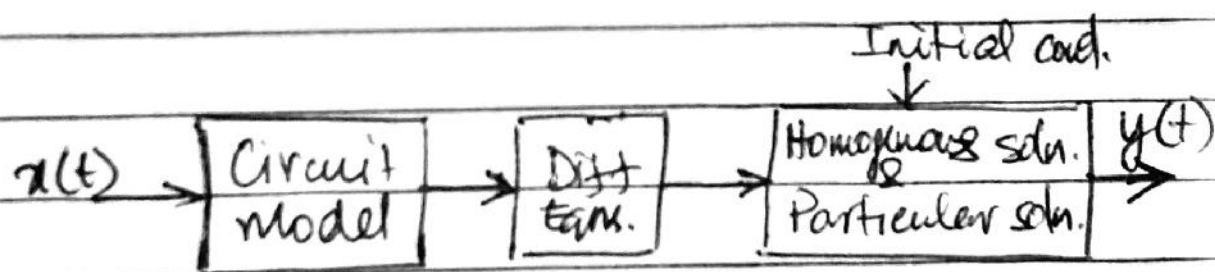


INTRODUCTION TO FREQUENCY ANALYSIS

Review of Dynamic Circuit Analysis Method. (For Linear circuits)

- * Simplify circuit if possible (series res, etc)
- * Formulate the differential equation using nodal equation method.
- * Solve the equations. Find the homogenous and particular soln.
- * Use initial conditions to evaluate the constants in the homogenous solution.



- * This method is basic yet powerful.
- * But math is complex when input is complex

So we look at a method which simplifies the analysis of dynamic systems considerably for certain domain of linear & LTI systems.

The method involves steady state analysis of sinusoidal drive. for any arbitrary signal can be constructed from linear combination of sinesoids. Since we are analyzing linear systems we can solve the problem of arbitrary input with superposition of sinusoid responses.

Moreover, a large variety of analog circuits (eg. audio amps, filters, etc.) are characterized by sinusoid response.

Solution to linear dynamic circuits are greatly simplified by assuming a drive of the form e^{st} .

Because the differential eqns. transform to a set of algebraic equation.

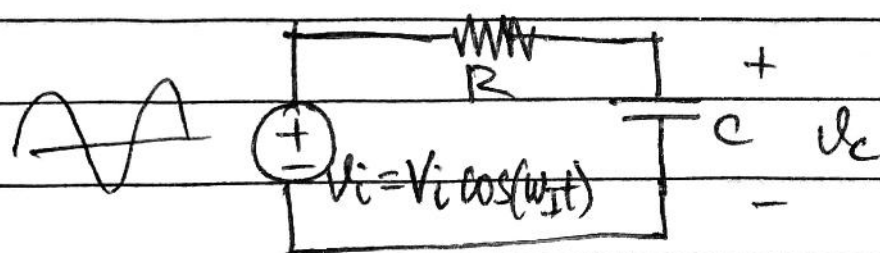
Why e^{st} ?

e^{st} (for $s = j\omega$) $\xRightarrow{\text{Euler Identity}}$ $\underbrace{\cos(\omega t)}_{\text{real}} + j \underbrace{\sin(\omega t)}_{\text{imaginary}}$

After finding the solution,

- the real part of the soln is due to $\cos(\omega t)$
- the imaginary part of the soln is due to $\sin(\omega t)$

EXAMPLE: FIRST-ORDER RC CIRCUIT



The differential equation for the ckt:

$$V_i = V_c + RC \frac{dV_c}{dt}$$

Solution: $V_c = V_{ch} + V_{cp}$

homogeneous

particular

On solving (with some trigonometric stuff.)

we get:

$$V_{ch} = K_1 e^{-t/RC}$$

&

$$V_{cp} = K_2 \sin(\omega t) + K_3 \cos(\omega t)$$

Nothing wrong with this method, just some math work involved which gets increasingly hard for complex obs.

Alternatively:

The Euler relation.

$$e^{j\omega t} = \underbrace{\cos(\omega t)}_{\text{wanted}} + j \underbrace{\sin(\omega t)}_{\text{unwanted.}}$$

Let's use the stimulus

$$\tilde{V}_i = V_i e^{s_1 t} \quad \text{where } s_1 = j\omega_1$$

"~" signifies the input stimulus carries both the wanted (real) & unwanted (imag)

The homogeneous solution is the same.

The diff eqn. for the particular solution

$$\vec{v}_c = V_c e^{s_1 t} = \vec{v}_{cp} + RC \frac{d\vec{v}_{cp}}{dt}$$

A reasonable assumption for \vec{v}_{cp}

$$v_{cp} = V_c e^{st}$$

$$V_c e^{s_1 t} = V_c e^{st} + RC \cdot V_c e^{st}$$

s must equal s_1 to satisfy the equation for all t .

☞ This where magic happens!

☞ $e^{s_1 t}$ cannot be zero for positive values of t , it can be factored out.
Resulting in,

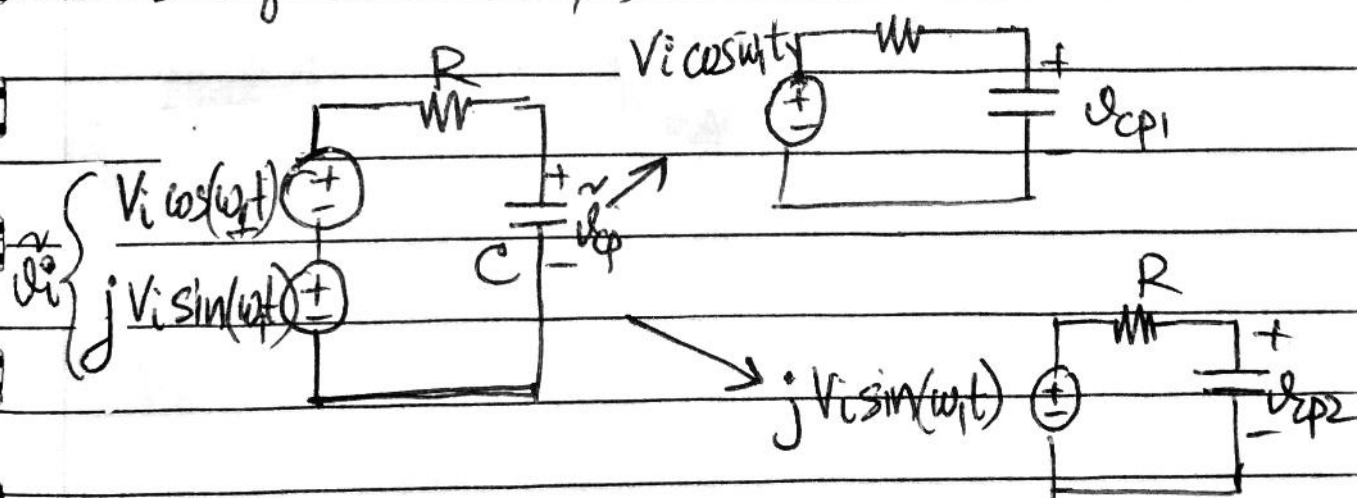
$$V_c = V_c + V_c RC s_1$$

So a differential equation gets converted to algebraic equation which simplifies the analysis dramatically.

The particular solution

$$\tilde{v}_{cp} = v_{cp1} + j v_{cp2} = \frac{V_i}{1 + j\omega_1 RC} e^{j\omega_1 t}$$

Diagrammatically,



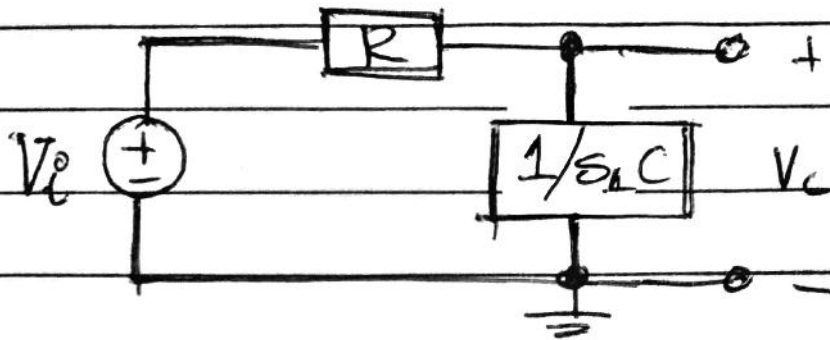
The real part of the particular soln.

$$v_{cp1} = \frac{V_i}{\sqrt{1 + (\omega_1 RC)^2}} \cos(\omega_1 t + \phi)$$

which is the steady state solution to $V_i \cos(\omega_1 t)$.

$$\phi = \tan^{-1}\left(\frac{-\omega_1 RC}{1}\right)$$

A circuit interpretation of the above equation:



This is the basis of doing circuit analysis using Laplace transform (s-domain, frequency domain all diff names.).