

1.a)

$$\begin{aligned}x(k+1) &= Ax(k) + Bu(k) + B_w w(k) \\w(k) &= C_w x_w(k), \quad x_w(k+1) = A_w x_w(k) + B_n \eta(k) \\E\{x(0)\} &= x_o, \quad E\{(x(0) - x_o)(x(0) - x_o)^T\} = X_o \\E\{x_w(0)\} &= x_{wo}, \quad E\{(x_w(0) - x_{wo})(x_w(0) - x_{wo})^T\} = X_{wo} \\E\{\eta(k)\} &= 0, \quad E\{\eta(k)\eta(k+l)^T\} = \Gamma \delta(l) \\J &= \frac{1}{2} E \left\{ x^T(N) S x(N) + \sum_{k=0}^{N-1} [x^T(k) Q x(k) + u^T(k) R u(k)] \right\}\end{aligned}$$

If $x(k)$ and $x_w(k)$ are both measurable for all k , let extended state $x_e(k) = \begin{bmatrix} x(k) \\ x_w(k) \end{bmatrix}$

$$\begin{aligned}J &= \frac{1}{2} E \left\{ \begin{bmatrix} x(N) \\ x_w(N) \end{bmatrix}^T \begin{bmatrix} S & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(N) \\ x_w(N) \end{bmatrix} + \sum_{k=0}^{N-1} \left(\begin{bmatrix} x(k) \\ x_w(k) \end{bmatrix}^T \begin{bmatrix} Q & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(k) \\ x_w(k) \end{bmatrix} + u^T(k) R u(k) \right) \right\} \\x_e(k+1) &= \begin{bmatrix} x(k+1) \\ x_w(k+1) \end{bmatrix} = \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} \begin{bmatrix} x(k) \\ x_w(k) \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u(k) + \begin{bmatrix} 0 \\ B_n \end{bmatrix} \eta(k)\end{aligned}$$

Optimal control solution is as in deterministic LQR, $u^o(k) = -K(k+1)x_e(k) = -K(k+1) \begin{bmatrix} x(k) \\ x_w(k) \end{bmatrix}$

$$\begin{aligned}K(k+1) &= \left(R + \begin{bmatrix} B \\ 0 \end{bmatrix}^T P(k+1) \begin{bmatrix} B \\ 0 \end{bmatrix} \right)^{-1} \begin{bmatrix} B \\ 0 \end{bmatrix}^T P(k+1) \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} \\P(k-1) &= \begin{bmatrix} Q & 0 \\ 0 & 0 \end{bmatrix} + \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix}^T P(k) \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} - \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix}^T P(k) \begin{bmatrix} B \\ 0 \end{bmatrix} K(k)\end{aligned}$$

$P(N) = \begin{bmatrix} S & 0 \\ 0 & 0 \end{bmatrix}$, and assuming initial values $x(0)$ and $x_w(0)$ are uncorrelated, the optimal cost is:

$$\begin{aligned}J^o &= \frac{1}{2} \begin{bmatrix} x_o \\ x_{wo} \end{bmatrix}^T P(0) \begin{bmatrix} x_o \\ x_{wo} \end{bmatrix} + \frac{1}{2} \text{trace} \left(P(0) \begin{bmatrix} X_o & 0 \\ 0 & X_{wo} \end{bmatrix} \right) + b(0) \\b(k) &= b(k+1) + \text{trace} \left(\begin{bmatrix} 0 \\ B_n \end{bmatrix}^T P(k+1) \begin{bmatrix} 0 \\ B_n \end{bmatrix} \Gamma \right), \quad b(N) = 0\end{aligned}$$

1.b)

$$\begin{aligned}y(k) &= C x(k) + v(k), \quad E\{v(k)\} = 0, \quad E\{v(k)v(k+l)^T\} = V \delta(l) \\v(k) &\text{ independent from } x(0), \quad x_w(0), \text{ and } \eta(k)\end{aligned}$$

$$y(k) = \begin{bmatrix} C & 0 \end{bmatrix} \begin{bmatrix} x(k) \\ x_w(k) \end{bmatrix} + v(k)$$

Construct Kalman filter observer $\hat{x}_e(k) = \begin{bmatrix} \hat{x}(k) \\ \hat{x}_w(k) \end{bmatrix} = \hat{x}_e^o(k) + F(k) \tilde{y}^o(k)$

$$\hat{x}_e^o(k+1) = \begin{bmatrix} \hat{x}^o(k+1) \\ \hat{x}_w^o(k+1) \end{bmatrix} = \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} \hat{x}_e^o(k) + \begin{bmatrix} B \\ 0 \end{bmatrix} u(k)$$

$$\tilde{y}^o(k) = y(k) - \begin{bmatrix} C & 0 \end{bmatrix} \hat{x}_e^o(k)$$

$$E(k) = [C \ 0] M(k) [C \ 0]^T + V$$

$$F(k) = M(k) [C \ 0]^T E(k)^{-1}$$

$$Z(k) = M(k) - F(k) [C \ 0] M(k)$$

$$M(k+1) = \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} Z(k) \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix}^T + \begin{bmatrix} 0 \\ B_n \end{bmatrix} \Gamma \begin{bmatrix} 0 \\ B_n \end{bmatrix}^T$$

$$M(0) = \begin{bmatrix} X_o & 0 \\ 0 & X_{wo} \end{bmatrix}, \text{ optimal control based on a-posteriori state estimate: } u^o(k) = -K(k+1) \hat{x}_e(k)$$

where $K(k+1)$ is defined exactly as in part a.

$$\text{optimal cost } J^o = \hat{J}^o + \sum_{j=0}^{N-1} \text{trace} \left(\begin{bmatrix} Q & 0 \\ 0 & 0 \end{bmatrix} Z(j) \right) + \text{trace} \left(\begin{bmatrix} S & 0 \\ 0 & 0 \end{bmatrix} Z(N) \right)$$

$$\hat{J}^o = \frac{1}{2} \begin{bmatrix} x_o \\ x_{wo} \end{bmatrix}^T P(0) \begin{bmatrix} x_o \\ x_{wo} \end{bmatrix} + \frac{1}{2} \text{trace} \left(P(0) \begin{bmatrix} X_o & 0 \\ 0 & X_{wo} \end{bmatrix} \right) + \hat{b}(0)$$

$$\hat{b}(k) = \hat{b}(k+1) + \text{trace} [F(k+1)^T P(k+1) F(k+1) E(k+1)], \hat{b}(N) = 0$$

$$\hat{b}(k) = \hat{b}(k+1) + \text{trace} [F(k+1)^T P(k+1) M(k+1) [C \ 0]^T]$$

1.c)

If $x(k)$ is measurable for all k , we have $x(k) - Ax(k-1) - Bu(k-1) = B_w C_w x_w(k-1)$

Let $y_w(k-1) = B_w C_w x_w(k-1)$, so at step k we know $y_w(k-1)$ with zero measurement noise

$$x_w(k) = A_w x_w(k-1) + B_n \eta(k-1)$$

Construct Kalman filter observer just for $\hat{x}_w(k-1 | k) = \hat{x}_w^o(k-1) + F_w(k-1) \tilde{y}_w^o(k-1)$

$$\hat{x}_w^o(k) = A_w \hat{x}_w(k-1 | k), \tilde{y}_w^o(k-1) = y_w(k-1) - B_w C_w \hat{x}_w^o(k-1)$$

$$E_w(k-1) = B_w C_w M_w(k-1) C_w^T B_w^T + 0$$

$F_w(k-1) = M_w(k-1) C_w^T B_w^T E_w(k-1)^{\#}$, using the pseudoinverse since $E_w(k-1)$ may be singular

$$Z_w(k-1) = M_w(k-1) - F_w(k-1) B_w C_w M_w(k-1)$$

$$M_w(k) = A_w Z_w(k-1) A_w^T + B_n \Gamma B_n^T, \text{ initial condition } M_w(0) = X_{wo}$$

$$\text{Let } Y_w(k-1) = \{y_w(0), \dots, y_w(k-1)\}$$

$$E\{x_w(k) | Y_w(k-1)\} = E\{A_w x_w(k-1) + B_n \eta(k-1) | Y_w(k-1)\}$$

$$E\{x_w(k) | Y_w(k-1)\} = A_w E\{x_w(k-1) | Y_w(k-1)\} + B_n E\{\eta(k-1) | Y_w(k-1)\}$$

$$y_w(k-1) = B_w C_w x_w(k-1) = B_w C_w A_w x_w(k-2) + B_w C_w B_n \eta(k-2)$$

Therefore $y(k-1)$ and earlier don't contain any information about $\eta(k-1)$

$$\text{so } E\{\eta(k-1) | Y_w(k-1)\} = E\{\eta(k-1)\} = 0$$

$$\hat{x}_w(k | k) = E\{x_w(k) | Y_w(k-1)\} = A_w E\{x_w(k-1) | Y_w(k-1)\} = A_w \hat{x}_w(k-1 | k) = \hat{x}_w^o(k)$$

$$\text{LQR control based on measured } x(k) \text{ and estimated } \hat{x}_w^o(k), \text{ so } u^o(k) = -K(k+1) \begin{bmatrix} x(k) \\ \hat{x}_w^o(k) \end{bmatrix}$$

where again $K(k+1)$ is defined exactly as in part a. In the special circumstances of this problem with the one-step observation delay, the value used by the controller happens to be the a-priori estimate instead of the usual a-posteriori estimate. (I wonder if this also holds in more general cases?)

Luckily the cost $J = \frac{1}{2} E \left\{ x^T(N) S x(N) + \sum_{k=0}^{N-1} [x^T(k) Q x(k) + u^T(k) R u(k)] \right\}$ is equivalent to:

$$J = \frac{1}{2} E \left\{ \begin{bmatrix} x(N) \\ \hat{x}_w^o(N) \end{bmatrix}^T \begin{bmatrix} S & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(N) \\ \hat{x}_w^o(N) \end{bmatrix} + \sum_{k=0}^{N-1} \left(\begin{bmatrix} x(k) \\ \hat{x}_w^o(k) \end{bmatrix}^T \begin{bmatrix} Q & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(k) \\ \hat{x}_w^o(k) \end{bmatrix} + u^T(k) R u(k) \right) \right\}$$

$$\hat{x}_w^o(k) = A_w \hat{x}_w(k-1 | k) = A_w \hat{x}_w^o(k-1) + A_w F_w(k-1) \tilde{y}_w^o(k-1)$$

$$\begin{aligned}
\tilde{y}_w^o(k-1) &= y_w(k-1) - B_w C_w \hat{x}_w^o(k-1) = B_w C_w x_w(k-1) - B_w C_w \hat{x}_w^o(k-1) \\
x(k) &= A x(k-1) + B u(k-1) + B_w C_w x_w(k-1) = A x(k-1) + B u(k-1) + B_w C_w \hat{x}_w^o(k-1) + \tilde{y}_w^o(k-1) \\
\begin{bmatrix} x(k) \\ \hat{x}_w^o(k) \end{bmatrix} &= \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} \begin{bmatrix} x(k-1) \\ \hat{x}_w^o(k-1) \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u(k-1) + \begin{bmatrix} I \\ A_w F_w(k-1) \end{bmatrix} \tilde{y}_w^o(k-1) \\
\begin{bmatrix} x(k+1) \\ \hat{x}_w^o(k+1) \end{bmatrix} &= \begin{bmatrix} A & B_w C_w \\ 0 & A_w \end{bmatrix} \begin{bmatrix} x(k) \\ \hat{x}_w^o(k) \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u(k) + \begin{bmatrix} I \\ A_w F_w(k) \end{bmatrix} \tilde{y}_w^o(k) \\
E \{ \tilde{y}_w^o(k) \tilde{y}_w^o(k)^T \} &= E_w(k) = B_w C_w M_w(k) C_w^T B_w^T
\end{aligned}$$

So using this form, the optimal cost is $J^o = \frac{1}{2} \begin{bmatrix} x_o \\ x_{wo} \end{bmatrix}^T P(0) \begin{bmatrix} x_o \\ x_{wo} \end{bmatrix} + \frac{1}{2} \text{trace} \left(P(0) \begin{bmatrix} X_o & 0 \\ 0 & X_{wo} \end{bmatrix} \right) + b(0)$

$$b(k) = b(k+1) + \text{trace} \left(\begin{bmatrix} I \\ A_w F_w(k) \end{bmatrix}^T P(k+1) \begin{bmatrix} I \\ A_w F_w(k) \end{bmatrix} E_w(k) \right), \quad b(N) = 0$$

2.a)

$$Z_1(s) = \frac{1.367 \cdot 10^4}{s^2 + 134s + 4.52 \cdot 10^9} W(s)$$

$$s^2 Z_1(s) + 134s Z_1(s) + 4.52 \cdot 10^9 \cdot Z_1(s) = 1.367 \cdot 10^4 \cdot W(s)$$

Inverse Laplace transform, $\ddot{z}_1(t) + 134 \dot{z}_1(t) + 4.52 \cdot 10^9 \cdot z_1(t) = 1.367 \cdot 10^4 \cdot w(t)$

Assuming the given transfer function was for the suspension by itself, since we know $w(t)$ is in units of force we have $m_1 = 1/(1.367 \cdot 10^4)$, $b_1 = 134/(1.367 \cdot 10^4)$, $k_1 = 4.52 \cdot 10^9/(1.367 \cdot 10^4)$

For the actuator, $m \ddot{z}(t) + b(\dot{z}(t) - \dot{z}_1(t)) + k(z(t) - z_1(t)) = K_e u(t)$

$z(t) = z_1(t) + z_2(t)$, so $m(\ddot{z}_1(t) + \ddot{z}_2(t)) + b\dot{z}_2(t) + k z_2(t) = K_e u(t)$

$$2 \cdot 10^{-6} \ddot{z}_1(t) + 2 \cdot 10^{-6} \ddot{z}_2(t) + 5 \cdot 10^{-3} \dot{z}_2(t) + 400 z_2(t) = 2.5 \cdot 10^{-5} u(t)$$

Taking into account the reaction forces on the suspension from the actuator spring, damper, and input,

$$\ddot{z}_1(t) + 134 \dot{z}_1(t) + 4.52 \cdot 10^9 \cdot z_1(t) = 1.367 \cdot 10^4 \cdot [w(t) + 5 \cdot 10^{-3} \dot{z}_2(t) + 400 z_2(t) - 2.5 \cdot 10^{-5} u(t)]$$

$$\ddot{z}_1(t) = -134 \dot{z}_1(t) - 4.52 \cdot 10^9 \cdot z_1(t) + 1.367 \cdot 10^4 \cdot w(t) + 68.35 \dot{z}_2(t) + 5.468 \cdot 10^6 z_2(t) - 0.34175 u(t)$$

$$\ddot{z}_2(t) = -\ddot{z}_1(t) - 2500 \dot{z}_2(t) - 2 \cdot 10^8 z_2(t) + 12.5 u(t)$$

$$\ddot{z}_2(t) = 134 \dot{z}_1(t) + 4.52 \cdot 10^9 \cdot z_1(t) - 1.367 \cdot 10^4 \cdot w(t) - 2568.35 \dot{z}_2(t) - 2.05468 \cdot 10^8 z_2(t) + 12.84175 u(t)$$

Let $x(t) = [z_1(t) \quad z_2(t) \quad \dot{z}_1(t) \quad \dot{z}_2(t)]^T$, then $\dot{x}(t) = [\dot{z}_1(t) \quad \dot{z}_2(t) \quad \ddot{z}_1(t) \quad \ddot{z}_2(t)]^T$

$$\dot{x}(t) = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -4.52 \cdot 10^9 & 5.468 \cdot 10^6 & -134 & 68.35 \\ 4.52 \cdot 10^9 & -2.05468 \cdot 10^8 & 134 & -2568.35 \end{bmatrix} \begin{bmatrix} z_1(t) \\ z_2(t) \\ \dot{z}_1(t) \\ \dot{z}_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ -0.34175 \\ 12.84175 \end{bmatrix} u(t) + \begin{bmatrix} 0 \\ 0 \\ 1.367 \cdot 10^4 \\ -1.367 \cdot 10^4 \end{bmatrix} w(t)$$

$$y(t) = z(t) + v(t) = z_1(t) + z_2(t) + v(t) = [1 \quad 1 \quad 0 \quad 0] x(t) + v(t)$$

2.b)

TF from u to x is $G(s) = C(sI - A)^{-1}B = \frac{12.5s^2 + 1675s + 5.65 \cdot 10^{10}}{s^4 + 2702s^3 + 4.726 \cdot 10^9 s^2 + 1.133 \cdot 10^{13} s + 9.04 \cdot 10^{17}}$

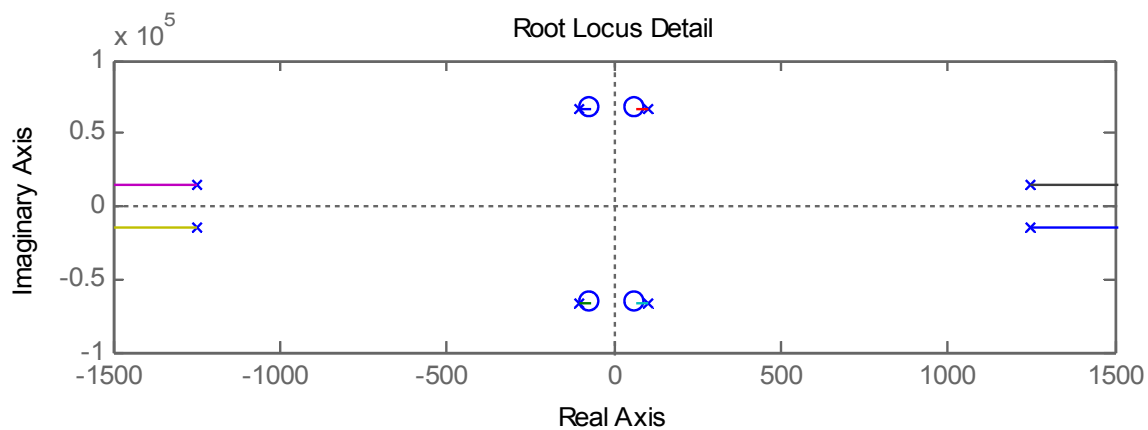
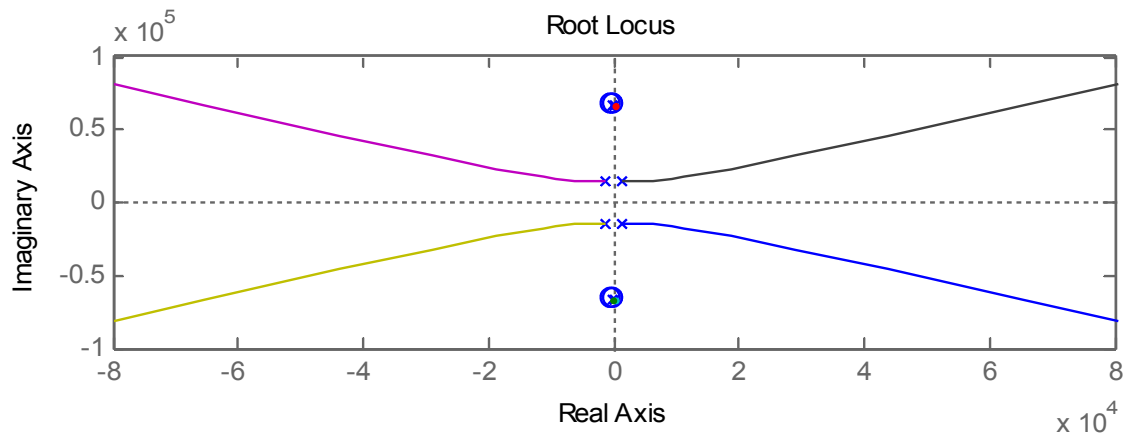
Symmetric root locus of $G(-s)G(s)$ shown on next page

2.c)

stationary LQG regulator minimizing cost $J = E\{z^2(t) + Ru^2(t)\}/2 = E\{x^T(t)C^T C x(t) + Ru^2(t)\}/2$

optimal controller $u^o(t) = -K \hat{x}(t)$, $K = R^{-1} B^T P$, $A^T P + P A + C^T C - P B R^{-1} B^T P = 0$

KF estimator $\dot{\hat{x}}(t) = A \hat{x}(t) + B u(t) + L(y(t) - C \hat{x}(t))$, $L = M C^T V^{-1}$



$$A M + M A^T = -B_w W B_w^T + M C^T V^{-1} C M, \quad W = E\{w(t)^2\} = 1.01 \cdot 10^{-7}, \quad V = E\{v(t)^2\} = 1.66 \cdot 10^{-14}$$

$$\text{Matlab says } M = \begin{bmatrix} 9.9388 \cdot 10^{-12} & -1.0381 \cdot 10^{-11} & 5.895 \cdot 10^{-12} & 2.7248 \cdot 10^{-8} \\ -1.0381 \cdot 10^{-11} & 1.0902 \cdot 10^{-11} & -2.7262 \cdot 10^{-8} & 8.1553 \cdot 10^{-12} \\ 5.895 \cdot 10^{-12} & -2.7262 \cdot 10^{-8} & 4.4979 \cdot 10^{-2} & -4.6987 \cdot 10^{-2} \\ 2.7248 \cdot 10^{-8} & 8.1553 \cdot 10^{-12} & -4.6987 \cdot 10^{-2} & 4.9168 \cdot 10^{-2} \end{bmatrix}, \quad L = \begin{bmatrix} -26.65 \\ 31.346 \\ -1.642 \cdot 10^6 \\ 1.642 \cdot 10^6 \end{bmatrix}$$

from $\text{care}(A^T, C^T, B_w W B_w^T, V)$

$$\text{For } R=1, \text{ Matlab gives } P = \begin{bmatrix} 2.288 \cdot 10^{-4} & 2.098 \cdot 10^{-4} & 6.7825 \cdot 10^{-12} & -1.0384 \cdot 10^{-10} \\ 2.098 \cdot 10^{-4} & 2.0674 \cdot 10^{-4} & 2.7321 \cdot 10^{-9} & 2.5062 \cdot 10^{-9} \\ 6.7825 \cdot 10^{-12} & 2.7321 \cdot 10^{-9} & 1.1008 \cdot 10^{-12} & 1.0502 \cdot 10^{-12} \\ -1.0384 \cdot 10^{-10} & 2.5062 \cdot 10^{-9} & 1.0502 \cdot 10^{-12} & 1.0037 \cdot 10^{-12} \end{bmatrix}$$

and $K = [-1.3358 \cdot 10^{-9} \quad 3.125 \cdot 10^{-8} \quad 1.311 \cdot 10^{-11} \quad 1.2531 \cdot 10^{-11}]$, both from $\text{care}(A, B, C^T C, R)$

optimal cost $J^o = \text{trace}\{P[B K M + B_w W B_w^T]\}$, for $R=1$ we have $J^o = 7.8156 \cdot 10^{-14}$

$$\text{For } R=10^{-10}, \text{ LQR result is } P = \begin{bmatrix} 2.2873 \cdot 10^{-4} & 2.0973 \cdot 10^{-4} & 7.6393 \cdot 10^{-12} & -1.0298 \cdot 10^{-10} \\ 2.0973 \cdot 10^{-4} & 2.0668 \cdot 10^{-4} & 2.7321 \cdot 10^{-9} & 2.5062 \cdot 10^{-9} \\ 7.6393 \cdot 10^{-12} & 2.7321 \cdot 10^{-9} & 1.1005 \cdot 10^{-12} & 1.0499 \cdot 10^{-12} \\ -1.0298 \cdot 10^{-10} & 2.5062 \cdot 10^{-9} & 1.0499 \cdot 10^{-12} & 1.0034 \cdot 10^{-12} \end{bmatrix}$$

and $K = [-13.251 \quad 312.5 \quad 0.13106 \quad 0.12527]$, with optimal cost $J^o = 7.8156 \cdot 10^{-14}$

2.d)

$$J^o = \text{trace}\{C_Q^T C_Q M + L^T P L V\}$$

First let $Q = C_Q^T C_Q$. Next, note that the matrices Q , M , and V are positive semidefinite.

$Q \geq 0$ by construction from C_Q , $M \geq 0$ since it is the solution of a Kalman filter Riccati equation, and $V \geq 0$ since it is a covariance matrix. These positive semidefinite matrices are symmetric and real-valued so they are each diagonalizable. Let $M = T_M \Lambda_M T_M^{-1}$ and $V = T_V \Lambda_V T_V^{-1}$

The transformations T_M and T_V of the symmetric real matrices are orthogonal: $T_M^{-1} = T_M^T$, $T_V^{-1} = T_V^T$
 $C_Q^T C_Q M = Q T_M \Lambda_M T_M^T$

The trace of a matrix is the sum of its eigenvalues, so the trace is invariant to a similarity transform.

$$\text{trace}\{C_Q^T C_Q M\} = \text{trace}\{T_M^T Q T_M \Lambda_M\}$$

Since Λ_M is diagonal, the columns with zero eigenvalues will result in columns of all zeros in the product matrix $T_M^T Q T_M \Lambda_M$, and those columns will not contribute to the trace.

So defining $\Lambda_M^{1/2}$ as the matrix with positive square roots of the eigenvalues of M on its diagonal, and $(\Lambda_M^{1/2})^\#$ as the pseudoinverse matrix with reciprocal values at the nonzero diagonals of $\Lambda_M^{1/2}$, but zeros everywhere else, we can take a pseudo-similarity transform without altering the trace.

$$\text{trace}\{C_Q^T C_Q M\} = \text{trace}\{\Lambda_M^{1/2} T_M^T Q T_M \Lambda_M^{1/2} (\Lambda_M^{1/2})^\#\} = \text{trace}\{\Lambda_M^{1/2} T_M^T Q T_M \Lambda_M^{1/2}\}$$

For any vector x , let $z = T_M \Lambda_M^{1/2} x$. Then $x^T \Lambda_M^{1/2} T_M^T Q T_M \Lambda_M^{1/2} x = z^T Q z$

By positive semidefiniteness $z^T Q z \geq 0$, so that implies $\Lambda_M^{1/2} T_M^T Q T_M \Lambda_M^{1/2}$ is positive semidefinite, and therefore has nonnegative eigenvalues and a nonnegative trace.

$$\text{trace}\{C_Q^T C_Q M\} = \text{trace}\{\Lambda_M^{1/2} T_M^T Q T_M \Lambda_M^{1/2}\} \geq 0$$

$$\text{Likewise } \text{trace}\{L^T P L V\} = \text{trace}\{L^T P L T_V \Lambda_V T_V^T\} = \text{trace}\{T_V^T L^T P L T_V \Lambda_V\}$$

Using corresponding definitions of $\Lambda_V^{1/2}$ and $(\Lambda_V^{1/2})^\#$ and the same zero-eigenvalue argument,

$$\text{trace}\{L^T P L V\} = \text{trace}\{\Lambda_V^{1/2} T_V^T L^T P L T_V \Lambda_V^{1/2} (\Lambda_V^{1/2})^\#\} = \text{trace}\{\Lambda_V^{1/2} T_V^T L^T P L T_V \Lambda_V^{1/2}\}$$

For any vector x , let $z = L T_V \Lambda_V^{1/2} x$. Then $x^T \Lambda_V^{1/2} T_V^T L^T P L T_V \Lambda_V^{1/2} x = z^T P z$

$z^T P z \geq 0$ by positive semidefiniteness of P , so $\Lambda_V^{1/2} T_V^T L^T P L T_V \Lambda_V^{1/2}$ is positive semidefinite

$$\text{trace}\{L^T P L V\} = \text{trace}\{\Lambda_V^{1/2} T_V^T L^T P L T_V \Lambda_V^{1/2}\} \geq 0$$

$$\text{For this example } \text{trace}\{C_Q^T C_Q M\} = 7.7945 \cdot 10^{-14}$$

$$\text{With } R=1, \text{ trace}\{L^T P L V\} = 2.1091 \cdot 10^{-16}$$

$$\text{With } R=10^{-10}, \text{ trace}\{L^T P L V\} = 2.1090 \cdot 10^{-16}$$

This is essentially at roundoff error, definitely two orders of magnitude smaller than the other term.

Difficult to tell for sure with only this data, but it appears as $R \rightarrow 0$, $J^o \rightarrow \text{trace}\{C_Q^T C_Q M\}$

2.e)

Now z_1 does not depend on z_2 or $u(t)$, so $\ddot{z}_1(t) + 134 \dot{z}_1(t) + 4.52 \cdot 10^9 \cdot z_1(t) = 1.367 \cdot 10^4 \cdot w(t)$

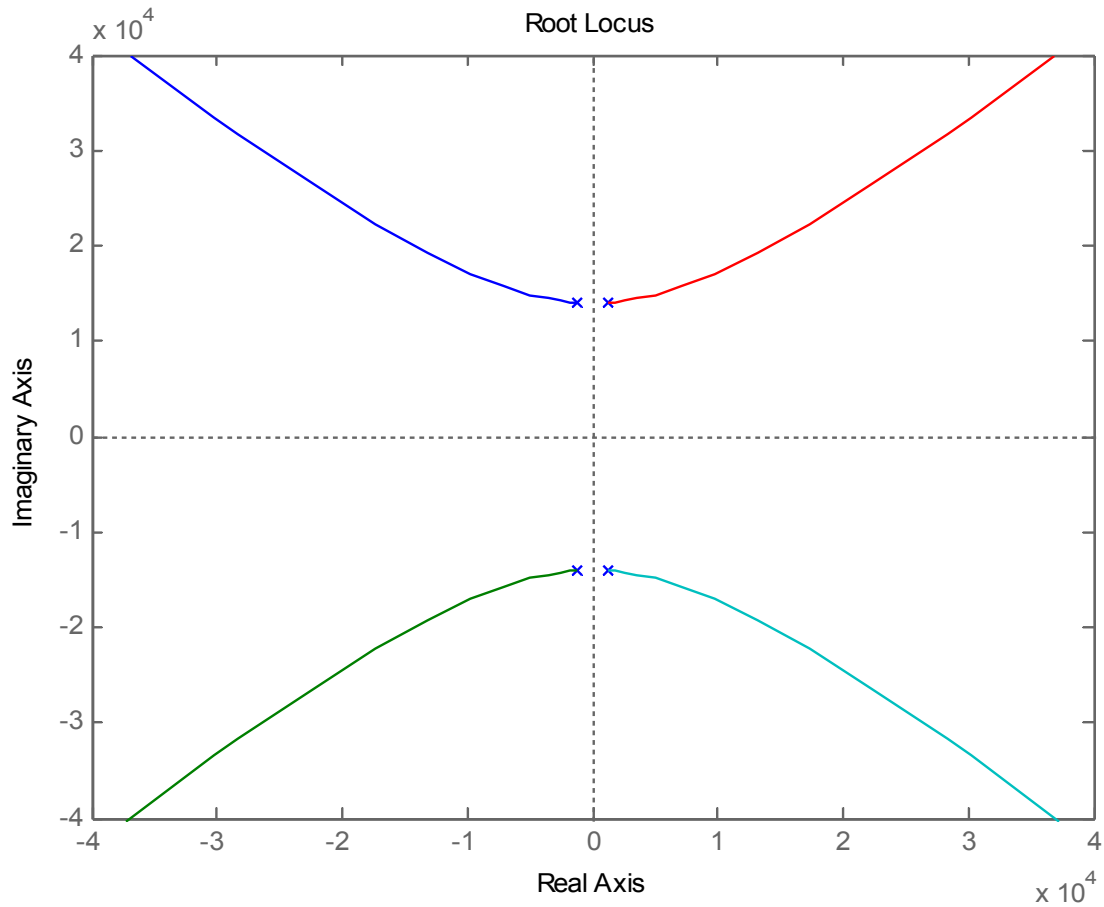
As before, $\ddot{z}_2(t) = -\ddot{z}_1(t) - 2500 \dot{z}_2(t) - 2 \cdot 10^8 z_2(t) + 12.5 u(t)$

$$\ddot{z}_2(t) = 134 \dot{z}_1(t) + 4.52 \cdot 10^9 z_1(t) - 1.367 \cdot 10^4 w(t) - 2500 \dot{z}_2(t) - 2 \cdot 10^8 z_2(t) + 12.5 u(t)$$

$$\dot{x}(t) = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -4.52 \cdot 10^9 & 0 & -134 & 0 \\ 4.52 \cdot 10^9 & -2 \cdot 10^8 & 134 & -2500 \end{bmatrix} \begin{bmatrix} z_1(t) \\ z_2(t) \\ \dot{z}_1(t) \\ \dot{z}_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 12.5 \end{bmatrix} u(t) + \begin{bmatrix} 0 \\ 0 \\ 1.367 \cdot 10^4 \\ -1.367 \cdot 10^4 \end{bmatrix} w(t)$$

$$\text{TF from } u \text{ to } x \text{ is } G(s) = C(sI - A)^{-1} B = \frac{12.5}{s^2 + 2500s + 2 \cdot 10^8}$$

Symmetric root locus of $G(-s)G(s)$ shown on next page



$$\text{Now } M = \begin{bmatrix} 1.5386 \cdot 10^{-11} & -1.6073 \cdot 10^{-11} & 1.4198 \cdot 10^{-11} & 4.2129 \cdot 10^{-8} \\ -1.6073 \cdot 10^{-11} & 1.6857 \cdot 10^{-11} & -4.2162 \cdot 10^{-8} & 1.8534 \cdot 10^{-11} \\ 1.4198 \cdot 10^{-11} & -4.2162 \cdot 10^{-8} & 6.9546 \cdot 10^{-2} & -7.2655 \cdot 10^{-2} \\ 4.2129 \cdot 10^{-8} & 1.8534 \cdot 10^{-11} & -7.2655 \cdot 10^{-2} & 7.6027 \cdot 10^{-2} \end{bmatrix}, L = \begin{bmatrix} -41.359 \\ 47.255 \\ -2.539 \cdot 10^6 \\ 2.539 \cdot 10^6 \end{bmatrix}$$

$$\text{For } R=1, P = \begin{bmatrix} 2.3319 \cdot 10^{-4} & 2.093 \cdot 10^{-4} & 6.9127 \cdot 10^{-12} & -1.0371 \cdot 10^{-10} \\ 2.093 \cdot 10^{-4} & 2.0625 \cdot 10^{-4} & 2.7258 \cdot 10^{-9} & 2.5 \cdot 10^{-9} \\ 6.9127 \cdot 10^{-12} & 2.7258 \cdot 10^{-9} & 1.098 \cdot 10^{-12} & 1.0464 \cdot 10^{-12} \\ -1.0371 \cdot 10^{-10} & 2.5 \cdot 10^{-9} & 1.0464 \cdot 10^{-12} & 1 \cdot 10^{-12} \end{bmatrix}$$

$$\text{and } K = [-1.2963 \cdot 10^{-9} \quad 3.125 \cdot 10^{-8} \quad 1.308 \cdot 10^{-11} \quad 1.25 \cdot 10^{-11}], J^o = 9.8598 \cdot 10^{-14}$$

$$\text{For } R=10^{-10}, \text{ LQR result is } P = \begin{bmatrix} 2.3312 \cdot 10^{-4} & 2.0924 \cdot 10^{-4} & 7.7655 \cdot 10^{-12} & -1.0285 \cdot 10^{-10} \\ 2.0924 \cdot 10^{-4} & 2.0619 \cdot 10^{-4} & 2.7258 \cdot 10^{-9} & 2.5 \cdot 10^{-9} \\ 7.7655 \cdot 10^{-12} & 2.7258 \cdot 10^{-9} & 1.0976 \cdot 10^{-12} & 1.046 \cdot 10^{-12} \\ -1.0285 \cdot 10^{-10} & 2.5 \cdot 10^{-9} & 1.046 \cdot 10^{-12} & 9.9968 \cdot 10^{-13} \end{bmatrix}$$

$$\text{and } K = [-12.857 \quad 312.5 \quad 0.13075 \quad 0.12496], \text{ with optimal cost } J^o = 9.8598 \cdot 10^{-14}$$

LQG results quite similar between the two models. Biggest difference was in the transfer function and therefore root locus as well. There was a pole/zero cancellation and a reduction of order.

3.a)

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ m \end{bmatrix} u(t) + \begin{bmatrix} 1 \\ 1 \end{bmatrix} w(t), \quad y(t) = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + v(t)$$

$$J = E \{ x^T(\tau) Q x(\tau) + u^2(\tau) \} = \lim_{T \rightarrow \infty} E \left\{ \frac{1}{T} \int_0^T [x^T(\tau) Q x(\tau) + u^2(\tau)] d\tau \right\}$$

$$Q = \rho \begin{bmatrix} 1 & 1 \\ 1 & 1 \end{bmatrix} = \rho \begin{bmatrix} 1 & 1 \\ 1 & 1 \end{bmatrix}, \quad R = 1, \quad \text{nominal } m_o = 1$$

$$\text{optimal controller } u(t) = -K_{LQ} x(t), \quad K_{LQ} = R^{-1} B^T P, \quad A^T P + P A + Q - P B R^{-1} B^T P = 0$$

$$0 = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} + \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} + \rho \begin{bmatrix} 1 & 1 \\ 1 & 1 \end{bmatrix} - \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} \begin{bmatrix} 0 \\ m_o \end{bmatrix} \begin{bmatrix} 0 & m_o \end{bmatrix} \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix}$$

$$0 = \begin{bmatrix} p_{11} & p_{12} \\ p_{11} + p_{12} & p_{12} + p_{22} \end{bmatrix} + \begin{bmatrix} p_{11} & p_{11} + p_{12} \\ p_{12} & p_{12} + p_{22} \end{bmatrix} + \rho \begin{bmatrix} 1 & 1 \\ 1 & 1 \end{bmatrix} - \begin{bmatrix} p_{12} m_o \\ p_{22} m_o \end{bmatrix} \begin{bmatrix} m_o p_{12} & m_o p_{22} \end{bmatrix}$$

$$0 = \begin{bmatrix} 2 p_{11} + \rho & p_{11} + 2 p_{12} + \rho \\ p_{11} + 2 p_{12} + \rho & 2 p_{12} + 2 p_{22} + \rho \end{bmatrix} - \begin{bmatrix} p_{12}^2 & p_{12} p_{22} \\ p_{12} p_{22} & p_{22}^2 \end{bmatrix}$$

$$p_{11} = (p_{12}^2 - \rho)/2, \quad p_{11} + 2 p_{12} + \rho - p_{12} p_{22} = 0, \quad p_{12} = -p_{22} + (p_{22}^2 - \rho)/2$$

$$p_{11} = (p_{22}^2 - p_{22}(p_{22}^2 - \rho) + (p_{22}^2 - \rho)^2/4 - \rho)/2$$

$$p_{22}^2/2 - p_{22}(p_{22}^2 - \rho)/2 + (p_{22}^2 - \rho)^2/8 - \rho/2 + (-p_{22} + (p_{22}^2 - \rho)/2)(2 - p_{22}) + \rho = 0$$

$$p_{22}^2/2 - p_{22}^3/2 + \rho p_{22}/2 + p_{22}^4/8 - \rho p_{22}^2/4 + \rho^2/8 - \rho/2 - 2 p_{22} + p_{22}^2 - \rho + p_{22}^2 - p_{22}^3/2 + \rho p_{22}/2 + \rho = 0$$

$$p_{22}^4/8 - p_{22}^3 + (5/2 - \rho/4) p_{22}^2 + (\rho - 2) p_{22} + \rho^2/8 - \rho/2 = 0$$

$$p_{22}^4 - 8 p_{22}^3 + (20 - 2\rho) p_{22}^2 + (8\rho - 16) p_{22} + \rho^2 - 4\rho = 0$$

$$(p_{22}^2 - 4 p_{22} - \rho)(p_{22}^2 - 4 p_{22} + 4 - \rho) = 0$$

$$p_{22} = 2 \pm \sqrt{4 + \rho} \quad \text{or} \quad p_{22} = 2 \pm \sqrt{\rho}$$

$$p_{12} = -p_{22} + (p_{22}^2 - \rho)/2, \quad \text{for the possible solutions } p_{22} = 2 \pm \sqrt{\rho} \quad \text{we have } p_{12} = \pm \sqrt{\rho}$$

$$p_{11} = (p_{12}^2 - \rho)/2, \quad \text{for the possible solutions } p_{22} = 2 \pm \sqrt{\rho} \quad \text{we have } p_{11} = 0$$

$$\det(sI - P) = (s - p_{11})(s - p_{22}) - p_{12}^2 = s^2 - (2 \pm \sqrt{\rho})s - \rho$$

$$\lambda = (2 + \sqrt{\rho} \pm \sqrt{(2 + \sqrt{\rho})^2 + 4\rho})/2 \quad \text{or} \quad \lambda = (2 - \sqrt{\rho} \pm \sqrt{(2 - \sqrt{\rho})^2 + 4\rho})/2$$

and both $p_{22} = 2 \pm \sqrt{\rho}$ options have 2 eigenvalues of different signs, not positive definite

$$p_{12} = -p_{22} + (p_{22}^2 - \rho)/2, \quad \text{for the possible solutions } p_{22} = 2 \pm \sqrt{4 + \rho} \quad \text{we have } p_{12} = p_{22} = 2 \pm \sqrt{4 + \rho}$$

$$p_{11} = (p_{12}^2 - \rho)/2, \quad \text{for the possible solutions } p_{22} = 2 \pm \sqrt{4 + \rho} \quad \text{we have } p_{11} = 2 p_{22} = 4 \pm 2 \sqrt{4 + \rho}$$

$$\det(sI - P) = (s - p_{11})(s - p_{22}) - p_{12}^2 = s^2 - 3 p_{22} s + p_{22}^2$$

$$\lambda = (3 p_{22} \pm \sqrt{5 p_{22}^2})/2, \quad \text{both eigenvalues same sign as } p_{22}$$

$$p_{22} = 2 + \sqrt{4 + \rho} \quad \text{is therefore the positive definite solution, } P = \begin{bmatrix} 4 + 2\sqrt{4 + \rho} & 2 + \sqrt{4 + \rho} \\ 2 + \sqrt{4 + \rho} & 2 + \sqrt{4 + \rho} \end{bmatrix}$$

$$K_{LQ} = R^{-1} B^T P = \begin{bmatrix} 0 & m_o \end{bmatrix} \begin{bmatrix} 4 + 2\sqrt{4 + \rho} & 2 + \sqrt{4 + \rho} \\ 2 + \sqrt{4 + \rho} & 2 + \sqrt{4 + \rho} \end{bmatrix} = m_o [2 + \sqrt{4 + \rho} \quad 2 + \sqrt{4 + \rho}]$$

$$K_{LQ} = (2 + \sqrt{4 + \rho}) [1 \quad 1] = \alpha(\rho) [1 \quad 1]$$

3.b)

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \left(\begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} 0 \\ m \end{bmatrix} (2 + \sqrt{4 + \rho}) [1 \quad 1] \right) \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ -2m - m\sqrt{4 + \rho} & 1 - 2m - m\sqrt{4 + \rho} \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}$$

$$\text{Eigenvalues given by } (s - 1)(s - 1 + 2m + m\sqrt{4 + \rho}) + 2m + m\sqrt{4 + \rho} = 0$$

$$s^2 + (-2 + 2m + m\sqrt{4+\rho})s + 1 = 0$$

$\sqrt{4+\rho} \geq 2$ so for $m > 0.5$ the middle coefficient is positive, and the real parts of the eigenvalues will be negative, giving a stable closed-loop system

$$\text{For } \rho = 5 \text{ and } m = 1, s^2 + \sqrt{9}s + 1 = 0$$

$$\lambda = (-3 \pm \sqrt{9-4})/2 = (-3 \pm \sqrt{5})/2 < 0$$

3.c)

$$\frac{d}{dt} \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ m \end{bmatrix} u(t) + \begin{bmatrix} L_1 \\ L_2 \end{bmatrix} \tilde{y}(t), \quad \tilde{y}(t) = y(t) - [1 \quad 0] \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix}$$

$$A M + M A^T = -B_w W B_w^T + M C^T V^{-1} C M, \quad W = 1, \quad V = \sigma$$

$$\text{Note that } B_w W B_w^T = \begin{bmatrix} 1 \\ 1 \end{bmatrix} [1 \quad 1] = \begin{bmatrix} 1 & 1 \\ 1 & 1 \end{bmatrix}, \text{ and let } M = \begin{bmatrix} M_{11} & M_{12} \\ M_{12}^T & M_{22} \end{bmatrix}$$

$$\frac{d}{dt} \begin{bmatrix} \hat{x}_2(t) \\ \hat{x}_1(t) \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} \hat{x}_2(t) \\ \hat{x}_1(t) \end{bmatrix} + \begin{bmatrix} m \\ 0 \end{bmatrix} u(t) + \begin{bmatrix} L_2 \\ L_1 \end{bmatrix} \tilde{y}(t), \quad \tilde{y}(t) = y(t) - [0 \quad 1] \begin{bmatrix} \hat{x}_2(t) \\ \hat{x}_1(t) \end{bmatrix}$$

$$\text{Let } M_{\text{flip}} = \begin{bmatrix} M_{22} & M_{12}^T \\ M_{12} & M_{11} \end{bmatrix}, \quad C_{\text{flip}} = [0 \quad 1]$$

$$A^T M_{\text{flip}} + M_{\text{flip}} A = -B_w W B_w^T + M_{\text{flip}} C_{\text{flip}}^T V^{-1} C_{\text{flip}} M_{\text{flip}}$$

recall the LQR Riccati equation from part a was: $A^T P + P A + Q - P B R^{-1} B^T P = 0$

$B = C_{\text{flip}}^T$ for $m = 1$, so if we multiply every term by σ^{-1} we get the same Riccati equation:

$$A^T (\sigma^{-1} M_{\text{flip}}) + (\sigma^{-1} M_{\text{flip}}) A + B_w \sigma^{-1} B_w^T - (\sigma^{-1} M_{\text{flip}}) C_{\text{flip}}^T C_{\text{flip}} (\sigma^{-1} M_{\text{flip}}) = 0$$

Where $Q = B_w \sigma^{-1} B_w^T$ for $\rho = \sigma^{-1}$

$$\sigma^{-1} M_{\text{flip}} = \sigma^{-1} \begin{bmatrix} M_{22} & M_{12}^T \\ M_{12} & M_{11} \end{bmatrix} = P = \begin{bmatrix} 4 + 2\sqrt{4+\sigma^{-1}} & 2 + \sqrt{4+\sigma^{-1}} \\ 2 + \sqrt{4+\sigma^{-1}} & 2 + \sqrt{4+\sigma^{-1}} \end{bmatrix}$$

$$M = \begin{bmatrix} M_{11} & M_{12} \\ M_{12}^T & M_{11} \end{bmatrix} = \sigma \begin{bmatrix} 2 + \sqrt{4+\sigma^{-1}} & 2 + \sqrt{4+\sigma^{-1}} \\ 2 + \sqrt{4+\sigma^{-1}} & 4 + 2\sqrt{4+\sigma^{-1}} \end{bmatrix}$$

$$K_{KF} = M C^T V^{-1} = \sigma \begin{bmatrix} 2 + \sqrt{4+\sigma^{-1}} & 2 + \sqrt{4+\sigma^{-1}} \\ 2 + \sqrt{4+\sigma^{-1}} & 4 + 2\sqrt{4+\sigma^{-1}} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \sigma^{-1} = \begin{bmatrix} 2 + \sqrt{4+\sigma^{-1}} \\ 2 + \sqrt{4+\sigma^{-1}} \end{bmatrix} = \beta \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

3.d)

Assuming expectations over all noise variables in the following:

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ m \end{bmatrix} u(t) = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} - \begin{bmatrix} 0 \\ m \end{bmatrix} \alpha [1 \quad 1] \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix}$$

$$\frac{d}{dt} \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ m_o \end{bmatrix} u(t) + \begin{bmatrix} L_1 \\ L_2 \end{bmatrix} \tilde{y}(t), \quad \tilde{y}(t) = y(t) - [1 \quad 0] \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix}$$

$$\frac{d}{dt} \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} - \begin{bmatrix} 0 \\ m_o \end{bmatrix} \alpha [1 \quad 1] \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} + \beta \begin{bmatrix} 1 \\ 1 \end{bmatrix} \left([1 \quad 0] \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} - [1 \quad 0] \begin{bmatrix} \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} \right)$$

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \\ \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 & 0 \\ 0 & 1 & -m\alpha & -m\alpha \\ \beta & 0 & 1-\beta & 1 \\ \beta & 0 & -\beta-m_o\alpha & 1-m_o\alpha \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 & 0 \\ 0 & 1 & -m\alpha & -m\alpha \\ \beta & 0 & 1-\beta & 1 \\ \beta & 0 & -\beta-\alpha & 1-\alpha \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ \hat{x}_1(t) \\ \hat{x}_2(t) \end{bmatrix}$$

3.e)

For $\rho=\sigma=5$, $m=1$, $\alpha=2+\sqrt{4+5}=5$, $\beta=2+\sqrt{4+5^{-1}}=4.0494$

the eigenvalues of the part d matrix in this case are -2.618 , -1.2483 , -0.80109 , and -0.38197 .

the eigenvalues of full-state LQR from part b were -2.618 and -0.38197 so the separation principle holds in this $m=1$ example

3.f)

With $m=1.1$, the eigenvalues of the above matrix are -3.4846 , $-0.83368 \pm 1.4736i$, and 0.10258 .

The eigenvalues are not where we expected them to be from the LQR and KF results, especially since we now have a complex pair, and even worse one of the closed-loop eigenvalues is unstable.