

ME 233 Advance Control II

Lecture 17

Stability Analysis Using The Hyperstability Theorem

Adaptive Control

Basic Adaptive Control Principle

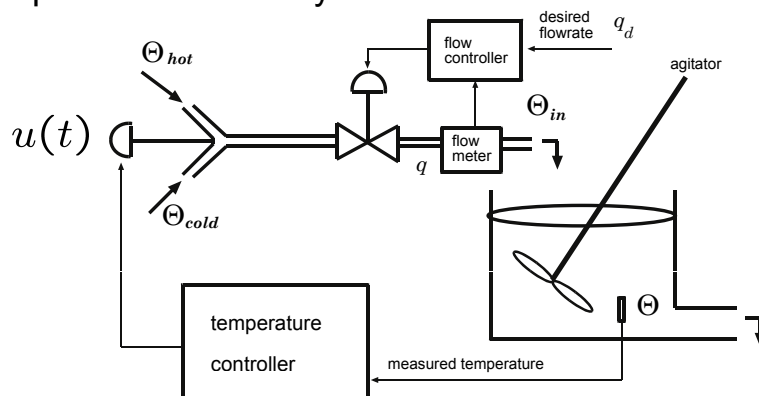
Controller parameters **are not constant**, rather, they are adjusted in an online fashion by a ***Parameter Adaptation Algorithm (PAA)***

When is adaptive control used?

- Plant parameters are unknown
- Plant parameters are time varying

Example of a system with varying parameters

- Temperature control system

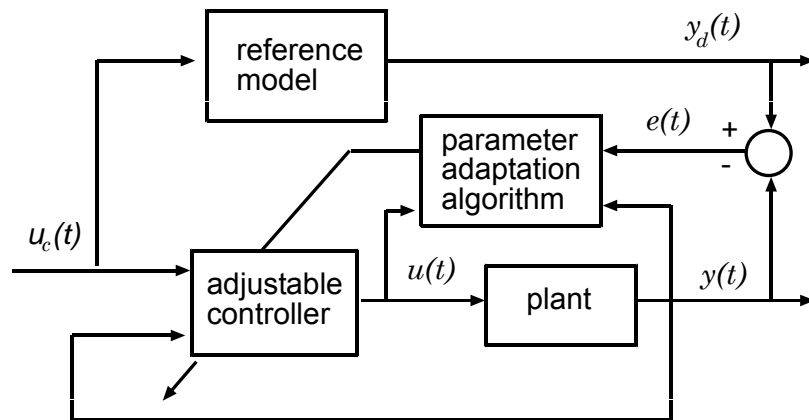


$$\frac{d}{dt}\theta(t) = - \underbrace{\frac{q}{V}}_{a(q)} \theta(t) + \underbrace{\frac{kq}{V}}_{b(q)} u(t - t_d)$$

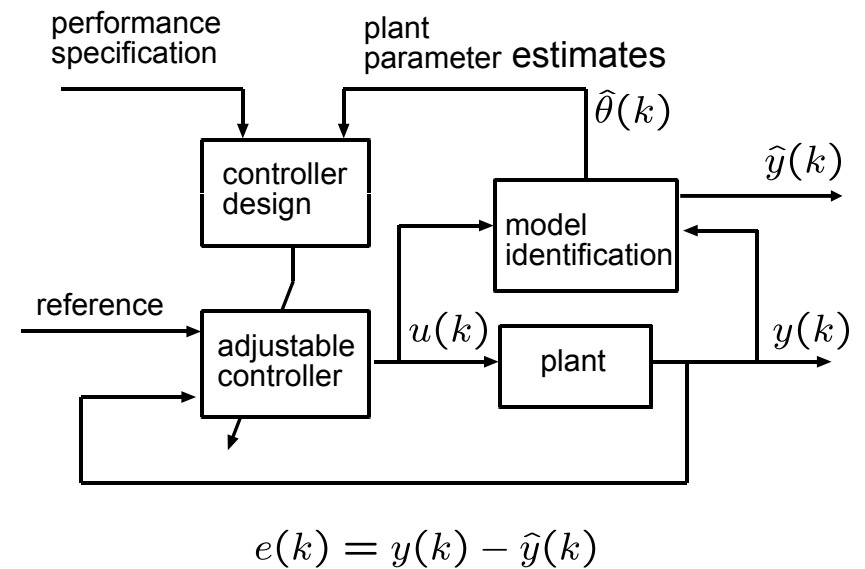
Adaptive Control Classification

- Continuous time VS **discrete time**
- Direct VS indirect
- MRAS VS **STR**

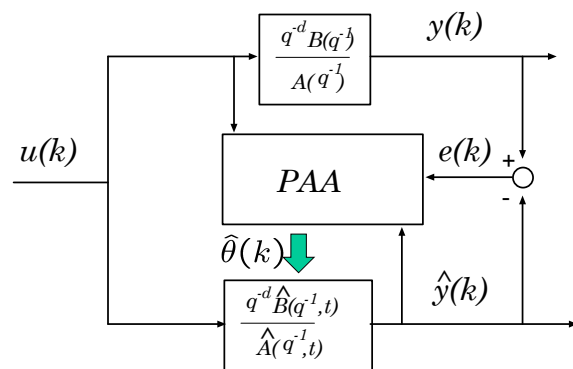
Model Reference Adaptive Systems (MRAS)



Self-Tuning Regulators (STR)

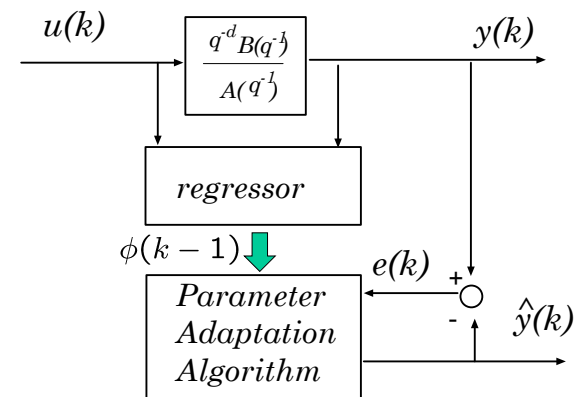


Identification of a LTI system



Parallel model

Identification of a LTI system



Series-parallel model

Plant ARMA Model

Plant model

$$A(q^{-1})y(k) = q^{-d}B(q^{-1})u(k)$$

where

$$A(q^{-1}) = 1 + a_1q^{-1} + \dots + a_nq^{-n}$$

$$B(q^{-1}) = b_o + b_1q^{-1} + \dots + b_mq^{-m}$$

Unknown plant parameters

Assume ARMA model parameters are unknown

$$y(k) = - \underline{a_1} y(k-1) \dots - \underline{a_n} y(k-n) \\ + \underline{b_o} u(k-d) \dots + \underline{b_m} u(k-d-m)$$

Define:

$$\theta = \left[\underline{a_1} \quad \dots \quad \underline{a_n} \quad \underline{b_o} \quad \dots \quad \underline{b_m} \right]^T$$

As the unknown parameter vector

Regressor vector

Collect all measurable signals in one vector

$$y(k) = - a_1 \underline{y(k-1)} \dots - a_n \underline{y(k-n)} \\ + b_o \underline{u(k-d)} \dots + b_m \underline{u(k-d-m)}$$

We define

$$\phi(k-1) = \left[- \underline{y(k-1)} \dots - \underline{y(k-n)} \right. \\ \left. \underline{u(k-d)} \dots \underline{u(k-d-m)} \right]^T$$

as the known regressor vector

Plant ARMA Model

Plant model

$$y(k) = \phi^T(k-1) \theta$$

where

$$\theta = \left[a_1 \quad \dots \quad a_n \quad b_o \quad \dots \quad b_m \right]^T$$

$$\phi(k-1) = \left[- y(k-1) \dots - y(k-n) \right. \\ \left. u(k-d) \dots u(k-d-m) \right]^T$$

Plant ARMA Model

Plant model

$$\hat{y}(k) = \phi^T(k-1) \hat{\theta}(k)$$

where

$$\hat{\theta}(k) = \begin{bmatrix} \hat{a}_1(k) & \cdots & \hat{a}_n(k) & \hat{b}_o(k) & \cdots & \hat{b}_m(k) \end{bmatrix}^T$$

$$\phi(k-1) = [-y(k-1) \cdots -y(k-n) \\ u(k-d) \cdots u(k-d-m)]^T$$

Plant output estimate

Plant a-posteriori estimate

$$\hat{y}(k) = \phi^T(k-1) \hat{\theta}(k)$$

Plant a-priori estimate

$$\hat{y}^o(k) = \phi^T(k-1) \hat{\theta}(k-1)$$

Plant a-posteriori error

$$y(k) = \phi^T(k-1) \theta$$

$$\hat{y}(k) = \phi^T(k-1) \hat{\theta}(k)$$

error:
$$e(k) = y(k) - \hat{y}(k)$$

$$e(k) = \phi^T(k-1) [\theta - \hat{\theta}(k)]$$

$$= \phi^T(k-1) \tilde{\theta}(k)$$

$$\tilde{\theta}(k) = \theta - \hat{\theta}(k)$$

Parameter Adaptation Algorithm

PAA
$$F = F^T \succ 0$$

$$\hat{\theta}(k) = \hat{\theta}(k-1) + F \phi(k-1) e(k)$$

Parameter error update law:
$$\tilde{\theta}(k) = \theta - \hat{\theta}(k)$$

$$\tilde{\theta}(k) = \tilde{\theta}(k-1) - F \phi(k-1) e(k)$$

Adaptation Dynamics

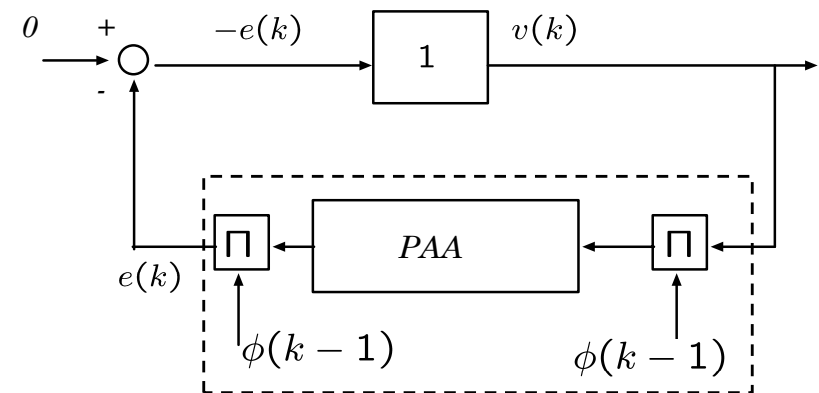
a-posteriori error: $e(k) = y(k) - \hat{y}(k)$

$$e(k) = \phi^T(k-1)\tilde{\theta}(k)$$

Parameter error update law: $\tilde{\theta}(k) = \theta - \hat{\theta}(k)$

$$\tilde{\theta}(k) = \tilde{\theta}(k-1) - F \phi(k-1)e(k)$$

Adaptation Dynamics



$$PAA: \quad \tilde{\theta}(k) = \tilde{\theta}(k-1) + F \phi(k-1)v(k)$$

Convergence of Adaptive Systems

Adaptive systems are nonlinear

We need to prove that the algorithms converge:

- **Output error convergence**

$$e(k) \rightarrow 0$$

$$e(k) = y(k) - \hat{y}(k)$$

- **Parameter error convergence**

$$\tilde{\theta}(k) \rightarrow 0$$

$$\tilde{\theta}(k) = \theta - \hat{\theta}(k)$$

Output error Convergence

Our first goal will be to prove the asymptotic convergence of the output error:

$$e(k) \rightarrow 0$$

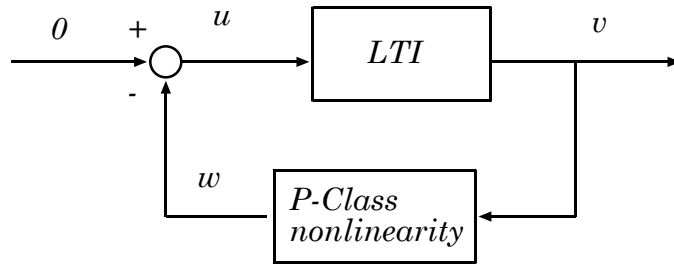
Two frequently used methods of stability analysis are:

- **Stability analysis using Lyapunov's direct method**
 - State space approach
- **Stability analysis using the Passivity or Hyperstability theorems**
 - Input/output approach

Hyperstability

Hyperstability Theory

- Developed by V.M. Popov to analyze the stability of a class of feedback systems (monograph published in 1973)

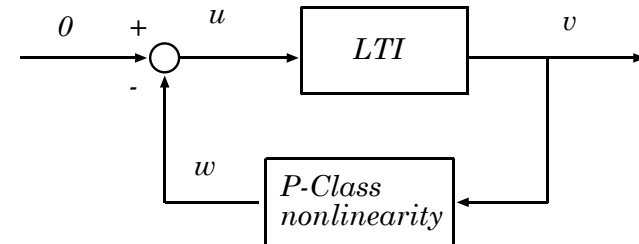


- Popularized by I.D. Landau for the analysis of adaptive systems (first book published in 1979)

Hyperstability Theory

Hyperstability Theory

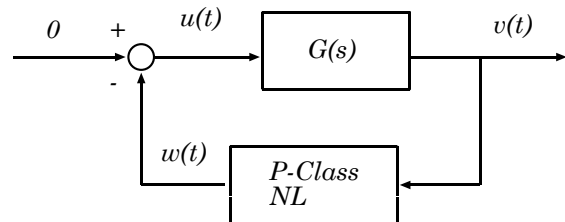
- Applies to both continuous time and discrete time systems



- Abuse of notation:** We will denote the LTI block by its transfer function

CT Hyperstability Theory

$$G(s) = C(sI - A)^{-1}B + D$$

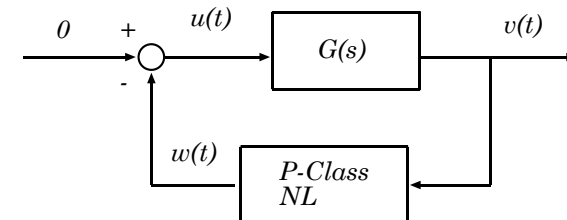


- A state space description of the LTI Block:

$$\frac{d}{dt}x(t) = Ax(t) + Bu(t)$$

$$v(t) = Cx(t) + Du(t)$$

CT Hyperstability Theory

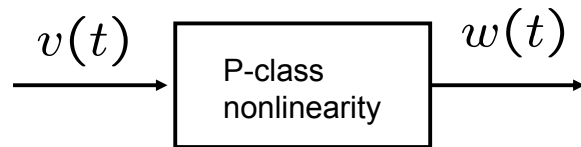


- P-class nonlinearity: (passive nonlinearities)

$$\int_0^t w^T v d\tau \geq -\gamma_o^2 \quad \forall t \geq 0$$

Where γ_o is a constant which is a function of the initial conditions

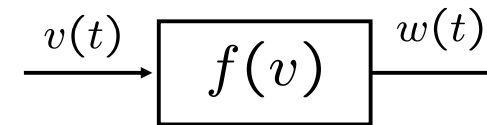
CT Hyperstability Theory



$$\int_0^t w^T v \, d\tau \geq -\gamma_o^2 \quad \forall t \geq 0$$

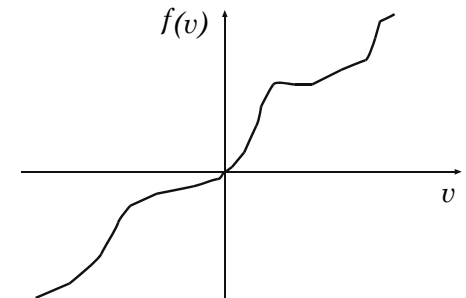
Where γ_o is a constant which is a function of the initial conditions

Example: Static P-class NL



$$w = f(v)$$

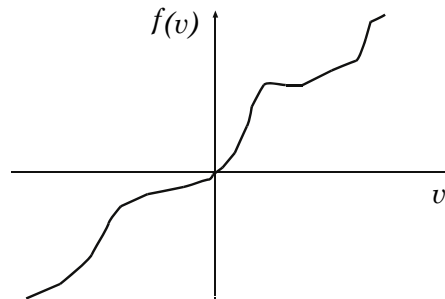
$$v f(v) \geq 0$$



Example: Static P-class NL

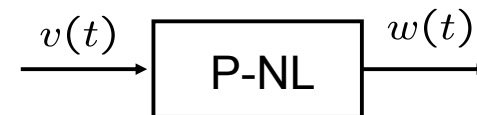
$$w = f(v)$$

$$v f(v) \geq 0$$



$$\int_0^t w v \, d\tau = \int_0^t \underbrace{f(v) v}_{\geq 0} \, d\tau \geq 0 > -\gamma_o^2$$

Example: Dynamic P-class block



$$\begin{cases} \frac{d}{dt} \tilde{\theta}(t) = F \phi(t) v(t) \\ w(t) = \phi^T(t) \tilde{\theta}(t) \end{cases}$$

$$\phi(t) \in \mathcal{R}^n$$

$$\tilde{\theta}(0) \in \mathcal{R}^n$$

$$|\tilde{\theta}(0)| < \infty$$

$$|\phi(t)| < \infty$$

$$F = F^T \succ 0$$

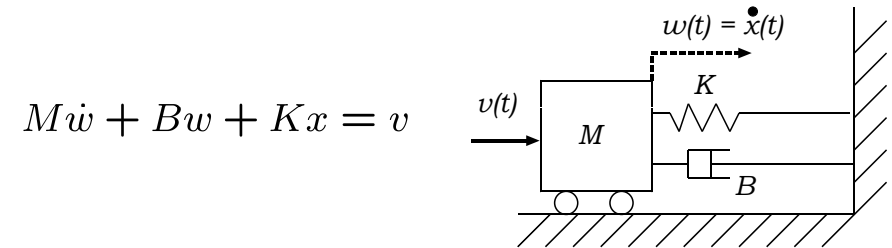
Example: Dynamic P-class block

$$w(t) = \phi^T(t)\tilde{\theta}(t) \quad \dot{\tilde{\theta}}(t) = F\phi(t)v(t)$$

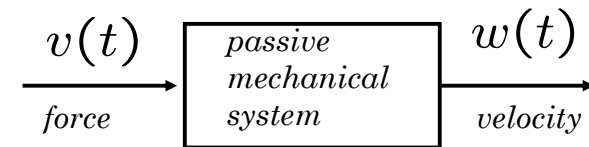
$$\begin{aligned} \int_0^t w(\tau)v(\tau) d\tau &= \int_0^t \phi^T(\tau)\tilde{\theta}(\tau)v(\tau) d\tau \\ &= \int_0^t \tilde{\theta}^T(\tau) \underbrace{[\phi(\tau)\tilde{v}(\tau)]}_{F^{-1}\dot{\tilde{\theta}}(\tau)} d\tau \\ &= \frac{1}{2} \int_0^t \frac{d}{d\tau} \{ \tilde{\theta}^T(\tau) F^{-1} \tilde{\theta}(\tau) \} d\tau \\ &= \frac{1}{2} \tilde{\theta}^T(t) F^{-1} \tilde{\theta}(t) - \underbrace{\frac{1}{2} \tilde{\theta}^T(0) F^{-1} \tilde{\theta}(0)}_{\gamma_o^2} \\ &\geq -\gamma_o^2 \end{aligned}$$

Example: Passive mechanical system

Input is force and output is velocity

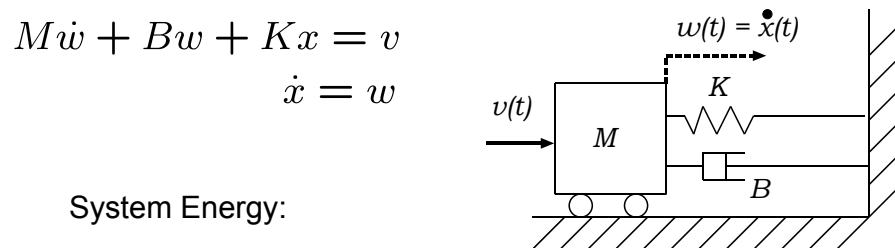


$$M\dot{w} + Bw + Kx = v$$



Example: Passive mechanical system

Input is force and output is velocity

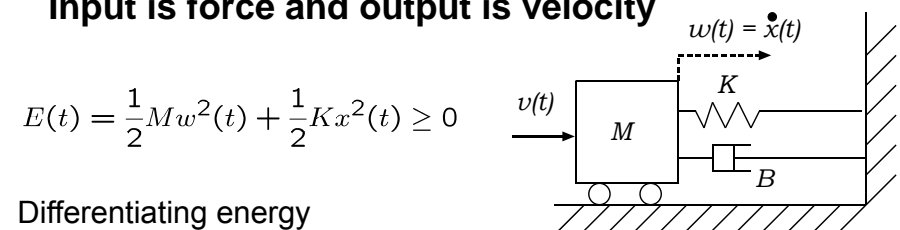


System Energy:

$$E(t) = \frac{1}{2} M w^2(t) + \frac{1}{2} K x^2(t) \geq 0$$

Example: Passive mechanical system

Input is force and output is velocity



$$E(t) = \frac{1}{2} M w^2(t) + \frac{1}{2} K x^2(t) \geq 0$$

Differentiating energy

$$\dot{E} = M\dot{w}w + Kxw$$

$$= [-Kx - Bw + v]w + Kxw$$

$$= \cancel{-Kxw} - Bw^2 + ww + \cancel{Kxw}$$

Example: Passive mechanical system

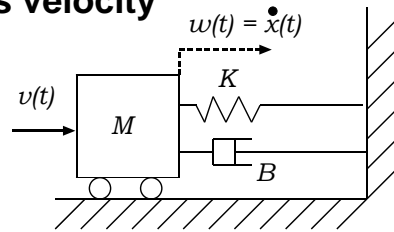
Input is force and output is velocity

$$E(t) = \frac{1}{2}Mw^2(t) + \frac{1}{2}Kx^2(t) \geq 0$$

Differentiating energy

$$\dot{E} = -Bw^2 + wv$$

$$\underbrace{wv}_{\text{power input}} = \dot{E} + Bw^2$$



Example: Passive mechanical system

Input is force and output is velocity

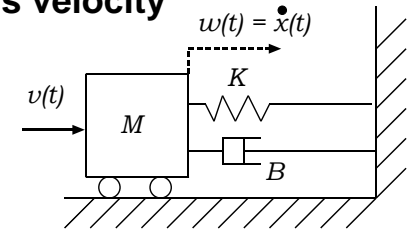
$$E(t) = \frac{1}{2}Mw^2(t) + \frac{1}{2}Kx^2(t) \geq 0$$

integrating power,

$$\int_0^t wv \, d\tau = E(t) - E(0) + \int_0^t Bw^2(\tau) \, d\tau$$

$$\geq -\gamma_o^2$$

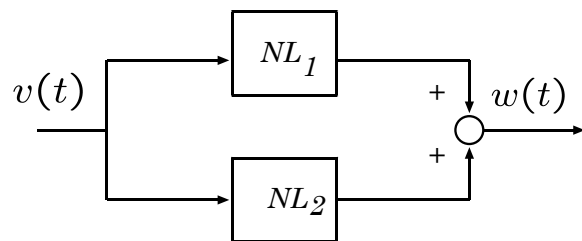
$$\gamma_o^2 = E(0) \geq 0$$



Examples of P-class NL

Lemma:

- The parallel combination of two P-class nonlinearities is also a P-class nonlinearity.

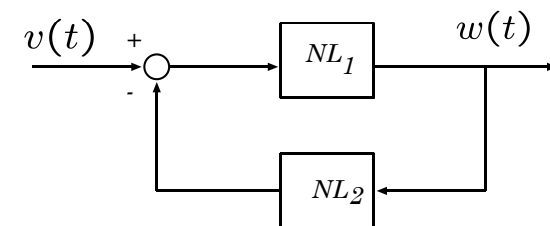


$$\int_0^t w^T v \, d\tau \geq -\gamma_o^2$$

Examples of P-class NL

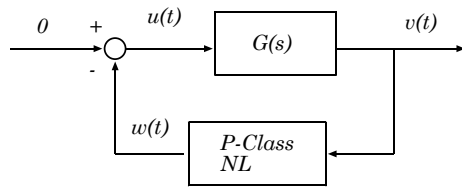
Lemma:

- The feedback combination of two P-class nonlinearities is also a P-class nonlinearity.



$$\int_0^t w^T v \, d\tau \geq -\gamma_o^2$$

CT Hyperstability

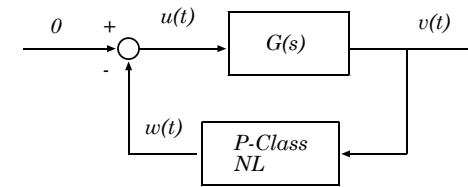


Hyperstability: The above feedback system is hyperstable if there exist positive bounded constants δ_1, δ_2 such that, for any state space realization of $G(s)$,

$$|x(t)| < \delta_1 [|x(0)| + \delta_2] \quad \forall t \geq 0$$

FOR ALL P-class nonlinearities

CT Asymptotic Hyperstability

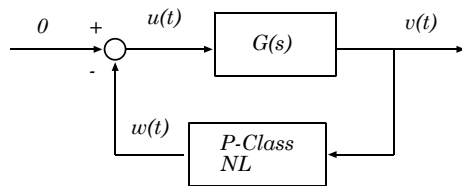


Asymptotic Hyperstability: The above feedback system is asymptotically hyperstable if

1. It is hyperstable
2. For all signals $|w(t)| < \infty$ (i.e. bounded output of any P-class nonlinearity), and any state space realization of $G(s)$,

$$\lim_{t \rightarrow \infty} x(t) = 0$$

CT Hyperstability Theorems



Hyperstability Theorem: The above feedback system is hyperstable **iff** the transfer function $G(s)$ of the LTI block is **Positive Real**.

Asymptotical Hyperstability Theorem: The above feedback system is asymptotically hyperstable **iff** the transfer function $G(s)$ of the LTI block is **Strictly Positive Real**.

CT Positive Real TF

$$G(s) = C(sI - A)^{-1}B + D$$

Is **Positive Real** iff:

1. $G(s)$ does not have any unstable poles (i.e. no $\text{Re}\{s\} > 0$).
2. Any pole of $G(s)$ that is in the imaginary axis does not repeat and its associated residue (i.e. the coefficient appearing in the partial fraction expansion) is non-negative.

$$2 \text{Re}\{G(j\omega)\} = G(j\omega) + G^T(-j\omega) \geq 0$$

for all real ω 's for which $s = j\omega$ is not a pole of $G(s)$

Strictly Positive Real (SPR) TF

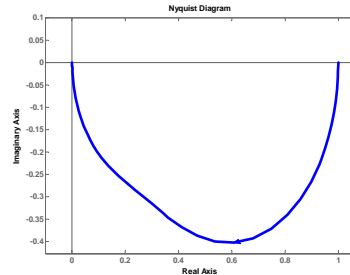
$$G(s) = C(sI - A)^{-1}B + D$$

Is **Strictly Positive Real (SPR)** iff:

1. All poles of $G(s)$ are asymptotically stable.
2. $2 \operatorname{Re}\{G(j\omega)\} = G(j\omega) + G^T(-j\omega) > 0$
for all ω , $0 \leq \omega < \infty$

Example:

$$G(s) = \frac{s + 1}{s^2 + 3s + 1}$$



Strictly Positive Real (SPR) TF

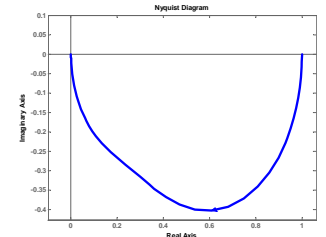
For scalar rational transfer functions

$$G(s) = \frac{B(s)}{A(s)}$$

1. All poles of $G(s)$ are asymptotically stable.
2. $\operatorname{Re}\{G(j\omega)\} > 0$ for all ω , $0 \leq \omega < \infty$

Note:

A necessary (but not sufficient) condition for $G(s)$ to be SPR is that its relative degree must be less than or equal to 1.



Kalman Yakubovich Popov Lemma

$$G(s) = C(sI - A)^{-1}B + D$$

Is **Strictly Positive Real (SPR)** if and only if

- there exist a symmetric and positive definite matrix P ,
- matrices L and K ,
- and a constant $\epsilon > 0$ such that

$$A^T P + P A = -L^T L - \epsilon P$$

$$B^T P - C = -K^T L$$

$$D + D^T = K^T K$$

Kalman Yakubovich Popov Lemma

$$G(s) = C(sI - A)^{-1}B$$

Is **Strictly Positive Real (SPR)** iff there exist symmetric and positive definite matrices P and Q , such that:

$$A^T P + P A = -Q$$

$$B^T P = C$$

SPR TF implies Possitivity

Let $G(s) = C(sI - A)^{-1}B + D$ be SPR

Then there exist positive definite functions

$$V(x) \succ 0 \quad \lambda_1(x) \succ 0$$

and a positive semi-definite function $\lambda_2(x, u) \succeq 0$

Such that the input $u(t)$ output $y(t)$ pair satisfies

$$\begin{aligned} \int_0^t y^T u \, d\tau &= V(x(t)) - V(x(0)) + \int_0^t (\lambda_1(x) + \lambda_2(x, u)) \, d\tau \\ &\geq -\gamma_o^2 \end{aligned}$$

$\gamma_o^2 = V(x(0))$

SPR TF implies Passivity

Proof: We consider a strictly causal transfer function

$$G(s) = C(sI - A)^{-1}B$$

which is SPR, with state space realization

$$\begin{aligned} \frac{d}{dt}x &= Ax + Bu \\ v &= Cx \end{aligned}$$

By the Kalman Yakubovich, Popov lemma, there exist symmetric and positive definite matrices P and Q , such that

$$\begin{aligned} A^T P + PA &= -Q \\ B^T P &= C \end{aligned}$$

SPR TF implies Passivity

Proof: Define the PD function $V(x) = \frac{1}{2}x^T Px$

and compute:

$$\begin{aligned} 2\dot{V}(x) &= \dot{x}^T Px + x^T P\dot{x} \\ &= (Ax + Bu)^T Px + x^T P(Ax + Bu) \\ &= x^T \left[\underbrace{A^T P + PA}_{-Q} \right] x + 2u^T \underbrace{B^T P x}_v \end{aligned}$$

by the Kalman Yakubovich, Popov lemma.

$$\begin{aligned} A^T P + PA &= -Q \\ B^T P &= C \end{aligned}$$

SPR TF implies Passivity

Proof: Thus, since $v = Cx$

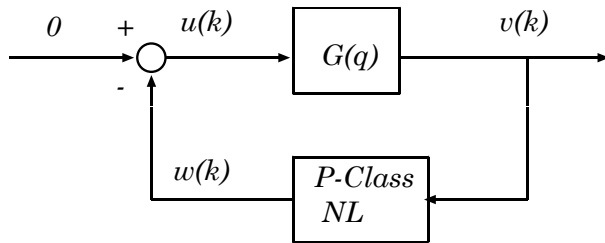
$$u^T v = \dot{V} + \frac{1}{2}x^T Qx$$

Define the PD function $\lambda_1(x) = \frac{1}{2}x^T Qx$ and integrate

$$\begin{aligned} \int_0^t u^T v \, d\tau &= \int_0^t \dot{V} \, d\tau + \int_0^t \lambda_1(x) \, d\tau \\ &= V(x(t)) - V(x(0)) + \int_0^t \lambda_1(x) \, d\tau \end{aligned}$$

DT Hyperstability Theory

$$G(z) = C(zI - A)^{-1}B + D$$

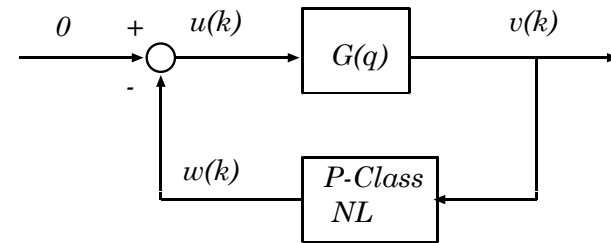


- State space description of the LTI Block:

$$x(k+1) = Ax(k) + Bu(k)$$

$$v(k) = Cx(k) + Du(k)$$

DT Hyperstability Theory

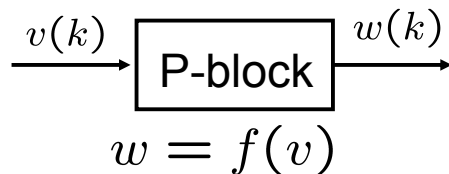


- P-class nonlinearity: (passive nonlinearities)

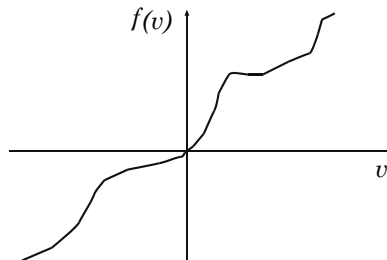
$$\sum_{j=0}^k w^T(j)v(j) \geq -\gamma_o^2 \quad \forall k \geq 0$$

Where γ_o is a bounded constant.

Example: Static nonlinearity:

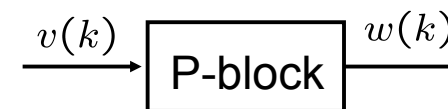


$$v f(v) \geq 0$$



$$\sum_{j=0}^k w^T(j)v(j) = \sum_{j=0}^k \underbrace{f(v(j))v(j)}_{\geq 0} \geq 0 > -\gamma_o^2$$

Example: Dynamic P-class block



$$\begin{cases} \tilde{\theta}(k) = \tilde{\theta}(k-1) + F \phi(k)v(k) \\ w(k) = \phi^T(k)\tilde{\theta}(k) \end{cases}$$

$$F = F^T \succ 0$$

$$\phi(k) \in \mathcal{R}^n$$

$$\tilde{\theta}(-1) \in \mathcal{R}^n$$

$$|\tilde{\theta}(-1)| < \infty$$

$$|\phi(k)| < \infty$$

Example: Dynamic P-class block

$$w(k) = \phi^T(k) \tilde{\theta}(k) \quad \tilde{\theta}(k) = \tilde{\theta}(k-1) + F \phi(k) v(k)$$

$$\begin{aligned} \sum_{j=0}^k w(j)v(j) &= \sum_{j=0}^k \phi^T(j) \tilde{\theta}(j) v(j) \\ &= \sum_{j=0}^k \tilde{\theta}^T(j) \underbrace{[\phi(j)v(j)]}_{F^{-1}[\tilde{\theta}(j) - \tilde{\theta}(j-1)]} \\ &= \sum_{j=0}^k \tilde{\theta}^T(j) F^{-1} [\tilde{\theta}(j) - \tilde{\theta}(j-1)] \\ &= \sum_{j=0}^k \{ \tilde{\theta}^T(j) F^{-1} \tilde{\theta}(j) - \tilde{\theta}^T(j) F^{-1} \tilde{\theta}(j-1) \} \end{aligned}$$

Example: Dynamic P-class block

$$\begin{aligned} \sum_{j=0}^k w(j)v(j) &= \sum_{j=0}^k \{ \tilde{\theta}^T(j) F^{-1} \tilde{\theta}(j) - \tilde{\theta}^T(j) F^{-1} \tilde{\theta}(j-1) \} \\ &\quad + \frac{1}{2} \sum_{j=0}^k \tilde{\theta}^T(j-1) F^{-1} \tilde{\theta}(j-1) - \frac{1}{2} \sum_{j=0}^k \tilde{\theta}^T(j-1) F^{-1} \tilde{\theta}(j-1) \\ \sum_{j=0}^k w(j)v(j) &= \frac{1}{2} \sum_{j=0}^k \{ \tilde{\theta}^T(j) F^{-1} \tilde{\theta}(j) - \tilde{\theta}^T(j-1) F^{-1} \tilde{\theta}(j-1) \} \\ &\quad + \underbrace{\frac{1}{2} \sum_{j=0}^k [\tilde{\theta}(j) - \tilde{\theta}(j-1)]^T F^{-1} [\tilde{\theta}(j) - \tilde{\theta}(j-1)]}_{\geq 0} \end{aligned}$$

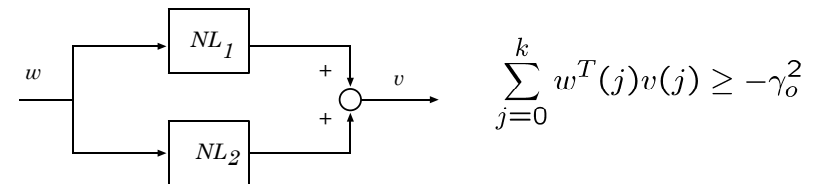
Example: Dynamic P-class block

$$\begin{aligned} \sum_{j=0}^k w(j)v(j) &\geq \frac{1}{2} \sum_{j=0}^k \{ \tilde{\theta}^T(j) F^{-1} \tilde{\theta}(j) - \tilde{\theta}^T(j-1) F^{-1} \tilde{\theta}(j-1) \} \\ &\geq \frac{1}{2} \tilde{\theta}^T(k) F^{-1} \tilde{\theta}(k) - \underbrace{\frac{1}{2} \tilde{\theta}^T(-1) F^{-1} \tilde{\theta}(-1)}_{\gamma_o^2} \\ &\geq -\gamma_o^2 \end{aligned}$$

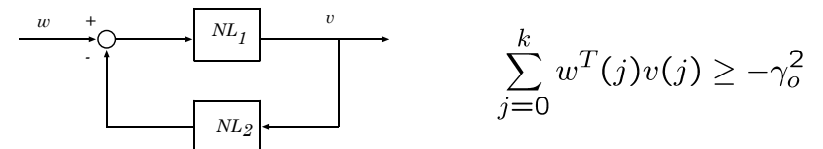
Examples of P-class NL

Lemma:

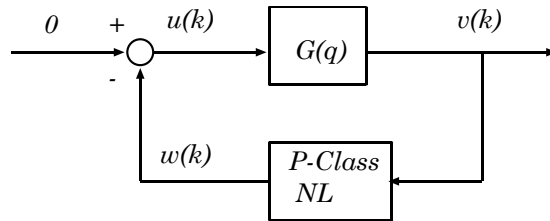
- The parallel combination of two P-class nonlinearities is also a P-class nonlinearity.



- The feedback combination of two P-class nonlinearities is also a P-class nonlinearity.



DT Hyperstability

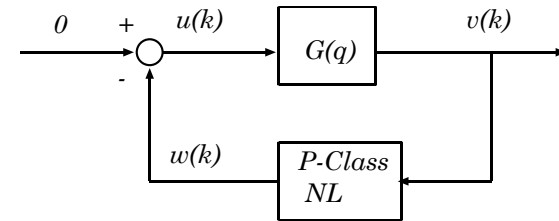


Hyperstability: The above feedback system is hyperstable if there exist positive bounded constants δ_1, δ_2 such that, for any state space realization of $G(q)$,

$$|x(k)| < \delta_1 [|x(0)| + \delta_2] \quad \forall k \geq 0$$

FOR ALL P-class nonlinearities

DT Asymptotic Hyperstability

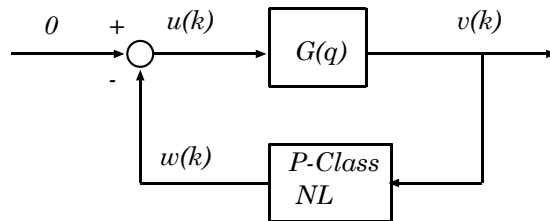


Asymptotic Hyperstability: The above feedback system is asymptotically hyperstable if

1. It is hyperstable
2. for any state space realization of $G(s)$,

$$\lim_{k \rightarrow \infty} x(k) = 0$$

DT Hyperstability Theorems



Hyperstability Theorem: The above feedback system is hyperstable **iff** the transfer function $G(z)$ of the LTI block is **Positive Real**.

Asymptotical Hyperstability Theorem: The above feedback system is asymptotically hyperstable **iff** the transfer function $G(z)$ of the LTI block is **Strictly Positive Real**.

Positive Real TF

$$G(z) = C(zI - A)^{-1}B + D$$

Is **Positive Real** iff:

1. $G(z)$ does not have any unstable poles (i.e. no $|z| > 1$).
2. Any pole of $G(z)$ that is in the unit circle does not repeat and its associated residue (i.e. the coefficient appearing in the partial fraction expansion) is non-negative.
3. $2 \operatorname{Re}\{G(e^{j\omega})\} = G(e^{j\omega}) + G^T(e^{-j\omega}) \geq 0$

for all ω , $0 \leq \omega \leq \pi$ for which $z = e^{j\omega}$ is not a pole of $G(s)$

Strictly Positive Real (SPR) TF

$$G(z) = C(zI - A)^{-1}B + D$$

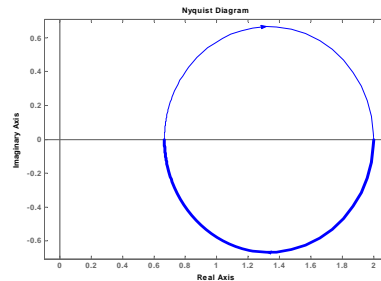
Is **Strictly Positive Real (SPR)** iff:

1. All poles of $G(z)$ are asymptotically stable.
2. $2 \operatorname{Re}\{G(e^{j\omega})\} = G(e^{j\omega}) + G^T(e^{-j\omega}) > 0$

for all $0 \leq \omega \leq \pi$

Example:

$$G(z) = \frac{z}{z + 0.5}$$



Strictly Positive Real (SPR) TF

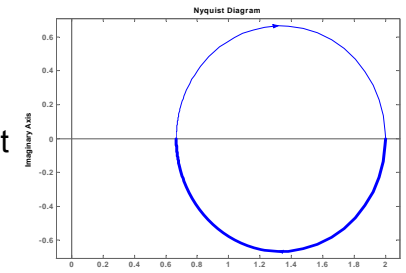
For scalar rational transfer functions

$$G(z) = \frac{B(z)}{A(z)}$$

1. All poles of $G(s)$ are asymptotically stable.
2. $\operatorname{Re}\{G(e^{j\omega})\} > 0$ for all ω , $0 \leq \omega \leq \pi$

Note:

A necessary (but not sufficient) condition for $G(z)$ to be SPR is that its relative degree must be 0.



Kalman Szegö Popov Lemma

$$G(z) = C(zI - A)^{-1}B + D$$

Is **Strictly Positive Real (SPR)** if and only if

- there exist a symmetric and positive definite matrix P ,
- matrices L and K ,
- and a constant $\epsilon > 0$ such that

$$A^T P A - P = -L^T L - \epsilon P$$

$$B^T P A - C = -K^T L$$

$$D + D^T - B^T P B = K^T K$$

SPR TF implies Possitivity

Let $G(z) = C(zI - A)^{-1}B + D$ be SPR

Then there exist positive definite functions

$$V(x) \succ 0 \quad \lambda_1(x) \succ 0$$

and a positive semi-definite function $\lambda_2(x, u) \succeq 0$

Such that the input $u(k)$ output $y(k)$ pair satisfies

$$\begin{aligned} \sum_{j=0}^k y^T(j)u(j) &= V(x(k+1)) - V(x(0)) + \sum_{j=0}^k \lambda_1(x(j)) \\ &\quad + \sum_{j=0}^k \lambda_2(x(j), u(j)) \\ &\geq -\gamma_o^2 \quad \gamma_o^2 = V(x(0)) \end{aligned}$$

Proof

Let $G(z) = C(zI - A)^{-1}B + D$ be SPR

Then by the Kalman Szegö Popov Lemma

$$A^T P A - P = -L^T L - \epsilon P \quad P = P^T \succ 0$$

$$B^T P A - C = -K^T L \quad \epsilon P \succ 0$$

$$D + D^T - B^T P B = K^T K$$

Define the Lyapunov function

$$V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0$$

Proof

$$V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0$$

$$V(x_{k+1}) - V(x_k) = \frac{1}{2} (Ax_k + Bu_k)^T P (Ax_k + Bu_k) - \frac{1}{2} x_k^T P x_k$$

$$= \frac{1}{2} x_k^T (A^T P A - P) x_k$$

$$\boxed{A^T P A - P = -L^T L - \epsilon P} \longrightarrow -\epsilon P - L^T L$$

$$+ \frac{1}{2} x_k^T \underbrace{A^T P B}_{C^T - L^T K} u_k + \frac{1}{2} u_k^T \underbrace{B^T P A}_{C - K^T L} x_k$$

$$\boxed{B^T P A - C = -K^T L} \longrightarrow$$

$$+ \frac{1}{2} u_k^T \underbrace{B^T P B}_{D + D^T - K^T K} u_k$$

$$\boxed{D + D^T - B^T P B = K^T K} \longrightarrow$$

Proof

$$V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0$$

$$V(x_{k+1}) - V(x_k) = - \underbrace{\epsilon \frac{1}{2} x_k^T P x_k}_{\lambda_1(x_k) \succ 0}$$

$$+ \frac{1}{2} \underbrace{(Cx_k + Du_k)^T}_{y_k^T} u_k + \frac{1}{2} u_k^T \underbrace{(Cx_k + u_k)}_{y_k}$$

$$- \underbrace{\frac{1}{2} \begin{bmatrix} x_k^T & u_k^T \end{bmatrix} \begin{bmatrix} L^T \\ K^T \end{bmatrix} \begin{bmatrix} L & K \end{bmatrix} \begin{bmatrix} x_k \\ u_k \end{bmatrix}}_{\lambda_2(x_k, u_k) \succeq 0}$$

Proof

$$V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0$$

$$V(x_{k+1}) - V(x_k) = \underbrace{+ y_k^T u_k}_{\lambda_1(x_k) \succ 0} - \underbrace{\epsilon \frac{1}{2} x_k^T P x_k}_{\lambda_2(x_k, u_k) \succeq 0} - \frac{1}{2} \|Lx_k + Ku_k\|^2$$

$$V(x_{k+1}) - V(x_k) = + y_k^T u_k - \lambda_1(x_k) - \lambda_2(x_k, u_k)$$

Proof

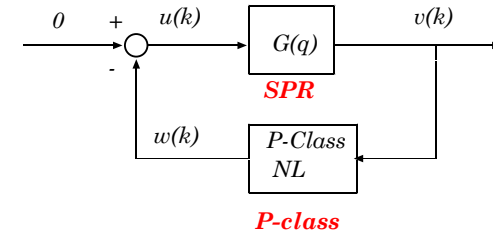
$$V(x_{k+1}) - V(x_k) = + y_k^T u_k - \lambda_1(x_k) - \lambda_2(x_k, u_k)$$

Taking summation:

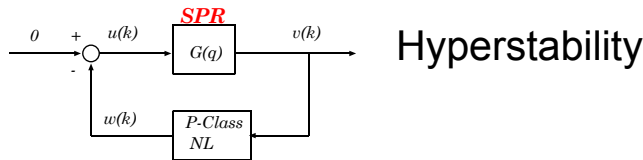
$$\begin{aligned} \sum_{j=0}^k y_j^T u_j &= \sum_{j=0}^k \{V(x_{j+1}) - V(x_j)\} \\ &\quad + \sum_{j=0}^k \lambda_1(x_j) + \sum_{j=0}^k \lambda_2(x_j, u_j) \end{aligned}$$

$$\sum_{j=0}^k y_j^T u_j = V(x_{k+1}) - V(x_0) + \sum_{j=0}^k \lambda_1(x_j) + \sum_{j=0}^k \lambda_2(x_j, u_j)$$

Proof of the sufficiency part of the Asymptotic Hyperstability Theorem - Discrete Time



- Since the nonlinearity is P-class, $\sum_{j=0}^k w_j^T v_j \geq -\gamma_1^2$
- Since LTR block is SPR, we can use the Kalman Szegő Popov Lemma



Hyperstability

Using the Kalman Szegő Popov Lemma, for any minimal realization,

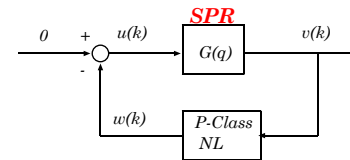
$$G(z) = C(zI - A)^{-1}B + D$$

we have

$$\sum_{j=0}^k v_j^T u_j = V(x_{k+1}) - V(x_0) + \sum_{j=0}^k \lambda_1(x_j) + \sum_{j=0}^k \lambda_2(x_j, u_j)$$

where

$$\lambda_1(x_k) = \epsilon V(x_k) \succ 0 \quad V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0 \quad \lambda_2(x_k, u_k) \geq 0$$



Hyperstability

$$G(z) = C(zI - A)^{-1}B + D$$

rearranging terms,

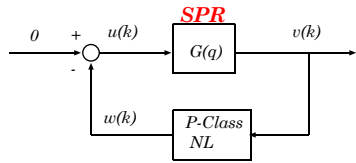
$$V(x_{k+1}) = V(x_0) + \sum_{j=0}^k v_j^T u_j - \sum_{j=0}^k \lambda_1(x_j) - \sum_{j=0}^k \lambda_2(x_j, u_j)$$

• P-class:

$$\sum_{j=0}^k w_j^T v_j \geq -\gamma_1^2 \quad \Rightarrow \quad \sum_{j=0}^k v_j^T u_j \leq \gamma_1^2$$

Therefore,

$$V(x_{k+1}) \leq V(x_0) + \gamma_1^2 - \sum_{j=0}^k \lambda_1(x_j) - \sum_{j=0}^k \lambda_2(x_j, u_j)$$



Hyperstability

$$\lambda_1(x_k) = \epsilon V(x_k) \succ 0$$

$$\lambda_2(x_k, u_k) \succeq 0$$

$$V(x_{k+1}) \leq V(x_0) + \gamma_1^2 - \underbrace{\sum_{j=0}^k \lambda_1(x_j) - \sum_{j=0}^k \lambda_2(x_j, u_j)}_{\leq 0}$$

Therefore,

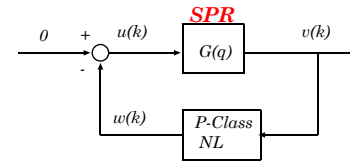
$$V(x_{k+1}) \leq V(x_0) + \gamma_1^2$$

Since $V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0$

$$\Rightarrow \lambda_{\min}(P)|x_k|^2 \leq \lambda_{\max}(P)|x_0|^2 + 2\gamma_1^2$$

$$\Rightarrow |x_k|^2 \leq \underbrace{\frac{\lambda_{\max}(P)}{\lambda_{\min}(P)}}_{\delta_1} \left[|x_0|^2 + \underbrace{\frac{2}{\lambda_{\max}(P)} \gamma_1^2}_{\delta_2} \right]$$

Thus, the feedback system is Hyperstable



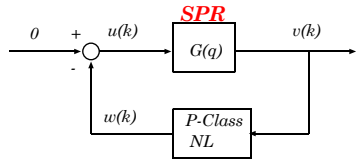
Asymptotic Hyperstability

$$G(z) = C(zI - A)^{-1}B + D$$

$$V(x_{k+1}) \leq V(x_0) + \gamma_1^2 - \sum_{j=0}^k \lambda_1(x_j) - \sum_{j=0}^k \lambda_2(x_j, u_j)$$

Taking the limit as $k \rightarrow \infty$

$$\lim_{k \rightarrow \infty} V(x_{k+1}) \leq V(x_0) + \gamma_1^2 - \sum_{j=0}^{\infty} \lambda_1(x_j) - \sum_{j=0}^{\infty} \lambda_2(x_j, u_j)$$



Asymptotic Hyperstability

$$G(z) = C(zI - A)^{-1}B + D$$

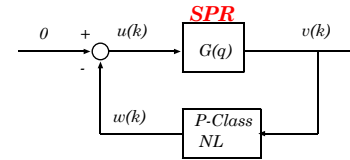
$$\lim_{k \rightarrow \infty} V(x_{k+1}) \leq \underbrace{V(x_0) + \gamma_1^2}_{\gamma_2^2} - \sum_{j=0}^{\infty} \lambda_1(x_j) - \underbrace{\sum_{j=0}^{\infty} \lambda_2(x_j, u_j)}_{\leq 0}$$

$$\lim_{k \rightarrow \infty} V(x_{k+1}) \leq \gamma_2^2 - \sum_{j=0}^{\infty} \lambda_1(x_j)$$

where

$$V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0$$

$$\lambda_1(x_k) = \epsilon V(x_k) \succ 0$$



Asymptotic Hyperstability

$$\lambda_1(x_k) = \epsilon V(x_k) \succ 0$$

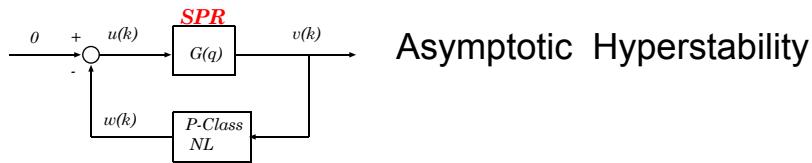
From our previous page, $\lim_{k \rightarrow \infty} V(x_{k+1}) \leq \gamma_2^2 - \underbrace{\sum_{j=0}^{\infty} \lambda_1(x_j)}_{\geq 0}$

Since $V(x_k) = \frac{1}{2} x_k^T P x_k \succ 0 \Rightarrow$ it cannot become negative

Moreover, the term $-\sum_{j=0}^{\infty} \lambda_1(x_j)$ can only become more negative or converge to a constant

Therefore,

$$0 \leq \lim_{k \rightarrow \infty} V(x_k) = V_{\infty} \leq \gamma_2^2 \Rightarrow \sum_{j=0}^{\infty} \lambda_1(x_j) = C_1 \leq \gamma_2^2$$

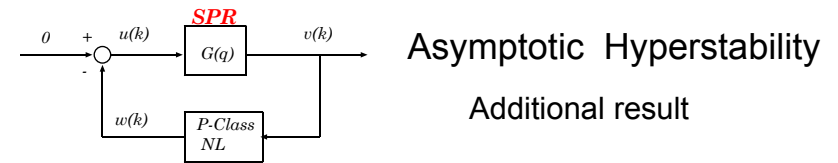


Since

$$\sum_{j=0}^{\infty} \lambda_1(x_j) = \epsilon \sum_{j=0}^{\infty} \underbrace{V(x_k)}_{\geq 0} = C_1 \quad \Rightarrow \quad \lim_{k \rightarrow \infty} V(x_k) = 0$$

Moreover, since $V(x_k) = \frac{1}{2} \underbrace{x_k^T P x_k}_{\succ 0} \quad \Rightarrow \quad \lim_{k \rightarrow \infty} x_k = 0$

Q.E.D.



Additional result

If in addition

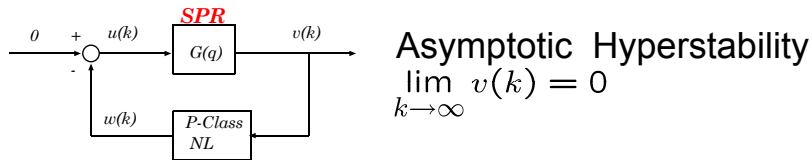
$$D + D^T = K^T K + B^T P B \succ 0$$

$$\begin{aligned} \lim_{k \rightarrow \infty} v(k) &= 0 \\ \lim_{k \rightarrow \infty} u(k) &= 0 \end{aligned}$$

We have already shown that $\lim_{k \rightarrow \infty} x_k = 0$

Since $v_k = Cx_k + Du_k$

We need to prove that: $\lim_{k \rightarrow \infty} Du_k = 0$



Using the Kalman Szegő Popov Lemma, we obtained

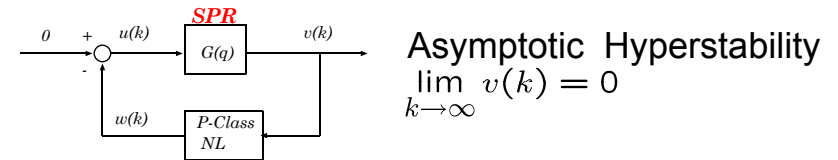
$$\lim_{k \rightarrow \infty} V(x_{k+1}) \leq V(x_0) + \gamma_1^2 - \sum_{j=0}^{\infty} \lambda_1(x_j) - \sum_{j=0}^{\infty} \lambda_2(x_j, u_j)$$

where: $\lambda_1(x_j) = \epsilon \frac{1}{2} x_k^T P x_k \succ 0$

$$\lambda_2(x_j, u_j) = \|Lx_k + Ku_k\|^2 \succeq 0$$

Thus: $\sum_{j=0}^{\infty} \lambda_1(x_j) < \infty \quad \Rightarrow \quad \lim_{k \rightarrow \infty} \lambda_1(x_k) = 0$

$\sum_{j=0}^{\infty} \lambda_2(x_j, u_j) < \infty \quad \Rightarrow \quad \lim_{k \rightarrow \infty} \lambda_2(x_k, u_k) = 0$



$$\lim_{k \rightarrow \infty} \lambda_1(x_k) = 0$$

$$\lambda_1(x_j) = \epsilon \frac{1}{2} x_k^T P x_k \succ 0$$

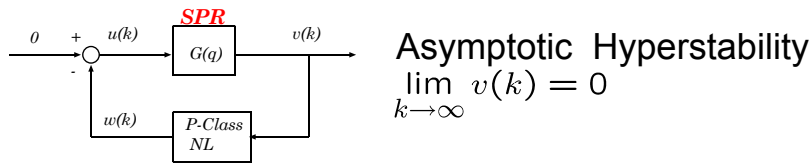
$$\lim_{k \rightarrow \infty} x_k = 0$$

$$\lim_{k \rightarrow \infty} \lambda_2(x_k, u_k) = 0$$

$$\lambda_2(x_j, u_j) = \|Lx_k + Ku_k\|^2 \succeq 0$$

$$\lim_{k \rightarrow \infty} \|Lx_k + Ku_k\|^2 = 0$$

$$\Rightarrow \lim_{k \rightarrow \infty} Ku_k = 0$$



So far we have: $\lim_{k \rightarrow \infty} x_k = 0$ $\lim_{k \rightarrow \infty} K u_k = 0$

The state equation $x(k+1) = A x(k) + B u(k)$

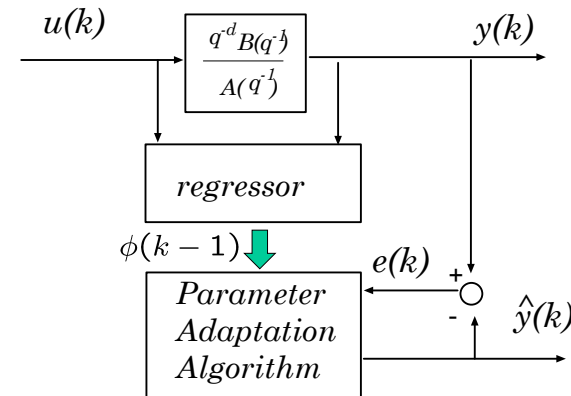
➡ $\lim_{k \rightarrow \infty} B u_k = 0$

From the Kalman Szegő Popov Lemma: $D + D^T = K^T K + B^T P B$

Thus, $u_k^T (D + D^T) u_k = u_k^T (K^T K + B^T P B) u_k = 0$

$(D + D^T) \succ 0$ ➡ $\lim_{k \rightarrow \infty} u_k = 0$ ➡ $\lim_{k \rightarrow \infty} v_k = 0$
 Q.E.D.

Stability analysis of Series-parallel ID



Series-Parallel ID Dynamics

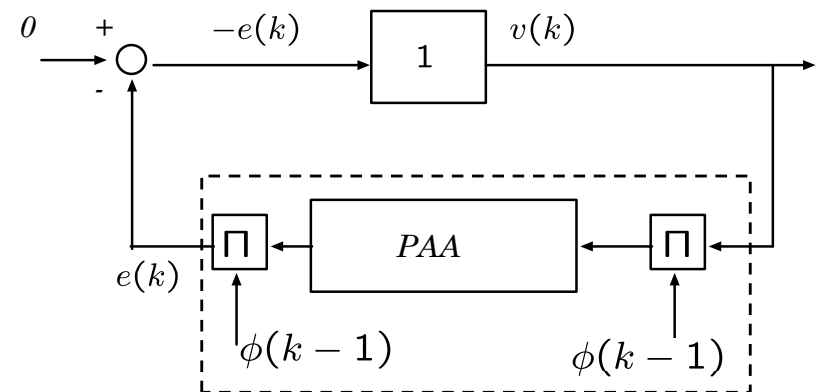
a-posteriori error: $e(k) = y(k) - \hat{y}(k)$

$$e(k) = \phi^T(k-1) \tilde{\theta}(k)$$

Parameter error update law: $\tilde{\theta}(k) = \theta - \hat{\theta}(k)$

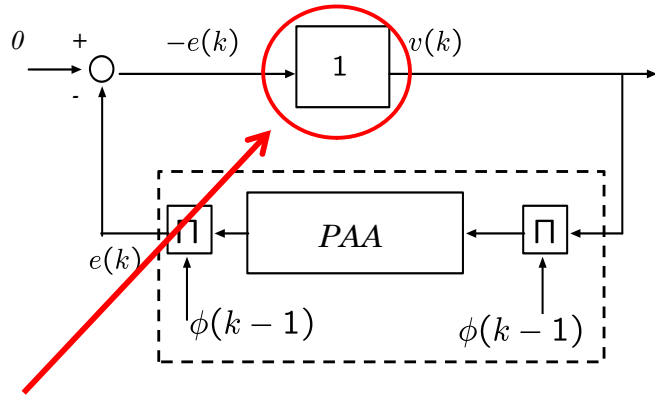
$$\tilde{\theta}(k) = \tilde{\theta}(k-1) - F \phi(k-1) e(k)$$

Series-Parallel ID Dynamics



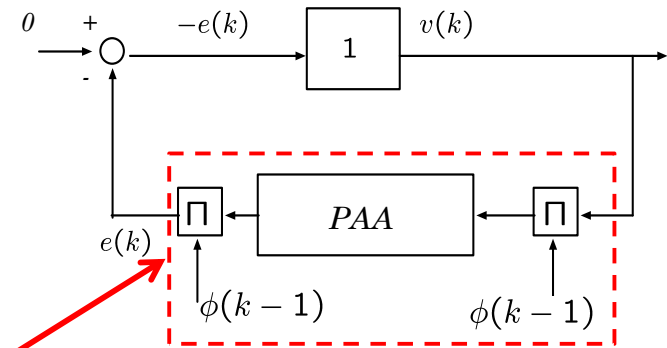
PAA: $\tilde{\theta}(k) = \tilde{\theta}(k-1) + F \phi(k-1) v(k)$

Stability analysis of Series-parallel ID



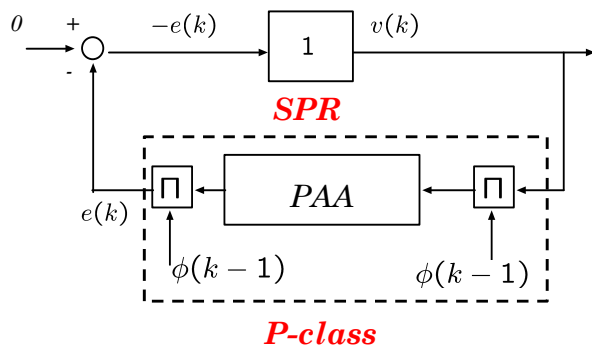
Strictly Positive Real

Stability analysis of Series-parallel ID



$$\begin{cases} \tilde{\theta}(k) = \tilde{\theta}(k-1) + F \phi(k-1)v(k) \\ e(k) = \phi^T(k-1)\tilde{\theta}(k) \end{cases} \Rightarrow \sum_{j=0}^k e^T(j)v(j) \geq -\gamma_o^2$$

Stability analysis of Series-parallel ID

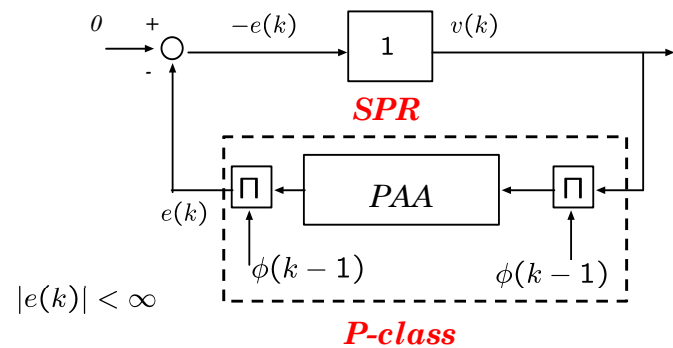


By the sufficiency portion of Hyperstability Theorem:

$$|v(k)| < \infty$$

$$|e(k)| < \infty$$

Stability analysis of Series-parallel ID



By the sufficiency portion of Asymptotic Hyperstability Theorem:

$$|v(k)| \rightarrow 0$$

$$|e(k)| \rightarrow 0$$

Q.E.D.

How to we implement the PAA?

a-posteriori error & PAA:

$$\left. \begin{aligned} e(k) &= \phi^T(k-1)\tilde{\theta}(k) \\ \tilde{\theta}(k) &= \tilde{\theta}(k-1) - F\phi(k-1)e(k) \end{aligned} \right\} \text{Static coupling}$$

Solution: Use the a-priori error

$$e^o(k) = \phi^T(k-1)\tilde{\theta}(k-1)$$

How to we implement the PAA?

$$\tilde{\theta}(k) = \tilde{\theta}(k-1) - F\phi(k-1)e(k)$$

Multiply by $\phi^T(k-1) = \phi_{k-1}^T$

$$\underbrace{\phi_{k-1}^T \tilde{\theta}(k)}_{e(k)} = \underbrace{\phi_{k-1}^T \tilde{\theta}(k-1)}_{e^o(k)} - \phi_{k-1}^T F \phi_{k-1} e(k)$$

$$e(k) = e^o(k) - \phi_{k-1}^T F \phi_{k-1} e(k)$$

Therefore,

$$e(k) = \frac{e^o(k)}{1 + \phi^T(k-1)F\phi(k-1)}$$

How to we implement the PAA?

$$e^o(k) = \phi^T(k-1)\tilde{\theta}(k-1)$$

$$e(k) = \frac{e^o(k)}{1 + \phi^T(k-1)F\phi(k-1)}$$

$$\tilde{\theta}(k) = \tilde{\theta}(k-1) - F\phi(k-1)e(k)$$

Stability analysis of Series-parallel ID

We have shown that

$$e(k) \rightarrow 0$$

Now we will shown that

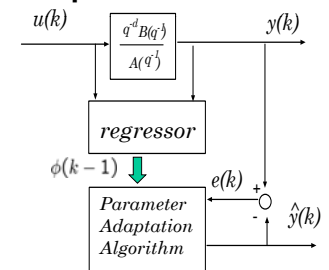
$$e^o(k) \rightarrow 0$$

Under the following assumptions:

$$u(k) < \infty \quad A(q^{-1}) \text{ is Schur}$$

Since $y(k) = \frac{q^{-d}B(q^{-1})}{A(q^{-1})}u(k) \Rightarrow y(k) < \infty$

Since $\phi(k-1) = \begin{bmatrix} y(k-1) \\ \vdots \\ u(k-d) \\ \vdots \end{bmatrix} \Rightarrow |\phi(k-1)| < \infty$



Stability analysis of Series-parallel ID

Thus,

$$e(k) \rightarrow 0$$

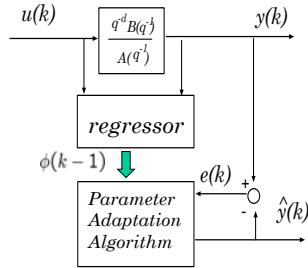
$$|\phi(k-1)| < \infty$$

remember that,

$$e(k) = \frac{e^o(k)}{1 + \phi^T(k-1)F\phi(k-1)}$$

$$\Rightarrow e^o(k) = \underbrace{e(k)}_{\rightarrow 0} \underbrace{\{1 + \phi^T(k-1)F\phi(k-1)\}}_{< \infty}$$

$$\Rightarrow e^o(k) \rightarrow 0$$



Stability analysis of Series-parallel ID

We have shown that

$$e(k) \rightarrow 0 \quad e^o(k) \rightarrow 0$$

$$|\phi(k-1)| < \infty$$

What about the parameter error $\tilde{\theta}(k)$?

since

$$\underbrace{e^o(k)}_{\rightarrow 0} = \phi^T(k-1)\tilde{\theta}(k-1) \quad \Rightarrow \quad |\phi^T(k)\tilde{\theta}(k)| \rightarrow 0$$

However, this **does not imply** that the parameter error goes to zero

We need to impose another condition on $u(k)$ (persistence of excitation) to guarantee that the parameter error goes to zero.

