Assignment 2 (ML for TS) - MVA 2023/2024

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1 Introduction

Objective. The goal is to better understand the properties of AR and MA processes, and do signal denoising with sparse coding.

Warning and advice.

- Use code from the tutorials as well as from other sources. Do not code yourself well-known procedures (e.g. cross validation or k-means), use an existing implementation.
- The associated notebook contains some hints and several helper functions.
- Be concise. Answers are not expected to be longer than a few sentences (omitting calculations).

Instructions.

- Fill in your names and emails at the top of the document.
- Hand in your report (one per pair of students) by Tuesday 5th December 11:59 PM.
- Rename your report and notebook as follows:
 FirstnameLastname1_FirstnameLastname1.pdf and
 FirstnameLastname2_FirstnameLastname2.ipynb.
 For instance, LaurentOudre_CharlesTruong.pdf.
- Upload your report (PDF file) and notebook (IPYNB file) using this link: docs.google.com/forms/d/e/1FAIpQLSfCqMXSDU9jZJbYUMmeLCXbVeckZYNiDpPl4hRUwcJ2cBHQM

2 General questions

A time series $\{y_t\}_t$ is a single realisation of a random process $\{Y_t\}_t$ defined on the probability space (Ω, \mathcal{F}, P) , i.e. $y_t = Y_t(w)$ for a given $w \in \Omega$. In classical statistics, several independent realisations are often needed to obtain a "good" estimate (meaning consistent) of the parameters of the process. However, thanks to a stationarity hypothesis and a "short-memory" hypothesis, it is still possible to make "good" estimates. The following question illustrates this fact.

Question 1

An estimator $\hat{\theta}_n$ is consistent if it converges in probability when the number n of samples grows to ∞ to the true value $\theta \in \mathbb{R}$ of a parameter, i.e. $\hat{\theta}_n \stackrel{\mathcal{D}}{\longrightarrow} \theta$.

- Recall the rate of convergence of the sample mean for i.i.d. random variables with finite variance.
- Let $\{Y_t\}_{t\geq 1}$ a wide-sense stationary process such that $\sum_k |\gamma(k)| < +\infty$. Show that the sample mean $\bar{Y}_n = (Y_1 + \cdots + Y_n)/n$ is consistent and enjoys the same rate of convergence as the i.i.d. case. (Hint: bound $\mathbb{E}[(\bar{Y}_n \mu)^2]$ with the $\gamma(k)$ and recall that convergence in L_2 implies convergence in probability.)

Answer 1

Let us consider iid random variables X_n with mean θ and finite variance σ^2 . The sample mean is none other than a simple Monte Carlo estimator $S_n = \frac{1}{n} \sum_{k=1}^n X_k$, which converges strongly to $\mathbb{E}(X) = \theta$ in virtue of the law of averages. The sample mean hence converges in probability to θ . Let us now evaluate the rate of this convergence. Let $\epsilon \geq 0$, then the Bienaymé-Tchebychev inequality gives us:

$$\mathbb{P}(|S_n - \theta| \ge \epsilon) \le \frac{\sigma^2}{n\epsilon^2}$$

Let us rewrite $\eta = \frac{\sigma^2}{n\epsilon^2}$, then, with probability $1 - \eta$, we have that:

$$|S_n - \theta| \le \frac{\sigma}{\sqrt{n\eta}}$$

Finally, we get a convergence rate of $\frac{1}{\sqrt{n}}$ for the sample mean.

We could obtain a similar result by exploiting directly the central limit theorem, which states that $\sqrt{n}(S_n - \theta)$ converges in law to the normal distribution.

Let us now show that the sample mean \bar{Y}_n is a consistent estimator.

$$(\bar{Y}_n - \theta)^2 = \frac{1}{n^2} \sum_{i=1}^n \sum_{k=-i+1}^{n-i} (Y_i - \theta)(Y_{k+i} - \theta)$$

$$= \frac{1}{n^2} \sum_{i=1}^n \sum_{k=-i+1}^{n-i} Y_i Y_{k+i} - \theta(Y_i + Y_{k+1}) + \theta^2$$

$$\mathbb{E}[(\bar{Y}_n - \theta)^2] = \frac{1}{n^2} \sum_{i=1}^n \sum_{k=-i+1}^{n-i} \gamma(k) - \theta^2$$

$$\leq \frac{1}{n^2} \sum_{i=1}^n \sum_{k=-i+1}^{n-i} \gamma(k)$$

$$\leq \frac{1}{n} \sum_{k=-n+1}^{n-1} \gamma(k)$$

$$\leq \frac{1}{n} \sum_{k=-\infty}^{+\infty} \gamma(k)$$

Indeed, we have seen that the $\gamma(k)$ family is summable. This proves that the quantity \bar{Y}_n converges (L2) to θ . Again, if we use Bienaymé-Tchebychev's inequality, we can show that for any $\epsilon \geq 0$:

$$\mathbb{P}(\left|\bar{Y}_n - \theta\right| \ge \epsilon) \le \frac{\mathbb{E}[(\bar{Y}_n - \theta)^2]}{\epsilon^2}$$

and we have just proven that:

$$\lim_{n=+\infty} \mathbb{E}[(\bar{Y}_n - \theta)^2] = 0$$

This proves that the estimator \bar{Y}_n converges in probability to θ and is consistent. .

3 AR and MA processes

Question 2 *Infinite order moving average MA*(∞)

Let $\{Y_t\}_{t\geq 0}$ be a random process defined by

$$Y_t = \varepsilon_t + \psi_1 \varepsilon_{t-1} + \psi_2 \varepsilon_{t-2} + \dots = \sum_{k=0}^{\infty} \psi_k \varepsilon_{t-k}$$
 (1)

where $(\psi_k)_{k\geq 0} \subset \mathbb{R}$ ($\psi=1$) are square summable, i.e. $\sum_k \psi_k^2 < \infty$ and $\{\varepsilon_t\}_t$ is a zero mean white noise of variance σ_{ε}^2 . (Here, the infinite sum of random variables is the limit in L_2 of the partial sums.)

- Derive $\mathbb{E}(Y_t)$ and $\mathbb{E}(Y_tY_{t-k})$. Is this process weakly stationary?
- Show that the power spectrum of $\{Y_t\}_t$ is $S(f) = \sigma_{\varepsilon}^2 |\phi(e^{-2\pi i f})|^2$ where $\phi(z) = \sum_j \psi_j z^j$. (Assume a sampling frequency of 1 Hz.)

The process $\{Y_t\}_t$ is a moving average of infinite order. Wold's theorem states that any weakly stationary process can be written as the sum of the deterministic process and a stochastic process which has the form (1).

Answer 2

Let us denote $Y_t^n = \sum_{k=0}^n \psi_k \varepsilon_{t-k}$. We will first show that Y_t^n converges to Y_t for the L2 norm. Let $n \in \mathbb{N}$ and $N \ge n+1$.

$$Y_t^N - Y_t^n = \sum_{k=n+1}^N \psi_k \varepsilon_{t-k}$$

$$(Y_t^N - Y_t^n)^2 = \sum_{k=n+1}^N \sum_{l=n+1}^N \psi_k \psi_l \varepsilon_{t-k} \varepsilon_{t-l}$$

$$\mathbb{E}(Y_t^N - Y_t^n)^2 = \sum_{k=n+1}^N \psi_k^2 \sigma_{\epsilon}^2$$

Indeed, ε_t is a Gaussian white noise with 0 mean so $\gamma(k) = \sigma_{\varepsilon}^2 \delta_0(k)$. We have also seen that the family ϕ_k is square summable, so the sum is definite when N goes to infinity. We can hence rewrite:

$$\mathbb{E}(Y_t - Y_t^n)^2 = \sum_{k=n+1}^{+\infty} \psi_k^2 \sigma_{\epsilon}^2$$

$$\lim_{n \to +\infty} \mathbb{E}(Y_t - Y_t^n)^2 = 0$$

Because the sum of ϕ is convergent. We have hence proven that Y_t^n converges to Y_t in L2. However, this convergence implies the L1 convergence (indeed, the Cauchy-Schwarz inequality gives us that:

$$\lim_{n\to+\infty}\mathbb{E}\left|Y_t-Y_t^n\right|=0$$

For all $n \in \mathbb{N}$, we have that:

$$\mathbb{E}(Y_t^n) = \sum_{k=0}^n \psi_k \mathbb{E}[\varepsilon_{t-k}] = 0$$

This means that:

$$\mathbb{E}(Y_t) = \lim_{n \to +\infty} \mathbb{E}(Y_t^n) = 0$$

Let us now compute $\mathbb{E}(Y_t Y_{t-k})$.

$$\begin{split} Y_{t}Y_{t-k} &= \sum_{i=0}^{+\infty} \psi_{i}\varepsilon_{t-i} \sum_{j=0}^{+\infty} \psi_{j}\varepsilon_{t-(j+k)} \\ &= \sum_{i=0}^{+\infty} \psi_{i}\varepsilon_{t-i} \sum_{l=k}^{+\infty} \psi_{l-k}\varepsilon_{t-l} \\ &= \sum_{i=k}^{+\infty} \sum_{l=k}^{+\infty} \psi_{i}\psi_{l-k}\varepsilon_{t-i}\varepsilon_{t-l} + \sum_{i=0}^{+k-1} \sum_{l=k}^{+\infty} \psi_{i}\psi_{l-k}\varepsilon_{t-i}\varepsilon_{t-l} \\ \mathbb{E}[Y_{t}Y_{t-k}] &= \sum_{i=k}^{+\infty} \psi_{i}\psi_{i-k}\mathbb{E}[\varepsilon_{t-i}^{2}] \\ \mathbb{E}[Y_{t}Y_{t-k}] &= \sigma_{\epsilon}^{2} \sum_{i=k}^{+\infty} \psi_{i}\psi_{i-k} = \sigma_{\epsilon}^{2} \sum_{i=0}^{+\infty} \psi_{i}\psi_{i+k} \end{split}$$

We can also show by reproducing the same calculation and splitting the sums in a similar way that $\gamma(k) = \gamma(-k)$, which finishes to prove that $\gamma(k)$ depends only on the absolute value of the difference of indexes between the two Y. This proves that the process Y_t is weakly stationary:

- For all t, $\mathbb{E}[Y_t] = 0$
- For all $t, k, \gamma(k) = \mathbb{E}[Y_t Y_{t-k}] = \sigma_{\epsilon}^2 \sum_{i=k}^{+\infty} \psi_i \psi_{i-k} = \sigma_{\epsilon}^2 \sum_{i=0}^{+\infty} \psi_i \psi_{i+k}$ is only a function of k
- $\gamma(-k) = \gamma(k)$
- $\mathbb{E}[Y_t + nY_{t+m}]$ is only a function of |m-n|

Let us now compute the spectrum of this signal, for which we take $f_s = 1$.

$$S(f) = \sum_{k=-\infty}^{+\infty} \gamma(k) e^{-2ik\pi f}$$

$$= \sum_{k=-\infty}^{+\infty} \sigma_{\epsilon}^{2} \sum_{i=0}^{+\infty} \psi_{i} \psi_{i+|k|} e^{-2ik\pi f}$$

$$= \sum_{k=1}^{+\infty} \sigma_{\epsilon}^{2} \sum_{i=0}^{+\infty} \psi_{i} \psi_{i+k} (e^{2ik\pi f} + e^{-2ik\pi f}) + \sigma_{\epsilon}^{2} \sum_{i=0}^{+\infty} \psi_{i}^{2}$$

Let us now take a look at:

$$\left| \phi(e^{-2\pi i f}) \right|^2 = \sum_{k=0}^{+\infty} \sum_{l=0}^{+\infty} \psi_k \psi_l z^{l-k}$$

$$= \sum_{l=0}^{+\infty} \sum_{k=1}^{+\infty} \psi_k \psi_{l+k} (z^k + z^{-k}) + \sum_{l=0}^{+\infty} \psi_l^2$$

With $z = e^{-2\pi i f}$, we finally conclude that:

$$S(f) = \sigma_{\varepsilon}^2 \left| \phi(e^{-2\pi i f}) \right|^2$$

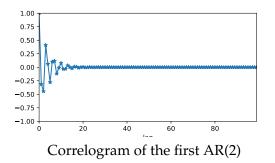
Question 3 *AR*(2) *process*

Let $\{Y_t\}_{t\geq 1}$ be an AR(2) process, i.e.

$$Y_t = \phi_1 Y_{t-1} + \phi_2 Y_{t-2} + \varepsilon_t \tag{2}$$

with $\phi_1, \phi_2 \in \mathbb{R}$. The associated characteristic polynomial is $\phi(z) := 1 - \phi_1 z - \phi_2 z^2$. Assume that ϕ has two distinct roots (possibly complex) r_1 and r_2 such that $|r_i| > 1$. Properties on the roots of this polynomial drive the behaviour of this process.

- Express the autocovariance coefficients $\gamma(\tau)$ using the roots r_1 and r_2 .
- Figure 1 shows the correlograms of two different AR(2) processes. Can you tell which one has complex roots and which one has real roots?
- Express the power spectrum S(f) (assume the sampling frequency is 1 Hz) using $\phi(\cdot)$.
- Choose ϕ_1 and ϕ_2 such that the characteristic polynomial has two complex conjugate roots of norm r=1.05 and phase $\theta=2\pi/6$. Simulate the process $\{Y_t\}_t$ (with n=2000) and display the signal and the periodogram (use a smooth estimator) on Figure 2. What do you observe?



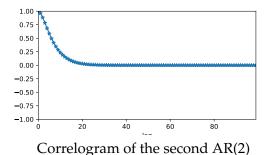


Figure 1: Two AR(2) processes

Answer 3

Expressing the autocovariance

First, note that the initial condition is $Y_1 = \varepsilon_1$. Since r_1 and r_2 are outside the unit circle, the process $\{Y_t\}$ is wide-sense stationary. Therefore, we have that the mean $\mu = \mathbb{E}[Y_t] = \mathbb{E}[Y_1] = \mathbb{E}[\varepsilon_1] = 0$.

Let us now compute the autocovariance function, for a delay $\tau \in \mathbb{N}$, we have

$$\begin{split} \gamma(\tau) &= \mathbb{E}[(Y_t - \mu)(Y_{t-\tau} - \mu)] = \mathbb{E}[Y_t Y_{t-\tau}] \\ &= \mathbb{E}[(\phi_1 Y_{t-1} + \phi_2 Y_{t-2} + \varepsilon_t) Y_{t-\tau}] \\ &= \phi_1 \mathbb{E}[Y_{t-1} Y_{t-\tau}] + \phi_2 \mathbb{E}[Y_{t-2} Y_{t-\tau}] + \mathbb{E}[\varepsilon_t Y_{t-\tau}] \\ &= \begin{cases} \phi_1 \gamma(\tau - 1) + \phi_2 \gamma(\tau - 2), \text{ for } \tau \neq 0, \\ \phi_1 \gamma(\tau - 1) + \phi_2 \gamma(\tau - 2) + \sigma^2, \text{ for } \tau = 0, \end{cases} \end{split}$$

where σ^2 is the variance of the white noise ε . Let us consider the following associated initial

conditions

$$\gamma(0) = \phi_1 \gamma(1) + \phi_2 \gamma(2) + \sigma^2,
\gamma(1) = \phi_1 \gamma(0) + \phi_2 \gamma(1),
\gamma(2) = \phi_1 \gamma(1) + \phi_2 \gamma(0)$$

Recall the known formulas on the sum and product of the roots of a second order polynomial, which yields $\phi_1 = r_1^{-1} + r_2^{-1}$ and $\phi_2 = -r_1^{-1}r_2^{-1}$. Thus, the above system becomes

$$\gamma(0) = (r_1^{-1} + r_2^{-1})\gamma(1) - r_1^{-1}r_2^{-1}\gamma(2) + \sigma^2,$$

$$\gamma(1) = (r_1^{-1} + r_2^{-1})\gamma(0) - r_1^{-1}r_2^{-1}\gamma(1),$$

$$\gamma(2) = (r_1^{-1} + r_2^{-1})\gamma(1) - r_1^{-1}r_2^{-1}\gamma(0).$$
(3)

In virtue of a result of a result [2] (Section 3.2.1), we look for solutions of the above system of the form

$$\gamma(\tau) = \alpha_1 r_1^{-\tau} + \alpha_2 r_2^{-\tau}, \quad \tau \in \mathbb{N},\tag{4}$$

where α_1 , α_2 are arbitrary constants. Plugging (4) into (3), we get

$$\begin{split} \alpha_1 + \alpha_2 &= (r_1^{-1} + r_2^{-1})(\alpha_1 r_1^{-1} + \alpha_2 r_2^{-1}) - r_1^{-1} r_2^{-1}(\alpha_1 r_1^{-2} + \alpha_2 r_2^{-2}) + \sigma^2, \\ \alpha_1 r_1^{-1} + \alpha_2 r_2^{-1} &= (r_1^{-1} + r_2^{-1})(\alpha_1 + \alpha_2) - r_1^{-1} r_2^{-1}(\alpha_1 r_1^{-1} + \alpha_2 r_2^{-1}), \\ (\alpha_1 r_1^{-2} + \alpha_2 r_2^{-2}) &= (r_1^{-1} + r_2^{-1})(\alpha_1 r_1^{-1} + \alpha_2 r_2^{-1}) - r_1^{-1} r_2^{-1}(\alpha_1 + \alpha_2). \end{split}$$

By linear combination and substitution, we solve this system with only 2 unknowns (α_1 , α_2), and get

$$\gamma(\tau) = \frac{\sigma^2 r_1^2 r_2^2}{(r_1 r_2 - 1)(r_2 - r_1)} [(r_1^2 - 1)^{-1} r_1^{1 - \tau} - (r_2^2 - 1)^{-1} r_2^{1 - \tau}]. \tag{5}$$

Identifying the correlogram depending on the roots of the characteristic polynomial

From [2] again, we get that, if $r_1 = re^{i\theta}$ and $r_2 = re^{-i\theta}$, $0 < \theta < \pi$, we can write (5) as

$$\gamma(\tau) = \frac{\sigma^2 r^{4-\tau} \sin(\tau \theta + \psi)}{(r^2 - 1)(r^4 - 2r^2 \cos(2\theta) + 1)\sin(\theta)}.$$
 (6)

where

$$\tan(\psi) = \frac{r^2 + 1}{r^2 - 1} \tan(\theta) \tag{7}$$

and $\cos \psi$ has the same sign as $\cos \theta$. Therefore in this case of complex roots, γ has the form of a damped sinusoid with damping factor r^{-1} and period $2\pi/\theta$. If the roots are close to the unit circle, then r is close to 1, the damping is slow, and we obtain a nearly sinusoidal autocovariance function.

Since an (AR(p)) process is wide-sense stationary, we compute the autocorrelation function ρ of such a process from its autocovariance function γ as

$$\rho(\tau) = \frac{\gamma(\tau)}{\gamma(0)}.\tag{8}$$

Thus, the correlogram is just a rescaling of the plot of the autocovariance function.

Therefore, since the correlogram of the second AR(2) is not a sinusoid, we know it has real roots. Thus, the first AR(2) has to be the one with complex roots.

Expressing the power spectral density

The power spectral density is defined for wide-sense stationary processes as the Fourier transform of the autocovariance function. Thus, the power spectral density of the AR(2) process Y with sampling frequency $f_s = 1$ Hz is

$$S(f) = \sum_{k=-\infty}^{+\infty} \gamma(k) e^{-2ik\pi f}$$

$$= \frac{\sigma^2 r_1^2 r_2^2}{(r_1 r_2 - 1)(r_2 - r_1)} \sum_{k=-\infty}^{+\infty} [(r_1^2 - 1)^{-1} r_1^{1-|k|} - (r_2^2 - 1)^{-1} r_2^{1-|k|}] e^{-2ik\pi f}$$

$$= \frac{\sigma^2 r_1^2 r_2^2}{(r_1 r_2 - 1)(r_2 - r_1)} \left[(r_1^2 - 1)^{-1} r_1 \sum_{k=-\infty}^{+\infty} r_1^{-|k|} e^{-2i\pi f k} - (r_2^2 - 1)^{-1} r_2 \sum_{k=-\infty}^{+\infty} r_2^{-|k|} e^{-2i\pi f k} \right].$$
(9)

Let us focus on the first term, where the geometric series equals

$$\sum_{k=-\infty}^{+\infty} r_1^{-|k|} e^{-2i\pi fk} = 1 + \sum_{k=1}^{+\infty} \frac{\left[e^{-2i\pi fk} + e^{2i\pi fk}\right]}{r_1^k} = 1 + \frac{e^{-2i\pi f}}{r_1 \left(1 - \frac{e^{-2i\pi f}}{r_1}\right)} + \frac{e^{2i\pi f}}{r_1 \left(1 - \frac{e^{2i\pi f}}{r_1}\right)}$$

$$= 1 + \frac{e^{-2i\pi f}}{r_1 - e^{-2i\pi f}} + \frac{e^{2i\pi f}}{r_1 - e^{2i\pi f}} = 1 + \frac{e^{-2i\pi f}r_1 - 1 + e^{2i\pi f}r_1 - 1}{r_1^2 - r_1 e^{-2i\pi f} - e^{2i\pi f}r_1 + 1}.$$
(10)

Similarly, we get

$$\sum_{k=-\infty}^{+\infty} r_2^{-|k|} e^{-2i\pi fk} = 1 + \frac{e^{-2i\pi f} r_2 - 1 + e^{2i\pi f} r_2 - 1}{r_2^2 - r_2 e^{-2i\pi f} - e^{2i\pi f} r_2 + 1}.$$
 (11)

On the other hand, we have $\phi(z) = -\phi_2(z - r_1)(z - r_2)$ so

$$|\phi(e^{-2i\pi f})|^2 = \phi_2^2(e^{-2i\pi f} - r_1)(e^{-2i\pi f} - r_2)(e^{2i\pi f} - r_1)(e^{2i\pi f} - r_2)$$
(12)

which, using $\phi_2 = -r_1^{-1}r_2^{-1}$, yields

$$|\phi(e^{-2i\pi f})|^2 = r_1^{-2} r_2^{-2} (e^{-2i\pi f} - r_1)(e^{-2i\pi f} - r_2)(e^{2i\pi f} - r_1)(e^{2i\pi f} - r_2)$$
(13)

and therefore

$$\frac{\sigma^{2}}{|\phi(e^{-2\pi if})|^{2}} = \sigma^{2} r_{1}^{2} r_{2}^{2} \left[(e^{-2i\pi f} - r_{1})(e^{-2i\pi f} - r_{2})(e^{2i\pi f} - r_{1})(e^{2i\pi f} - r_{2}) \right]^{-1}
= \sigma^{2} r_{1}^{2} r_{2}^{2} \times
\left[1 - (r_{2} + r_{1})e^{-2i\pi f} + r_{1}r_{2}e^{-4i\pi f} - r_{2}e^{2i\pi f} + r_{2}^{2} + r_{1}r_{2} - r_{2}^{2}r_{1}e^{-2i\pi f} +
-r_{1}e^{2i\pi f} + r_{1}r_{2} + r_{1}^{2} - r_{1}^{2}r_{2}e^{-2i\pi f} + r_{1}r_{2}e^{4i\pi f} - r_{1}r_{2}^{2}e^{2i\pi f} - r_{1}^{2}r_{2}e^{2i\pi f} + r_{1}^{2}r_{2}^{2} \right]^{-1}$$
(14)

Plugging (10) and (11) in (9), and subsquently putting at the same denominator, the result equals (14). Thus, we conclude

$$S(f) = \frac{\sigma^2}{|\phi(e^{-2\pi i f})|^2}$$
 (15)

Plotting the periodogram in the case of complex roots

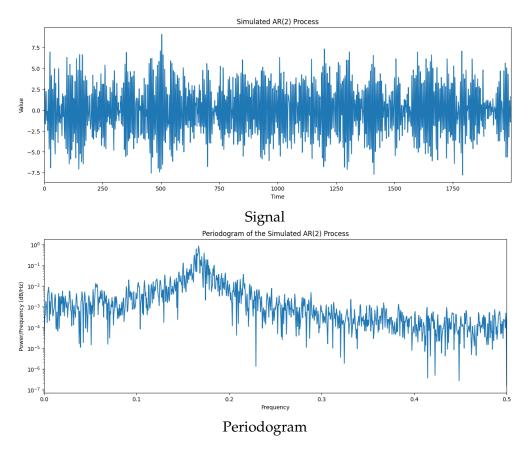


Figure 2: AR(2) process

For the last part of the question, we simulate an AR(2) process and plot its periodogram. One can see the plots in 2. To have complex conjuguate roots equal to $r_1 = re^{-i\pi/3}$, $r_2 = re^{i\pi/3}$, we used $\phi_1 = r_1^{-1} + r_2^{-1} = \frac{2}{r}\cos(\pi/3) = r^{-1}$ and $\phi_2 = -r_1^{-1}r_2^{-1} = -r^{-2}$.

One can see a peak in the periodogram. We printed the frequency corresponding to this peak and obtained The frequency associated to the peak on the spectrum is 0.167. This is expected because we have complex conjuguate roots of ϕ and, as said above in this answer, this corresponds to an autocovariance function being a damped sinusoid of frequency $\frac{\theta}{2\pi}$. Here, $\theta=\pi/3$ so the frequency of the sinusoid is $\frac{\pi}{6\pi}=\frac{1}{6}=0.1666\ldots$ Since the periodogram is the Fourier transform of the autocovariance function, it should have a peak at the fundamental frequency of the autocovariance function, which is precisely $\frac{1}{6}$. This is indeed the case.

4 Sparse coding

The modulated discrete cosine transform (MDCT) is a signal transformation often used in sound processing applications (for instance to encode a MP3 file). A MDCT atom $\phi_{L,k}$ is defined for a length 2L and a frequency localisation k (k = 0, ..., L - 1) by

$$\forall u = 0, \dots, 2L - 1, \quad \phi_{L,k}[u] = w_L[u] \sqrt{\frac{2}{L}} \cos\left[\frac{\pi}{L} \left(u + \frac{L+1}{2}\right) (k + \frac{1}{2})\right]$$
 (16)

where w_L is a modulating window given by

$$w_L[u] = \sin\left[\frac{\pi}{2L}\left(u + \frac{1}{2}\right)\right]. \tag{17}$$

Question 4 Sparse coding with OMP

For the signal provided in the notebook, learn a sparse representation with MDCT atoms. The dictionary is defined as the concatenation of all shifted MDCDT atoms for scales L in [32,64,128,256,512,1024].

- For the sparse coding, implement the Orthogonal Matching Pursuit (OMP). (Use convolutions to compute the correlations coefficients.)
- Display the norm of the successive residuals and the reconstructed signal with 10 atoms.

Answer 4

Contrary to the previous assignment, we are now looking for sparse coding that is not convolutional, and with l_0 -regularization (where we constrain the activation vectors to have K_0 non-null entries). In our case, this yields

$$\min_{(\mathbf{z}_{k})_{k}, \|\mathbf{z}\|_{0} = K_{0}} \left\| \mathbf{x} - \sum_{L \in \mathcal{L}} \sum_{k=0}^{L-1} \mathbf{z}_{L,k} \phi_{L,k} \right\|_{2}^{2} , \qquad (18)$$

where $\mathbf{z}_{L,k} \in \mathbb{R}$ are activations signals, with $\mathcal{L} = \{32,64,128,256,512,1024\}$. Thus, the size of the learnable activation vector \mathbf{z} is $MDCT(\mathcal{L})$ with $MDCT(\mathcal{L}) = \sum_{L \in \mathcal{L}} \sum_{k=0}^{L-1} 1 = 1008$. Denote $K = MDCT(\mathcal{L})$ in what follows.

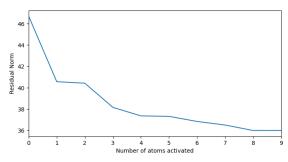
In the signal given in the notebook, we have a number of samples N = 2048 and we are asked to use OMP with a sparsity constraint of $K_0 = 10$.

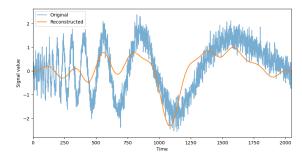
For OMP algorithm we take inspiration from the pseudo-code in [1] to produce our own Algorithm 1. We implement this pseudo-code and obtain Figure 3. We test what the OMP algorithm outputs for 50 atoms and display it in 4.

Algorithm 1: OMP for MDCT

```
Data: Time series \mathbf{x} \in \mathbb{R}^N
Dictionary \mathbf{D} \in \mathbb{R}^{N \times K}, recall K = MDCT(\mathcal{L})
Number of non-null coefficients K_0
Initial support (index set) \mathcal{S} = \emptyset
Result: Activation vector \mathbf{z} \in \mathbb{R}^K with K_0 non-null entries Initialisation
```

```
\mathbf{z} \leftarrow \mathbf{0}_K;
                                                                                                                           /* Initial solution */
\mathbf{r} \leftarrow \mathbf{x};
                                                                                                                           /* Initial residual */
                                                                                                                                          /* Iteration */
it \leftarrow 1;
while it \leq K_0 do
     \forall j \in \{1, \dots, K\}, t_j \leftarrow \frac{(\mathbf{D}_j^\top \mathbf{r}^{(it-1)})^2}{\|\mathbf{D}_j\|^2};
idx \leftarrow \arg\max_{k \notin \mathcal{S}} t_k;
                                                                                                                  /* Compute correlations */
                                                                /* Find an atom in D with maximum correlations */
                                                                                                                               /* Support update */
     \mathbf{z}(\mathcal{S}) \leftarrow \mathbf{D}(\mathcal{S})^+ \mathbf{x} = \arg\min_{\mathbf{z}} \|\mathbf{x} - \mathbf{D}(\mathcal{S})\mathbf{z}\|_2^2; /* Sparse coefficient update */
     \mathbf{r}^{(it)} = \mathbf{x} - \mathbf{D}(\mathcal{S})\mathbf{z}(\mathcal{S});
                                                                                                                             /* Residual update */
      it \leftarrow it + 1;
end
```

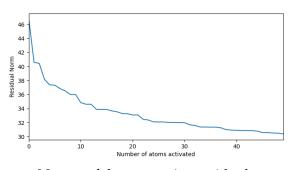


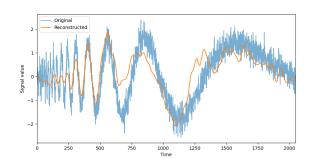


Norms of the successive residuals

Reconstruction with 10 atoms

Figure 3: Question 4





Norms of the successive residuals

Reconstruction with 50 atoms

Figure 4: Question 4

References

- [1] Ilkay Oksuz. Chapter 5 dictionary learning for medical image synthesis. In Ninon Burgos and David Svoboda, editors, *Biomedical Image Synthesis and Simulation*, The MICCAI Society book Series, pages 79–89. Academic Press, 2022.
- [2] Richard A. Davis Peter J. Brockwell. *Introduction to Time Series and Forecasting, Third Edition*. Springer, 2016.