## ECE 351 - 52

# SIGNALS AND SYSTEMS 1 LAB 11

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#### 1 Important Notes

It is important to note that plt.show() is needed at the end of the python code to properly show the plots of each function. It was not included in every section of code that plots a function(s) because it is assumed that its included at the end the code. It is also assumed that the following is included at the beginning of the program:

```
import numpy as np
import matplotlib.pyplot as plt
import scipy.signal as sig
import scipy.fftpack as fft
import math as m

The following package is not included with the Anaconda distribution
# and needed to be installed separately. The control package also has issues
# working on macs and a PC or a linux distribution is needed
import control as con
```

A note of caution for Mac users is that the control package has issues with the mac operating systems and needs to be run on a windows or Linux machine.

The web address to the GitHub where LaTeXcode is stored is here: https://github.com/Blairis123/ECE351\_Reports

The web address to the GitHub where the Python code is here: https://github.com/Blairis123/ECE351\_Code

#### 2 Part 1 Description

#### 2.1 Part 1

Part 1 is about analyzing a discrete system using python and a function developed by Christopher Felton and by hand. The function to be analyzed is assumed to be at rest and is:

$$y[k] = 2x[k] - 40x[k-1] + 10y[k-1] - 16y[-2]$$

The hand calculated derivation for the z-transform and solution can be found in the Results section with the subheading H(z) Hand Derivation and h[k] Hand Derivation. Then the scipy.signal.residuez() function was used to verify the partial fraction decomposition. The output can from this can be seen in the Results section under subheading Residuez Function Output. The code used to generate the output can be seen in Listing 1: Residuez Function Output Code. H(z) was also used in the zplane() function and the zeros and poles were plotted with respect to the unit circle. This can be seen in Figure 1: H(z) Pole Plot and the code used to do so can be seen in Listing 2.

The magnitude and phase responses of H(z) were then plotted using the scipy.signal.freqz() function. This function calculates the frequency response of a filter and returns an array of complex numbers. This array can then be used to calculate the phase and magnitude response of the filter. It is important that the input includes whole = True within the command. The frequency response plots can be found in Figure 2: H(z) Frequency Response. The code used to do this can be seen in Listing 3: H(z) Frequency Response Code.

#### 3 Results

#### 3.1 H(z) Hand Derivation

After transforming the function to the z-domain the result is:

$$Y(z) = 2X(z) - 40(z^{-1}X(z) - x[-1] + 10(z^{-1}Y(z) - y[-1]) - 16(Y(z)z^{-2} - z^{-1}y[-1] - y[-2])$$

Knowing that the system is initially at rest, the function can be further simplified to:

$$Y(z) = 2X(z) - 40z^{-1}X(z) + 10z^{-1}Y(z) - 16z^{-2}Y(z)$$
$$(40z^{-1} - 2)X(z) = (-16z^{-2} + 10z^{-1} - 1)Y(z)$$

$$\frac{Y(z)}{X(z)} = H(z) = \frac{40Z^{-1} - 2}{-16z^{-2} + 10z^{-1} - 1}$$
$$H(z) = \frac{2z^2 - 40z}{(z - 8)(z - 2)}$$

#### 3.2 h[k] Hand Derivation

$$H(z) = \frac{2z^2 - 40z}{(z - 8)(z - 2)}$$
$$\frac{H(z)}{z} = \frac{2z - 40}{(z - 8)(z - 2)}$$

Using partial fraction decomposition it can be found that:

$$\frac{H(z)}{z} = \frac{-4}{z - 8} + \frac{6}{z - 2}$$

Multiplying by z the result is:

$$H(z) = 6\frac{z}{z-8} - 4\frac{z}{z-2}$$

The inverse Z transform of H(z) is:

$$h[k] = (6(2)^k - 4(8)^k)u[k]$$

#### 3.3 Residuez Function Output

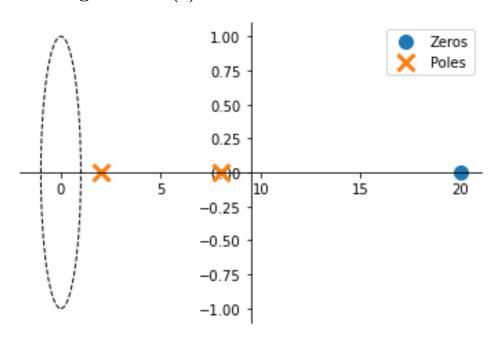
### 4 Questions

#### 4.1 Question 1

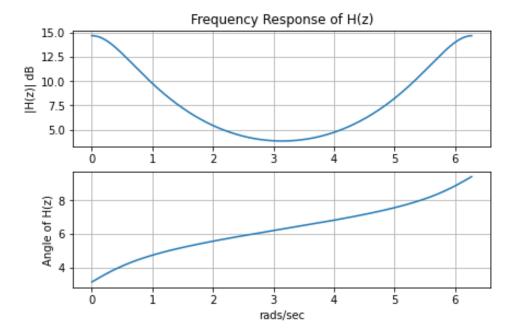
Looking at the plot generated by the zplane() function it can be determined that H(z) is not stable. This is because the poles of H(z) do not reside within the unit circle. The unit circle is the dashed oval on the right side of the plot in *Figure 1*.

## 5 Figures

## 5.1 Figure 1: H(z) Pole Plot



## 5.2 Figure 2: H(z) Frequency Response



## 6 Listings

#### **6.1** Listing 1:

```
# Verify partial fraction expantion
num = [2, -40]
den = [1, -10, 16]

[res, poles, coeff] = sig.residuez( num, den)

print(res)
print(poles)
print(coeff)
```

#### **6.2** Listing 2:

```
# Copyright (c) 2011 Christopher Felton
2 #
3 # This program is free software: you can redistribute it and/or
    modify
4 # it under the terms of the GNU Lesser General Public License as
     published by
5 # the Free Software Foundation, either version 3 of the License, or
6 # (at your option) any later version.
8 # This program is distributed in the hope that it will be useful,
9 # but WITHOUT ANY WARRANTY; without even the implied warranty of
10 # MERCHANTABILITY or FITNESS FOR A PARTICULAR PURPOSE. See the
11 # GNU Lesser General Public License for more details.
13 # You should have received a copy of the GNU Lesser General Public
14 # along with this program. If not, see <http://www.gnu.org/
     licenses/>.
15 #
16 # The following is derived from the slides presented by
17 # Alexander Kain for CS506/606 "Special Topics: Speech Signal
     Processing"
# CSLU / OHSU, Spring Term 2011.
19 #
20 #
22 # Modified by Drew Owens in Fall 2018 for use in the University of
     Idaho's
23 # Department of Electrical and Computer Engineering Signals and
     Systems I Lab
24 # (ECE 351)
25 #
```

```
26 # Modified by Morteza Soltani in Spring 2019 for use in the ECE 351
      of the U of
27 # I.
28 #
29 # Modified by Phillip Hagen in Fall 2019 for use in the University
     of Idaho's
30 # Department of Electrical and Computer Engineering Signals and
     Systems I Lab
  # (ECE 351)
32
  def zplane(b,a,filename=None):
33
      """Plot the complex z-plane given a transfer function.
34
      0.00
35
36
      import numpy as np
37
      import matplotlib.pyplot as plt
      from matplotlib import patches
40
      # get a figure/plot
41
      ax = plt.subplot(111)
42
43
      # create the unit circle
44
      uc = patches.Circle((0,0), radius=1, fill=False,
45
                            color='black', ls='dashed')
46
      ax.add_patch(uc)
47
48
      # The coefficients are less than 1, normalize the coeficients
49
      if np.max(b) > 1:
50
          kn = np.max(b)
51
           b = np.array(b)/float(kn)
52
      else:
53
          kn = 1
      if np.max(a) > 1:
56
57
          kd = np.max(a)
           a = np.array(a)/float(kd)
58
      else:
59
          kd = 1
60
61
      # Get the poles and zeros
      p = np.roots(a)
63
      z = np.roots(b)
64
      k = kn/float(kd)
65
      # Plot the zeros and set marker properties
67
      t1 = plt.plot(z.real, z.imag, 'o', ms=10, label='Zeros')
68
      plt.setp( t1, markersize=10.0, markeredgewidth=1.0)
69
70
      # Plot the poles and set marker properties
71
```

```
t2 = plt.plot(p.real, p.imag, 'x', ms=10, label='Poles')
      plt.setp( t2, markersize=12.0, markeredgewidth=3.0)
73
74
      ax.spines['left'].set_position('center')
75
      ax.spines['bottom'].set_position('center')
      ax.spines['right'].set_visible(False)
77
      ax.spines['top'].set_visible(False)
      plt.legend()
81
      # set the ticks
82
      \# r = 1.5; plt.axis('scaled'); plt.axis([-r, r, -r, r])
83
      # ticks = [-1, -.5, .5, 1]; plt.xticks(ticks); plt.yticks(ticks
84
85
      if filename is None:
86
          plt.show()
      else:
88
          plt.savefig(filename)
89
90
      return z, p, k
```

#### **6.3** Listing 3:

```
num = [2, -40]
_{2} den = [1, -10, 16]
      \# Plot magnitude and phase response of H(z)
5 w, H = sig.freqz(num, den, whole = True)
      #This is for the angle!
  angles = np.unwrap(np.angle(H))
      # Convert H into dB
_{11} H = 20 * np.log10(H)
      # Do some ploting!
plt.figure(figsize=(10,7))
plt.figure(constrained_layout=True)
16 plt.subplot(2,1,1)
17 plt.title("Frequency Response of H(z)")
18 plt.ylabel("|H(z)| dB")
19 plt.plot(w,H)
20 plt.grid()
22 plt.subplot(2,1,2)
plt.ylabel("Angle of H(z)")
24 plt.xlabel("rads/sec")
```

```
plt.plot(w,angles)
plt.grid()
```