

Subject Name: DIGITAL COMMUNICATION (DCOM) EC209

Admission Number: U19CS012

Full Name: BHAGYA VINOD RANA

Division: A

Total Pages: 11

Q1) Transmitter and Receiver of two separate balance modulators  
→ two carrier waves of same freq, but differing in phase by  $90^\circ$ .

① QAM (Quadrature Amplitude Modulation)

A signal in which two carriers shifted in phase by  $90^\circ$  degrees (i.e. sine and cosine) are modulated and combined.

② One signal is called In-phase or "I" signal (sine)  
other signal is called quadrature or "Q" signal (cosine)

QAM Transmitter (Figure 1(a))

③ Idea derived from Basic QAM theory, two carrier signals with phase shift  $90^\circ$ . These are then amplitude modulated with two data streams known as I (In-phase) or Q (Quadrature)

(Generated in Baseband processing area)

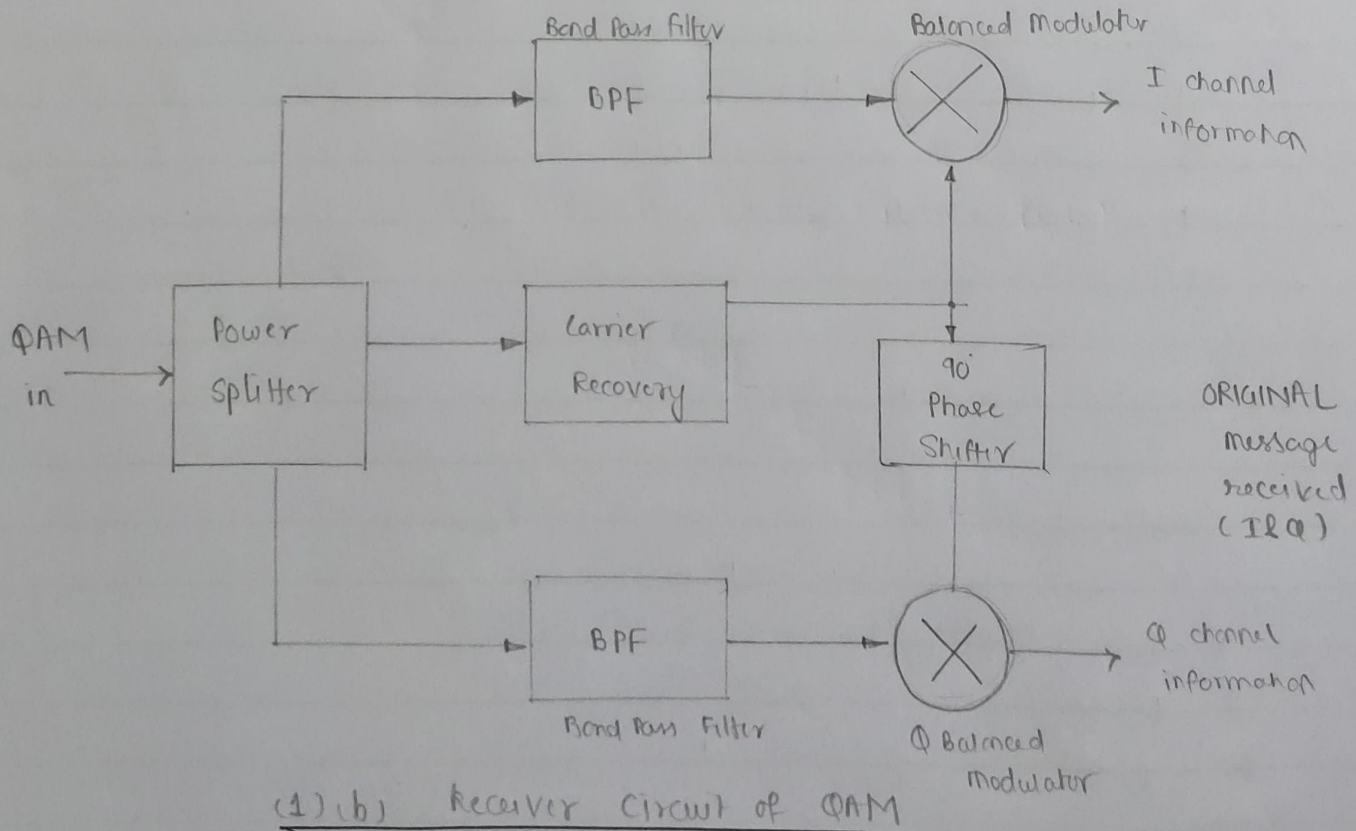
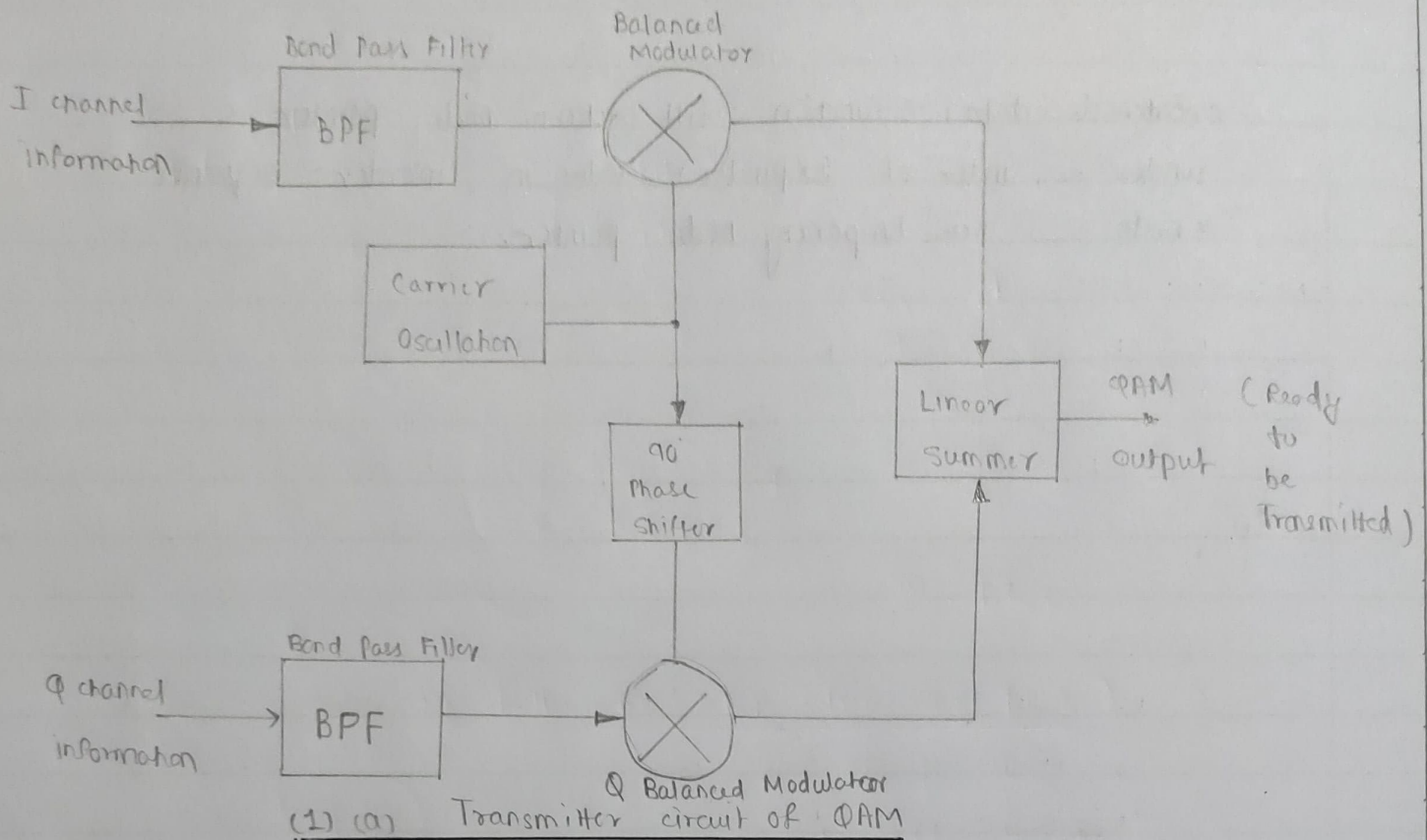
④ The two resultant signals are summed and then processed as required in RF signal chain.

RF Amplifier must be linear to preserve the integrity of signal.

Any non-linearities will alter → relative level of signals

→ phase difference

Possibility of Error.





let two carrier signals be

$$V_{c1} = V_c \cos(\omega_c t)$$

$$V_{c2} = V_c \sin(\omega_c t)$$

Corresponding QAM (After transmitter summation)

$$V_{QAM} = \underbrace{V_{m1} V_c \cos(\omega_c t)}_{\text{In phase}} + \underbrace{V_{m2} V_c \sin(\omega_c t)}_{\text{Quadrature}}$$

RECEIVER (QAM) (Figure 1(b))

⑤ message signals are recovered using coherent detection. (w.c.t)  $\sin(\omega_c t) \cos$

Output of Analog Multiplier + I phase

$$S_I = V_{QAM} V_c' \cos(\omega_c t) = \underbrace{\frac{V_{m1} V_c V_c'}{2}}_{\text{Scaled version of } V_{m1}} + \frac{V_{m1} V_c V_c'}{2} \cos(2\omega_c t) + \frac{V_{m2} V_c V_c'}{2}$$

Scaled version  
of  $V_{m1}$

Low pass filter

Output of Analog Multiplier + Q phase

$$S_Q = V_{QAM} V_c' \sin(\omega_c t) = \frac{V_{m2} V_c V_c'}{2} + \frac{V_{m2} V_c V_c'}{2} \sin(2\omega_c t) + \underbrace{V_{m1} V_c V_c'}_{\text{(w.c.t) } \cos}$$

In this way, we can transmit two independent message signals on the same bandwidth with help of two carrier (with 90 phase).

2.7 Transmitter & Receiver block single bit per sample digital mod.

① Waveform with mathematical expression

② Working

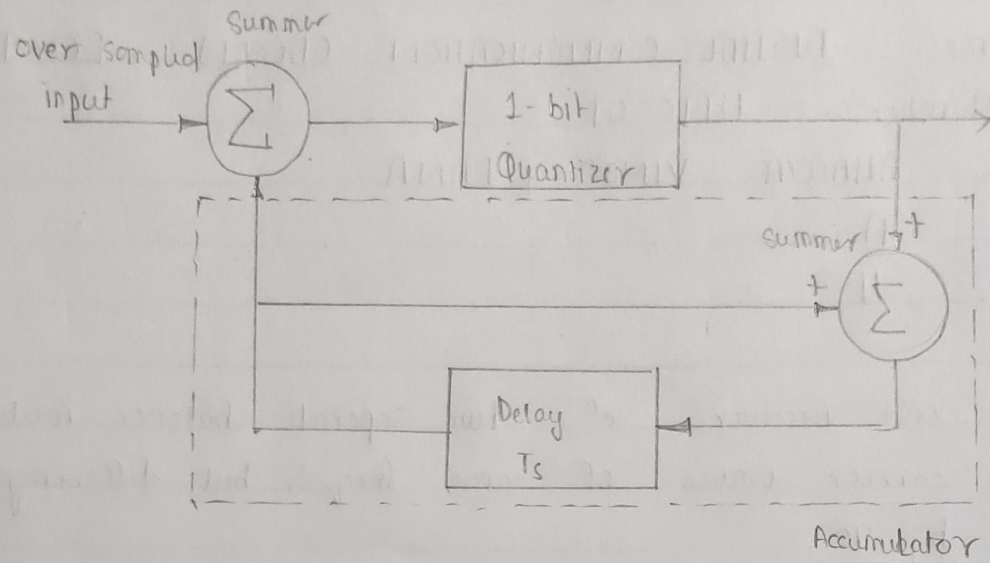
① Delta modulation (sampling rate  $\gg$  stepsize after quantization)

> The design of modulator & demodulator is simple

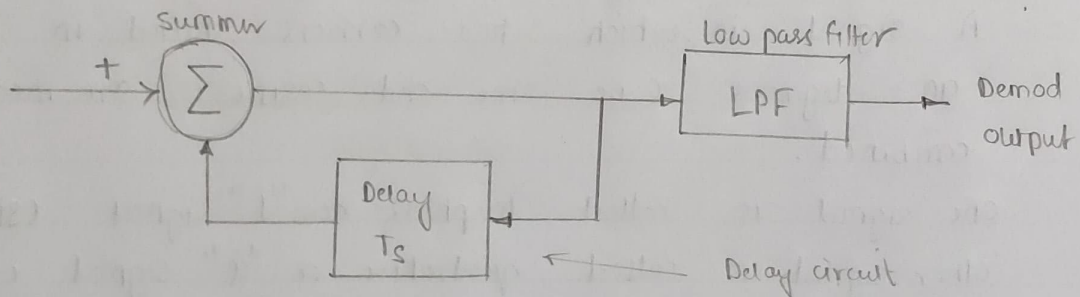
> Staircase approx of output waveform.

> bit rate can be decided by user. (Single bit/sample)

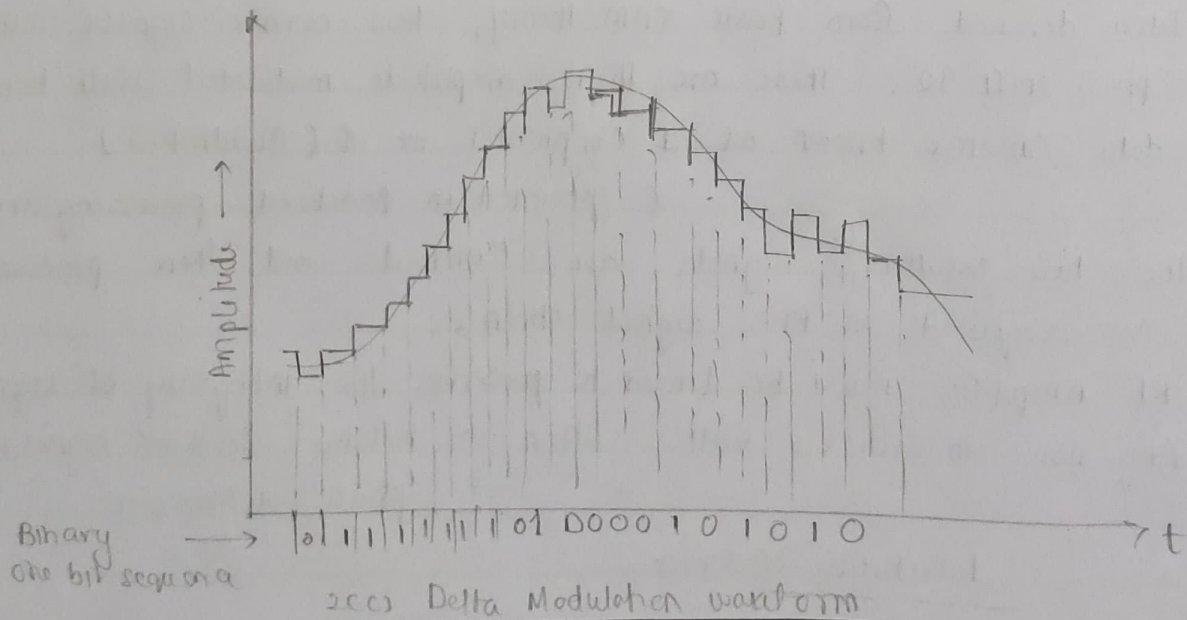
(1 bit DPCM scheme)



2(a) Delta Modulator (Transmitter Block)



2(b) Delta Demodulator (Receiver Block)





## ② Working Principle

- (i) Delta modulation transmits one bit per sample
  - (ii) Here present sample value is compared with the previous sample value and this results whether Amplitude will increase or decrease is transmitted.
  - (iii) Input signal  $x(t)$   $\approx$  approximate to step signal by delta (modulator)  
Step size = fixed
  - (iv) The difference between input signal  $x(t)$  & staircase approximate signal  $\rightarrow$  2 levels  $+\Delta$  &  $-\Delta$  ① transmitted.
- if  $\left\{ \begin{array}{l} \text{difference} = \text{Positive} = \text{Approximated signal} \uparrow \text{increase by 1 step} \\ \text{difference} = \text{Negative} = \text{Approx. signal} \downarrow \text{decrease by 1 step} \end{array} \right.$   
① transmitted

## ③ Mathematical Expressions

Principle can be explained using following eqn's

The error between sampled value and last approximated sample

$$e(nT_s) = x(nT_s) - \hat{x}(nT_s)$$

$e(nT_s)$  = error at present sample

$x(nT_s)$  = sampled signal of  $x(t)$

$\hat{x}(nT_s)$  = last sample approximation of the staircase waveform.

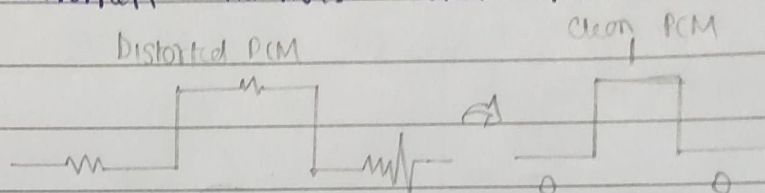
UI9C5012

## Q. 1 PCM control Noise and Distortion

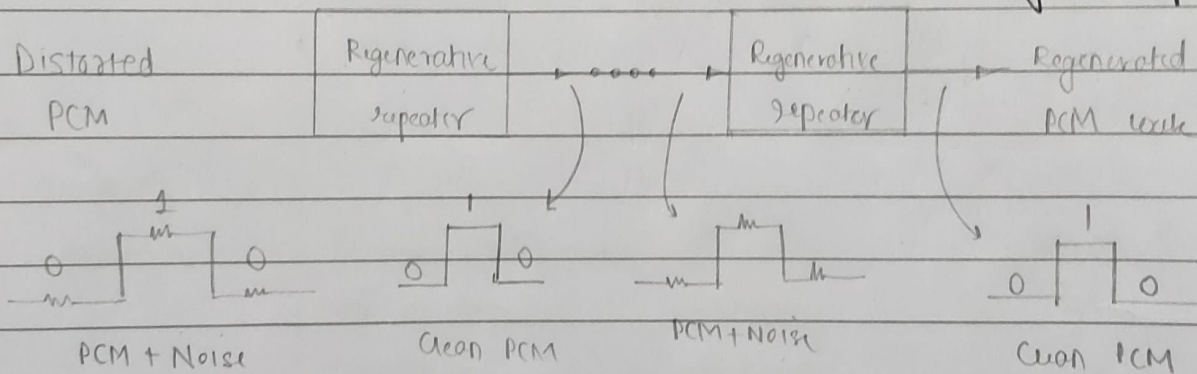
The PCM is a method to <sup>convert</sup> sampled analog signal  $\rightarrow$  digital code

The most important feature of PCM system is its ability to control the effects of distortion and noise when PCM wave travels on the channel.

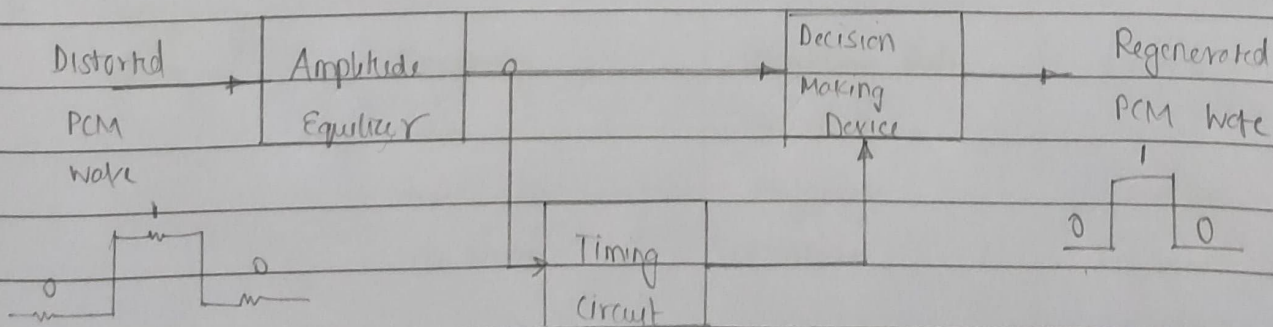
PCM Transmitted PATH



(3) (c) Waveform of regenerative repeater



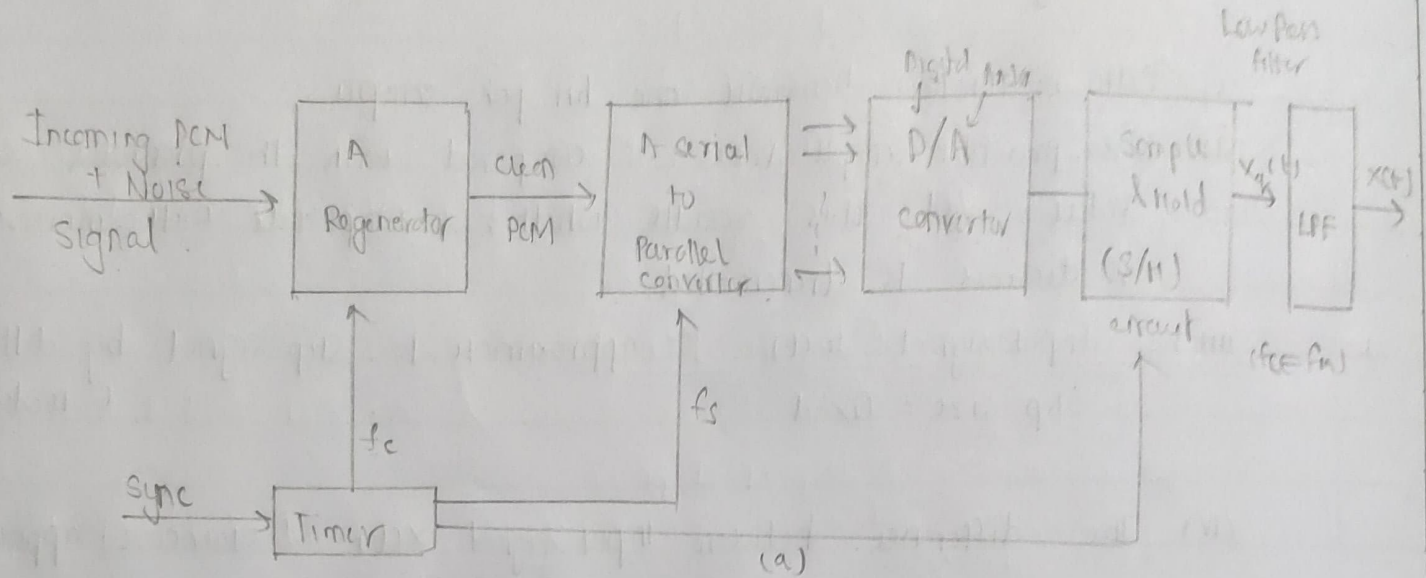
## (3) (a) PCM Path



## (3) (b) Block diagram of Regenerative Repeater



(3cd) Whole process



3cd) PCM Receiver

$$[1.111]_2 = [1.111]_2 = [1.111]_2$$

47	Sr No.	Parameter of comparison	Ideal or Instantaneous Sampling	Natural sampling	Flat top Sampling
①		Generation Circuit (Circuit of sampler)			
②		Waveforms involved			
③		Sampling principle	It uses multiplication by an impulse function	It uses chopping principle	Sample & Hold circuit
④		Sampling Rate	Sampling rate tends to Infinity.	Sampling Rate satisfies Nyquist criteria.	Sampling Rate Satisfies Nyquist criteria.



5

$$m(t) = 10 \cos(2\pi \times 10^3 t)$$

$$c(t) = 20 \sin(2\pi \times 10^4 t)$$

$$A_m = 10 \quad f_m = 10^3 \text{ Hz} = 1 \text{ kHz}$$

$$A_c = 20 \quad f_c = 10^4 \text{ Hz} = 10 \text{ kHz}$$

(i)  $\mu = \text{Modulation index} = \frac{A_m}{A_c} = \frac{10}{20} = 0.5$

Percentage modulation =  $\mu \times 100\% = \boxed{50\%}$

(ii) frequencies of sidebands =  $f_c - f_m = 9 \text{ kHz}$

=  $f_c + f_m = 11 \text{ kHz}$

Ans:  $\boxed{9 \text{ kHz}}$   $\boxed{f_{USB} = 11 \text{ kHz}}$

(lower side band) (upper side band)

(iii) Amplitude of side bands =  $\left(\frac{\mu}{2}\right) (A_c)$

=  $\frac{(0.5)}{(2)} \times (20) 10$

=  $\boxed{5 \text{ unit}}$

(iv) Bandwidth =  $2 \times f_m$   
of =  $2 \times 1 \text{ kHz}$

modulating signal =  $\boxed{2 \text{ kHz}}$

⑥

$E_c = 10V$

$f_c = 30 \text{ kHz}$

$E_m = 3V$

$f_m = 1 \text{ kHz}$

$R = 50 \Omega$

$$(i) \text{ Modulation Index} = \mu = \frac{E_m}{E_c} = \frac{3}{10} = \boxed{0.3}$$

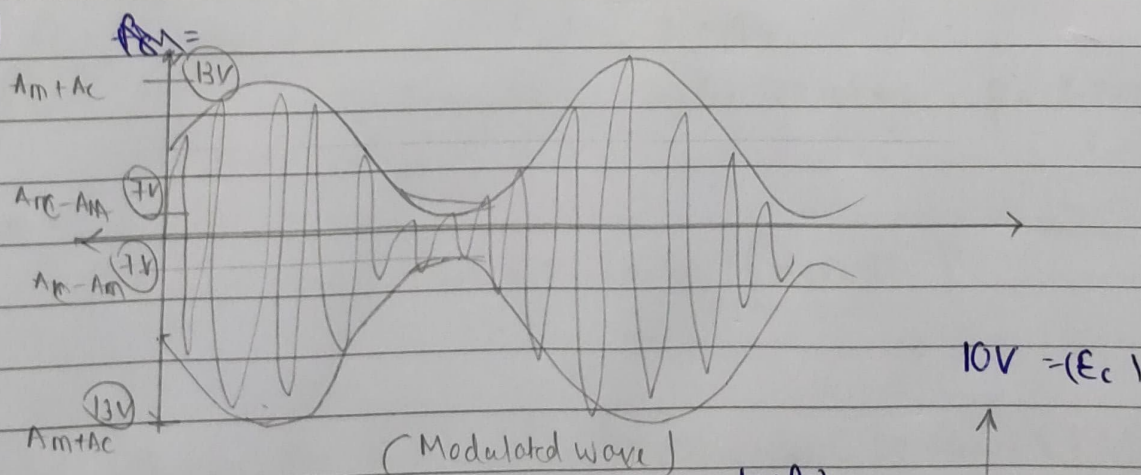
(ii) Equation of Modulated wave

$$S(t) = A_c [1 + \mu \cos(2\pi f_m t)] \cos(2\pi f_c t)$$

$$= 10 [1 + 0.3 \cos(2\pi(1000)(t))] \cos(2\pi 30 \times 10^3 t)$$

$$= 10 [1 + 0.3 \cos(2\pi(10^3) t)] \cos(60 \times 10^3 \pi t)$$

(iii)



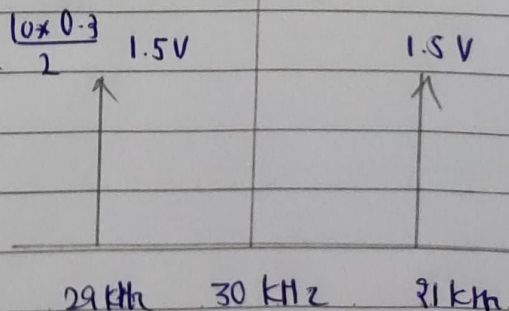
(iv)

$f_c = 30 \text{ kHz}$

$f_{LSR} = 29 \text{ kHz}$

$f_{USR} = 31 \text{ kHz}$

$$\left( \frac{\mu A_c}{2} \right)$$





⑦

$f_c = 20 \text{ MHz}$

$E_c = 5 \text{ V}$

$f_m = 400 \text{ Hz}$

$\Delta f = 10 \text{ KHz}$   
 $= k_f A_m$

$\beta = \frac{\Delta f}{f_m} = \frac{k_f A_m}{f_m} = 25$

FM =

$$s(t) = A_c \cos(2\pi f_c t + 2\pi k_f \int m(t) dt)$$

$$= 5 \cos(2\pi (20 \text{ MHz}) t + 2\pi (25) \int m(t) dt)$$

PM

$$s(t) = A_c \cos(2\pi f_c t + \beta \cos(2\pi f_m t))$$

$(\beta = k_p A_m)$

$$= [5 \text{ V} \cos(2\pi (20 \text{ MHz}) t + 25 \cos(2\pi (400) t))]$$

$f_c = 20 \times 10^6 \text{ Hz}$

$E_c = 5 \text{ V}$

$f_m = 2000 + 400 = 2400 \text{ Hz}$

$\Delta f = 10 \times 10^3 \text{ Hz}$

$\beta = \frac{10 \text{ KHz}}{2.4 \text{ KHz}}$

FM =

$\beta = 4.167$

$$5 \cos(2\pi (20 \text{ MHz}) t + 2\pi k_f \int m(t) dt)$$

$$\text{PM} = 5 \cos(2\pi (20 \text{ MHz}) t + 4.167 \cos(2\pi (2400) t))$$

X