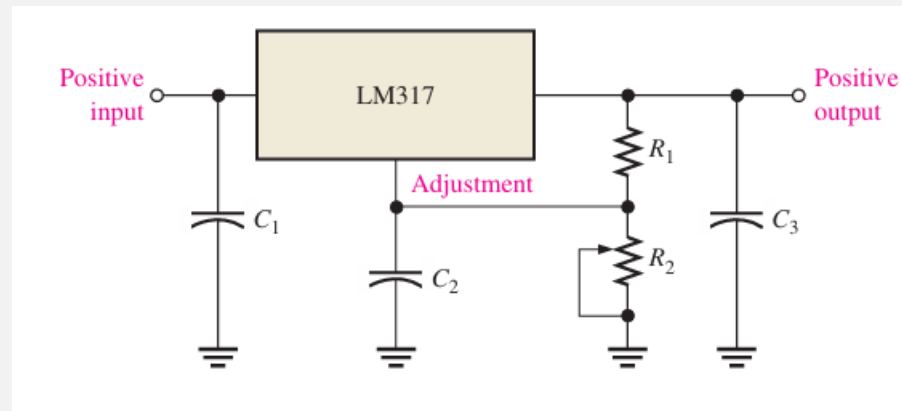
# ADJUSTABLE POSITIVE LINEAR VOLTAGE REGULATORS

- The LM317 is an example of a three-terminal positive regulator with an adjustable output voltage.
- The capacitors are for decoupling and do not affect the DC operation.
- There is an input, an output, and an adjustment terminal in this type of regulator.
- The external fixed resistor  $R_1$  provide the output voltage adjustment.
- $V_{OUT}$  and the external variable resistor  $R_2$  can be varied from 1.2 V to 37 V depending on the resistor values.
- The LM317 can provide over 1.5 A of output current to a load.



# ADJUSTABLE NEGATIVE LINEAR VOLTAGE REGULATORS

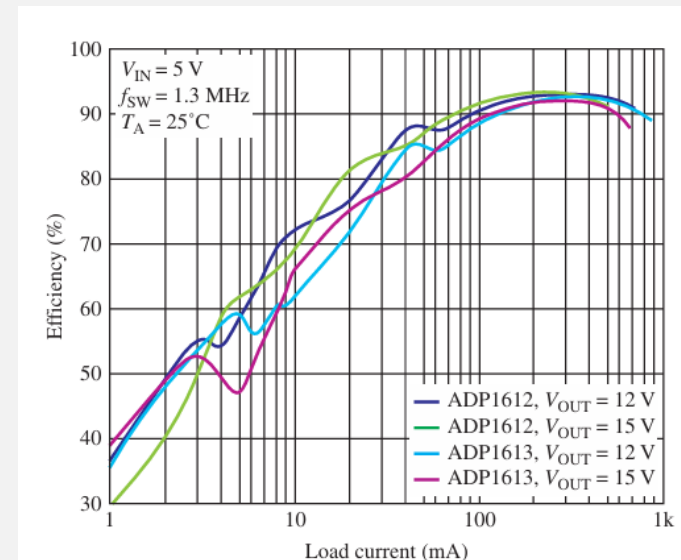
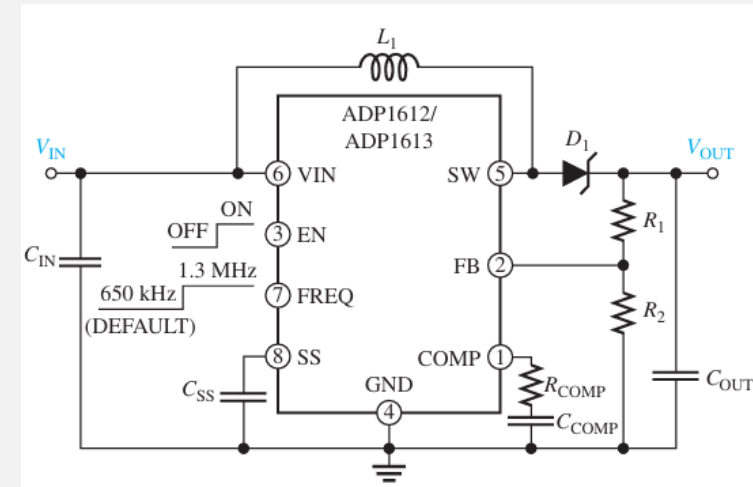
- The LM337 is the negative output counterpart of the LM317 and is a good example of this type of IC regulator.
- Like the LM317, the LM337 requires two external resistors for output voltage adjustment.
- The output voltage can be adjusted from  $-1.2\text{ V}$  to  $-37\text{ V}$ , depending on the external resistor values.
- The capacitors are for decoupling and do not affect the dc operation.



# SWITCHING VOLTAGE REGULATORS

## ❖ The Step-Up Switching Regulator

- The step-up regulator configuration using an ADP1612/ADP1613 is shown in the Figure.
- The ADP1612 and the ADP1613 are essentially the same except for their switching frequency, which is used in the pulse-width modulation (PWM) operation.
- This regulator operates with PWM and exhibits an efficiency of up to 94% at the higher switch frequency, depending on the output current and voltage.
- The load current increases, the efficiency increases.
- The output voltage has a much smaller effect.
- The operating frequency of the PWM is pin-selectable for 650 kHz or 1.3 MHz.
- The lower frequency results in better efficiency, and the higher frequency allows the use of smaller external components.



# SWITCHING VOLTAGE REGULATORS

## ❖ The Step-Down Switching Regulator

- The step-down regulator configuration using an ADP2300/ADP2301 is shown in the Figure.
- The ADP2300 and the ADP2301 are essentially the same except for their switching frequency.
- Unlike the ADP1612/ADP1613, this device does not have pin-selectable frequencies. Instead, each has a fixed internal oscillator.
- This regulator operates with PWM and exhibits an efficiency of up to 91%, depending on the output current.
- This device has thermal shutdown (TSP) protection in case temperature exceeds 140°C and turns back on when the temperature drops to 150°C.
- Also, it has an under-voltage lock-out (UVLO) feature and short-circuit protection.
- As the load current increases above about 0.2 A, the efficiency remains relatively constant (between about 91% and about 88%) and drops off a little as the output current increases.

