Network Analysis & Systems

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UE18EC201: Network Analysis & Systems





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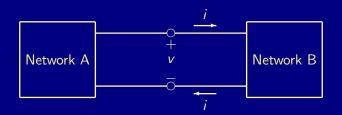


Network Analysis and Synthesis

Unit III: Network Theorems







- Network B is the focus of interest.
- Network A to be replaced by an equivalent network s.t. the current *i* and the voltage *v* remain invariant.





Assumptions on Network A:



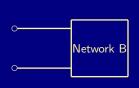
- Linear elements.
- Independent or dependent sources.
- Initial conditions on passive elements.
- No magnetic or controlled-source coupling to Network B.







Assumptions on Network B:

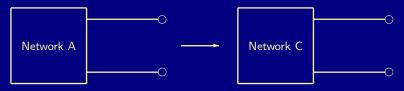


- Linear, nonlinear, or time-varying elements.
- Independent or dependent sources.
- Initial conditions on passive elements.
- No magnetic or controlled-source coupling to Network A.





Step A: Derive Network C from Network A



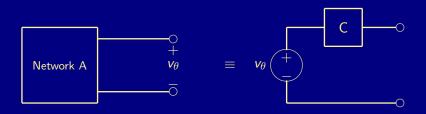
- Set all initial conditions to zero: $v_C = 0$ and $i_L = 0$.
- Turn off all independent sources: v = 0 (i.e., s.c.) for voltage sources and i = 0 (i.e., o.c.) for current sources.
- Controlled sources continue to operate.
- Determine the driving-point impedance or admittance.







Thévenin Equivalent Network:

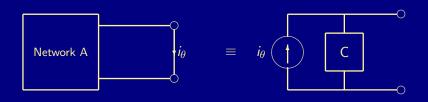


- \blacksquare v_{θ} is the voltage at the open terminals of network A with network B removed.
- \blacksquare v_{θ} is connected in series with network C.





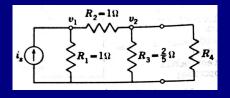
Norton Equivalent Network:



- \bullet i_{θ} is the current in the shorted terminals of network A with network B removed.
- \bullet i_{θ} is connected in parallel with network C.

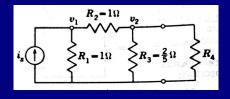










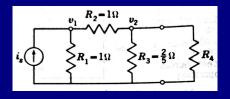


Source: Van Valkenburg, 1975.

lacksquare Find $v_{ heta}=v_2$: Apply nodal analysis to obtain



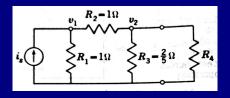




- Find $v_{\theta} = v_2$: Apply nodal analysis to obtain $v_2 = i_s/6$.
- Find R_{θ} after open circuiting the current source:







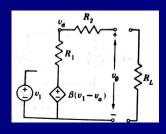
- Find $v_{\theta} = v_2$: Apply nodal analysis to obtain $v_2 = i_s/6$.
- Find R_{θ} after open circuiting the current source: $R_{\theta} = 1/3$.





Examples

Network Theorems — Thévenin and Norton (8)



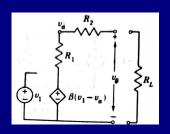
■ Find v_{θ} :





Examples

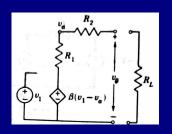
Network Theorems — Thévenin and Norton (8)



■ Find v_{θ} : Clearly, $v_{\theta} = v_a = \beta(v_1 - v_a)$. Therefore,



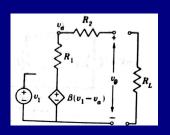




- Find v_{θ} : Clearly, $v_{\theta} = v_a = \beta(v_1 v_a)$. Therefore, $v_{\theta} = \frac{\beta}{1+\beta}v_1$.
- Find R_{θ} :

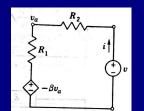






- Find v_{θ} : Clearly, $v_{\theta} = v_a = \beta(v_1 v_a)$. Therefore, $v_{\theta} = \frac{\beta}{1+\beta}v_1$.
- Find R_{θ} : Set $v_1 = 0$ and connect a voltage source v to the output terminals.

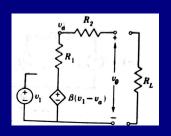
Source: Van Valkenburg, 1975.



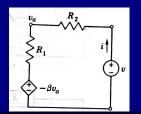
For the loop,







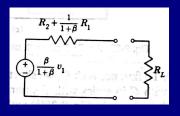
- Find v_{θ} : Clearly, $v_{\theta} = v_a = \beta(v_1 v_a)$. Therefore, $v_{\theta} = \frac{\beta}{1+\beta}v_1$.
- Find R_{θ} : Set $v_1 = 0$ and connect a voltage source v to the output terminals.



- For the loop, $v = iR_2 + v_a = iR_2 + iR_1 - \beta v_a$.
- Therefore,

$$v = i \left(R_2 + R_1 \left(1 - \frac{\beta}{1 + \beta} \right) \right)$$





Thus,

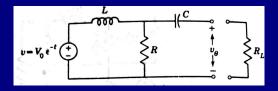
$$extstyle v_ heta = rac{eta}{1+eta} extstyle v_1$$

and

$$R_{ heta} = R_2 + R_1 \left(1 - rac{eta}{1+eta} \right) = R_2 + R_1 rac{1}{1+eta}$$





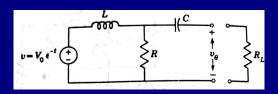


Source: Van Valkenburg, 1975.

■ Find $V_{\theta}(s)$:







Source: Van Valkenburg, 1975.

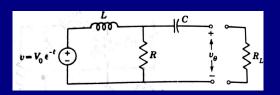
■ Find $V_{\theta}(s)$: Clearly,

$$V_{ heta}(s) = rac{R}{Ls+R}V(s) = rac{V_0R}{(Ls+R)(s+1)}$$

■ Find Z_θ :







Source: Van Valkenburg, 1975

■ Find $V_{\theta}(s)$: Clearly,

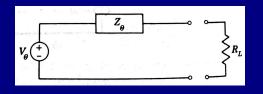
$$V_{ heta}(s) = rac{R}{Ls+R}V(s) = rac{V_0R}{(Ls+R)(s+1)}$$

■ Find Z_{θ} :

$$Z_{ heta}(s) = rac{1}{Cs} + rac{1}{rac{1}{R} + rac{1}{Ls}} = rac{R(s^2 + (1/RC)s + (1/LC))}{s(s + (R/L))}$$



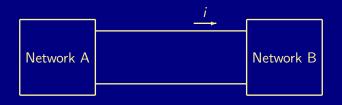




$$V_{\theta}(s) = rac{V_{0}R}{(Ls+R)(s+1)}$$
 $Z_{\theta}(s) = rac{R(s^{2}+(1/RC)s+(1/LC))}{s(s+(R/L))}$



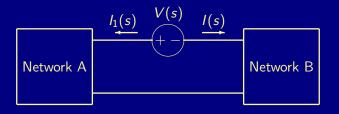




■ The current i(t) (or, I(s)) flows into Network B.



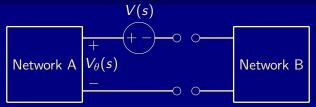




- Include a voltage source V(s) so that the resulting current $I_1(s)$ is such that $I_1(s) = -I(s)$.
- Therefore, the net current flowing into network B is zero.



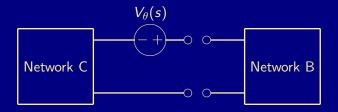




- Since the net current flowing into network B is zero, the two networks can be broken without affecting the conditions in network A.
- The voltage at the broken terminals is clearly zero.
- Also, the network B can be shorted.
- By KVL, $V_{\theta}(s) = V(s)$.
- That is, the open circuit voltage at the terminals is V(s).



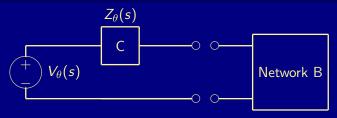




- All independent sources are reduced to zero.
- Dependent sources are not changed.
- Thus network A is transformed into network C.
- The polarity of the source $V_{\theta}(s)$ is reversed so that when connected, the current I(s) will flow into network B.





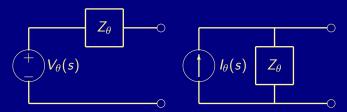


- Impedance of C is $Z_{\theta}(s)$; the voltage source is $V_{\theta}(s)$.
- If A has only independent sources, then $Z_{\theta}(s)$ is the impedance of the passive network; otherwise, $Z_{\theta}(s)$ must be determined for the active network.
- This is the Thévenin equivalent network.
- If $Z_B(s)$ is the impedance of B, then the current is

$$I(s) = rac{V_{ heta}(s)}{Z_{ heta}(s) + Z_{B}(s)}$$







■ Perform source transformation to obtain the current source

$$I_{ heta}(s) = rac{V_{ heta}(s)}{Z_{ heta}(s)}$$

- This is the Norton equivalent network.
- The voltage of network B is

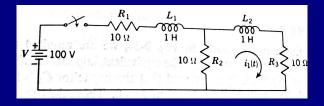
$$V_B(s) = rac{I_{ heta}(s)}{Y_{ heta}(s) + Y_B(s)}$$





Examples (1)

Example 6, Van Valkenburg, pp. 267–268:

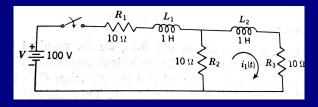


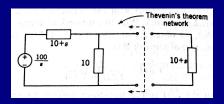




Examples (1)

Example 6, Van Valkenburg, pp. 267–268:



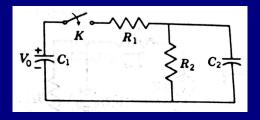






Examples (2)

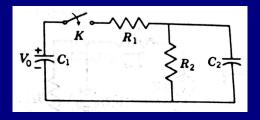
Example 7, Van Valkenburg, pp. 268–269:





Examples (2)

Example 7, Van Valkenburg, pp. 268–269:





Examples (3)

Example 8, Van Valkenburg, p. 270:

