#### Network Analysis & Systems

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Network Analysis and Synthesis

Part V: Network Synthesis





#### Network Functions (1)

■ Let H(s) be a network function:

$$H(s) = \frac{M_1(s) + N_1(s)}{M_2(s) + N_2(s)}$$

where  $M_1$  and  $M_2$  are the even parts, and  $N_1$  and  $N_2$  are the odd parts.

Also,

$$H(s) = \frac{M_1(s) + N_1(s)}{M_2(s) + N_2(s)} \frac{M_2(s) - N_2(s)}{M_2(s) - N_2(s)}$$

$$= \frac{M_1M_2 - N_1N_2}{M_2^2 - N_2^2} + \frac{N_1M_2 - M_1N_2}{M_2^2 - N_2^2}$$

$$\stackrel{\triangle}{=} \text{Even } [H(s)] + \text{Odd } [H(s)]$$





# Network Functions (2)

On the imaginary axis,

$$H(j\omega) = \text{Re} [H(j\omega)] + j \text{Im} [H(j\omega)]$$

Clearly,

$$Re [H(j\omega)] = Even [H(s)]|_{s=j\omega}$$

and

$$j$$
Im  $[H(j\omega)] = \text{Odd } [H(s)]|_{s=j\omega}$ 

That is, the even part is real and the odd part is imaginary.





■ Let Z(s) be the impedance of a passive one-port.





- Let Z(s) be the impedance of a passive one-port.
- For a pure reactive network, the average power dissipated is zero.
- Therefore,

Re 
$$[Z(j\omega)] = \text{Even } [Z(j\omega)] = 0$$

where

Even 
$$[Z(s)] = \frac{M_1(s)M_2(s) - N_1(s)N_2(s)}{M_2^2(s) - N_2^2(s)}$$

■ Clearly, Even  $[Z(j\omega)] = 0$  implies,

$$M_1(j\omega)M_2(j\omega) - N_1(j\omega)N_2(j\omega) = 0$$





$$M_1(j\omega)M_2(j\omega) - N_1(j\omega)N_2(j\omega) = 0$$

#### implies

- $Z(s) = \frac{M_1 M_2 N_1 N_2}{M_2^2 N_2^2} + \frac{N_1 M_2 M_1 N_2}{M_2^2 N_2^2} = \frac{N_1 M_2 M_1 N_2}{M_2^2 N_2^2}.$
- Either  $M_1 = 0 = N_2$  or  $M_2 = 0 = N_1$ .





$$M_1(j\omega)M_2(j\omega) - N_1(j\omega)N_2(j\omega) = 0$$

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- Either  $M_1 = 0 = N_2$  or  $M_2 = 0 = N_1$ .

#### Thus,

■ If 
$$M_1 = 0 = N_2$$
, then  $Z(s) = \frac{N_1(s)}{M_2(s)}$ .





$$M_1(j\omega)M_2(j\omega) - N_1(j\omega)N_2(j\omega) = 0$$

#### implies

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#### Thus,

- If  $M_1 = 0 = N_2$ , then  $Z(s) = \frac{N_1(s)}{M_2(s)}$ .
- If  $M_2 = 0 = N_1$ , then  $Z(s) = \frac{M_1(s)}{M_2(s)}$ .





$$M_1(j\omega)M_2(j\omega) - N_1(j\omega)N_2(j\omega) = 0$$

#### implies

- $Z(s) = \frac{M_1 M_2 N_1 N_2}{M_2^2 N_2^2} + \frac{N_1 M_2 M_1 N_2}{M_2^2 N_2^2} = \frac{N_1 M_2 M_1 N_2}{M_2^2 N_2^2}.$
- Either  $M_1 = 0 = N_2$  or  $M_2 = 0 = N_1$ .

#### Thus,

- If  $M_1 = 0 = N_2$ , then  $Z(s) = \frac{N_1(s)}{M_2(s)}$ .
- If  $M_2 = 0 = N_1$ , then  $Z(s) = \frac{M_1(s)}{N_2(s)}$ .
- Thus LC immittance functions are a ratio of odd to even polynomials or even to odd polynomials.
- All poles and zeros lie on the imaginary axis.





Consider an LC immittance function

$$H(s) = \frac{a_4 s^4 + a_2 s^2 + a_0}{b_5 s^5 + b_3 s^3 + b_1 s}$$

- All coefficients are real and positive.
- All poles and zeros lie on the imaginary axis.
- There are no multiple poles or zeros on the imaginary axis.





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- All coefficients are real and positive.
- All poles and zeros lie on the imaginary axis.
- There are no multiple poles or zeros on the imaginary axis.
- The highest powers must differ by unity: If 2n is the degree of the numerator, then the denominator has degree either 2n-1 or 2n+1.
- If the degree of denominator is 2n-1, then H(s) has a simple pole at  $s=\infty$ .
- If the degree of denominator is 2n + 1, then H(s) has a simple zero at  $s = \infty$ .



$$H(s) = \frac{a_4s^4 + a_2s^2 + a_0}{b_5s^5 + b_3s^3 + b_1s}$$

■ The lowest terms of numerator and denominator must differ by one; otherwise, there would be multiple poles or zeros at s = 0, a contradiction.





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- The lowest terms of numerator and denominator must differ by one; otherwise, there would be multiple poles or zeros at s = 0, a contradiction.
- If the polynomial is even, it should contain all even powers.
- If the polynomial is odd, it should contain all odd powers.
- Otherwise, there would be poles or zeros not on the imaginary axis, a contradiction.





Therefore,

$$H(s) = \frac{K(s^2 + \omega_1^2)(s^2 + \omega_3^2) \cdots (s^2 + \omega_i^2) \cdots}{s(s^2 + \omega_2^2)(s^2 + \omega_4^2) \cdots (s^2 + \omega_j^2) \cdots}$$
$$= \frac{k_0}{s} + \frac{2k_2s}{s^2 + \omega_2^2} + \frac{2k_4s}{s^2 + \omega_4^2} + \cdots + k_{\infty}s$$

- The residues must be real and positive.
- Clearly,  $H(j\omega)$  is a pure reactance  $jX(\omega)$ , where

$$X(\omega) = -\frac{k_0}{\omega} + \frac{2k_2\omega}{-\omega^2 + \omega_2^2} + \frac{2k_4\omega}{-\omega^2 + \omega_4^2} + \dots + k_\infty\omega$$

Moreover.

$$\frac{dX(\omega)}{d\omega} = \frac{k_0}{\omega^2} + k_\infty + \frac{2k_2(\omega^2 + \omega_2^2)}{(\omega^2 - \omega_2^2)^2} + \dots \ge 0$$



Let

$$H(s) = \frac{Ks(s^2 + \omega_3^2)}{(s^2 + \omega_2^2)(s^2 + \omega_4^2)}$$

Therefore,

$$H(j\omega) = jX(\omega) =$$





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$$H(s) = \frac{Ks(s^2 + \omega_3^2)}{(s^2 + \omega_2^2)(s^2 + \omega_4^2)}$$

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$$H(j\omega) = jX(\omega) = j\frac{K\omega(-\omega^2 + \omega_3^2)}{(-\omega^2 + \omega_2^2)(-\omega^2 + \omega_4^2)}$$

■ What is the plot of  $X(\omega)$  vs.  $\omega$ ?



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Therefore,

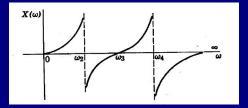
$$H(j\omega) = jX(\omega) = j\frac{K\omega(-\omega^2 + \omega_3^2)}{(-\omega^2 + \omega_2^2)(-\omega^2 + \omega_4^2)}$$

- What is the plot of  $X(\omega)$  vs.  $\omega$ ?
- Recall that  $\frac{dX(\omega)}{d\omega} \ge 0$ .
- $\blacksquare$   $\omega_2$ ,  $\omega_3$ , and  $\omega_4$  are called internal critical frequencies.
- ullet s=0 and  $s=\infty$  are called external critical frequencies.





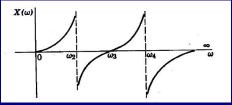
$$H(s) = rac{Ks(s^2 + \omega_3^2)}{(s^2 + \omega_2^2)(s^2 + \omega_4^2)}$$

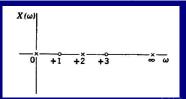






$$H(s) = rac{Ks(s^2 + \omega_3^2)}{(s^2 + \omega_2^2)(s^2 + \omega_4^2)}$$









#### Summary of Properties:

- It is the ratio of odd to even polynomials or even to odd polynomials.
- The poles and zeros are simple and lie on the imaginary axis.
- The poles and zeros interlace on the imaginary axis.
- The highest powers of numerator and denominator must differ by unity; the lowest powers also differ by unity.
- There must be a zero or a pole at the origin and infinity.





$$H(s) = \frac{Ks(s^2 + 4)}{(s^2 + 1)(s^2 + 3)}$$





$$H(s) = \frac{Ks(s^2 + 4)}{(s^2 + 1)(s^2 + 3)}$$

$$H(s) = \frac{s^5 + 4s^3 + 5s}{3s^4 + 6s^2}$$





$$H(s) = \frac{Ks(s^2+4)}{(s^2+1)(s^2+3)}$$

$$H(s) = \frac{s^5 + 4s^3 + 5s}{3s^4 + 6s^2}$$

$$H(s) = \frac{K(s^2+1)(s^2+9)}{(s^2+2)(s^2+10)}$$





$$H(s) = \frac{k_0}{s} + \frac{2k_2s}{s^2 + \omega_2^2} + \frac{2k_4s}{s^2 + \omega_4^2} + \dots + k_\infty s$$

Let H(s) be an impedance.

■ The first term is a  $1/k_0$  F capacitor and the last term is a  $k_{\infty}$  H inductor.





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Let H(s) be an impedance.

- The first term is a  $1/k_0$  F capacitor and the last term is a  $k_{\infty}$  H inductor.
- The term  $\frac{2k_is}{s^2+\omega_i^2}$  is a parallel tank circuit with  $1/2k_i$  F capacitor in parallel with a  $2k_i/\omega_i^2$  H inductor.
- This is known as Foster series network.

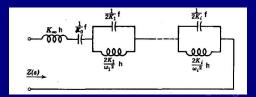




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$$Z(s) = \frac{2(s^2+1)(s^2+9)}{s(s^2+4)} =$$



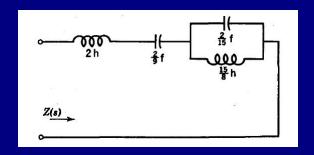


$$Z(s) = \frac{2(s^2+1)(s^2+9)}{s(s^2+4)} = 2s + \frac{9/2}{s} + \frac{15s/2}{s^2+4}$$





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Let H(s) be an admittance.

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- The term  $\frac{2k_is}{s+\omega_i^2}$  is a series circuit with  $1/2k_i$  H inductor in series with a  $2k_i/\omega_i^2$  F capacitor.
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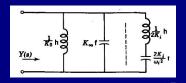




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$$Y(s) = \frac{s(s^2+2)(s^2+4)}{s(s^2+1)(s^2+3)} =$$



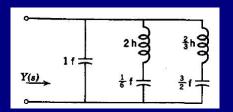


$$Y(s) = \frac{s(s^2+2)(s^2+4)}{s(s^2+1)(s^2+3)} = s + \frac{3s/2}{s^2+1} + \frac{s/2}{s^2+3}$$





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- Recall: the degrees of the numerator and denominator always differ by unity.
- Therefore, there is always a pole or zero at  $s = \infty$ .





- Recall: the degrees of the numerator and denominator always differ by unity.
- Therefore, there is always a pole or zero at  $s = \infty$ .
- Thus, for an impedance function Z(s), the degree of the numerator is 2n and the degree of the denominator is 2n-1, implying that there is a pole at  $s=\infty$ .
- $\blacksquare$  Clearly, we can remove an impedance  $L_1s$  so that  $Z_2(s) \stackrel{\Delta}{=} Z(s) - L_1 s$  is still an LC impedance function.



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- Clearly, we can remove an impedance  $L_1s$  so that  $Z_2(s) \stackrel{\Delta}{=} Z(s) L_1s$  is still an LC impedance function.
- The degree of the denominator of  $Z_2(s)$  is 2n-1 making the degree of the numerator 2n-2!
- Clearly  $Z_2(s)$  has a zero at  $s = \infty$ ; or  $1/Z_2(s)$  has a pole at  $s = \infty$ .
- This process can be continued, leading to a CFE with a ladder network.



$$Z(s) = \frac{2s^5 + 12s^3 + 16s}{s^4 + 4s^2 + 3}$$





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$$s^{4} + 4s^{2} + 3)2s^{5} + 12s^{3} + 16s(2s \leftrightarrow Z)$$

$$\underline{2s^{5} + 8s^{3} + 6s}$$

$$\underline{4s^{3} + 10s)s^{4} + 4s^{2} + 3(\frac{1}{4}s \leftrightarrow Y)}$$

$$\underline{s^{4} + \frac{5}{2}s^{2}}$$

$$\underline{\frac{3}{2}s^{2} + 3)4s^{3} + 10s(\frac{9}{3}s \leftrightarrow Z)}$$

$$\underline{4s^{3} + 8s}$$

$$\underline{2s)\frac{3}{2}s^{2} + 3(\frac{3}{4}s \leftrightarrow Y)}$$

$$\underline{\frac{3}{2}s^{2}}$$

$$\underline{3)2s(\frac{2}{3}s \leftrightarrow Z)}$$



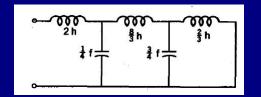


$$Z(s) = 2s + rac{1}{rac{s}{4} + rac{1}{rac{8s}{4} + rac{1}{rac{3s}{4} + rac{1}{2s/3}}}$$





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- If the initial function is an impedance, the first quotient is an impedance, the second is an admittance (due to the expansion of the reciprocal), and so on.
- This is a Cauer ladder network.





- If the initial function is an impedance, the first quotient is an impedance, the second is an admittance (due to the expansion of the reciprocal), and so on.
- This is a Cauer ladder network.
- In contrast, if H(s) has a pole at s=0, the aforementioned procedure can be used to successively remove the poles at s=0.
- For this the polynomials are arranged in ascending order.





$$Z(s) = \frac{(s^2+1)(s^2+3)}{s(s^2+2)}$$

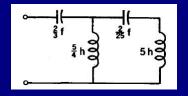




$$Z(s) = \frac{(s^2+1)(s^2+3)}{s(s^2+2)}$$







■ This is also a Cauer ladder network.





#### Canonical Forms

- Foster Series Network or Foster I form: PFE of an impedance function.
- Foster Parallel Network or Foster II form: PFE of an admittance function.
- Cauer I form: CFE when the polynomials are arranged in descending order.
- Cauer II form: CFE when the polynomials are arranged in ascending order.
- Cauer I and II forms result in ladder networks and are applicable for an immittance function.



