

⑥ $\left[\left(\frac{s}{\omega_p} \right)^2 + 2\zeta \left(\frac{s}{\omega_p} \right) + 1 \right]$ terms in the denominator of $H(s)$.

4 possible scenarios: $\left\langle \begin{array}{l} \omega_p = \text{natural frequency, } \zeta = \text{damping factor/coefficient,} \\ \text{also related to } Q = \text{quality factor} = \frac{1}{2\zeta} \end{array} \right\rangle$

1) $\zeta = 0, Q \rightarrow \infty \Rightarrow$ undamped; roots $s_{1,2} = \pm j\omega_p \Leftrightarrow$ oscillatory, unstable (not covered in this course)

2) $\zeta = 1, Q = 0.5 \Rightarrow$ critically damped; real, repeated roots; this is just like having $(1 + s/\omega_m)^2$ in the denominator

3) $\zeta > 1, 0 < Q < 0.5 \Rightarrow$ overdamped; two distinct, real roots; this is just like having $(1 + s/\omega_{m1})(1 + s/\omega_{m2})$ in the denominator

4) $0 < \zeta < 1, 0.5 < Q < \infty \Rightarrow$ underdamped; complex-conjugate pole pair
 \Rightarrow this is the main focus here; for example, consider transfer

function $H(s) = \frac{30(s+10)}{s^2 + 3s + 50} = \frac{(30)(10)(1 + s/10)}{(50)[\frac{s^2}{50} + \frac{3s}{50} + 1]}$ (\Leftarrow putting $H(s)$ into proper form by approximation)

$\Rightarrow H(s) = \frac{\overset{\text{K term}}{\underbrace{6}_{\text{zero term}}}(1 + s/10)}{(\frac{s}{\sqrt{50}})^2 + 2\zeta(\frac{s}{\sqrt{50}}) + 1} \Rightarrow \omega_p = \sqrt{50} \text{ rps}, \zeta = ?$

\Rightarrow Set $\frac{3s}{50} = 2\zeta(\frac{s}{\sqrt{50}}) \Rightarrow \zeta = \frac{3\omega_p}{2\omega_p^2} = \frac{3\sqrt{50}}{2(50)} \approx 0.2121$ (unitless)

\Rightarrow Note that $\zeta = 0.2121$ satisfies $0 < \zeta < 1 \Rightarrow$ underdamped; so we have a complex-conjugate pole pair in the denominator of transfer function $H(s)$;

poles are $s_{1,2} = -\frac{\zeta}{\omega_p} \omega_p \pm j\omega_p \sqrt{1 - \zeta^2}$

\Rightarrow in this particular example, we have $\omega_p \approx 7.071 \text{ rps}$, $\zeta \approx 0.2121$ with poles $s_{1,2} \approx -1.5 \pm j6.91 \text{ rps}$ & $Q \approx 2.3570$

\rightarrow

OK, So how do we sketch a Bode approximation of a complex-conjugate pole pair?

⇒ Let's start with Bode approximation of the magnitude response:

$$\begin{aligned} \text{(here, } \omega_0 = \omega_p) \quad |H(j\omega)| &= \left| \frac{1}{(j\omega/\omega_0)^2 + 2\zeta(j\omega/\omega_0) + 1} \right| = \frac{1}{|-(\omega/\omega_0)^2 + j2\zeta(\omega/\omega_0) + 1|} \\ &= \frac{1}{\sqrt{(1 - (\omega/\omega_0)^2)^2 + (2\zeta\omega/\omega_0)^2}} \end{aligned}$$

$$\begin{aligned} \Rightarrow |H(j\omega)|_{dB} &= 20 \log_{10} \left(\frac{1}{\sqrt{(1 - (\omega/\omega_0)^2)^2 + (2\zeta\omega/\omega_0)^2}} \right) \\ &= -20 \log_{10} \left(\sqrt{(1 - (\omega/\omega_0)^2)^2 + (2\zeta\omega/\omega_0)^2} \right) \end{aligned}$$



Now, let's consider 3 ranges of ω values:

- 1) $\omega \ll \omega_0$, 2) $\omega \gg \omega_0$, 3) $\omega \approx \omega_0$
low-frequencies , high-frequencies , $\omega \approx$ natural frequency

1) Low-frequencies, $\omega \ll \omega_0$: $|H|_{dB} \approx -20 \log_{10}(1) = 0 \text{ dB} \leftarrow$ makes sense, like a "regular" pole.

2) High-frequencies, $\omega \gg \omega_0$: $|H|_{dB} \approx -20 \log_{10}((\omega/\omega_0)^2) = -40 \log_{10}(\omega/\omega_0) \text{ dB}$

⇒ the high-frequency approximation is a straight-line with a slope of -40 dB/decade ($\approx -12 \text{ dB/octave}$), going through ω_0 at $0 \text{ dB} \leftarrow -40 \text{ dB/decade}$ is as expected, since a 2nd-order denominator term has (essentially) 2 poles activated at high frequencies, at -20 dB/decade per pole.

3) $\omega \approx \omega_0$: It can be shown that a magnitude peak occurs near the natural frequency ω_0 ; here, we make the approximation that a peak exists only when $0 < \zeta < 0.5$, or alternatively $1 < Q < \infty$, and that the peak occurs at ω_0 with a height of $Q = \frac{1}{2\zeta}$ (\leftarrow not in dB!); to be clear, no appreciable peak at ω_0 for $0.5 \leq \zeta < 1$, or alternatively $0.5 \leq Q \leq 1$.

⊛ To draw a piecewise-linear Bode approximation of the magnitude response (in dB!) for a complex-conjugate pole pair: Use the low-frequency asymptote (0dB) up to the natural (or corner, or break) frequency and use the high-frequency asymptote (with slope $-40\text{dB/decade} \approx -12\text{dB/octave}$) thereafter. If $0 < \zeta < 0.5$, or alternatively $1 < Q < \infty$, then draw a peak of amplitude $20 \log_{10}(Q) = Q\text{dB}$; draw a smooth curve between the low- and high-frequency asymptote that goes through the peak value. (It can also be shown that the -3dB frequency will be $\omega_0 \sqrt{X^*}$, where $X^* = 1 - 2\zeta^2 + \sqrt{(2\zeta^2 - 1)^2 + 1}$.)

⇒ Now, let's show the details of sketching a Bode approximation of the phase response for a complex-conjugate pole pair:

$$\angle H(j\omega) = \angle \frac{1}{(j\omega/\omega_0)^2 + 2\zeta(j\omega/\omega_0) + 1} = -\angle(1 - (\omega/\omega_0)^2 + j2\zeta(\omega/\omega_0))$$

$$\Downarrow$$

$$= -\tan^{-1}\left(\frac{2\zeta\omega/\omega_0}{1 - (\omega/\omega_0)^2}\right)$$

($\pm 180^\circ$, if $1 - (\omega/\omega_0)^2 < 0$)

Again, we consider 3 ranges of angular frequency ω values:

- 1) Low-frequencies, $\omega \ll \omega_0$: $\angle H(j\omega) \approx -\tan^{-1}(2\zeta\frac{\omega}{\omega_0}) \approx -\tan^{-1}(0) = 0^\circ \leftarrow \text{as expected (like a "regular" pole)}$
- 2) High-frequencies, $\omega \gg \omega_0$: $\angle H(j\omega) \approx -180^\circ \leftarrow \text{as expected, for (essentially) 2 poles activated at high frequencies}$
- 3) $\omega = \omega_0$: $\angle H(j\omega_0) = -90^\circ \leftarrow \text{as expected, since one pole has a phase rolloff of } -45^\circ \text{ at the pole frequency, and here we have two poles}$

⊛ There are several approaches to drawing a piecewise-linear Bode approximation of the phase response (in degrees!) for a complex-conjugate pole pair; here is my suggestion: Connect low-frequency asymptote at 0° to the high-frequency asymptote at -180° starting at $\omega_1 = \frac{\omega_0}{10^{\frac{1}{2}}} = \omega_0 10^{-\frac{1}{2}}$ and ending at $\omega_2 = \omega_0 10^{\frac{1}{2}}$; \Rightarrow

the slope of the line between ω_1 and ω_2 will depend on the ω_1 and ω_2 values, but the total phase roll-off from ω_1 to ω_2 will be -180° . If $\zeta < 0.02$, just draw a vertical line at ω_0 (i.e., omit ω_1 and ω_2 and draw a vertical line at ω_0).

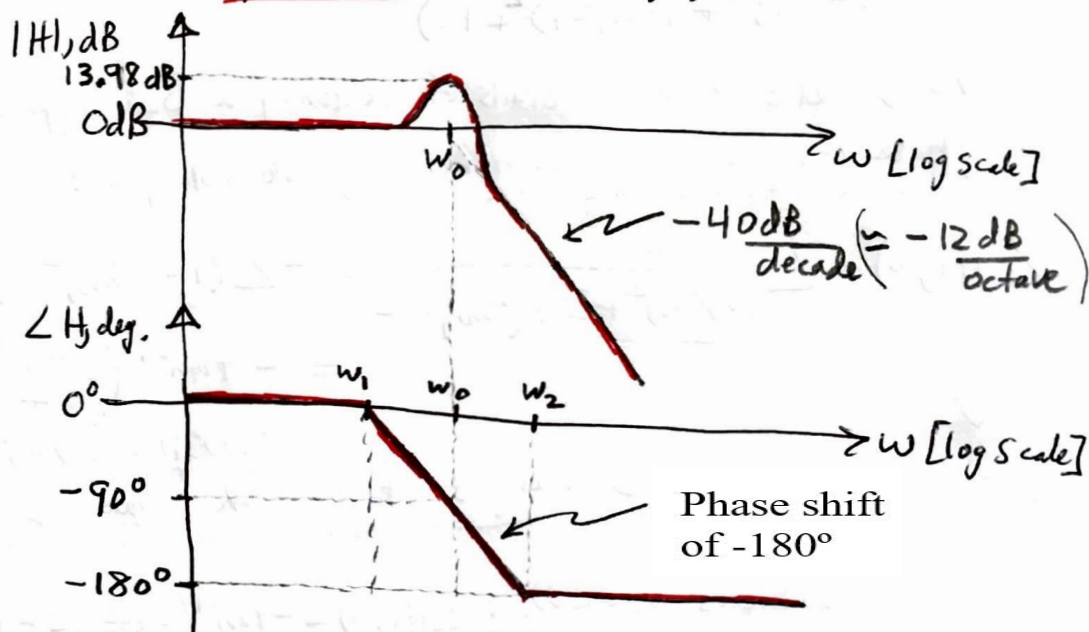
An example: Suppose $H(s) = \frac{1}{(s/\omega_0)^2 + 2\zeta(s/\omega_0) + 1}$, with $\zeta = 0.1$ and $\omega_0 = 10$ rps. Sketch Bode approximation of the frequency response, $H(j\omega)$. 4

\Rightarrow Note that $0 < \zeta < 0.5$, so there will be an appreciable peak in the magnitude response at ω_0 of Q dB; what is Q ?

$$\Rightarrow Q = \frac{1}{2\zeta} = \frac{1}{2(0.1)} = 5, \text{ and } Q_{dB} = 20 \log_{10}(Q) \approx 13.98 \text{ dB}.$$

$$\Rightarrow \omega_1 = \frac{10}{10^{0.1}} \approx 7.94 \text{ rps} \quad \& \quad \omega_2 = 10(10^{0.1}) \approx 12.59 \text{ rps}$$

\Rightarrow Here is the Bode approximation sketch of $H(j\omega)$:



⑦ $[(s/\omega_0)^2 + 2\zeta(s/\omega_0) + 1]$ terms in the numerator of $H(s)$:

To Bode approximate this kind of term, the magnitude and phase responses flip around in the usual, expected way, when we take a term in the denominator & move it to the numerator; consider the previous example (with $\omega_0 = 10$ rps, $\zeta = 0.1$), with the complex-conjugate pole pair in the numerator of $H(s)$ instead of the denominator:

