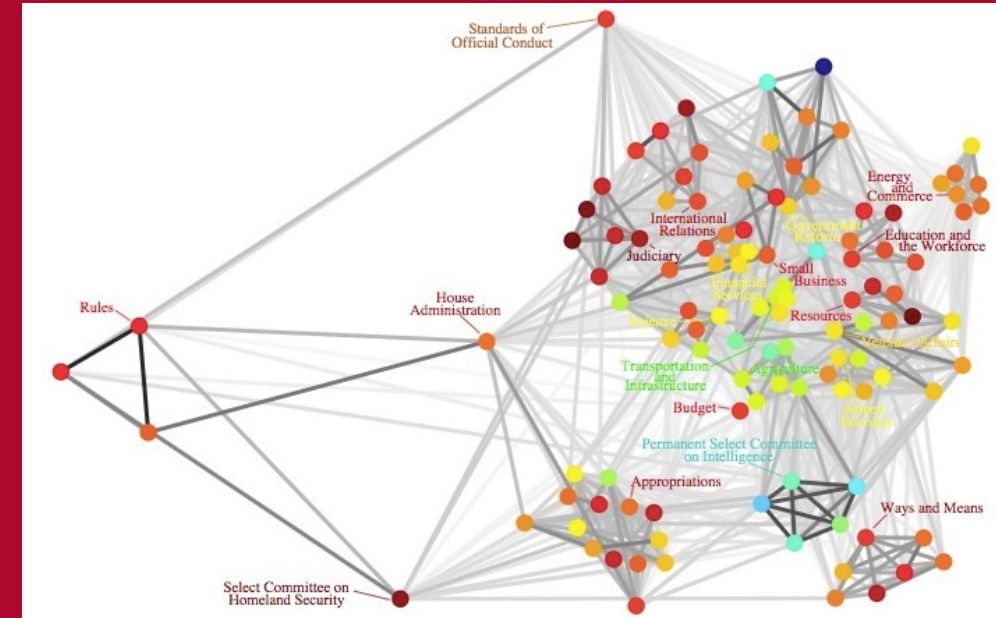


# Automatic Control Theory

## Chapter 3



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# The performance of feedback control systems

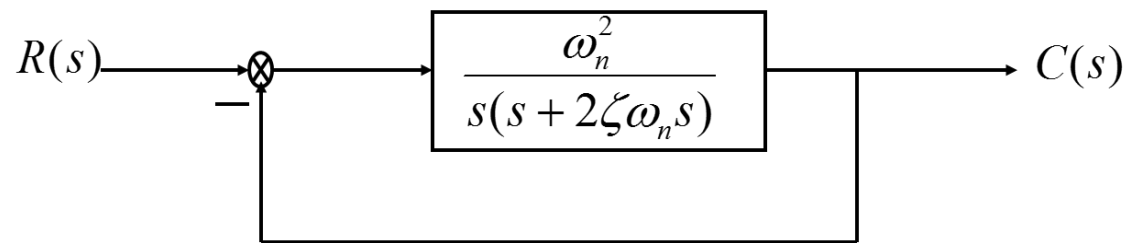
## Main contents

1. Typical test signals for the time response of control systems.
2. The unit-step response and time-domain specifications.
3. Time response of first-order and second-order systems.
4. Improvement performance of second systems.
5. Condition for a feedback system to be stable
6. Routh-Hurwitz criterion
7. The steady-state error of feedback control system.

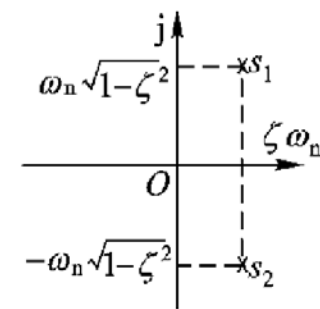


# The performance of feedback control systems

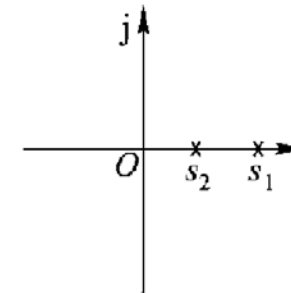
## Review



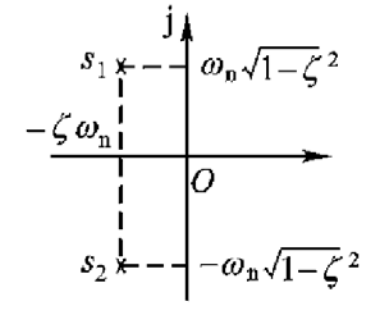
$$T(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$



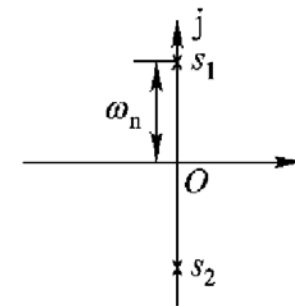
(a)  $-1 < \zeta < 0$



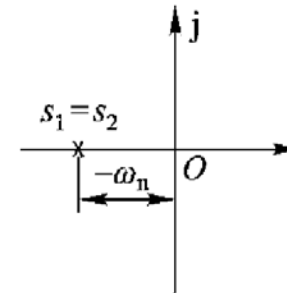
(b)  $\zeta < -1$



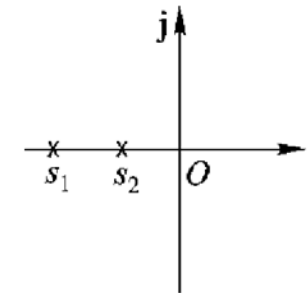
(c)  $0 < \zeta < 1$



(d)  $\zeta = 0$



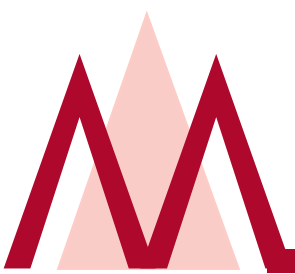
(e)  $\zeta = 1$



(f)  $\zeta > 1$

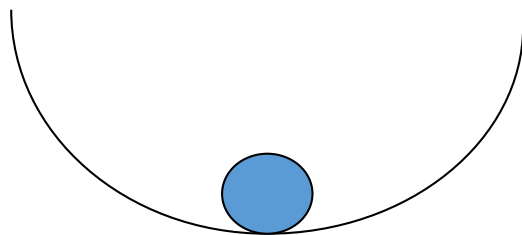
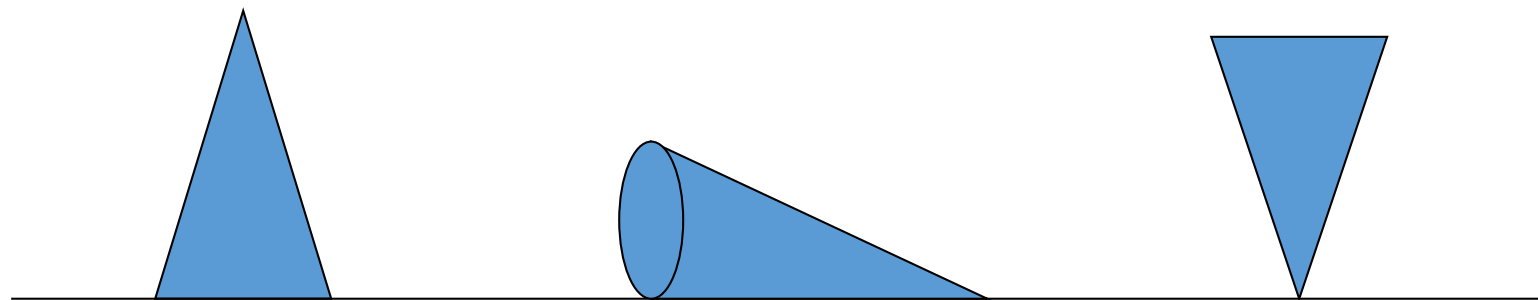
## what is next

Condition for a feedback system to be stable (Routh-Hurwitz criterion)

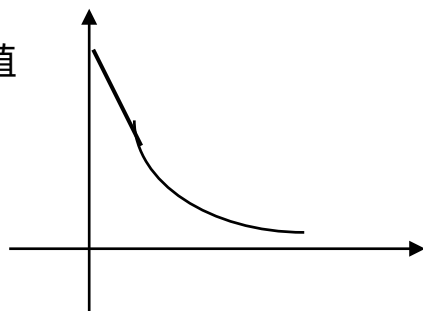


# Condition for a feedback system to be stable

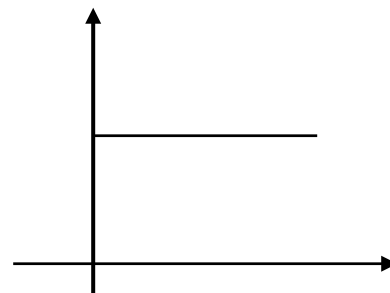
## Example



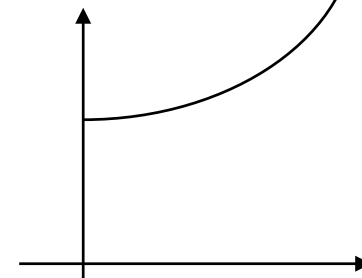
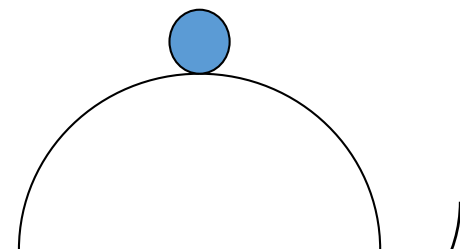
纵坐标为偏差值



stable



neutral



unstable

# Condition for a feedback system to be stable

Consider the transfer function of a closed-loop system as

$$T(s) = \frac{N(s)}{D(s)} = \frac{N(s)}{\prod_{i=1}^n (s - s_i)}$$

⚠ 写法  
不管分子  
分母一定是因式相乘

稳定性和分子无关 和极点分别有关

Where  $s_i, i = 1, 2 \cdots n$ , are roots of the characteristic equation  $D(s)$ . Assume that all roots are simple, then we have

$$C(s) = T(s)R(s) = \frac{N(s)}{\prod_{i=1}^n (s - s_i)} R(s) = \sum_{j=1}^l \frac{B_j}{s - s_{rj}} + \sum_{i=1}^n \frac{A_i}{s - s_i}$$

由输入r决定      由传递函数决定  
分析稳定性时主要分析



# Condition for a feedback system to be stable

$$c(t) = \sum_{j=1}^l B_j e^{s_{rj}t} + \sum_{i=1}^n A_i e^{s_i t}$$

Steady-state response

Transient response

$$\lim_{t \rightarrow \infty} \sum_{i=1}^n A_i e^{s_i t} = 0$$

$$\sigma_i < 0$$

$$\lim_{t \rightarrow \infty} A_i e^{s_i t} = 0$$

When  $s_i$  is real number, letting  $s_i = \sigma_i$  we have

$$\sigma_i = 0$$

$$\lim_{t \rightarrow \infty} A_i e^{s_i t} = A_i$$

$$\sigma_i > 0$$

$$\lim_{t \rightarrow \infty} A_i e^{s_i t} = \infty$$



# Condition for a feedback system to be stable

When  $s_i$  is complex number, letting  $s_i = \sigma_i \pm j\omega_i$ , we have

$$A_i e^{(\sigma_i + j\omega_i)t} + A_{i+1} e^{(\sigma_i - j\omega_i)t} = A e^{\sigma_i t} \cos(\omega_i t + \psi)$$

$$\sigma_i < 0 \quad \lim_{t \rightarrow \infty} A e^{\sigma_i t} \cos(\omega_i t + \psi) = 0$$

$$\sigma_i = 0 \quad \lim_{t \rightarrow \infty} A e^{\sigma_i t} \cos(\omega_i t + \psi) = A \cos(\omega_i t + \psi)$$

$$\sigma_i > 0 \quad \lim_{t \rightarrow \infty} A e^{\sigma_i t} \cos(\omega_i t + \psi) = \infty$$

A **necessary and sufficient** condition for a feedback system to be stable is that all the **poles** of the system transfer function have **negative real parts**.



# Routh-hurwitz criterion

## The characteristic equation of a feedback system

$$a_n s^n + a_{n-1} s^{n-1} + \cdots a_1 s + a_0 = 0$$

Assume:  $a_n > 0$ , when  $n=5$ , Routh table is

$s^5$	$a_5$	$a_3$	$a_1$
$s^4$	$a_4$	$a_2$	$a_0$
$s^3$	$\frac{a_3 a_4 - a_2 a_5}{a_4} = b_1$	$\frac{a_1 a_4 - a_0 a_5}{a_4} = b_2$	0
$s^2$	$\frac{b_1 a_2 - b_2 a_4}{b_1} = c_1$	$a_0$	
$s^1$	$\frac{c_1 b_2 - b_1 a_0}{c_1} = d_1$		
$s^0$	$a_0$		





# Routh-hurwitz criterion

## Example 1 :

**Determine the stability of the closed-loop system that has the following characteristic equations.**

$$(1) \quad s^4 + 2s^3 + 3s^2 + 4s + 5 = 0$$

$$(2) \quad s^3 - 3s + 4 = 0$$

$$(3) \quad s^5 + 2s^4 + 2s^3 + 4s^2 + 11s + 10 = 0$$



# Routh-hurwitz criterion

## Answer

$$(1) \quad s^4 + 2s^3 + 3s^2 + 4s + 5 = 0$$

$$s^4 \quad 1 \quad 3 \quad 5$$

$$s^3 \quad 2 \quad 4$$

$$s^2 \quad 1 \quad 5$$

$$s^1 \quad -6$$

$$s^0 \quad 5$$

$$0.2878 + 1.4161i$$

$$0.2878 - 1.4161i$$

$$-1.2878 + 0.8579i$$

$$-1.2878 - 0.8579i$$



# Routh-hurwitz criterion

Answer

$$(2) \quad s^3 - 3s + 4 = 0$$

$$\begin{array}{ccc} s^3 & 1 & -3 \\ s^2 & \varepsilon & 4 \\ s^1 & \frac{-3\varepsilon - 4}{\varepsilon} & \\ s^0 & 4 & \end{array}$$

$$(s^3 - 3s + 4)(s + 1) = s^4 + s^3 - 3s^2 + s + 4 = 0$$

$$\begin{array}{ccc} s^4 & 1 & -3 & 4 \\ s^3 & 1 & 1 & \\ s^2 & -4 & 4 & \\ s^1 & 2 & & \\ s^0 & 4 & & \end{array}$$



# Routh-hurwitz criterion

$$(3) \quad s^5 + 2s^4 + 2s^3 + 4s^2 + 11s + 10 = 0$$

$$\begin{array}{l} s^5 \quad 1 \quad 2 \quad 11 \\ s^4 \quad 2 \quad 4 \quad 10 \\ s^3 \quad 0 \quad 6 \rightarrow \varepsilon \quad 6 \\ s^2 \quad \frac{4\varepsilon - 12}{\varepsilon} \quad 10 \\ s^1 \quad \frac{24}{\varepsilon} - \frac{72}{\varepsilon^2} - 10 \\ s^0 \quad 10 \end{array}$$

```
a=[1 2 2 4 11 10];  
>> p=roots(a)  
  
p =  
  
    0.8950 + 1.4561i  
    0.8950 - 1.4561i  
   -1.2407 + 1.0375i  
   -1.2407 - 1.0375i  
   -1.3087
```



# Routh-hurwitz criterion

应用 $\varepsilon$  法要注意，否则可能得出错误的结论

例  $s^6 + s^5 + 3s^4 + 3s^3 + 3s^2 + 2s + 1 = 0$

$s^6$	1	3	3	1
$s^5$	1	3	2	
$s^4$	$\varepsilon$	1	1	
$s^3$	$3 - \frac{1}{\varepsilon}$	$2 - \frac{1}{\varepsilon}$		
$s^2$	$v$	1		
$s^1$	$u$			
$s^0$	1			

$$v = 1 - \frac{2\varepsilon - 1}{3 - 1/\varepsilon}$$

$$u = 2 - \frac{1}{\varepsilon} - \frac{3 - 1/\varepsilon}{1 - \frac{2\varepsilon - 1}{3 - 1/\varepsilon}}$$

原因是该方程有虚轴上的根。

# Routh-hurwitz criterion

$$(4) \quad s^6 + s^5 - 2s^4 - 3s^3 - 7s^2 - 4s - 4 = 0$$

$$\begin{array}{l} s^6 \quad 1 \quad -2 \quad -7 \quad -4 \\ s^5 \quad 1 \quad -3 \quad -4 \\ s^4 \quad 1 \quad -3 \quad -4 \\ s^3 \quad 0 \quad 0 \quad 0 \end{array}$$

$$F(s) = s^4 - 3s^2 - 4 = 0$$

$$\frac{dF(s)}{ds} = 4s^3 - 6s = 0$$

辅助方程  
求导  
得到并替代

$$\begin{array}{l} s^6 \quad 1 \quad -2 \quad -7 \quad -4 \\ s^5 \quad 1 \quad -3 \quad -4 \\ s^4 \quad 1 \quad -3 \quad -4 \\ s^3 \quad 4 \quad -6 \quad 0 \\ s^2 \quad -1.5 \quad -4 \\ s^1 \quad -16.7 \quad 0 \\ s^0 \quad -4 \end{array}$$

但不能直接反应稳定性的几个极点  
只能反应实根  
虚根被解决掉了 要从辅助方程里求

$$\pm 2, \pm j, (-1 \pm j\sqrt{3})/2$$



# The performance of feedback control systems

## 核心

- **Condition for a feedback system to be stable**
  - **Routh-Hurwitz criterion**
- $$c(t) = \sum_{j=1}^l B_j e^{s_{rj}t} + \sum_{i=1}^n A_i e^{s_i t}$$

## 续

- **The steady-state error of feedback control system**



## Routh-hurwitz criterion

Example 2 : Consider that the characteristic equation of a closed-loop control system is

$$s^3 + 3Ks^2 + (K + 2)s + 4 = 0$$

Find the rang of  $K$  so that the system is stable.

Answer:  $K > 0.528$

Example 3: Determine the following characteristic equations, how many roots are to the right of the line  $s=-1$  in the  $s$ -plane.

(1)  $s^3 + 3s^2 + 3s + 1 = 0$

(2)  $s^3 + 4s^2 + 3s + 10 = 0$