

5-09-19

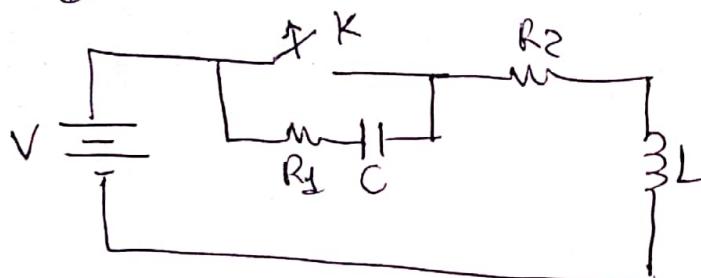
3rd SEM-PESU

Problems on Initial Conditions and Transient Behavior

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 Dept.

Network Analysis and Synthesis
 [VUE18EC201]

- (1) In the n/w shown the switch is opened at $t=0$ after n/w has attained steady state with switch closed. (a) Find an expression for voltage across switch at $t=0^+$; (b) If parameters are adjusted such that $i(0^+) = 1A$ and $\frac{di(0^+)}{dt} = -1A/\text{sec}$. What is the value of derivative across the switch.



Q) When K is closed 'L' acts like S.C. and 'C' acts like o.c.

$$i(0^-) = \frac{V}{R_2} = i(0^+) \quad (1)$$

$$v_c(0^-) = v_c(0^+) = 0V \quad (2)$$

Q) When 'K' is opened at $t=0^+$:

$$v_K = R_1 i + \frac{1}{C} \int_{-\infty}^t i \, d\tau \quad (3)$$

$$\therefore v_K(0^+) = R_1 i(0^+) + \frac{1}{C} \int_{-\infty}^{0^+} i \, d\tau = R_1 \frac{V}{R_2} + v_c(0^+) = \frac{VR_1}{R_2} \quad (4)$$

We have,

$$v_K = R_1 i + \frac{1}{C} \int_{-\infty}^t i \, d\tau \quad (5)$$

→ Differentiating

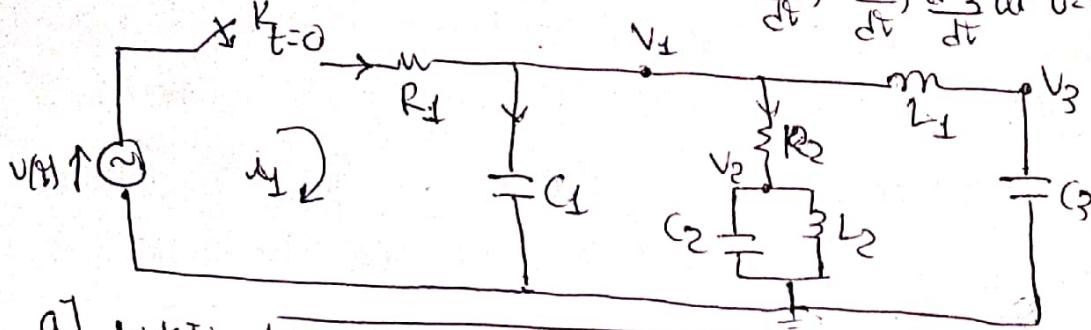
$$\frac{dv_K}{dt} = R_1 \frac{di}{dt} + \frac{1}{C} i \quad (6)$$

$$\therefore \frac{dv_K(0^+)}{dt} = R_1 \frac{di(0^+)}{dt} + \frac{1}{C} i(0^+) = +R_1(-1) + \frac{1}{C}(1) = \frac{1}{C} - R_1 \quad (7)$$

(1)

Q. 5.27 In the network shown below, the switch 'K' is closed at $t=0^-$. At $t=0^+$ all capacitor voltages and inductor currents are zero.

Find (a) V_1, V_2, V_3 at $t=0^+$ (b) $\frac{dV_1}{dt}, \frac{dV_2}{dt}, \frac{dV_3}{dt}$ at $t=0^+$



a) W.K.T.: - $[i_{L_1}(0^+) = i_{L_2}(0^+) = i_{L_3}(0^+) = 0A] - [1]$

b) At node V_1 :

$$\frac{V(t) - V_1}{R_1} = C_1 \frac{dV_1}{dt} + \frac{V_1 - V_2}{R_2} \quad [1] + \frac{1}{L_1} \int_{-\infty}^t V d\tau = [1]$$

At $t=0^+$:-

$$\frac{V(0^+) - V_1(0^+)}{R_1} = C_1 \frac{dV_1(0^+)}{dt} + \frac{V_1(0^+) - V_2(0^+)}{R_2} + \frac{1}{L_1} \int_{-\infty}^{0^+} V d\tau - i_{L_1}(0^+) \quad [2]$$

$$\therefore \frac{V(0^+)}{R_1} = C_1 \frac{dV_1(0^+)}{dt} + \frac{0}{R_2} + 0$$

$$\therefore \boxed{\frac{dV_1(0^+)}{dt} = \frac{V(0^+)}{R_1 C_1}} - [2]$$

c) At node V_2 :

$$\frac{V_1 - V_2}{R_2} = C_2 \frac{dV_2}{dt} + \frac{1}{L_2} \int_{-\infty}^t V_2 d\tau - i_{L_2}(0^+) \quad [3]$$

$$\frac{V_1(0^+) - V_2(0^+)}{R_2} = C_2 \frac{dV_2(0^+)}{dt} + \frac{1}{L_2} \int_{-\infty}^{0^+} V_2 d\tau$$

$$\therefore \frac{0}{R_2} = C_2 \frac{dV_2(0^+)}{dt} + 0$$

$$\boxed{\frac{dV_2(0^+)}{dt} = 0 \text{ V/s}} - [4]$$

(2)

At node V_3 : -

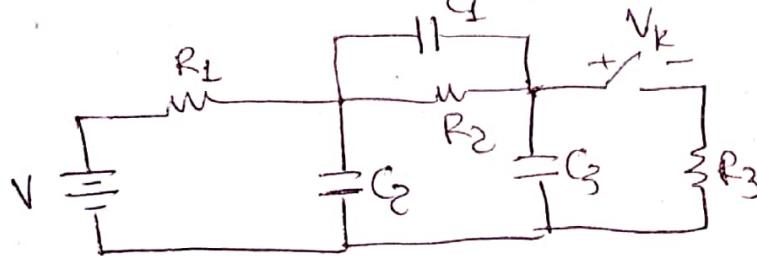
$$\frac{1}{L_1} \int_{-\infty}^t (V_L - V_3) d\tau = C_3 \frac{dV_3}{dt} \quad [5]$$

At $t=0^+$

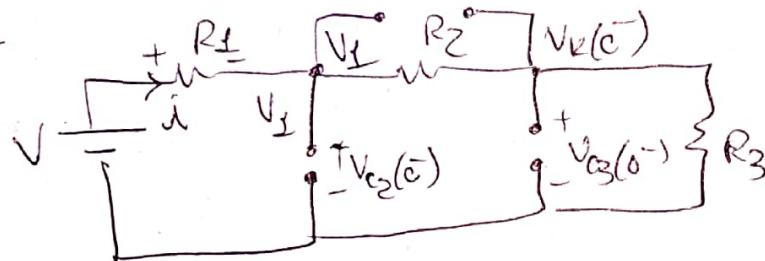
$$i_L(0^+) = C_3 \frac{dV_3(0^+)}{dt} \quad \therefore \frac{dV_3(0^+)}{dt} = 0 \text{ V/S}$$

$$0 = C_3 \frac{dV_3(0^+)}{dt}$$

- (3) In the network shown, steady state is reached, and at $t=0$, switch 'K' is opened. (a) Find the voltage across switch, V_K at $t=0^+$; (b) Find $\frac{dV_K}{dt}$ at $t=0^+$.



a) At $t=0^-$:-



At $t=0^-$ steady state is reached and all capacitors acts like O.C.

$$V(0^-) = V_{C_2}(0^-) \quad \therefore i(0^-) = \frac{V}{R_1 + R_2 + R_3} \quad [6]$$

$$+ V_{C_2}(0^-) + R_1 i - V = 0 \quad \therefore V_2(0^-) = V - R_1 i \quad \text{cancel}$$

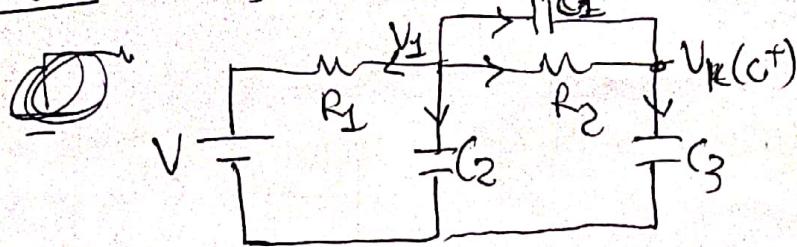
$$\therefore V_2(0^-) = V_1(0^-) = V_{C_2}(0^+) = V - \frac{VR_1}{R_1 + R_2 + R_3}$$

$$V_1 = \left[V_2(0^-) = \frac{V(R_2 + R_3)}{R_1 + R_2 + R_3} \right] = V_1 \quad [7]$$

$$V_{C_3}(0^-) = V_K(0^-) = V_K(0^+) = V - R_3 i(0^-) R_3 = \frac{VR_3}{R_1 + R_2 + R_3} \quad [8]$$

(3)

5] At $t=c^+$:- R_3 is not included in circuit.



Applying KCL at node V_1 :

$$\left[\frac{V_1 - V}{R_1} + C_2 \frac{dV_1}{dt} + C_1 \frac{d(V_1 - V_R)}{dt} + \frac{V_R - V_1}{R_2} = 0 \right] \rightarrow [4]$$

$$\therefore [C_1 + C_2] \frac{dV_1}{dt} = C_1 \frac{dV_R}{dt} + \frac{V}{R_1} - \frac{V_1}{R_1} - \frac{V_1}{R_2} + \frac{V_R}{R_2}$$

$$\therefore [C_1 + C_2] \frac{dV_1}{dt} = C_1 \frac{dV_R}{dt} + \frac{V}{R_1} - \frac{V(R_2 + R_3)}{R_1(R_1 + R_2 + R_3)} - \frac{V(R_2 + R_3)}{R_2(R_1 + R_2 + R_3)} + \frac{VR_3}{R_2(R_1 + R_2 + R_3)}$$

$$\therefore [C_1 + C_2] \frac{dV_1}{dt} = C_1 \frac{dV_R}{dt} + \frac{V}{R_1} - \frac{V(R_2 + R_3)}{R_1(R_1 + R_2 + R_3)} - \frac{VR_2}{R_2(R_1 + R_2 + R_3)}$$

$$\therefore [C_1 + C_2] \frac{dV_1}{dt} = C_1 \frac{dV_R}{dt} + \frac{1}{R_1} \left[\frac{VR_1}{R_1 + R_2 + R_3} \right] - \frac{V}{R_1 + R_2 + R_3}$$

$$\therefore [C_1 + C_2] \frac{dV_1}{dt} = C_1 \frac{dV_R}{dt} \quad [5]$$

continuation of problem 3(5.28)

Applying KCL at node V_k :

$$C_1 \frac{d(V_1 - V_k)}{dt} + \frac{V_1 - V_k}{R_2} = C_3 \frac{dV_k}{dt} \quad (6)$$

$$\therefore C_1 \frac{dV_k}{dt} + \frac{1}{R_2} [V_1 - V_k] = [C_1 + C_3] \frac{dV_k}{dt}$$

$$\therefore C_1 \frac{dV_1}{dt} + \frac{1}{R_2} \left[\frac{V(R_2 + R_3)}{R_1 + R_2 + R_3} - \frac{VR_3}{R_1 + R_2 + R_3} \right] = [C_1 + C_3] \frac{dV_k}{dt}$$

$$\therefore C_1 \frac{dV_1}{dt} + \frac{1}{R_2} \left[\frac{VR_2}{R_1 + R_2 + R_3} \right] = [C_1 + C_3] \frac{dV_k}{dt}$$

$$\therefore C_1 \frac{dV_k}{dt} + \frac{V}{R_1 + R_2 + R_3} = [C_1 + C_3] \frac{dV_k}{dt} \quad (7)$$

Substitute eq-(5) in eq-(7) for $\frac{dV_1}{dt}$:

$$\therefore \frac{C_1^2}{C_1 + C_3} \frac{dV_k}{dt} + \frac{V}{R_1 + R_2 + R_3} = [C_1 + C_3] \frac{dV_k}{dt}$$

$$\therefore \frac{dV_k}{dt} \left[\left(C_1 + C_3 \right) - \frac{C_1^2}{C_1 + C_3} \right] = \frac{V}{R_1 + R_2 + R_3}$$

$$\therefore \frac{dV_k}{dt} \left[\frac{C_1^2 + C_1 C_2 + C_3 C_1 + C_2 C_3}{C_1 + C_3} - C_1^2 \right] = \frac{V}{R_1 + R_2 + R_3}$$

$$\therefore \frac{dV_k(t)}{dt} = \frac{V(C_1 + C_2)}{[(C_1 C_2 + C_2 C_3 + C_3 C_1)(R_1 + R_2 + R_3)]}$$

Q.23 A bolt of lightning having a waveform which is approximated as $V(t) = t e^t$ strikes a transmission line having resistance $R = 0.1 \Omega$ and inductance $L = 0.1 H$ (the line-to-line capacitance is assumed to be negligible). An equivalent circuit is shown in the accompanying diagram. What is the form of the current as a function of time? (This current will be in ampere per unit volt of lightning; likewise the time-base is normalized).

$$V(t) = L \frac{di}{dt} + Ri \quad [1] \text{---a}$$

$$\frac{di}{dt} + \frac{Ri}{L} = \frac{1}{L} V(t) \quad [1] \text{---b}$$

$$\frac{di}{dt} + i = 10t e^{-t} \quad [1] \text{---c}$$

Direct method: - Let, complementary sol. in: - $i_c(t) = K e^{st}$ - [2] - a

$$\frac{di}{dt} + i = 0 \quad \therefore K s e^{st} + K e^{st} = 0$$

$$\therefore s + 1 = 0 \quad \therefore s = -1 \quad i_c(t) = K e^{-t} \quad [2] \text{---b}$$

2) Particular solution:-

Given $V(t) = t e^{-t}$

$$i_p(t) = [A_0 t + A_1] e^{-t}$$

$$i_p(t) = A_0 t e^{-t} + A_1 e^{-t} \rightarrow \text{Same form as } i_c(t)$$

↳ Change the form

$$i_p(t) = A_0 t^2 e^{-t} + A_1 t e^{-t} = [A_0 t + A_1] t e^{-t} \quad [3]$$

Substituting eq-(3) in eq-[1]-c:-

$$\therefore \frac{d}{dt} \{ [A_0 t^2 e^{-t} + A_1 t e^{-t}] \} + [A_0 t^2 e^{-t} + A_1 t e^{-t}] = 10 t e^{-t}$$

$$\therefore -A_0 t^2 e^{-t} + A_0 2t e^{-t} + A_1 e^{-t} - A_1 t e^{-t} + A_0 t^2 e^{-t} + A_1 t e^{-t} = 10 t e^{-t} + 0 \cdot e^{-t}$$

$$2A_0 = 10 \quad (A_0 = 5) \quad A_1 = 0 \quad \therefore i_p(t) = 5t^2 e^{-t} \quad [4]$$

complete response: - $i(t) = i_c(t) + i_p(t)$ - [5]

$$i(t) = K e^{-t} + 5t^2 e^{-t} \quad [6]$$

(5)

Initial condition $i(0) = 0 \Rightarrow A = i(0)$ → used to find the constant in [6]

$$\therefore i(0) = 0 = K e^{-0} + 0 \therefore K = 0$$

$$\therefore I(Y) = 5t^2 e^{st} A - [7]$$

Q) Laplace Transform method:- Taking L.T. of eq. f.i.g.c :-

$$SI(S) - i(0) + I(S) = \frac{10}{(S+1)^2} \rightarrow [8]$$

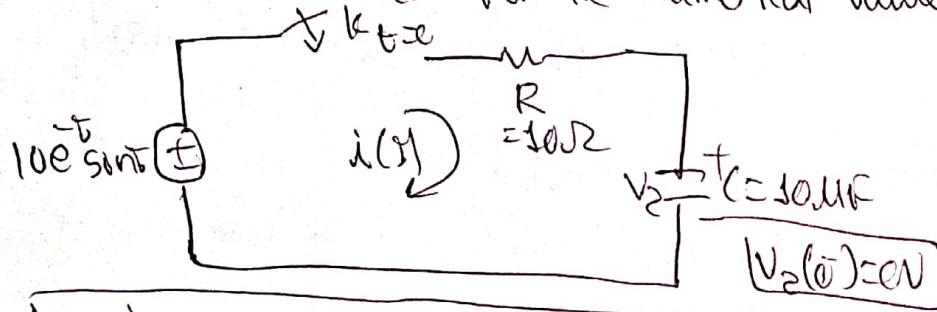
$$\therefore SI(S) + I(S) = \frac{10}{(S+1)^2}$$

$$\therefore i(Y) = \frac{10 + t^2 e^t}{2!}$$

$$\therefore I(S) = \frac{10}{(S+1)^3} \rightarrow [9]$$

$$\therefore I(Y) = 5t^2 e^{st} A$$

(5) In the network shown switch 'K' is closed at $t=0$, with capacitor initially uncharged. For the numerical values given find $i(t)$.



$$10e^{-t} \sin(5t) = R i(t) + \frac{1}{C} \int_{-\infty}^t i(z) dz \rightarrow [1]$$

$$R \frac{di}{dt} + \frac{1}{C} i = 10 \cos t e^{-t} - 10e^{-t} \sin t$$

$$\therefore \frac{di}{dt} + 10^4 i = e^{-t} \cos t - e^{-t} \sin t \rightarrow [2]$$

a) Complementary soln:- $i_c(t) = K e^{st}$ → [3] a $s + 10^4 = 0 \quad (s = -10^4)$

$$\therefore i_c(t) = K e^{-10^4 t} \rightarrow [3] b$$

⑥

particular solution:-

$$i_p(t) = F_1 e^{-t} \cos t + F_2 e^{-t} \sin t \quad (6)$$

Substituting Eq-(6) in Eq-(2):-

$$10^4 F_1 e^{-t} \cos t + 10^4 F_2 e^{-t} \sin t - F_1 e^{-t} \cos t - F_2 e^{-t} \sin t - F_2 e^{-t} \sin t \\ + F_2 e^{-t} \cos t = e^{-t} \cos t - e^{-t} \sin t$$

$$e^{-t} \cos t [9999 F_1 + F_2] + [-F_1 + 9999 F_2] = e^{-t} \cos t - e^{-t} \sin t$$

$$\therefore 9999 F_1 + F_2 = 1 \quad (5a)$$

$$-F_1 + 9999 F_2 = -1 \quad (5b) \times 9999$$

$$\therefore 9999 F_1 + F_2 = 1$$

$$-9999 F_1 + 99.98 \times 10^6 F_2 = -9999$$

$$99.98 \times 10^6 F_2 = 9999$$

$$\therefore F_2 = -10^{-4}$$

$$\therefore F_1 = 10^{-4}$$

$$\therefore i(t) = i_c(t) + i_p(t) \quad (6)$$

$$\therefore i(t) = K e^{-10^4 t} + 10^{-4} e^{-t} \cos t - 10^{-4} e^{-t} \sin t \quad (7)$$

$$i(0) \text{ w.r.t. } \quad i(0) = 0 \quad (8a)$$

$$\therefore i(0) = 0 = K + 10^{-4}$$

$$\therefore K = -10^{-4} \quad (8b)$$

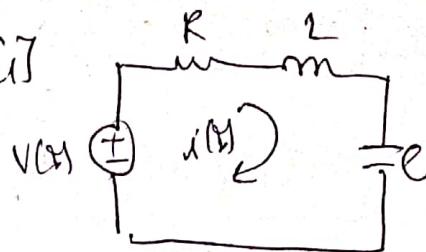
$$\therefore i(t) = -10^{-4} e^{-10^4 t} + 10^{-4} e^{-t} (\cos t - 10^{-4} e^{-t} \sin t) \quad (9)$$

(7)

- ⑥ 6.29 Consider a series R-L-C Ckt. which is excited by a voltage $V(t)$.
 (a) Determine the characteristic eqn. corresponding to differential eqns. $i(t)$; (b) Suppose, L and C are fixed in value but 'R' varies from 0 to ∞ . What will be the locus & the roots of the characteristic eqn.? (c) Plot the roots of characteristic eqn. in s-plane if $L=1H$; $C=1MF$ and 'R' has the values (i) 500Ω ; (ii) $1k\Omega$; (iii) $3k\Omega$; (iv) $5k\Omega$

$$V(s) = R i(s) + L \frac{di}{dt} + \frac{1}{C} \int_0^t i dt \quad [1]$$

Dif. Eq - [1]:-



$$\frac{L di}{dt} + R \frac{di}{dt} + \frac{1}{C} i = \frac{1}{C} V(t)$$

$$\frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = \frac{1}{L} \frac{dV(t)}{dt} \quad [2]$$

Characteristic eqn. is given for homogeneous eqn.

$$\frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = 0 \quad [3] \quad \text{Put, } i = k e^{st} \text{ we get the}$$

$$\text{Characteristic eqn.:- } s^2 + \frac{R}{L} s + \frac{1}{LC} = 0 \quad [4]$$

$$s_{1,2} = -\frac{R}{2L} \pm \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}} \quad [5] \quad \text{Put, } 2\zeta \omega_n = \frac{R}{L} \text{ and } \omega_n = \frac{1}{\sqrt{LC}}$$

$$\therefore s_{1,2} = -\zeta \omega_n \pm j \omega_n$$

$$\text{we define, } \xi = \frac{R}{2\sqrt{LC}} \quad [6]$$

$$\therefore s_{1,2} = -\xi \omega_n \pm \omega_n \sqrt{\xi^2 - 1} \quad [6]$$

When R varies from $0 \rightarrow \infty \Rightarrow \xi$ varies from $0 \rightarrow \infty$.

(8)

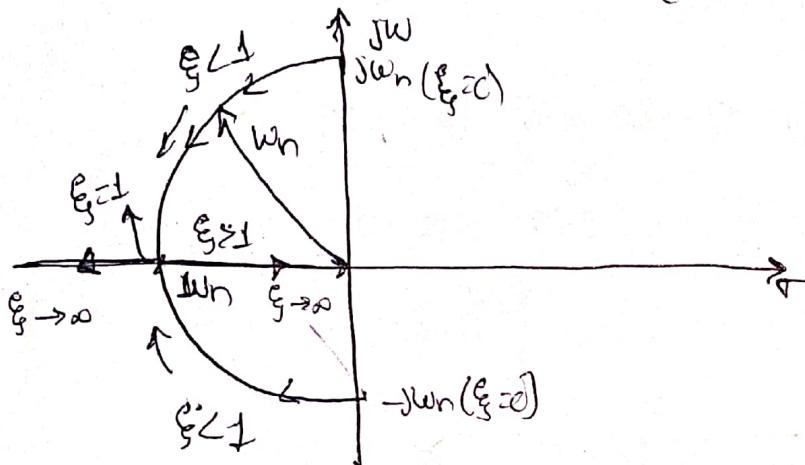
When $\xi \rightarrow 0$ $s_1, 2 = \pm j\omega_n$ roots are complex conjugate and purely imaginary. The response is oscillatory. The roots lie on imaginary axis $j\omega$.

case-i) When $\xi \rightarrow 0 < \xi < 1$ $s_{1,2} = -\xi\omega_n \pm j\omega_n\sqrt{1-\xi^2} = \tau \pm j\omega$
 $\tau^2 + \omega^2 = \xi^2\omega_n^2 + \omega_n^2(1-\xi^2) = \omega_n^2$, the roots move on a circle described by a circle of radius ω_n .

case-ii) When $\xi = 1$:— $s_{1,2} = -\omega_n$ The roots are real, negative and equal where, $\tau = -\omega_n$ on real line.

case-iv) When $\xi > 1$:— $s_{1,2} = -\xi\omega_n \pm j\omega_n\sqrt{\xi^2 - 1}$ as $\xi^2 > 1$

$\therefore s_{1,2} = -\xi\omega_n \pm j\omega_n\xi$ $s_1 = -2\xi\omega_n$ $s_2 = 0$
 As $\xi \rightarrow \infty$, $s_1 \rightarrow -\infty$ on one end $s_2 = 0$ on another end.



(C) Location of roots:-

$$R = 500 \Omega$$

$$L = 1H$$

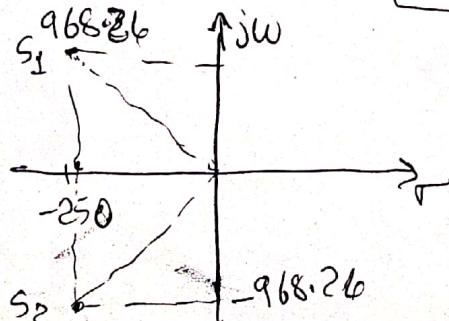
$$C = 1 \mu F$$

$$\omega_n = \frac{1}{\sqrt{LC}} = 1000 \text{ rad/s}$$

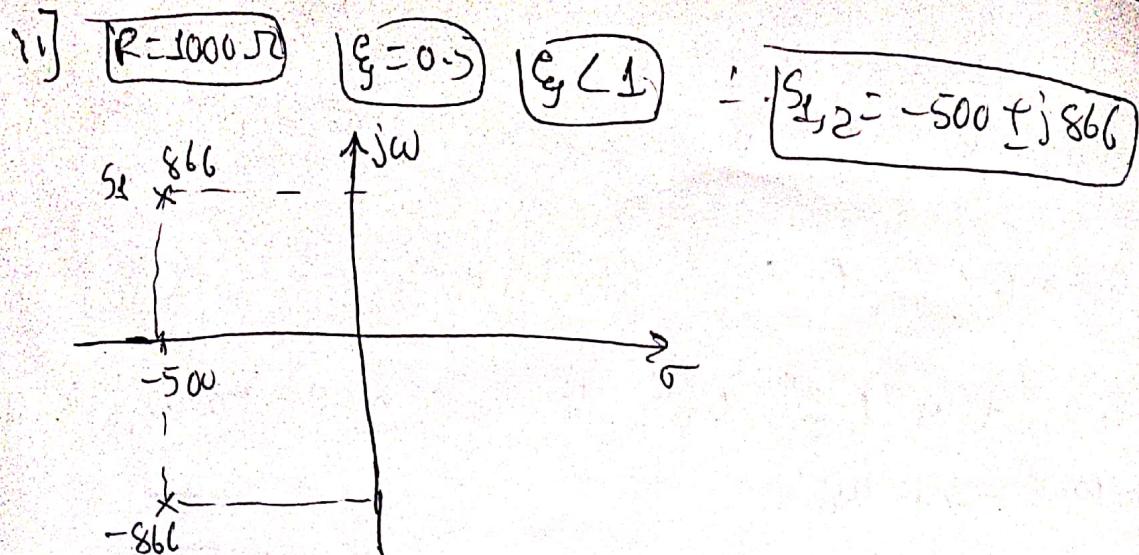
$$\xi = \frac{R}{2\sqrt{LC}} = \frac{500}{2\sqrt{1 \times 10^{-6}}} = 10.25$$

$\xi > 1 \rightarrow$ Underdamped

$$s_{1,2} = -\xi\omega_n \pm j\omega_n\sqrt{1-\xi^2} = [-250 \pm j968.24]$$

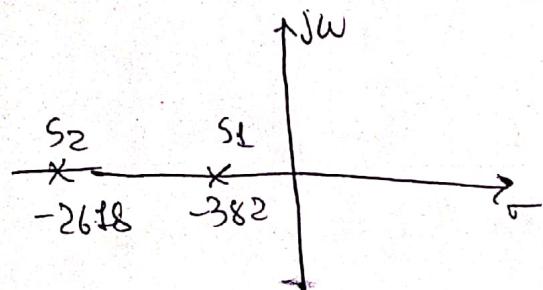


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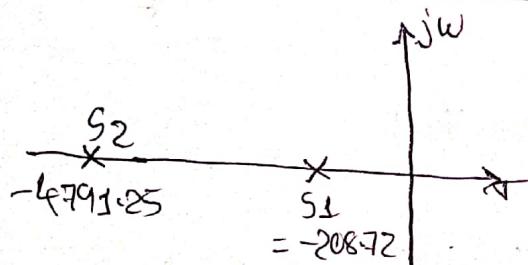
ii) $R = 3000 \Omega$ $\xi = \frac{3000}{2} \times 10^{-3} = 1.5$ $\xi > 1$ $S_{1,2} = -\xi u_n \pm u_n \sqrt{\xi^2 - 1}$

$$S_{1,2} = -1500 \pm 1118 = -382, -2618$$



iii) $R = 5000 \Omega$ $\xi = \frac{5000}{2} \times 10^{-3} = 2.5$ $\xi > 1$

$$S_{1,2} = -2500 \pm 2291.28 = -208.72, -4791.25$$



- 33 A switch is closed at $t=0$, connecting a battery of voltage V with a series RL -circuit. (a) Show that, energy in the resistor is ~~in time~~ as a function of time is $w_R = \frac{V^2}{R} \left(t + \frac{2L}{R} e^{-tR/L} - \frac{L}{2R} e^{-2tR/L} - \frac{3L}{2R} \right)$ Joules
- (b) Find an expression for energy in magnetic field as a function of time; (c) Sketch w_R and w_L as a function of time. Show the steady state asymptotes, that is, the values that w_R and w_L approach as $t \rightarrow \infty$; (d) Find the total energy supplied by the voltage source in steady state as function of time.

$$\therefore i(t) = i(\infty) + [i(0) - i(\infty)] e^{-tR/L}$$

$$i(0) = CA + f_2 \quad i(\infty) = \frac{V}{R} A - f_2$$

$$\therefore i(t) = \frac{V}{R} [1 - e^{-tR/L}] A - f_2$$

$$\boxed{i(t) = i(\infty) + [i(0) - i(\infty)] e^{-tR/L}}$$

$$\boxed{i(0) = CA + f_2}$$

$$\boxed{i(\infty) = \frac{V}{R} A - f_2}$$

$$\boxed{i(t) = \frac{V}{R} [1 - e^{-tR/L}] A - f_2}$$

$$(a) w_R = \int_0^t i^2(t) R dt = \frac{V^2}{R^2} \times R \int_0^t [1 - e^{-tR/L}]^2 dt$$

$$\therefore w_R = \frac{V^2}{R} \int_0^t [1 + e^{-2tR/L} - 2e^{-tR/L}] dt$$

$$\therefore w_R = \frac{V^2}{R} \left[t + \frac{e^{-2tR/L}}{2R} \times L \Big|_0^t + 2 \frac{e^{-tR/L}}{R} \times L \Big|_0^t \right]$$

$$\therefore w_R = \frac{V^2}{R} \left[t - \frac{L}{2R} [\frac{e^{-2Rt}}{2} - 1] + \frac{2L}{R} [e^{-tR/L} - 1] \right]$$

$$\therefore w_R = \frac{V^2}{R} \left[t + \frac{2L}{R} e^{-tR/L} - \frac{L}{2R} e^{-2tR/L} + \frac{L}{2R} - \frac{2L}{R} \right]$$

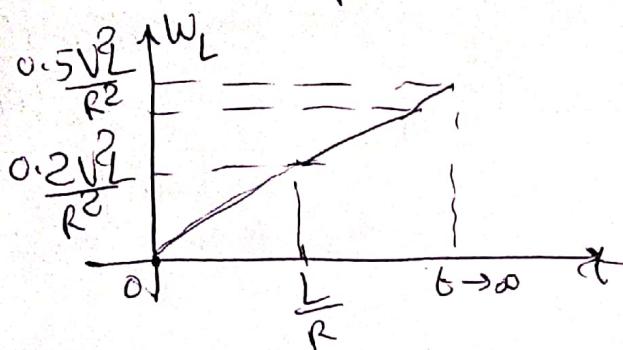
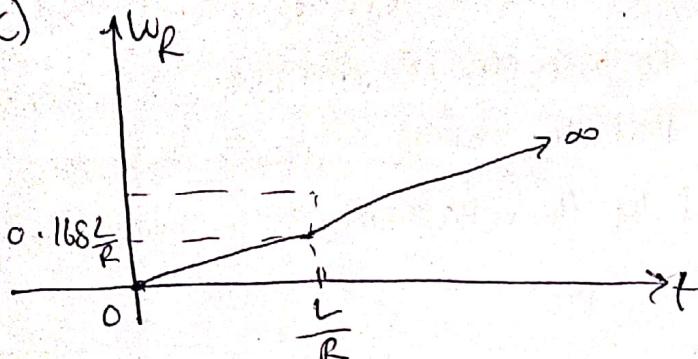
$$\boxed{\therefore w_R = \frac{V^2}{R} \left[t + \frac{2L}{R} e^{-tR/L} - \frac{L}{2R} e^{-2tR/L} - \frac{3L}{2R} \right] \text{ Joules}}$$

⑪

$$(C) \quad w_L = \frac{1}{2} L \dot{\theta}_0^2 (t)$$

$$\therefore \boxed{w_L = \frac{1}{2} \times L \times \frac{V^2}{R^2} [1 - e^{-tRL}]^2}$$

(C)



(d)

$$\text{In steady state: } - w_s = \int_0^t V_i \cdot \frac{V}{R} dt + \left(\frac{R}{R} t \right)$$

$$\text{As a function of time: } - w_s = \int_0^t V_i(t) dt = \frac{V^2}{R} \int_0^t [1 - e^{-tRL}] dt$$

$$\therefore w_s = \frac{V^2}{R} \left[t + \frac{e^{-tRL}}{RL} \right] \Big|_0^t$$

$$\therefore w_s = \frac{V^2}{R} \left[t + \frac{L}{R} [e^{-RL} - 1] \right]$$

$$\therefore \boxed{w_s = \frac{V^2}{R} \left[t - \frac{L}{R} (1 - e^{-RL}) \right]}$$