

⑥ $\left[\left(\frac{s}{w_p} \right)^2 + 2 \zeta \left(\frac{s}{w_p} \right) + 1 \right]$ terms in the denominator of $H(s)$.

4 possible scenarios: $w_p = \text{natural frequency}$, $\zeta = \text{damping factor/coefficient}$
 also related to $Q = \text{quality factor} = \frac{1}{2\zeta}$

1) $\zeta = 0, \Rightarrow \text{undamped}$; roots $s_{1,2} = \pm j w_p \Leftrightarrow \text{oscillatory, unstable}$
 $Q \rightarrow \infty$ (not covered in this course)

2) $\zeta = 1, Q = 0.5 \Rightarrow \text{critically damped}$; real, repeated roots; this is just like having $(1 + s/w_m)^2$ in the denominator

3) $\zeta > 1, 0 < Q < 0.5 \Rightarrow \text{overdamped}$; two distinct, real roots; this is just like having $(1 + s/w_{m_1})(1 + s/w_{m_2})$ in the denominator

4) $0 < \zeta < 1, 0.5 < Q < \infty \Rightarrow \text{underdamped}$; complex-conjugate pole pair
 ↳ this is the main focus here; for example, consider transfer

function $H(s) = \frac{30(s+10)}{s^2 + 3s + 50} = \frac{(30)(10)(1 + 4/s_0)}{(50)[s^2/50 + 3s/50 + 1]}$ (→ putting $H(s)$ into proper form)

⇒ $H(s) = \frac{6/(1 + 4/s_0)}{(s/\sqrt{50})^2 + 2\zeta(s/\sqrt{50}) + 1} \Rightarrow w_p = \sqrt{50} \text{ rps}, \zeta = ?$

⇒ Set $\frac{3s}{50} = 2\zeta(s/\sqrt{50}) \Rightarrow \zeta = \frac{3w_p}{2w_p^2} = \frac{3\sqrt{50}}{2(50)} \approx 0.2121$ (unless)

⇒ Note that $\zeta = 0.2121$ satisfies $0 < \zeta < 1 \Rightarrow \text{underdamped}$; so we have a complex-conjugate pole pair in the denominator of transfer function $H(s)$.

Poles are $s_{1,2} = -\zeta w_p \pm w_p \sqrt{\zeta^2 - 1} = -\zeta w_p \pm j w_p \sqrt{1 - \zeta^2}$

⇒ in this particular example, we have $w_p \approx 7.07$ rps, $\zeta \approx 0.2121$ with poles $s_{1,2} \approx -1.5 \pm j 6.91$ rps & $Q \approx 2.3570$



OK, so how do we sketch a Bode approximation of a complex-conjugate pole pair?

⇒ Let's start with Bode approximation of the magnitude response:

(here, $w_0 = w_p$)

$$|H(j\omega)| = \left| \frac{1}{(\omega/\omega_0)^2 + 2\zeta(\omega/\omega_0) + 1} \right| = \frac{1}{\sqrt{(1 - (\omega/\omega_0)^2)^2 + (2\zeta\omega/\omega_0)^2}}$$

$$\Rightarrow |H(j\omega)|_{dB} = 20 \log_{10} \left(\frac{1}{\sqrt{(1 - (\omega/\omega_0)^2)^2 + (2\zeta\omega/\omega_0)^2}} \right)$$

$$= -20 \log_{10} \left(\sqrt{(1 - (\omega/\omega_0)^2)^2 + (2\zeta\omega/\omega_0)^2} \right)$$



Now, let's consider 3 ranges of ω values:

- 1) $\underbrace{\omega \ll \omega_0}_{\text{low-frequencies}}$, 2) $\underbrace{\omega \gg \omega_0}_{\text{high-frequencies}}$, 3) $\underbrace{\omega \approx \omega_0}_{\omega \approx \text{natural frequency}}$

1) Low-frequencies, $\omega \ll \omega_0$: $|H|_{dB} \approx -20 \log_{10}(1) = 0 \text{ dB} \Leftarrow$ makes sense, like a "regular" pole.

2) High-frequencies, $\omega \gg \omega_0$: $|H|_{dB} \approx -20 \log_{10}((\omega/\omega_0)^2) = -40 \log_{10}(\omega/\omega_0) \text{ dB}$

⇒ the high-frequency approximation is a straight-line with a slope of -40 dB/decade ($\approx -12 \text{ dB/octave}$), going through ω_0 at $0 \text{ dB} \Leftarrow -40 \text{ dB/decade}$ is as expected, since a 2nd-order denominator term has (essentially) 2 poles activated at high frequencies, at $-20 \text{ dB/decade per pole}$.

3) $\omega \approx \omega_0$: It can be shown that a magnitude peak occurs near the natural frequency ω_0 ; here, we make the approximation that a peak exists only when $0 < \zeta < 0.5$, or alternatively $1 < Q < \infty$, and that the peak occurs at ω_0 with a height of $Q = \frac{1}{2\zeta}$ (\Leftarrow not in dB!); to be clear, no appreciable peak at ω_0 for $0.5 \leq \zeta < 1$, or alternatively $0.5 \leq Q \leq 1$. \rightsquigarrow

* To draw a piecewise-linear Bode approximation of the magnitude response (in dB!) for a complex-conjugate pole pair: Use the low-frequency asymptote (0 dB) up to the natural (or corner, or break) frequency and use the high-frequency asymptote (with slope $-40 \text{ dB/decade} \approx -12 \text{ dB/octave}$) thereafter. If $0 < \zeta < 0.5$, or alternatively $1 < Q < \infty$, then draw a peak of amplitude $20 \log_{10}(Q) = Q \text{ dB}$; draw a smooth curve between the low- and high-frequency asymptote that goes through the peak value. (It can also be shown that the -3dB frequency will be $\omega_0 \sqrt{x^*}$, where $x^* = 1 - 2\zeta^2 + \sqrt{(2\zeta^2 - 1)^2 + 1}$.)

⇒ Now, let's show the details of sketching a Bode approximation of the phase response for a complex-conjugate pole pair:

$$\begin{aligned} \angle H(j\omega) &= \angle \frac{1}{(\omega/\omega_0)^2 + 2\zeta(\omega/\omega_0) + 1} = -\angle (1 - (\omega/\omega_0)^2 + j2\zeta(\omega/\omega_0)) \\ &= -\tan^{-1} \left(\frac{2\zeta\omega/\omega_0}{1 - (\omega/\omega_0)^2} \right) \\ &\quad (\pm 180^\circ, \text{ if } 1 - (\omega/\omega_0)^2 < 0) \end{aligned}$$

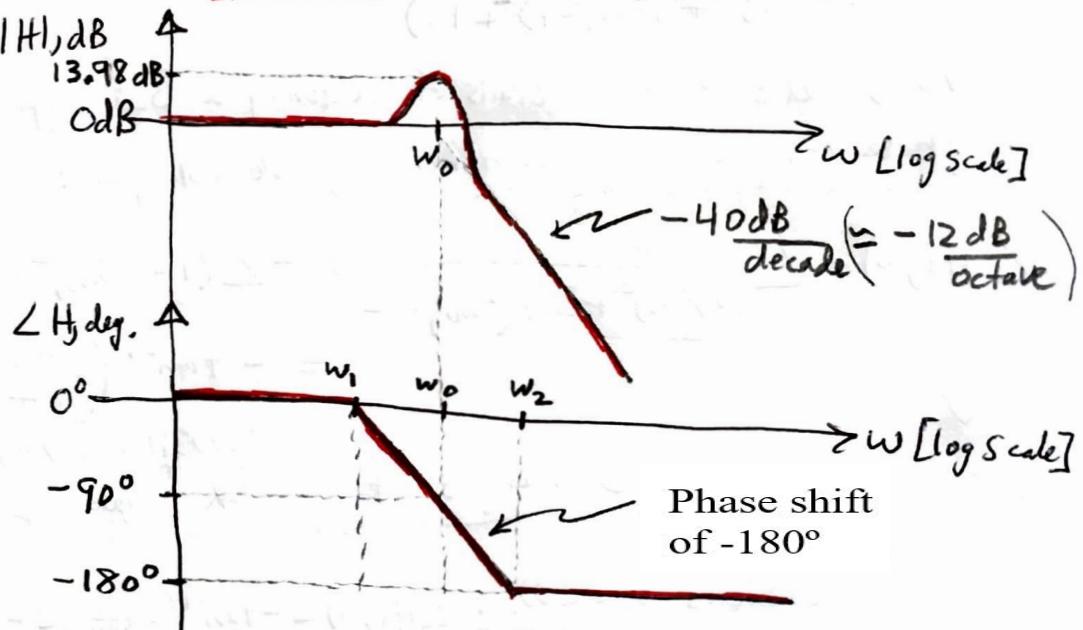
Again, we consider 3 ranges of angular frequency ω values:

- 1) Low-frequencies, $\omega \ll \omega_0$: $\angle H(j\omega) \approx -\tan^{-1}(2\zeta\frac{\omega}{\omega_0}) \approx -\tan^{-1}(0) = 0^\circ$ as expected (like a "regular" pole)
- 2) High-frequencies, $\omega \gg \omega_0$: $\angle H(j\omega) \approx -180^\circ$ as expected, for (essentially) 2 poles activated at high frequencies
- 3) $\omega = \omega_0$: $\angle H(j\omega_0) = -90^\circ$ as expected, since one pole has a phase roll-off of -45° at the pole frequency, and here we have two poles

* There are several approaches to drawing a piecewise-linear Bode approximation of the phase response (in degrees!) for a complex-conjugate pole pair; here is my suggestion: Connect low-frequency asymptote at 0° to the high-frequency asymptote at -180° starting at $\omega_1 = \frac{\omega_0}{10^{\frac{1}{4}}} = \omega_0 10^{-\frac{1}{4}}$ and ending at $\omega_2 = \omega_0 10^{\frac{1}{4}}$; ↗

the slope of the line between ω_1 and ω_2 will depend on the ω_1 and ω_2 values, but the total phase roll-off from ω_1 to ω_2 will be -180° . If $\zeta < 0.02$, just draw a vertical line at ω_0 (i.e., omit ω_1 and ω_2 and draw a vertical line at ω_0).

An example: Suppose $H(s) = \frac{1}{(\frac{s}{w_0})^2 + 2\zeta(\frac{s}{w_0}) + 1}$, with 4
 $\zeta = 0.1$ and $w_0 = 10 \text{ rps}$. Sketch Bode approximation of the frequency response, $H(j\omega)$. \Rightarrow Note that $0 < \zeta < 0.5$, so there will be an appreciable peak in the magnitude response at w_0 of Q_{dB} ; what is Q ?
 $\Rightarrow Q = \frac{1}{2\zeta} = \frac{1}{2(0.1)} = 5$, and $Q_{dB} = 20 \log_{10}(Q) \approx 13.98 \text{ dB}$.
 $\Rightarrow w_1 = \frac{10}{10^{0.1}} \approx 7.94 \text{ rps}$ & $w_2 = 10(10^{0.1}) \approx 12.59 \text{ rps}$
 \Rightarrow Here is the Bode approximation sketch of $H(j\omega)$:



(7) $[(\frac{s}{w_0})^2 + 2\zeta(\frac{s}{w_0}) + 1]$ terms in the numerator of $H(s)$:

To Bode approximate this kind of term, the magnitude and phase responses flip around in the usual, expected way, when we take a term in the denominator & move it to the numerator; consider the previous example (with $w_0 = 10 \text{ rps}$, $\zeta = 0.1$), with the complex-conjugate pole pair in the numerator of $H(s)$ instead of the denominator:

