ECEN452: ULTRA HIGH FREQUENCY TECHNIQUE

LAB01

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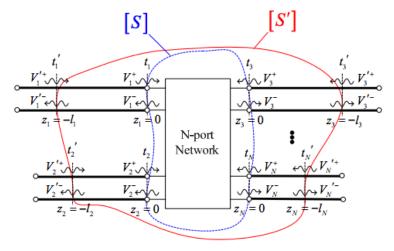
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Calculating the two-port S and ABCD matrices for a series impedance  $Z=10+j25\,\Omega$  using a system impedance  $Z_o=50\Omega$  and the frequency sweep parameters from the simulations.

Sparameters from the simulations. 
$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 & Z \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 1 & 10 + j25 \\ 0 & 1 \end{bmatrix}$$
$$S = \begin{bmatrix} \frac{A + BY_0 - CZ_0 - D}{\Delta} & \frac{2(AD - BC)}{\Delta} \\ \frac{2}{\Delta} & \frac{-A + BY_0 - CZ_0 + D}{\Delta} \end{bmatrix}$$

 $\Delta = A + BY_0 + CZ_0 + D = 2.2 + j0.5$   $S = \begin{bmatrix} 0.13556 + j0.19646 & 0.86444 - j0.19646 \\ 0.86444 - j0.19646 & 0.13556 + j0.19646 \end{bmatrix} = \begin{bmatrix} 0.2387e^{j55.4} & 0.8865e^{-j12.804} \\ 0.8865e^{-j12.804} & 0.2387e^{j55.4} \end{bmatrix}$ 

Now, for calculating shifted reference plane, it looks like the following:



Figure

$$[S'] = \begin{bmatrix} e^{-j\theta_1} & 0 \\ & \ddots & \\ 0 & e^{-j\theta_N} \end{bmatrix} \cdot [S] \cdot \begin{bmatrix} e^{-j\theta_1} & 0 \\ & \ddots & \\ 0 & e^{-j\theta_N} \end{bmatrix}$$

Figure

Now, we have all the ingredients to calculate the new S-parameter.

$$\theta = \beta l$$

Port 1 shifted by  $0.8\lambda : \theta_1 = 288^\circ$ 

Port 2 shifted by 
$$0.25\lambda$$
 :  $\theta_2=90^\circ$  
$$S'=\begin{bmatrix}e^{-j288}&0\\0&e^{-j90}\end{bmatrix}S\begin{bmatrix}e^{-j288}&0\\0&e^{-j90}\end{bmatrix}$$
 >> shift\*S\*shift

ans =

Figure. MATLAB Calculation

$$S' = \begin{bmatrix} 0.2387e^{j176.56} & 0.8845e^{j228.75} \\ 0.8845e^{j228.75} & 0.2387e^{j176.56} \end{bmatrix}$$

this network is reciprocal because the S-parameter is symmetric.

#### Substrates Table:

|                          | FR4  | Duroid | Duroid | Duroid |  |
|--------------------------|------|--------|--------|--------|--|
|                          |      | 5880   | 6006   | 6010.2 |  |
| $oldsymbol{arepsilon_r}$ | 4.4  | 2.2    | 6.15   | 10.2   |  |
| Tanδ                     | 0.02 | 0.0009 | 0.0027 | 0.0023 |  |
|                          |      | TI     |        |        |  |

Figure

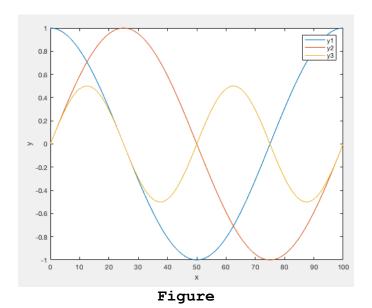
### Connector Types:

|        | Type N | SMA | 3.5mm | 2.92mm | 2.4mm | 1.85mm |
|--------|--------|-----|-------|--------|-------|--------|
| Type N | Y      |     |       |        |       |        |
| SMA    |        | Y   | Y     | Y      |       |        |
| 3.5mm  |        | Y   | Y     | Y      |       |        |
| 2.92mm |        | Y   | Y     | Y      |       |        |
| 2.4mm  |        |     |       |        | Y     |        |
| 1.85mm |        |     |       |        | Y     | Y      |

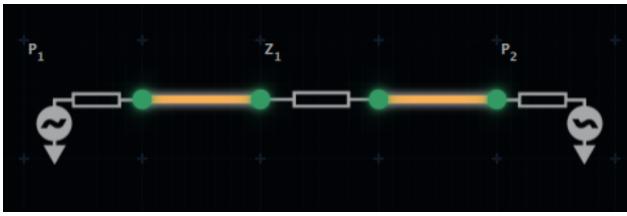
Figure

Matching Rule: Mechanical & Electrical Characteristics conditions such as the operating frequency.

### PLOT:



# FIRST CIRCUIT:



Figure

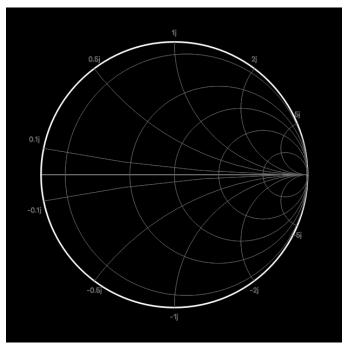


Figure. Smith Chart

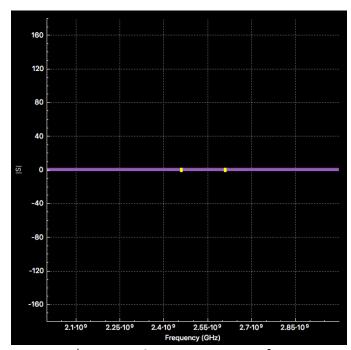


Figure. S-parameter Plot

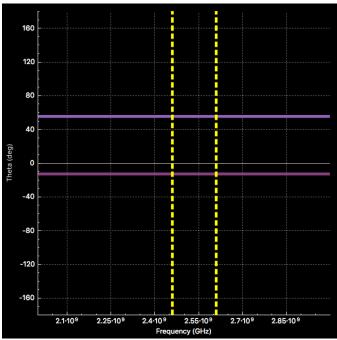


Figure. Phase Plot

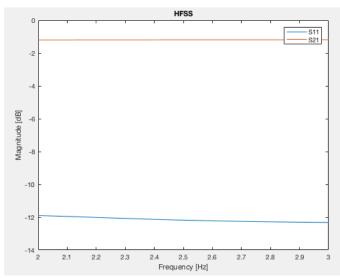
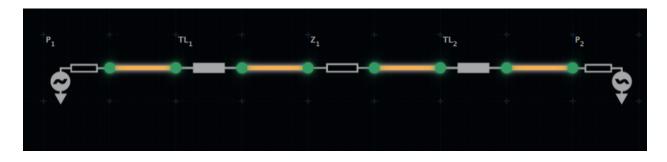


Figure. HFSS dB plot

## SECOND CIRCUIT:



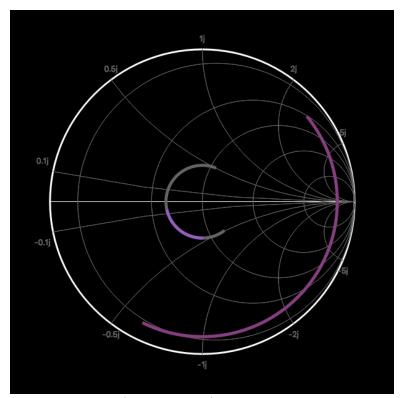
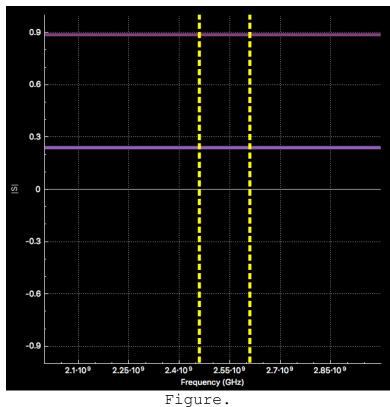


Figure. Smith Chart



Around the design frequency (f = 2.45 GHz),  $S_{11} = 0.238693$ 

$$S_{12} = 0.886484$$
  
 $S_{21} = 0.886484$   
 $S_{22} = 0.238693$ 

Thus, we can simply put them all together into a pretty matrix.

$$S = \begin{bmatrix} 0.238693 & 0.886484 \\ 0.886484 & 0.238693 \end{bmatrix}$$

Note that these are just the magnitude portion of S-parameter and the matrix is symmetric. The network could have been a reciprocal network if phase matrix is also symmetric; however, the phase matrix shows the following:

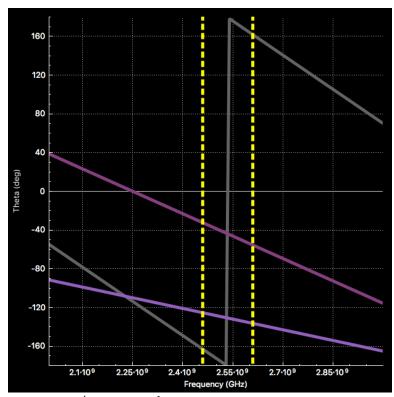


Figure. Phase vs. Frequency

$$S_{11} = -162.961$$

$$S_{12} = -32.3651$$

$$S_{21} = -32.3651$$

$$S_{22} = -125.372$$

$$S = \begin{bmatrix} -162.961 & -32.3651 \\ -32.3651 & -125.372 \end{bmatrix}$$

These values are just the phase portion of the S-parameter. Thus, the network is not reciprocal. Reciprocity of a circuit has a somewhat meaningful property because the circuit looks 'unchanged' no matter from which port is was seen in a two network case.