



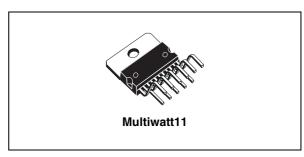
10 + 10 W stereo amplifier for car radio

Features

- Low distortion
- Low noise
- Protection against:
 - Output AC short circuit to ground
 - Overrating chip temperature
 - Load dump voltage surge
 - Fortuitous open ground
 - Very inductive loads

Description

The TDA2004R is a class B dual audio power amplifier in Multiwatt11 package specifically designed for car radio applications.



Power booster amplifiers can be easily designed using this device that provides a high current capability (up to 3.5 A) and can drive very low impedance loads (down to 1.6 Ω).

The TDA2004R allows very compact applications because few external components are required and it doesn't need electrical insulation between the package and the heatsink.

Table 1. Device summary

Order code	Package	Packing
TDA2004R	Multiwatt11	Tube

Contents TDA2004R

Contents

1	Pins	descrip	ption	5
2	Elec	trical sp	pecifications	6
	2.1	Absolu	ute maximum ratings	6
	2.2	Therm	al data	6
	2.3	Electric	cal characteristics	6
	2.4	Test ar	nd application circuit	8
	2.5	Electric	cal characteristics curves	9
3	Appl	lication	suggestion	12
	3.1	Built-in	n protection systems	12
		3.1.1	Load dump voltage surge	12
		3.1.2	Short circuit (AC condition)	13
		3.1.3	Polarity inversion	13
		3.1.4	Open ground	13
		3.1.5	Inductive load	13
		3.1.6	DC voltage	13
		3.1.7	Thermal shut-down	14
4	Pack	age info	ormation	15
5	Revi	sion his	story	16

TDA2004R List of tables

List of tables

Table 1.	Device summary	1
	Absolute maximum ratings	
Table 3.	Thermal data	
Table 4.	Electrical characteristics	6
Table 5.	Recommended values of the component of the application circuit	. 12
Table 6.	Document revision history	. 16

List of figures TDA2004R

List of figures

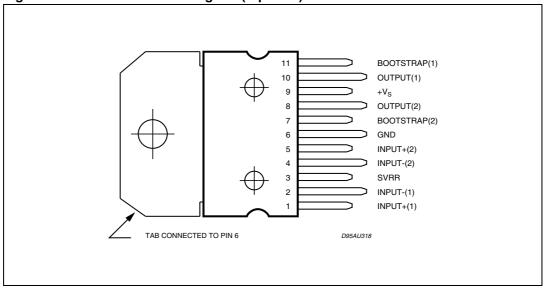
Figure 1.	Pins connection diagram (top view)	5
Figure 2.	Test and application circuit	
Figure 3.	Printed circuit board and components layout of the figure 2	
Figure 4.	Quiescent output voltage vs. supply voltage	
Figure 5.	Quiescent drain current vs. supply voltage	
Figure 6.	Distortion vs. output power	9
Figure 7.	Output power vs. supply voltage, $R_L = 2$ and $4 \Omega \dots \Omega$	
Figure 8.	Output power vs. supply voltage, $R_{\rm l} = 1.6$ and 3.2Ω	
Figure 9.	Distortion vs. frequency, $R_L = 2$ and $4 \Omega \dots$	9
Figure 10.	Distortion vs. frequency, $R_L = 1.6$ and 3.2Ω	10
Figure 11.	Supply voltage rejection vs. C3	10
Figure 12.	Supply voltage rejection vs. frequency	10
Figure 13.	Supply voltage rejection vs. C2 and C3, $G_V = 390/1\Omega$	10
Figure 14.	Supply voltage rejection vs. C2 and C3, $G_V = 1000/10\Omega$	10
Figure 15.	Gain vs. input sensitivity	10
Figure 16.	Total power dissipation and efficiency vs. output power ($R_L = 2 \Omega$)	11
Figure 17.	Total power dissipation and efficiency vs. output power ($R_L = 3.2 \Omega$)	11
Figure 18.	Maximum allowable power dissipation vs. ambient temperature	11
Figure 19.	Suggested LC network circuit	
Figure 20.	Voltage gain bridge configuration	13
Figure 21	Multiwatt11 mechanical data and package dimensions	15

4/17 Doc ID 17614 Rev 2

TDA2004R Pins description

1 Pins description

Figure 1. Pins connection diagram (top view)



2 Electrical specifications

2.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
	Operating supply voltage	18	
V _S	DC supply voltage	28	V
	Peak supply voltage (50 ms)	40	
lo ⁽¹⁾	Output peak current (non repetitive t = 0.1 ms)	4.5	۸
10(1)	Output peak current (repetitive f ≥10 Hz)	3.5	А
P _{tot}	Power dissipation at T _{case} = 60 °C	30	W
T _{stg} , T _j	Storage and junction temperature	-40 to 150	°C

^{1.} The max. output current is internally limited.

2.2 Thermal data

Table 3. Thermal data

Symbol	Parameter		Value	Unit
R _{th-j-case}	Thermal resistance junction-to-case	max	3	°C/W

2.3 Electrical characteristics

Refer to the stereo application circuit T_{amb} = 25 °C; G_v = 50 dB; $R_{th(heatsink)}$ = 4 °C/W unless otherwise specified

Table 4. Electrical characteristics

Symbol	Parameter	Test condition	Min.	Тур.	Max.	Unit
V _S	Supply voltage		8		18	V
V _o	Quiescent offset voltage	V _S = 14.4 V V _S = 13.2 V	6.6 6	7.2 6.6	7.8 7.2	V V
I _d	Total quiescent drain current	V _S = 14.4 V V _S = 13.2 V	-	65 62	120 120	mA mA
P _o	Output power (each channel)	$f = 1 \text{ kHz}; \text{ THD} = 10 \%$ $V_S = 14.4 \text{ V}; \text{ R}_L = 4 \Omega$ $V_S = 14.4 \text{ V}; \text{ R}_L = 3.2 \Omega$ $V_S = 14.4 \text{ V}; \text{ R}_L = 2 \Omega$ $V_S = 14.4 \text{ V}; \text{ R}_L = 1.6 \Omega$	6 7 9 10	6.5 8 10 11	-	W

Table 4. Electrical characteristics (continued)

Symbol	Parameter Test condition		Min.	Тур.	Max.	Unit
P _o	Output power (each channel)	$f = 1 \text{ kHz; THD} = 10 \%$ $V_{S} = 13.2 \text{ V; R}_{L} = 3.2 \Omega$ $V_{S} = 13.2 \text{ V; R}_{L} = 1.6 \Omega$ $V_{S} = 16 \text{ V; R}_{L} = 2 \Omega$	6 9	6.5 10 ⁽¹⁾ 12	-	W
		$f = 1 \text{ kHz}; V_S = 14.4 \text{ V};$ $R_L = 4 \Omega; P_O = 50 \text{ mW to } 4 \text{ W};$	-	0.2	1	%
TUD	Total house onic alieta stice.	$f = 1 \text{ kHz}; V_S = 14.4 \text{ V};$ $R_L = 2 \Omega; P_0 = 50 \text{ mW to } 6 \text{ W};$	-	0.3	1	%
THD	Total harmonic distortion	$f = 1 \text{ kHz}; V_S = 13.2 \text{ V}; \\ R_L = 3.2 \Omega; P_o = 50 \text{ mW to 3W};$	-	0.2	1	%
		$f = 1KHz; V_S = 13.2V;$ $R_L = 1.6Ω; P_o = 40mW to 6W;$	-	0.3	1	%
СТ	Cross talk	$V_S = 14.4 \text{ V; } V_o = 4 \text{ V}_{RMS};$ $R_g = 5 \text{ k}\Omega; \text{ R}_L = 4 \Omega;$ $f = 1 \text{ kHz}$ $f = 10 \text{ kHz}$	50 40	60 45	-	mW mW
V _i	Input saturation voltage	-	300		-	mW
R _i	Input resistance	f = 1 kHz	70	200	-	kΩ
f _L	Low frequency roll off (-3 dB)	$R_L = 4 \Omega$ $R_L = 2 \Omega$ $R_L = 3.2 \Omega$ $R_L = 1.6 \Omega$	-	-	35 50 40 55	Hz
f _H	High frequency roll off (-3 dB)	$R_L = 1.6 \Omega \text{ to } 4 \Omega$	15	_	-	kHz
•••	Open loop voltage gain	f = 1 kHz	-	90	_	
G_{v}	Closed loop voltage gain	f = 1 kHz	48	50	51	dB
ΔG _v	Closed loop gain matching	-	-	0.5	-	dB
e _N	Total input noise voltage	$R_{q} = 10 \text{ k}\Omega^{(2)}$	-	1.5	5	μV
SVR	Supply voltage rejection	V_{ripple} = 0.5 Vrms; f_{ripple} =100 Hz; R _g = 10 kΩ; C_3 = 10 μF	35	45	-	dB
		$f = 1 \text{ kHz}; V_S = 14.4 \text{ V};$ $R_L = 4 \Omega; P_o = 6.5 \text{ W};$ $R_L = 2\Omega; P_o = 10 \text{ W};$	-	70 60	-	%
η	Efficiency	$f = 1 \text{ kHz}; V_S = 13.2 \text{ V};$ $R_L = 3.2 \Omega; P_0 = 6.5 \text{ W};$ $R_L = 1.6 \Omega; P_0 = 10 \text{ W};$	-	70 60	-	%
T_J	Thermal shutdown junction temperature	-	-	145	-	°C

^{1. 9.3} W without bootstrap.

577

^{2.} Bandwidth filter: 22 Hz to 22 kHz.

2.4 Test and application circuit

Figure 2. Test and application circuit

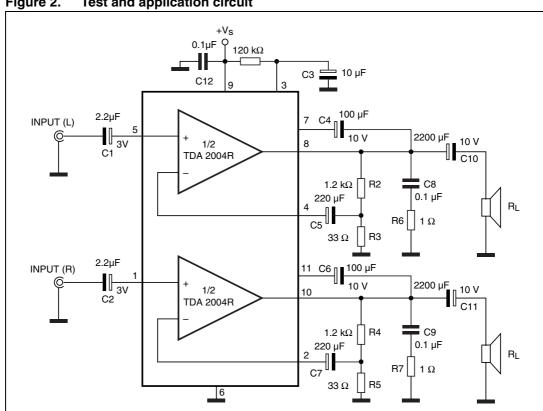
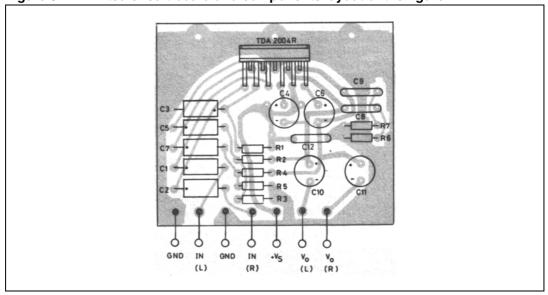


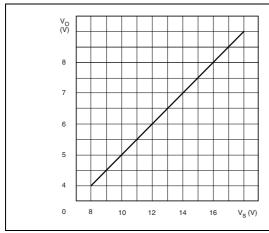
Figure 3. Printed circuit board and components layout of the figure 2



8/17 Doc ID 17614 Rev 2

2.5 Electrical characteristics curves

Figure 4. Quiescent output voltage vs. sup- Figure 5. Quiescent drain current vs. supply ply voltage voltage



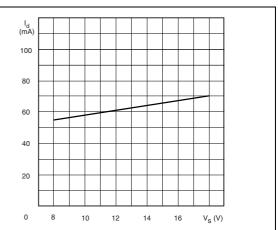
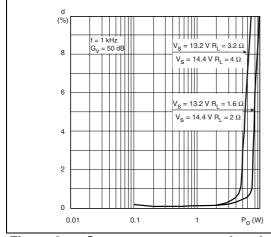


Figure 6. Distortion vs. output power

Figure 7. Output power vs. supply voltage, R_L = 2 and 4 Ω



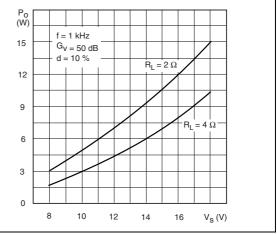
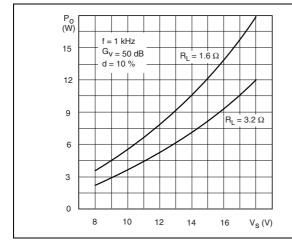
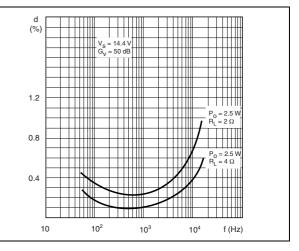


Figure 8. Output power vs. supply voltage, $R_L = 1.6$ and 3.2Ω

Figure 9. Distortion vs. frequency, R_L = 2 and 4 Ω





577

Figure 10. Distortion vs. frequency, R_L = 1.6 Figure 11. Supply voltage rejection vs. C3 and 3.2 Ω

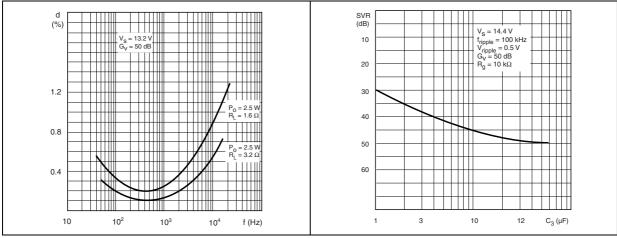


Figure 12. Supply voltage rejection vs. frequency

Figure 13. Supply voltage rejection vs. C2 and C3, $G_V = 390/1\Omega$

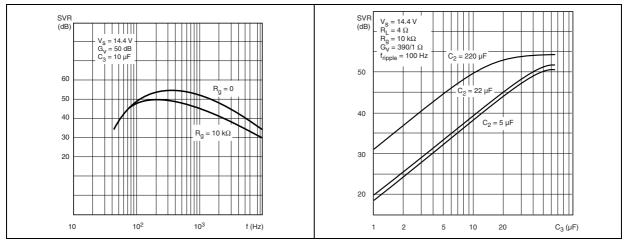
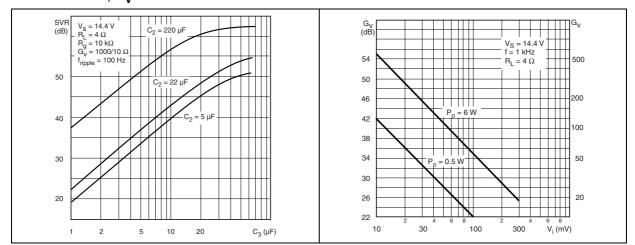


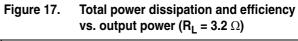
Figure 14. Supply voltage rejection vs. C2 and C3, $G_V = 1000/10\Omega$

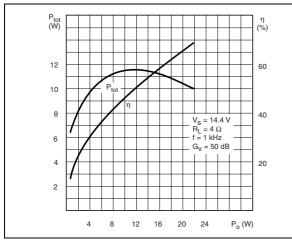
Figure 15. Gain vs. input sensitivity



10/17 Doc ID 17614 Rev 2

Figure 16. Total power dissipation and efficiency vs. output power ($R_L = 2 \Omega$)





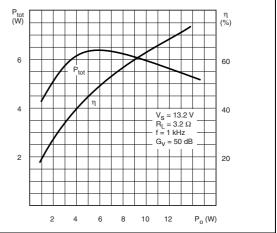
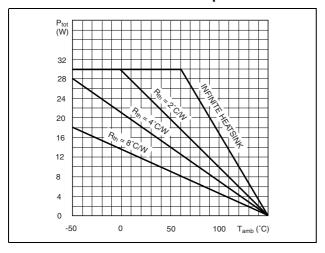


Figure 18. Maximum allowable power dissipation vs. ambient temperature



3 Application suggestion

The recommended values of the components are those shown on application circuit of *Figure 2*. Different values can be used; the following table can help the designer.

Table 5. Recommended values of the component of the application circuit

Component	Recommended value	Purpose	Larger than	Smaller than r
R1	120 kΩ	Optimization of the output signal symmetry	Smaller P _{omax}	Smaller P _{omax}
R2, R4	1 kΩ	Closed loop gain setting	Increase of gain	Decrease of gain
R3, R5	3.3 Ω	(1)	Decrease of gain	Increase of gain
R6, R7	1 Ω	Frequency stability	Danger of oscillation at high frequency with inductive load	
C1, C2	2.2 μF	Input DC decoupling	High turn-on delay	High turn-on pop, higher low frequency cutoff. Increase of noise
C3	10 μF	Ripple rejection	Increase of SVR, Increase of the switch-on time	Degradation of SVR
C4, C6	100 μF	Bootstrapping	-	Increase of distortion at low frequency
C5, C7	100 μF	Feedback input DC decoupling	-	-
C8, C9	0.1 μF	Frequency stability	-	Danger of oscillation
C10, C11	1000 to 2200 μF	Output DC decoupling	-	Higher low-frequency cut-off

^{1.} The closed loop gain must be higher than 26 dB.

3.1 Built-in protection systems

3.1.1 Load dump voltage surge

The TDA2004R has a circuit which enables it to withstand voltage pulse train, on Pin 9, of the type shown in *Figure 20*. If the supply voltage peaks to more than 40 V, then an LC filter must be inserted between the supply and pin 9, in order to assure that the pulses at pin 9 will be held within the limits shown.

A suggested LC network is shown in *Figure 19*. With this network, a train of pulses with amplitude up to 120 V and width of 2 ms can be applied at point A. This type of protection is ON when the supply voltage (pulse or DC) exceeds 18 V. For this reason the maximum operating supply voltage is 18 V.

12/17 Doc ID 17614 Rev 2

Figure 19. Suggested LC network circuit

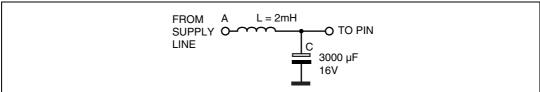
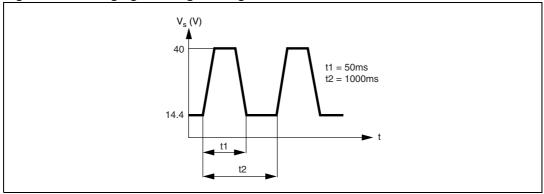


Figure 20. Voltage gain bridge configuration



3.1.2 Short circuit (AC condition)

The TDA2004R can withstand a permanent short-circuit from the output to ground caused by a wrong connection during normal working.

3.1.3 Polarity inversion

High current (up to 10 A) can be handled by the device with no damage for a longer period than the blow-out time of a quick 2 A fuse (normally connected in series with the supply). This feature is added to avoid destruction, if during fitting to the car, a mistake on the connection of the supply is made.

3.1.4 Open ground

When the ratio is in the ON condition and the ground is accidentally opened, a standard audio amplifier will be damaged. On the TDA2004R protection diodes are included to avoid any damage.

3.1.5 Inductive load

A protection diode is provided to allow use of the TDA2004R with inductive loads.

3.1.6 DC voltage

The maximum operating DC voltage for the TDA2004R is 18 V. However the device can withstand a DC voltage up to 28 V with no damage. This could occur during winter if two batteries are series connected to crank the engine.

3.1.7 Thermal shut-down

The presence of a thermal limiting circuit offers the following advantages:

- 1. an overload on the output (even if it is permanent), or an excessive ambient temperature can be easily withstood.
- 2. the heatsink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of excessive junction temperature: all that happens is that P_o (and therefore P_{tot}) and I_d are reduced.

The maximum allowable power dissipation depends upon the size of the external heatsink (i.e. its thermal resistance); *Figure 18* shows the power dissipation as a function of ambient temperature for different thermal resistance.

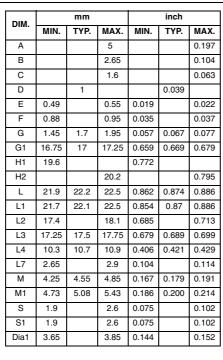
TDA2004R Package information

4 Package information

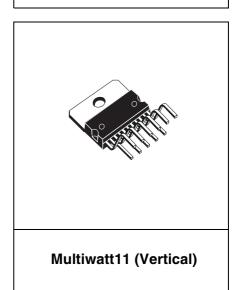
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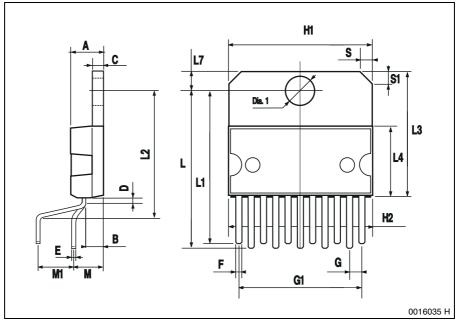
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Figure 21. Multiwatt11 mechanical data and package dimensions



OUTLINE AND MECHANICAL DATA





Revision history TDA2004R

5 Revision history

Table 6. Document revision history

Date	Revision	Changes
18-Jun-2010	1	Initial release.
18-Sep-2013	2	Updated Disclaimer.

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