EE2011 Engineering ElectromagneticsSemester II of Academic Year 2011/2012

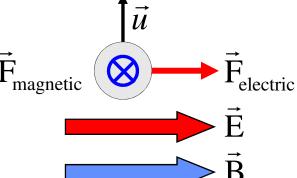
Prof Yeo Swee Ping eleyeosp@nus.edu.sg

Lorentz Force

force on charge q moving with velocity \vec{u}

$$\vec{F} = q\vec{E} + q\vec{u} \times \vec{B}$$

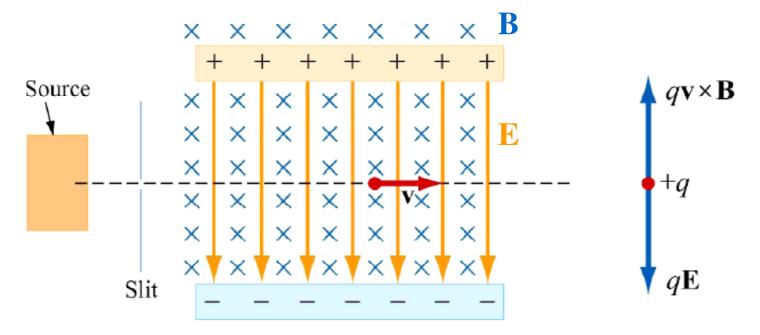
notes for $\vec{F}_{magnetic}$:



- (a) normal to $\vec{u} \Rightarrow$ no work done circular trajectory with constant kinetic energy
- (b) zero for stationary charge (i.e when $|\vec{u}| = 0$) no magnetic force in electrostatics
- (c) independent of $\vec{F}_{electric}$
- (d) used for defining Tesla = $\frac{N}{C \text{ (m/s)}} = \frac{N}{(C/s) \text{ m}} = \frac{N}{A \text{ m}}$
- (e) applicable to point charges but can extend to charge systems

Lorentz Force

force on electron flow in vacuum (Thomson's experiment)



adjust E and B for electrons to flow in straight path \rightarrow $V = \frac{E}{B}$ used for measuring velocity and charge-to-mass ratio of electron

Magnetic Force

force on electron flow in wire (i.e. current)

$$d\vec{F} = \vec{u} \times \vec{B} dq$$

$$= \vec{u} \times \vec{B} \sigma dV$$

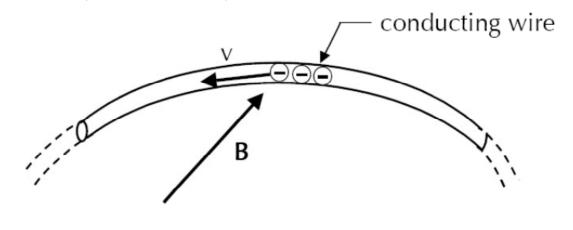
$$= \vec{J} \times \vec{B} A ds$$

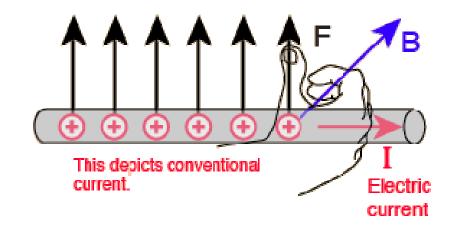
$$= I d\vec{s} \times \vec{B}$$

$$\Rightarrow \vec{F} = I \int d\vec{s} \times \vec{B}$$

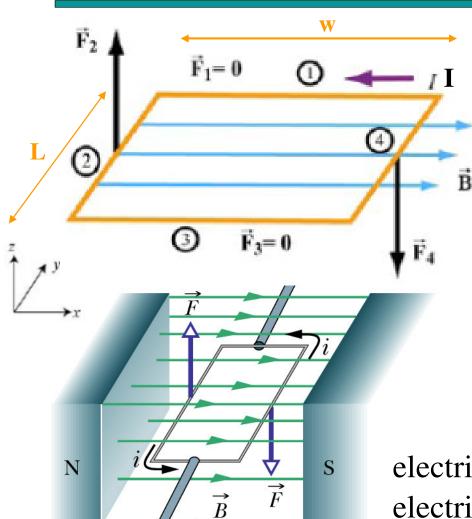
force on straight wire (where $d\vec{s} = \hat{u} ds$)

$$\vec{F} = IL\hat{u} \times \vec{B}$$





Magnetic Force



no forces on arms 1 and 3

no net force from arms 2 and 4

because
$$\left| \vec{F}_2 \right| = \left| \vec{F}_4 \right| = \left| \vec{B} \right| IL$$

resultant torque on wire loop

$$\tau = |\vec{B}| ILw = |\vec{B}| IA$$

torque vector on any wire loop

$$\vec{\tau} = I\vec{A} \times \vec{B} = \vec{m} \times \vec{B}$$

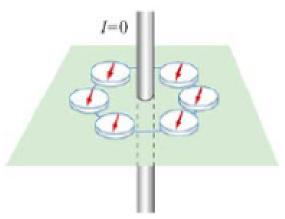
define magnetic moment $\vec{m} = IA$

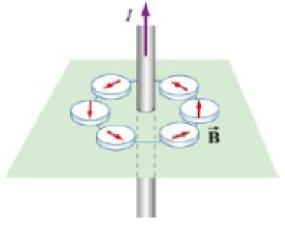
electric motor — transducer converting electrical energy into mechanical energy

magnetic field around current-carrying wire(s)









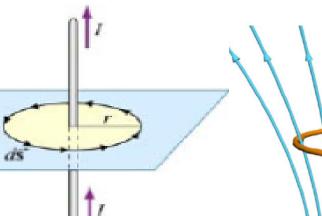
Ampere's Law

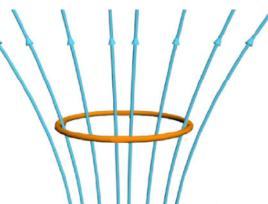
$$\oint \vec{B} \bullet d\vec{s} \, \propto \, \mathbf{I}$$

$$\oint \vec{B} \bullet d\vec{s} = \mu_0 \sum I_k$$

$$\oint \vec{B} \cdot d\vec{s} = \mu_0 \sum_k I_k$$

$$\oint \vec{B} \cdot d\vec{s} = \mu_0 \iint \vec{J} \cdot d\vec{A}$$





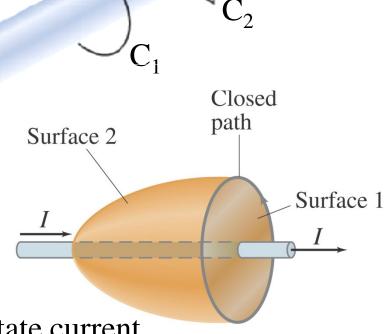
notes on Ampere's Law:

- (a) valid only for DC (*i.e.* when $\frac{\partial}{\partial t} = 0$)
- (b) valid for any closed contour
- (c) valid for non-planar surfaces
- (d) practical utility for cases with symmetry

$$\iint \nabla \times \vec{B} \cdot d\vec{A} = \mu_0 \iint \vec{J} \cdot d\vec{A}$$
$$\Rightarrow \nabla \times \vec{B} = \mu_0 \vec{J}$$

take div of both sides:

$$\nabla \bullet \vec{J} = \frac{1}{\mu_0} \nabla \bullet \nabla \times \vec{B} = 0 \longrightarrow \text{steady-state current}$$



no isolated magnetic charge c.f. Gauss's Law for electric charges

$$\iint_{S} \vec{B} \cdot d\vec{A} = 0 \qquad \qquad \iint_{S} \vec{D} \cdot d\vec{A} = Q_{\text{total in enclosure}}$$

$$\Rightarrow \iiint_{V} \nabla \cdot \vec{B} \, dV = 0 \text{ for application to any volume}$$

PDE version $\nabla \cdot \vec{B} = 0$ for application to particular point implications:

- (a) \vec{B} fields always in closed loops (due to lack of sources/sinks)
- (b) basis for introducing magnetic potential (vector) use null identity $\nabla \cdot (\nabla \times \vec{A}) = 0$ useful to express $\vec{B} = \nabla \times \vec{A}$ during analytical derivation but not common to leave answer in magnetic-potential form

re-visit curl equation in \vec{B} (from Ampere's Law)

PDE in
$$\vec{A}$$

$$\nabla \times (\nabla \times \vec{A}) = \mu_0 \vec{J}$$

expanding curl curl
$$\nabla \cdot \vec{A} - \nabla^2 \vec{A} = \mu_0 \vec{J}$$

choosing
$$\nabla \cdot \vec{A} = 0$$

$$\nabla^2 \vec{A} = -\mu_0 \vec{J}$$

decomposing vector Poisson's equation in (x, y, z) coordinates

$$\begin{array}{c} \nabla^2 A_x = -\mu_0 J_x \\ \nabla^2 A_y = -\mu_0 J_y \\ \nabla^2 A_z = -\mu_0 J_z \end{array} \right\} \ c.f. \ \ \nabla^2 V = -\frac{1}{\varepsilon_0} \sigma \ \rightarrow \ V = \frac{1}{4\pi\varepsilon_0} \iiint \frac{\sigma}{r''} dV'$$

$$\Rightarrow$$
 $\vec{A} = \frac{\mu_0}{4\pi} \iiint \frac{1}{r''} \vec{J} \ dV'$ and thereafter obtain $\vec{B} = \nabla \times \vec{A}$

short wire (carrying current I) with length $\Delta s \rightarrow 0$ in the limit

$$\vec{J} \, dV' = \vec{J} \, A \, ds' = I \, d\vec{s}'$$

$$\vec{A} = \frac{\mu_o}{4\pi} \iiint \frac{1}{r''} \vec{J} \, dV = \frac{\mu_o I}{4\pi} \int \frac{1}{r''} \, d\vec{s}'$$

$$\vec{B} = \nabla \times \left(\frac{\mu_o I}{4\pi} \int \frac{1}{r''} \, d\vec{s}'\right) = \frac{\mu_o I}{4\pi} \int \nabla \times \left(\frac{1}{r''} \, d\vec{s}'\right)$$

$$= \frac{\mu_o I}{4\pi} \int \frac{d\vec{s}' \times \hat{u}_{r''}}{(r'')^2}$$

Biot-Savart's Law to evaluate \vec{B} from wire (for $d\vec{s}$ with $\theta \neq 0$) check for solenoidal property

$$\nabla \bullet \vec{B} = \frac{\mu_{o} I}{4\pi} \int \nabla \bullet \left(\frac{d\vec{s}' \times \hat{\mathbf{u}}_{r''}}{(r'')^{2}} \right) = -\frac{\mu_{o} I}{4\pi} \int d\vec{s}' \bullet \nabla \times \nabla \left(\frac{1}{r''} \right) = 0$$

Example #1A: straight wire of length L (with origin at center)

• x-y plane normal to $d\vec{s} = dz \hat{k}$ and \vec{r}

• dB in same direction for all dz elements

• r and θ dependent on z

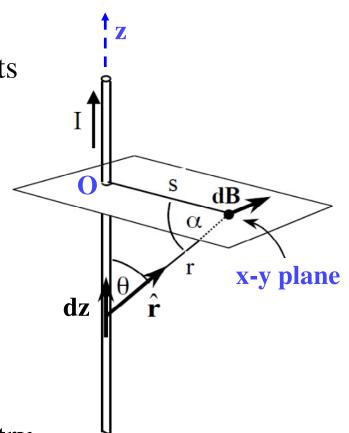
apply Biot-Savart's Law

$$\vec{B} = \frac{\mu_o I}{4\pi} \int_{\frac{L}{2}}^{\frac{L}{2}} \frac{dz \, \hat{k} \times \hat{r}}{r^2} = \frac{\mu_o I}{4\pi} \, \hat{u}_{\phi} \int_{\frac{L}{2}}^{\frac{L}{2}} \frac{\sin\theta \, dz}{z^2 + s^2}$$

$$= \frac{\mu_o I}{4\pi s} \frac{L}{\sqrt{\frac{1}{4}L^2 + s^2}} \, \hat{u}_{\phi}$$

$$\rightarrow \frac{\mu_o I}{4\pi s} \, \hat{u}_{\phi} \quad \text{when } L \rightarrow \infty$$

useful tool for structures without symmetry



Example #1B: long straight wire of radius R

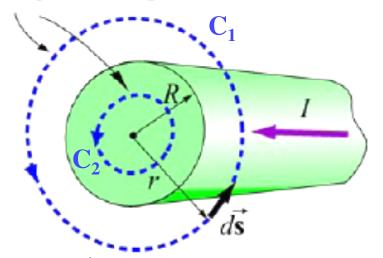
infer from symmetry:

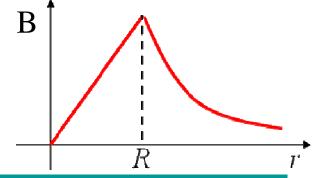
- azimuthal component $(\vec{B} = B_{\phi} \hat{u}_{\phi})$
- B_{ϕ} constant of ϕ and z coordinates
- (a) apply Ampere's Law to C₁ loop: $2\pi r B_{\phi} = \mu_{o} I \implies B_{\phi} = \frac{\mu_{o} I}{2\pi r}$
- (b) apply Ampere's Law to C₂ loop:

$$2\pi r B_{\phi} = \mu_{o} \frac{\pi r^{2}}{\pi R^{2}} I \implies B_{\phi} = \frac{\mu_{o} r I}{2\pi R^{2}}$$
 valid only for DC (*i.e.* $\frac{dI}{dt} = 0$)

I flowing within skin-depth for RF

Amperian loops





Example #2: long solenoid (with *n* turns of wire per meter)

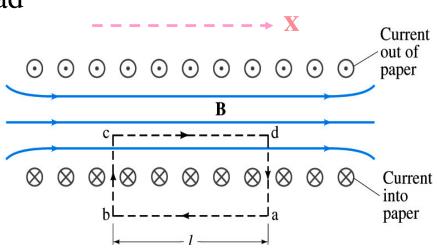
infer from structure:

- axial component $(\vec{B} = B_x \hat{u}_x)$
- uniform B within solenoid

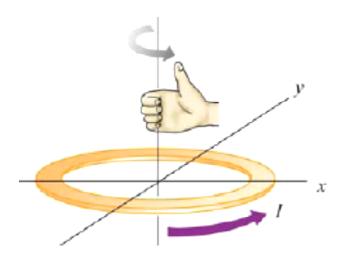
apply Ampere's Law to abcd:

- (a) no contributions from bc and ad
- (b) $B \approx 0$ outside solenoid

$$\begin{cases} \oint \vec{B} \cdot d\vec{s} = B_x l + 0 + 0 + 0 \\ \mu_0 \sum_k I_k = \mu_0 (nl) I \\ \Rightarrow \vec{B} = \mu_0 n I \hat{u}_x \end{cases}$$



Example #3: small loop (magnetic dipole with $\vec{m} = \pi r_o^2 I \hat{u}_z$)



$$\vec{A} = \frac{\mu_o I}{4\pi} \int \frac{1}{r''} d\vec{s}' = \frac{\mu_o}{4\pi r^2} \vec{m} \times \hat{u}_r \text{ when } r >> r_o$$

$$\vec{B} = \nabla \times \vec{A} = \frac{\mu_o}{4\pi} \frac{m}{r^3} \left(2\cos\theta \,\hat{\mathbf{u}}_r + \sin\theta \,\hat{\mathbf{u}}_\theta \right)$$

similarity with electric dipole (duality)

$$\vec{E} = \frac{1}{4\pi\varepsilon_0} \frac{p}{r^3} \left(2\cos\theta \,\hat{\mathbf{u}}_r + \sin\theta \,\hat{\mathbf{u}}_\theta \right)$$

extend to electron in orbit with $r_o \approx 0.5 \text{ x } 10^{-10} \text{ m}$ travel time for one orbit $\Delta t \approx 1.5 \text{ x } 10^{-16} \text{ s}$

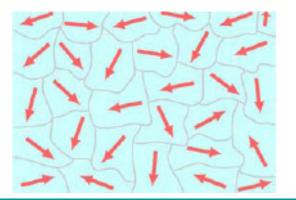
:. current
$$I = \frac{e}{\Delta t} \approx 1 \text{ mA}$$

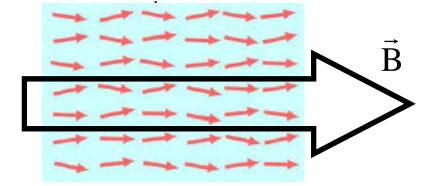
moment $m = \pi r_o^2 I \approx 9.3 \text{ x } 10^{-24} \text{ A m}^2$

field-matter interactions in magnetic material

- (a) vectorial (instead of scalar in dielectrics)
- (b) very weak for large majority of materials very strong in certain materials (*e.g.* Fe, Co, Ni)

not possible to consider individual magnetic dipoles need to consider average density instead for bulk behaviour define magnetization vector $\vec{M} = \frac{1}{\text{volume}} \sum \vec{m}_{k}$





Ampere's Law in differential form

$$\nabla \times \vec{B} = \mu_0 \vec{J}$$

 $\nabla \times \vec{B} = \mu_0 \vec{J}$ for free space

add magnetization current density $\vec{J}_m = \nabla \times \vec{M}$

$$\nabla \times \vec{B} = \mu_0 (\vec{J} + \vec{J}_m)$$
 for magnetic material

$$\Rightarrow \nabla \times \left(\frac{1}{\mu_0} \vec{B} - \vec{M}\right) = \vec{J}$$



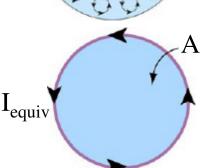
 $H = \text{magnetic field intensity vector } (A \,\text{m}^{-1}) \text{ to account for currents}$

$$\vec{B} = \mu_0 (\vec{H} + \vec{M}) = \text{magnetic flux density vector}$$

$$= \mu_0 (1 + \chi_m) \vec{H} = \mu_0 \mu_r \vec{H}$$

 μ_r relative permeability

 χ_m magnetic susceptibility

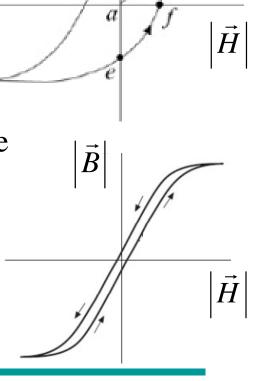


hysteresis behavior

- a: initially unmagnetized state
- b: increasing magnetization till onset of saturation
- c: retention after reducing source to zero
- d: reversing polarity of source to saturation
- e: retention with reversed polarity
- f: not coincident with initial unmagnetized state

implications

- non-linear behavior
- large retentivity for permanent magnets
- reduced hysteresis loop for memory core



relative permeability values:

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1.00 for air, water, copper, gold, aluminium, water, etc
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250 for cobalt
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> 5,000 for iron
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> 50,000 for MuMetal® at DC

> 35,000 for MuMetal® at 50 Hz with low magnetization

200,000 for MuMetal® at 50 Hz with high magnetization

- equal either to 1 (for most materials) or very large value
- variable due to hysteresis (check operating conditions)
- anisotropic (use tensor instead of scalar representation)

dual of electric boundary conditions at interface between materials:

(a) from "Gauss's Law" applied to pill-box structure

$$(B_1)_n = (B_2)_n$$

(b) from Ampere's Law applied to closed contour $(H_1)_t = (H_2)_t$ valid for non-conducting materials

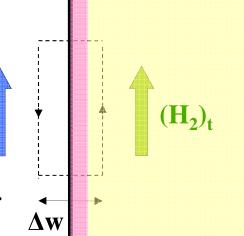
variation for perfect conductor (Region #2)

due to large current densities in RHS

$$\oint \vec{H} \cdot d\vec{s} = \iint \vec{J} \cdot d\vec{A}$$

$$(H_2)_t - (H_1)_t = (J_2)_s$$

linear current density $J_s(Am^{-1})$ defined later



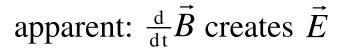
conservative \vec{E} only for electrostatics (*i.e.* when $\frac{\partial}{\partial t} = 0$)

Faraday's Law
$$V_{emf} = -\frac{d}{dt}\Phi$$

$$\oint \vec{E} \cdot d\vec{s} = -\frac{d}{dt} \iint \vec{B} \cdot d\vec{A}$$

$$\iint \nabla \times \vec{E} \bullet d\vec{A} = -\iint_{\frac{d}{dt}} \vec{B} \bullet d\vec{A}$$

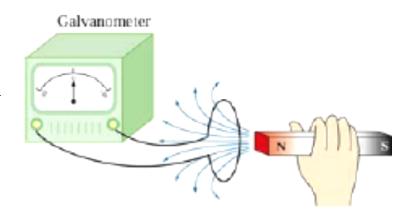
$$\nabla \times \vec{E} = -\frac{\mathrm{d}}{\mathrm{dt}} \vec{B}$$

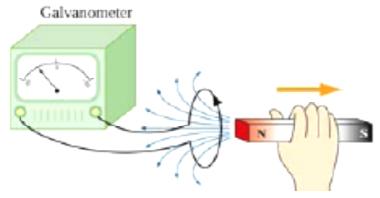


actual: $\frac{d}{dt}\vec{E}$ creates \vec{B} in turn

→ both inter-coupled

(together with propagation-velocity vector)





rate of change of flux $\Phi = \vec{B} \cdot \vec{A}$

Faraday's Law
$$V_{emf} = -\frac{d}{dt}(B_n A)$$

$$= -\frac{dB_n}{dt}A - B_n \frac{dA}{dt}$$
 for time-varying area

example: rod moving along wire rails

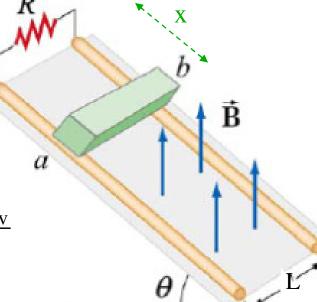
with
$$\frac{dB_n}{dt} = 0$$

and
$$\frac{dA}{dt} = \frac{d(Lx)}{dt} = L\frac{dx}{dt} = Lv$$

induced emf between ab and R

current in
$$R$$
 $I = \frac{V_{emf}}{R} = -\frac{B}{R} \frac{dA}{dt} = -\frac{BLv}{R}$

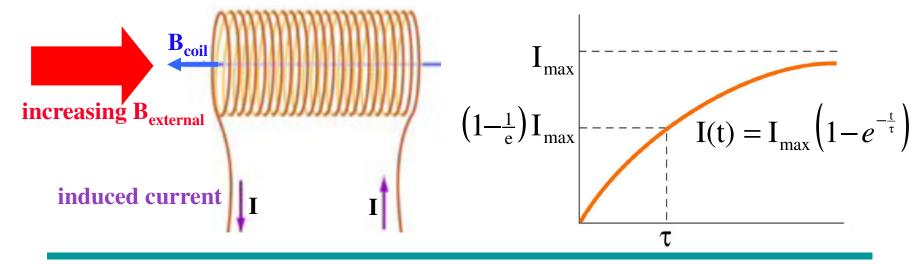
(note: rod accelerating with $\frac{dv}{dt} = g \sin \theta$)



Lenz's Law (i.e. negative sign in Faraday's Law)

- change of flux \rightarrow induced voltage $V_{emf} = -\frac{d}{dt}\Phi$
- current induced by back emf $\rightarrow B$ to counter change in Φ (*i.e.* slow down rate of increase/decrease of Φ)

illustration: increasing Φ for solenoid (inductor)



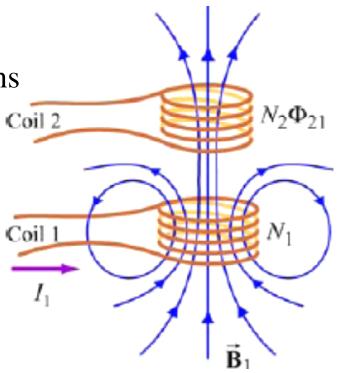
mutual inductance between loops

illustration: two coils with N_1 and N_2 turns

- magnetic field from I_1 flow in coil 1
- magnetic flux linkage with coil 2 induced emf $V_2 = -\frac{d}{dt} (N_2 \Phi_{21})$ direction determined by Lenz's Law

definition: mutual inductance $M = \frac{N_2 \Phi_{21}}{I_1}$ (a) unit of Henry $\left(H = \frac{T m^2}{A} = \frac{V s}{A}\right)$

- (b) same result if current in other loop $M = \frac{N_1 \Phi_{12}}{I_2}$
- (c) disregard N for single-turn loops
- (d) use matrix M for circuit system with more than two loops



mutual-inductance example: coil wrapped around solenoid recommend current in solenoid (instead of coil)

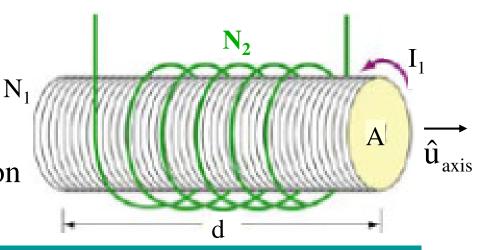
$$\rightarrow$$
 magnetic field within solenoid $\vec{B}_1 = \frac{\mu_o N_1 I_1}{d} \hat{u}_{axis}$

$$\rightarrow$$
 flux linkage with coil $\Phi_{21} = (\vec{B}_1 \cdot \vec{A}_2) N_2 = \frac{\mu_o N_1 I_1}{d} N_2 A_2$

$$\rightarrow$$
 mutual inductance $M = \frac{\Phi_{21}}{I_1} = \frac{\mu_0 N_1 N_2 A}{d}$

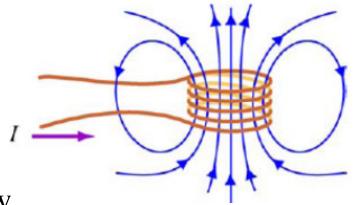
starting with current in coil?

- non-uniform \vec{B}_2
- more difficult derivation
- different analytical expression
- but same numerical answer



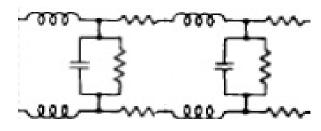
self-inductance of current loop

- magnetic field set up by I in coil
- (partial) magnetic flux linkage induced emf $V = -\frac{d}{dt}(N\Phi)$ direction determined by Lenz's Law



definition: $L = \frac{N\Phi}{I}$ (same unit of Henry as mutual inductance)

- (a) interpreted as measure of magnetic flux/energy storage
- (b) adopted as series component in transmission-line model
- (c) can minimize L by removing loops
- (d) may still have to consider $L_{internal}$ due to magnetic field in conductor



self-inductance example: coaxial cable current in central conductor (need not be uniform)

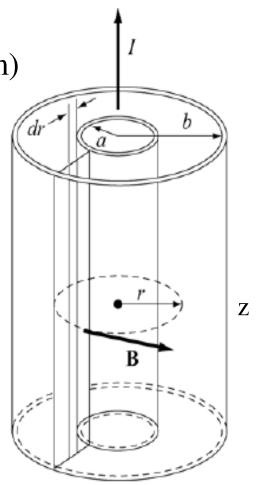
- \rightarrow Ampere's Law for deriving \vec{B} expression in space between conductors (a < r < b)
- \rightarrow flux over narrow strip dr at r

$$\begin{split} d\Phi &= \vec{B} \bullet d\vec{A} = \tfrac{\mu_o I}{2\pi r} z dr \\ \Phi &= \tfrac{\mu_o Iz}{2\pi} \int_a^b \tfrac{1}{r} dr = \tfrac{\mu_o Iz}{2\pi} \ln \tfrac{b}{a} \end{split}$$

→ inductance per unit length

$$\hat{L} = \frac{L}{z} = \frac{\mu_o}{2\pi} \ln \frac{b}{a}$$

usually disregard $L_{interior}$ due to magnetic field inside central conductor (r < a)



transients for circuit with coil

Lenz's Law → opposes change in flux

(a) close S₁ (but leave S₂ open) back emf initially blocking current flow

(b) close S₂ (but leave S₁ open)back emf initially blocking current decay

