

ECEN452: ULTRA HIGH FREQUENCY TECHNIQUE
LAB02
SAMBONG JANG
DR. HUFF

Reference:

"Learn Stub Tuning with a Smith Chart" by Benjamin Crabtree, October 09, 2015
 Microwave Engineering 4th Edition, David M. Pozar

Quarter Wave Transformer

$$\epsilon_r = 2.2 \text{ (Duroid 5880)}$$

$$h = 1.5748\text{mm (thickness of substrate)}$$

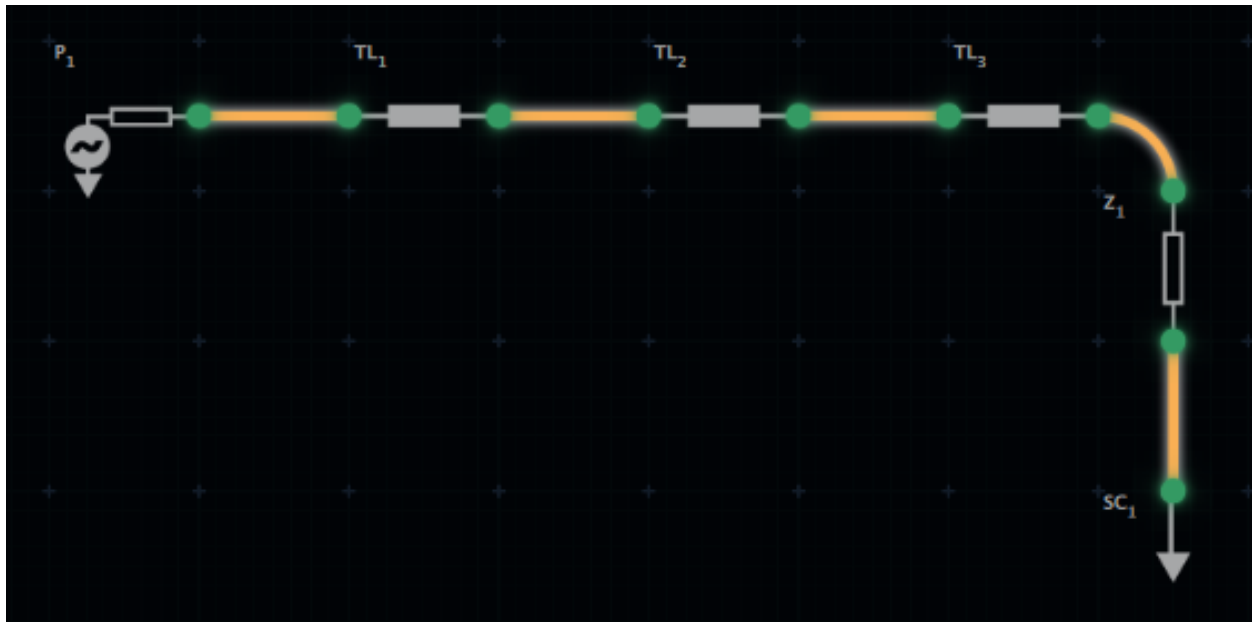
$$f = 2.45\text{GHz}$$

Z_0 [Ω]	W [mm]	λ [mm]	$\lambda/4$ [mm]	ϵ_{eff}
50	4.7725	89.068		1.8899
100	1.396		22.993	1.773
200	0.1654	94.268		1.6877

Figure. Microstrip line Calculation

CURRENT NETLIST ⓘ					
Component	Real Impedance		Imag Impedance		Length
P1					
TL1	50	none ▾			89.10 m ▾
TL2	100	none ▾			23.00 m ▾
TL3	200	none ▾			94.25 m ▾
Z1	200	none ▾	0	f ▾	
SC1					

Figure



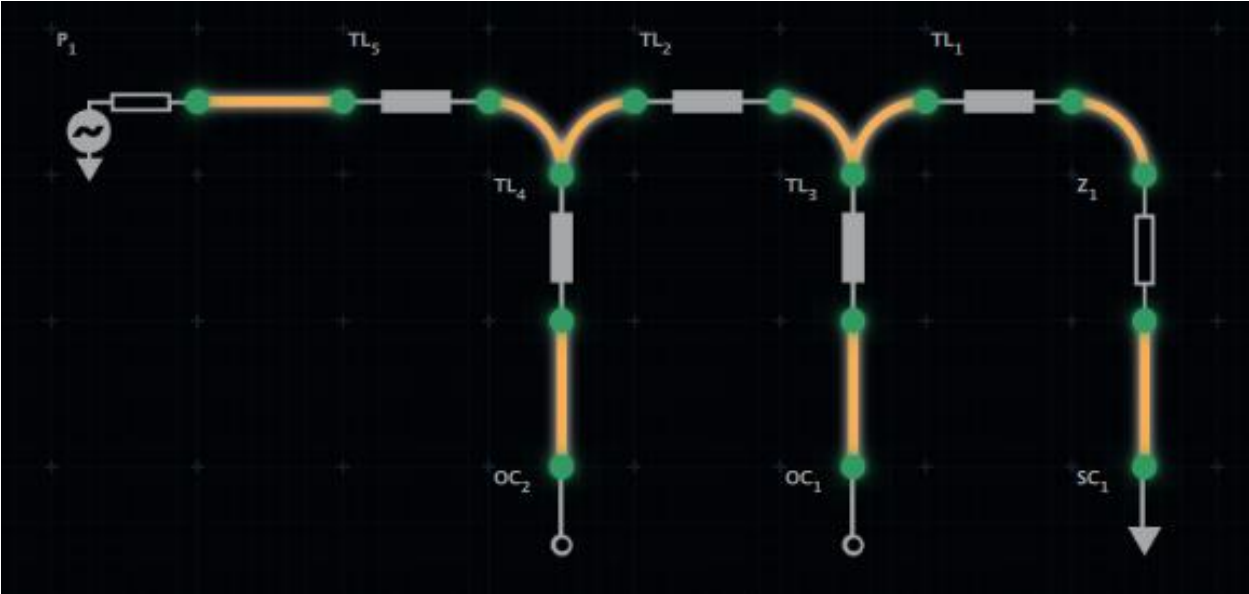
Figure

DOUBLE STUB MATCHING

BACKGROUND :

From single stub matching technique, we can obtain impedance matching on that particular frequency that we started with and perhaps, quite a small range of bandwidth around that particular frequency due to an addition of a single stub to cancel out the reactive component of the load.

In double stub matching, we can end up with a fairly larger bandwidth. In addition, we can place the first stub at an arbitrary location from the load. However, there is forbidden zone when it comes to double stub matching which means not double stub tuning technique cannot match every load.



Figure

CURRENT NETLIST ⓘ					
Component	Real Impedance		Imag Impedance		Physical Length
P1					
TL1	50	none ▾		33.59 m ▾	2.32 m ▾
TL2	50	none ▾		8.40 m ▾	0.58 m ▾
TL3	50	none ▾		10.86 m ▾	0.75 m ▾
TL4	50	none ▾		13.47 m ▾	0.93 m ▾
Z1	100	none ▾	-50	none ▾	
OC1					
OC2					
SC1					
TL5	50	none ▾		16.80 m ▾	1.16 m ▾

Figure

CALCULATION:

I have noticed that the dielectric constant designed in hfss was 4.4