

Dual and Quad Micropower Rail-to-Rail Input and Output Op Amps

FEATURES

- Rail-to-Rail Input and Output
- Single Supply Input Range: -0.4V to 44V
- Micropower: 50µA/Amplifier Max
- MSOP Package
- Specified on 3V, 5V and ±15V Supplies
- High Output Current: 20mA
- Output Drives 10,000pF with Output Compensation
- Reverse Battery Protection to 18V
- No Supply Sequencing Problems
- High Voltage Gain: 1500V/mV
- High CMRR: 98dBNo Phase Reversal
- Gain Bandwidth Product: 200kHz

APPLICATIONS

- Battery- or Solar-Powered Systems
 Portable Instrumentation
 Sensor Conditioning
- Supply Current Sensing
- Battery Monitoring
- Micropower Active Filters
- 4mA to 20mA Transmitters

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DESCRIPTION

The dual LT®1490 and quad LT1491 op amps operate on all single and split supplies with a total voltage of 2V to 44V drawing only $40\mu A$ of quiescent current per amplifier. These amplifiers are reverse supply protected; they draw no current for reverse supply up to 18V. The input range of the LT1490/LT1491 includes both supplies and the output swings to both supplies. Unlike most micropower op amps, the LT1490/LT1491 can drive heavy loads; their rail-to-rail outputs drive 20mA. The LT1490/LT1491 are unity-gain stable and drive all capacitive loads up to 10,000pF when optional 0.22µF and 150 Ω compensation is used.

The LT1490/LT1491 have a unique input stage that operates and remains high impedance when above the positive supply. The inputs take 44V both differential and common mode even when operating on a 3V supply. Built-in resistors protect the inputs for faults below the negative supply up to 22V. There is no phase reversal of the output for inputs 22V below V^- or 44V above V^- , independent of V^+ .

The LT1490 dual op amp is available in the 8-pin MSOP, SO and PDIP packages. The quad LT1491 is available in the 14-pin SO and PDIP packages.

For new designs it is recommended that the LT1490A/LT1491A be used instead of the LT1490/LT1491. See the LT1490A/LT1491A data sheet for details.

TYPICAL APPLICATION

Battery Monitor 01 CHARGER 2N3904 A 1/4 LT1491 I_{BATT} C 1/4 LT1491 LOGIC LOGIC HIGH (5V) = CHARGING LOGIC LOW (0V) = DISCHARGING B 1/4 LT1491 LOAD 1/4 LT1491 V_{OUT} V_{BATT} = 12V $\frac{v_{0UT}}{(R_S)(R_G/R_A)(GAIN)}$ S1 = OPEN, GAIN = 1 S1 = CLOSED, GAIN = 10 $R_A = R_B$ $V_S = 5V, 0V$

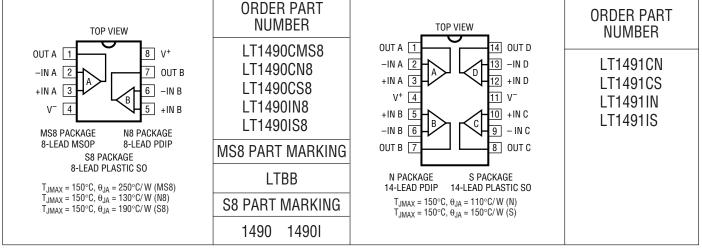


ABSOLUTE MAXIMUM RATINGS (Note 1)

| Total Supply Voltage (V + to V -) | 44V |
|--|---------------|
| Input Differential Voltage | 44V |
| Input Current | ±25mA |
| Output Short-Circuit Duration (Note 2) | Continuous |
| Operating Temperature Range | -40°C to 85°C |

| Junction Temperature 150°C |
|--|
| Specified Temperature Range (Note 3)40°C to 85°C |
| Storage Temperature Range65°C to 150°C |
| Lead Temperature (Soldering, 10 sec)300°C |

PACKAGE/ORDER INFORMATION (For new designs, use the LT1490A or LT1491A)



Consult factory for Military grade parts.

| SYMBOL | PARAMETER | CONDITIONS | | MIN | TYP | MAX | UNITS |
|--------------------|----------------------------|--|---|-----|---------------|----------------------|-------------------|
| V _{OS} | Input Offset Voltage | LT1490 N Package $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$ | • | | 220 | 800 1000 1100 | μV μV μV |
| | | LT1490 S Package $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$ | • | | 220 | 950 1200 1300 | μV μV μV |
| | | LT1491 N Package $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$ | • | | 300 | 1100 1350 1450 | μV μV μV |
| | | LT1490CMS8 Package, LT1491 S Package $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$ | • | | 350 | 1450 1650 1750 | μV μV μV |
| V _{OS} TC | Input Offset Voltage Drift | $0^{\circ}C \le T_A \le 70^{\circ}C$ (Note 7) -40°C $\le T_A \le 85^{\circ}C$ (Note 7) | • | | 2 2 | 4 4 | μV/°C μV/°C |
| I _{OS} | Input Offset Current | V _{CM} = 44V (Note 4) | • | | 0.2 | 0.8 0.8 | nA μA |
| I _B | Input Bias Current | V _{CM} = 44V (Note 4) V _S = 0V | • | | 4 4 0.1 | 8 10 | nA μA nA |
| | Input Noise Voltage | 0.1Hz to 10Hz | | | 1 | | μV _{P-P} |

ELECTRICAL CHARACTERISTICS The \bullet denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $V_{CM} = V_{OUT} = half supply, unless otherwise noted. (Note 3)$

| SYMBOL | PARAMETER | CONDITIONS | | MIN | TYP | MAX | UNITS |
|------------------|---------------------------------------|--|---|-------------------------|------------------|------------------|----------------------|
| e _n | Input Noise Voltage Density | f = 1kHz | | | 50 | | nV/√Hz |
| i _n | Input Noise Current Density | f = 1kHz | | | 0.03 | | pA/√Hz |
| R _{IN} | Input Resistance | Differential Common Mode, V _{CM} = 0V to 44V | | 6 4 | 17 11 | | MΩ MΩ |
| C _{IN} | Input Capacitance | | | | 4.6 | | pF |
| | Input Voltage Range | | • | 0 | | 44 | V |
| CMRR | Common Mode Rejection Ratio (Note 4) | $V_{CM} = 0V \text{ to } V_{CC} - 1V$ $V_{CM} = 0V \text{ to } 44V$ | • | 84 80 | 98 98 | | dB dB |
| A _{VOL} | Large-Signal Voltage Gain | $V_S = 3V$, $V_O = 500$ mV to 2.5V, $R_L = 10$ k $0^{\circ}\text{C} \le T_A \le 70^{\circ}\text{C}$ $-40^{\circ}\text{C} \le T_A \le 85^{\circ}\text{C}$ | • | 200 133 100 | 1500 | | V/mV V/mV V/mV |
| | | $V_S = 5V$, $V_0 = 500$ mV to 4.5V, $R_L = 10$ k $0^{\circ}\text{C} \le T_A \le 70^{\circ}\text{C}$ $-40^{\circ}\text{C} \le T_A \le 85^{\circ}\text{C}$ | • | 400 250 200 | 1500 | | V/mV V/mV V/mV |
| V_{OL} | Output Voltage Swing Low | $V_S = 3V$, No Load $V_S = 3V$, $I_{SINK} = 5mA$ | • | | 22 250 | 50 450 | mV mV |
| | | $V_S = 5V$, No Load $V_S = 5V$, $I_{SINK} = 5mA$ $V_S = 5V$, $I_{SINK} = 10mA$ | • | | 22 250 330 | 50 500 500 | mV mV mV |
| V _{OH} | Output Voltage Swing High | $V_S = 3V$, No Load $V_S = 3V$, $I_{SOURCE} = 5mA$ | • | 2.95 2.55 | 2.978 2.6 | | V |
| | | V _S = 5V, No Load V _S = 5V, I _{SOURCE} = 10mA | • | 4.95 4.30 | 4.978 4.6 | | V |
| I _{SC} | Short-Circuit Current (Note 2) | V _S = 3V, Short to GND V _S = 3V, Short to V _{CC} | | 10 10 | 15 30 | | mA mA |
| | | V _S = 5V, Short to GND V _S = 5V, Short to V _{CC} | | 15 15 | 25 30 | | mA mA |
| PSRR | Power Supply Rejection Ratio | $V_S = 2.5V$ to 12.5V, $V_{CM} = V_0 = 1V$ | • | 84 | 98 | | dB |
| | Minimum Operating Supply Voltage | | • | | 2 | 2.5 | V |
| | Reverse Supply Voltage | $I_S = -100\mu A$ per Amplifier | • | 18 | 27 | | V |
| Is | Supply Current per Amplifier (Note 5) | | • | | 40 | 50 55 | μA μA |
| GBW | Gain Bandwidth Product (Note 4) | $ f = 1 \text{kHz} $ $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C} $ $-40^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85^{\circ}\text{C} $ | • | 110 100 90 | 180 | | kHz kHz kHz |
| SR | Slew Rate (Note 6) | $ \begin{aligned} A_V &= -1, \ R_L = \infty \\ 0^{\circ} C &\leq T_A \leq 70^{\circ} C \\ -40^{\circ} C &\leq T_A \leq 85^{\circ} C \end{aligned} $ | • | 0.035 0.031 0.030 | 0.06 | | V/μs V/μs V/μs |



ELECTRICAL CHARACTERISTICS The ullet denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25 \,^{\circ}\text{C}$. $V_S = \pm 15 \,\text{V}$, $V_{CM} = 0 \,\text{V}$, $V_{OUT} = 0 \,\text{V}$, unless otherwise noted. (Note 3)

| SYMBOL | PARAMETER | CONDITIONS | | MIN | TYP | MAX | UNITS |
|---------------------|-----------------------------------|---|---|------------|--------------|----------|-------------------|
| $\overline{V_{OS}}$ | Input Offset Voltage | LT1490 N, S Package | | | 250 | 1200 | μV |
| | | $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C}$ | • | | | 1400 | μV |
| | | $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$ | • | | | 1500 | μV |
| | | LT1491 N Package | | | 350 | 1250 | μV |
| | | $0^{\circ}\text{C} \leq \text{T}_{A} \leq 70^{\circ}\text{C}$ | • | | | 1500 | μV |
| | | $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$ | • | | | 1600 | μV |
| | | LT1490CMS8 Package, LT1491 S Package | | | 400 | 1600 | μV |
| | | $0^{\circ}C \leq T_A \leq 70^{\circ}C$ | • | | | 1850 | μV |
| | | $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$ | • | | | 1950 | μV |
| V_{OS} TC | Input Offset Voltage Drift | $0^{\circ}C \le T_A \le 70^{\circ}C \text{ (Note 7)}$ | • | | 3 | 6 | μV/°C |
| | | -40 °C \leq T _A \leq 85°C (Note 7) | • | | 3 | 6 | μV/°C |
| l _{os} | Input Offset Current | | • | | 0.2 | 0.8 | nA |
| I _B | Input Bias Current | | • | | 4 | 8 | nA |
| | Input Noise Voltage | 0.1Hz to 10Hz | | | 1 | | μV _{P-P} |
| en | Input Noise Voltage Density | f = 1kHz | | | 50 | | nV/√Hz |
| i _n | Input Noise Current Density | f = 1kHz | | | 0.03 | | pA/√Hz |
| R_{IN} | Input Resistance | Differential | | 6 | 17 | | MΩ |
| | | Common Mode, $V_{CM} = -15V$ to 14V | | | 15000 | | MΩ |
| C _{IN} | Input Capacitance | | | | 4.6 | | pF |
| | Input Voltage Range | | • | -15 | | 29 | V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = -15V \text{ to } 29V$ | • | 80 | 98 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_0 = \pm 14V, R_L = 10k$ | | 100 | 250 | | V/mV |
| | | $0^{\circ}\text{C} \leq \text{T}_{A} \leq 70^{\circ}\text{C}$ | • | 75 | | | V/mV |
| | | $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$ | • | 50 | | | V/mV |
| V_0 | Output Voltage Swing | No Load | • | ±14.9 | ± 14.978 | | V |
| | | $I_{OUT} = \pm 5 \text{mA}$ | • | ±14.5 | ±14.750 | | V |
| | | $I_{OUT} = \pm 10 \text{mA}$ | | ±14.5 | ±14.670 | | V |
| I_{SC} | Short-Circuit Current (Note 2) | Short to GND | | ±20 | ±25 | | mA |
| | | $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C}$ | • | ±15 | | | mA |
| DODD | Develop Overally Delegation Detic | $-40^{\circ}\text{C} \le T_{\text{A}} \le 85^{\circ}\text{C}$ | • | ±10 | 00 | | mA |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 1.25 V \text{ to } \pm 22 V$ | • | 88 | 98 | 70 | dB |
| I _S | Supply Current per Amplifier | | | | 50 | 70 85 | μΑ |
| GBW | Coin Dondwidth Droduct | f = 1kHz | _ | 105 | 200 | 00 | μA kHz |
| GBW | Gain Bandwidth Product | $T = TKHZ$ $0^{\circ}C \le T_{A} \le 70^{\circ}C$ | | 125 110 | 200 | | kHz kHz |
| | | $-40^{\circ}\text{C} \le T_{A} \le 70^{\circ}\text{C}$ $-40^{\circ}\text{C} \le T_{A} \le 85^{\circ}\text{C}$ | | 100 | | | kHz |
| SR | Clay Data | | _ | | 0.07 | | |
| δĦ | Slew Rate | $A_V = -1$, $R_L = \infty$, $V_0 = \pm 10V$, Measure at $V_0 = \pm 5V$ | | 0.0375 | 0.07 | | V/µs |
| | | $0^{\circ}C \le T_A \le 70^{\circ}C$ | | 0.0330 | | | V/µs |
| | | $-40^{\circ}\text{C} \le \text{T}_{A} \le 70^{\circ}\text{C}$ | | 0.0300 | | | V/μs |
| - | | 10 0 = 1A = 00 0 | | 0.0000 | | | Ι ν/μο |

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

Note 2: A heat sink may be required to keep the junction temperature below absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted.

Note 3: The LT1490C/LT1491C are guaranteed to meet specified performance from 0°C to 70°C. The LT1490C/LT1491C are designed, characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures.

The LT1490I/LT1491I are guaranteed to meet specified performance from -40°C to $85^{\circ}\text{C}.$

Note 4: $V_S = 5V$ limits are guaranteed by correlation to $V_S = 3V$ and $V_S = \pm 15V$ tests.

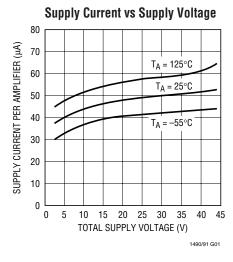
Note 5: $V_S = 3V$ limits are guaranteed by correlation to $V_S = 5V$ and $V_S = \pm 15V$ tests.

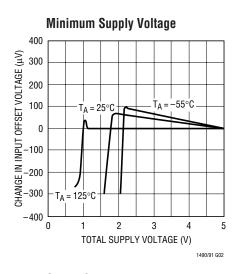
Note 6: Guaranteed by correlation to slew rate at $V_S=\pm 15V$ and GBW at $V_S=3V$ and $V_S=\pm 15V$ tests.

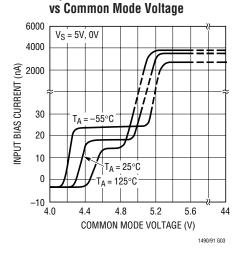
Note 7: This parameter is not 100% tested.



TYPICAL PERFORMANCE CHARACTERISTICS

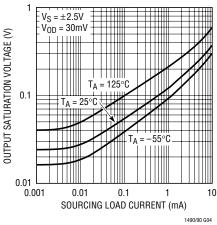


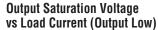


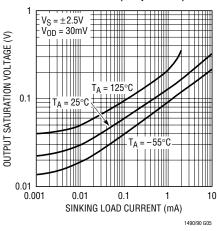


Input Bias Current

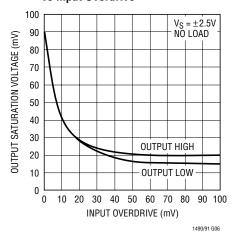




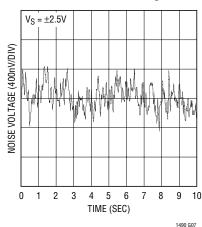




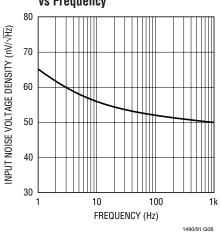
Output Saturation Voltage vs Input Overdrive



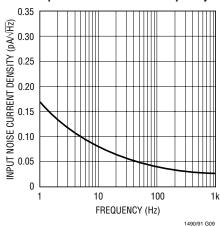
0.1Hz to 10Hz Noise Voltage





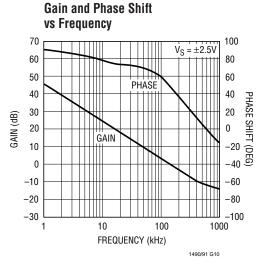


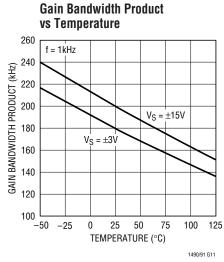
Input Noise Current vs Frequency

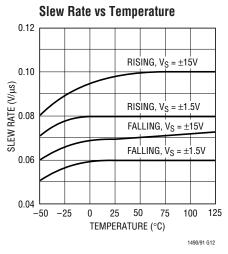


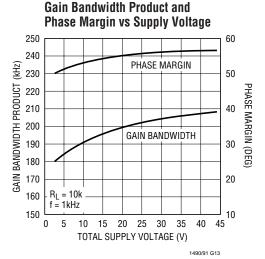


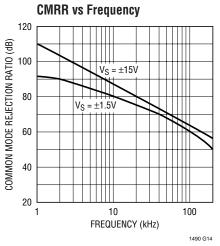
TYPICAL PERFORMANCE CHARACTERISTICS

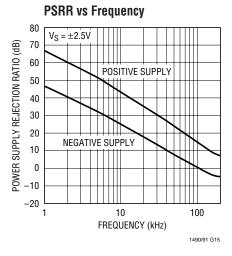


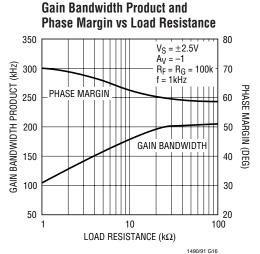


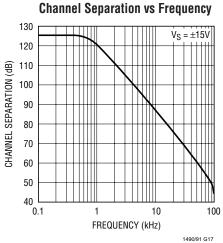


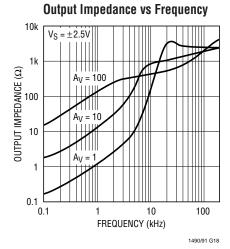






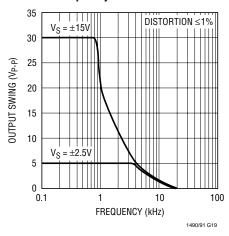




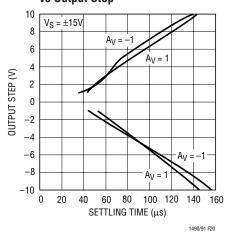


TYPICAL PERFORMANCE CHARACTERISTICS

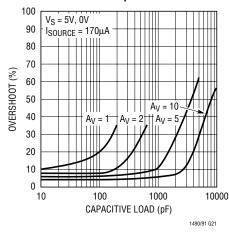
Undistorted Output Swing vs Frequency



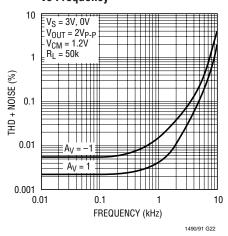
Settling Time to 0.1% vs Output Step



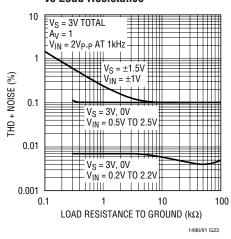
Capacitive Load Handling, Overshoot vs Capacitive Load



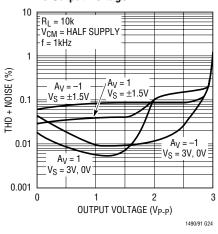
Total Harmonic Distortion + Noise vs Frequency



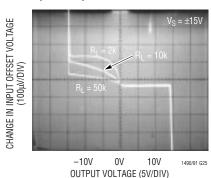
Total Harmonic Distortion + Noise vs Load Resistance



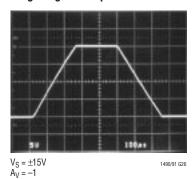
Total Harmonic Distortion + Noise vs Output Voltage



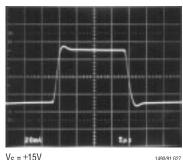
Open-Loop Gain



Large-Signal Response



Small-Signal Response



 $V_S = \pm 15V$ 1490/91 G27 $A_V = 1$



APPLICATIONS INFORMATION

Supply Voltage

The positive supply pin of the LT1490/LT1491 should be bypassed with a small capacitor (about $0.01\mu F$) within an inch of the pin. When driving heavy loads an additional $4.7\mu F$ electrolytic capacitor should be used. When using split supplies, the same is true for the negative supply pin.

The LT1490/LT1491 are protected against reverse battery voltages up to 18V. In the event a reverse battery condition occurs, the supply current is less than 1nA.

The LT1490/LT1491 can be shut down by removing V⁺. In this condition the input bias current is less than 0.1nA, even if the inputs are 44V above the negative supply.

When operating the LT1490/LT1491 on total supplies of 30V or more, the supply must not be brought up faster than 1 μ s. This is especially true if low ESR bypass capacitors are used. A series RLC circuit is formed from the supply lead inductance and the bypass capacitor. 5 Ω of resistance in the supply or the bypass capacitor will dampen the tuned circuit enough to limit the rise time.

Inputs

The LT1490/LT1491 have two input stages, NPN and PNP (see the Simplified Schematic), resulting in three distinct operating regions as shown in the Input Bias Current vs Common Mode typical performance curve.

For input voltages about 0.8V or more below V+, the PNP input stage is active and the input bias current is typically -4nA. When the input voltage is about 0.5V or less from V+, the NPN input stage is operating and the input bias current is typically 18nA. Increases in temperature will cause the voltage at which operation switches from the PNP stage to the NPN stage to move towards V+. The input offset voltage of the NPN stage is untrimmed and is typically $600\mu V$.

A Schottky diode in the collector of each NPN transistor of the NPN input stage allows the LT1490/LT1491 to operate with either or both of its inputs above V⁺. At about 0.3V above V⁺ the NPN input transistor is fully saturated and the input bias current is typically $4\mu A$ at room temperature. The input offset voltage is typically $700\mu V$ when operating above V⁺. The LT1490/LT1491 will operate with its inputs 44V above V⁻ regardless of V⁺.

The inputs are protected against excursions as much as 22V below V^- by an internal 1k resistor in series with each input and a diode from the input to the negative supply. There is no output phase reversal for inputs up to 22V below V^- . There are no clamping diodes between the inputs and the maximum differential input voltage is 44V.

Output

The output voltage swing of the LT1490/LT1491 is affected by input overdrive as shown in the typical performance curves. When monitoring voltages within 100mV of either rail, gain should be taken to keep the output from clipping.

The output of the LT1490/LT1491 can be pulled up to 18V beyond V^+ with less than 1nA of leakage current, provided that V^+ is less than 0.5V.

The normally reverse-biased substrate diode from the output to V^- will cause unlimited currents to flow when the output is forced below V^- . If the current is transient and limited to 100mA, no damage will occur.

The LT1490/LT1491 is internally compensated to drive at least 200pF of capacitance under any output loading conditions. A 0.22µF capacitor in series with a 150 Ω resistor between the output and ground will compensate these amplifiers for larger capacitive loads, up to 10,000pF, at all output currents.

Distortion

There are two main contributors of distortion in op amps: output crossover distortion as the output transitions from sourcing to sinking current and distortion caused by nonlinear common mode rejection. Of course, if the op amp is operating inverting there is no common mode induced distortion. When the LT1490 switches between input stages there is significant nonlinearity in the CMRR. Lower load resistance increases the output crossover distortion, but has no effect on the input stage transition distortion. For lowest distortion the LT1490/LT1491 should be operated single supply, with the output always sourcing current and with the input voltage swing between ground and (V $^+$ – 0.8V). See the Typical Performance Characteristics curves.



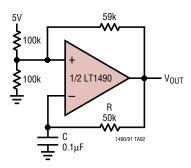
APPLICATIONS INFORMATION

Gain

The open-loop gain is almost independent of load when the output is sourcing current. This optimizes performance in single supply applications where the load is returned to ground. The typical performance photo of Open-Loop Gain for various loads shows the details.

TYPICAL APPLICATIONS

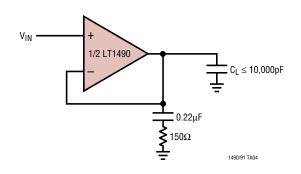
Square Wave Oscillator



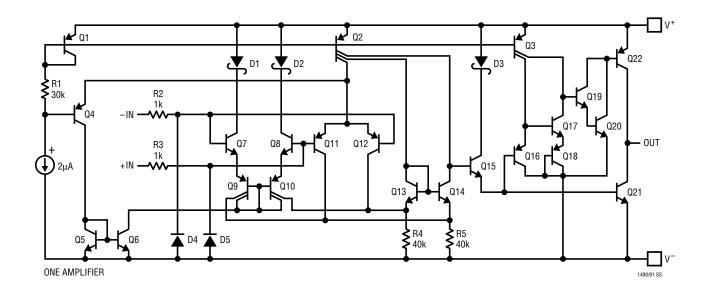
 $f = \frac{1}{2RC}$ $V_{OUT} = 5V_{P-P} \text{ WITH 5V SUPPLY }$ $I_S = 200 \mu \text{A}$

AT $V_S = 5V$, R = 50k, C = 1nFOUTPUT IS 5kHz SLEW LIMITED TRIANGLE WAVE

Optional Output Compensation for Capacitive Loads Greater Than 200pF



SIMPLIFIED SCHEMATIC

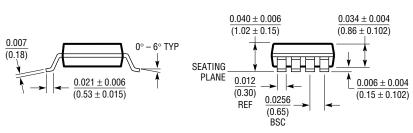


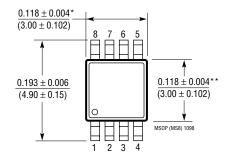


PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

MS Package 8-Lead Plastic MSOP

(LTC DWG # 05-08-1660)

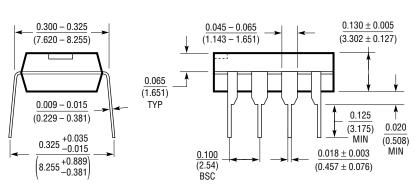


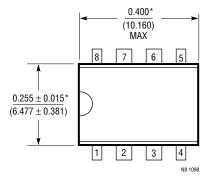


- * DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE
- ** DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
 INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

N8 Package 8-Lead PDIP (Narrow 0.300)

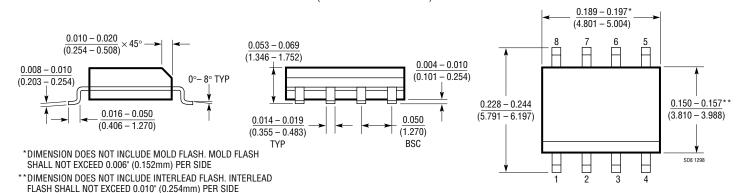
(LTC DWG # 05-08-1510)





S8 Package 8-Lead Plastic Small Outline (Narrow 0.150)

(LTC DWG # 05-08-1610)

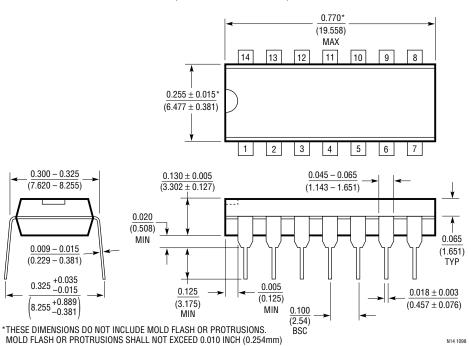


^{*}THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

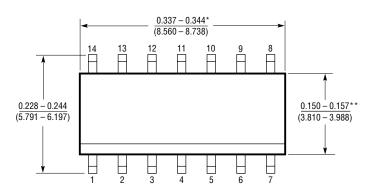
N Package 14-Lead PDIP (Narrow 0.300)

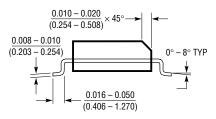
(LTC DWG # 05-08-1510)

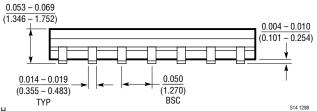


S Package 14-Lead Plastic Small Outline (Narrow 0.150)

(LTC DWG # 05-08-1610)







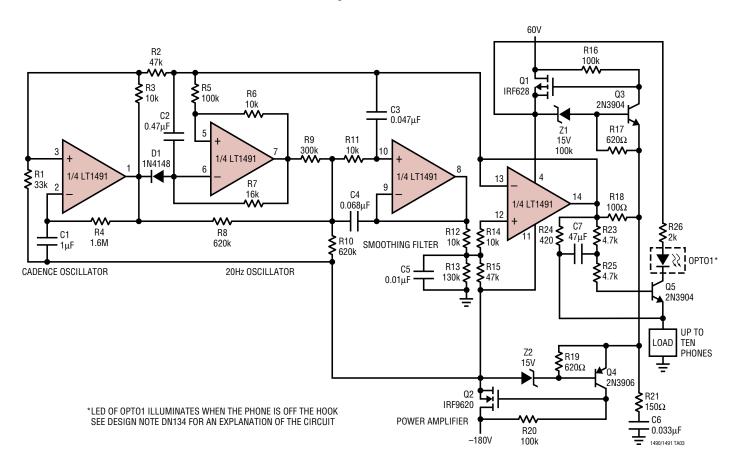
^{*}DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

^{**}DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE



TYPICAL APPLICATION

Ring-Tone Generator



RELATED PARTS

| PART NUMBER | DESCRIPTION | COMMENTS |
|-----------------|---|---|
| LT1078/LT1079 | Dual/Quad 55μA Max, Single Supply, Precision Op Amps | Input/Output Common Mode Includes Ground, 70μV V _{OS(MAX)} and 2.5μV/°C Drift (Max), 200kHz GBW, 0.07V/μs Slew Rate |
| LTC1152 | Rail-to-Rail Input, Rail-to-Rail Output, Zero-Drift Amplifier | High DC Accuracy, 10μV V _{OS(MAX)} , 100nV/°C, 1MHz GBW, 1V/μs Slew Rate, Supply Current 2.2mA (Max), Single Supply, Can Be Configured for C-Load [™] Operation |
| LT1178/LT1179 | Dual/Quad 17μA Max, Single Supply, Precison Op Amps | Input/Output Common Mode Includes Ground, $70\mu V$ $V_{OS(MAX)}$ and $4\mu V/^{\circ}C$ Drift (Max), $85kHz$ GBW, $0.04V/\mu s$ Slew Rate |
| LT1366/LT1367 | Dual/Quad Precision, Rail-to-Rail Input and Output Op Amps | 475μV V _{OS(MAX)} , 500V/mV A _{VOL(MIN)} , 400kHz GBW |
| LT1490A/LT1491A | Dual/Quad Micropower Rail-to-Rail Precision Op Amps | 500μV V _{OS(MAX)} , Improved Version of the LT1490/LT1491 |

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