

# Formularium - Electromagnetism I

## 1 Introduction

$$AB + BA = \heartsuit \quad (1)$$

## 2 Maxwell's Equations

## 3 Electrostatics

## 4 Magnetostatics

## 5 Plane Waves

### 5.1 Plane waves in a lossless dielectric

**Assumptions** Source free, homogenous, lossless and isotropic dielectric characterized by  $\epsilon = \epsilon_0 \epsilon_r$  and  $\mu = \mu_0 \mu_r$ .

Plane wave propagation to an arbitrary direction  $\vec{u}$ :

$$\vec{e}(\vec{r}) = \vec{A} e^{-jk\vec{u} \cdot \vec{r}}, \quad (2)$$

$$\vec{h}(\vec{r}) = \frac{1}{Z_c} \vec{u} \times \vec{e}(r), \quad (3)$$

with  $\vec{A} \cdot \vec{u} = 0$ ,  $Z_c = \frac{\mu}{\epsilon}$ ,  $k^2 = \omega^2 \epsilon \mu$  and Poynting's vector:

$$\vec{p}(\vec{r}) = \frac{|\vec{A}|^2}{2Z_c} \vec{u}. \quad (4)$$

### Plane waves in a lossy dielectric

Equations 2 and 3 with  $\vec{A} \cdot \vec{u} = 0$  remain unchanged but  $k$  and  $Z_c$  are now complex valued (in VI). Denote  $k = \beta - j\alpha$ , the new Poynting vector becomes

$$\vec{p}(\vec{r}) = \frac{|\vec{A}|^2}{2Z_c} e^{-2\alpha\vec{u} \cdot \vec{r}} \vec{u}. \quad (5)$$

This shows that when a wave propagates over a distance  $d$  in a lossy medium, its power decreases by a factor  $e^{-2\alpha d}$ . Expressed in decibel (dB) gives

$$-10 \log_{10}(e^{-2\alpha d}) \approx 8.686\alpha d, \quad (6)$$

with  $L = 8.686\alpha$  the relative power loss in (dB/m).

Considering a material with conduction losses ( $\epsilon + \frac{\sigma}{j\omega}$ ):

$$k = \omega \sqrt{\epsilon \mu} \sqrt{1 + \frac{\sigma}{j\omega\epsilon}}, \quad (7)$$

$$Z_c = \sqrt{\frac{\mu}{\epsilon}} \frac{1}{\sqrt{1 + \frac{\sigma}{j\omega\epsilon}}}, \quad (8)$$

gives rise to two interesting situations:

1. Low-loss dielectric ( $\sigma \ll \omega\epsilon$ ):

$$k = \beta - j\alpha \approx \omega\sqrt{\epsilon\mu} - j\frac{\sigma}{2}\sqrt{\frac{\mu}{\epsilon}}, \quad (9)$$

$$Z_c \approx \sqrt{\frac{\mu}{\epsilon}}, \quad (10)$$

2. Good conductor ( $\sigma \gg \omega\epsilon$ ):

$$k = \beta - j\alpha \approx \frac{1-j}{\delta}, \quad (11)$$

$$Z_c \approx \frac{1+j}{\sigma\delta}, \quad (12)$$

with  $\delta = \sqrt{\frac{2}{\omega\mu\sigma}}$  known as the *skin depth*.

## 5.2 Reflection and transmission at a plane interface

Given the incident fields in medium 1:

$$\vec{e}_i = \vec{A}e^{-jk_1\vec{u}_i \cdot \vec{r}}, \quad (13)$$

$$\vec{h}_i = \frac{1}{Z_1}\vec{u}_i \times \vec{e}_i, \quad (14)$$

with  $u_i = \cos(\theta_i)\vec{u}_z + \sin(\theta_i)\vec{u}_x$  and  $\vec{A} \cdot \vec{u}_i = 0$ . Then:

1. The reflected wave becomes

$$\vec{e}_r = \vec{B}e^{-jk_1\vec{u}_r \cdot \vec{r}}, \quad (15)$$

$$\vec{h}_r = \frac{1}{Z_1}\vec{u}_r \times \vec{e}_r, \quad (16)$$

with  $u_r = -\cos(\theta_i)\vec{u}_z + \sin(\theta_i)\vec{u}_x$  and  $\vec{B} \cdot \vec{u}_r = 0$ .

2. The transmitted wave becomes

$$\vec{e}_t = \vec{C}e^{-jk_2\vec{u}_t \cdot \vec{r}}, \quad (17)$$

$$\vec{h}_t = \frac{1}{Z_2}\vec{u}_t \times \vec{e}_t, \quad (18)$$

with  $u_t = \cos(\theta_t)\vec{u}_z + \sin(\theta_t)\vec{u}_x$ ,  $\vec{C} \cdot \vec{u}_t = 0$  and where Snell's law states that

$$k_2 \sin \theta_t = k_1 \sin \theta_i. \quad (19)$$