

#### 25V 10A GATE DRIVER IN SOT26

#### **Description and Applications**

The ZXGD3005E6 is a high-speed non-inverting single gate driver capable of driving up to 10A into a MOSFET or IGBT gate capacitive load from supply voltages up to 25V. With propagation delay times down to <10ns and correspondingly rise/fall times of <20ns.

This gate driver ensures rapid switching of the MOSFET or IGBT to minimize power losses and distortion in high current switching applications. It is ideally suited to act as a voltage buffer between the typically high output impedances of a controller IC and the effectively low impedance on the gate of a power MOSFET or IGBT during switching. Its low input voltage requirement and high current gain allows high current driving from low voltage controller ICs.

The ZXGD3005E6 has separate source and sink outputs that enables the turn-on and turn-off times of the MOSFET or IGBT to be independently controlled. In addition, the wide supply voltage range allows full enhancement of the MOSFET or IGBT to minimize on-state losses and permits +15V to -5V gate drive voltage to prevent dV/dt induced false triggering of IGBTs. The ZXGD3005E6 has been designed to be inherently rugged to latch-up and shoot-through issues. The optimized pin-out SOT26 package eases board layout, enabling reduced parasitic inductance of traces.

#### Power MOSFET and IGBT Gate Driving in:

- Synchronous switch-mode power supplies
- Power Factor Correction (PFC) in power supplies
- · Secondary side synchronous rectification
- Plasma Display Panel power modules
- 1, 2 and 3-phase motor control circuits
- · Audio switching amplifier power output stages
- Solar inverters

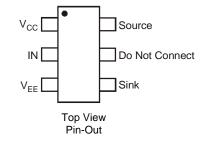
#### **Features and Benefits**

- Emitter-follower configuration for ultra-fast switching
  - <10ns propagation delay time
  - <20ns rise/fall time</p>
- Non-inverting voltage buffer stage
- Wide supply voltage up to 25V to minimize on-losses
- 10A peak current drive into capacitive loads
- Low input current of 1mA to deliver 4A output current
- Separate source and sink outputs for independent control of rise and fall time
- Optimized pin-out to ease board layout and minimize parasitic inductance of traces
- Rugged design that avoids latch-up or shoot-through issues
- Near Zero quiescent supply current
- "Lead-Free", RoHS Compliant (Note 1)
- "Green" Devices (Note 2)
- Qualified to AEC-Q101 Standards for High Reliability

#### **Mechanical Data**

- Case: SOT26
- Case material: Molded Plastic. "Green" Molding Compound.
- UL Flammability Rating 94V-0
- Moisture Sensitivity: Level 1 per J-STD-020
- Terminals: Matte Tin Finish
- Weight: 0.018 grams (approximate)





Pin Name	Pin Function
Vcc	Supply voltage high
IN	Driver input pin
V <sub>EE</sub>	Supply voltage low
SOURCE	Source current output
SINK	Sink current output

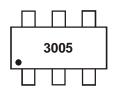
#### Ordering Information (Note 3)

Product	Marking	Reel size (inches)	Tape width (mm)	Quantity per reel
ZXGD3005E6TA	3005	7	8	3000

Notes:

- 1. No purposefully added lead
- 2. "Green" devices, Halogen and Antimony Free, Diodes Inc's "Green" Policy can be found on our website at http://www.diodes.com
- 3. For packaging details, go to our website at http://www.diodes.com

### **Marking Information**

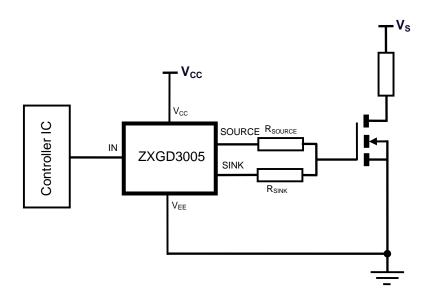


3005 = Product Type Marking Code





## **Typical Application Circuit**



# Maximum Ratings @T<sub>A</sub> = 25°C unless otherwise specified

Characteristic	Symbol	Value	Unit
Supply voltage, with respect to V <sub>EE</sub>	V <sub>CC</sub>	25	V
Input voltage, with respect to V <sub>EE</sub>	V <sub>IN</sub>	25	V
Output difference voltage (Source – Sink)	$\Delta V_{(source-sink)}$	±7.5	V
Peak output current	I <sub>PK</sub>	±10	Α
Input current	I <sub>IN</sub>	±100	mA

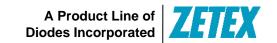
### Thermal Characteristics @T<sub>A</sub> = 25°C unless otherwise specified

Characteristic	Symbol	Value	Unit	
Power Dissipation (Notes 4 & 5)	В	1.1	W	
Linear derating factor	PD	8.8	mW/°C	
Thermal Resistance, Junction to Ambient (Notes 4 & 5)	$R_{ heta JA}$	113	00/14/	
Thermal Resistance, Junction to Lead (Note 6)	$R_{\theta JL}$	105	°C/W	
Operating and Storage Temperature Range	T <sub>J.</sub> T <sub>STG</sub>	-55 to +150	°C	

Notes:

- 4. For a device surface mounted on 25mm x 25mm x 0.6mm FR4 PCB with high coverage of single sided 1oz copper, in still air conditions; the device is measured when operating in a steady-state condition. The heatsink is split in half with the pin 1 (V<sub>CC</sub>) and pin 3 (V<sub>EE</sub>) connected separately to each half.
- 5. For device with two active die running at equal power.
- 6. Thermal resistance from junction to solder-point at the end of each lead on pin 1 (V<sub>CC</sub>) and pin 3 (V<sub>EE</sub>).

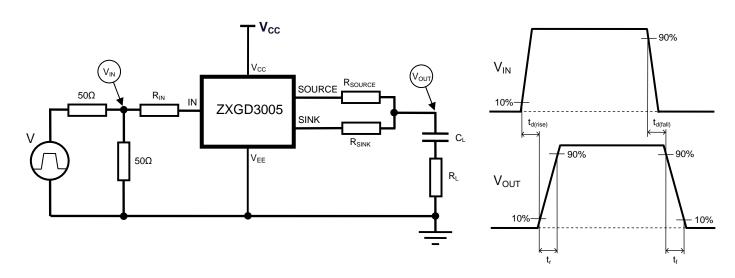




### Electrical Characteristics @T<sub>A</sub> = 25°C unless otherwise specified

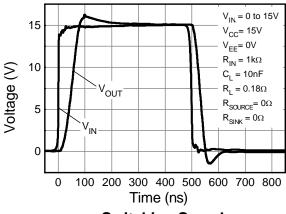
Characteristic	Symbol	Min	Тур	Max	Unit	Test Condition
Output voltage, high	V <sub>OH</sub>	-	V <sub>CC</sub> - 0.8	-	V	$V_{IN} = V_{CC}$
Output voltage, low	V <sub>OL</sub>	-	V <sub>EE</sub> + 0.2	V <sub>EE</sub> + 0.5	V	VIN = VEE
Supply breakdown voltage	D)/	25	-	-	V	$I_Q = 100\mu A$ , $V_{IN} = V_{CC}$
Supply breakdown voltage	BV <sub>CC</sub>	25	-	-	٧	$I_Q = 100 \mu A$ , $V_{IN} = V_{EE} = 0 V$
Quiescent supply current		1	-	50	nA	$V_{CC}=20V, V_{IN}=V_{CC}$
Quiescent supply current	ΙQ	-	-	50	ПА	$V_{CC}=20V$ , $V_{IN}=V_{EE}=0V$
Source current	I <sub>(source)</sub>	-	4.0	-	Α	V <sub>CC</sub> = 5V, I <sub>IN</sub> = 1mA, V <sub>OUT</sub> = 0V
Sink current	I <sub>(sink)</sub>	-	3.8	-	Λ	V <sub>CC</sub> = 5V, I <sub>IN</sub> =-1mA, V <sub>OUT</sub> = 5V
Source current with varying input resistances	I <sub>(source)</sub>	1	6.4 5.5 3.9 2.2 0.44	-	Α	$ \begin{aligned} R_{IN} &= 200\Omega \\ R_{IN} &= 1k\Omega \\ R_{IN} &= 10k\Omega \\ R_{IN} &= 100k\Omega \\ R_{IN} &= 1000k\Omega \end{aligned}  \begin{aligned} &V_{CC} &= 15V,  V_{EE} = 0V \\ &V_{IN} &= 15V \\ &C_{L} &= 100nF,  R_{L} = 0.18\Omega \\ &R_{SOURCE} &= 0\Omega,  R_{SINK} = 0\Omega \end{aligned}$
Sink current with varying input resistances	I <sub>(sink)</sub>	-	7.7 6.5 4.4 2.3 0.46	-	Α	$ \begin{aligned} R_{IN} &= 200\Omega \\ R_{IN} &= 1k\Omega \\ R_{IN} &= 10k\Omega \\ R_{IN} &= 100k\Omega \\ R_{IN} &= 1000k\Omega \end{aligned}  \begin{aligned} &V_{CC} &= 15V,  V_{EE} = 0V \\ &V_{IN} &= 15V \\ &C_{L} &= 100nF,  R_{L} = 0.18\Omega \\ &R_{SOURCE} &= 0\Omega,  R_{SINK} = 0\Omega \end{aligned}$
Switching times with low load capacitance C <sub>L</sub> = 10nF	t <sub>d(rise)</sub> t <sub>r</sub> t <sub>d(fall)</sub>	-	8 48 16 35	-	ns	$\begin{aligned} &V_{CC} = 15V,  V_{EE} = 0V \\ &V_{IN} = 0 \text{ to } 15V \\ &R_{IN} = 1k\Omega \\ &C_L = 10nF,  R_L = 0.18\Omega \\ &R_{SOURCE} = 0\Omega,  R_{SINK} = 0\Omega \end{aligned}$
Switching times with high load capacitance $C_L = 100 nF$	t <sub>d(rise)</sub> t <sub>r</sub> t <sub>d(fall)</sub>	-	46 419 47 467	-	ns	$\begin{split} &V_{CC}=15V,V_{EE}=0V\\ &V_{IN}=0\text{to}15V\\ &R_{IN}=1k\Omega\\ &C_L=100nF,R_L=0.18\Omega\\ &R_{SOURCE}=0\Omega,R_{SINK}=0\Omega \end{split}$
Switching times with asymmetric source and sink resistors	$\begin{array}{c} t_{\text{d(rise)}} \\ t_{\text{r}} \\ t_{\text{d(fall)}} \\ t_{\text{f}} \end{array}$		24 133 16 37		ns	$\begin{split} &V_{CC}=15\text{V}, \text{ V}_{EE}=\text{-}5\text{V} \\ &V_{IN}=\text{-}5 \text{ to } 15\text{V} \\ &R_{IN}=1\text{k}\Omega \\ &C_{L}=10\text{nF}, R_{L}=0.18\Omega \\ &R_{SOURCE}=4.7\Omega, R_{SINK}=0\Omega \end{split}$

## **Switching Test Circuit and Timing Diagram**



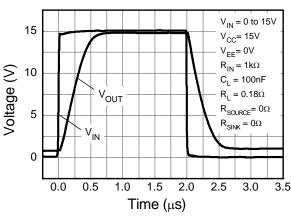


#### Typical Switching Characteristics @TA = 25°C unless otherwise specified



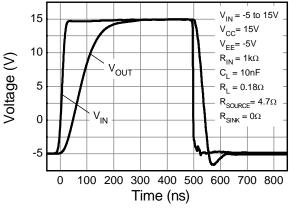
## Switching Speed

Low Load Capacitance  $C_I = 10nF$ 



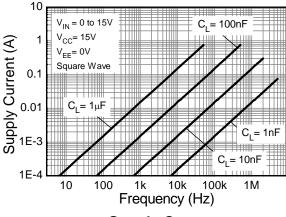
**Switching Speed** 

High Load Capacitance  $C_I = 100nF$ 



## Switching Speed

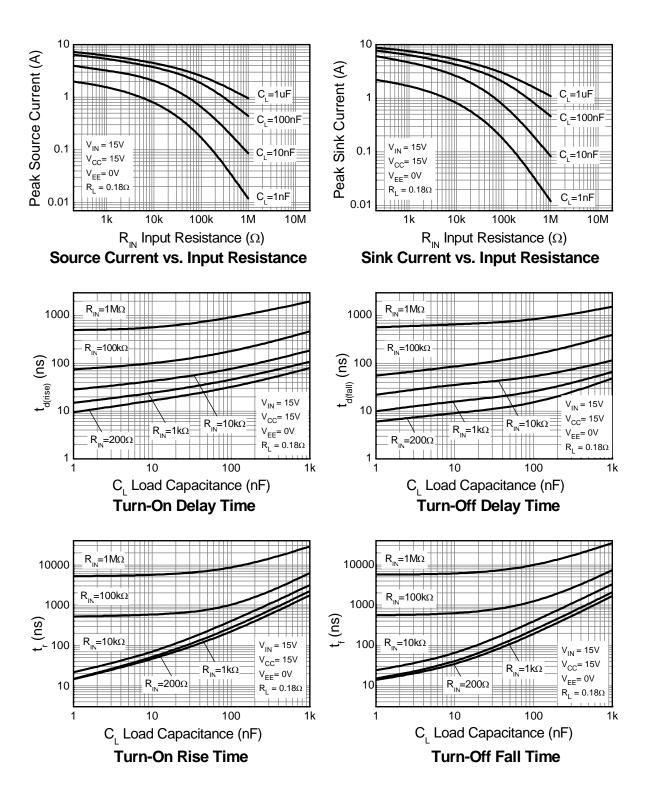
Asymmetric Source and Sink Resistors



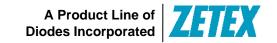
**Supply Current** 



### Typical Switching Characteristics @TA = 25°C unless otherwise specified





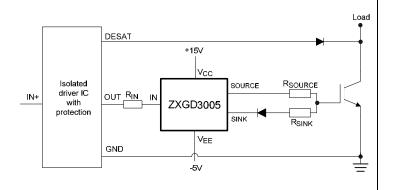


### **Application Notes**

#### Independent control of rise and fall time

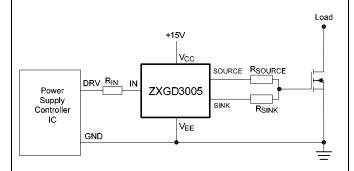
An application may require the turn-on (ton) and turn-off (toff) time to be independently controlled, which can be achieved by setting different R<sub>SOURCE</sub> and R<sub>SINK</sub> values. With asymmetric R<sub>SOURCE</sub> and R<sub>SINK</sub> resistors, then a potential difference will occur between the SOURCE and SINK pins during the switching transition. If the potential difference across the SOURCE and SINK pins is greater than 7.5V, then it could damage the ZXGD3005.

In this circuit example of driving an IGBT, a blocking diode is added in series with R<sub>SINK</sub> to protect against excess reverse current being induced into the SINK pin.

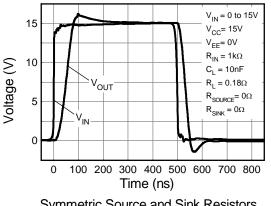


#### Circuit example of driving a MOSFET

Application example of gate driving a MOSFET from 0 to 15V with  $R_{SOURCE} = R_{SINK} = 0\Omega$ 



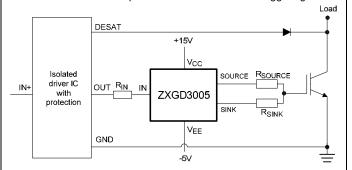
### **Switching Time Characteristic**



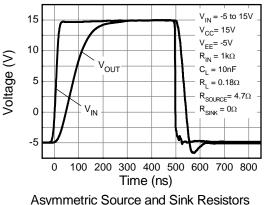
Symmetric Source and Sink Resistors

### Circuit example of driving an IGBT

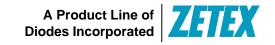
Application example of gate driving an IGBT with independent ton and toff using asymmetric R<sub>SOURCE</sub> and R<sub>SINK</sub> In addition, the gate is driven from -5 to +15V to prevent dV/dt induced false triggering.



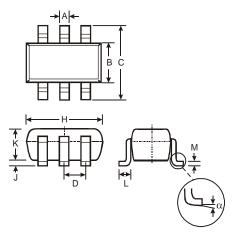
## **Switching Time Characteristic**





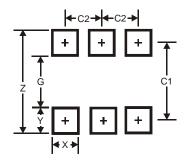


## **Package Outline Dimensions**



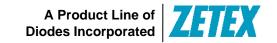
SOT26					
Dim	Min	Max	Тур		
Α	0.35	0.50	0.38		
В	1.50	1.70	1.60		
С	2.70	3.00	2.80		
D	_	_	0.95		
Н	2.90	3.10	3.00		
J	0.013	0.10	0.05		
K	1.00	1.30	1.10		
L	0.35	0.55	0.40		
M	0.10	0.20	0.15		
α	0°	8°			
All Dimensions in mm					

## **Suggested Pad Layout**



Dimensions	Value (in mm)
Z	3.20
G	1.60
Х	0.55
Y	0.80
C1	2.40
C2	0.95





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